

Ex 1 .  $\tilde{P}(s)$  stable

Step 1 FACTORIZATION OF  $\tilde{P}(s) = 100 \frac{s+1}{(s+10)^2}$

IMC

EXERCISES

$$\tilde{P}_+(s) = 1$$

(no non-minimum phase terms)

$$\tilde{P}_-(s) = \frac{1+s}{(1+s/10)^2}$$

$$\Rightarrow \tilde{Q}(s) = \tilde{P}_-(s) = \frac{(1+s/10)^2}{1+s} \quad \text{← NOT PROPER}$$

Step 2 FILTER  $f(s) = \frac{1}{(1+\lambda s)^m}, m=1$

$$Q(s) = \tilde{Q}(s) f(s) = \frac{(1+s/10)^2}{(1+s)(1+\lambda s)}$$

$$* \quad \tilde{T}(s) = \tilde{P}(s) Q(s) = \tilde{P}_+(s) \tilde{P}_-(s) \tilde{Q}(s) f(s) = \frac{\tilde{P}_+(s)}{1+\lambda s} \Big|_{s=0} = 1 \quad \checkmark$$

$\Rightarrow$  Type 1 system

### STEP 3 ROBUSTNESS

- MULTIPLICATIVE UNCERTAINTY

$$P(s) = \tilde{P}(s) \left(1 + \Delta_m(s)\right)^+; \quad \text{ROBUST STABILITY CONDITION}$$

$$|\tilde{\ell}_m'(j\omega) \tilde{F}(j\omega)| < 1, \forall \omega$$

upper bound of  $\Delta_m$

$$\boxed{\Delta_m(j\omega)} = \frac{P(j\omega) - \tilde{P}(j\omega)}{\tilde{P}(j\omega)} = \frac{K \frac{j\omega - 2}{(j\omega + 10)^2} - 100 \frac{j\omega + 1}{(j\omega + 10)^2}}{100 \frac{j\omega + 1}{(j\omega + 10)^2}} =$$

$$= \frac{K(j\omega - 2) - 100(j\omega + 1)}{100(j\omega + 1)} = \xrightarrow{\quad} \left\{ \begin{array}{l} K = 100(1 + \delta) \\ 2 = -\frac{1}{1 + \delta} \end{array} \right.$$

$$\therefore \frac{(1 + \delta)j\omega + 1 - j\omega - 1}{j\omega + 1} = \delta \frac{j\omega}{j\omega + 1}, \quad \delta \in [0, 2]$$

$$h_m(j\omega) = 2 \frac{j\omega}{1+j\omega}$$

Robust stability

$$|h_m(j\omega) \tilde{H}(j\omega)| < 1, \forall \omega$$

$$\left| 2 \frac{j\omega}{1+j\omega} \frac{1}{1+\lambda j\omega} \right| < 1, \forall \omega$$

HP filter       $\tilde{H}$  LP filter

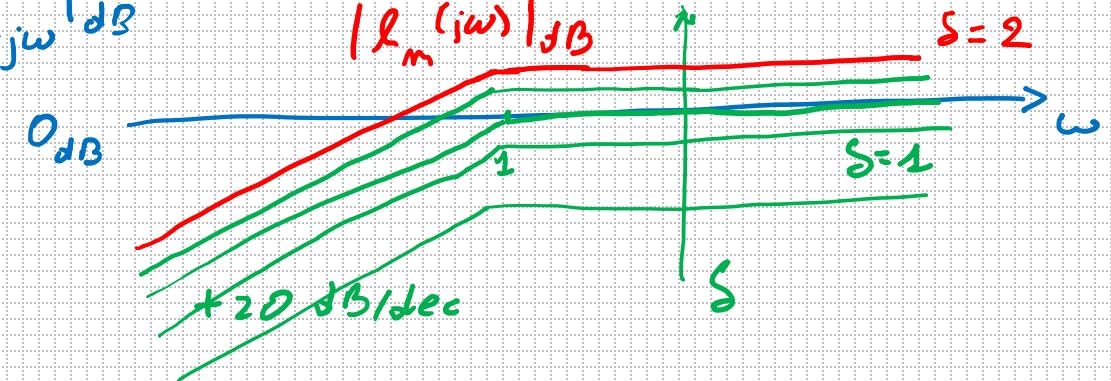
For  $\lambda > 1$  the condition is verified

$$\boxed{\lambda = 2}$$

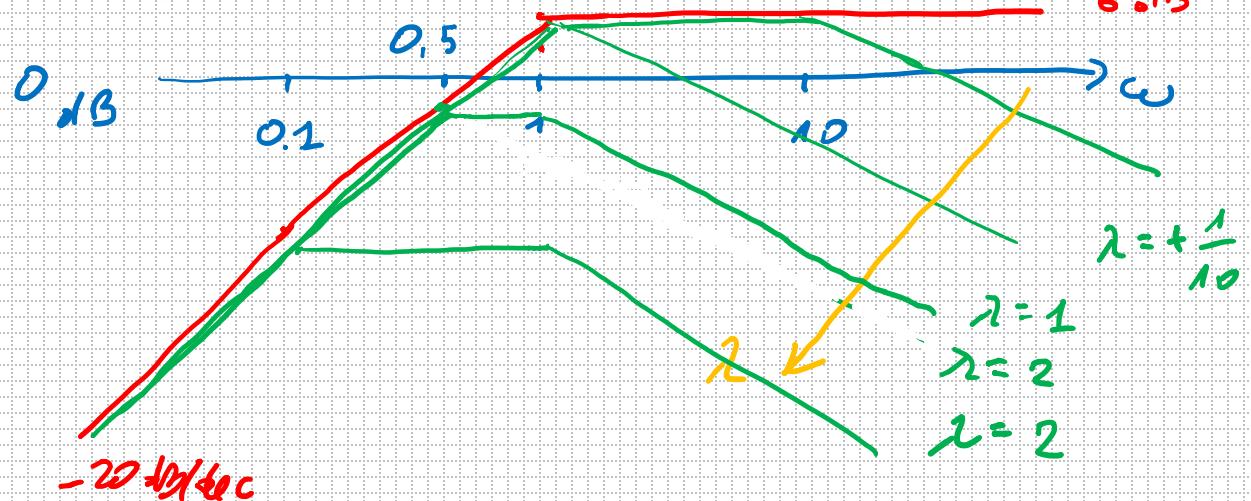
$$\boxed{Q(s) = \frac{(1+s/10)^2}{(1+s)(1+2s)}}$$

