

EEM 323

ELECTROMAGNETIC WAVE THEORY II

TIME VARYING FIELDS

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Önemli not: Ders notlarındaki şekillerin hazırlanmasında internet ortamından faydalانılmıştır. Özellikle belirtilmeyen tüm şekil, tablo, eşitlik ve denklemler vb. “D. K, Fundamentals of Engineering Electromagnetics, Addison-Wesley Inc.” ile “D. K, Field and Wave Electromagnetics, Mc-Graw Hill Inc.” kitabından taranarak elde edilmiştir. Alıntıların kaynağına kolay ulaşılabilmesi maksadıyla numarası ve altyazılıları da gösterilmektedir.

DERS KİTABI

- [1] David Keun Cheng, *Fundamentals of Engineering Electromagnetics*, Addison-Wesley Publishing, Inc., 1993.
veya David Keun Cheng, Çeviri: Adnan Köksal, Birsen Saka, *Mühendislik Elektromanyetiginin Temelleri – Fundamentals of Engineering Electromagnetics*, Palme Yayıncılık.

KAYNAK / YARDIMCI KİTAPLAR:

- [2] David Keun Cheng, *Field and Wave Electromagnetics*, Addison-Wesley Publishing, Inc. veya David Keun Cheng, Çeviri: Mithat İdemen, *Elektromanyetik Alan Teorisinin Temelleri – Field and Wave Electromagnetics*, Literatür Yayıncılık.
- [3] Stanley V. Marshall, Richard E. DuBroff, Gabriel G. Skitek, *Electromagnetic Concepts and Applications*, Dördüncü Basım, Prentice Hall International, Inc., 1996.
- [4] Joseph A. Edminister, Elektromanyetik, 2. Baskıdan çeviri, Çevirenler: M. Timur Aydemir, E. Afacan, K. C. Nakipoğlu, Schaum's Outlines, McGraw Hill Inc., Nobel Yayın Dağıtım, Ankara, 2000.

TIME VARYING FIELDS AND MAXWELL'S EQUATIONS

Static \bar{E}

$$\nabla \times \bar{E} = 0$$

$$\nabla \cdot \bar{D} = \rho$$

$$\bar{D} = \epsilon \bar{E} \quad (\text{linear media})$$

Static \bar{H}

$$\nabla \cdot \bar{B} = 0$$

$$\nabla \times \bar{H} = \bar{J}$$

$$\bar{B} = \mu \bar{H}$$

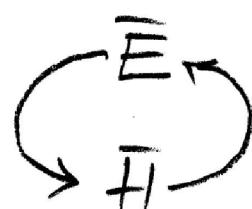
Static CASE
Uncoupled Equations

ρ constant $\rightarrow \bar{E}$ —

\bar{J} constant $\rightarrow \bar{H}$ unrelated
Independent

ρ : time varying $\rightarrow \bar{E}$ is time varying

\bar{J} : time varying $\rightarrow \bar{H}$ is time varying



\bar{E} and \bar{H} are
governed by
coupled equations.

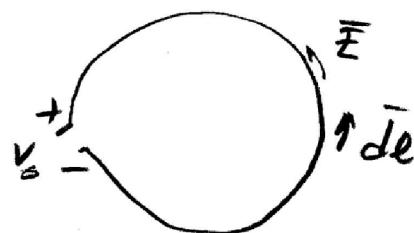
EMF - Electromotive Force (V)
AND FARADAY'S LAW OF INDUCTION

Static E and H

$$\oint \vec{E} \cdot d\vec{l} = 0$$

| E is conservative.

$$\Rightarrow V_0 = 0$$



FARADAY'S INDUCTION LAW

$$\nabla \times \vec{E} = - \frac{\partial}{\partial t} \vec{B} \quad \dots (1)$$

$$\nabla \times \vec{E} \neq 0 \Rightarrow \vec{E} \neq \nabla V$$

$$\text{Why?} \Rightarrow \nabla \times \nabla V = 0 !$$

FARADAY LAW OF EM INDUCTION

Integrate Egn. (1)



$$\int_S \nabla \times \vec{E} \cdot d\vec{s} = \boxed{\oint \vec{E} \cdot d\vec{l} = - \int_S \frac{\partial \vec{B}}{\partial t} \cdot d\vec{s}}$$

$$\text{Flux: } \Phi = \int \vec{B} \cdot d\vec{s} \Rightarrow -N \frac{d\Phi}{dt}$$

67A

EXAMPLE

Stationary Circuit - Time-varying Magnetic field

$$\mathcal{V} = \oint_C \vec{E} \cdot d\vec{l} = - \frac{d}{dt} \int_S \vec{B} \cdot d\vec{s}$$

