#### DSP First, 2/e

Lecture 23
Frequency Response,
H(z), Poles and Zeros
for IIR and FIR Systems

#### READING ASSIGNMENTS

- This Lecture:
  - Chapter 9, Sects. 9-5 and 9-6
  - Chapter 10, Sects. 10-5 and 10-7

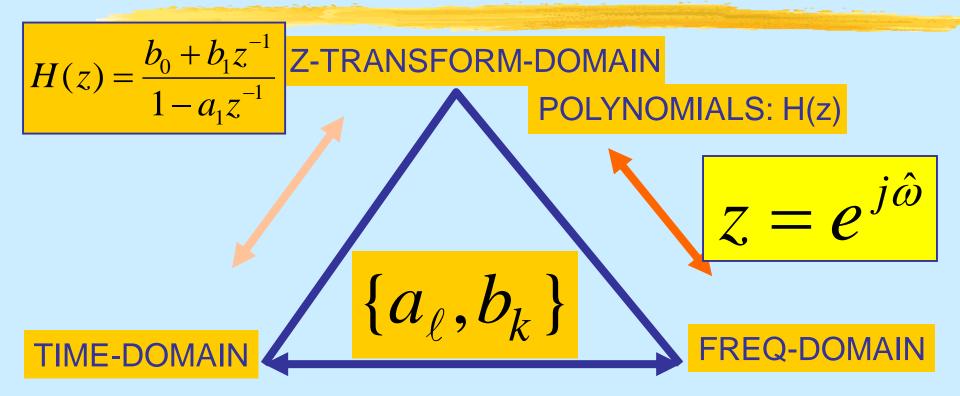
#### LECTURE OBJECTIVES

- ZEROS and POLES
- Relate H(z) to FREQUENCY RESPONSE

$$H(e^{j\hat{\omega}}) = H(z)\big|_{z=e^{j\hat{\omega}}}$$

- Four demos: PeZ, 3-Domain movies
  - Placing Poles and Zeros
- Bandpass Filters: IIR
- Nulling Filters: FIR Notch Filters: IIR

## THREE DOMAINS: $H(e^{j\hat{\omega}})$



$$y[n] = a_1 y[n-1] + b_0 x[n] + b_1 x[n-1]$$

Impulse response, h[n]

$$H(e^{j\hat{\omega}}) = \frac{b_0 + b_1 e^{-j\hat{\omega}}}{1 - a_1 e^{-j\hat{\omega}}}$$

## **Motivation: Filter Design**

- Some tasks/analysis easier in one domain
  - Freq domain: system response to sinusoids
  - Time domain: calculate output to any signal
  - Z-domain: given specs, build a filter
- Can we design a filter that removes DC and sinusoids at frequency  $\hat{\omega} = \pi/3$ ?
- Z-domain reduces this to polynomial roots

## H(z) = Rational Function

First Order:

$$H(z) = \frac{b_0 + b_1 z^{-1}}{1 - a_1 z^{-1}}$$

We can also study Second-Order Systems:

$$H(z) = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2}}{1 - a_1 z^{-1} - a_2 z^{-2}} = \frac{B(z)}{A(z)}$$

Numerator & Denominator Polynomials

## POLES & ZEROS of H(z)

- Zeros of H(z), i.e., where is H(z)=0?
  - Look for Roots of Numerator Polynomial

$$H(z) = \frac{B(z)}{A(z)}$$
, so  $B(z_0) = 0 \Rightarrow H(z_0) = 0$   
if  $A(z_0) \neq 0$ 

- Poles of H(z), i.e., where is H(z)=infinity?
  - Look for Roots of Denominator Polynomial

$$H(z) = \frac{B(z)}{A(z)}$$
, so  $A(z_0) = 0 \Rightarrow H(z_0) \rightarrow \infty$   
if  $B(z_0) \neq 0$ 

