

# RF Processing

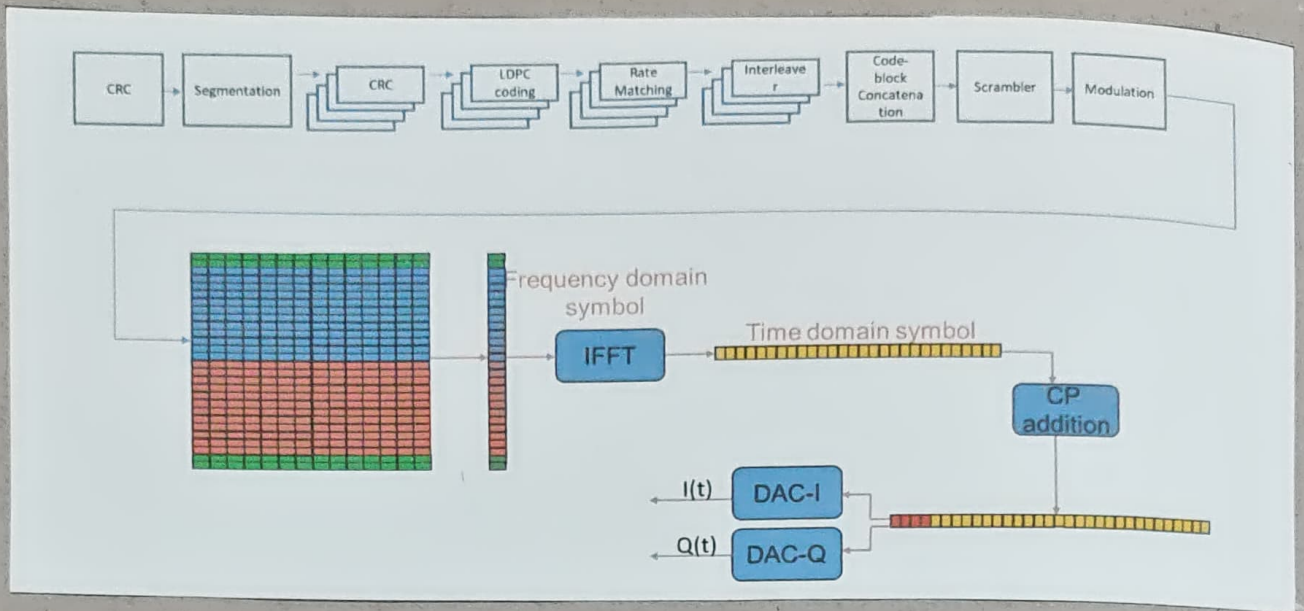


Fig: 5G Transmit Chain

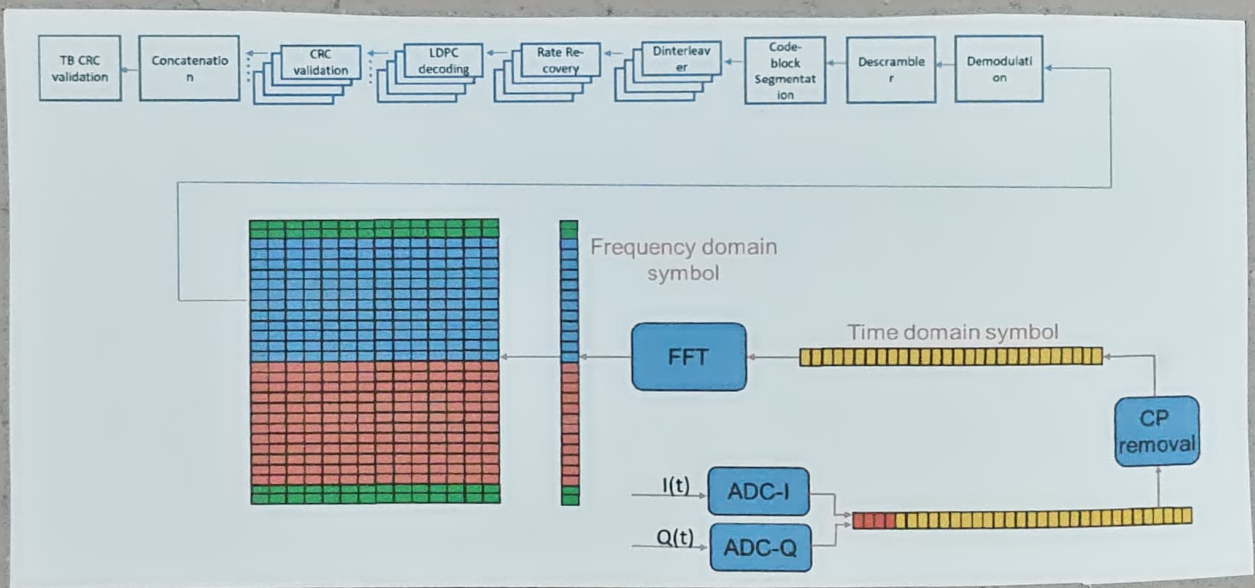


Fig: 5G Receive Chain

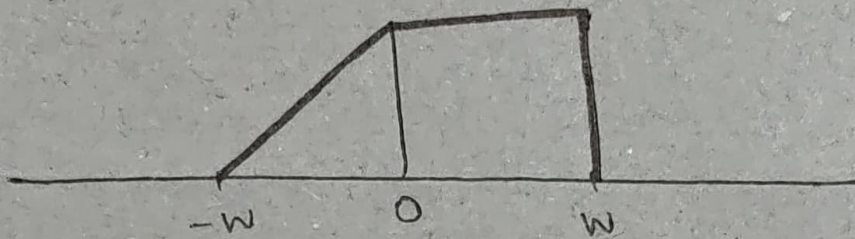
## Baseband - passband system model

- ① Transmit complex waveform

$$s(t) = I(t) + j Q(t)$$

