

#### BLM3620 Digital Signal Processing\*

Erkan Uslu

euslu@yildiz.edu.tr

Yıldız Technical University – Computer Engineering \*Based on lecture notes from Ali Can Karaca & Ahmet Elbir



#### Lecture #13 – z - Transform

- Introduce the z-Trasnform
- Examples
- Transform Example
- MATLAB Applications

#### Review & Recall



- The FFT is simply an algorithm for efficiently calculating the DFT
- Computational efficiency of an N-Point FFT:

DFT: N<sup>2</sup> Complex Multiplications
 FFT: (N/2) log<sub>2</sub>(N) Complex Multiplications

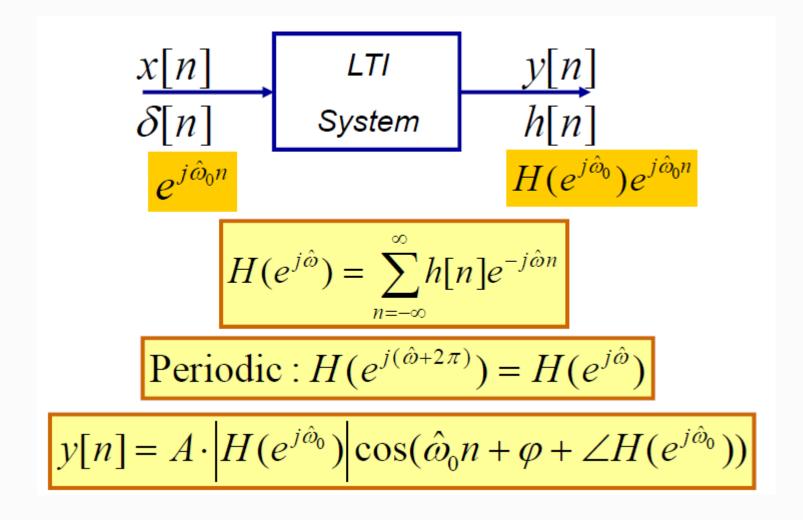
N	DFT Multiplications	FFT Multiplications	FFT Efficiency		
256	65,536	1,024	64:1		
512	262,144	2,304	114 : 1		
1,024	1,048,576	5, 120	205:1		
2,048	4, 194, 304	11,264	372:1		
4,096	16,777,216	24,576	683:1		

#### **Discrete Fourier Transform (DFT)**

$$X[k] = \sum_{n=0}^{N-1} x[n]e^{-j(2\pi/N)kn}$$

# Recap: Frequency Response $H(e^{j\widehat{\omega}})$

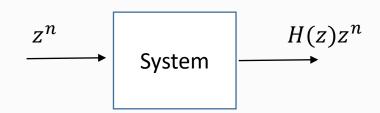






Given impulse response of an LTI system h[n], the transfer function can be found as:

$$H[z] = \sum_{k=-\infty}^{\infty} h[k] z^{-k}$$



$$=e^{j\widehat{w}}$$
)

Z transform can be computed if FT exists or not.

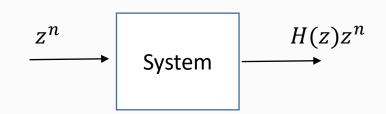
Easy to compute the system output.

Mostly used with block diagrams to illustrate the system.



Given impulse response of an LTI system h[n], the transfer function can be found as:

$$H[z] = \sum_{k=-\infty}^{\infty} h[k] z^{-k}$$



Z-transform is a Generalized version of Discrete Time Fourier  $= e^{j\widehat{w}}$ )

Z transform can be computed if FT exists or not.

Easy to compute the system output.

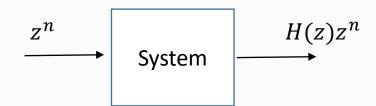
Mostly used with block diagrams to illustrate the system.



ejw ee ejw jjwwww ejw)

Given impulse response of an LTI system h[n], the transfer function can be found as:

$$H[z] = \sum_{k=-\infty}^{\infty} h[k] z^{-k}$$



Z-transform is a Generalized version of Discrete Time Fourier Transform ( $z=e^{j\widehat{w}}$ )

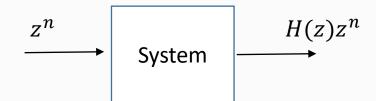
Z transform can be computed if FT exists or not.



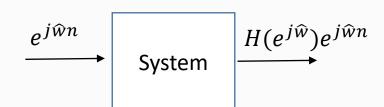
ejw ee ejw jjwwww ejw)

Given impulse response of an LTI system h[n], the transfer function can be found as:

$$H[z] = \sum_{k=-\infty}^{\infty} h[k] z^{-k}$$



Z-transform is a Generalized version of Discrete Time Fourier Transform ( $z=e^{j\widehat{w}}$ )



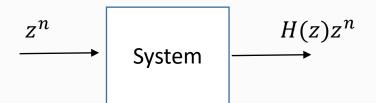
Z transform can be computed if FT exists or not.



ejw ee ejw jjwwww ejw)

Given impulse response of an LTI system h[n], the transfer function can be found as:

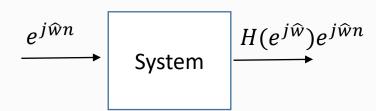
$$H[z] = \sum_{k=-\infty}^{\infty} h[k] z^{-k}$$



Z-transform is a Generalized version of Discrete Time Fourier Transform ( $z=e^{j\widehat{w}}$ )

Z transform can be computed if FT exists or not.

Z transform can be computed if FT exists or not.

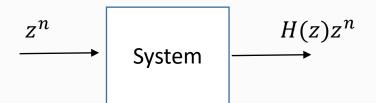




ejw ee ejw jjwwww ejw)

Given impulse response of an LTI system h[n], the transfer function can be found as:

$$H[z] = \sum_{k=-\infty}^{\infty} h[k] z^{-k}$$

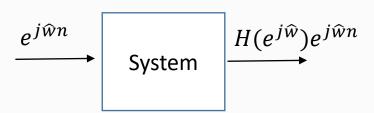


Z-transform is a Generalized version of Discrete Time Fourier Transform ( $z=e^{j\widehat{w}}$ )

Z transform can be computed if FT exists or not.

Z transform can be computed if FT exists or not.

Easy to compute the system output.

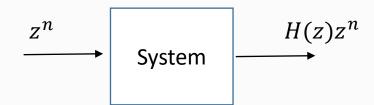




ejw ee ejw jjwwww ejw)

Given impulse response of an LTI system h[n], the transfer function can be found as:

$$H[z] = \sum_{k=-\infty}^{\infty} h[k] z^{-k}$$



Z-transform is a Generalized version of Discrete Time Fourier Transform ( $z=e^{j\widehat{w}}$ )

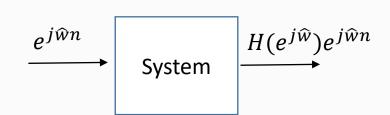
Z transform can be computed if FT exists or not.

Z transform can be computed if FT exists or not.

Easy to compute the system output.

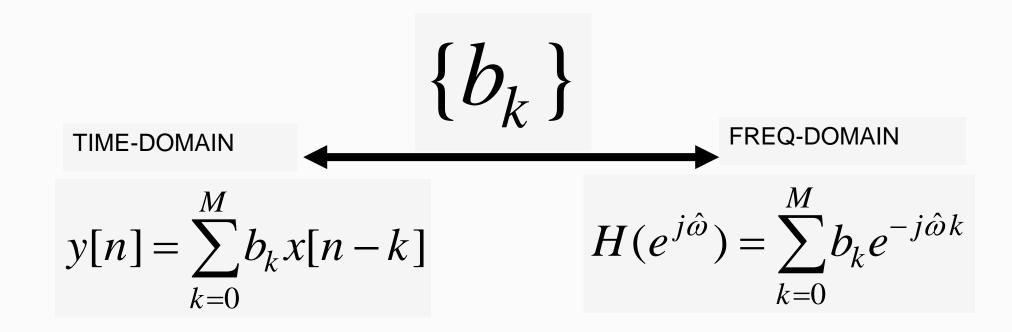
Easy to compute the system output.

Mostly used with block diagrams to illustrate the system. Mostly used with block diagrams to illustrate



#### Three Domains

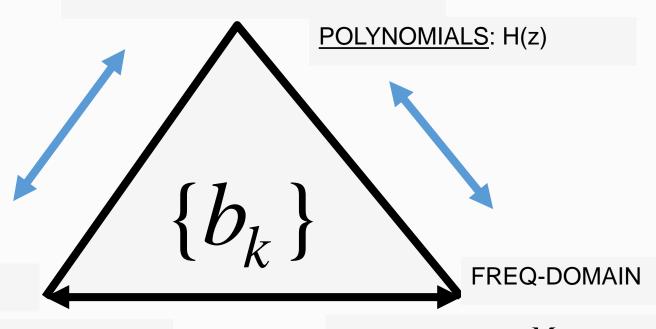




#### Three Domains







**TIME-DOMAIN** 

$$y[n] = \sum_{k=0}^{M} b_k x[n-k]$$

$$H(e^{j\hat{\omega}}) = \sum_{k=0}^{M} b_k e^{-j\hat{\omega}k}$$

#### Three Domains

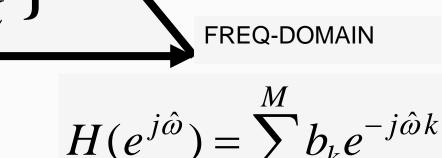




Best of both worlds?

**TIME-DOMAIN** 

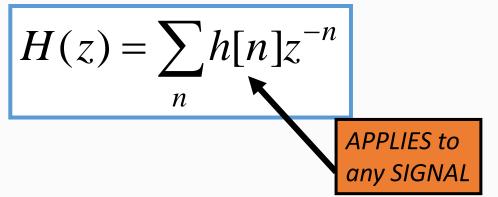
$$y[n] = \sum_{k=0}^{M} b_k x[n-k]$$





• POLYNOMIAL Representation of LTI SYSTEM:

• EXAMPLE:





POLYNOMIAL Representation of LTI SYSTEM:

• EXAMPLE:

$$H(z) = \sum_{n} h[n]z^{-n}$$

$${h[n]} = {2,0,-3,0,2}$$

APPLIES to any SIGNAL



POLYNOMIAL Representation of LTI SYSTEM:

• EXAMPLE:

$$H(z) = \sum_{n} h[n] z^{-n}$$

APPLIES to

any SIGNAL

$${h[n]} = {2,0,-3,0,2}$$

$$H(z) = 2z^{-0} + 0z^{-1} - 3z^{-2} + 0z^{-3} + 2z^{-4}$$



POLYNOMIAL Representation of LTI SYSTEM:

• EXAMPLE:

EXAMPLE: 
$$H(z) = \sum_{n} h[n]z^{-n}$$

$$\{h[n]\} = \{2, 0, -3, 0, 2\}$$

$$APPLIES to any SIGNAL$$

$$H(z) = 2z^{-0} + 0z^{-1} - 3z^{-2} + 0z^{-3} + 2z^{-4}$$
$$= 2 - 3z^{-2} + 2z^{-4}$$



POLYNOMIAL Representation of LTI SYSTEM:

• EXAMPLE:

$$H(z) = \sum_{n} h[n]z^{-n}$$

$${h[n]} = {2,0,-3,0,2}$$

$$H(z) = 2z^{-0} + 0z^{-1} - 3z^{-2} + 0z^{-3} + 2z^{-4}$$
$$= 2 - 3z^{-2} + 2z^{-4}$$

$$=2-3(z^{-1})^2+2(z^{-1})^4$$

POLYNOMIAL in  $z^{-1}$ 

APPLIES to

any SIGNAL

### Transfer Fonksiyonu



$$y[n] - y[n-1] = x[n] + 2x[n-1]$$
 sisteminin transfer fonksiyonunu bulunuz.

