

1

$$G(s) = \frac{6}{(s+3)^2}, \quad H(s) = K_p + \frac{K_i}{s}$$

$$= K_p \left(\frac{s + \frac{K_i}{K_p}}{s} \right)$$

a) i: $(-\infty, -3)$ $L(s) = K \left(\frac{s+z}{(s+3)^2 s} \right), K=6K_p, z = \frac{K_i}{K_p}$
 ii: $(-3, 0)$ $D(s) + KN(s) = 0$
 iii: $(0, +\infty)$ $s^3 + 6s^2 + 9s + Ks + Kz = 0$

$$n=3, m=1 \Rightarrow \sigma_a = \frac{-3-3+0 - (-\frac{K_i}{K_p})}{2} = -3 + \frac{K_i}{2K_p}$$

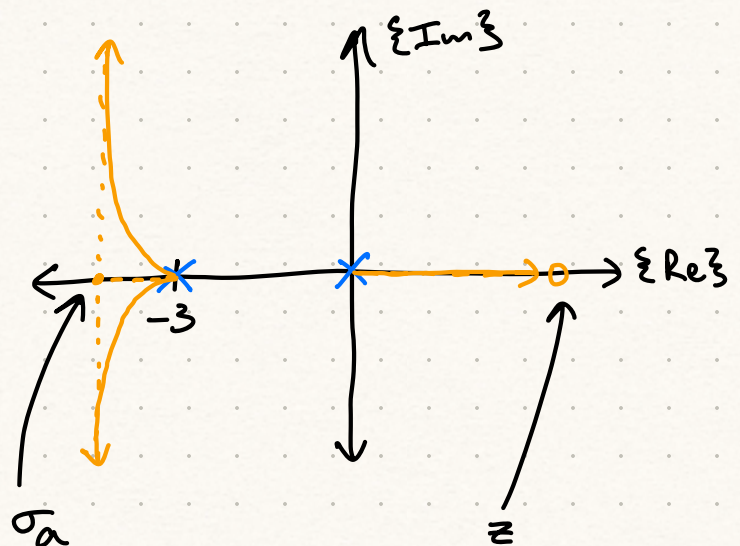
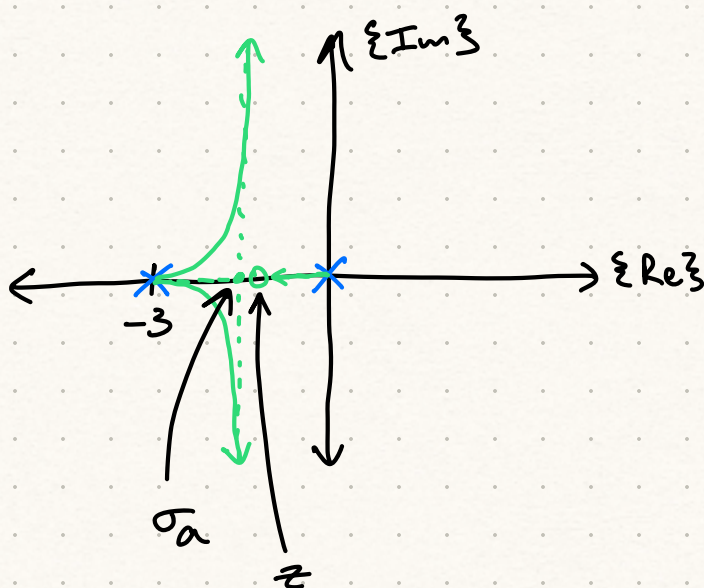
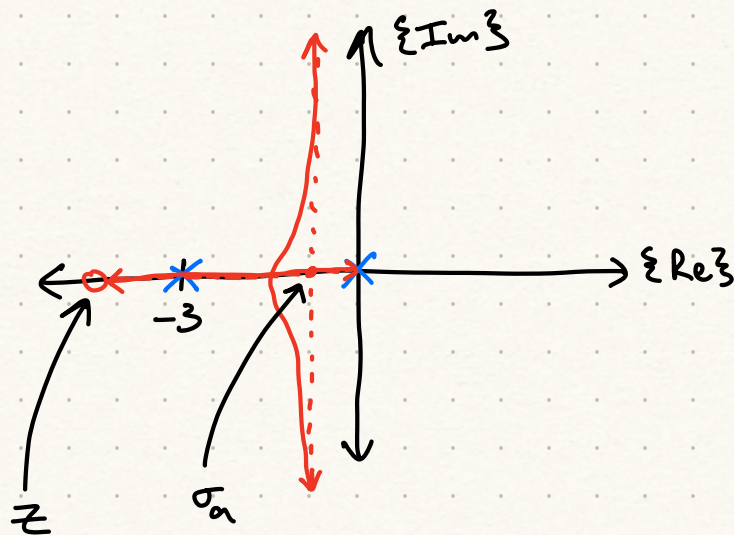
$$\alpha_L = \pm 90^\circ$$

OL poles

case i

case ii

case iii



b) case iii is unstable, as it mandates a singular RHP zero and its LHP branch is unstable

c) Want: $\alpha < 0$

$$\alpha = \frac{-6+z}{2} \Rightarrow z < 6$$

$z > 0$ to avoid case iii

$$\hookrightarrow 0 < z < 6$$

\hookrightarrow zero must lie between $(-6, 0)$

Need: $\nexists L(s) \neq -180^\circ$

$$\text{Let: } \nexists L(s) = 180^\circ, \quad \alpha \tan\left(\frac{\omega}{z}\right) - \frac{\pi}{2} - 2\alpha \tan\left(\frac{\omega}{3}\right) = -\pi$$

$$-\frac{\pi}{2} = \alpha \tan\left(\frac{\omega}{z}\right) - 2\alpha \tan\left(\frac{\omega}{3}\right)$$

$$\Rightarrow \frac{\omega}{z} = \frac{9-\omega^2}{-6\omega} \Rightarrow \omega^2 = -\frac{9z}{6-z} > 0$$

