

OTFS Linear Receivers and MAP

OTFS System model

- Recall, the end-to-end DD-domain relationship.

$$\bar{Y}_{DD} = H_{DD} \bar{\pi}_{DD} + \bar{V}_{DD}$$

$\uparrow M \times 1$ $\uparrow M \times 1$

where,

$$H = \sum_{i=1}^{L_p} h_i (\pi)^{l_i} (\Delta)^{k_i}$$

- For rectangular pulse : $P_{Rx} = P_{Tx} = I_M$

Problem 1 : What is the expression for H ?

$$H = \sum_{i=1}^{L_p} h_i (\pi)^{l_i} (\Delta)^{k_i}$$

where,

$\pi_i = l_i \Delta \tau \rightarrow$ Integer multiple of
Delay resolution ($\Delta \tau$)

$v_i = k_i \Delta v \rightarrow$ Integer multiple of
Doppler resolution (Δv)

Problem 2 :

Net end-to-end system model is.

$$\bar{Y}_{DD} = H_{DD} \bar{\pi}_{DD} + \bar{V}_{DD}$$

What is the expression for H_{DD} ?

$$H_{DD} = (F_N \otimes I_M) H (F_N^H \otimes I_M)$$

Signal Detection in OTFS

- Linear Receivers
 - ① ZF Receiver
 - ② LMMSE Receiver

- Non-linear Receiver

- ① Message-Passing detector

ZF Receiver

- ① ZF Receiver minimizes Least-Squares (LS) cost function.

$$\min_{\hat{x}_{DD}} \left\| \bar{y}_{DD} - H_{DD}^H \hat{x}_{DD} \right\|^2$$

- ② The ZF estimate is

$$\hat{x}_{DD}^{ZF} = \underbrace{\left(H_{DD}^H H_{DD} \right)^{-1} H_{DD}^H \bar{y}_{DD}}_{\text{Pseudo inverse of } H_{DD}}$$

$\left(H_{DD}^H H_{DD} \right)^{-1} H_{DD}^H$ is the pseudo-inverse of H_{DD} .

$$(ii) \quad \left(H_{DD}^H H_{DD} \right)^{-1} H_{DD}^H \times H_{DD} = I$$

LMMSE Receiver

- ③ LMMSE Receiver minimizes the cost function

$$E \left\{ \left\| \hat{x}_{DD}^{LMMSE} - \bar{x}_{DD} \right\|^2 \right\}$$

- ④ Consider independent information symbols, with average power P. (ii)

$$R_{xx} = E \left[\bar{x}_{DD} \bar{x}_{DD}^H \right] = P I_{MN}$$

- ⑤ LMMSE Receiver is

$$\hat{x}_{DD}^{LMMSE} = \left(H_{DD}^H H_{DD} + \frac{\sigma^2}{P} I_{MN} \right)^{-1} H_{DD}^H \bar{y}_{DD}$$

$$\hat{x}_{DD}^{LMMSE} = \left(H_{DD}^H H_{DD} + \frac{1}{SNR} I_{MN} \right)^{-1} H_{DD}^H \bar{y}_{DD}$$

- ⑥ As $SNR \rightarrow \infty$, LMMSE \Rightarrow ZF receiver.

Non-Linear Receiver

- Note that the matrix $H = \sum_{i=1}^{L_p} h_i(\pi)^{l_i} (\Delta)^{k_i}$ is SPARSE, since it has only L_p non-zero entries in each row and column, and $L_p \ll MN$.

Note: SPARSE \rightarrow More entries are Zeros
Few entries are non-zeros } in Matrix

- Hence, H_{DD} is also SPARSE.
- ZF and LMMSE based linear receivers do not exploit this sparsity, which can significantly simplify detection.

