I authorize Prof. Pietrabissa to upload the test results on the mailing list of the Process Automation group.

FAMYLY NAME

GIVEN NAME

SIGNATURE

Process Automation (MCER), 2019-2020

TEST - A

19 Dec. 2019, 2h00

Exercise 1 (12 pts.)

Let the process be described by the transfer function: $P(s) = 0.1 \frac{1+\tau s}{(1+100s)^2} e^{-0.2s}$, with $\tau \in [1,11)$, and let the nominal value of the parameter be $\tilde{\tau} = 1$. Design a robust controller by following the IMC design to obtain a type 1 system under the ISE cost.

Exercise 2 (12 pts.)

Consider a process whose transfer function model is:

$$\begin{cases} x(t) = Mx(t-1) + Pu(t-1) \\ y(t) = Qx(t) \end{cases}, \text{ with } M = \begin{bmatrix} 0.5 & 0.5 \\ 0 & 0.5 \end{bmatrix}, P = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, Q = \begin{bmatrix} 1 & 0 \end{bmatrix}.$$

- Compute the control action at time t = 5 by implementing a Predictive Functional Control algorithm, considering:
 - prediction and control horizon p = m = 3;
 - number of base functions $n_B = 1$;
 - number of coincident points $n_H = 2$;
 - smooth response preferred;
 - cost function $J = e^T e$, where e is the vector of future errors between predicted output and reference trajectory;

known ramp reference signal
$$r(t) = t$$
 and reference trajectory $w(t+k|t) = \begin{cases} y_m(t) & \text{if } k = 0 \\ 0.5 & w(t+k-1|t) + 0.5 & r(t+k) \text{ otherwise} \end{cases}$

- predicted plant-model errors at time t equal to $\hat{n}(t + k|t) = k(y_m(t) y(t))$;
- available data at t = 5:
 - control action u(4) = 1.5;
 - state $x(4) = \begin{bmatrix} 6 \\ 4 \end{bmatrix}$;
 - measured output $y_m(5) = 7$.

Questions (6 pt.)

- Analyze the stability of the controlled system with open-loop transfer function $F(s) = 0.1 \frac{1-s}{s(1+s)} e^{-10s}$ (approximations are acceptable, 4pt)
- Explain why the stability of a system controlled by using a Smith's Predictor is not dependent on the value of the nominal delay. (1/2 pg. max, 2pt)

Solution of exercise 1

Since the time-delay of the process, 0.2s, is much smaller than the time constant – about 100s – of the process, we can use a

Padé approximation to write the delay term as a transfer function: $P^P(s) = 0.1 \frac{1+\tau s}{(1+100s)^2} \frac{1-0.1s}{1+0.1s}$. The process : P^P is stable, therefore it is possible to design a stable controller Q(s) for the closed-loop nominal system, with nominal process $\tilde{P}^P(s) = 0.1 \frac{1+s}{(1+100s)^2} \frac{1-0.1s}{1+0.1s}$.

The IMC design procedure to robustly stabilize the process $P^{P}(s)$ consists in the following 3 steps: Step 1)

a) Factorize the nominal process in a minimum-phase term and a non-minimum-phase term (ISE-optimal factorization):

$$P^{P}(s) = P_{+}(s)P_{-}(s)$$
, with $P_{+}(s) = \frac{1 - 0.1s}{1 + 0.1s}$ and $P_{-}(s) = 0.1 \frac{1 + s}{(1 + 100s)^{2}}$.

Define the preliminary controller as $\tilde{Q}(s) = P_{-}^{-1}(s) = 10 \frac{(1+100s)^2}{1+s}$.

Step 2)

Design the controller $Q(s) = \tilde{Q}(s)f(s)$, where the IMC filter f(s) must be such that a) the controller Q(s) is proper and b) the overall system is of type 1. Thus, we use the IMC filter $f(s) = \frac{1}{(1+\lambda s)^n}$ with n=1.