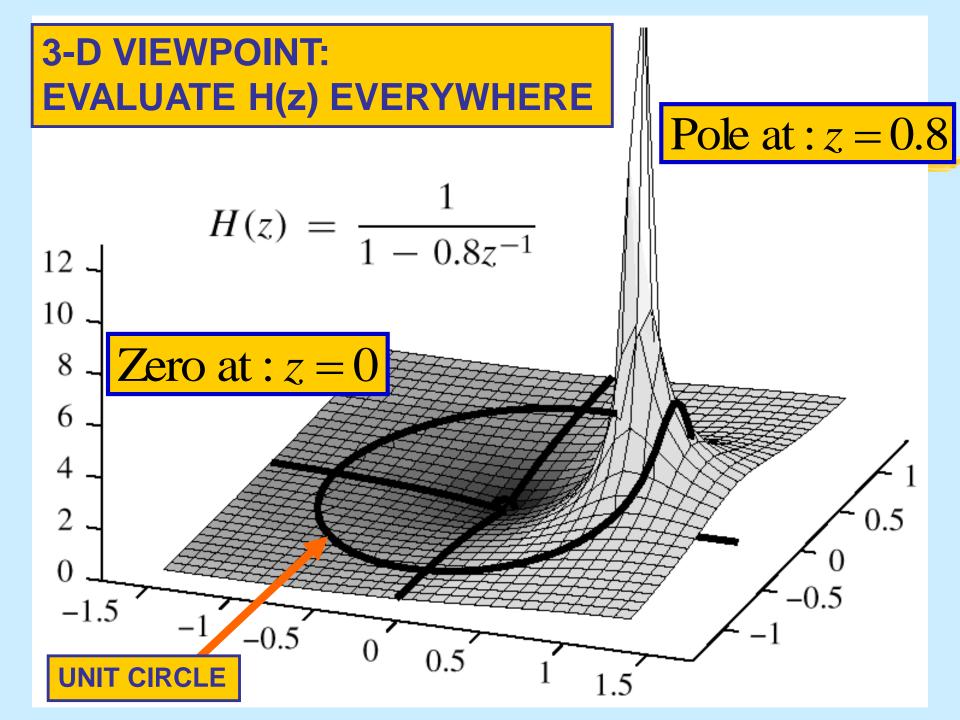
## Poles/Zeros of 1<sup>st</sup>-order H(z)

Roots of Numerator & Denominator Polys:

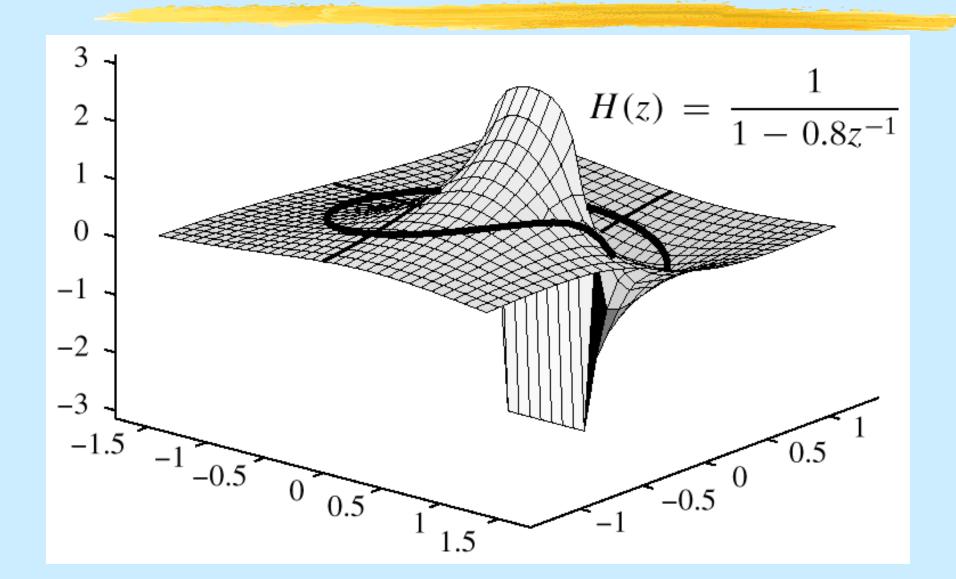
$$H(z) = \frac{1 + b_1 z^{-1}}{1 - 0.8 z^{-1}}$$