Message Passing (MP) Detector

- The MP detector exploits the sparsity of the matrix H_{DD} , employs the Maximum a Posteriori (MAP) detection rule.
- symbol-by-symbol MAP detection rule

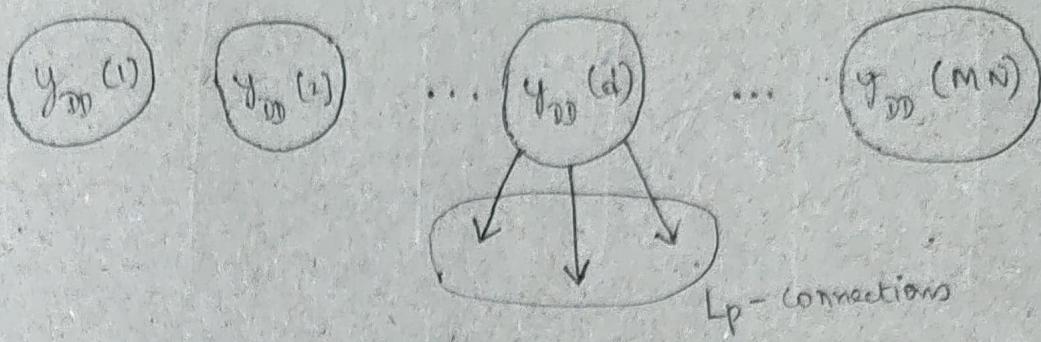
$$\hat{x}_{DD}(c) = \arg \max_{a_j \in A} \Pr(x_{DD}(c) = a_j | \bar{y}_{DD}, H_{DD})$$
- $\mathbb{I}(d)$ denotes the set of indices of the non-zero elements in the d^{th} row of H_{DD} .
- $\mathbb{J}(c)$ denotes the set of indices of the non-zero elements in the c^{th} column of H_{DD} .
- Number of elements in $\mathbb{I}(d)$ and $\mathbb{J}(c)$ is L_p each.

Note:

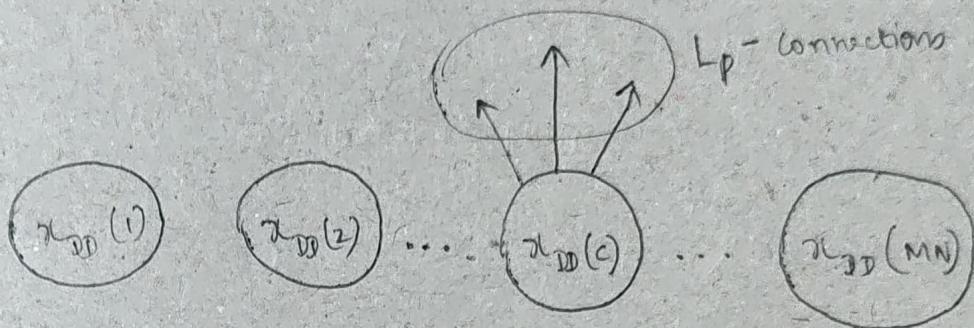
$$\mathbb{I}(d) = \{c_1, c_2, \dots, c_{L_p}\} \rightarrow \text{Observation Node messages}$$

$$\mathbb{J}(c) = \{d_1, d_2, \dots, d_{L_p}\} \rightarrow \text{Variable Node messages}$$

- ① Elements of \mathcal{Y}_{DD} become Observation nodes.



- ② Elements of \mathcal{X}_{DD} become Variable nodes.



- ③ The factor graph is sparingly connected.

- Any Observation node $y_{DD}(d)$ is connected to only L_p variable node $x_{DD}(c)$
- Any variable node $x_{DD}(c)$ is connected to only L_p observation nodes $y_{DD}(d)$

- ④ MP detector can solve this symbol-by-symbol MAP detection problem. with linear complexity in MN .

④ End-to-end model

$$\begin{bmatrix} y_{DD}(1) \\ y_{DD}(2) \\ \vdots \\ y_{DD}(MN) \end{bmatrix} = \sum_{c=1}^{MN} \begin{bmatrix} H_{DD}(1,c) \\ H_{DD}(2,c) \\ \vdots \\ H_{DD}(MN,c) \end{bmatrix} x_{DD}(c) + \begin{bmatrix} v_{DD}(1) \\ v_{DD}(2) \\ \vdots \\ v_{DD}(MN) \end{bmatrix}$$