EMF

$$\text{EMF, } \mathcal{V} = - \frac{d\phi}{dt}$$

* Induction Motor

* Current measurement loop
- clamp meter

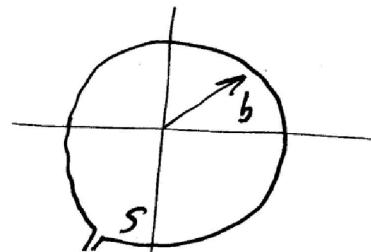
OR loop of N turns, xy plane,

$$\vec{B} = \hat{a}_z B_0 \cos\left(\frac{\pi}{2b} r\right) \sin(\omega t)$$

EMF=?

$$\phi = \int_S \vec{B} \cdot d\vec{l} = \int_0^b \left(\hat{a}_z B_0 \cos\left(\frac{\pi}{2b} r\right) \sin \omega t \right) \cdot (\hat{a}_z 2\pi r dr)$$

$$= \frac{8b^2}{\pi} \left(\frac{\pi}{2} - 1\right) B_0 \sin(\omega t)$$

 N turns ...

$$\mathcal{V} = -N \frac{d\phi}{dt}$$

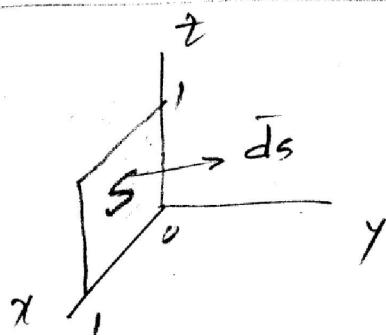
$$\text{EMF, } \mathcal{V} = - \frac{8N}{\pi} b^2 \left(\frac{\pi}{2} - 1\right) B_0 \cdot \omega \cdot \cos(\omega t) \quad (V)$$

67B

EXAMPLE

loop of N turns on the $x-z$ plane

$$\bar{B} = \hat{a}_z B_0 \cos\left(\frac{\pi}{2b} r\right) \sin(\omega t)$$



$$\bar{ds} = \hat{a}_y dx dz$$

$$\bar{B} \cdot \bar{ds} = 0$$

$$\Phi = \int_S \bar{B} \cdot \bar{ds} = 0 \quad (\text{v})$$

No induction

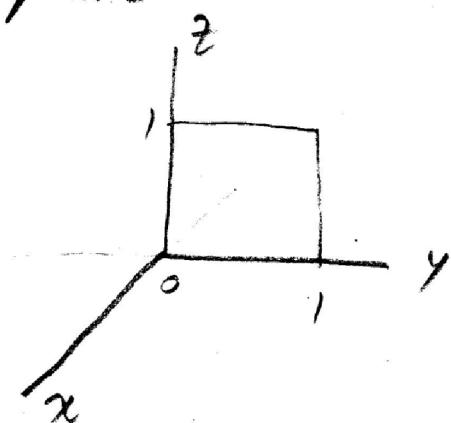
EXAMPLE

N turns, S on the $y-z$ plane

$$\bar{B} = \hat{a}_x B_0 \sin(\omega t)$$

$$+ \hat{a}_y B_1 \sin(\omega t)$$

$$+ \hat{a}_z B_2 \sin(\omega t)$$



$$\bar{ds} = \hat{a}_x dy dz$$

67c

$$\bar{B} \cdot \bar{ds} = \bar{B} \cdot \hat{a}_x dy dz$$

$$= B_0 \sin(\omega t) dy dz$$

$$\Phi = \int_S \bar{B} \cdot \bar{ds}$$

$$= \iiint_0^l B_0 \sin(\omega t) dy dz$$

$$= B_0 \sin(\omega t)$$

$$\text{EMF} = \mathcal{V} = -N \frac{d\Phi}{dt} = -\omega N B_0 \cos(\omega t) \text{ (V)}$$