The IMC controller is then $Q(s) = 10 \frac{(1+100s)^2}{(1+s)(1+\lambda s)}$

Step 3)

Determine the value of λ such that the sufficient condition for robust stability holds:

$$|l_m(j\omega)\tilde{T}^P(j\omega)| < 1, \forall \omega$$

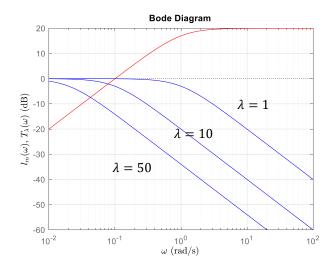
where

$$\left|\tilde{T}^P(j\omega)\right| = |Q(j\omega)P^P(j\omega)| = |P_-^{-1}(j\omega)f(j\omega)P_+(j\omega)P_-(j\omega)| = \left|\frac{1 - 0.1j\omega}{(1 + 0.1j\omega)(1 + j\omega\lambda)}\right| = \left|\frac{1}{1 + j\omega\lambda}\right|$$

and $l_m(j\omega)$ is an upper-bound of the multiplicative uncertainty $\Delta_m(j\omega)$, i.e., a function such that $|l_m(j\omega)| >$ $|\Delta_m(j\omega)|$, $\forall \omega$. By definition, the multiplicative uncertainty is defined as

$$\Delta_m(s) \coloneqq \frac{P^P(s) - \bar{P}^P(s)}{\bar{P}^P(s)} = \frac{0.1 \frac{1 + s\tau}{(1 + 100s)^2 1 + 0.1s} - 0.1 \frac{1 + s}{(1 + 100s)^2 1 + 0.1s}}{0.1 \frac{1 + s}{(1 + 100s)^2 1 + 0.1s}} = (\tau - 1) \frac{s}{1 + s}.$$

The upper-bound is clearly obtained when the gain magnitude is maximized, that is for $\tau = 11$: $l_m = 10 \frac{s}{1+s}$. The figures below show that, at least, for $\lambda \ge 10$ the condition is met. For instance, we choose $\lambda = 10$, obtaining $Q(s) = 10 \frac{(1+100s)^2}{(1+s)(1+10s)}$



Solution of exercise 2

To develop the PFC controller, we need to select the $n_B = 1$ base function: we select the polynomial (step) function $B_1(k) = 1$ $k^0 = 1, k = 0,1,2,...$

From the process model we note that there is an input-output delay d = 1, as, using the model equations,

$$y(t) = Qx(t) = x_1(t) = 0.5x_1(t-1) + 0.5x_2(t-1) = 0.5x_1(t-1) + 0.5(0.5x_2(t-2) + u(t-2)).$$

Since the control horizon is 3, the only coincident points that can be selected are then $h_1 = 2$ and $h_2 = 3$.

The model responses to the base functions in the coincidence points, considering null initial conditions $x(0) = \begin{bmatrix} 0 \\ n \end{bmatrix}$ are computed as follows:

$$\begin{split} B_1(k) &= 1 \\ \underline{t = h_1 = 2} & y_{B_1}(2) = QMPB_1(0) + QPB_1(1) = (QMP + QP)1 = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} 0.5 & 0.5 \\ 0 & 0.5 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix} 1 + \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix} 1 = 0.5; \\ \underline{t = h_2 = 3} & y_{B_1}(3) = QM^2PB_1(0) + QMPB_1(1) + QPB_1(2) = (QM^2P + QMP + QP)1 = 2. \end{split}$$

Thus,

$$y_B(h_1) = y_{B_1}(h_2) = [1.5];$$

 $y_B(h_2) = y_{B_1}(h_2) = [2].$

The matrix
$$Y_B \in \mathbb{R}^{n_H \times n_B}$$
 is then $Y_B = \begin{pmatrix} y_B(h_1) \\ y_B(h_2) \end{pmatrix} = \begin{bmatrix} 1.5 \\ 2 \end{bmatrix}$