⑤ dth equation

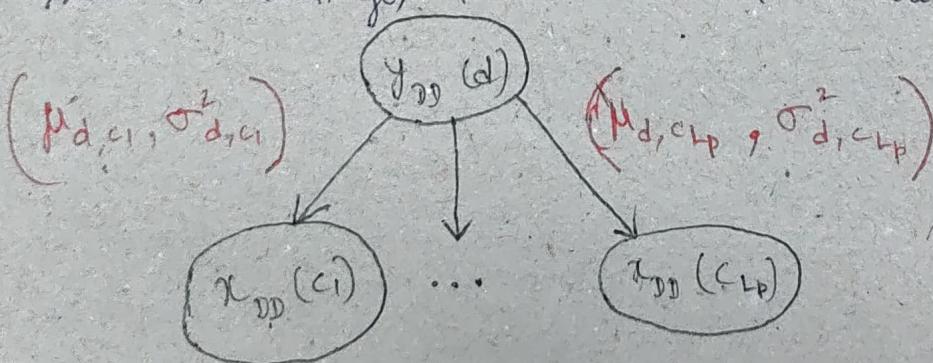
$$y_{DD}(d) = H_{DD}(d,1) x_{DD}(1) + H_{DD}(d,2) x_{DD}(2) + \dots + H_{DD}(d,MN) x_{DD}(MN) + v_{DD}(d)$$

$$= \sum_{c=1}^{MN} H_{DD}(d,c) x_{DD}(c) + v_{DD}(d)$$

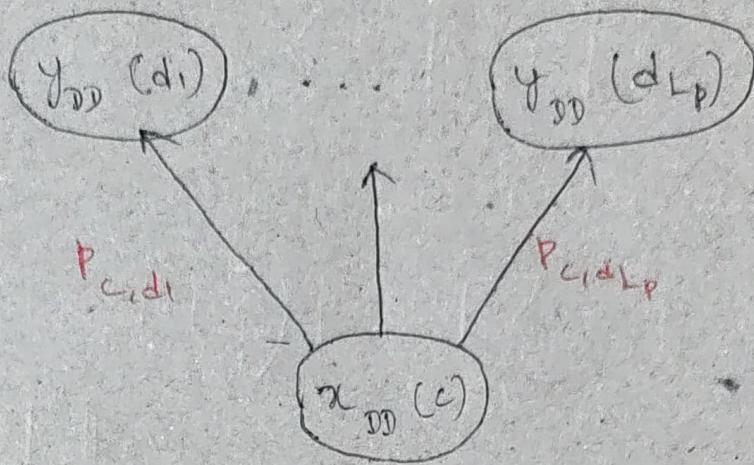
⑥ For a given $x_{DD}(c)$, the interference term $\zeta_{d,c}$ is assumed to be Gaussian.

$$y_{DD}(d) = H_{DD}(d,c) x_{DD}(c) + \underbrace{\sum_{\substack{e \neq c \\ e \in I(d)}} H_{DD}(d,e) x_{DD}(e) + v_{DD}(d)}_{\zeta_{d,c}}$$

⑦ From any observation mode $y_{DD}(d)$, the mean $\mu_{d,c}$ and variance $\sigma_{d,c}^2$ of the interference terms are sent as messages to the connected variable nodes $x_{DD}(c)$.



- ① Messages from any variable node $x_{DD}(c)$ (a) the probability mass function (pmf) of the alphabet, is passed as message to the observation mode $y_{DD}(d)$.



② Observation mode message

- Message from the observation mode $y_{DD}(d)$ to any connected variable node $x_{DD}(c)$, are the mean $\mu_{d,c}^{(i)}$ and variance $(\sigma_{d,c}^{(i)})^2$ of the interference

$$\zeta_{d,c}$$

- Recall,

$$y_{DD}(d) = H_{DD}(d,c) x_{DD}(c)$$

$$+ \sum_{\substack{e \neq c \\ e \in \mathbb{I}(d)}} H_{DD}(d,e) x_{DD}(e) + \mathcal{V}_{DD}(d)$$

$\zeta_{d,c}$

- Let i denote the iteration. It follows that

$$\mathbb{E}\left[\zeta_{d,c}^{(i)}\right] = \mu_{d,c}^{(i)} = \sum_{\substack{e \neq c \\ e \in \mathbb{I}(d)}} H_{DD}(d,e) \mathbb{E}[x_{DD}(e)]$$

