Task 1:

damp(denreg);

```
clear all
clc
```

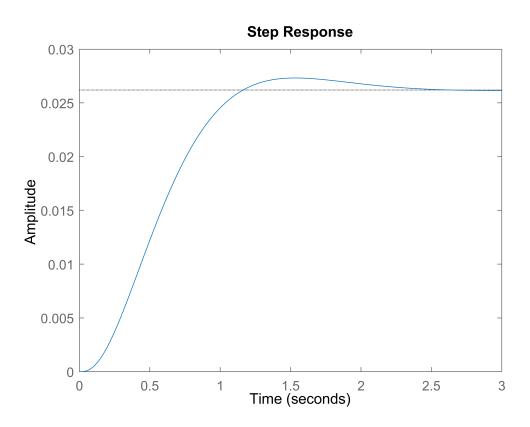
```
a)
 % First enter the transfer function G(s)
 numG = 1;
 denG = conv ( conv ( [1 0], [1 1] ), [0.2 1] );
 % Convert to state-space model
 [ Ag, Bg, Cg, Dg ] = tf2ss ( numG, denG );
 % Check the controllability and observability of the system
     system order = length(Ag) % equals "3"
 system order = 3
 M = ctrb(Ag, Bg);
 rank of M = rank(M) % equals "3"
 rank_of_M = 3
 N = obsv(Ag, Cg);
 rank_of_N = rank(N) % equals "3"
 rank_of_N = 3
 % Compute the poles of a second-order system
 damping = 0.707;
 wn = 3;
 [ num2, den2 ] = ord2 (wn, damping);
 % Select desired poles to include poles of the second-order system
 dominant = roots(den2);
 desiredpoles = [dominant' 10 * real( dominant(1) ) ];
 % Compute the controller gain K
 K = acker (Ag, Bg, desiredpoles);
 % Compute the closed-loop state variable feedback system
 Asf = Ag - Bg * K; Bsf = Bg; Csf = Cg; Dsf = 0;
 [numsf, densf] = ss2tf (Asf, Bsf, Csf, Dsf);
 % Select observer poles to be 10 times faster than controller poles
 observerpoles = 10 * desiredpoles;
 % Compute the observer gain L
 L = acker (Ag', Cg', observerpoles);
 % Compute the closed-loop system with both controller and observer
 Areg = [ (Ag - Bg * K) Bg * K; zeros( size(Ag) ) (Ag - L' * Cg) ];
 Breg = [ Bg; zeros( size(Bg) ) ];
 Creg = [ Cg zeros( size(Cg) ) ];
 Dreg = 0;
 [numreg, denreg] = ss2tf ( Areg, Breg, Creg, Dreg );
```

Pole Damping Frequency Time Constant (rad/TimeUnit) (TimeUnit)

```
-2.12e+01 + 2.12e+01i
                          7.07e-01
                                         3.00e+01
                                                           4.71e-02
-2.12e+01 - 2.12e+01i
                          7.07e-01
                                         3.00e+01
                                                           4.71e-02
-2.12e+01
                          1.00e+00
                                         2.12e+01
                                                           4.71e-02
-2.12e+00 + 2.12e+00i
                          7.07e-01
                                         3.00e+00
                                                           4.71e-01
-2.12e+00 - 2.12e+00i
                          7.07e-01
                                         3.00e+00
                                                           4.71e-01
G=tf(numsf,densf);
step(numsf,densf);
```

4.71e-03

2.12e+02



1.00e+00

```
stepinfo(G, 'RiseTimeLimits',[0.1 0.9])
```

```
ans = struct with fields:
    RiseTime: 0.7242
TransientTime: 2.0357
SettlingTime: 2.0357
SettlingMax: 0.0239
SettlingMax: 0.0273
Overshoot: 4.2749
Undershoot: 0
Peak: 0.0273
PeakTime: 1.5416

K
```

```
K = 1×3
19.4520 93.9728 190.8900
```