Fark denkleminde her iki tarafın z-dönüşümü alındığında

$$Y(z) - z^{-1}Y(z) = X(z) + 2z^{-1}X(z)$$

elde edilmektedir. Sistemin transfer fonksiyonu

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1 + 2z^{-1}}{1 - z^{-1}}$$
 olarak bulunmaktadır.



y[n]-4.5y[n-1]+2y[n-2]=x[n] şeklinde tanımlanmış sistemin blok diyagramını çizelim.



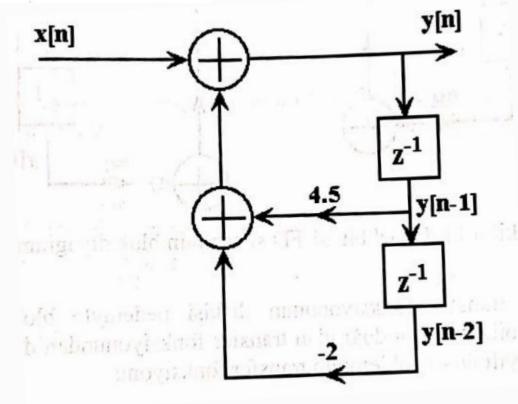
y[n]-4.5y[n-1]+2y[n-2]=x[n] şeklinde tanımlanmış sistemin blok diyagramını çizelim.

$$y[n] = x[n] + 4.5y[n-1] - 2y[n-2]$$



y[n]-4.5y[n-1]+2y[n-2]=x[n] şeklinde tanımlanmış sistemin blok diyagramını çizelim.

$$y[n] = x[n] + 4.5y[n-1] - 2y[n-2]$$



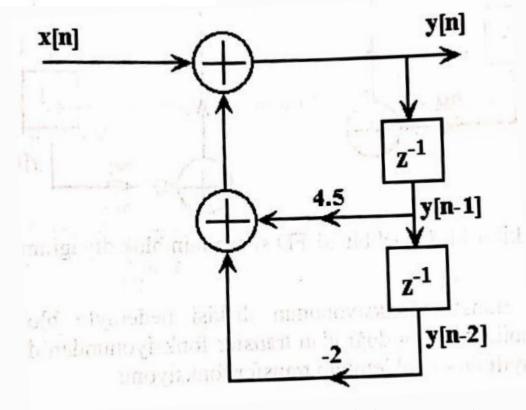
Credit by Sarp Ertürk



y[n]-4.5y[n-1]+2y[n-2]=x[n] şeklinde tanımlanmış sistemin blok diyagramını çizelim.

$$y[n] = x[n] + 4.5y[n-1] - 2y[n-2]$$

$$H(z) = \frac{1}{1 - 4.5z^{-1} + 2z^{-2}}$$



Credit by Sarp Ertürk



ANY SIGNAL has a z-Transform:

$$X(z) = \sum_{n} x[n]z^{-n}$$

$$H(z) = \sum_{n} h[n]z^{-n}$$



ANY SIGNAL has a z-Transform:

$$H(z) = \sum_{n} h[n] z^{-n}$$

n	n < -1	-1	0	1	2	3	4	5	n > 5
x[n]	0	0	2	4	6	4	2	0	0

 $X(z) = \sum x[n]z^{-n}$ 



ANY SIGNAL has a z-Transform:

$$H(z) = \sum_{n} h[n]z^{-n}$$

n	n < -1	-1	0	1	2	3	4	5	<i>n</i> > 5
x[n]	0	0	2	4	6	4	2	0	0

 $X(z) = \sum x[n]z^{-n}$ 

$$X(z) = ?$$



ANY SIGNAL has a z-Transform:

$$H(z) = \sum_{n} h[n]z^{-n}$$

$$X(z) = \sum_{n} x[n]z^{-n}$$

#### Example 7.1

n	n < -1	-1	0	1	2	3	4	5	<i>n</i> > 5
x[n]	0	0	2	4	6	4	2	0	0

$$X(z) = ?$$

$$X(z) = ? X(z) = 2 + 4z^{-1} + 6z^{-2} + 4z^{-3} + 2z^{-4}$$



ANY SIGNAL has a z-Transform:

$$H(z) = \sum_{n} h[n]z^{-n}$$

$$X(z) = \sum_{n} x[n]z^{-n}$$

#### Example 7.1

n	n < -1	-1	0	1	2	3	4	5	<i>n</i> > 5
x[n]	0	0	2	4	6	4	2	0	0
Y(7) - 2 $Y(-)$ $2 + 4 - 1 + 6 - 2 + 4 - 3 + 2 - 4$									

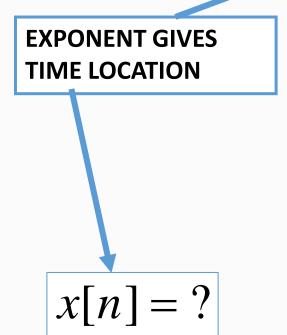
$$X(z) = ?$$

$$X(z) = ? | X(z) = 2 + 4z^{-1} + 6z^{-2} + 4z^{-3} + 2z^{-4}$$

## Example 9.2



$$X(z) = 1 - 2z^{-1} + 3z^{-3} - z^{-5}$$



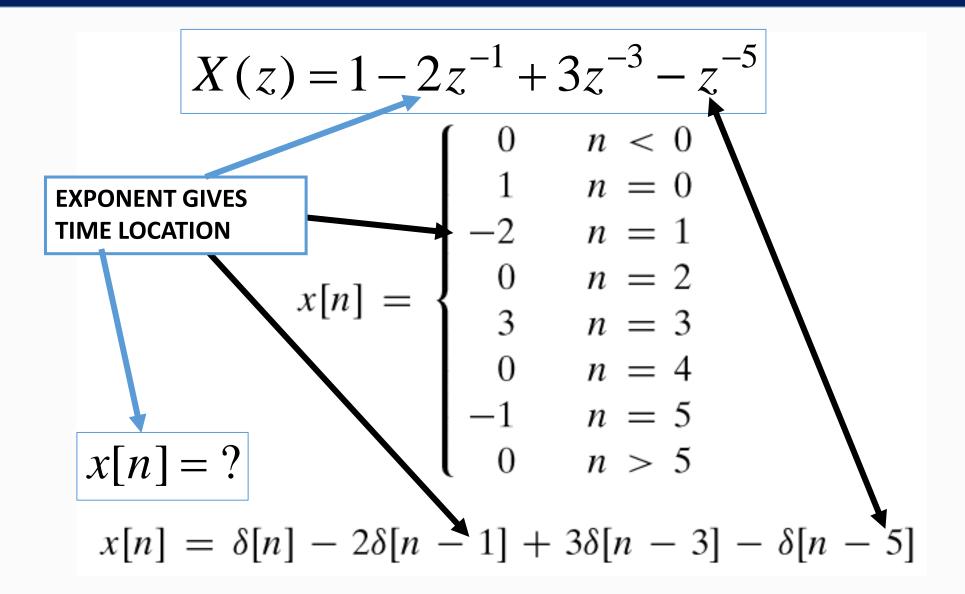
### Example 9.2



$$X(z) = 1 - 2z^{-1} + 3z^{-3} - z^{-5}$$
 EXPONENT GIVES TIME LOCATION 
$$x[n] = \begin{cases} 0 & n < 0 \\ 1 & n = 0 \\ -2 & n = 1 \\ 0 & n = 2 \\ 3 & n = 3 \\ 0 & n = 4 \\ -1 & n = 5 \\ 0 & n > 5 \end{cases}$$
 
$$x[n] = 8[n] - 2\delta[n-1] + 3\delta[n-3] - \delta[n-5]$$

### Example 9.2





### Z-Transform Property: Delay Property



A delay of one sample multiplies the z-transform by  $z^{-1}$ .

$$x[n-1]$$

$$\iff$$

$$x[n-1] \iff z^{-1}X(z)$$

Time delay of  $n_0$  samples multiplies the z-transform by  $z^{-n_0}$ 

$$x[n-n_0]$$

$$\iff$$

$$x[n-n_0] \iff z^{-n_0}X(z)$$

### Example:



Find z-transform of shifted impulse  $x[n] = 3\delta[n-5]$ .

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

$$X(z) = \sum_{n=-\infty}^{\infty} 3\delta[n-5] z^{-n} = 3\delta[0] z^{-5} = 3z^{-5}$$

## Example:



Find z-transform of shifted impulse  $x[n] = 3\delta[n-5]$ .

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

$$X(z) = \sum_{n=-\infty}^{\infty} 3\delta[n-5] z^{-n} = 3\delta[0] z^{-5} = 3z^{-5}$$

Fourier transform?

## Example:



Find z-transform of shifted impulse  $x[n] = 3\delta[n-5]$ .

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

$$X(z) = \sum_{n=-\infty}^{\infty} 3\delta[n-5] z^{-n} = 3\delta[0] z^{-5} = 3z^{-5}$$

Fourier transform?

$$X(e^{j\Omega}) = X[z]\Big|_{z=e^{jw}} = 3e^{-j5w}$$

### Z-Transform of FIR Filter



- CALLED the <u>SYSTEM FUNCTION</u>
  - because h[n] is same as {b<sub>k</sub>}

$$H(z) = \sum_{k=0}^{M} b_k z^{-k} = \sum_{k=0}^{M} h[k] z^{-k}$$

$$y[n] = \sum_{k=0}^{M} b_k x[n-k] = \sum_{k=0}^{M} h[k] x[n-k]$$

FIR DIFFERENCE EQUATION

CONVOLUTION

### Z-Transform of FIR Filter



- Get H(z) DIRECTLY from the {b<sub>k</sub>}
- Example 7.3 in the book:

$$y[n] = 6x[n] - 5x[n-1] + x[n-2]$$

$$\{b_k\} = \{6, -5, 1\}$$

$$H(z) = \sum b_k z^{-1} = 6 - 5z^{-1} + z^{-2}$$

### GENERAL I/O PROBLEM



- Input is x[n], find y[n] (for FIR, h[n])
- How to combine X(z) and H(z)?