- let $P_{e,d}^{(i-1)}(a_j)$ denote the probability of symbol a_j , passed by the e^{th} variable node $x_{DD}(e)$ to the d^{th} observation node $y_{DD}(d)$ after iteration $i-1$.
- Hence, $E[x_{DD}(e)] = \sum_{j=1}^Q a_j P_{e,d}^{(i-1)}(a_j)$

- The Variance of interference term is

$$\left(\sigma_{e,c}^{(i)}\right)^2 = \sum_{e \neq c} |H_{DD}(d,e)|^2 \left(\sum_{j=1}^Q |a_j|^2 P_{e,d}^{(i-1)}(a_j) - E^2[x_{DD}(e)] \right)$$

② Variable Node Messages

- Message from variable node $x_{DD}(c)$ to any connected observation node $y_{DD}(d)$, is the pmf vector $P_{c,d}^{(i)}$, which contains the probabilities $P_{c,d}^{(i)}(a_j)$ of the symbols a_j .
- Note that, for any $e \in J(c)$

$$P(y_{DD}(e) | x_{DD}(c) = a_j, H_{DD})$$

$$\propto \exp \left(\frac{-|y_{DD}(e) - \bar{\mu}_{e,c}^{(i)} - H_{DD}(e,c)a_j|^2}{(\sigma_{e,c}^{(i)})^2} \right)$$

$$= \xi^{(i)}(e, c, j)$$

- Upon normalization $\sum_{k=1}^q \xi^{(i)}(e, c, k) = 1$
 - $P\left(Y_{Dj}(e) \mid n_{Dj}(c) = a_j, H_{Dj}\right) = \frac{\xi^{(i)}(e, c, j)}{\sum_{k=1}^q \xi^{(i)}(e, c, k)}$
 - Hence,
- $$\tilde{P}_{c,d}^{(i)}(a_j) = \frac{\prod_{e \in J(c)} e^{a_j} \frac{\xi^{(i)}(e, c, j)}{\sum_{k=1}^q \xi^{(i)}(e, c, k)}}{\sum_{j=1}^q \prod_{e \in J(c)} e^{a_j} \frac{\xi^{(i)}(e, c, j)}{\sum_{k=1}^q \xi^{(i)}(e, c, k)}}$$
- The j^{th} element $\tilde{P}_{c,d}^{(i)}(a_j)$ of the pmf vector $\tilde{P}_{c,d}^{(i)}$ is updated as

$$P_{c,d}^{(i)}(a_j) = \Delta \tilde{P}_{c,d}^{(i)}(a_j) + (1-\Delta) P_{c,d}^{(i-1)}(a_j)$$

where, $\Delta \in (0, 1]$ is the damping factor.

② Convergence Indicator $\eta^{(i)}$

- At node $c \in C$ define

$$P_c^{(i)}(a_j) = \frac{\prod_{e \in J(c)} \left(\frac{\xi^{(i)}(e, c, j)}{\sum_{k=1}^q \xi^{(i)}(e, c, k)} \right)}{\sum_{j=1}^q \prod_{e \in J(c)} \left(\frac{\xi^{(i)}(e, c, j)}{\sum_{k=1}^q \xi^{(i)}(e, c, k)} \right)}$$

- Convergence indicator: For some small $\gamma > 0$

$$\eta^{(i)} = \frac{1}{NM} \sum_{c=1}^{NM} \text{Ind}\left(\max_j P_c^{(i)}(a_j) \geq 1 - \gamma\right)$$

Update Decision

When $\eta^{(i)} > \eta^{(i-1)}$, the decision $\hat{x}(c)$ of each transmitted symbol is updated as

$$\hat{x}(c) = \arg \max_{a_j} p_c^{(i)}(a_j)$$

Stopping Rule

- The MP algorithm stops when any of the following conditions holds

$$\eta^{(i)} \approx 1$$

- Maximum number of iterations is reached
- $\eta^{(i)} < \eta^{(i^*)} - \epsilon$, where $0 \leq i^* \leq i-1$ is the iteration such that $\eta^{(i^*)}$ is maximum.