$$H(z) = \frac{z(1+b_1z^{-1})}{z(1-0.8z^{-1})} = \frac{z+b_1}{z-0.8}$$

Pole at: z = 0.8 Zero at:  $z = -b_1$ 



#### **PHASE from 3-D PLOT**



#### FREQ. RESPONSE from H(z)

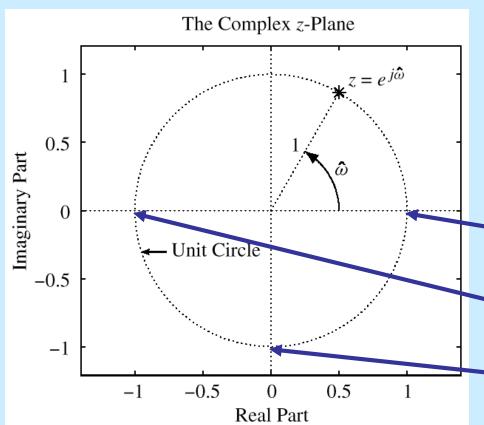
$$H(e^{j\hat{\omega}}) = H(z)\big|_{z=e^{j\hat{\omega}}}$$

- Relate H(z) to FREQUENCY RESPONSE
- EVALUATE H(z) on the <u>UNIT CIRCLE</u>
  - ANGLE is same as FREQUENCY

$$z = e^{j\hat{\omega}}$$
 (as  $\hat{\omega}$  varies)  
defines a CIRCLE, radius = 1

#### **UNIT CIRCLE: RECAP**

## - MAPPING BETWEEN z and $\hat{\omega}$



$$z = e^{j\hat{\omega}}$$

$$z = 1 \quad \leftrightarrow \quad \hat{\omega} = 0$$

$$z = -1 \quad \leftrightarrow \quad \hat{\omega} = \pm \pi$$

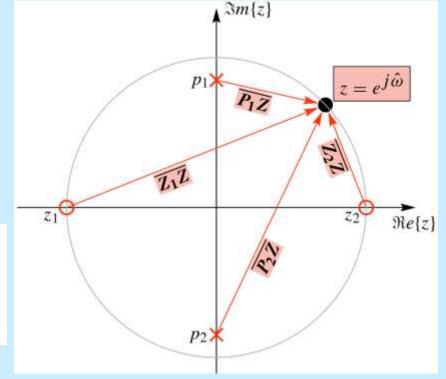
$$z = \pm j \quad \leftrightarrow \quad \hat{\omega} = \pm \frac{1}{2}\pi$$

## Frequency Response from poles and zeros

$$|H(e^{j\hat{\omega}})| = G \frac{|e^{j\hat{\omega}} - z_1| |e^{j\hat{\omega}} - z_2|}{|e^{j\hat{\omega}} - p_1| |e^{j\hat{\omega}} - p_2|}$$

$$|H(e^{j\hat{\omega}})| = G \frac{\overline{Z_1 Z} \cdot \overline{Z_2 Z}}{\overline{P_1 Z} \cdot \overline{P_2 Z}}$$

$$H(z) = G \frac{(z - z_1)(z - z_2)}{(z - p_1)(z - p_2)}$$



## IIR H(z) example: two poles

Poles just inside the unit circle (for stability)

$$H(z) = \frac{1}{1 + 0.97z^{-1} + 0.9409z^{-2}}$$

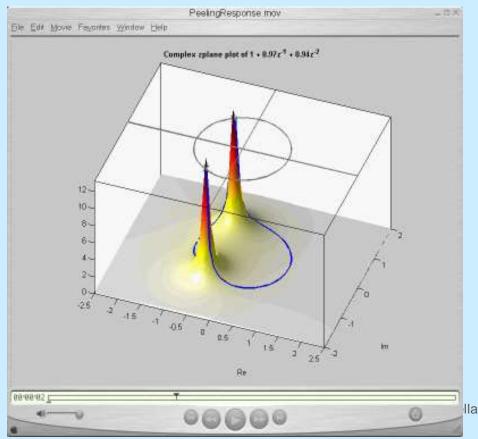
2 Poles : 
$$z = 0.97e^{\pm j2\pi/3}$$
 | 2 Zeros :  $z = 0.0$ 

