

## Precision, Low Noise, CMOS, Rail-to-Rail, Input/Output Operational Amplifier

**FEATURES**

- ▶ **Low offset voltage: 300 $\mu$ V maximum**
- ▶ **Low input bias currents: 5pA maximum**
- ▶ **Low noise: 9nV/ $\sqrt{\text{Hz}}$**
- ▶ **Wide bandwidth: 10MHz**
- ▶ **High slew rate: 5V/ $\mu$ s**
- ▶ **High open-loop gain: 120dB**
- ▶ **Unity gain stable**
- ▶ **Single-supply operation: 5.5V**
- ▶ **Low supply current: 1.15mA/amplifier**

**APPLICATIONS**

- ▶ **Photodiode amplification**
- ▶ **Optical control loops**
- ▶ **Battery-powered instrumentation**
- ▶ **Loop-powered instrumentation**
- ▶ **Multipole filters**
- ▶ **Sensors**
- ▶ **Barcode scanners**
- ▶ **Audio**

**GENERAL DESCRIPTION**

The MAX74821 is a dual rail-to-rail input and output, single-supply amplifier. The MAX74821 features very low offset voltage, low input voltage and current noise, and wide signal bandwidth. The MAX74821 uses the Analog Devices Inc. patented DigiTrim<sup>®</sup> trimming technique<sup>1</sup>, which achieves superior precision without the switching noise associated with chopping or auto-zeroing techniques.

The combination of low offsets, low noise, very low input bias currents, and high speed makes this amplifier useful in a wide variety of applications. Filters, integrators, photodiode amplifiers, and high impedance sensors all benefit from the combination of performance features. Audio and other AC applications benefit from the wide bandwidth and low distortion. Applications for these amplifiers include optical control loops, portable and loop-powered instrumentation, and audio amplification for portable devices.

The MAX74821 is specified over the extended industrial temperature range (–40°C to +125°C). The MAX74821 dual is available in an 8-lead MSOP package.

<sup>1</sup> Protected by U.S. Patent No. 5,969,657.

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## SPECIFICATIONS

## 5V Electrical Specifications

Table 1. Electrical Characteristics

(V<sub>S</sub> = 5V, V<sub>CM</sub> = V<sub>S</sub>/2, T<sub>A</sub> = 25°C, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
INPUT CHARACTERISTICS						
Offset Voltage	V <sub>OS</sub>	V <sub>S</sub> = 5V, V <sub>CM</sub> = 0V to 5V		80	300	μV
		−40°C < T <sub>A</sub> < +125°C		750		μV
Input Bias Current	I <sub>B</sub>			0.25	5	pA
		−40°C < T <sub>A</sub> < +125°C		175		pA
Input Offset Current	I <sub>OS</sub>			0.15	2.5	pA
		−40°C < T <sub>A</sub> < +125°C		85		pA
Input Voltage Range			−0.1		5.1	V
Common-Mode Rejection Ratio	CMRR	V <sub>CM</sub> = 0V to 5V	85	100		dB
		−40°C < T <sub>A</sub> < +125°C		90		dB
Large Signal Voltage Gain	A <sub>VO</sub>	R <sub>L</sub> = 2kΩ, V <sub>O</sub> = 0.5V to 4.5V	108	120		dB
Offset Voltage Drift	ΔV <sub>OS</sub> /ΔT	−40°C < T <sub>A</sub> < +125°C		1		μV/°C
INPUT CAPACITANCE						
Common-Mode Input Capacitance	C <sub>COM</sub>			8.8		pF
Differential Input Capacitance	C <sub>DIFF</sub>			2.6		pF
OUTPUT CHARACTERISTICS						
Output Voltage High	V <sub>OH</sub>	I <sub>L</sub> = 1mA	4.96	4.98		V
		−40°C < T <sub>A</sub> < +125°C		4.6		V
Output Voltage Low	V <sub>OL</sub>	I <sub>L</sub> = 1mA		20	40	mV
		−40°C < T <sub>A</sub> < +125°C		290		mV
Output Current	I <sub>OUT</sub>			±80		mA
Closed-Loop Output Impedance	Z <sub>OUT</sub>	f = 1MHz, A <sub>v</sub> = +1		1		Ω
POWER SUPPLY						
Power Supply Rejection Ratio	PSRR	V <sub>S</sub> = 2.7V to 5.5V	80	95		dB
		−40°C < T <sub>A</sub> < +125°C		90		dB
Supply Current/Amplifier	I <sub>SY</sub>	I <sub>OUT</sub> = 0mA		1	1.15	mA
		−40°C < T <sub>A</sub> < +125°C		1.4		mA
DYNAMIC PERFORMANCE						
Slew Rate	SR	R <sub>L</sub> = 2kΩ, C <sub>L</sub> = 16pF		5		V/μs