(A)

$$|s \frac{j\omega}{1+j\omega}| dB$$



$$|h_m \tilde{H}| dB$$



$$s \in [0, \infty)$$

$$\underline{B.} \quad G(s) = \frac{Q(s)}{1 - \tilde{P}(s)Q(s)} = \frac{\frac{(1 + \frac{s}{T_0})^2}{(1+s)(1+2s)}}{1 - \frac{1+s}{\frac{(1+\frac{s}{T_0})^2}{(1+s)(1+2s)}}} = \frac{(1 + \frac{s}{T_0})^2}{(1+s)(1+2s) - (1+s)} =$$

$$= \frac{\frac{1}{100}s^2 + \frac{1}{5}s + 1}{2s(1+s)} = \frac{1}{10} \left( 1 + \frac{1}{0.5s} + \frac{1}{20}s \right) \frac{1}{1+s} = K_p \left( 1 + \frac{1}{T_i s} + T_0 s \right) \frac{1}{1 + \beta_s s}$$

$\text{PID} + f$  (2 zeros,  
1 pole in  $s=0$ ,  
1 pole )

$$\begin{cases} K_p = 0.1 \\ T_i = 0.5 \\ T_0 = 0.05 \\ \beta_s < 1 \end{cases}$$

ROBUST PID + f CONTROLLER

[2]

S RHP zeros

$$P(s) = K \frac{s - z}{(s + 10)^2}, \quad z = \frac{1}{1+s}, \quad K = 100(1+z), \quad \delta \in [0, 2)$$

$$\tilde{P}(s) = 100 \frac{s-1}{(s+10)^2}$$

A.  $Q(s)$  D.F.

ROBUST STABILITY  
 TYPES  
 IAE



,  $\tilde{P}(s)$  is STABLE  $\Rightarrow$  IMC design

- STEP 1.

$$\tilde{P}_+(s) = s - 1$$

~~W/ wrong V~~

$$\tilde{P}_+(0) = -1$$

$$\tilde{P}(s) = -\frac{1-s}{(1+s/10)^2};$$

$$\tilde{P}_+(s) = 1-s,$$

$$\tilde{P}_- = -\frac{1}{(1+\frac{s}{10})^2}$$

IAE

$$\Rightarrow \boxed{\tilde{Q}(s) = -\left(1 + \frac{s}{10}\right)^2}$$

$$\left( \tilde{P}_+(s) = \frac{1-s}{1+s}, \quad \tilde{P}_- = -\frac{1+s}{(1+\frac{s}{10})^2} \right) \text{ISE}$$

STEP 2. FILTER  $\Rightarrow$  Type 1  $\Rightarrow f(s) = \frac{1}{(1+2s)^m}$ ,  $m = 2$

$$Q(s) = \hat{Q}(s) f(s) = -\frac{(1+s/10)^2}{(1+2s)^2}$$

$$\times \quad \tilde{P}(s) = \tilde{P}_+(s) | f(s) = (1-s) \frac{1}{(1+2s)^2} \Big|_{s=0} = 1 \quad \checkmark$$

STEP 3 ROBUSTNESS

ADDITIVE

UNCERTAINTY

$$\Delta(j\omega) = P(j\omega) - \tilde{P}(j\omega) \stackrel{*}{=}$$

$$(K=100(1+\zeta), 2 \cdot \frac{1}{1+\zeta})$$

ROBUST STAB. CONDITION

$$\underbrace{|h_a(j\omega) Q(j\omega)|}_{\text{UPPER-BOUND}} < 1, \forall \omega$$

of  $\Delta(j\omega)$

$$\begin{aligned} * &= K \frac{j\omega - 2}{(j\omega + 10)^2} - 100 \frac{j\omega - 1}{(j\omega + 10)^2} = 100 \frac{(1+\zeta)(j\omega - \frac{1}{1+\zeta}) - j\omega + 1}{(j\omega + 10)^2} = \zeta \frac{j\omega}{(1 + \frac{j\omega}{1+\zeta})^2} \end{aligned}$$