The matrix $Y_B \in \mathbb{R}^{n_H \times n_B}$ is then $Y_B = \begin{pmatrix} y_B(h_1) \\ y_B(h_2) \end{pmatrix} = \begin{bmatrix} 1.5 \\ 2 \end{bmatrix}$. The matrix Y_B is used to compute the solution of the unconstrained optimization problem $\mu^* = (Y_B^T Y_B)^{-1} Y_B^T (w - f)$, with $(Y_B^T Y_B)^{-1} Y_B^T = (\begin{bmatrix} 0.5 & 2 \end{bmatrix} \begin{bmatrix} 0.5 \\ 2 \end{bmatrix})^{-1} \begin{bmatrix} 0.5 & 2 \end{bmatrix} = \begin{bmatrix} 0.12 & 0.47 \end{bmatrix},$

where μ^* is the vector of the optimal parameters at time t. The control action is the computed as $u(t) = \mu^{*T} B(0)$, where B(0)is the column vector of base functions $B_i(k)$, $i=1,2,...,n_B$, evaluated for k=0. In our problem, since $n_B=1$, we need to find one parameter $\mu^*(t)$.

At t = 5, considering that r(t) = t, the vector of the future reference values evaluated in the coincidence points $t + h_1 = 7$ and $t + h_2 = 8$ is computed as follows:

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w(5|5) = y_m(5) = 4;
w(6|5) = 0.5 \cdot w(5|5) + 0.5 \cdot r(6) = 5;
w(7|5) = 0.5 \cdot w(6|5) + 0.5 \cdot r(7) = 6.
w(8|5) = 0.5 \cdot w(7|5) + 0.5 \cdot r(8) = 7.
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By considering the given state and measured output values, we can compute the free response at time t = 5 over the coincidence points. With PFC, the output prediction is

$$\hat{y}(t+k|t) = y(t+k) + \hat{n}(t+k|t) = \sum_{i=1,\dots,n_B} y_{B_i}(k)\mu_i(t) + QM^k x(t) + \hat{n}(t+k|t),$$

where the last two terms constitute the free response:

$$f(t+k|t) = QM^{k}x(t) + \hat{n}(t+k|t) = QM^{k}x(t) + k(y_{m}(t) - y(t)).$$

$$\underline{t=5}$$
 $y_m(5) = 7; x(4) = \begin{bmatrix} 4 \\ 6 \end{bmatrix}; u(4) = 1.5;$

$$\begin{cases} x(5) = Mx(4) + Pu(4) = \begin{bmatrix} 0.5 & 0.5 \\ 0 & 0.5 \end{bmatrix} \begin{bmatrix} 4 \\ 6 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} 1 = \begin{bmatrix} 5 \\ 4 \end{bmatrix} \\ y(5) = Qx(5) = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} 4 \\ 6 \end{bmatrix} = 6 \end{cases};$$

$$\begin{split} \underline{h_1 = 2} & f(7) = QM^{h_1}x(5) + h_1\big(y_m(5) - y(5)\big) = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} 0.5 & 0 \\ 0 & 0.5 \end{bmatrix}^2 \begin{bmatrix} 6 \\ 4 \end{bmatrix} + 2(7-6) = 3; \\ \underline{h_2 = 3} & f(8) = QM^{h_2}x(5) + h_1\big(y_m(5) - y(5)\big) = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} 0.5 & 0 \\ 0 & 0.5 \end{bmatrix}^3 \begin{bmatrix} 6 \\ 4 \end{bmatrix} + 3(7-6) = 3.5. \end{split}$$

$$d(5) = \begin{bmatrix} \widehat{w}(7) \\ \widehat{w}(8) \end{bmatrix} - \begin{bmatrix} f(7) \\ f(8) \end{bmatrix} = \begin{bmatrix} 7 \\ 8 \end{bmatrix} - \begin{bmatrix} 3 \\ 3.5 \end{bmatrix} = \begin{bmatrix} 4 \\ 4.5 \end{bmatrix}.$$

$$\mu(5) = \begin{bmatrix} 0.12 & 0.47 \end{bmatrix} \begin{bmatrix} 4 \\ 4.5 \end{bmatrix} = 2.6.$$

$$u(5) = \mu(5)B_1(0) = 2.6.$$