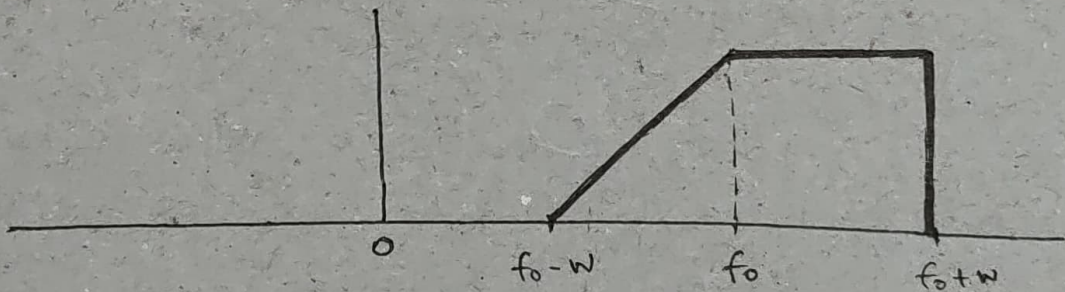
is called as baseband signal

- ②  $s(t)$  will have asymmetric spectrum around origin with bandwidth  $-W$  to  $W$



- ③ Upconvert the baseband signal to desired center frequency  $f_0$  with  $\omega_0 = 2\pi \cdot f_0$