( $V_S = 5V$ ,  $V_{CM} = V_S/2$ ,  $T_A = 25^\circ C$ , unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Settling Time	$t_s$	To 0.01%, 0V to 2V step $A_V = +1$		<1		$\mu s$
Unity Gain Bandwidth Product	GBP			10		MHz
Phase Margin	$\Phi_M$			65		Degrees
<b>NOISE PERFORMANCE</b>						
Peak-to-Peak Noise	$e_n$ p-p	$f = 0.1Hz$ to $10Hz$		2.3		$\mu V$ p-p
Voltage Noise Density	$e_n$	$f = 1kHz$		9		$nV/\sqrt{Hz}$
		$f = 10kHz$		7.3		$nV/\sqrt{Hz}$
Current Noise Density	$i_n$	$f = 1kHz$		10		$fA/\sqrt{Hz}$

## ABSOLUTE MAXIMUM RATINGS

$T_A = 25^\circ\text{C}$ , unless otherwise specified.

**Table 2. Absolute Maximum Ratings**

PARAMETER	RATING
Supply Voltage	6V
Input Voltage	GND – 0.3V to $V_S + 0.3\text{V}$
Differential Input Voltage	$\pm 5\text{V}$
Input Current (Indefinite)	$\pm 10\text{mA}$
Input Current (Duration < 1 sec)	$\pm 100\text{mA}$
Output Short-Circuit Duration to GND	Indefinite
Operating Temperature Range	$-40^\circ\text{C}$ to $+125^\circ\text{C}$
Storage Temperature Range	$-65^\circ\text{C}$ to $+150^\circ\text{C}$
Junction Temperature Range	$-65^\circ\text{C}$ to $+150^\circ\text{C}$
Lead Temperature (Soldering, 60 sec)	$300^\circ\text{C}$
ESD HBM	$\pm 4000\text{V}$
ESD FICDM 8-Lead MSOP	$\pm 1000\text{V}$

Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

**Table 3. Package Type**

PACKAGE TYPE	$\theta_{JA}^1$	$\theta_{JC}$	UNIT
8-Lead MSOP (RM)	206	44	$^\circ\text{C/W}$

<sup>1</sup>  $\theta_{JA}$  is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages.

## ESD Caution



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

## PIN CONFIGURATION

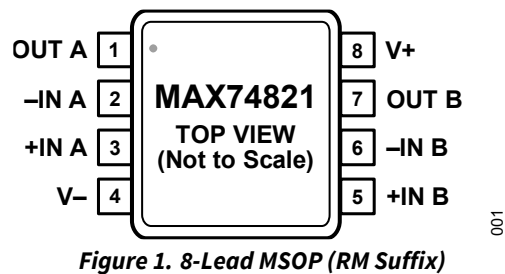


Figure 1. 8-Lead MSOP (RM Suffix)