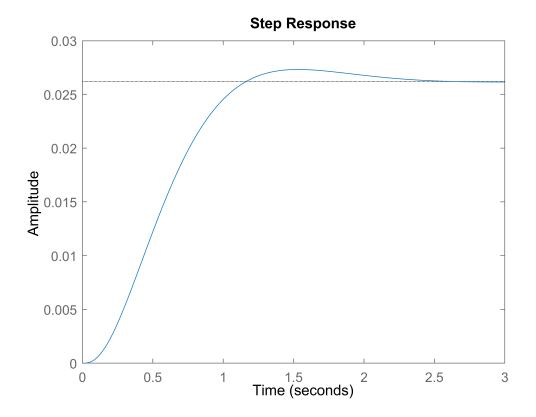
-2.12e+02

```
desiredpoles2 = [dominant' 20 * real( dominant(1) ) ];
% Compute the controller gain K
K2 = acker (Ag, Bg, desiredpoles2);
```

```
% Compute the closed-loop state variable feedback system
Asf = Ag - Bg * K; Bsf = Bg; Csf = Cg; Dsf = 0;
[numsf2, densf2] = ss2tf (Asf, Bsf, Csf, Dsf);
% Select observer poles to be 10 times faster than controller poles
observerpoles2 = 10 * desiredpoles2;
% Compute the observer gain L
L2 = acker (Ag', Cg', observerpoles2);
% Compute the closed-loop system with both controller and observer
Areg2 = [ (Ag - Bg * K) Bg * K; zeros( size(Ag) ) (Ag - L2' * Cg) ];
Breg2 = [ Bg; zeros( size(Bg) ) ];
Creg2 = [ Cg zeros( size(Cg) ) ];
Dreg2 = 0;
[numreg2, denreg2] = ss2tf ( Areg2, Breg2, Creg2, Dreg2 );
damp(denreg2);
```

Pole	Damping	Frequency (rad/TimeUnit)	Time Constant (TimeUnit)
-4.24e+02 -2.12e+01 + 2.12e+01i -2.12e+01 - 2.12e+01i -2.12e+01 -2.12e+00 + 2.12e+00i -2.12e+00 - 2.12e+00i	1.00e+00 7.07e-01 7.07e-01 1.00e+00 7.07e-01 7.07e-01	4.24e+02 3.00e+01 3.00e+01 2.12e+01 3.00e+00 3.00e+00	2.36e-03 4.71e-02 4.71e-02 4.71e-02 4.71e-01

```
G2=tf(numsf2,densf2);
step(numsf2,densf2);
```




```
K2 = 1 \times 3
40.6620 183.9456 381.7800
```

In this example we can observe that the gain calculated with the regulator.m file is approximately 2 times smaller, than the gain of the initial system. Correspondingly the numbers obtained were: 40.66, 183,94 & 381.78 without regulator.m and 19.45, 93.97 & 190.89 with it. The change in gain has an influence on the position of the poles of the system. We know that as the gain continues to increase the location of closed-loop poles moves from the location of open-loop poles to the location of open-loop zeros. The overall dynamics of the system with and without regulator.m have not been significantly affected with the exception of frequency, which is directly related to the gain of the poles.

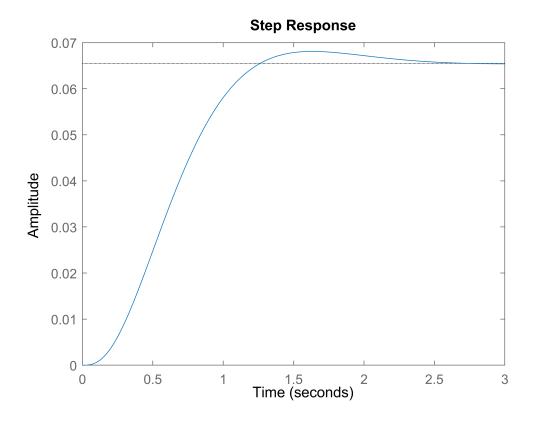
b)

```
desiredpoles = [dominant' 4 * real( dominant(1) ) ];
% Compute the controller gain K.
K = acker (Ag, Bg, desiredpoles);
% Compute the closed-loop state variable feedback system.
Asf = Ag - Bg * K; Bsf = Bg; Csf = Cg; Dsf = 0;
[numsf, densf] = ss2tf (Asf, Bsf, Csf, Dsf);
% Select observer poles to be 10 times faster than controller.
observerpoles = 10 * desiredpoles;
% Compute observer gain L.
L = acker (Ag', Cg', observerpoles);
% Compute the closed-loop system with controller and observer.
Areg = [ (Ag - Bg * K) Bg * K; zeros( size(Ag) ) (Ag - L' *Cg) ];
Breg = [ Bg; zeros( size(Bg) ) ];
Creg = [ Cg zeros ( size(Cg) ) ];
Dreg = 0;
[numreg, denreg] = ss2tf ( Areg, Breg, Creg, Dreg );
damp (denreg);
```