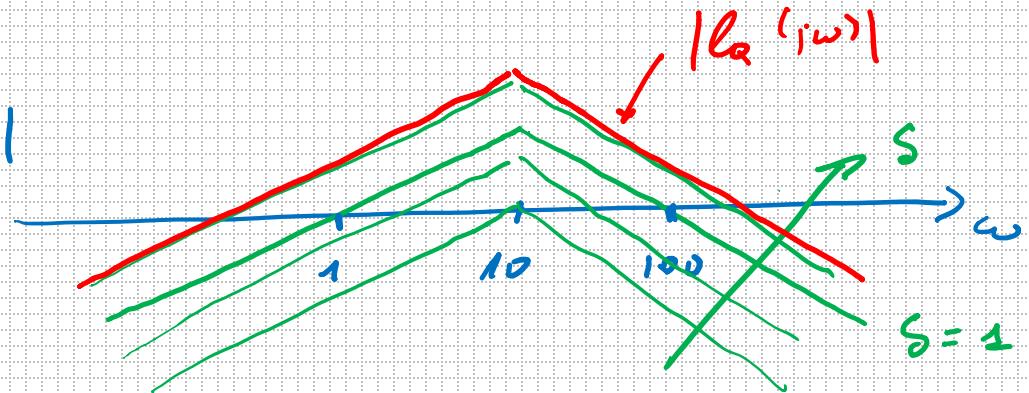
$$\Delta(j\omega) = \frac{j\omega}{(1 + j\omega/\tau_0)^2}$$

$$\zeta \in [0, 2)$$

$$l_a(j\omega) = 2 \frac{j\omega}{(1 + \frac{j\omega}{\tau_0})^2}$$

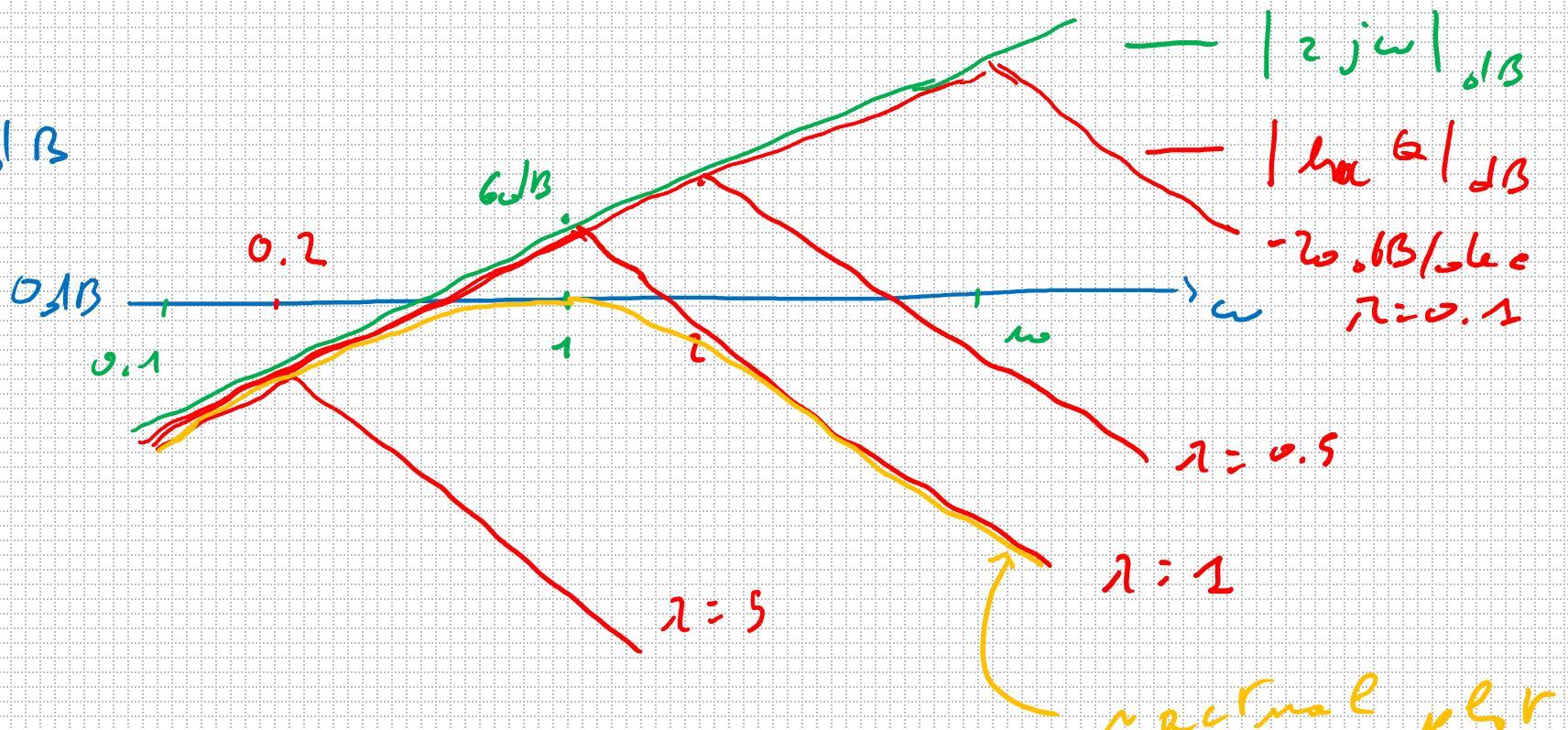
$$|l_a(j\omega) Q(j\omega)| = \left| 2 \frac{j\omega}{(1 + \frac{j\omega}{\tau_0})^2} \frac{(1 + \frac{j\omega}{\tau_0})^2}{(1 + 2j\omega)^2} \right| = \left| 2j\omega \frac{1}{(1 + 2j\omega)^2} \right| < 1 \text{ Bw}$$

$$|\Delta(j\omega)|$$



$$\left| \frac{2j\omega}{(1+\lambda j\omega)^2} \right| < 1 \quad \forall \omega$$