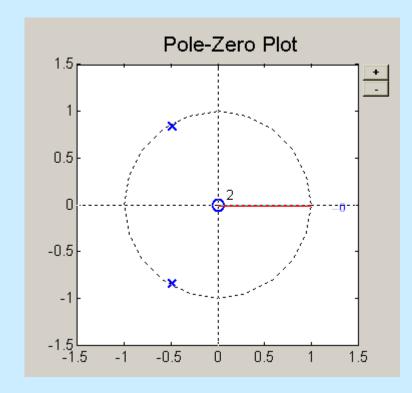
2 Zeros : 
$$z = 0.0$$

- MATLAB: roots ( ) and poly ( )
  - roots([1, 0.97, 0.9409])
  - -poly(0.97\*exp(j\*2\*pi\*[1,-1]/3))

## MOVIE for H(z) in 3-D

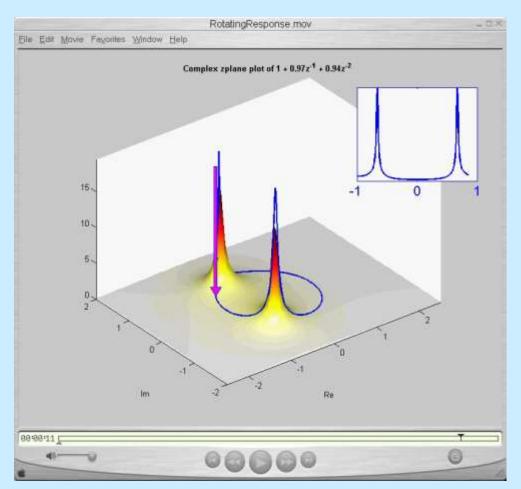
- POLES to H(z) to Frequency Reponse
  - TWO POLES SHOWN