## TYPICAL PERFORMANCE CHARACTERISTICS

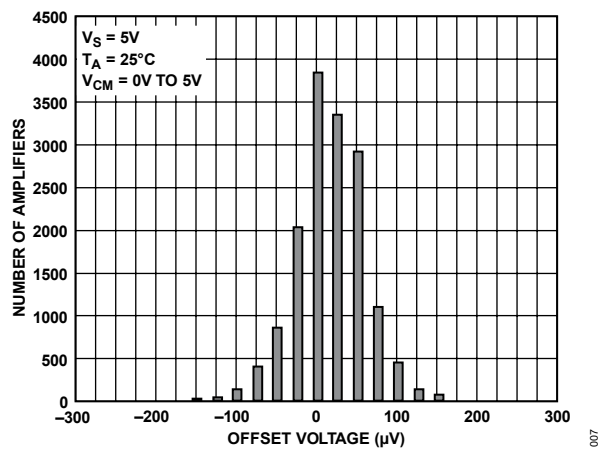


Figure 2. Input Offset Voltage Distribution

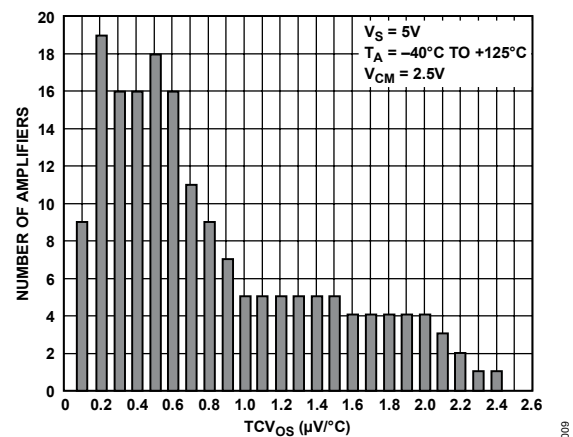


Figure 3. Input Offset Voltage Drift Distribution

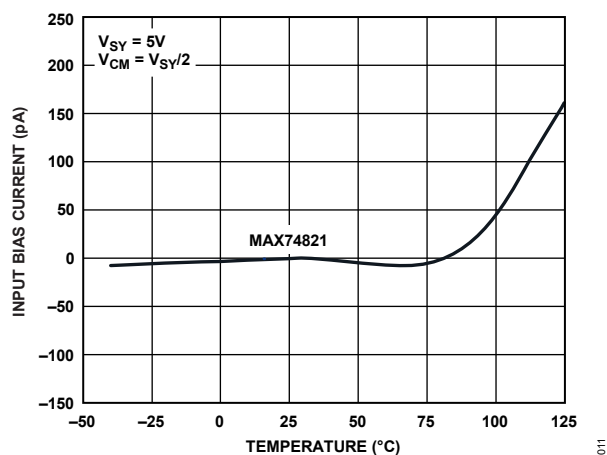


Figure 4. Input Bias Current vs. Temperature

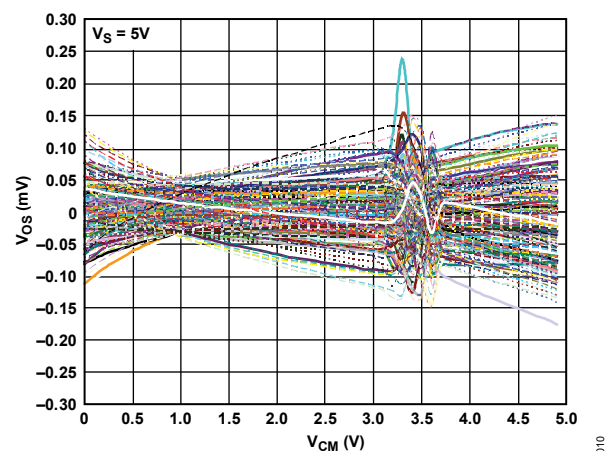
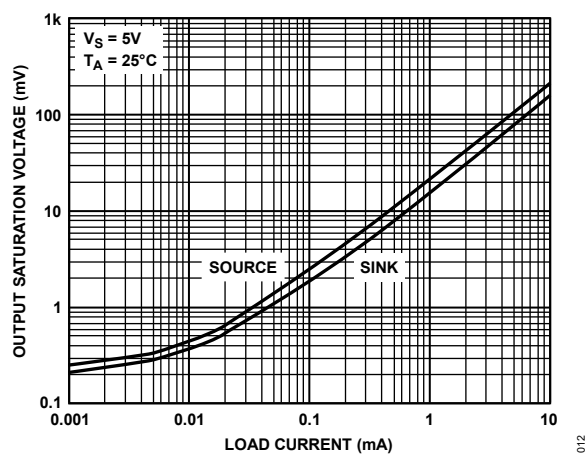
Figure 5. Input Offset Voltage vs. Common-Mode Voltage  
(200 Units, 5 Wafer Lots, Including Process Skews)