$$\left| \frac{2j\omega}{(1+\lambda j\omega)^2} \right|_0 \text{dB}$$



Condition is met for  $\lambda > 1$

conservative value

$\lambda = 5$

for  $\lambda = 1$

$$Q(s) = \frac{(1+j\omega/10)^2}{(1+j\omega 5)^2}$$

$$(\text{Ex. 2b}) \quad P(s) = K \frac{s - \tau}{(s + \tau_0)^2}, \quad \tau = \frac{1}{1+s}, \quad K = \log(1+s), \quad s \in [0, 2)$$

$$\tilde{P}(s) = \log \frac{s - \tau}{(s + \tau_0)^2}$$

• Design INC controller  $Q$ :

robust stability  
 type 1  
 TSE

Step 1. Factorization

$$\tilde{P}(s) = -\frac{1 - s}{(1 + s/\tau_0)^2}$$

$$P_+ = \prod_i \frac{1 - \beta_i s}{1 + \beta_i s} e^{-\theta_i}$$

$$\beta_i > 0$$

$$\tilde{P}_+(s) = \frac{1 - s}{1 + s},$$

$$(\tilde{P}_+(0) = 1)$$

$$\tilde{P}_-(s) = -\frac{1 + s}{(1 + s/\tau_0)^2}$$

$$\Rightarrow \tilde{Q}(s) = -\frac{(1 + s/\tau_0)^2}{1 + s}$$

Step 2. Filter

$$\text{Type 2} \rightarrow F(s) = \frac{1}{(1 + 2s)^m}, \quad m = 1$$

$$Q(s) = -\frac{(1 + s/\tau_0)^2}{(1 + s)(1 + 2s)}$$

(---)

Ex. 3 (IMC robust design, Type 2 system)

$$P(s) = \frac{(1+s/10)(1+s\tau+0.3)}{(1+s)^2(1+s\tau)}, \quad \tau \in [1, 10]$$

$$\tilde{P}(s) = \frac{1+s/10}{(1+s)^2}$$

• IMC controller Q : { robust stability  
Type 2 system  
IAE

check:  $\tilde{P}$  stable since it has 2 poles in -1

step 1.  $\tilde{P}_+(s) = 1, \quad \tilde{P}_-(s) = \frac{1+s/10}{(1+s)^2} \Rightarrow \tilde{Q}(s) = \frac{(1+s)^2}{1+s/10}$

step 2. type 2:  $\tilde{T}(0) = 1, \quad \left. \frac{d\tilde{T}(s)}{ds} \right|_{s=0} = 0$

$$f(s) = \frac{1 + (2\lambda - P'_+(0))s}{(1+\lambda s)^2} = \frac{1 + 2\lambda s}{(1+\lambda s)^2} \quad \leftarrow \frac{d\tilde{P}_+(s)}{ds} = 0$$

$$1) \quad Q(s) = \tilde{Q}(s) f(s) = \frac{(1+s)^2}{1+s/10} \frac{1+2s}{(1+2s)^2} \quad \text{PROPER V}$$

$$2) \quad \tilde{T}(s) = \tilde{P}(s) Q(s) = \underbrace{\tilde{P}_+(s)}_{\text{PROPER}} \cancel{\tilde{Q}(s)} f(s) = f(s)$$

$$\left\{ \begin{array}{l} \tilde{T}(0) = f(0) = 1 \quad \checkmark \\ \end{array} \right.$$

$$\left. \frac{d\tilde{T}(s)}{ds} \right|_{s=0} = \left. \frac{df(s)}{ds} \right|_{s=0} = \left[ -2 \cancel{s} \underbrace{\frac{1+2s^2}{(1+2s)^3}}_1 + 2 \cancel{2} \underbrace{\frac{1}{(1+2s)^2}}_1 \right]_{s=0} = 0 \quad \checkmark$$

