

TDOA Based Direct Positioning: MLE and CRLB

Extension of Vankayalapati et al.

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Introduction: Passive Localization

- ▶ **Objective:** Locate a stationary emitter using spatially separated passive sensors.
- ▶ **Conventional Approach (Two-Step):**
 1. Estimate Time Differences of Arrival (TDOAs) between sensor pairs.
 2. Intersect hyperbolas defined by TDOAs to find the location.
- ▶ **The Problem:** The two-step method is suboptimal. It processes pairs independently and ignores the constraint that all measurements originate from a single point.
- ▶ **Proposed Solution:** Direct Position Determination (DPD). Estimate location directly from raw signal data in a single step.

Problem Setup: Received Signal Model

We observe the signal at M stationary sensors over interval $(0, T)$.

The Received Signal $r_i(t)$

$$r_i(t) = A_i s(t - \tau_i) + w_i(t), \quad i = 0, 1, \dots, M - 1 \quad (1)$$

- ▶ $w_i(t)$: Zero-mean AWGN. Assumed independent and identical spectral density $(\frac{N_0}{2})$.
- ▶ A_i : Unknown attenuation factor.
- ▶ τ_i : Unknown Time of Arrival (TOA).

Goal: Estimate the emitter location $\mathbf{p} = [x_T, y_T]^T$ (2D) or $[x_T, y_T, z_T]^T$ (3D).

Relationship Between Parameters

The unknown TOA (τ_i) is linked to the unknown emitter position (**p**) and unknown transmission time (t_0).

Time of Arrival τ_i

$$\tau_i = \frac{\sqrt{(x_T - x_i)^2 + (y_T - y_i)^2}}{c} + t_0 \quad (2)$$

- ▶ c : Signal propagation speed.
- ▶ The problem involves many unknown/nuisance parameters: **p**, t_0 , A_i , and the signal waveform $s(t)$.

The Signal Approximation

To solve the Maximum Likelihood problem, the continuous signal waveform $s(t)$ must be represented by a finite set of parameters. As we are assuming signal lies inside observation interval $(0, T)$, we can assume signal is periodic with period T and can be written as a Fourier Series.

Fourier Series Approximation

$s(t)$ is approximated by $2N - 1$ Fourier coefficients

$$\phi = [a_0 \ a_1 \dots \ a_{N-1} \ b_1 \dots \ b_{N-1}]^T:$$

$$s(t) = \mathbf{h}^T(t)\phi \quad (3)$$

- ▶ $\mathbf{h}(t)$ is the vector of trigonometric basis functions (dimension $(2N - 1) \times 1$).
- ▶ Rewriting the model in (1) as:

$$r_i(t) = A_i \mathbf{h}^T(t - \tau_i) \phi + w_i(t), \quad 0 \leq t \leq T, \quad i = 0, 1, \dots, M-1. \quad (4)$$

Case 1: Signal Unknown + Transmission Time Unknown → Singular FIM

When both the signal waveform $s(t)$ and the transmission time t_0 are unknown, the Fisher Information Matrix (FIM) \mathcal{I}_θ becomes **singular** (at least rank two deficient).

Ambiguity in Parameters

It is impossible to uniquely determine all unknowns due to inherent relationships.

- ▶ **Amplitude Ambiguity:** Observations $r_i(t)$ generated by $(\bar{A}_i, \bar{s}(t))$ are identical to those generated by $(\bar{A}_i/\gamma, \gamma\bar{s}(t))$ for any non-zero constant γ .
- ▶ **Time Ambiguity:** The transmitted signal $s_T(t)$ generated by $(\bar{t}_0, \bar{s}(t))$ is indistinguishable from one generated by $(\bar{t}_0 - \gamma, \bar{s}(t - \gamma))$ for any constant γ .

Transformation - Eliminating Ambiguity

To remove the singularity of the FIM, we transform the parameters into a set of identifiable quantities. We use sensor $i = 0$ as the reference.