Figure 6. Output Saturation Voltage vs. Load Current

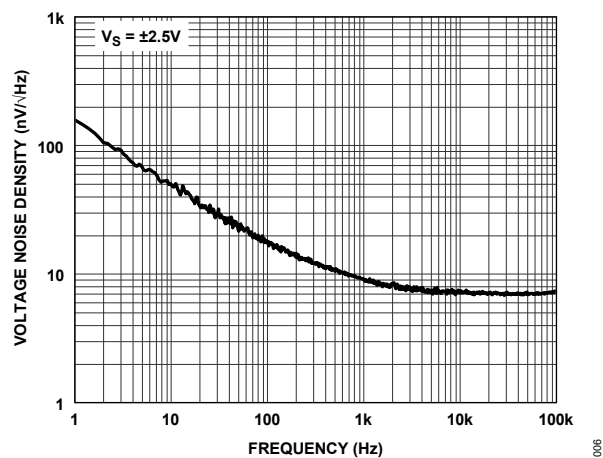


Figure 7. Voltage Noise Density vs. Frequency

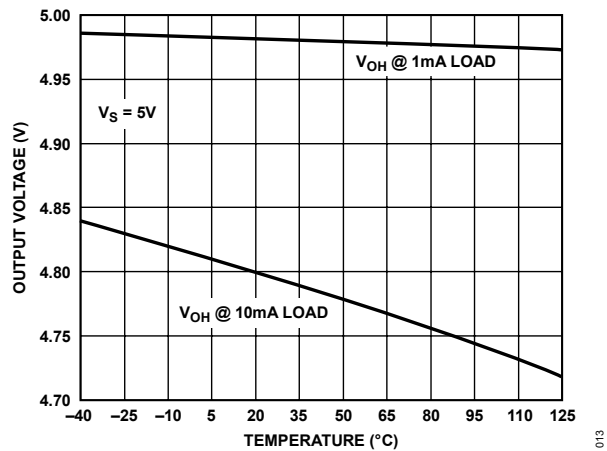


Figure 8. Output Voltage Swing High vs. Temperature

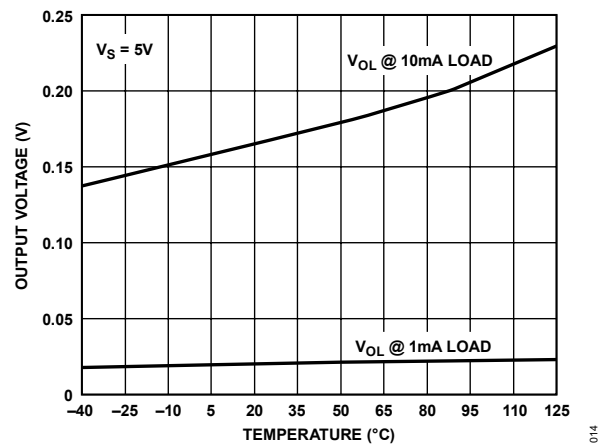


Figure 9. Output Voltage Swing Low vs. Temperature

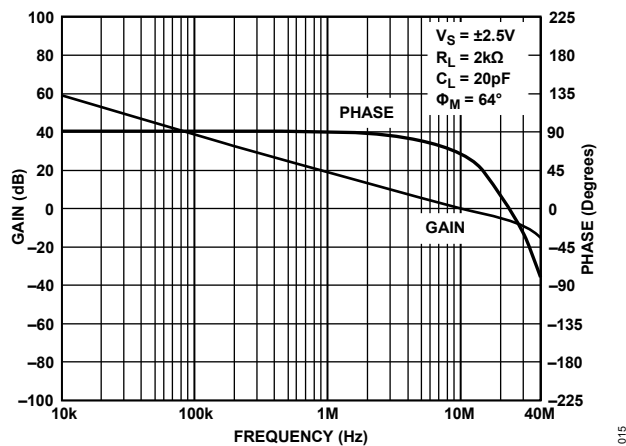


Figure 10. Open-Loop Gain and Phase vs. Frequency

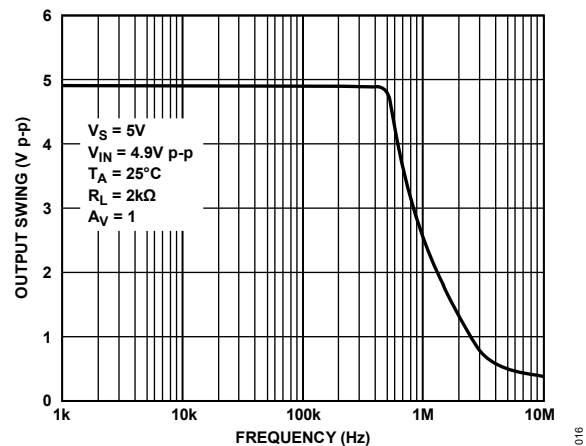


Figure 11. Closed-Loop Output Voltage Swing (FPBW)