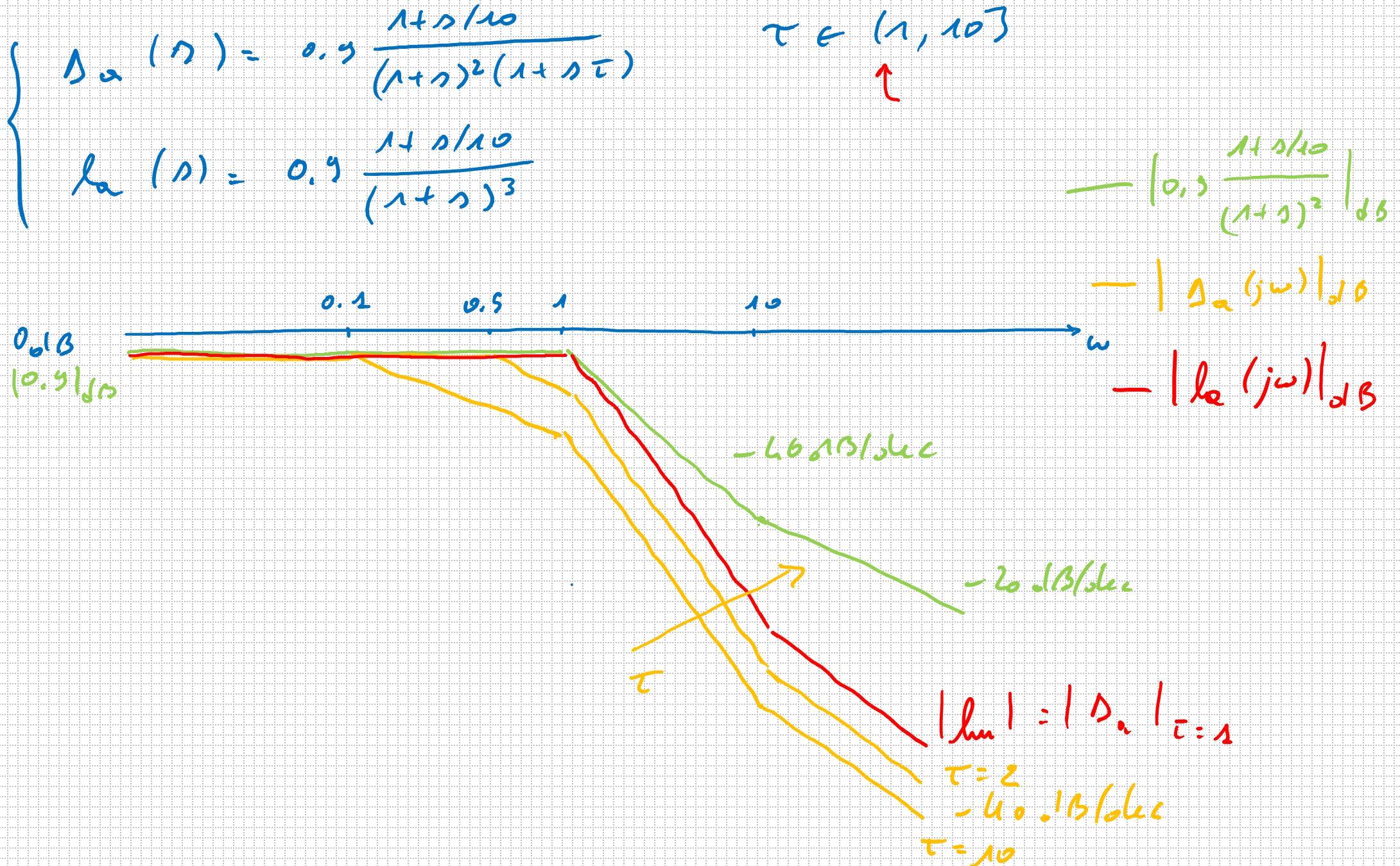
$$Q(s) = \frac{(1+s)^2 (1+2s)}{(1+s/10)(1+2s)^2}$$

STEP 3. ROBUST STABILITY

$$|\hat{h}_a(j\omega) Q(j\omega)| < 1, \forall \omega$$

$$\begin{aligned}
 \Delta_a(s) &= p(s) - \tilde{p}(s) = \frac{(1+s/10)((1+s\tau)+0.9)}{(1+s)^2(1+s\tau)} - \frac{1+s/10}{(1+s)^2} = \\
 &= \frac{(1+s/10)(1+s\tau) + 0.9(1+s/10) - (1+s/10)(1+s\tau)}{(1+s)^2(1+s\tau)} = \\
 &= 0.9 \frac{1+s/10}{(1+s)^2(1+s\tau)}, \quad \tau \in (2, 10]
 \end{aligned}$$

$\lambda_a(s) : |\lambda_a(j\omega)| > |\Delta_a(j\omega)|, \forall \omega, \forall \tau \in (2, 10]$



Robust stability condition  $|L_\omega(j\omega) Q(j\omega)| < 1, \forall \omega$

$$\Rightarrow \left| 0.9 \frac{1 + j\omega/\lambda_1}{(1 + j\omega)^2} \frac{(1 + j\omega)^2 (1 + j\omega/2\lambda_2)}{(1 + j\omega/\lambda_0) (1 + j\omega/2\lambda_2)^2} \right| = \left| 0.9 \frac{1 + j\omega/2\lambda_2}{(1 + j\omega)(1 + j\omega/2)^2} \right| < 1, \forall \omega$$

$$\left| 0.9 \frac{1}{1 + j\omega} \right|_{dB}$$



$$Q(j) = \frac{(1+j)^3}{(1+\lambda/\lambda_0)(1+\lambda/2)^2}$$

$\lambda = 0.5$

CONSERVATIVE  
CONDITION

$$\lambda \leq 0.5$$

-20 dB/decade

$$\lambda = 0.5$$

-40 dB/decade

$$\lambda = 5$$

\*  $\lambda_2 = -\frac{1}{2\lambda}, \lambda_1 = -\frac{1}{\lambda}, \lambda_2 = -\frac{1}{\lambda}$

Ex. 4

$$P(s) = K \frac{1+s/10}{(1+2s)^2} e^{-\theta s}, \quad \theta = 0.2, \quad K \in [10, 20]$$

$$\tilde{K} = 10$$

A) IMC controller  $Q(s)$ : { robust stability  
Type 1  
IAE}

B) Verify that  $Q(s)$  stabilizes  $P(s) = 10 \frac{1+s/10}{(1+2s)^2} e^{-0.2s}$

-  $\tilde{P}(s)$  is stable (it has two poles in  $-0.5$ )

-  $\theta = 0.2, \tau = 2 \Rightarrow$  Peake-approximation

1/1 Peake approx.

$$e^{-\theta s} \approx \frac{1 - s \frac{\theta}{2}}{1 + s \frac{\theta}{2}} = \frac{1 - s/10}{1 + s/10}$$

$$P^r(n) = K \frac{1 + \frac{2}{1-n}}{(1+2n)^2} \frac{1 - \frac{2}{1-n}}{1 + \frac{2}{1-n}} = K \frac{1 - \frac{2}{1-n}}{(1+2n)^2}$$

$\underbrace{1 + \frac{2}{1-n}}_{P(n)}$        $\underbrace{1 - \frac{2}{1-n}}_{\sim e^{-0.7}}$

$$\tilde{P}^r(n) = 10 \frac{1 - n/10}{(1+2n)^2}$$

Step 1.

$$\tilde{P}_+^r(n) = 1 - n/10 , \quad \tilde{P}_-^r(n) = 10 \frac{1}{(1+2n)^2}$$

$$\hat{Q}(n) = (\tilde{P}_-^r(n))^{-1} = \frac{1}{10} (1+2n)^2$$

Step 2

$$\text{Type 1} \quad f(n) = \frac{1}{(1+2n)^m}, m=2$$

$$Q(n) = \frac{1}{10} \frac{(1+2n)^2}{(1+2n)^2} ; \quad \tilde{T}(0) = \tilde{P}_+^r(0) f(0) = 1 \quad \checkmark$$

Step 3.

