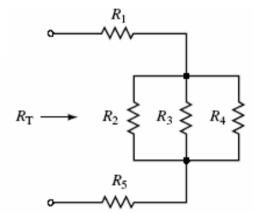
#### The Series-Parallel Network

In electric circuits, we define a **branch** as any portion of a circuit which can be simplified as having two terminals. The components between the two terminals may be any combination of resistors, voltage sources, or other elements. Many complex circuits may be separated into a combination of both series and/or parallel elements, while other circuits consist of even more elaborate combinations which are neither series nor parallel.

In order to analyze a complicated circuit, it is important to be able to recognize which elements are in series and which elements or branches are in parallel. Consider the network of resistors shown in Figure 7–1.

We immediately recognize that the resistors  $R_2$ ,  $R_3$ , and  $R_4$  are in parallel. This parallel combination is in series with the resistors  $R_1$  and  $R_5$ . The total resistance may now be written as follows:

$$R_{\rm T} = R_1 + (R_2 || R_3 || R_4) + R_5$$



**EXAMPLE 7–1** For the network of Figure 7–2, determine which resistors and branches are in series and which are in parallel. Write an expression for the total equivalent resistance,  $R_T$ .

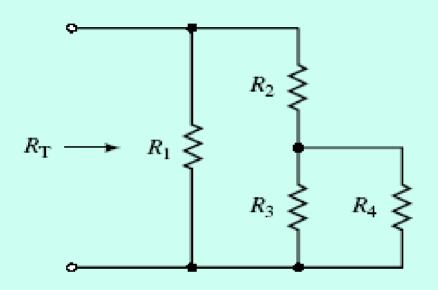


FIGURE 7-2

Solution First, we recognize that the resistors  $R_3$  and  $R_4$  are in parallel:  $(R_3||R_4)$ .

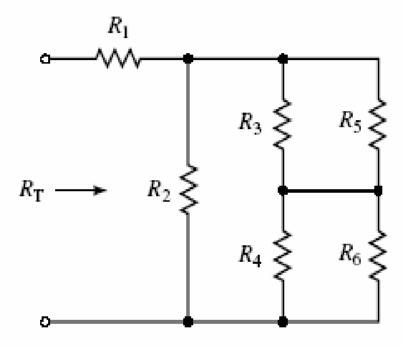
Next, we see that this combination is in series with the resistor  $R_2$ :  $[R_2 + (R_3||R_4)]$ .

Finally, the entire combination is in parallel with the resistor  $R_1$ . The total resistance of the circuit may now be written as follows:

$$R_{\rm T} = R_1 || [R_2 + (R_3 || R_4)]$$

For the network of Figure 7–3, determine which resistors and branches are in series and which are in parallel. Write an expression for the total resistance,  $R_T$ .



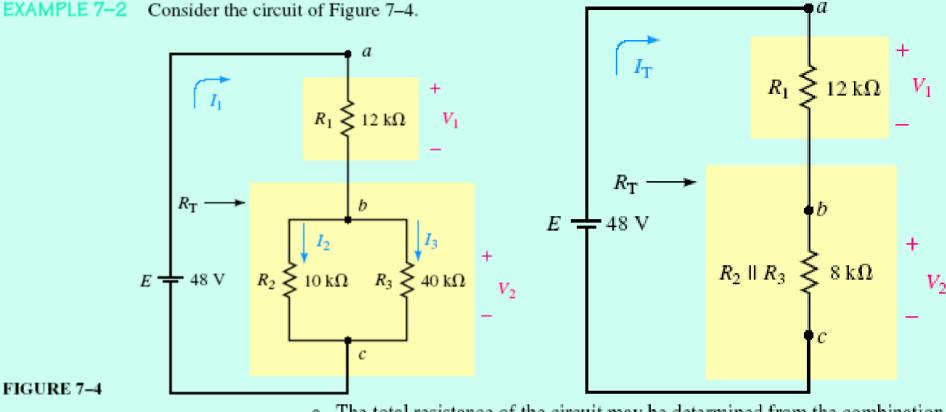


# **Analysis of Series-Parallel Circuits**

Series-parallel networks are often difficult to analyze because they initially appear confusing. However, the analysis of even the most complex circuit is simplified by following some fairly basic steps. By practicing (not memorizing) the techniques outlined in this section, you will find that most circuits can be reduced to groupings of series and parallel combinations. In analyzing such circuits, it is imperative to remember that the rules for analyzing series and parallel elements still apply.