$$s_1(t) = s(t) \sqrt{2} e^{j\omega_0 t}$$



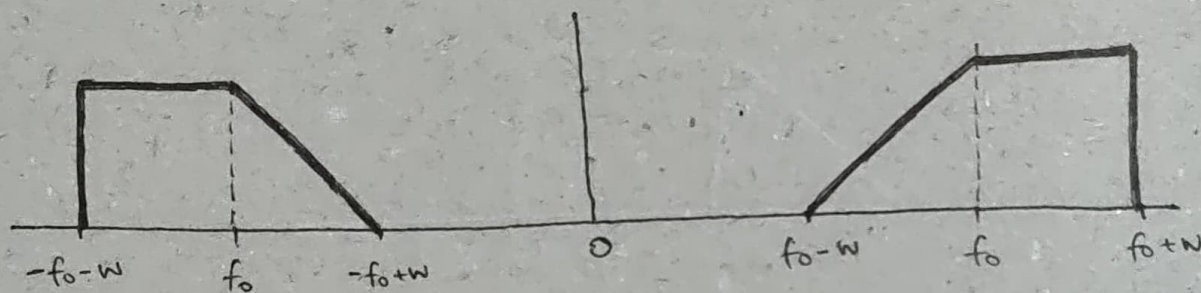


## Transmit Pass band system model

- ① We can transmit only real signals

$$s^o(t) = \Re(s_1(t))$$

$$= \sqrt{2} \Re(s_1(t) e^{j\omega_0 t})$$



- ①  $s_1(t)$  is called passband / RF / upconverted transmit signal

- ① Real transmit signal can equivalently be written as

$$s^o(t) = \Re(s_1(t)) = \sqrt{2} \Re(s_1(t) e^{j\omega_0 t})$$

$$= \sqrt{2} \Re([I(t) + jQ(t)] e^{j\omega_0 t})$$

$$= \sqrt{2} I(t) \cos(\omega_0 t) - \sqrt{2} Q(t) \sin(\omega_0 t)$$

- ① Architecture is called balanced homodyne transmitter

- ① We assume that channel is not faded.



## Receive Passband system model

① Receive signal is :

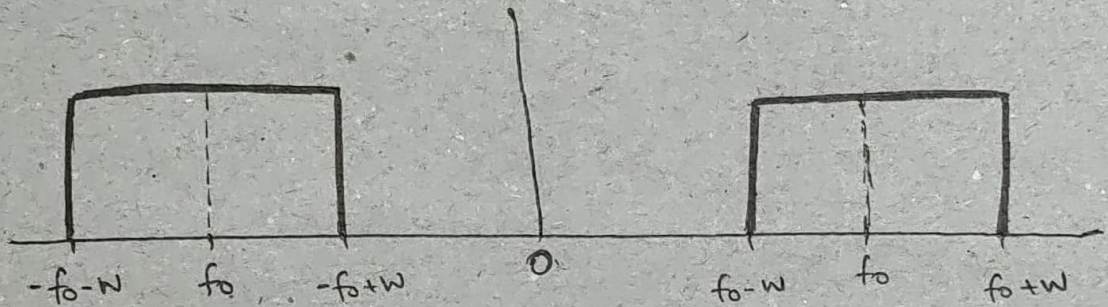
$$r_1(t) = s^o(t) + n_w(t)$$

$$= \sqrt{2} I(t) \cos(\omega_0 t) - \sqrt{2} Q(t) \sin(\omega_0 t) + n_w(t)$$

② First step in recovering baseband signal

- Limit the band pass noise  $n_w(t)$

③ Filter the receive signal  $r_1(t)$  using a band pass filter  $W_b(f)$



④ Received equivalent signal is

$$r(t) = r_1(t) \otimes w_0(t)$$

$$= (s^o(t) + n_w(t)) \otimes w_0(t)$$

$$= s^o(t) + (n_w(t) \otimes w_0(t))$$

$$= s^o(t) + m(t)$$



## Receive baseband system model

⊙ Demodulate inphase signal  $I(t)$

$$r_c(t) = [r(t) \sqrt{2} \cos(\omega_0 t)]_{\text{LPF}}$$

$$= \left[ \frac{\sqrt{2} I(t) \cos(\omega_0 t) - \sqrt{2} Q(t) \sin(\omega_0 t) + n(t)}{\sqrt{2} \cos(\omega_0 t)} \right]_{\text{LPF}}$$

$$= \left[ \frac{2 I(t) \cos^2(\omega_0 t) - Q(t) \sin(2\omega_0 t) + \sqrt{2} n(t) \cos(\omega_0 t)}{1} \right]_{\text{LPF}}$$

$$= \left[ I(t) + I(t) \cos(2\omega_0 t) - Q(t) \sin(2\omega_0 t) + \sqrt{2} n(t) \cos(\omega_0 t) \right]_{\text{LPF}}$$

$$= I(t) + n_c(t)$$

⊙ Demodulate quadrature signal  $Q(t)$   
(multiply  $r(t)$  with  $-\sqrt{2} \sin(\omega_0 t)$ )