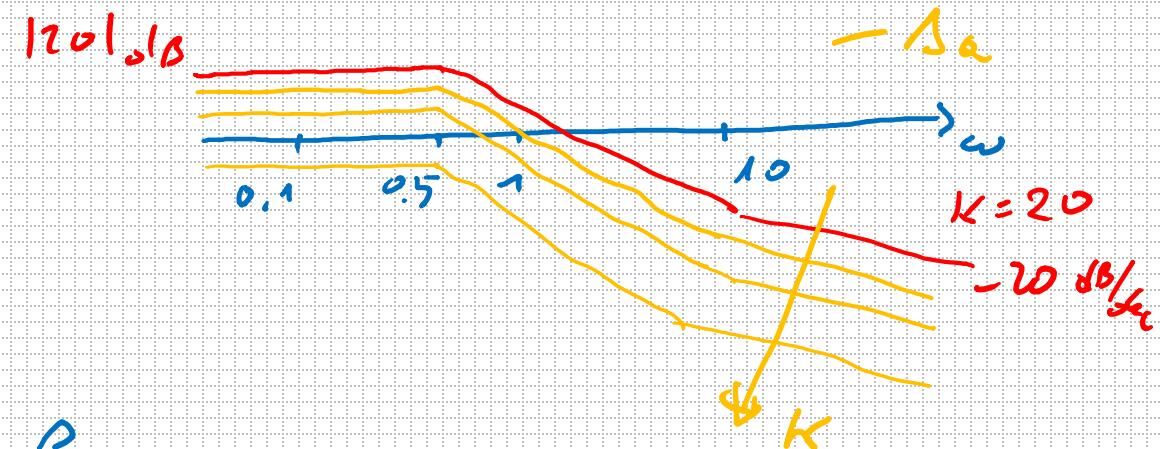
$$|\ln(j\omega) Q(j\omega)| < 1, \forall \omega$$

$$\Delta_a(s) = P^P(s) - \tilde{P}^P(s) = K \frac{1 - s/10}{(1+2s)^2} - 10 \frac{1 - s/10}{(1+2s)^2} = (K-10) \frac{1 - \frac{s}{10}}{(1+2s)^2}$$

$$\omega \in [10, \infty)$$

$$\mu_e(s) = 10 \frac{1 - s/10}{(1+2s)^2}$$

$$\ln Q = \cancel{10} \frac{1 - \cancel{s/10}}{(1+2s)^2} \cancel{\frac{1}{10}} \frac{(1+2s)^2}{(1+2s)^2} = \frac{1 - \frac{s}{10}}{(1+2s)^2}$$