The same current occurs through all series elements.

The same voltage occurs across all parallel elements.



a. The total resistance of the circuit may be determined from the combination a. Find  $R_{\rm T}$ .

b. Calculate 
$$I_1$$
,  $I_2$ , and  $I_3$ .
$$R_T = R_1 + R_2 || R_3$$

$$(10 \text{ k}\Omega)(40 \text{ k}\Omega)$$

c. Determine the voltages 
$$V_1$$
 and  $V_2$ . 
$$R_{\rm T} = 12 \text{ k}\Omega + \frac{(10 \text{ k}\Omega)(40 \text{ k}\Omega)}{10 \text{ k}\Omega + 40 \text{ k}\Omega}$$
$$= 12 \text{ k}\Omega + 8 \text{ k}\Omega = 20 \text{ k}\Omega$$

b. From Ohm's law, the total current is

$$I_{\rm T} = I_1 = \frac{48 \text{ V}}{2010} = 2.4 \text{ m/s}$$

The current  $I_1$  will enter node b and then split between the two resistors  $R_2$ and  $R_3$ . This current divider may be simplified as shown in the partial circuit of Figure 7-6.

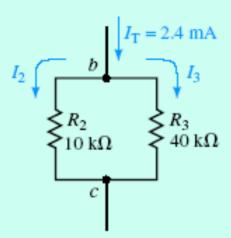


FIGURE 7-6

 $I_2 = \frac{(40 \text{ k}\Omega)(2.4 \text{ mA})}{10 \text{ k}\Omega + 40 \text{ k}\Omega} = 1.92 \text{ mA}$ 

Applying the current divider rule to these two resistors gives

$$I_3 = \frac{(10 \text{ k}\Omega)(2.4 \text{ mA})}{10 \text{ k}\Omega + 40 \text{ k}\Omega} = 0.48 \text{ A}$$
c. Using the above currents and Ohm's law, we determine the voltages:

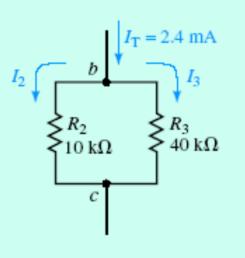
In order to check the answers, we may simply apply Kirchhoff's voltage law  $V_1 = (2.4 \text{ mA})(12 \text{ k}\Omega) = 28.8 \text{ V}$ 

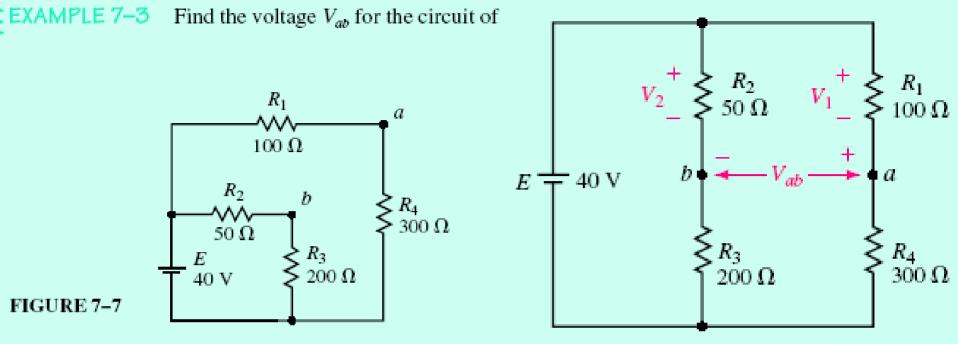
 $V_3 = (0.48 \text{ mA})(40 \text{ k}\Omega) = 19.2 \text{ V} = V_2$ 

around any closed loop which includes the voltage source:  $\sum V = E - V_1 - V_3$ = 48 V - 28.8 V - 19.2 V = 0 V (checks!)