Transformed Parameters θ'

$$\tau' = \begin{bmatrix} (\tau_1 - \tau_0) & (\tau_2 - \tau_0) & \cdots & (\tau_{M-1} - \tau_0) \end{bmatrix}^T$$
$$\mathbf{A}' = (1/A_0) \begin{bmatrix} A_1 & \cdots & A_{M-1} \end{bmatrix}^T$$

- ▶ The unknown transmission time t_0 is **eliminated** and is no longer a nuisance parameter.
- ▶ Let $\theta' = [\tau'^T \mathbf{A}'^T \phi'^T]^T$ be $(2M + 2N - 3) \times 1$ be the unknown transformed parameter vector.
- ▶ The vector of interest is $\alpha' = [\eta'^T \mathbf{A}'^T \phi'^T]^T$, where $\eta' = [x_T, y_T]^T$ is the location vector.

CRLB Derivation Procedure: Signal Unknown

1. Joint Log-Likelihood (LL) Function:

$$I(\boldsymbol{\theta}') = -\frac{1}{N_0} \int_0^T (r_0(t) - \mathbf{h}^T(t)\boldsymbol{\phi}')^2 dt \\ - \frac{1}{N_0} \int_0^T \sum_{i=1}^{M-1} (r_i(t) - A'_i \mathbf{h}^T(t - \tau'_i)\boldsymbol{\phi}')^2 dt \quad (5)$$

2. Derive FIM $\mathcal{I}_{\boldsymbol{\theta}'}$: The FIM elements are computed as:

$$[\mathcal{I}_{\boldsymbol{\theta}'}]_{ij} = -E \left[\frac{\partial^2 I(\boldsymbol{\theta}')}{\partial \theta'_i \partial \theta'_j} \right]$$

3. Resulting FIM $\mathcal{I}_{\boldsymbol{\theta}'}$ is transformed to FIM $\mathcal{I}_{\boldsymbol{\alpha}'}$: The FIM is calculated using the Jacobian matrix:

$$\mathcal{I}_{\boldsymbol{\alpha}'} = \left(\frac{\partial \boldsymbol{\theta}'}{\partial \boldsymbol{\alpha}'^T} \right)^T \mathcal{I}_{\boldsymbol{\theta}'} \left(\frac{\partial \boldsymbol{\theta}'}{\partial \boldsymbol{\alpha}'^T} \right) \quad (6)$$

Maximum Likelihood Estimator (MLE): Signal Unknown

- ▶ **Correlation Vector \mathbf{y}'_i :** The estimation procedure defines a correlation vector \mathbf{y}'_i for each sensor i , correlating the received signal with the basis vector $\mathbf{h}(t)$ aligned by the hypothesized TDOA τ'_i .

$$\mathbf{y}'_i = \int_0^T r_i(t) \mathbf{h}(t - \tau'_i) dt \quad (7)$$

- ▶ **Cross-Correlation Matrix \mathbf{B}' :** The matrix \mathbf{Y}' concatenates these vectors, $\mathbf{Y}' = [\mathbf{y}'_0 \ \mathbf{y}'_1 \ \cdots \ \mathbf{y}'_{M-1}]$. \mathbf{B}' is calculated as $\mathbf{B}' = \mathbf{Y}' \mathbf{Y}'^T$.
- ▶ **The MLE Solution:** The maximum likelihood estimate of the location (\hat{x}_T, \hat{y}_T) is found by maximizing the maximum eigenvalue of \mathbf{B}' over all possible locations.

$$(\hat{x}_T, \hat{y}_T) = \arg \max_{(x_T, y_T)} \lambda_{max}(\mathbf{B}') \quad (8)$$

Case 2: Signal Known + Transmission Time Unknown

- ▶ **Unknowns Reduced:** The number of unknowns is reduced to $2M$. The unknown parameter vector is $\zeta = [\tau^T \mathbf{A}^T]^T$. The signal parameters ϕ are known.
- ▶ **FIM Not Singular:** Since the signal waveform is known, the unknown parameters can be uniquely determined. Therefore, the FIM \mathcal{I}_ζ is **not singular** \implies No transformation
- ▶ **MLE Simplifies:** The location and transmission time are estimated by maximizing the sum of squared correlation values across all sensors.