67D

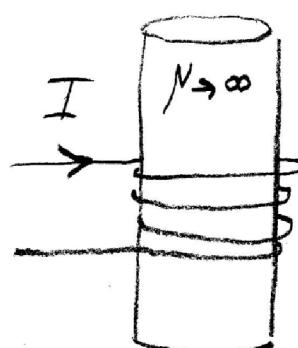
Exercise 6.1 Omitted lecture notes

$$\int \bar{B} \cdot \bar{ds} = \Phi \quad \text{Flux (magnetic)}$$

For linear media: $\bar{B} = \mu \bar{H}$

$$\int \mu \bar{H} \cdot \bar{ds} = \Phi$$

$$\int \bar{H} \cdot \bar{ds} = \left(\frac{1}{\mu} \right)_R \Phi$$



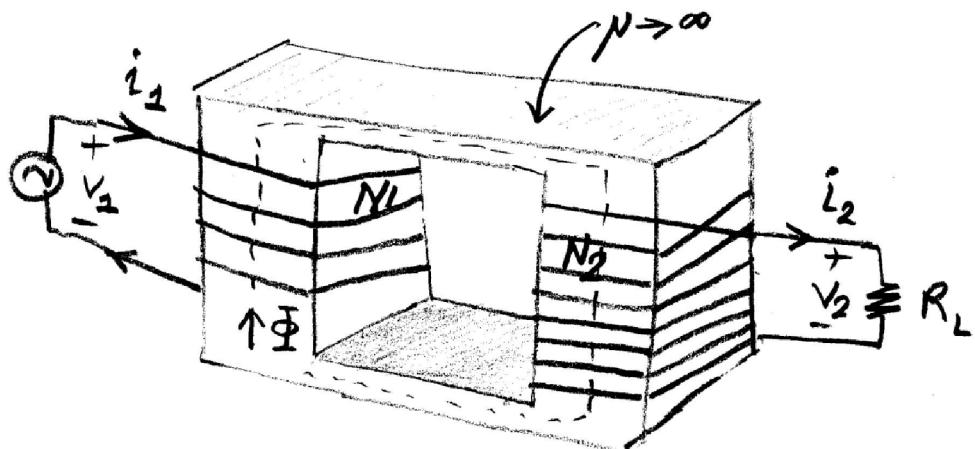
$$\int \bar{H} \cdot \bar{ds} = \oint \bar{H} \cdot \bar{dl} = R \Phi$$

\Rightarrow
Stokes

If $N \rightarrow \infty$, $R \rightarrow 0 \Rightarrow R \Phi \approx 0$

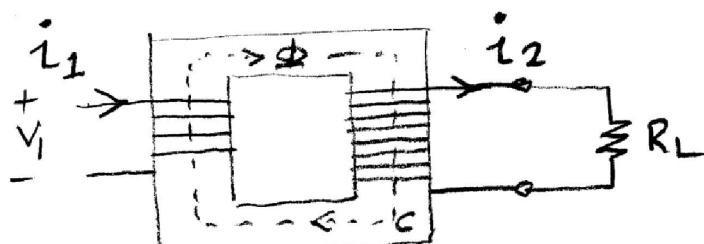
$\oint \bar{H} \cdot \bar{dl} = 0$ when coil is
wound around a
very high μ material

EXAMPLE: Transformers



We now have two coils wound on a transformer.

$$\mu \rightarrow \infty \Rightarrow \oint \bar{H} \cdot d\bar{l} = I_{enc} = 0$$

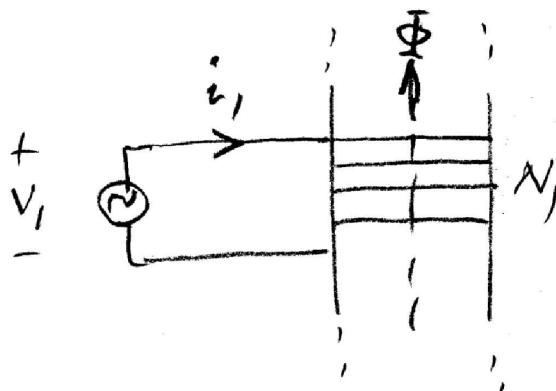


$$\oint_C \bar{H} \cdot d\bar{l} = I_{enc} = N_1 i_1 - N_2 i_2 = \cancel{\oint \Phi} = 0$$

$$\frac{i_1}{i_2} = \frac{N_2}{N_1}$$

GFF

What about the ratio between the voltages?



$$\text{Recall } \text{EMF} = \mathcal{E} = - \frac{d\Phi}{dt} \quad (\text{V})$$