### Example 7.5

$$x[n] = \delta[n-1] - \delta[n-2] + \delta[n-3] - \delta[n-4]$$

and 
$$h[n] = \delta[n] + 2\delta[n-1] + 3\delta[n-2] + 4\delta[n-3]$$

$$X(z) = 0 + 1z^{-1} - 1z^{-2} + 1z^{-3} - 1z^{-4}$$

and 
$$H(z) = 1 + 2z^{-1} + 3z^{-2} + 4z^{-3}$$

### FIR Filter = CONVOLUTION

$$y[n] = \sum_{k=0}^{M} b_k x[n-k] = \sum_{k=0}^{M} h[k]x[n-k]$$
CONVOLUTION

## FIR Filter = CONVOLUTION

```
x[n], X(z) 0 +1 -1 +1 -1
h[n], H(z) 1 2 3 4
         0
            0 +2 -2 +2 -2
                0 +3 -3 +3 -3
y[n], Y(z) 0 +1 +1 +2 +2 -3 +1 -4
```

$$y[n] = \sum_{k=0}^{M} b_k x[n-k] = \sum_{k=0}^{M} h[k]x[n-k]$$
CONVOLUTION

### CONVOLUTION PROPERTY



#### PROOF:

$$y[n] = x[n] * h[n] = \sum_{k=0}^{M} h[k]x[n-k]$$

$$Y(z) = \sum_{k=0}^{M} h[k] (z^{-k}X(z))$$

$$= \left(\sum_{k=0}^{M} h[k]z^{-k}\right) X(z) = H(z)X(z).$$

### CONVOLUTION EXAMPLE



#### • **MULTIPLY** the z-TRANSFORMS:

#### Example 7.5

$$x[n] = \delta[n-1] - \delta[n-2] + \delta[n-3] - \delta[n-4]$$
and 
$$h[n] = \delta[n] + 2\delta[n-1] + 3\delta[n-2] + 4\delta[n-3]$$

$$X(z) = 0 + 1z^{-1} - 1z^{-2} + 1z^{-3} - 1z^{-4}$$
  
and 
$$H(z) = 1 + 2z^{-1} + 3z^{-2} + 4z^{-3}$$

$$y[n] = h[n] * x[n] \leftrightarrow Y(z) = H(z)X(z)$$

### CONVOLUTION EXAMPLE



- Finite-Length input x[n]
- FIR Filter (L=4)

MULTIPLY Z-TRANSFORMS

$$Y(z) = H(z)X(z)$$

$$= (1 + 2z^{-1} + 3z^{-2} + 4z^{-3})(z^{-1} - z^{-2} + z^{-3} - z^{-4})$$

$$= z^{-1} + (-1 + 2)z^{-2} + (1 - 2 + 3)z^{-3} + (-1 + 2 - 3 + 4)z^{-4}$$

$$+ (-2 + 3 - 4)z^{-5} + (-3 + 4)z^{-6} + (-4)z^{-7}$$

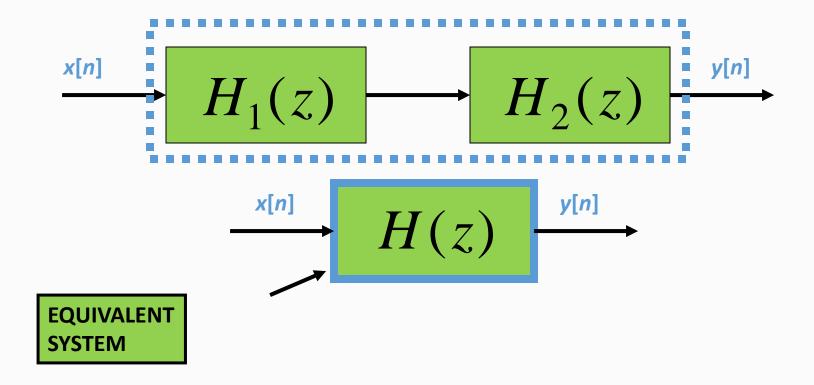
$$= z^{-1} + z^{-2} + 2z^{-3} + 2z^{-4} - 3z^{-5} + z^{-6} - 4z^{-7}$$

$$y[n] = ?$$

### Z-Transform Property: CASCADE EQUIVALENT



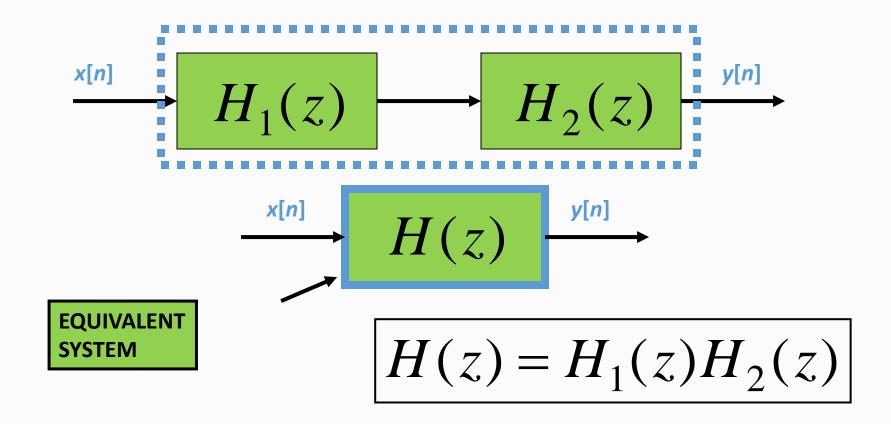
Multiply the System Functions



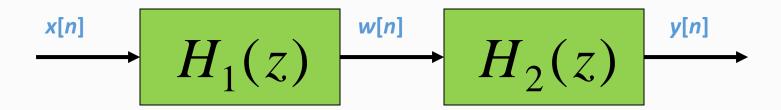
### Z-Transform Property: CASCADE EQUIVALENT

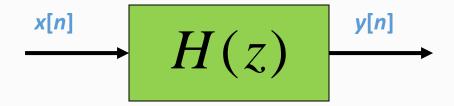


Multiply the System Functions

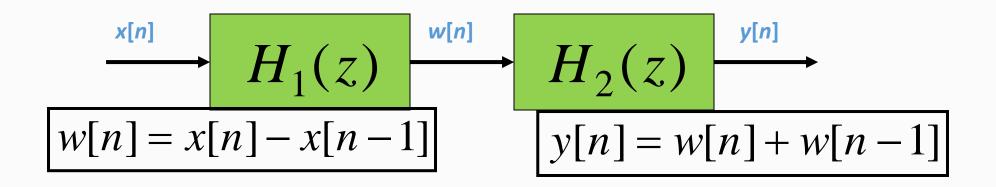


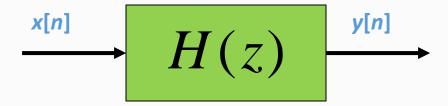




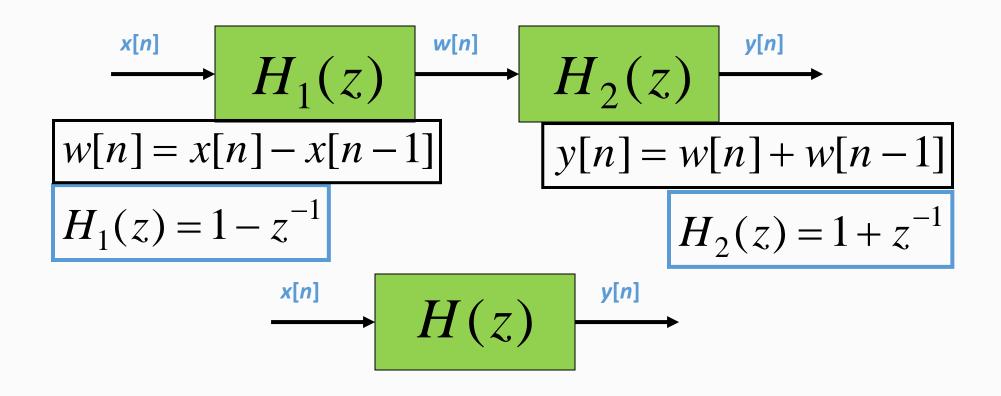




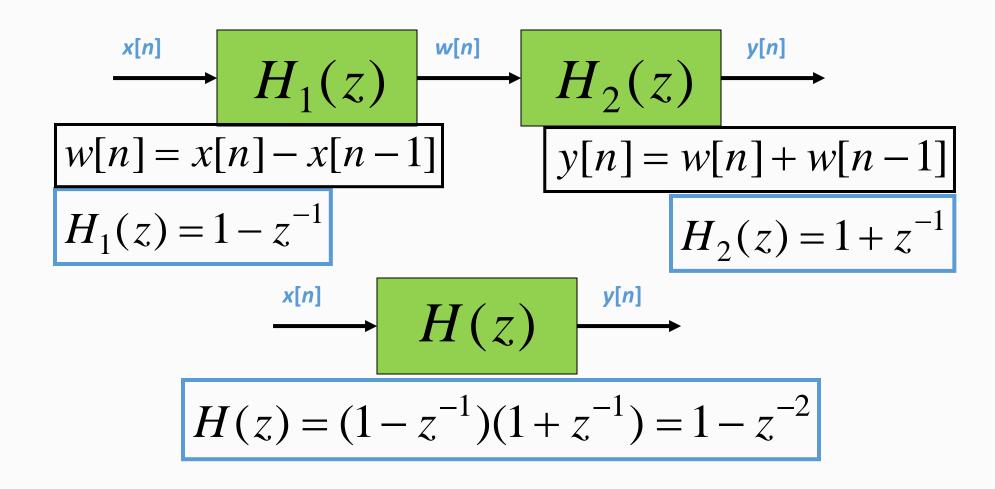




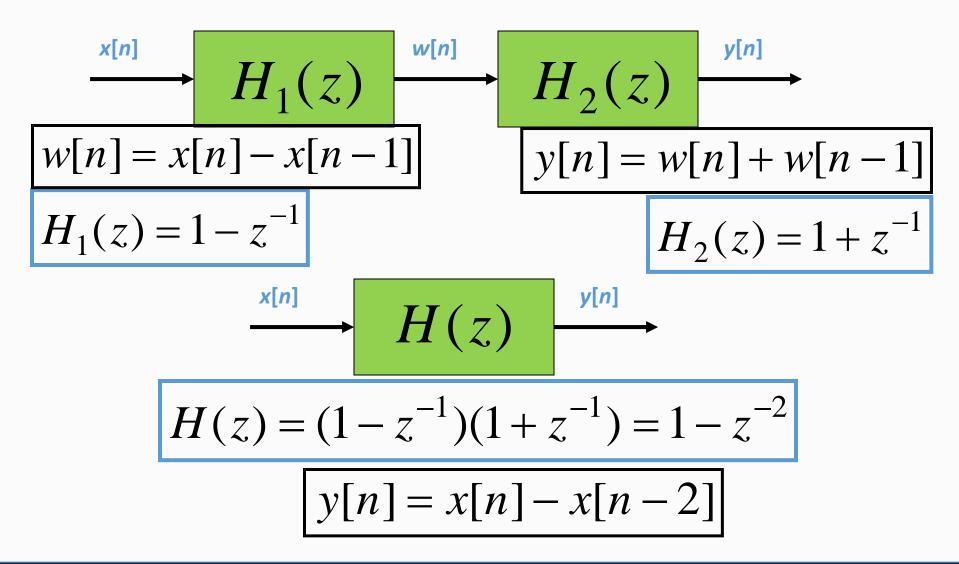












## Region of Convergence (ROC)



• Give a sequence, the set of values of z for which the z-transform converges, i.e.,  $|X(z)| < \infty$ , is called the region of convergence.

$$ROC = \{ z = re^{j\omega} \mid R_{x-} < r < R_{x+} \}$$

## Region of Convergence (ROC)

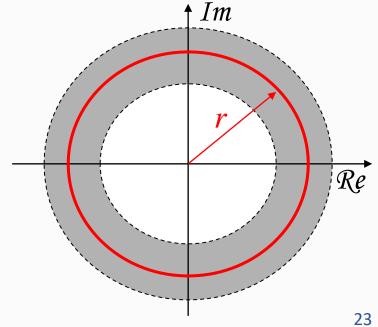


• Give a sequence, the set of values of z for which the z-transform converges, i.e.,  $|X(z)| < \infty$ , is called the region of convergence.

$$|X(z)| = \left| \sum_{n=-\infty}^{\infty} x(n) z^{-n} \right| = \sum_{n=-\infty}^{\infty} |x(n)| |z|^{-n} < \infty$$

ROC is centered on origin and consists of a set of rings.