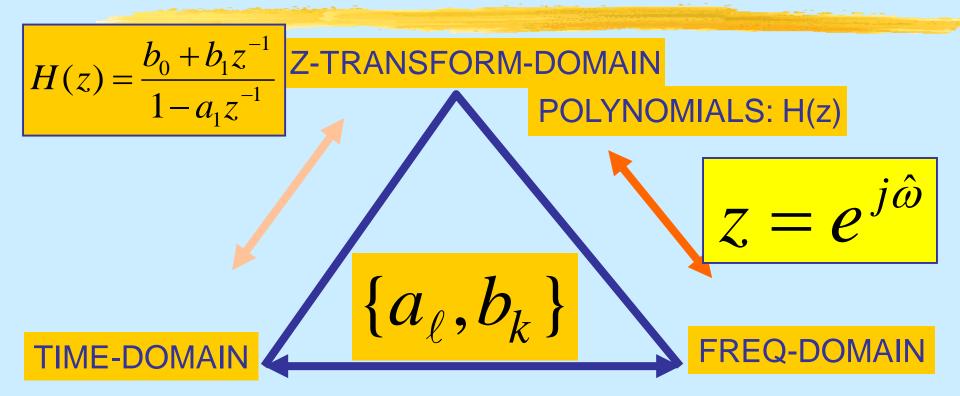


## Frequency Response from H(z)

#### Walking around the Unit Circle



## THREE DOMAINS: $H(e^{j\hat{\omega}})$

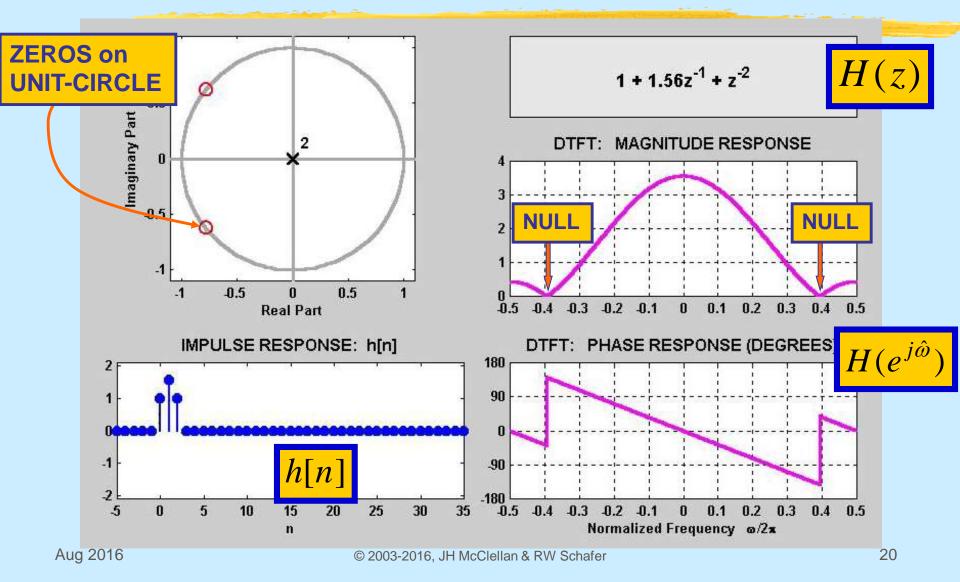


$$y[n] = a_1 y[n-1] + b_0 x[n] + b_1 x[n-1]$$

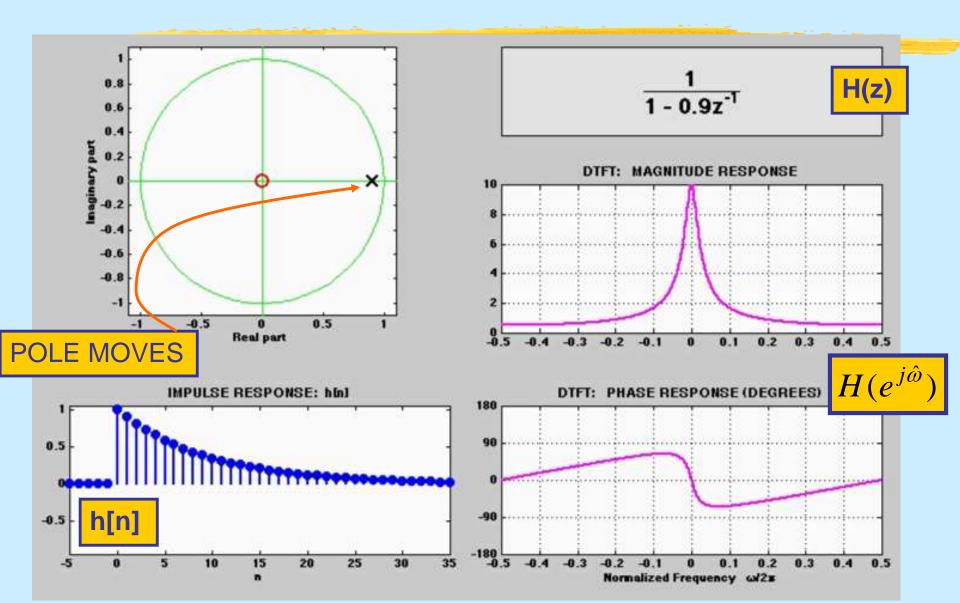
Impulse response, h[n]

$$H(e^{j\hat{\omega}}) = \frac{b_0 + b_1 e^{-j\hat{\omega}}}{1 - a_1 e^{-j\hat{\omega}}}$$