$$(\hat{x}_T, \hat{y}_T, \hat{t}_0) = \arg \max_{(x_T, y_T, t_0)} \sum_{i=0}^{M-1} \left(\int_0^T r_i(t) s(t - \tau_i) dt \right)^2 \quad (9)$$

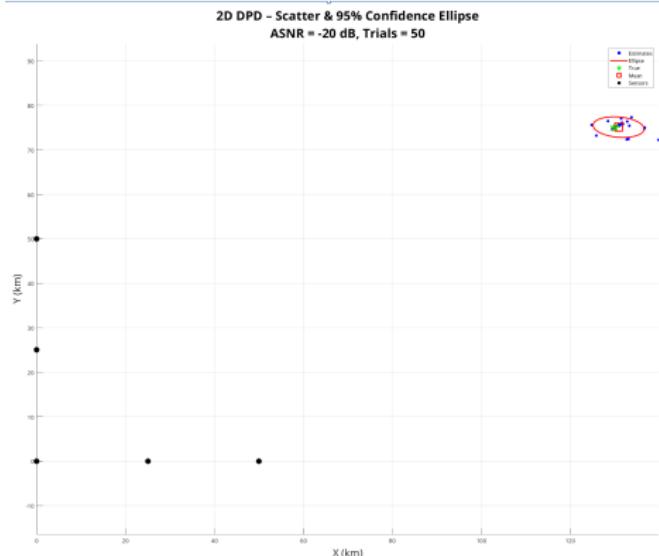
2D Simulation Setup

Configuration:

- ▶ 5 Sensors in a 2D L-shape.
- ▶ Emitter Location: (130, 75) km.
- ▶ Signal: Gaussian Chirp.
- ▶ ASNR Range: -35 dB to -10 dB.

Results:

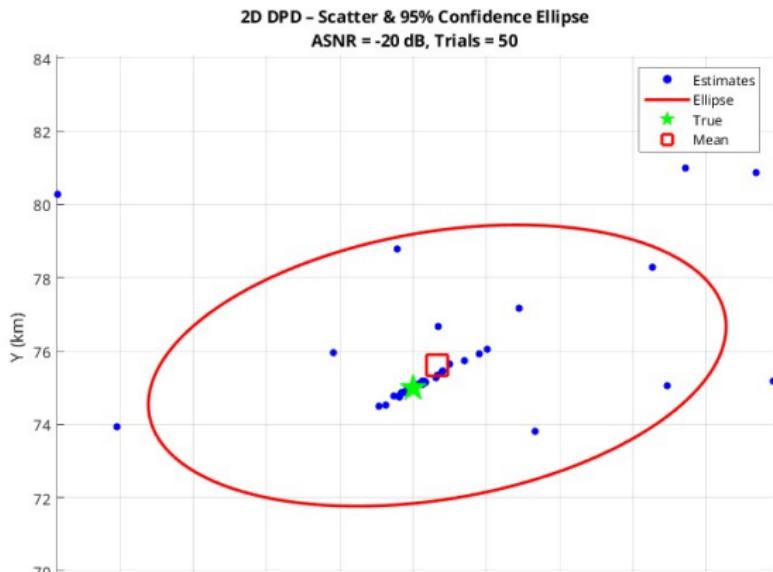
1. True Pos (X, Y): [130.00, 75.00] km
2. Mean Est (X, Y): [132.99, 76.11] km
3. Covariance Diagonal (Var X, Y): 0.3194380, 0.1273112



MLE Scatter and 95% confidence ellipse

2D Results: Scatter Plot

- ▶ Estimates are highly concentrated near the true location even at low ASNR (e.g., -20 dB).
- ▶ The 95% error ellipse confirms that the estimates are tightly bound, confirming the low variance.
- ▶ The estimator operates reliably far below the SNR threshold where the conventional TDOA method breaks down (≈ -17 dB).