The solution may be verified by ensuring that the power delivered by the voltage source is equal to the summation of powers dissipated by the resistors.

c





From Figure 7–8, we see that the original circuit consists of two parallel branches, where each branch is a series combination of two resistors.

If we take a moment to examine the circuit, we see that the voltage  $V_{ab}$  may be determined from the combination of voltages across  $R_1$  and  $R_2$ . Alternatively, the voltage may be found from the combination of voltages across  $R_3$  and  $R_4$ .

As usual, several methods of analysis are possible. Because the two branches are in parallel, the voltage across each branch must be 40 V. Using the voltage divider rules allows us to quickly calculate the voltage across each resistor. Although equally correct, other methods of calculating the voltages would be more lengthy.

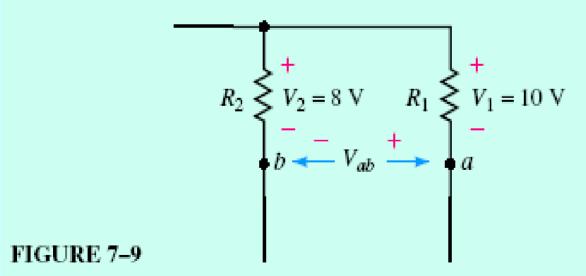
$$V_{2} = \frac{R_{2}}{R_{2} + R_{3}} E$$

$$= \left(\frac{50 \Omega}{50 \Omega + 200 \Omega}\right) (40 \text{ V}) = 8.0 \text{ V}$$

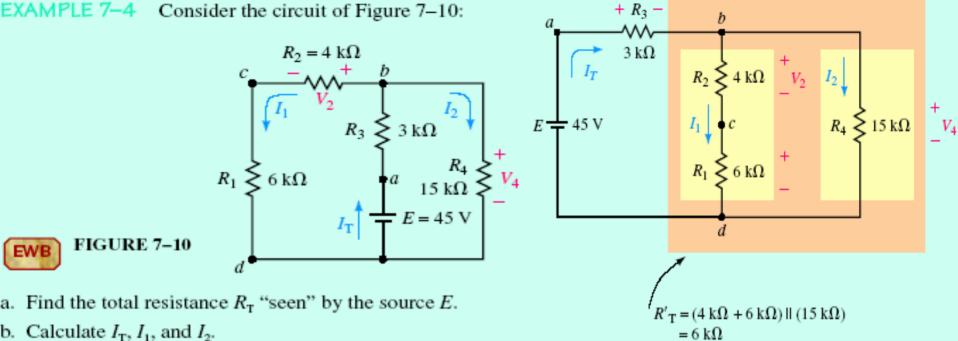
$$V_{1} = \frac{R_{1}}{R_{1} + R_{4}} E$$

$$= \left(\frac{100 \Omega}{100 \Omega + 300 \Omega}\right) (40 \text{ V}) = 10.0 \text{ V}$$

As shown in Figure 7–9, we apply Kirchhoff's voltage law to determine the voltage between terminals a and b.



$$V_{ab} = -10.0 \text{ V} + 8.0 \text{ V} = -2.0 \text{ V}$$



Determine the voltages V<sub>2</sub> and V<sub>4</sub>.

a. From the redrawn circuit, the total resistance of the circuit is

$$R_{\rm T} = R_3 + [(R_1 + R_2)||R_4]$$

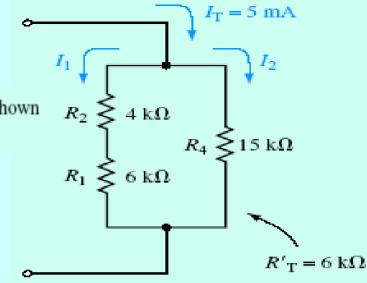
$$= 3 k\Omega + \frac{(4 k\Omega + 6 k\Omega)(15 k\Omega)}{(4 k\Omega + 6 k\Omega) + 15 k\Omega}$$

$$= 3 k\Omega + 6 k\Omega = 9.00 k\Omega$$

The current supplied by the voltage source is

$$I_{\rm T} = \frac{E}{R_{\rm T}} = \frac{45 \text{ V}}{9 \text{ k}\Omega} = 5.00 \text{ mA}$$

We see that the supply current divides between the parallel branches as shown in Figure 7–12.