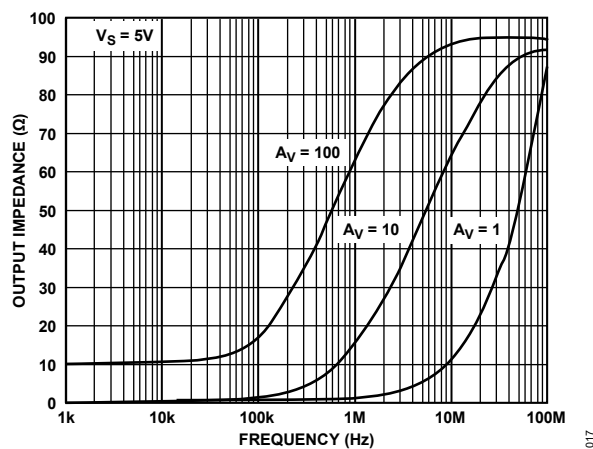


Figure 12. Output Impedance vs. Frequency

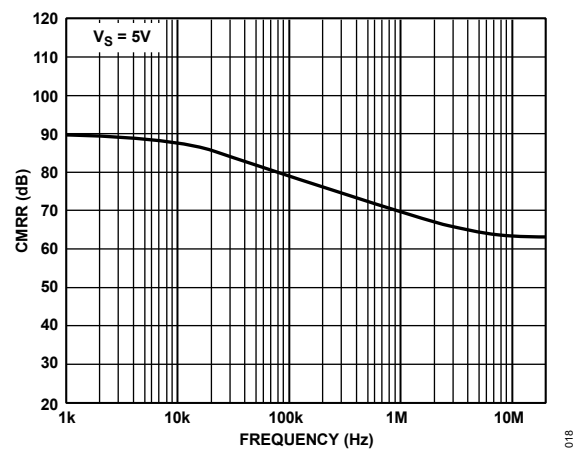


Figure 13. Common-Mode Rejection Ratio (CMRR) vs. Frequency



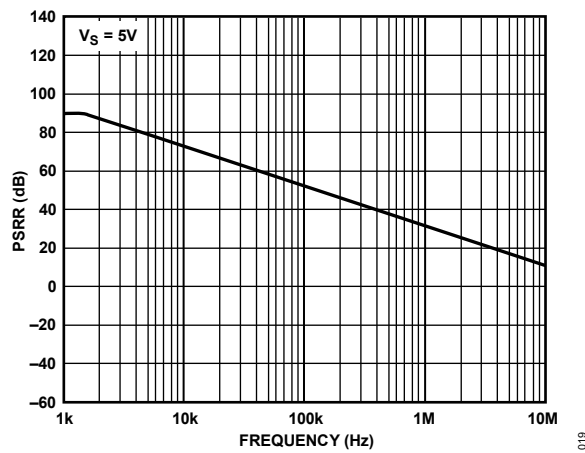


Figure 14. PSRR vs. Frequency

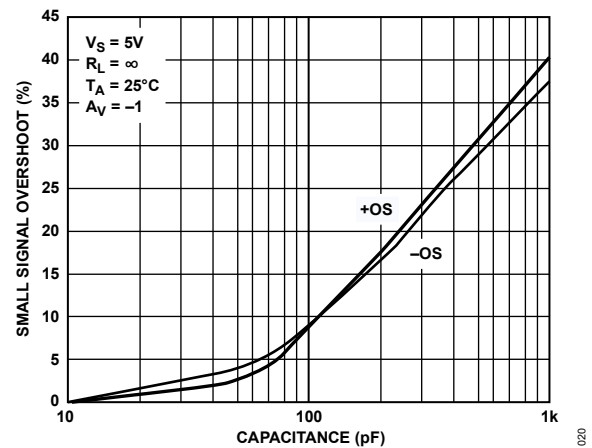


Figure 15. Small Signal Overshoot vs. Load Capacitance

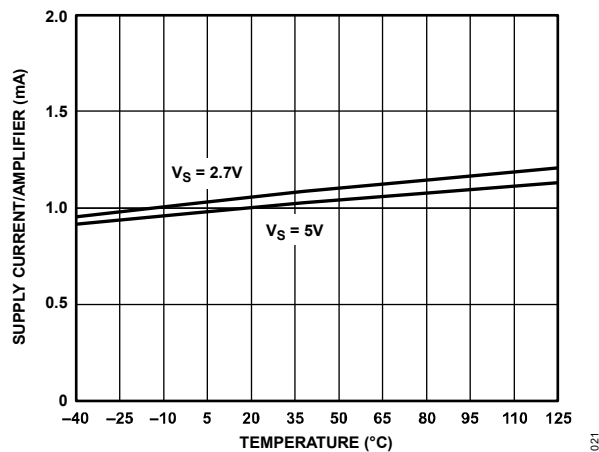