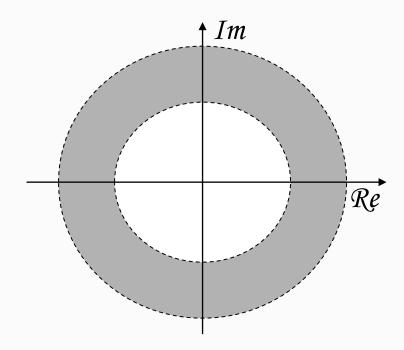
$$ROC = \{ z = re^{j\omega} \mid R_{x-} < r < R_{x+} \}$$



# Stable Systems



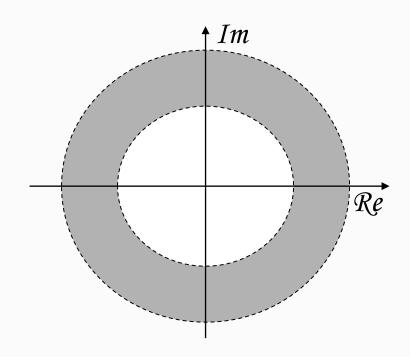
• A stable system requires that its Fourier transform is uniformly convergent.



# Stable Systems



• A stable system requires that its Fourier transform is uniformly convergent.

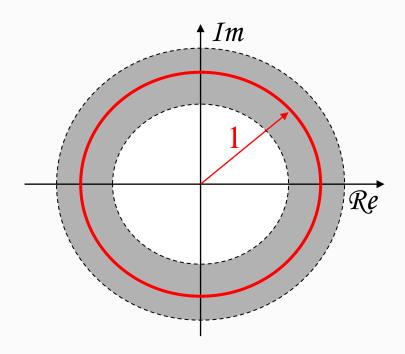


• Fact: Fourier transform is to evaluate *z*-transform on a unit circle.

# Stable Systems



• A stable system requires that its Fourier transform is uniformly convergent.



- Fact: Fourier transform is to evaluate *z*-transform on a unit circle.
- A stable system requires the ROC of *z*-transform to include the unit circle.



$$x(n) = a^n u(n)$$

$$X(z) = \sum_{n=-\infty}^{\infty} a^n u(n) z^{-n}$$



$$x(n) = a^n u(n)$$

$$X(z) = \sum_{n=-\infty}^{\infty} a^n u(n) z^{-n}$$
$$= \sum_{n=0}^{\infty} a^n z^{-n}$$



$$x(n) = a^n u(n)$$

$$X(z) = \sum_{n=-\infty}^{\infty} a^n u(n) z^{-n}$$

$$= \sum_{n=0}^{\infty} a^n z^{-n}$$

$$= \sum_{n=0}^{\infty} (az^{-1})^n$$



$$x(n) = a^n u(n)$$

$$X(z) = \sum_{n=-\infty}^{\infty} a^n u(n) z^{-n}$$

$$= \sum_{n=0}^{\infty} a^n z^{-n}$$

$$= \sum_{n=0}^{\infty} (az^{-1})^n$$

$$\sum_{n=0}^{\infty} |az^{-1}| < \infty$$



$$x(n) = a^n u(n)$$

$$X(z) = \sum_{n=-\infty}^{\infty} a^n u(n) z^{-n}$$

$$= \sum_{n=0}^{\infty} a^n z^{-n}$$

$$= \sum_{n=0}^{\infty} (az^{-1})^n$$

$$\sum_{n=0}^{\infty} |az^{-1}| < \infty \implies |az^{-1}| < 1$$



$$x(n) = a^n u(n)$$

$$X(z) = \sum_{n=-\infty}^{\infty} a^n u(n) z^{-n}$$

$$= \sum_{n=0}^{\infty} a^n z^{-n}$$

$$= \sum_{n=0}^{\infty} (az^{-1})^n$$

$$\sum_{n=0}^{\infty} |az^{-1}| < \infty \qquad |az^{-1}| < 1$$

$$|z| > |a|$$



$$x(n) = a^n u(n)$$

$$X(z) = \sum_{n=-\infty}^{\infty} a^n u(n) z^{-n}$$

$$= \sum_{n=0}^{\infty} a^n z^{-n}$$

$$= \sum_{n=0}^{\infty} (az^{-1})^n$$

$$\sum_{n=0}^{\infty} |az^{-1}| < \infty \qquad |az^{-1}| < 1$$

$$|z| > |a|$$

$$X(z) = \sum_{n=0}^{\infty} (az^{-1})^n = \frac{1}{1 - az^{-1}} = \frac{z}{z - a}$$

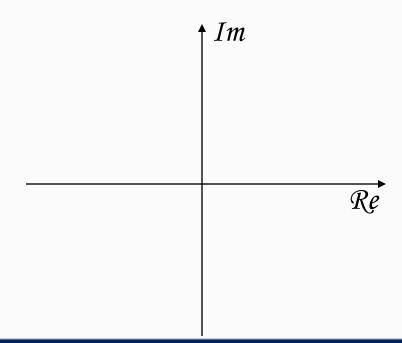
$$|z| > |a|$$



$$X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$

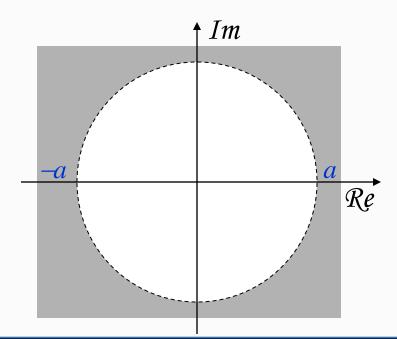


$$X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$



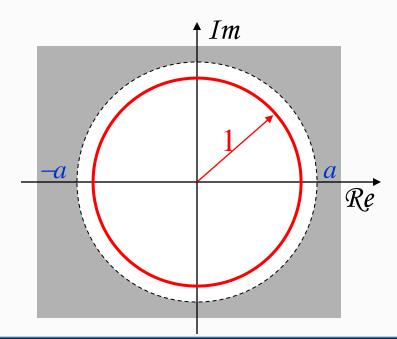


$$X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$



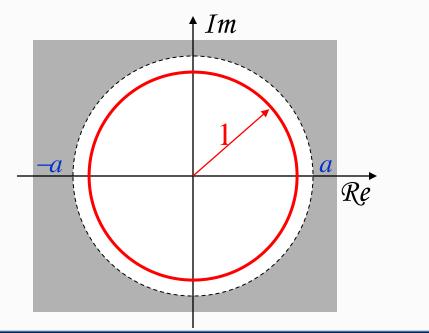


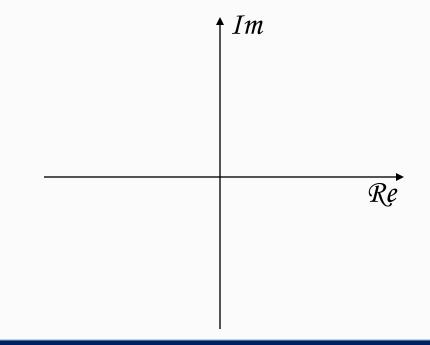
$$X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$





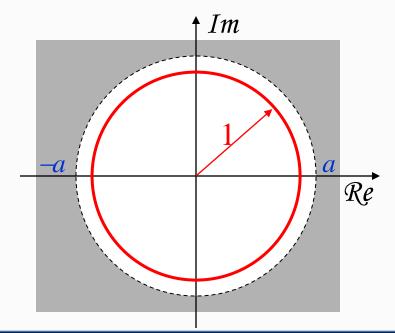
$$X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$

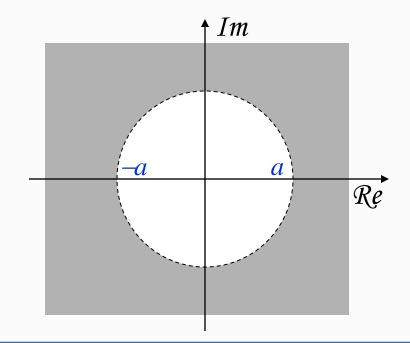






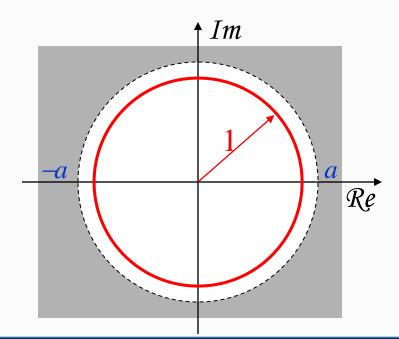
$$X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$

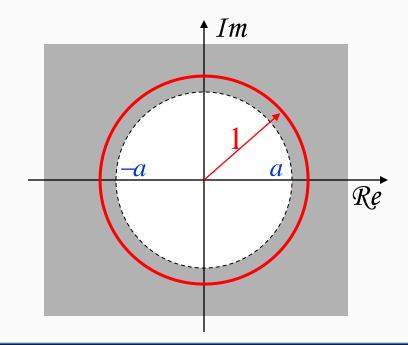






$$X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$

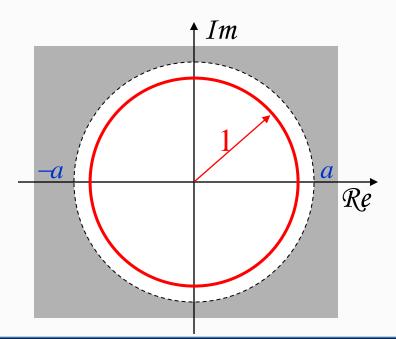


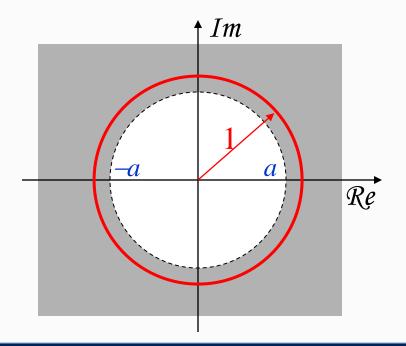




$$X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$

### Which one is stable?

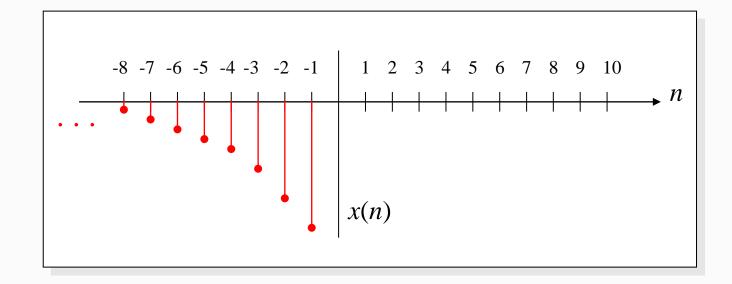




## Another Example



$$x(n) = -a^n u(-n-1)$$





$$x(n) = -a^n u(-n-1)$$

$$X(z) = -\sum_{n=-\infty}^{\infty} a^{n} u(-n-1) z^{-n}$$



$$x(n) = -a^n u(-n-1)$$

$$X(z) = -\sum_{n=-\infty}^{\infty} a^n u(-n-1)z^{-n}$$
$$= -\sum_{n=-\infty}^{-1} a^n z^{-n}$$



$$x(n) = -a^n u(-n-1)$$

$$X(z) = -\sum_{n=-\infty}^{\infty} a^n u(-n-1)z^{-n}$$
$$= -\sum_{n=-\infty}^{-1} a^n z^{-n}$$
$$= -\sum_{n=1}^{\infty} a^{-n} z^n$$