Applying the current divider rule, we calculate the branch currents as

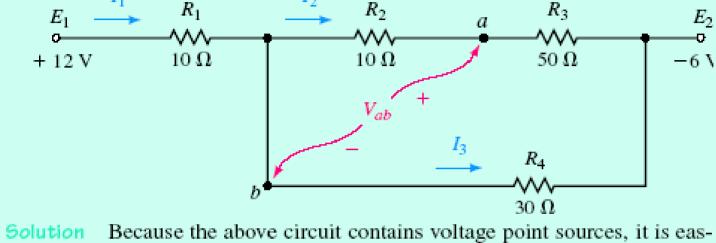
$$I_1 = I_T \frac{R'_T}{(R_1 + R_2)} = \frac{(5 \text{ mA})(6 \text{ k}\Omega)}{4 \text{ k}\Omega + 6 \text{ k}\Omega} = 3.00 \text{ mA}$$
$$I_2 = I_T \frac{R'_T}{R_*} = \frac{(5 \text{ mA})(6 \text{ k}\Omega)}{15 \text{ k}\Omega} = 2.00 \text{ mA}$$

Notice: When determining the branch currents, the resistance  $R'_{\rm T}$  is used in the calculations rather the total circuit resistance. This is because the current  $I_{\rm T}=5$  mA splits between the two branches of  $R'_{\rm T}$  and the split is not affected by the value of  $R_3$ .

c. The voltages  $V_2$  and  $V_4$  are now easily calculated by using Ohm's law:

$$V_2 = I_1 R_2 = (3 \text{ mA})(4 \text{ k}\Omega) = 12.0 \text{ V}$$
  
 $V_4 = I_2 R_4 = (2 \text{ mA})(15 \text{ k}\Omega) = 30.0 \text{ V}$ 

**EXAMPLE 7–5** For the circuit of Figure 7–13, find the indicated currents and voltages.



ier to analyze if we redraw the circuit to help visualize the operation.

The point sources are voltages with respect to ground, and so we begin by drawing a circuit with the reference point as shown in Figure 7–14.

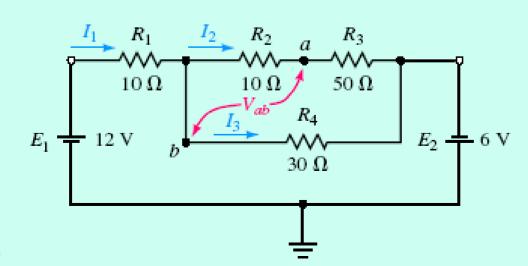
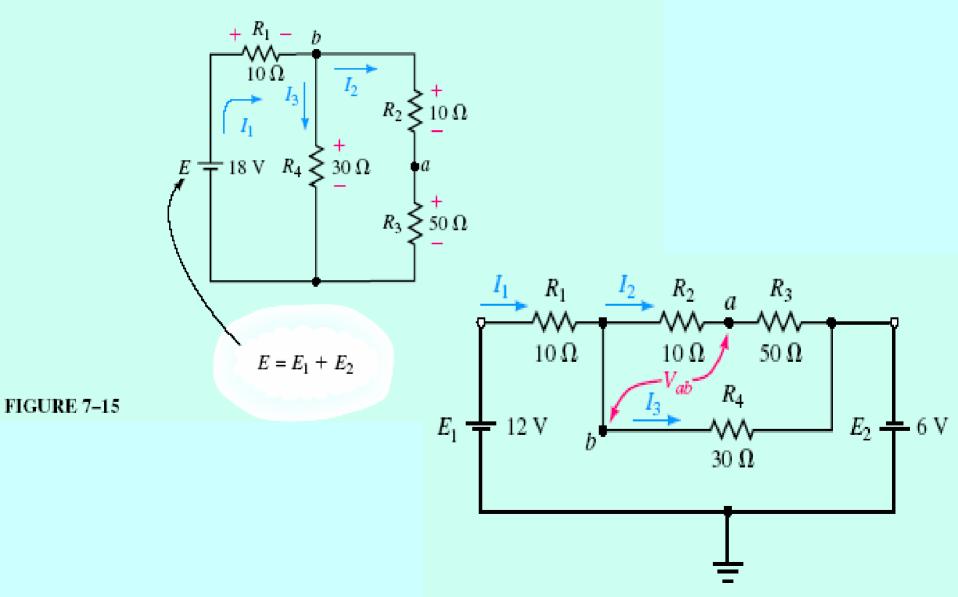


