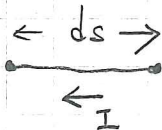


## Harmonic Electric Dipole

- What is it?
- Analytic solution
- Vector potential
- Fields
- Asymptotics
- Field transformations

### What is it?

An electric current dipole is an infinitesimal length of wire carrying a current  $I$ .



The strength of the dipole is given by  
electric dipole moment. =  $I ds$

To Do:

- more rigorous description
- can we talk about a current dipole without a generator and/or grounding
- differentiate between an electric dipole  $\begin{pmatrix} + & - \end{pmatrix}$  and electric current dipole.

?? - Devin

2016 05 06

Analytic Solution

(3)

The electric dipole is an elementary length current flowing in a single direction. It is convenient to solve the problem in terms of potentials as done in WH. To summarize, Schelkunoff

(D3)

for an electric current source  $\mathbf{J}_e^s$ , Maxwell's equations in frequency are

$$\begin{aligned}\nabla \times \mathbf{E}_e + i\omega\mu\mathbf{H}_e &= 0 \\ \nabla \times \mathbf{H}_e - (\sigma + i\omega\epsilon)\mathbf{E}_e &= \mathbf{J}_e^s\end{aligned}$$

where the subscript e reminds us that we're using an electric source. Define a vector potential

$$\mathbf{H}_e = \nabla \times \mathbf{A}$$

and manipulate to obtain the Helmholtz equation for  $\mathbf{A}$ .

$$\nabla^2 \mathbf{A} + k^2 \mathbf{A} = -\mathbf{J}_e^s \quad \text{where } k^2 = (\omega\mu\epsilon - i\omega\mu\sigma) *$$

This equation, with boundary conditions, is solved to generate  $\mathbf{A}$ . For ~~an~~ infinite media, the boundary condition is that  $\mathbf{A} \rightarrow 0$ , as  $r \rightarrow \infty$ .

If  $\mathbf{J}_e^s$  has only a single component,  $\mathbf{A}$  also has only that direction.

The scalar Green's function for  $*$  is

$$G(r) = \frac{e^{-ikr}}{4\pi r}$$

and hence the vector potential for an arbitrary electric current source is

$$\mathbf{A}(r) = \int_V \frac{e^{-ik|r-r'|}}{4\pi|r-r'|} \mathbf{J}_e(r') dV$$

(cont... Analytic solution)

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(4)

For an electric current dipole in the  $\hat{x}$  direction

(D4)

$$J_e(r) = \hat{x} I ds \delta(x) \delta(y) \delta(z)$$

and

$$A(r) = \frac{I ds}{4\pi r} e^{-ikr} \hat{x}$$

The electric and magnetic fields are

$$E_e = -i\omega\mu A + \frac{1}{(\sigma + i\omega\epsilon)} \nabla(\nabla \cdot A)$$

$$H_e = \nabla \times A$$

In component form (with eq. 2.39 - 2.42)

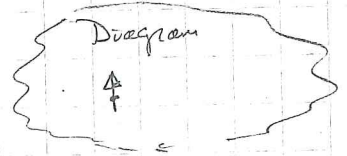
$$\left. \begin{array}{l} \leftarrow 2.39 \\ 2.40 \\ 2.41 \\ 2.42 \end{array} \right\}$$

## Vector Potential A.

The vector potential for an electric dipole has a simple mathematical representation yet it contains complete information about the electric and magnetic fields. We explore this function here.

For an electric current dipole in the  $\hat{x}$  direction, and with moment  $I ds$ , the vector potential is

$$A(r) = \frac{I ds}{4\pi r} e^{-ikr} \hat{x}$$



where

$$k = (\omega^2 \mu \epsilon - i\omega \mu \sigma)^{1/2}$$

To Do: Visualization and motivating questions

Plot:  $(Re, Im)$   $(A, \varphi)$   $\{\omega, \sigma, \mu, \epsilon\}$ ; Plan view + transect

## Physical Insight from A.

- $H = \nabla \times A$  :
- Extract data on a planar surface
  - Run a digital estimator for the curl and observe what a particular component of the magnetic field looks like
  - Relative sizes of  $Re, Im$  as a  $f(\omega)$  of frequency

$$E = -i\omega \mu A + \frac{1}{\epsilon} \nabla(\nabla \cdot A)$$

There are two terms. The first term involving  $i\omega$  arises from the  $\frac{\partial B}{\partial t}$  term in the initial equations  $\nabla \times E = -\frac{\partial B}{\partial t}$ .