2D Results: Performance Comparison

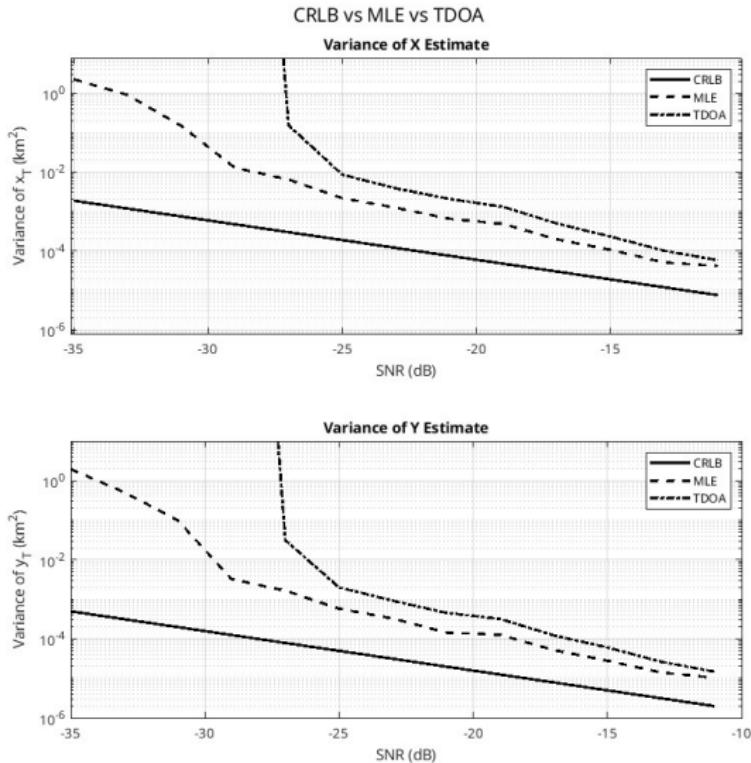


Figure 1: Variance Comparison: MLE vs. TDOA vs. CRLB

Extension to 3D: Modifying the Jacobian

Extending the DPD model to 3D requires modifying the geometry-dependent terms.

The Jacobian Matrix $\frac{\partial \tau'}{\partial \eta'^T}$ (Equation 40)

The Jacobian relates changes in the location vector $\eta' = [x_T, y_T, z_T]^T$ to changes in the TDOA vector τ' .

- ▶ **Change:** The Jacobian matrix $(\frac{\partial \tau'}{\partial \eta'^T})$ must be expanded from $(M - 1) \times 2$ to $(M - 1) \times 3$.
- ▶ **New Term:** Each row (TDOA) gains a column representing the derivative with respect to z :