FIGURE 7-14

Now, we can see that the circuit may be further simplified by combining the voltage sources ( $E = E_1 + E_2$ ) and by showing the resistors in a more suitable location. The simplified circuit is shown in Figure 7–15.



The total resistance "seen" by the equivalent voltage source is

tance "seen" by the equivalent voltage source is 
$$R_{\rm T} = R_1 + [R_4 || (R_2 + R_3)]$$

$$= 10 \ \Omega + \frac{(30 \ \Omega)(10 \ \Omega + 50 \ \Omega)}{30 \ \Omega + (10 \ \Omega + 50 \ \Omega)} = 30.0 \ \Omega$$
all current provided into the circuit is
$$I_1 = \frac{E}{R_{\rm T}} = \frac{18 \ \rm V}{30 \ \Omega} = 0.600 \ \rm A$$

And so the total current provided into the circuit is

$$I_1 = \frac{E}{R_T} = \frac{18 \text{ V}}{30 \Omega} = 0.600 \text{ A}$$

At node b this current divides between the two branches as follows:

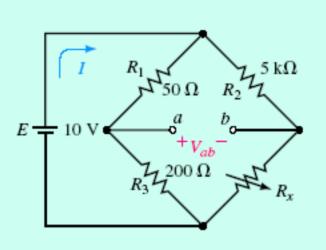
$$I_3 = \frac{(R_2 + R_3)I_1}{R_4 + R_2 + R_3} = \frac{(60 \Omega)(0.600 \text{ A})}{30 \Omega + 10 \Omega + 50 \Omega} = 0.400 \text{ A}$$

$$I_2 = \frac{R_4I_1}{R_4 + R_2 + R_3} = \frac{(30 \Omega)(0.600 \text{ A})}{30 \Omega + 10 \Omega + 50 \Omega} = 0.200 \text{ A}$$

The voltage  $V_{ab}$  has the same magnitude as the voltage across the resistor  $R_2$ , but with a negative polarity (since b is at a higher potential than a):

$$V_{ab} = -I_2 R_2 = -(0.200 \text{ A})(10 \Omega) = -2.0 \text{ V}$$

# **EXAMPLE 7–6** The circuit of Figure 7–17 is referred to as a *bridge circuit* and is used extensively in electronic and scientific instruments.



#### FIGURE 7-17

Calculate the current I and the voltage  $V_{ab}$  when

- a.  $R_x = 0 \Omega$  (short circuit)
- b.  $R_x = 15 \text{ k}\Omega$
- c.  $R_x = \infty$  (open circuit)

The voltage source "sees" a total resistance of

resulting in a source current of

$$I = \frac{10 \text{ V}}{238 \Omega} = 0.042 \text{ A} = 42.2 \text{ mA}$$

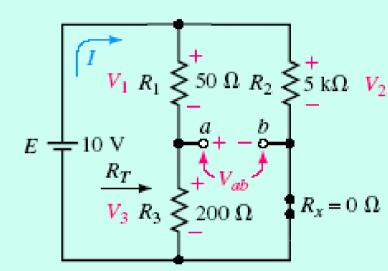
Now, since the variable resistor is a short circuit, the entire source voltage will appear across the resistor  $R_2$ , giving

$$V_2 = 10.0 \text{ V}$$

#### Solution

a.  $R_x = 0 \Omega$ :

The circuit is redrawn as shown in Figure 7–18.



The voltage  $V_{ab}$  may be determined by solving for voltage across  $R_1$  and  $R_2$ .