$$x(n) = -a^n u(-n-1)$$

$$X(z) = -\sum_{n=-\infty}^{\infty} a^n u(-n-1)z^{-n}$$

$$= -\sum_{n=-\infty}^{-1} a^n z^{-n}$$

$$= -\sum_{n=1}^{\infty} a^{-n} z^n$$

$$= 1 - \sum_{n=0}^{\infty} a^{-n} z^n$$



$$x(n) = -a^n u(-n-1)$$

$$X(z) = -\sum_{n=-\infty}^{\infty} a^n u(-n-1)z^{-n}$$

$$= -\sum_{n=-\infty}^{-1} a^n z^{-n}$$

$$= -\sum_{n=1}^{\infty} a^{-n} z^n$$

$$= 1 - \sum_{n=0}^{\infty} a^{-n} z^n$$

$$\sum_{n=0}^{\infty} |a^{-1}z| < \infty$$



$$x(n) = -a^n u(-n-1)$$

$$X(z) = -\sum_{n=-\infty}^{\infty} a^n u(-n-1)z^{-n}$$

$$= -\sum_{n=-\infty}^{-1} a^n z^{-n}$$

$$= -\sum_{n=1}^{\infty} a^{-n} z^n$$

$$= 1 - \sum_{n=0}^{\infty} a^{-n} z^n$$

$$\sum_{n=0}^{\infty} |a^{-1}z| < \infty \qquad |a^{-1}z| < 1$$



$$x(n) = -a^n u(-n-1)$$

$$X(z) = -\sum_{n=-\infty}^{\infty} a^n u(-n-1)z^{-n}$$

$$= -\sum_{n=-\infty}^{-1} a^n z^{-n}$$

$$= -\sum_{n=1}^{\infty} a^{-n} z^n$$

$$= 1 - \sum_{n=0}^{\infty} a^{-n} z^n$$

$$\sum_{n=0}^{\infty} |a^{-1}z| < \infty \qquad |a^{-1}z| < 1$$

$$|z| < |a|$$



$$x(n) = -a^n u(-n-1)$$

$$X(z) = -\sum_{n=-\infty}^{\infty} a^n u(-n-1)z^{-n}$$

$$= -\sum_{n=-\infty}^{-1} a^n z^{-n}$$

$$= -\sum_{n=1}^{\infty} a^{-n} z^n$$

$$= 1 - \sum_{n=0}^{\infty} a^{-n} z^n$$

$$\sum_{n=0}^{\infty} |a^{-1}z| < \infty \qquad |a^{-1}z| < 1$$

$$|z| < |a|$$

$$X(z) = 1 - \sum_{n=0}^{\infty} (a^{-1}z)^n = 1 - \frac{1}{1 - a^{-1}z} = \frac{z}{z - a}$$
$$|z| < |a|$$



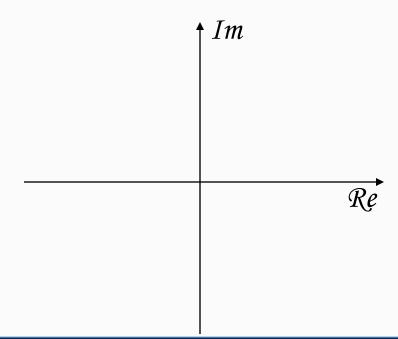


$$X(z) = \frac{z}{z - a}, \qquad |z| < |a|$$





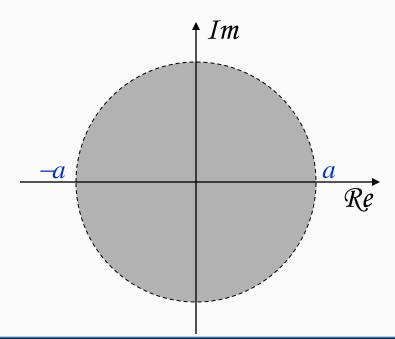
$$X(z) = \frac{z}{z - a}, \qquad |z| < |a|$$







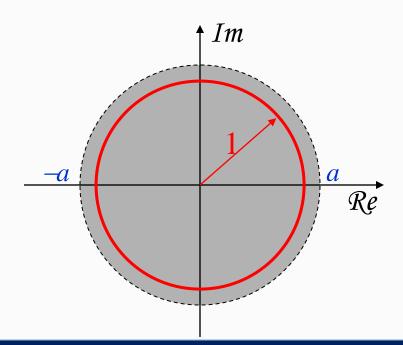
$$X(z) = \frac{z}{z - a}, \qquad |z| < |a|$$





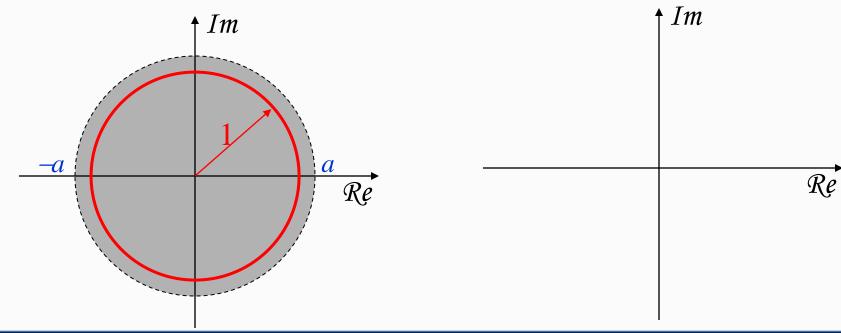


$$X(z) = \frac{z}{z - a}, \qquad |z| < |a|$$



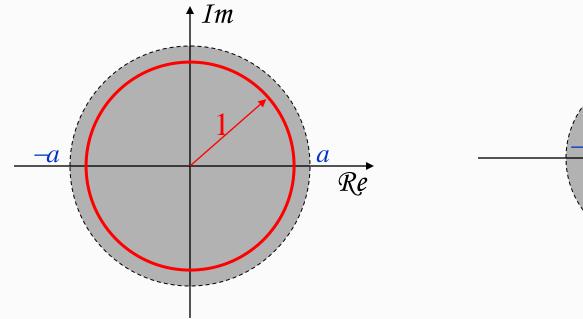


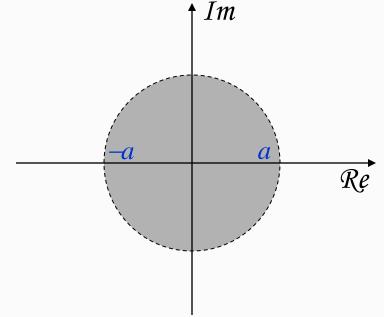
$$X(z) = \frac{z}{z - a}, \qquad |z| < |a|$$





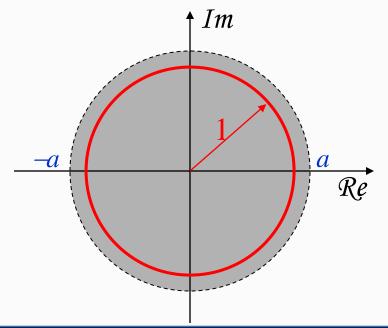
$$X(z) = \frac{z}{z - a}, \qquad |z| < |a|$$

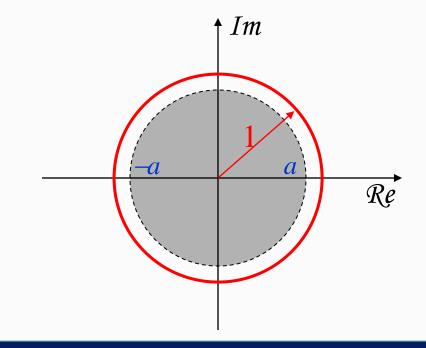






$$X(z) = \frac{z}{z - a}, \qquad |z| < |a|$$

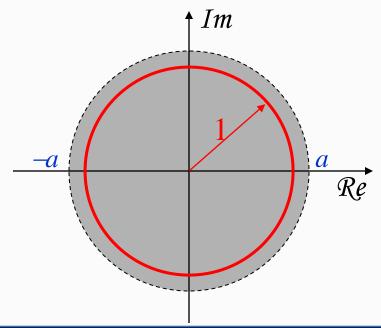


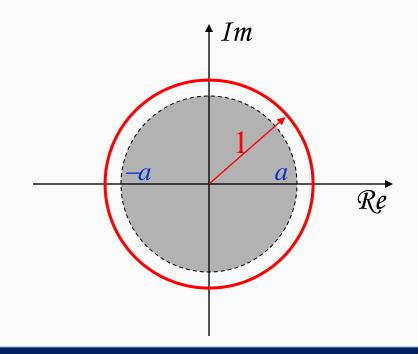




$$X(z) = \frac{z}{z - a}, \qquad |z| < |a|$$

#### Which one is stable?







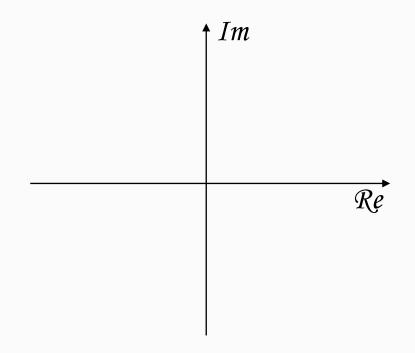
$$x(n) = a^n u(n)$$



$$X(n) = a^n u(n) \qquad X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$

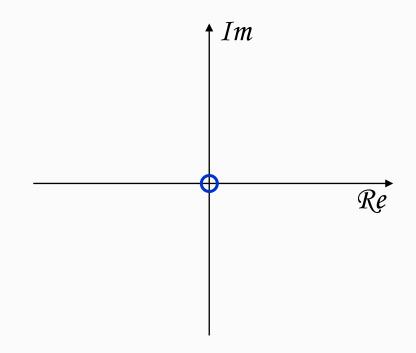


$$X(n) = a^n u(n) \qquad X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$



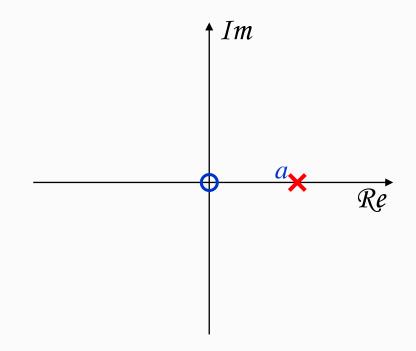


$$X(n) = a^n u(n) \qquad X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$



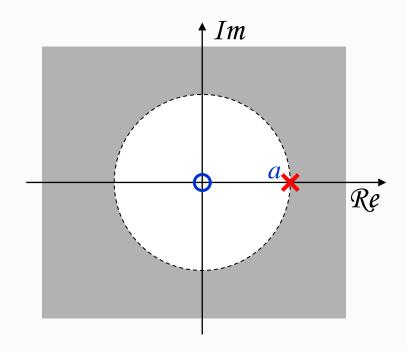


$$X(n) = a^n u(n) \qquad X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$



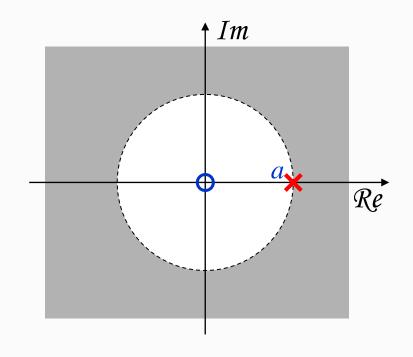


$$X(n) = a^n u(n) \qquad X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$





$$X(n) = a^n u(n) \qquad X(z) = \frac{z}{z - a}, \qquad |z| > |a|$$



ROC is bounded by the pole and is the exterior of a circle.