Pole	Damping	<pre>Frequency (rad/TimeUnit)</pre>	Time Constant (TimeUnit)
-8.48e+01	1.00e+00	8.48e+01	1.18e-02
-2.12e+01 + 2.12e+01i	7.07e-01	3.00e+01	4.71e-02
-2.12e+01 - 2.12e+01i	7.07e-01	3.00e+01	4.71e-02
-8.48e+00	1.00e+00	8.48e+00	1.18e-01

```
-2.12e+00 + 2.12e+00i 7.07e-01 3.00e+00 4.71e-01
-2.12e+00 - 2.12e+00i 7.07e-01 3.00e+00 4.71e-01

G2=tf(numsf,densf);
step(numsf,densf);
```



```
stepinfo(G2,'RiseTimeLimits',[0.1 0.9])
```

```
ans = struct with fields:
   RiseTime: 0.7647
TransientTime: 2.1069
SettlingTime: 2.1069
SettlingMin: 0.0593
SettlingMax: 0.0681
   Overshoot: 3.9657
Undershoot: 0
   Peak: 0.0681
   PeakTime: 1.6284
```

K

```
K = 1×3
6.7260 39.9891 76.3560
```

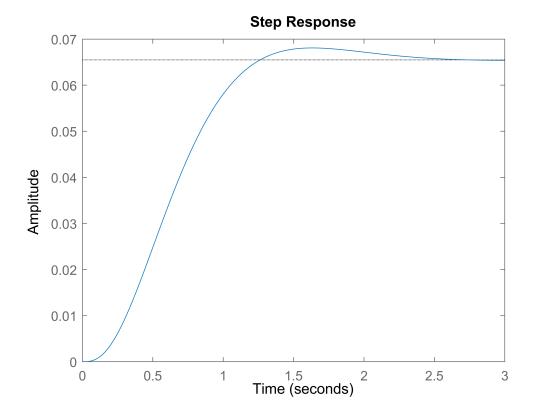
Compared to the other 2 set ups, in this case the obtained gain K is smaller, therefore the poles are closer to the imaginary axis. In general the closer non-dominant poles are to the dominant pole, the more significant effect they have on the overall system performance. In particular in our case they resulted in higher rise time and peak time compared to the previous instances.

c)

```
desiredpoles4 = [dominant' 10 * real( dominant(1) ) ];
% Compute the controller gain K.
K4 = acker (Ag, Bg, desiredpoles4);
% Compute the closed-loop state variable feedback system.
Asf = Ag - Bg * K; Bsf = Bg; Csf = Cg; Dsf = 0;
[numsf4, densf4] = ss2tf (Asf, Bsf, Csf, Dsf);
% Select observer poles to be 10 times faster than controller.
observerpoles4 = 20 * desiredpoles4;
% Compute observer gain L.
L4 = acker (Ag', Cg', observerpoles4);
% Compute the closed-loop system with controller and observer.
Areg4 = [ (Ag - Bg * K) Bg * K; zeros( size(Ag) ) (Ag - L4' *Cg) ];
Breg4 = [ Bg; zeros( size(Bg) ) ];
Creg4 = [ Cg zeros ( size(Cg) ) ];
Dreg4 = 0;
[numreg4, denreg4] = ss2tf ( Areg4, Breg4, Creg4, Dreg4 );
damp (denreg4);
```

```
Pole
                        Damping
                                     Frequency
                                                   Time Constant
                                    (rad/TimeUnit)
                                                     (TimeUnit)
-4.24e+02
                        1.00e+00
                                      4.24e+02
                                                       2.36e-03
-4.24e+01 + 4.24e+01i
                        7.07e-01
                                      6.00e+01
                                                       2.36e-02
                                                       2.36e-02
-4.24e+01 - 4.24e+01i
                       7.07e-01
                                      6.00e+01
                                     8.48e+00
-8.48e+00
                        1.00e+00
                                                       1.18e-01
-2.12e+00 + 2.12e+00i
                        7.07e-01
                                      3.00e+00
                                                       4.71e-01
-2.12e+00 - 2.12e+00i
                       7.07e-01
                                      3.00e+00
                                                       4.71e-01
```

```
G4=tf(numsf4,densf4);
step(numsf4,densf4);
```



```
stepinfo(G4,'RiseTimeLimits',[0.1 0.9])
```

```
ans = struct with fields:
    RiseTime: 0.7647
TransientTime: 2.1069
SettlingTime: 2.1069
SettlingMin: 0.0593
SettlingMax: 0.0681
    Overshoot: 3.9657
Undershoot: 0
    Peak: 0.0681
    PeakTime: 1.6284
```

```
K4 = 1 \times 3
19.4520 93.9728 190.8900
```

In this case, by increasing the distance of the observer poles (or their "speed") the values in the original L matrix are significantly smaller (2.7848e+04, 1.6802e+03, 49.7040 vs 2.6104e+05, 7.3132e+03, 100.6080). When the observer poles are faster, the state estimation is more responsive to changes in the system's dynamics. This can be beneficial in some cases because it allows the controller to respond quickly to changes in the system and improve the system's performance. However, faster observer poles can also introduce noise amplification and sensitivity to modeling errors, which can degrade the system's stability and performance.