 $R_{\rm T} = (R_1 + R_3) ||R_2| = 250 \ \Omega ||5000 \ \Omega = 238 \ \Omega$  The voltage across  $R_{\rm T}$  will be constant regardless of the value of the variable resistor  $R_{\rm T}$ . Hence

$$V_1 = \left(\frac{50 \Omega}{50 \Omega + 200 \Omega}\right) (10 \text{ V}) = 2.00 \text{ V}$$

$$V_{ab} = -V_1 + V_2 = -2.00 \text{ V} + 10.0 \text{ V} = +8.00 \text{ V}$$

b. 
$$R_x = 15 \text{ k}\Omega$$
:

FIGURE 7-19

The circuit is redrawn in Figure 7–19.

$$R_{\rm T} = (R_1 + R_3)||(R_2 + R_x)||$$
  
= 250  $\Omega$ ||20 k $\Omega$  = 247  $\Omega$ 

$$E = \begin{bmatrix} I & & & & \\ V_1 & R_1 & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & &$$

$$I = \frac{10 \text{ V}}{247 \Omega} = 0.0405 \text{ A} = 40.5 \text{ mA}$$

The voltages across  $R_1$  and  $R_2$  are

$$V_1 = 2.00 \text{ V} \quad \text{(as before)}$$

$$V_2 = \frac{R_2}{R_2 + R_x} E$$

$$= \left(\frac{5 \text{ k}\Omega}{5 \text{ k}\Omega + 15 \text{ k}\Omega}\right) (10 \text{ V}) = 2.50 \text{ V}$$

Now the voltage between terminals a and b is found as

$$V_{ab} = -V_1 + V_2$$
  
= -2.0 V + 2.5 V = +0.500 V

c.  $R_x = \infty$ :

The circuit is redrawn in Figure 7–20.

The voltages across  $R_1$  and  $R_2$  are

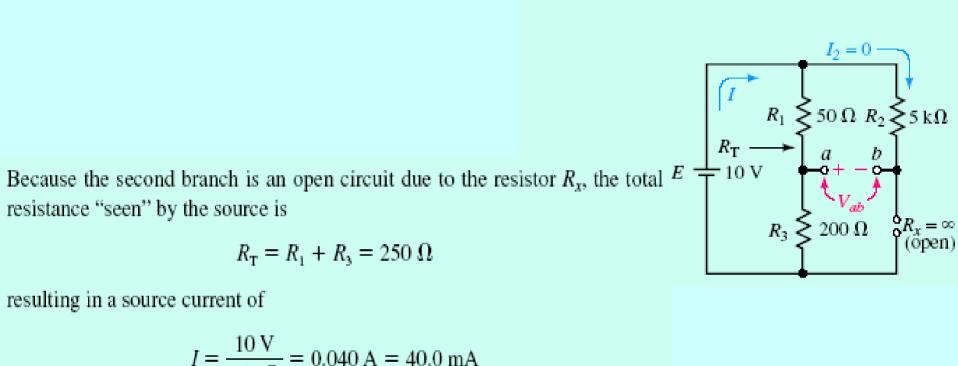
resulting in a source current of

 $V_1 = 2.00 \,\mathrm{V}$  (as before)

 $I = \frac{10 \text{ V}}{250 \Omega} = 0.040 \text{ A} = 40.0 \text{ mA}$ 

 $V_2 = 0 \text{ V}$  (since the branch is open)

And so the resulting voltage between terminals a and b is  $V_{ab} = -V_1 + V_2$ = -2.0 V + 0 V = -2.00 V



The previous example illustrates how voltages and currents within a circuit are affected by changes elsewhere in the circuit. In the example, we saw that the voltage  $V_{ab}$  varied from -2 V to +8 V, while the total circuit current varied from a minimum value of 40 mA to a maximum value of 42 mA. These changes occurred even though the resistor  $R_x$  varied from  $0 \Omega$  to  $\infty$ .

A transistor is a three-terminal device which may be used to amplify small signals. In order for the transistor to operate as an amplifier, however, certain dc conditions must be met. These conditions set the "bias point" of the transistor. The bias current of a transistor circuit is determined by a dc voltage source and several resistors. Although the operation of the transistor is outside the scope of this textbook, we can analyze the bias circuit of a transistor using elementary circuit theory.