Figure 16. Supply Current/Amplifier vs. Temperature

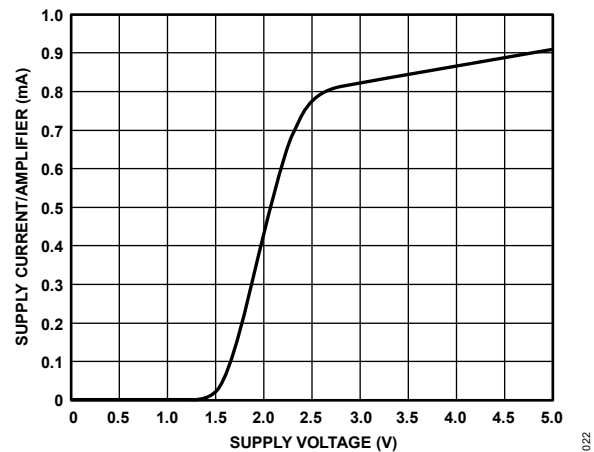


Figure 17. Supply Current/Amplifier vs. Supply Voltage

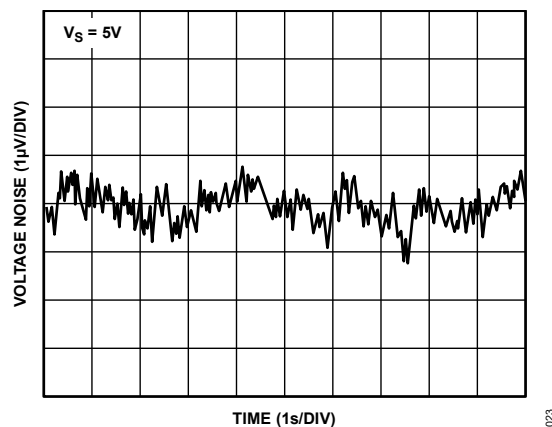


Figure 18. 0.1Hz to 10Hz Input Voltage Noise

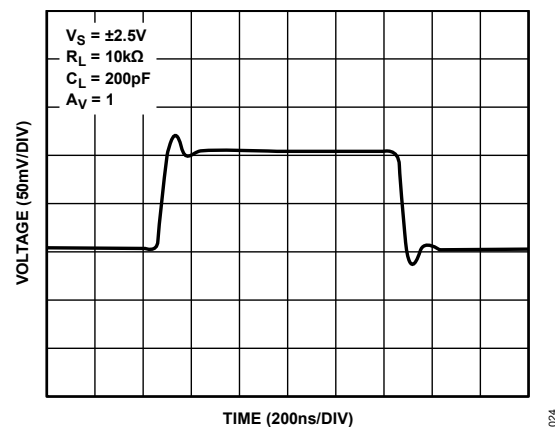


Figure 19. Small Signal Transient Response

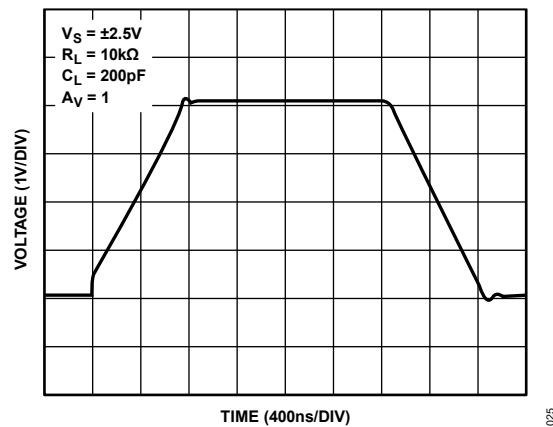


Figure 20. Large Signal Transient Response

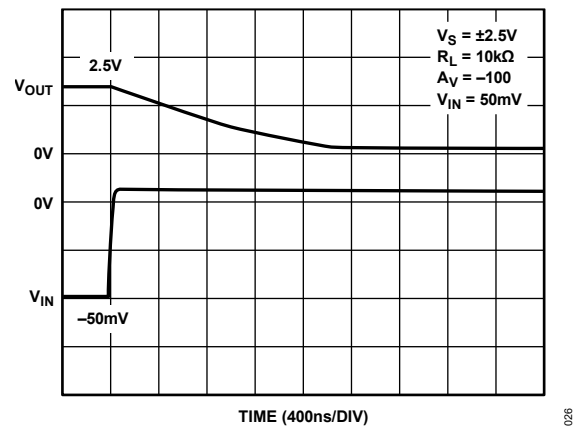