$$\left| \frac{1 - j\omega/10}{(1 + j\omega\lambda)^2} \right|_{dB}$$

$$\lambda > 0.1$$

conservative  
condition

$\lambda_B$

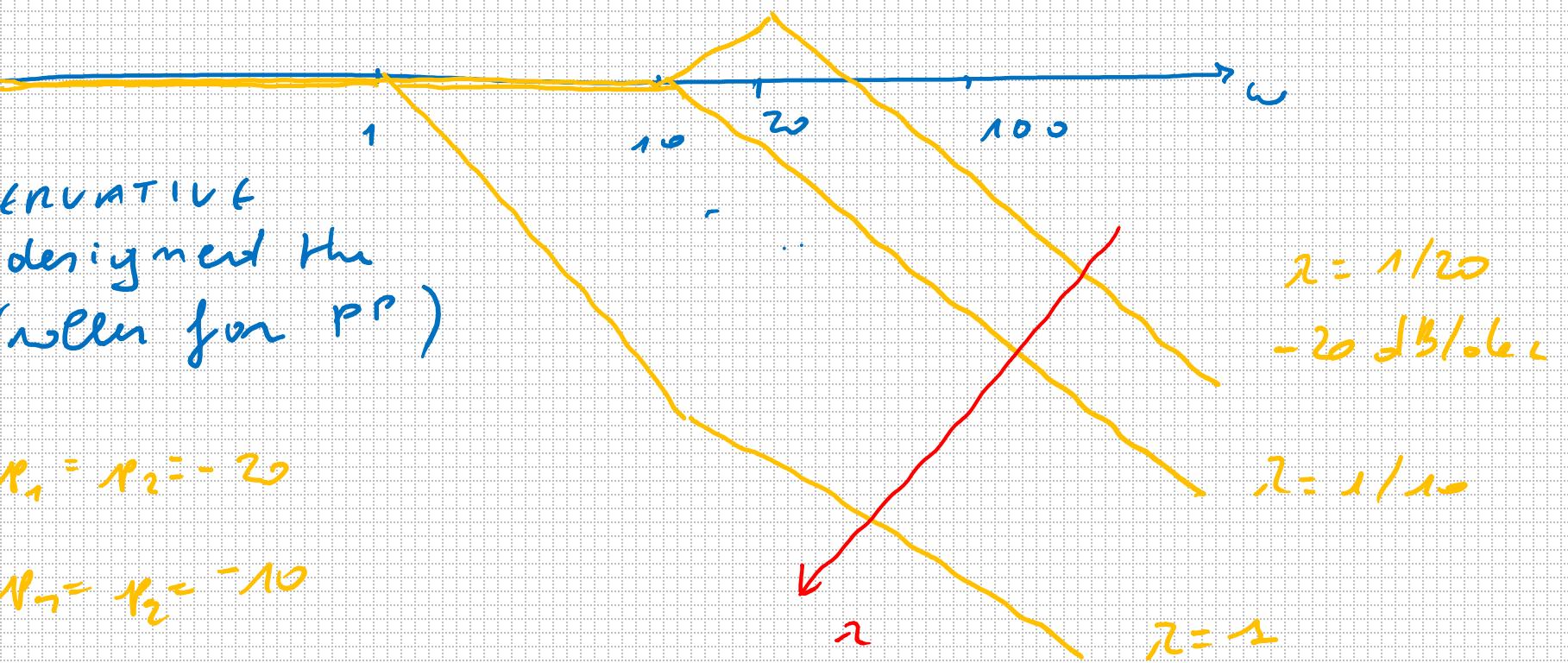
$$\boxed{\lambda = 0.2}$$

CONSERVATIVE  
(we designed the  
controller for P)

$$\lambda = 1/20 \Rightarrow p_1 = p_2 = -20$$

$$\lambda = 1/10 \Rightarrow p_1 = p_2 = -10$$

$$\lambda = 1 \Rightarrow p_1 = p_2 = -1$$

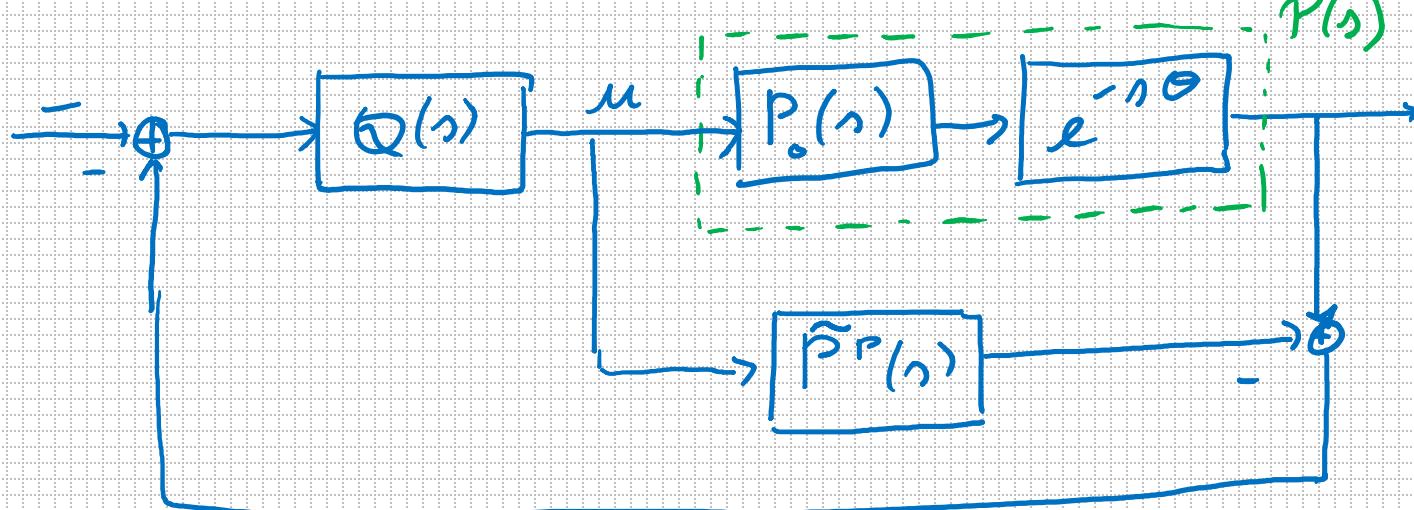


$$Q(s) = \frac{1}{10} \frac{(1+2s)^2}{(1+0.2s)^2}$$

CONTROLLER WHICH  
ROBUSTLY STABILIZES  
 $P^r(s)$

b) classic controller

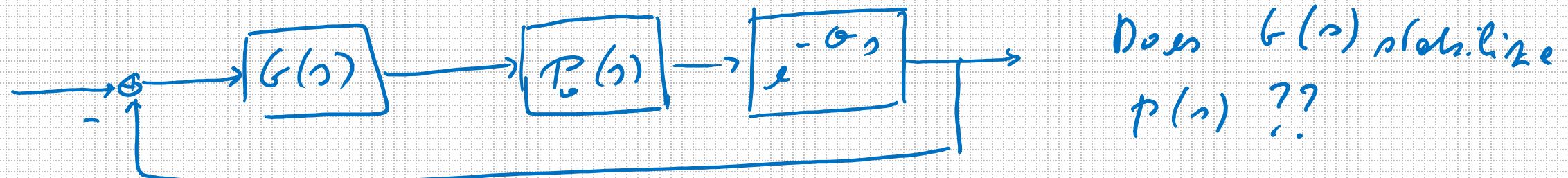
$$G(s) = \frac{Q(s)}{1 - \tilde{P}^r(s) Q(s)} = \frac{\frac{1}{10} \frac{(1+2s)^2}{(1+0.2s)^2}}{1 - \cancel{10} \frac{1-0.1s}{(1+2s)^2} \cancel{\frac{1}{10} \frac{(1+2s)^2}{(1+0.2s)^2}}} = *$$