#### 3 DOMAINS MOVIE: FIR



#### 3 DOMAINS MOVIE: IIR



#### 7 IIR MOVIES @ WEBSITE

- http://dspfirst.gatech.edu/chapters/08feedbac/demos/3\_domain/index.html
- 3 DOMAINS MOVIES: <u>IIR</u> Filters
  - One pole moving and a zero at the origin
  - One pole and one zero; both moving
  - Two complex-conjugate poles moving radially
  - Two complex-conjugate poles moving in angle
  - Movement of a zero in a two-pole Filter
  - Radial Movement of Two out of Four Poles
  - Angular Movement of Two out of Four Poles

# Reminder: 4 FIR MOVIES @ WEBSITE

- http://dspfirst.gatech.edu/chapters/08feedbac/demos/3\_domain/index.html
- 3 DOMAINS MOVIES: <u>FIR</u> Filters
  - Two zeros moving around UC and inside
  - Three zeros; one held fixed at z = -1
  - Ten zeros; 9 equally spaced around UC; one moving
  - Ten zeros; 8 equally spaced around UC; two moving

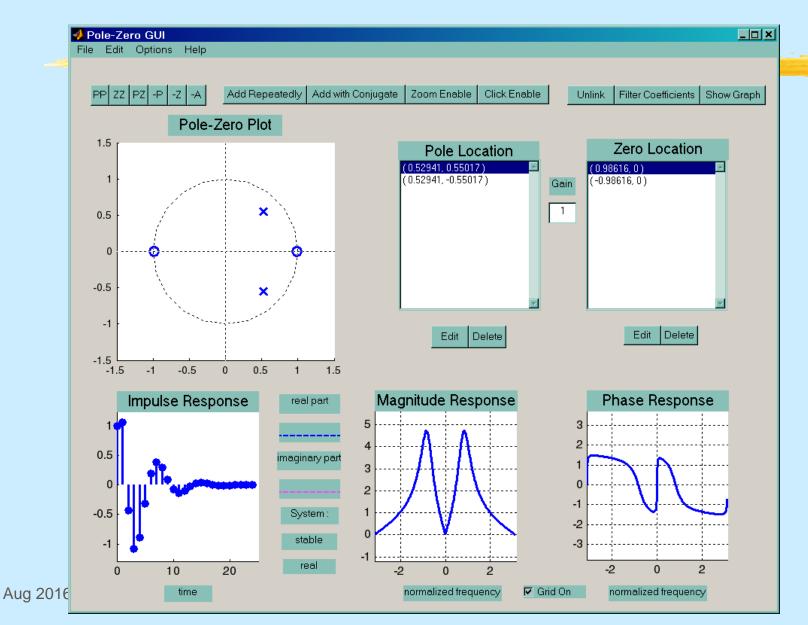
#### Remove Interference

- Design a NOTCH filter (Find a<sub>k</sub> and b<sub>k</sub>)
  - To Reject completely  $0.7\pi$ 
    - This is NULLING
    - Zeros on UC