### **Potentiometers**

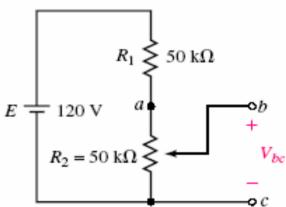
As mentioned in Chapter 3, variable resistors may be used as potentiometers as shown in Figure 7–28 to control voltage into another circuit.

The volume control on a receiver or amplifier is an example of a variable resistor used as a potentiometer. When the movable terminal is at the uppermost position, the voltage appearing between terminals b and c is simply calculated by using the voltage divider rule as

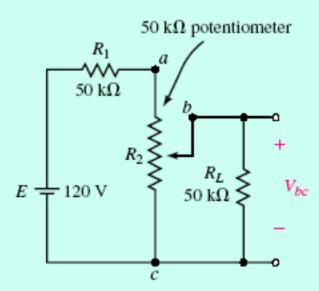
$$V_{bc} = \left(\frac{50 \text{ k}\Omega}{50 \text{ k}\Omega + 50 \text{ k}\Omega}\right) (120 \text{ V}) = 60 \text{ V}$$

Alternatively, when the movable terminal is at the lowermost position, E = 120 V the voltage between terminals b and c is  $V_{bc} = 0 \text{ V}$ , since the two terminals are effectively shorted and the voltage across a short circuit is always zero.

The circuit of Figure 7–28 represents a potentiometer having an output voltage which is adjustable between 0 and 60 V. This output is referred to as the **unloaded output**, since there is no load resistance connected between the terminals b and c. If a load resistance were connected between these terminals, the output voltage, called the **loaded output**, would no longer be the same. The following example is an illustration of circuit loading.



EXAMPLE 7–10 For the circuit of Figure 7–29, determine the range of the voltage  $V_{bc}$  as the potentiometer varies between its minimum and maximum values.



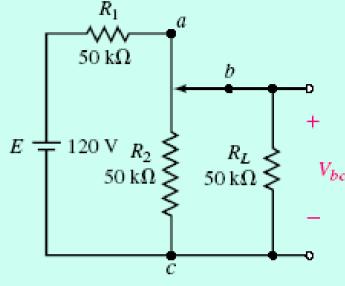


FIGURE 7-29

Solution The minimum voltage between terminals b and c will occur when the movable contact is at the lowermost contact of the variable resistor. In this position, the voltage  $V_{bc} = 0$  V, since the terminals b and c are shorted.

The maximum voltage  $V_{bc}$  occurs when the movable contact is at the uppermost contact of the variable resistor. In this position, the circuit may represented as shown in Figure 7–30.

$$V_{bc} = \frac{R_2 || R_L}{(R_2 || R_L) + R_1} E$$

$$= \left( \frac{25 \text{ k}\Omega}{25 \text{ k}\Omega + 50 \text{ k}\Omega} \right) (120 \text{ V}) = 40 \text{ V}$$

We conclude that the voltage at the output of the potentiometer is adjustable from 0 V to 40 V for a load resistance of  $R_L = 50 \text{ k}\Omega$ .

By inspection, we see that an unloaded potentiometer in the circuit of Figure 7-29 would have an output voltage of 0 V to 60 V.

## **Loading Effects of Instruments**

The degree to which the circuits are affected is called the **loading effect** of the instrument.

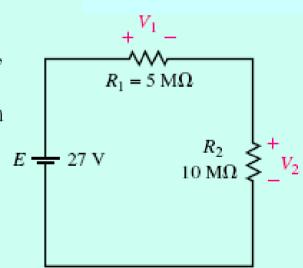
for an instrument to provide an accurate indication of how a circuit operates, the loading effect should ideally be zero. In practice, it is impossible for any instrument to have zero loading effect, since all instruments absorb some energy from the circuit under test, thereby affecting circuit operation.

EXAMPLE 7–11 Calculate the loading effects if a digital multimeter, having an internal resistance of  $10 \text{ M}\Omega$ , is used to measure  $V_1$  and  $V_2$  in the circuit of Figure 7–31.