Figure 21. Positive Overload Recovery

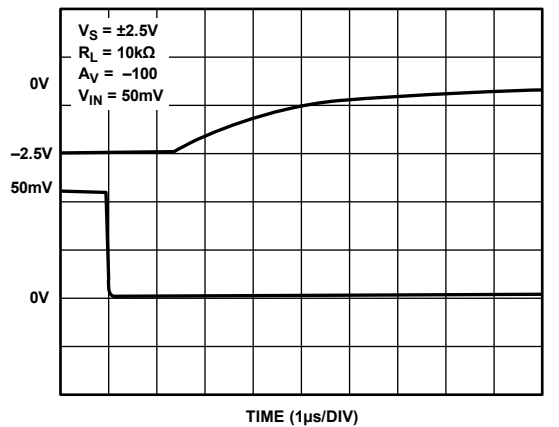


Figure 22. Negative Overload Recovery

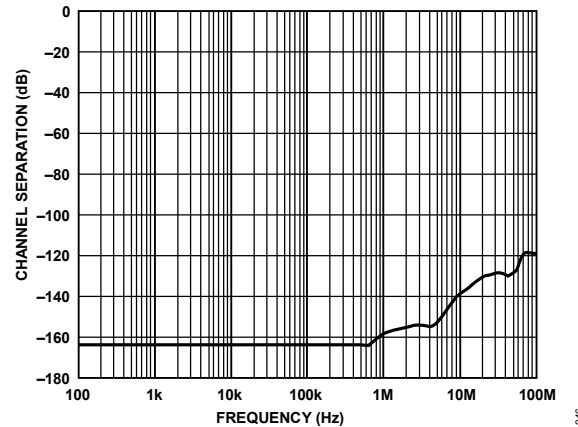


Figure 23. Channel Separation vs. Frequency

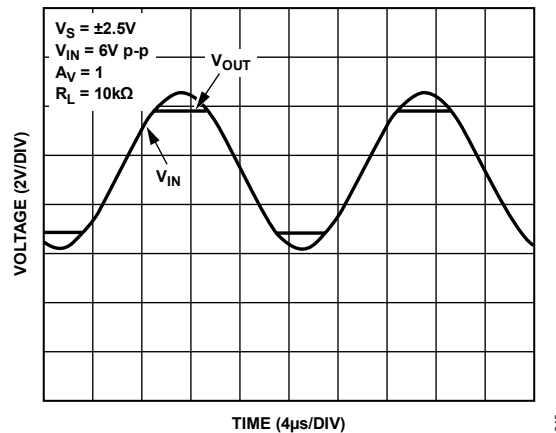


Figure 24. No Phase Reversal

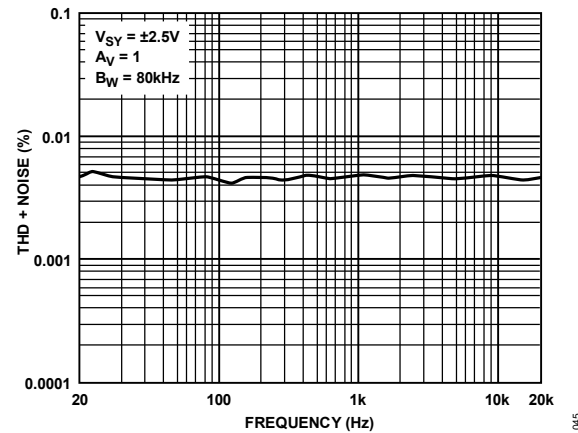
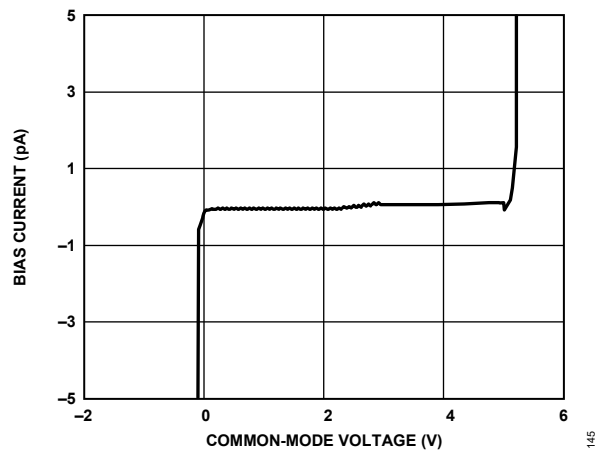
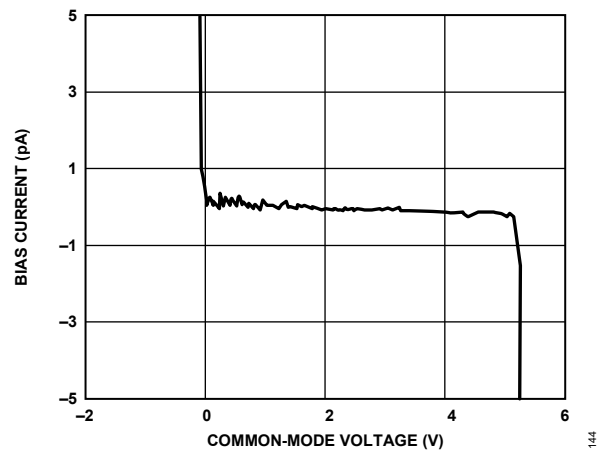