The second term results because of the decomposition of a  $\nabla \times \nabla \times$  system.

The app allows comparison of the relative contribution of these two terms.

# Harmonic electric dipole

## Fields

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Purpose: To gain insight about EM fields due to a harmonic electric dipole. ~~Both near field and far field behavior are explored...~~

For an electric dipole in the  $\hat{x}$  direction, the electric field is

$$\mathbf{E} = \frac{I ds}{4\pi\epsilon_0 r^3} e^{-ikr} \left[ \right]$$

WH 2.40

and 
$$\mathbf{H} = \frac{I ds}{4\pi r^2} (1 + ikr) e^{-ikr} \left( -\frac{z}{r} \hat{x} + \frac{y}{r} \hat{z} \right)$$

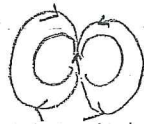
the current density, from Ohm's law, is  $\mathbf{J} = \sigma \mathbf{E}$ .

ToDo Some useful plots and what they say.

(1) Field lines for  $\mathbf{E}, \mathbf{J}$  (Plot #3, 8)

Currents flow from positive

end of the dipole into the negative end of the dipole.



(2) Magnetic field lines

$\mathbf{H}$  will have  $\hat{y}, \hat{z}$  components: Plot on a plane  $\perp \hat{x}$

? Talk about - Poloidal fields  
- Toroidal fields

(3) Movie (Plot 4)  $\mathbf{E}$ -fields propagating outwards as fields oscillate

## Asymptotics

Purpose: The character of EM fields for a harmonic electric dipole depend upon geometry and wave number. Here we explore "near-field" and "far-field" limits and gain insight about the fields. The DC fields, which are purely governed by geometry, are treated as a special case of "near-field".

- DC
- Near-field  $|kr| \ll 1$
- Far field  $|kr| \gg 1$

To do: I have attached some first pass notes. Work through, correct mistakes, extract informative equations, and physical insight, invite questions and suggest apps

# Asymptotics (harmonic electric dipole)

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## Electric dipole DC

At zero frequency, the wave number  $k=0$ . The  $E$  and  $H$  fields reduce to

$$E = \frac{I ds}{4\pi\sigma r^3} \left[ 3 \left( \frac{x^2}{r^2} \hat{x} + \frac{xy}{r^2} \hat{y} + \frac{xz}{r^2} \hat{z} \right) - \hat{x} \right]$$

$$H = \frac{I ds}{4\pi r^2} \left( -\frac{z}{r} \hat{y} + \frac{y}{r} \hat{z} \right)$$

(Note: We should also transfer these to cylindrical coordinates)

Plots: fields and currents.

## Geometric decay.

At zero frequency there is no attenuation caused by EM induction losses. The decrease in amplitude outward from the source is strictly dependent on geometry. The electric field falls off as  $1/r^3$ . This is anticipated from work with steady state currents in  $\{\text{link}\}$ . There we showed that for a point current in a uniform earth that

$$\vec{E} = \frac{I}{4\pi\sigma r^2} \hat{r}$$

The effect due to a dipole requires a spatial differentiation and hence the electric field falls as  $1/r^3$ .

TODO: Other stuff about a DC current dipole

omit } The magnetic field however falls off as  $1/r^2$ . We note that the magnetic field is independent of conductivity. Because of the different rates of decay of the  $E$  and  $H$  fields, the electromagnetic impedance  $Z = E/H$  is not easily converted to an apparent conductivity as it is when data are obtained in the far field.

• Near-field  $|k| \ll 1$

To Do: revisit attached notes, re-derive, assimilate

- Far-field  $|k| \gg 1$

To Do

Revisit attached notes, re-derive, assimilate



(b) E field = small  $|kr| \ll 1$  Near-Field

$$\frac{I ds}{4\pi r^3} e^{-ikr} \left[ \quad \right]$$

$$e^{-ikr} = 1 - ikr - \frac{k^2 r^2}{2}$$

$$\begin{aligned} \sim & (-k^2 r^2 + 3ikr + 3) \left(1 - ikr - \frac{k^2 r^2}{2}\right) \\ & = -k^2 r^2 + 3ikr + 3 + 3kr^2 - 3ikr - \frac{3}{2}k^2 r^2 \\ & = 3 + \frac{1}{2}k^2 r^2 \end{aligned}$$