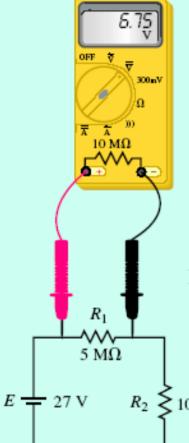
Solution In order to determine the loading effect for a particular reading, we need to calculate both the unloaded voltage and the loaded voltage.

For the circuit given in Figure 7-31, the unloaded voltage across each resistor is

$$V_1 = \left(\frac{5 \text{ M}\Omega}{5 \text{ M}\Omega + 10 \text{ M}\Omega}\right) (27 \text{ V}) = 9.0 \text{ V}$$
$$V_2 = \left(\frac{10 \text{ M}\Omega}{5 \text{ M}\Omega + 10 \text{ M}\Omega}\right) (27 \text{ V}) = 18.0 \text{ V}$$



When the voltmeter is used to measure  $V_1$ , the result is equivalent to connecting a  $10\text{-}\mathrm{M}\Omega$  resistor across resistor  $R_1$ , as shown in Figure 7–32.



The voltage appearing across the parallel combination of  $R_1$  and resistance of the voltmeter is calculated as

$$\begin{split} V_1 &= \left(\frac{5 \text{ M}\Omega \| 10 \text{ M}\Omega}{(5 \text{ M}\Omega \| 10 \text{ M}\Omega) + 10 \text{ M}\Omega}\right) (27 \text{ V}) \\ &= \left(\frac{3.33 \text{ M}\Omega}{13.3 \text{ M}\Omega}\right) (27 \text{ V}) \\ &= 6.75 \text{ V} \end{split}$$

Notice that the measured voltage is significantly less than the 9 V that we had expected to measure.

With the voltmeter connected across resistor  $R_2$ , the circuit appears as shown in Figure 7–33.

The voltage appearing across the parallel combination of  $R_2$  and resistance of the voltmeter is calculated as

$$\begin{split} V_2 &= \left(\frac{10 \text{ M}\Omega \| 10 \text{ M}\Omega}{5 \text{ M}\Omega + (10 \text{ M}\Omega \| 10 \text{ M}\Omega)}\right) (27 \text{ V}) \\ &= \left(\frac{5.0 \text{ M}\Omega}{10.0 \text{ M}\Omega}\right) (27 \text{ V}) \\ &= 13.5 \text{ V} \end{split}$$

Again, we notice that the measured voltage is quite a bit less than the 18 V that we had expected.

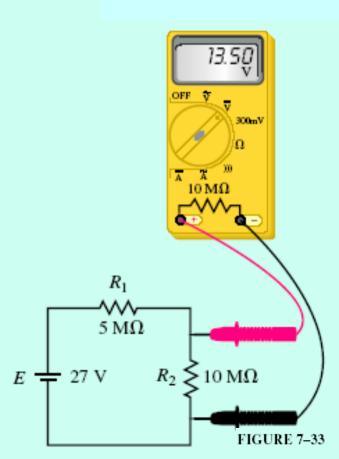
Now the loading effects are calculated as follows.

When measuring  $V_1$ :

loading effect = 
$$\frac{9.0 \text{ V} - 6.75 \text{ V}}{9.0 \text{ V}} \times 100\%$$
$$= 25\%$$

When measuring  $V_2$ :

loading effect = 
$$\frac{18.0 \text{ V} - 13.5 \text{ V}}{18.0 \text{ V}} \times 100\%$$
  
= 25%



This example clearly illustrates a problem that novices often make when they are taking voltage measurements in high-resistance circuits. If the measured voltages  $V_1 = 6.75$  V and  $V_2 = 13.50$  V are used to verify Kirchhoff's voltage law, the novice would say that this represents a contradiction of the law (since 6.75 V + 13.50 V  $\neq 27.0$  V). In fact, we see that the circuit is behaving exactly as predicted by circuit theory. The problem occurs when instrument limitations are not considered.

#### PRACTICAL NOTES...

Whenever an instrument is used to measure a quantity, the operator must always consider the loading effects of the instrument.