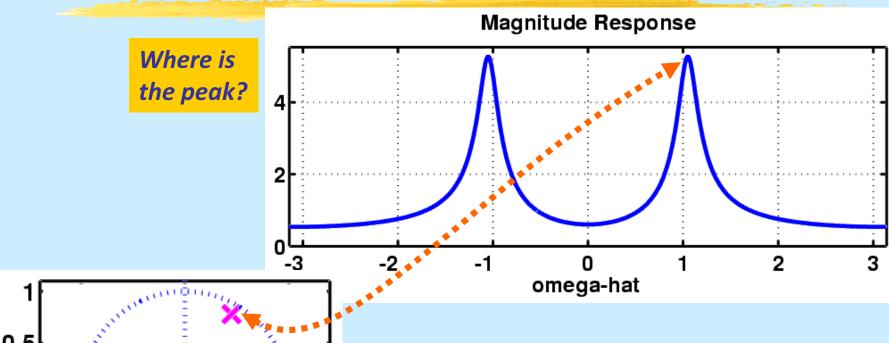
2 Zeros : 
$$z = e^{\pm j0.7\pi}$$

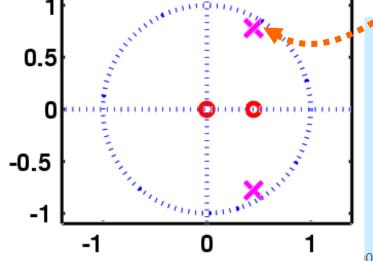
- Make the frequency response magnitude FLAT away from the notch. 2 Poles :  $z = 0.97e^{\pm j0.7\pi}$ 
  - Use poles at the <u>same angle</u>
- Z-POLYNOMIALS provide the TOOLS
  - PEZDEMO GUI