$$\begin{aligned} \text{and } (k^2 r^2 - ikr - 1) \left(1 - ikr - \frac{k^2 r^2}{2}\right) & = k^2 r^2 - ikr - 1 - k^2 r^2 + ikr + \frac{k^2 r^2}{2} \\ & = -1 + \frac{k^2 r^2}{2} \end{aligned}$$

$$\text{So } \mathbf{E} = \frac{I ds}{4\pi r^3} \left[ \left( \frac{x^2}{r^2} \hat{x} + \frac{xy}{r^2} \hat{y} + \frac{xz}{r^2} \hat{z} \right) \left( 3 + \frac{1}{2}k^2 r^2 \right) + \left( -1 + \frac{k^2 r^2}{2} \right) \hat{x} \right]$$

$$k^2 r^2 \approx -i\omega\mu\sigma r^2$$

$$? \text{ So } \text{Im } \mathbf{E} \sim \frac{I ds}{4\pi r^3} [\text{geometry}] \left( +\frac{i}{2}\omega\mu\sigma r^2 \right) \quad \text{Note } \sigma\text{'s cancel}$$

$$\sim \frac{\omega\mu I ds}{4\pi r^3} [\text{geometry}] \quad \underline{\text{Independent of } \sigma!}$$

Asymptotic H-field

$$\mathbf{H} \sim \frac{I ds}{4\pi r^2} (1 + ikr) e^{-ikr} \left( -\frac{z}{r} \hat{y} + \frac{y}{r} \hat{z} \right)$$

$$\begin{aligned} (1 + ikr) \left(1 - ikr - \frac{k^2 r^2}{2}\right) & \sim 1 - ikr - \frac{k^2 r^2}{2} + ikr + k^2 r^2 \\ & \sim 1 + \frac{k^2 r^2}{2} \end{aligned}$$

Notes

20/6 0526

Asymptotic H. (near-field)

$$H = \frac{I ds}{4\pi r^2} \left( 1 + \frac{k^2 r^2}{2} \right) \left( -\frac{3}{r} \hat{y} + \frac{y \hat{z}}{r} \right)$$

$$\frac{k^2 r^2}{2} = \frac{-i\omega\mu_0 r^2}{2}$$

So the imaginary part of H varies as  $\sigma$ .

Summary: Information about conductivity at low frequencies is in  $\text{Re}[E]$  and  $\text{Im}[H]$

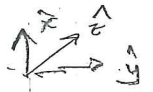
(\*) - Investigate the above statements using the app.

(c) Far Field

In the far field  $|k r| \gg 1$  amplitudes are small but may be measurable. The,

Aside: Far Field

$$E = \frac{I ds}{4\pi\epsilon_0 r^3} e^{-ikr} \left[ \left( \frac{x^2}{r^2} \hat{x} + \frac{xy}{r^2} \hat{y} + \frac{xz}{r^2} \hat{z} \right) (-k^2 r^2) + k^2 r^2 \hat{x} \right]$$

Suppose we want to evaluate at  $x=0$ ,  $y, z=0$ , (so ray)

$$? \quad E = \frac{I ds}{4\pi\epsilon_0 r^2} e^{-ikr} \left[ k^2 r^2 \hat{x} \right]$$

$$\approx \frac{I ds}{4\pi\epsilon_0 r^2} e^{-ikr} \left[ -i\omega\mu r^2 \right] \hat{x}$$

(quasi-static)

$$E \approx \frac{I ds (-i\omega\mu) e^{-ikr}}{4\pi r}$$

$$H = \frac{I ds}{4\pi r^3} (1 + ikr) e^{-ikr} \left( -\frac{z}{r} \hat{x} + \frac{y}{r} \hat{z} \right)$$

$$= \frac{I ds}{4\pi r^3} e^{-ikr} (iky) \frac{y}{r} \hat{z}$$

 $|kr| \gg 1$ 

$$H = \frac{I ds}{4\pi r^3} e^{-ikr}$$

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(1)

(D14)

## Field Transformations

Purpose: Generally the values of EM fields at specific locations do not provide insight about the value of a physical property of the medium. However scaled values of fields, or ratios of field components, can provide <sup>such</sup> information. The near-field and far field behavior of fields can often be used to estimate an apparent conductivity. The concept of EM impedance is useful.

### Outline

To Do

- Impedance
- Apparent conductivity
  - near field
  - far field

### Sensitivity

~~Purpose: To discover how the fields at a particular location are affected by changes in the location and orientation of the dipole and the values of the physical properties. This understanding is important in survey design and also is fundamental when solving an inverse source problem.~~