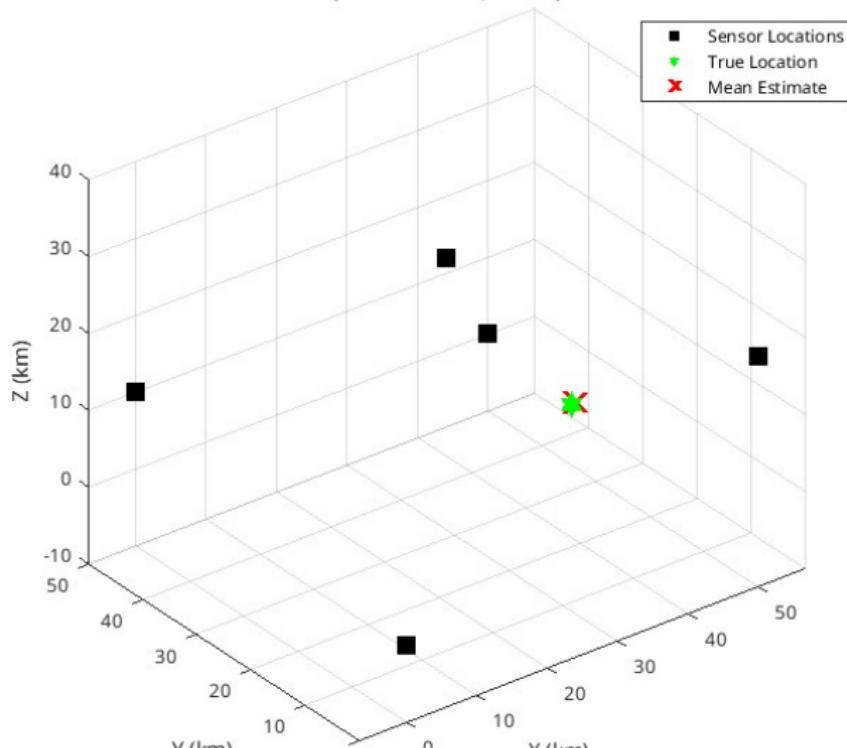
$$\frac{\partial \tau'_i}{\partial z} = \frac{1}{c} \left(\frac{z_T - z_i}{d_i} - \frac{z_T - z_0}{d_0} \right)$$

- ▶ This modified Jacobian ensures the FIM and subsequent CRLB calculations are valid for the 3D position vector.

3D Simulation Setup and Results

- **Setup:** 5 non-coplanar sensors are used to guarantee a non-singular FIM for the 3D problem.

Physical placement of sensors (for $M = 5$) and emitter position used for simulation.
(ASNR = -20 dB, N = 50)

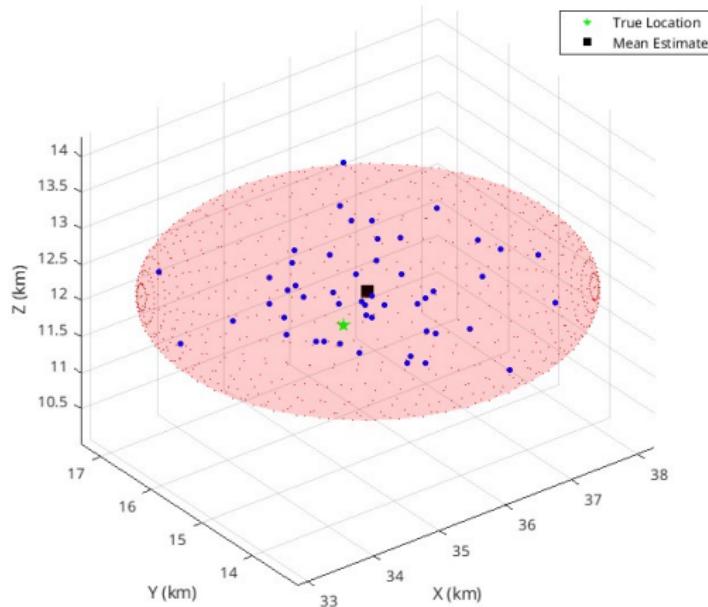


3D Simulation - 95% error ellipse for MLE

Results:

- ▶ True Pos: [35.00, 15.00, 12.00]
- ▶ Mean Est: [35.55, 15.24, 12.15]
- ▶ Covariance Diagonal (Var X, Y, Z): 0.9369315, 0.5124023, 0.3347965

3D Extension of Fig. 7: MLE Scatter & 95% Conf. Ellipsoid
(ASNR = -20 dB, N = 50)



Conclusion

- ▶ **Methodology:** DPD-MLE is the optimal single-step localization technique, correctly handling nuisance parameters.
- ▶ **Key Insight:** Ambiguities in the unknown signal case necessitate parameter transformation (using relative measurements) to ensure a non-singular FIM and valid CRLB.
- ▶ **Extension:** The DPD framework extends seamlessly to 3D by expanding the Jacobian to include derivatives w.r.t. the z-coordinate.