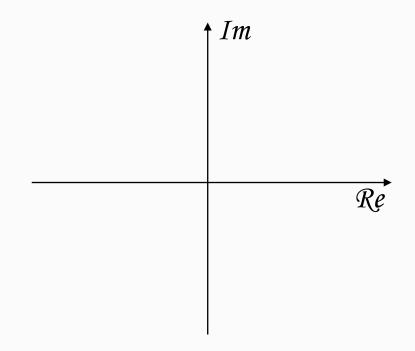
$$x(n) = -a^n u(-n-1)$$



$$x(n) = -a^n u(-n-1)$$
  $X(z) = \frac{z}{z-a}, |z| < |a|$ 

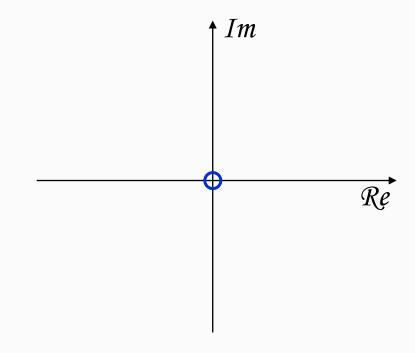


$$x(n) = -a^n u(-n-1)$$
  $X(z) = \frac{z}{z-a}, |z| < |a|$ 



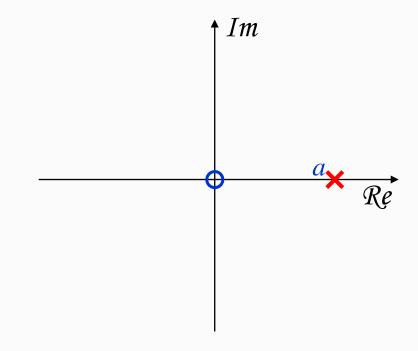


$$x(n) = -a^n u(-n-1)$$
  $X(z) = \frac{z}{z-a}, |z| < |a|$ 



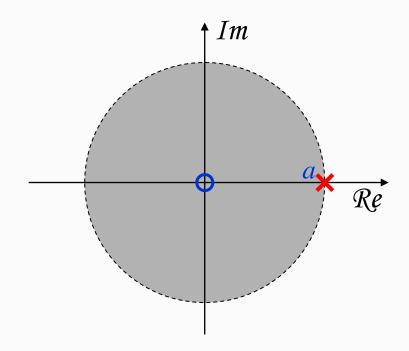


$$x(n) = -a^n u(-n-1)$$
  $X(z) = \frac{z}{z-a}, |z| < |a|$ 



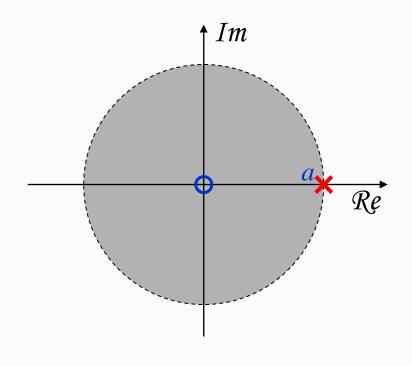


$$x(n) = -a^n u(-n-1)$$
  $X(z) = \frac{z}{z-a}, |z| < |a|$ 





$$x(n) = -a^n u(-n-1)$$
  $X(z) = \frac{z}{z-a}, |z| < |a|$ 



ROC is bounded by the pole and is the interior of a circle.

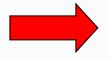




$$x(n) = (\frac{1}{2})^n u(n) + (-\frac{1}{3})^n u(n)$$



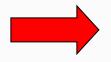
$$x(n) = (\frac{1}{2})^n u(n) + (-\frac{1}{3})^n u(n)$$



$$X(z) = \frac{z}{z - \frac{1}{2}} + \frac{z}{z + \frac{1}{3}}$$



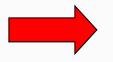
$$x(n) = (\frac{1}{2})^n u(n) + (-\frac{1}{3})^n u(n)$$



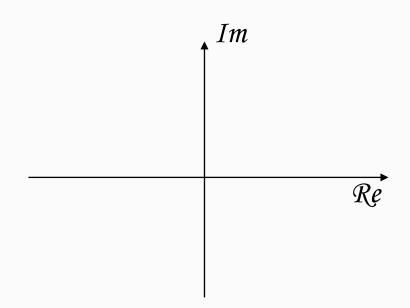
$$X(z) = \frac{z}{z - \frac{1}{2}} + \frac{z}{z + \frac{1}{3}} = \frac{2z(z - \frac{1}{12})}{(z - \frac{1}{2})(z + \frac{1}{3})}$$



$$x(n) = (\frac{1}{2})^n u(n) + (-\frac{1}{3})^n u(n)$$

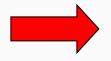


$$X(z) = \frac{z}{z - \frac{1}{2}} + \frac{z}{z + \frac{1}{3}} = \frac{2z(z - \frac{1}{12})}{(z - \frac{1}{2})(z + \frac{1}{3})}$$

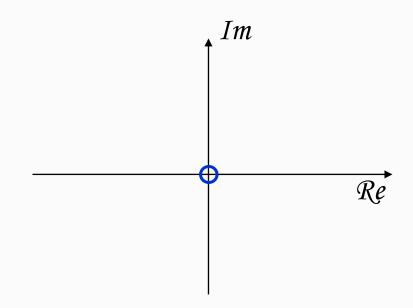




$$x(n) = (\frac{1}{2})^n u(n) + (-\frac{1}{3})^n u(n)$$

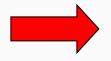


$$X(z) = \frac{z}{z - \frac{1}{2}} + \frac{z}{z + \frac{1}{3}} = \frac{2z(z - \frac{1}{12})}{(z - \frac{1}{2})(z + \frac{1}{3})}$$

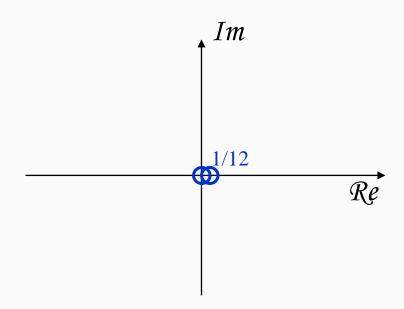




$$x(n) = (\frac{1}{2})^n u(n) + (-\frac{1}{3})^n u(n)$$

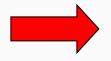


$$X(z) = \frac{z}{z - \frac{1}{2}} + \frac{z}{z + \frac{1}{3}} = \frac{2z(z - \frac{1}{12})}{(z - \frac{1}{2})(z + \frac{1}{3})}$$

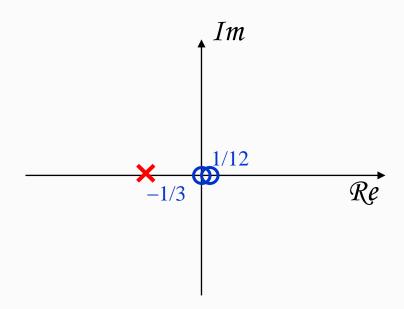




$$x(n) = (\frac{1}{2})^n u(n) + (-\frac{1}{3})^n u(n)$$



$$X(z) = \frac{z}{z - \frac{1}{2}} + \frac{z}{z + \frac{1}{3}} = \frac{2z(z - \frac{1}{12})}{(z - \frac{1}{2})(z + \frac{1}{3})}$$

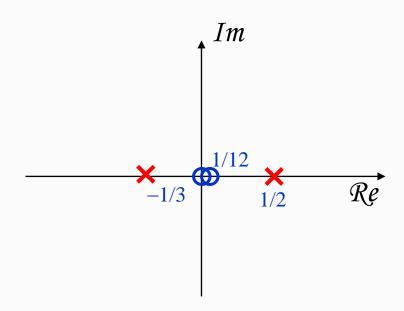




$$x(n) = (\frac{1}{2})^n u(n) + (-\frac{1}{3})^n u(n)$$

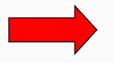


$$X(z) = \frac{z}{z - \frac{1}{2}} + \frac{z}{z + \frac{1}{3}} = \frac{2z(z - \frac{1}{12})}{(z - \frac{1}{2})(z + \frac{1}{3})}$$

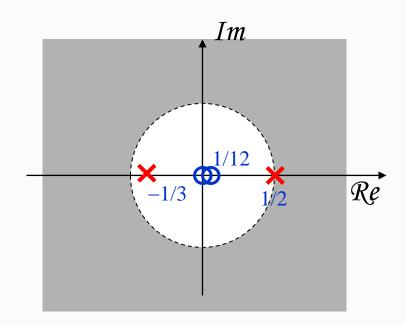




$$x(n) = (\frac{1}{2})^n u(n) + (-\frac{1}{3})^n u(n)$$

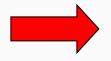


$$X(z) = \frac{z}{z - \frac{1}{2}} + \frac{z}{z + \frac{1}{3}} = \frac{2z(z - \frac{1}{12})}{(z - \frac{1}{2})(z + \frac{1}{3})}$$

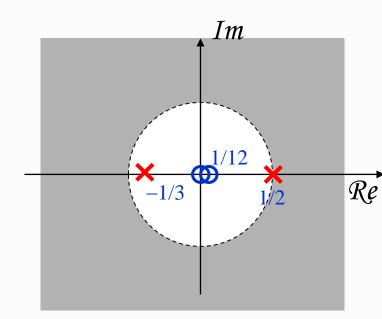




$$x(n) = (\frac{1}{2})^n u(n) + (-\frac{1}{3})^n u(n)$$



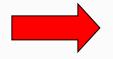
$$X(z) = \frac{z}{z - \frac{1}{2}} + \frac{z}{z + \frac{1}{3}} = \frac{2z(z - \frac{1}{12})}{(z - \frac{1}{2})(z + \frac{1}{3})}$$



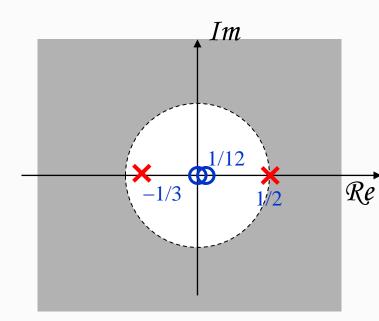
ROC is bounded by poles and is the exterior of a circle.



$$x(n) = (\frac{1}{2})^n u(n) + (-\frac{1}{3})^n u(n)$$



$$X(z) = \frac{z}{z - \frac{1}{2}} + \frac{z}{z + \frac{1}{3}} = \frac{2z(z - \frac{1}{12})}{(z - \frac{1}{2})(z + \frac{1}{3})}$$



ROC is bounded by poles and is the exterior of a circle.