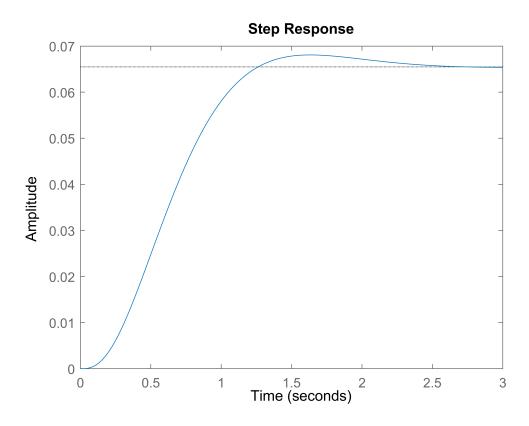
```
d)
```

```
desiredpoles5 = [dominant' 10 * real( dominant(1) ) ];
% Compute the controller gain K.
```

```
K5 = acker (Ag, Bg, desiredpoles5);
% Compute the closed-loop state variable feedback system.
Asf = Ag - Bg * K; Bsf = Bg; Csf = Cg; Dsf = 0;
[numsf5, densf5] = ss2tf (Asf, Bsf, Csf, Dsf);
% Select observer poles to be 10 times faster than controller.
observerpoles5 = 4 * desiredpoles5;
% Compute observer gain L.
L5 = acker (Ag', Cg', observerpoles5);
% Compute the closed-loop system with controller and observer.
Areg5 = [ (Ag - Bg * K) Bg * K; zeros( size(Ag) ) (Ag - L5' *Cg) ];
Breg5 = [ Bg; zeros( size(Bg) ) ];
Creg5 = [ Cg zeros ( size(Cg) ) ];
Dreg5 = 0;
[numreg5, denreg5] = ss2tf ( Areg5, Breg5, Creg5, Dreg5 );
damp (denreg5);
```

Pole	Damping	Frequency (rad/TimeUnit)	Time Constant (TimeUnit)
-8.48e+01	1.00e+00	8.48e+01	1.18e-02
-8.48e+00 + 8.49e+00i	7.07e-01	1.20e+01	1.18e-01
-8.48e+00 - 8.49e+00i	7.07e-01	1.20e+01	1.18e-01
-8.48e+00	1.00e+00	8.48e+00	1.18e-01
-2.12e+00 + 2.12e+00i	7.07e-01	3.00e+00	4.71e-01
-2.12e+00 - 2.12e+00i	7.07e-01	3.00e+00	4.71e-01

```
G5=tf(numsf5,densf5);
step(numsf5,densf5);
```



stepinfo(G5, 'RiseTimeLimits',[0.1 0.9]) ans = struct with fields: RiseTime: 0.7647 TransientTime: 2.1069 SettlingTime: 2.1069 SettlingMan: 0.0593 SettlingMax: 0.0681 Overshoot: 3.9657 Undershoot: 0 Peak: 0.0681 PeakTime: 1.6284 K5 K5 = 1×3

In this case observer gain L was equal to: 1.1431e+03 200.7434 19.1616. It's lower than in the previous case, because the "speed" of observer poles was smaller. There were no changes in RiseTime, Overshoot, PeakTime and so on. On the other hand we can see changes in poles, damping, frequency and time constant.

e)

19.4520

93.9728 190.8900

```
for wn=3:-0.001:0
    [ num2, den2 ] = ord2 (wn, damping);
    dominant=roots(den2);
    desiredpoles = [dominant' 10 * real( dominant(1) ) ];
    K = acker (Ag, Bg, desiredpoles);

if ((K(3)<10)&&(K(2)<10)&&(K(1)<10))
    wn
    K
    break
end</pre>
```

```
wn = 1.1220

K = 1 \times 3

3.5190 8.8439 9.9861
```

In this exercise we have shown that the maximum natural frequency omega n obtained while all the elements of matrix K are smaller than 10 is 1.122.