We now have

$$V_1 = N_1 \frac{d\Phi_1}{dt} \quad (\text{V}) \quad (\text{right hand rule})$$

Similarly

$$V_2 = N_2 \frac{d\Phi_2}{dt} \quad (\text{V})$$

$$\Phi_1 = \Phi_2 \text{ Then,}$$

$$\frac{V_1}{V_2} = \frac{N_1}{N_2}$$

EXAMPLE

Input impedance of a transformer

Calculate $(R_1)_{\text{effective}} = \frac{V_1}{i_1}$

Recall

$$V_1 = \frac{N_1}{N_2} V_2$$

$$i_1 = \frac{N_2}{N_1} i_2$$

Then

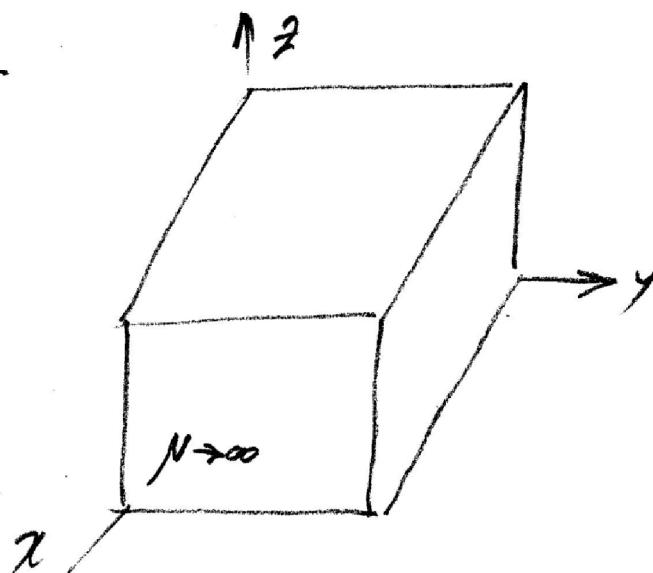
$$(R_1)_{\text{eff}} = \frac{(N_1/N_2) V_2}{(N_2/N_1) i_2} = \left(\frac{N_1}{N_2}\right)^2 \underbrace{\frac{V_2}{i_2}}_{R_L}$$

In general

$$(Z_1)_{\text{eff}} = \left(\frac{N_1}{N_2}\right)^2 Z_L$$

EXAMPLE

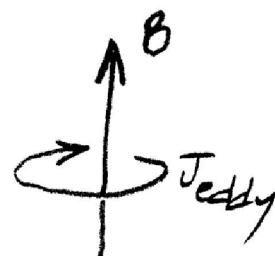
67 H



Given: $\bar{B} = \hat{x} B_0$

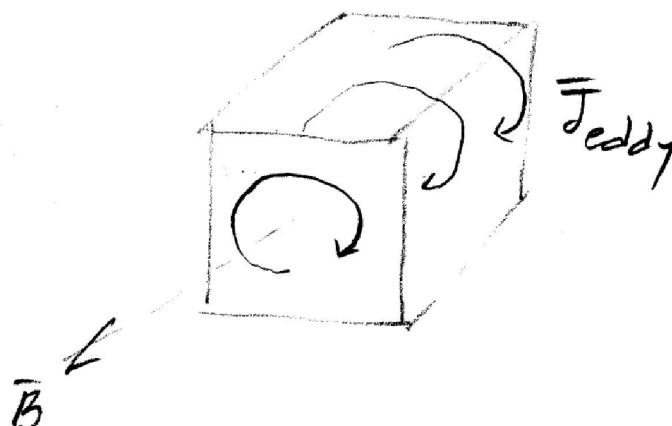
Eddy currents occur when $\sigma \neq 0$.

Also, given



opposite of
Right Hand Rule.

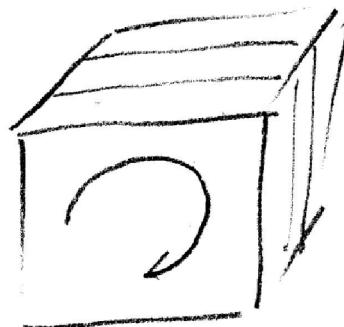
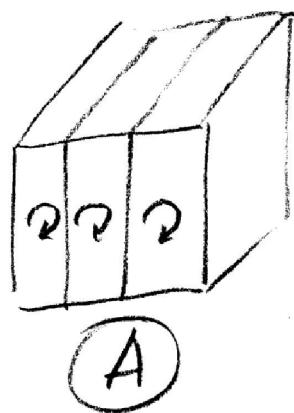
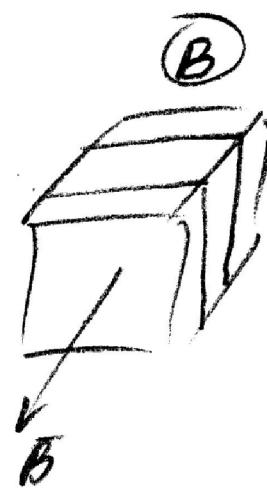
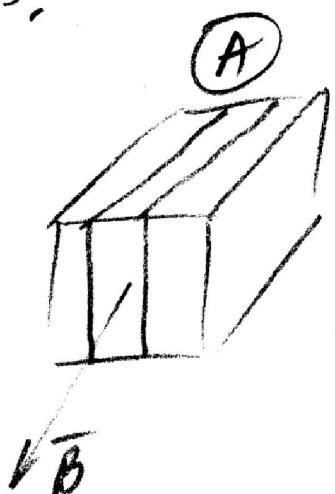
Draw Jeddy for the problem given figure.



67I

EXAMPLE Lamination of the magnetic core.

How should you slice the core with high μ and low σ to reduce the EDDY CURRENTS which cause OHMIC POWER LOSS and heat loss.



⇒ Total eddy-current power loss decreases as the number of laminations increases,