We computed  $Q$   
considering  $\tilde{P}^r$ , i.e.,  
according this  
scheme

$$G(s) = \frac{1}{5} \frac{(1+2s)^2}{s(1+\frac{2}{50})} = K_C \left(1 + \frac{1}{T_i s} + T_D s\right) \frac{1}{1+\beta_f s} (PID + f)$$

$$\begin{cases} K_C = 0.2 \\ T_i = 0.06 \\ T_D = 1 \\ \beta_f = 0.02 \end{cases}$$

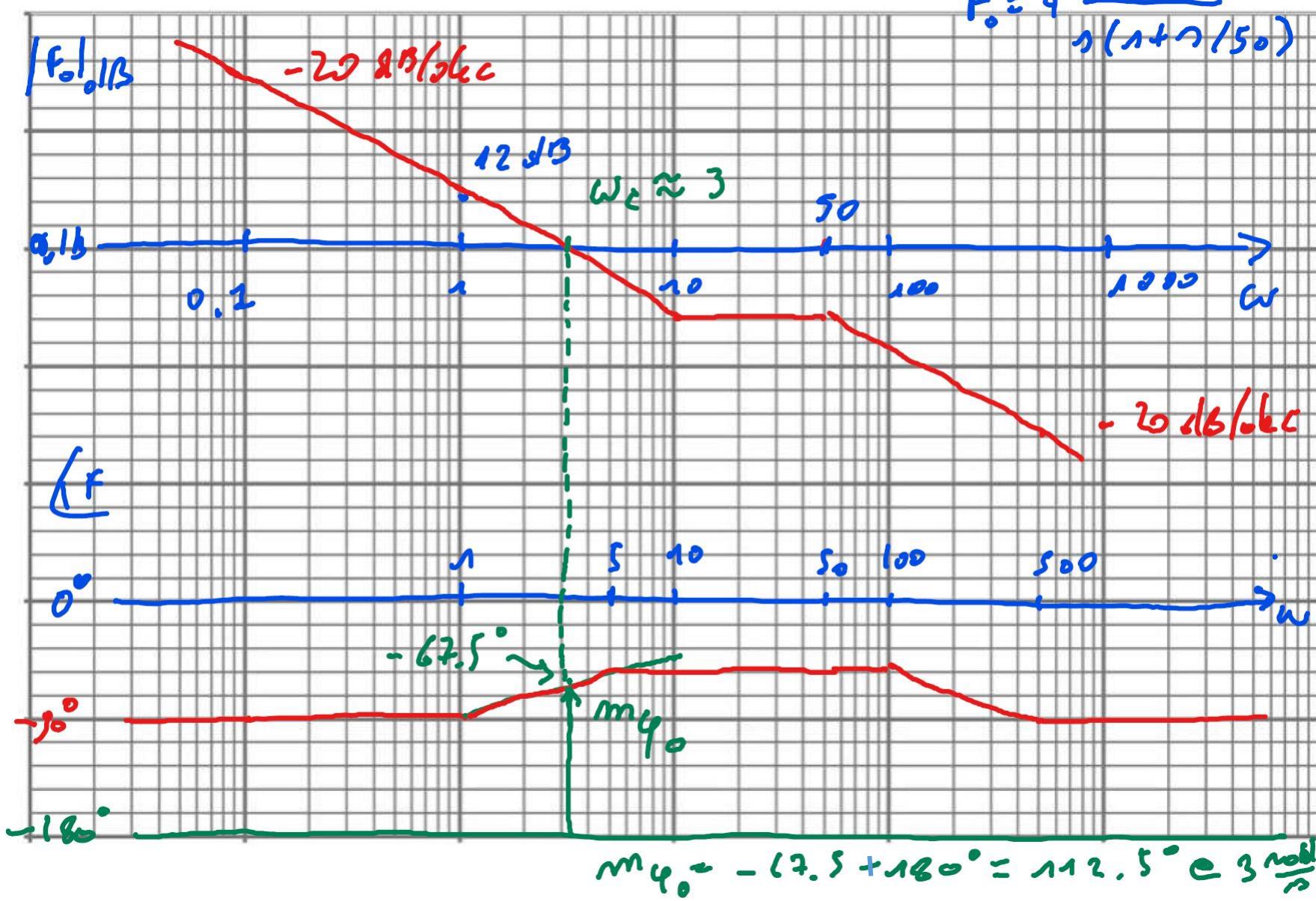


Step 1. Check if  $G$  stabilizes  $P_0$  and compute its  $m_\infty$ .

Step 2. Check if  $m_{\infty} > 0$

(delay margin for  $P$  is  $m_t = m_\infty - \sigma$ )

$$\bar{F}_0 = 4 \frac{\lambda + \gamma / \omega}{\gamma (\lambda + \gamma / 50)}$$



$$K = 20, \quad \bar{P}_o(s) = 20 \frac{1 - 0.1s}{(1 + 2s)^2}$$

$$\bar{F}_o(s) = G(s) \bar{P}_o(s) = 4 \frac{1 + 0.1s}{s(1 + 0.1s)}$$

$$m_{\tau_0} = \frac{m_{\varphi_0}}{\omega_c} \approx \frac{112.5^\circ}{3 \text{ rad/s}} \frac{\frac{\pi}{180^\circ}}{} = 0.65 \text{ s}$$

$$[m_x = m_{\tau_0} - 0 = 0.65 - 0.2 = 0.45]$$

maximum steering

$\bar{P}(s)$  is  
stabilized  
by  $G(s)$