ROC does not include any pole.



$$x(n) = \left(-\frac{1}{3}\right)^n u(n) - \left(\frac{1}{2}\right)^n u(-n-1)$$



$$x(n) = \left(-\frac{1}{3}\right)^n u(n) - \left(\frac{1}{2}\right)^n u(-n-1)$$

$$X(z) = \frac{z}{z + \frac{1}{3}} + \frac{z}{z - \frac{1}{2}}$$



$$x(n) = (-\frac{1}{3})^n u(n) - (\frac{1}{2})^n u(-n-1)$$

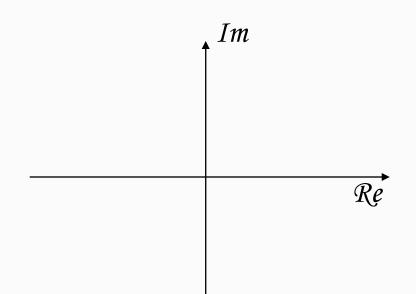
$$X(z) = \frac{z}{z + \frac{1}{3}} + \frac{z}{z - \frac{1}{2}} = \frac{2z(z - \frac{1}{12})}{(z + \frac{1}{3})(z - \frac{1}{2})}$$



$$x(n) = (-\frac{1}{3})^n u(n) - (\frac{1}{2})^n u(-n-1)$$



$$X(z) = \frac{z}{z + \frac{1}{3}} + \frac{z}{z - \frac{1}{2}} = \frac{2z(z - \frac{1}{12})}{(z + \frac{1}{3})(z - \frac{1}{2})}$$

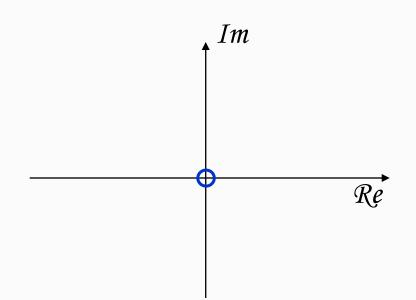




$$x(n) = (-\frac{1}{3})^n u(n) - (\frac{1}{2})^n u(-n-1)$$

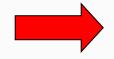


$$X(z) = \frac{z}{z + \frac{1}{3}} + \frac{z}{z - \frac{1}{2}} = \frac{2z(z - \frac{1}{12})}{(z + \frac{1}{3})(z - \frac{1}{2})}$$

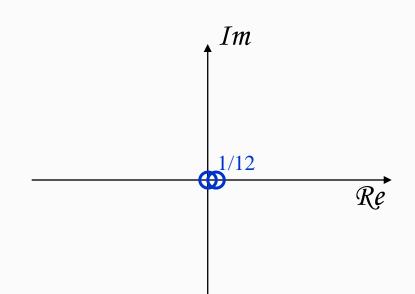




$$x(n) = (-\frac{1}{3})^n u(n) - (\frac{1}{2})^n u(-n-1)$$

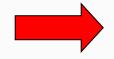


$$X(z) = \frac{z}{z + \frac{1}{3}} + \frac{z}{z - \frac{1}{2}} = \frac{2z(z - \frac{1}{12})}{(z + \frac{1}{3})(z - \frac{1}{2})}$$

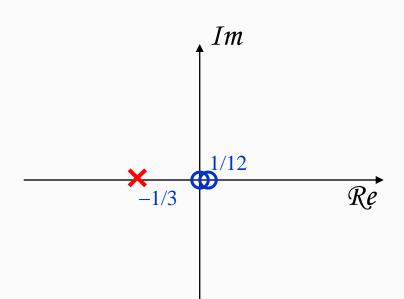




$$x(n) = (-\frac{1}{3})^n u(n) - (\frac{1}{2})^n u(-n-1)$$



$$X(z) = \frac{z}{z + \frac{1}{3}} + \frac{z}{z - \frac{1}{2}} = \frac{2z(z - \frac{1}{12})}{(z + \frac{1}{3})(z - \frac{1}{2})}$$

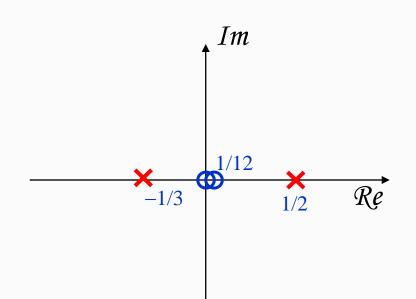




$$x(n) = (-\frac{1}{3})^n u(n) - (\frac{1}{2})^n u(-n-1)$$



$$X(z) = \frac{z}{z + \frac{1}{3}} + \frac{z}{z - \frac{1}{2}} = \frac{2z(z - \frac{1}{12})}{(z + \frac{1}{3})(z - \frac{1}{2})}$$

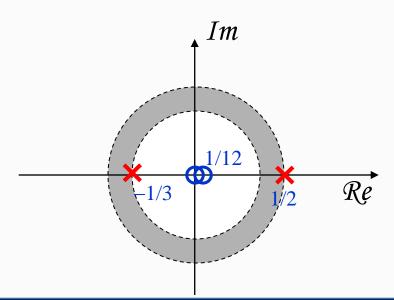




$$x(n) = (-\frac{1}{3})^n u(n) - (\frac{1}{2})^n u(-n-1)$$



$$X(z) = \frac{z}{z + \frac{1}{3}} + \frac{z}{z - \frac{1}{2}} = \frac{2z(z - \frac{1}{12})}{(z + \frac{1}{3})(z - \frac{1}{2})}$$

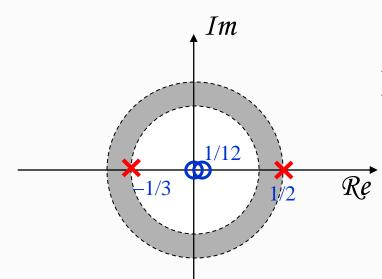




$$x(n) = (-\frac{1}{3})^n u(n) - (\frac{1}{2})^n u(-n-1)$$



$$X(z) = \frac{z}{z + \frac{1}{3}} + \frac{z}{z - \frac{1}{2}} = \frac{2z(z - \frac{1}{12})}{(z + \frac{1}{3})(z - \frac{1}{2})}$$



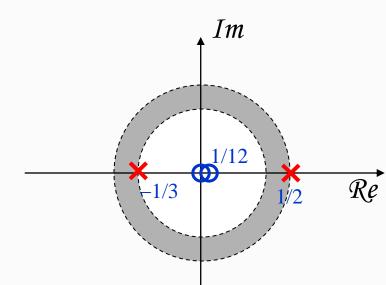
ROC is bounded by poles and is a ring.



$$x(n) = (-\frac{1}{3})^n u(n) - (\frac{1}{2})^n u(-n-1)$$



$$X(z) = \frac{z}{z + \frac{1}{3}} + \frac{z}{z - \frac{1}{2}} = \frac{2z(z - \frac{1}{12})}{(z + \frac{1}{3})(z - \frac{1}{2})}$$



ROC is bounded by poles and is a ring.

ROC does not include any pole.

### Represent z-transform as a Rational Function



$$X(z) = \frac{P(z)}{Q(z)}$$

where P(z) and Q(z) are polynomials in z.

### Represent z-transform as a Rational Function



$$X(z) = \frac{P(z)}{Q(z)}$$

where P(z) and Q(z) are polynomials in z.

Zeros: The values of z's such that X(z) = 0

### Represent z-transform as a Rational Function



$$X(z) = \frac{P(z)}{Q(z)}$$

where P(z) and Q(z) are polynomials in z.

Zeros: The values of z's such that X(z) = 0

Poles: The values of z's such that  $X(z) = \infty$ 

### Sıfır / Kutup Gösterimi



$$\chi(z) = \frac{1 - 0.64z^{-2}}{1 - 0.2z^{-1} - 0.08z^{-2}}$$
 için kutup-sıfır grafiğini çizelim.

### Sıfır / Kutup Gösterimi



$$\chi(z) = \frac{1 - 0.64z^{-2}}{1 - 0.2z^{-1} - 0.08z^{-2}}$$
 için kutup-sıfır grafiğini çizelim.

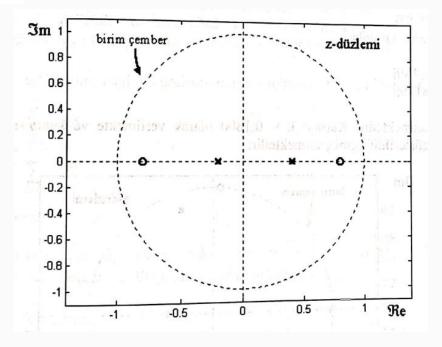
$$X(z) = \frac{\left(1 - 0.8z^{-1}\right)\left(1 + 0.8z^{-1}\right)}{\left(1 - 0.4z^{-1}\right)\left(1 + 0.2z^{-1}\right)}$$

### Sıfır / Kutup Gösterimi



$$\chi(z) = \frac{1 - 0.64z^{-2}}{1 - 0.2z^{-1} - 0.08z^{-2}}$$
 için kutup-sıfır grafiğini çizelim.

$$X(z) = \frac{\left(1 - 0.8z^{-1}\right)\left(1 + 0.8z^{-1}\right)}{\left(1 - 0.4z^{-1}\right)\left(1 + 0.2z^{-1}\right)}$$





- Find z, where
  - FIR only has poles at z=0

$$H(z) \rightarrow \infty$$



- Find z, where
  - FIR only has poles at z=0

$$H(z) \rightarrow \infty$$

$$H(z) = \frac{z^3 - 2z^2 + 2z - 1}{z^3}$$

Three Poles at : z = 0



- Find z, where
  - FIR only has poles at z=0

$$H(z) \rightarrow \infty$$

$$H(z) = \frac{z^3 - 2z^2 + 2z - 1}{z^3}$$

Three Poles at : z = 0

Roots: 
$$z = 1, \frac{1}{2} \pm j \frac{\sqrt{3}}{2}$$



- Find z, where
  - FIR only has poles at z=0

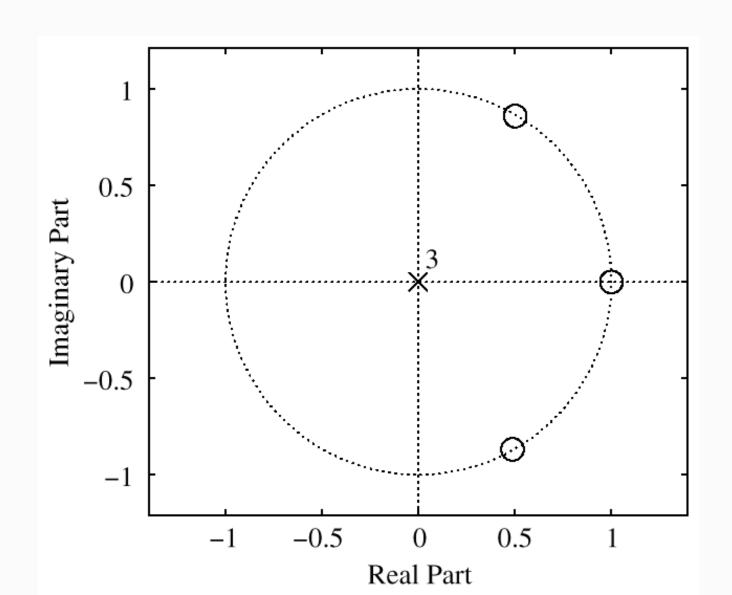
$$H(z) \rightarrow \infty$$

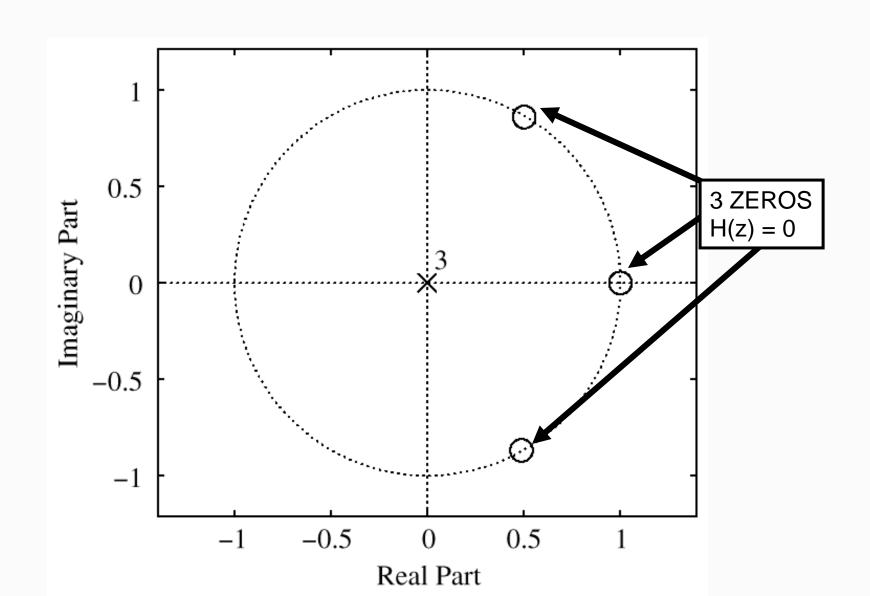
$$H(z) = \frac{z^3 - 2z^2 + 2z - 1}{z^3}$$