Figure 25. THD + Noise vs. Frequency



**Figure 26. Noninverting Input Bias Current vs. Common-Mode Voltage**



**Figure 27. Inverting Input Bias Current vs. Common-Mode Voltage**

## APPLICATIONS INFORMATION

### Output Phase Reversal

Phase reversal is defined as a change in polarity at the output of the amplifier when a voltage that exceeds the maximum input common-mode voltage drives the input.

Phase reversal can cause permanent damage to the amplifier; it can also cause system lockups in feedback loops. The MAX74821 does not exhibit phase reversal even for inputs exceeding the supply voltage by more than 2 V.

### Maximum Power Dissipation

Power dissipated in an IC causes the die temperature to increase, which can affect the behavior of the IC and the application circuit performance.

The absolute maximum junction temperature of the MAX74821 is 150°C. Exceeding this temperature can damage or destroy the device.

The maximum power dissipation of the amplifier is calculated according to:

$$P_{DISS} = \frac{T_J - T_A}{\theta_{JA}}$$

where:

$T_J$  is the junction temperature.

$T_A$  is the ambient temperature.

$\theta_{JA}$  is the junction-to-ambient thermal resistance.

### Input Overvoltage Protection

The MAX74821 has internal protective circuitry. However, if the voltage applied at either input exceeds the supplies by more than 0.3V, external resistors should be placed in series with the inputs. The resistor values can be determined by:

$$\frac{V_{IN} - V_S}{R_S} \leq 10\text{mA}$$

The remarkable low input offset current of the MAX74821 (<1pA) allows the use of larger value resistors. With a 10kΩ resistor at the input, the output voltage has less than 10nV of error voltage.

A 10kΩ resistor has less than 13nV/√Hz of thermal noise at room temperature.

## OUTLINE DIMENSIONS

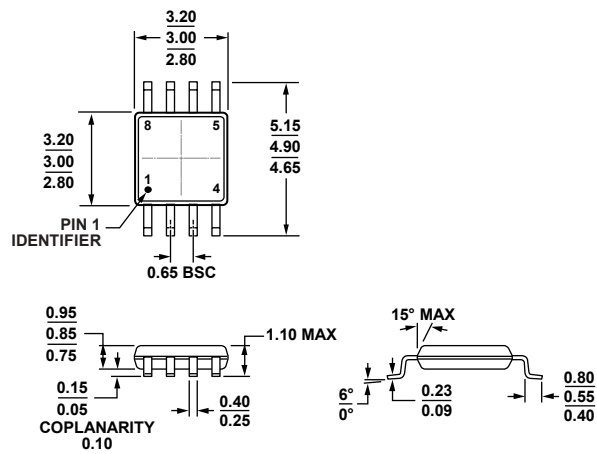


Figure 28. 8-Lead Mini Small Outline Package [MSOP] (RM-8) Dimensions shown in millimeters

## ORDERING GUIDE

Table 4. Ordering Guide

MODEL <sup>1</sup>	TEMPERATURE RANGE	PACKAGE DESCRIPTION	PACKAGE OPTION	Marking Code
MAX74821ARMZ	-40°C to +125°C	8-Lead MSOP	RM-8	A6Q
MAX74821ARMZ-R7	-40°C to +125°C	8-Lead MSOP	RM-8	A6Q
MAX74821ARMZ-RL	-40°C to +125°C	8-Lead MSOP	RM-8	A6Q

<sup>1</sup> Z = RoHS compliant part

## REVISION HISTORY

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGE NUMBER
0	04/25	Initial release	—

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