## **PeZ Demo: Pole-Zero Placing**



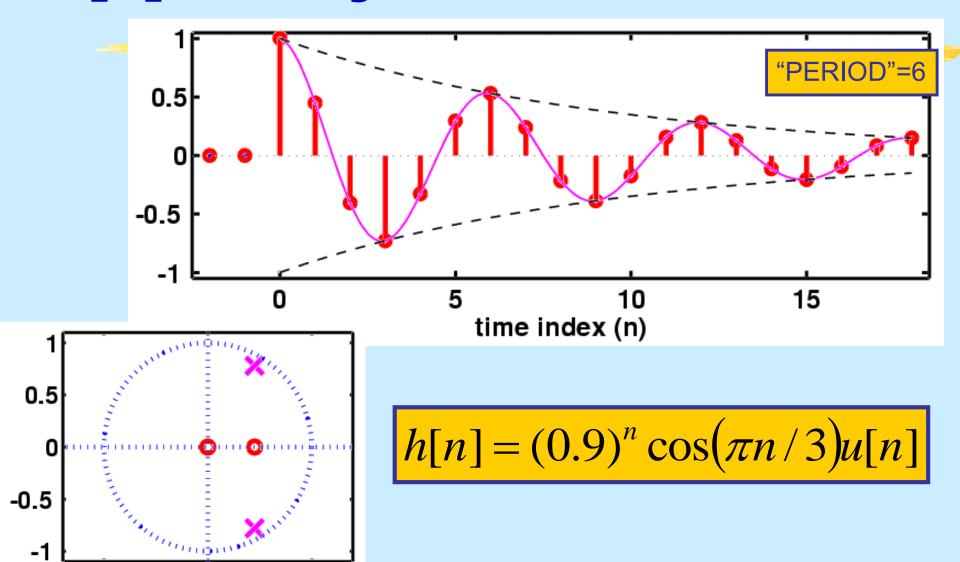
## **Complex POLE-ZERO PLOT**



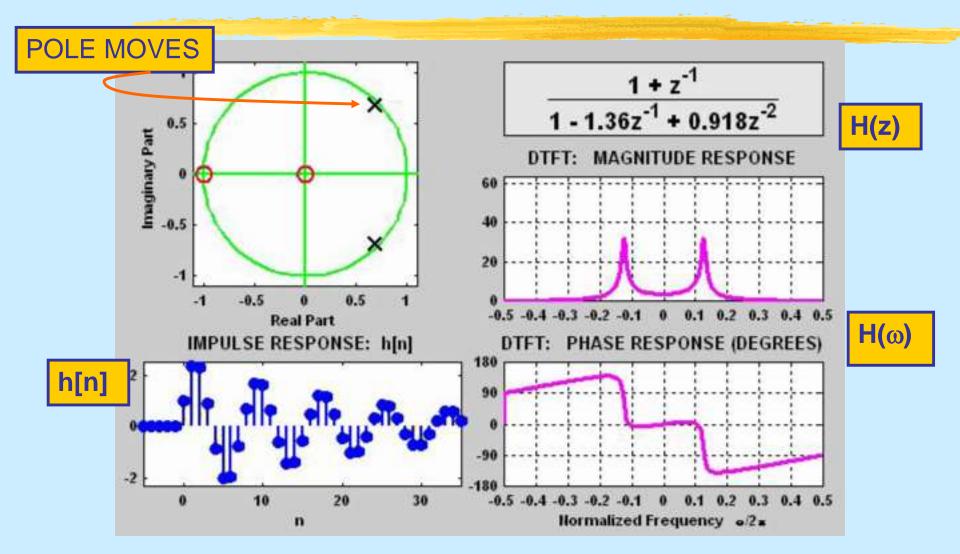


$$H(z) = \frac{1 - 0.45z^{-1}}{1 - 0.9z^{-1} + 0.81z^{-2}}$$

#### h[n]: Decays & Oscillates



#### 3 DOMAINS MOVIE: IIR



#### SINUSOIDAL RESPONSE

- x[n] = SINUSOID => y[n] is SINUSOID
- Get MAGNITUDE & PHASE from H(z)

if 
$$x[n] = e^{j\hat{\omega}n}$$
  
then  $y[n] = H(e^{j\hat{\omega}})e^{j\hat{\omega}n}$   
where  $H(e^{j\hat{\omega}}) = H(z)|_{z=e^{j\hat{\omega}}}$ 

#### POP QUIZ

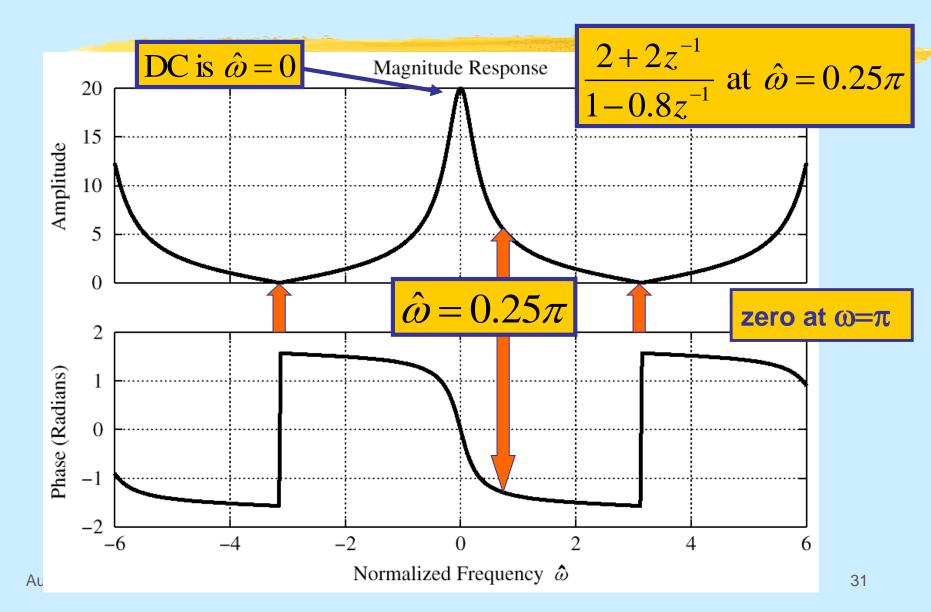
Given:

$$H(z) = \frac{2 + 2z^{-1}}{1 - 0.8z^{-1}}$$

- Find the Impulse Response, h[n]
- Find the output, y[n]

When 
$$x[n] = \cos(0.25\pi n)$$

#### **Evaluate FREQ. RESPONSE**



#### POP QUIZ: Eval Freq. Resp.

Given: 
$$H(z) = \frac{2 + 2z^{-1}}{1 - 0.8z^{-1}}$$

Find output, y[n], when  $x[n] = \cos(0.25\pi n)$ 

$$x[n] = \cos(0.25\pi n)$$

• Evaluate at 
$$z = e^{j0.25\pi}$$

$$H(z) = \frac{2 + 2(\frac{\sqrt{2}}{2} - j\frac{\sqrt{2}}{2})}{1 - 0.8e^{-j0.25\pi}} = 5.182e^{-j1.309}$$

$$y[n] = 5.182\cos(0.25\pi n - 0.417\pi)$$