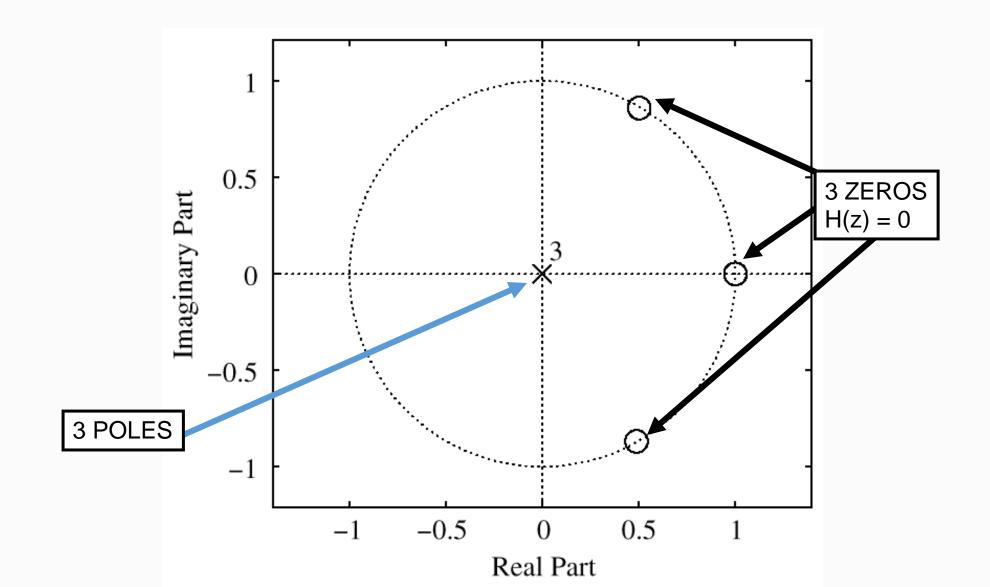
Three Poles at : z = 0

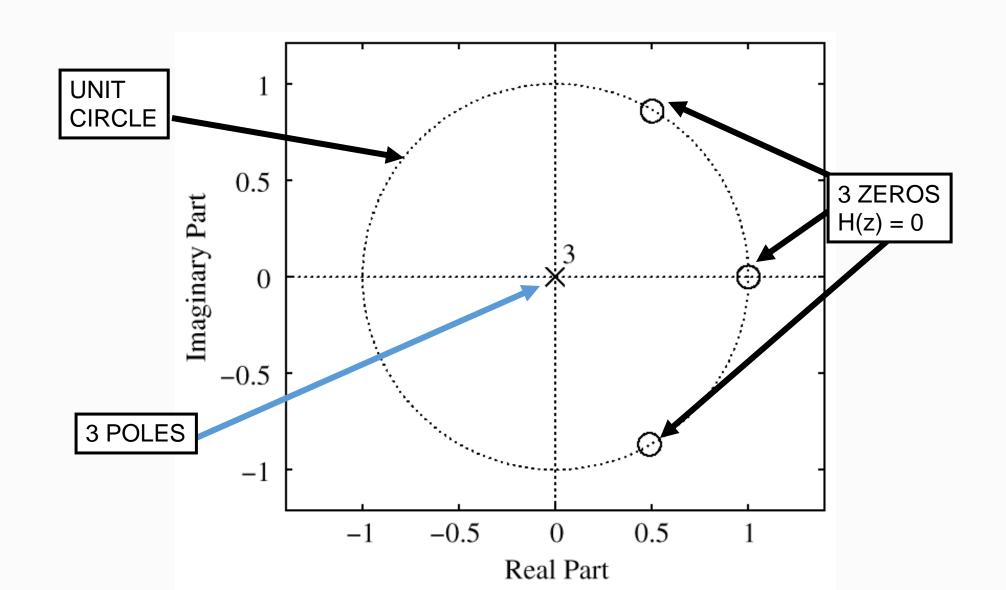
Roots: 
$$z = 1, (\frac{1}{2} \pm j \frac{\sqrt{3}}{2})$$

$$e^{\pm j\pi/3}$$





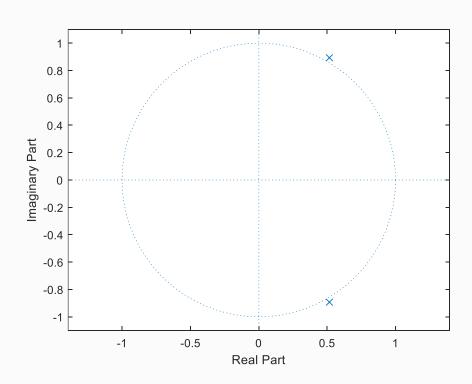




#### MATLAB Code



$$H(z) = \frac{z^2}{z^2 - 0.97z + 0.94}$$

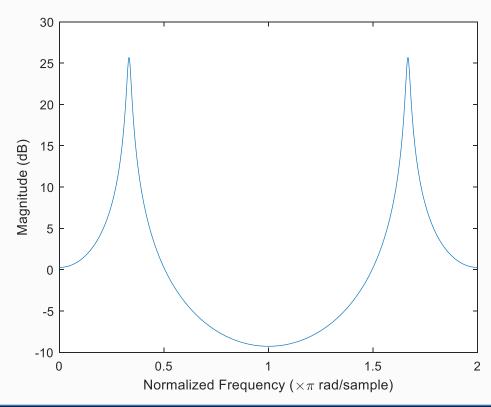


```
%% Z transform
sym z;
Hz = 1/( 1 - 0.97*z^(-1) + 0.94*z^(-2));
[N,D] = numden(Hz);
pay = roots([0 0 100]);
payda = roots([94 -97 100]);
%% Z-plane
zplane(pay, payda);
```

#### MATLAB Code



$$H(z) = \frac{z^2}{z^2 - 0.97z + 0.94}$$

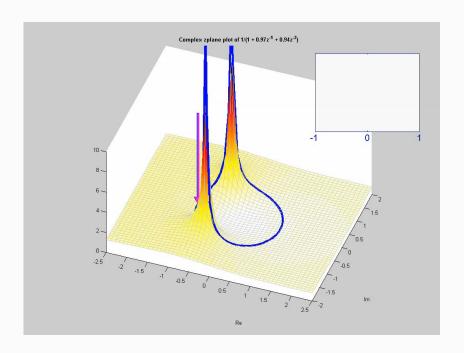


```
%% Z transform
sym z;
Hz = 1/(1 - 0.97*z^{(-1)} + 0.94*z^{(-2)});
[N,D] = numden(Hz);
pay = roots([0 \ 0 \ 100]);
payda = roots([94 - 97 100]);
%% Z-plane
zplane(pay, payda);
%% Freq. Response
[h, w] = freqz([0 \ 0 \ 100], [94 \ -97])
100], 'whole', 2001);
plot(w/pi,20*log10(abs(h)))
ax = qca;
ax.XTick = 0:.5:2;
xlabel('Normalized Frequency')
ylabel('Magnitude (dB)')
```

#### MATLAB Code



$$H(z) = \frac{z^2}{z^2 - 0.97z + 0.94}$$



```
%% Z transform
   syms a b;
   Hz = 1/(1 - 0.97*(a+1j*b)^{(-1)} + 0.94*(a+1j*b)^{(-1)}
   2));
   fsurf(abs(Hz));
   xlim([-1 1]);
   ylim([-1 1]);
   hold on;
   omega = 0:pi/100:2*pi;
   s = \exp(-1j*omega);
   plot3(real(s), imag(s), 2*ones(201,1), 'LineWidth', 3);
```



**EXERCISE 9.2:** Determine the system function H(z) of an FIR filter whose impulse response is

$$h[n] = \delta[n] - 7\delta[n-2] - 3\delta[n-3]$$

McClellan, Schafer, and Yoder, *DSP First*, *2e*, ISBN 0-13-065562-7. Prentice Hall, Upper Saddle River, NJ 07458. ©2016 Pearson Education, Inc.





**EXERCISE 9.2:** Determine the system function H(z) of an FIR filter whose impulse response is

$$h[n] = \delta[n] - 7\delta[n-2] - 3\delta[n-3]$$

#### **SOLUTION to EXERCISE 9.2:**

**DSP First 2e** 

$$H(z) = \sum_{n=0}^{3} h[n] \bar{z}^{n}$$

$$= \sum_{n=0}^{3} (\delta[n] - 7\delta[n-2] - 3\delta[n-3]) \bar{z}^{n}$$

$$= \bar{z}^{0} - 7\bar{z}^{2} - 3\bar{z}^{3}$$

$$H(z) = 1 - 7\bar{z}^{2} - 3\bar{z}^{3}$$

41



$$w[n] = x[n] + x[n-1]$$
  
$$y[n] = w[n] - w[n-1] + w[n-2]$$

into a single difference equation for y[n] in terms of x[n].

McClellan, Schafer, and Yoder, DSP First, 2e, ISBN 0-13-065562-7. Prentice Hall, Upper Saddle River, NJ 07458. ©2016 Pearson Education, Inc.



**EXERCISE 9.5:** Use *z*-transforms to combine the following cascaded systems

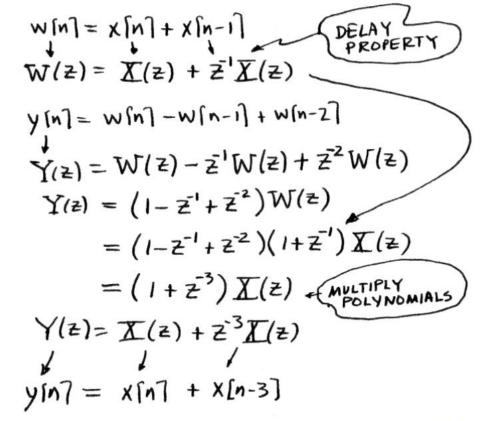
$$w[n] = x[n] + x[n-1]$$
  
$$y[n] = w[n] - w[n-1] + w[n-2]$$

into a single difference equation for y[n] in terms of x[n].

#### **SOLUTION to EXERCISE 9.5:**

DSP First 2e







#### **Example 9-9:** Consider a DTFT function expressed as

$$H(e^{j\hat{\omega}}) = (1 + \cos 2\hat{\omega})e^{-j3\hat{\omega}}$$

Using the inverse Euler formula for the cosine term gives

$$H(e^{j\hat{\omega}}) = \left(1 + \frac{e^{j2\hat{\omega}} + e^{-j2\hat{\omega}}}{2}\right)e^{-j3\hat{\omega}} = (e^{j\hat{\omega}})^{-3} + \frac{1}{2}(e^{j\hat{\omega}})^{-1} + \frac{1}{2}(e^{j\hat{\omega}})^{-5}$$

Making the substitution  $e^{j\hat{\omega}} = z$  gives

$$H(z) = z^{-3} + \frac{1}{2}z^{-1} + \frac{1}{2}z^{-5} = \frac{1}{2}z^{-1} + z^{-3} + \frac{1}{2}z^{-5}$$

If the impulse response of the system is needed, the inverse z-transform gives

$$h[n] = \frac{1}{2}\delta[n-1] + \delta[n-3] + \frac{1}{2}\delta[n-5]$$

McClellan, Schafer, and Yoder, *DSP First*, 2e, ISBN 0-13-065562-7. Prentice Hall, Upper Saddle River, NJ 07458. ©2016 Pearson Education, Inc.