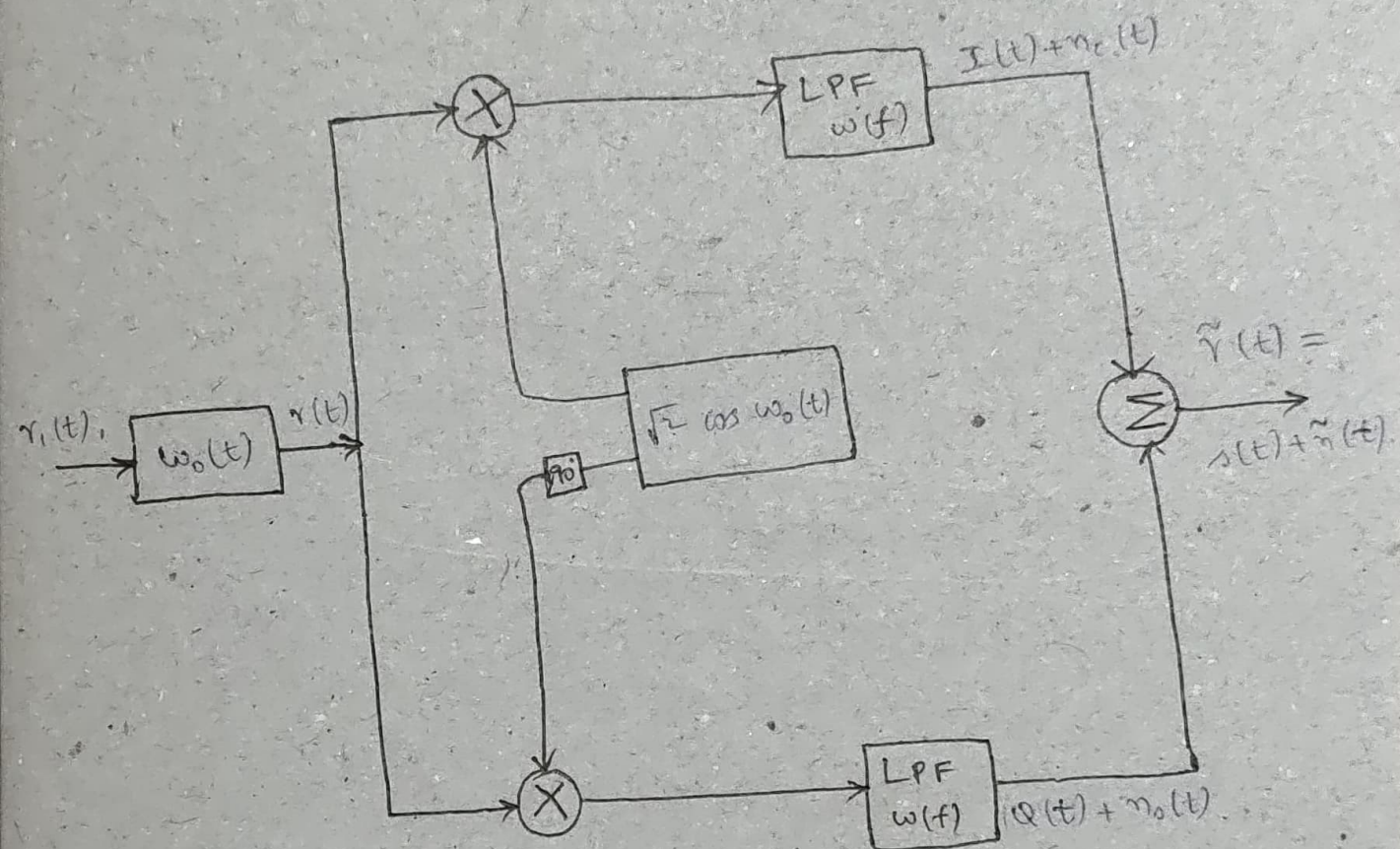
$$r_s(t) = -[r(t) \sqrt{2} \sin(\omega_0 t)]_{\text{LPF}}$$

$$= - \left[ \frac{I(t) \cos(2\omega_0 t) - Q(t) \sin^2(\omega_0 t) + \sqrt{2} n(t) \sin(\omega_0 t)}{1} \right]_{\text{LPF}}$$

$$= -Q(t) + n_s(t)$$

## Demodulator block diagram:

- Demodulated complex baseband receive signal
- $$\tilde{r}(t) = r_c(t) + j r_o(t) = I(t) + n_c(t) + j [Q(t) + n_o(t)]$$
- $$= I(t) + jQ(t) + n_c(t) + j n_o(t)$$
- $$= s(t) + \tilde{n}(t)$$



- Homodyne receiver architecture.