$\underbrace{\hspace{10em}}$

want this to be false

$$\hookrightarrow 0 < z < 6$$

\hookrightarrow zero must lie between $(-6, 0)$

d) $\zeta = \frac{1}{\sqrt{2}}$, cancel a pole: $z = 3$

$$\omega_n = \frac{z}{2\zeta}, \quad K = \omega_n^2 / G$$

$$K_p = K$$

\rightarrow

$$K_p = 0.75$$

$$K_i = K_p z$$

\rightarrow

$$K_i = 2.25$$

$$\underline{2} \quad G(s) = \frac{5(s-1)}{s-6}$$

a) $H(s) = \frac{K}{s-p}$, with $p > 0$

$$L(s) = \frac{5K(s-1)}{(s-6)(s-p)} \Rightarrow 1+L(s)=0$$

$$(s-p)(s-6) + 5K(s-1) = 0 \Rightarrow s^2 + (-p-6)s + 6p + 5Ks - 5K = 0$$

$$\hookrightarrow 5K - (p+6) > 0 \ \& \ 6p - 5K > 0$$

$$\frac{p+6}{5} < K < \frac{6}{5}p \Rightarrow \boxed{p > \frac{6}{5}}$$

b) Want: 2 real poles @ -2

$$CL: s^2 + (5K - (p+6))s + (6p - 5K)$$

$$T_{des}: (s+2)^2 = s^2 + 4s + 4$$

$$\begin{bmatrix} 1, & 5K - (p+6), & 6p - 5K \\ 1, & 4, & 4 \end{bmatrix} \begin{matrix} 0 \\ 0 \end{matrix}$$

$$p = \frac{5K+4}{6}$$

$$5K - \left(\frac{5K+4}{6} + 6 \right) = 4$$

$$5K = 10 + \frac{5K+4}{6}$$

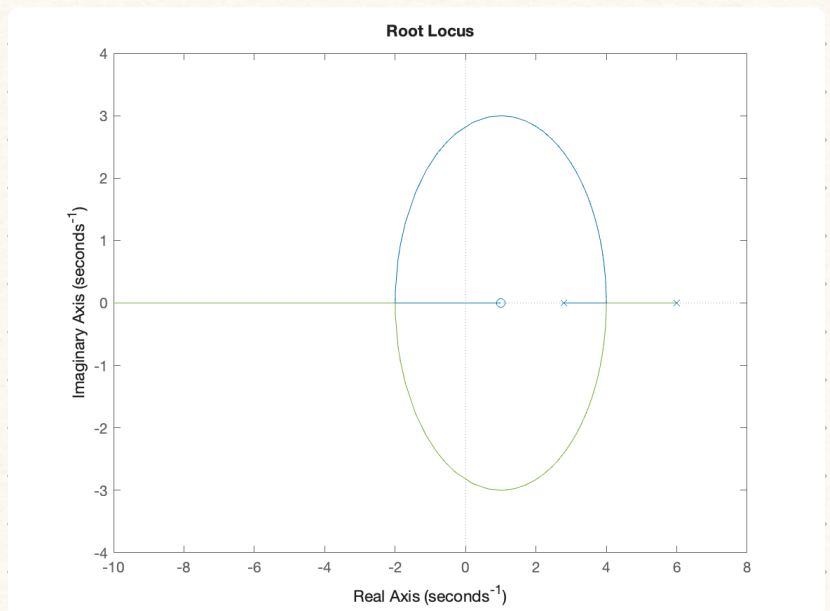
$$30K = 60 + 5K + 4$$

$$25K = 64$$

$$K = \left(\frac{8}{5} \right)^2 = \frac{64}{25}$$

$$p = \frac{\frac{64}{5} + 4}{6} = \frac{84}{30} = \frac{14}{5}$$

$$\boxed{H(s) = \frac{64}{25} \left(\frac{1}{s - \frac{14}{5}} \right)}$$

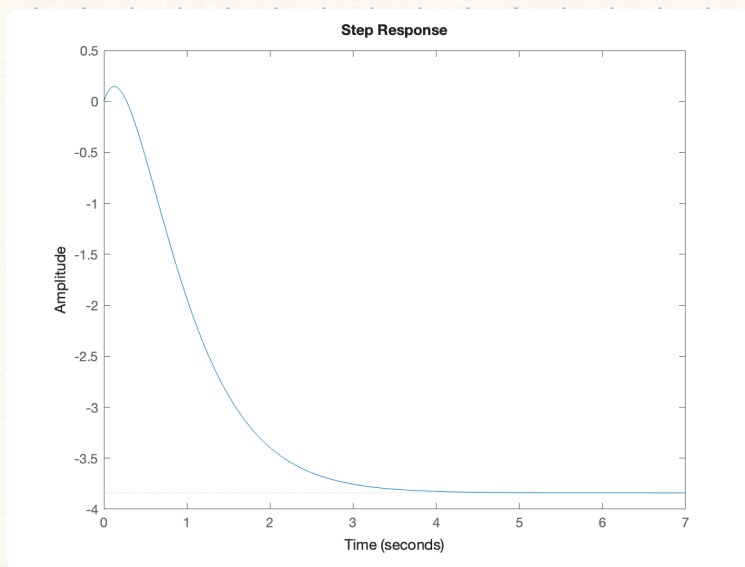


c) $y_d(t) = 1(t)$, $R(s) = \frac{H(s)}{1+L(s)}$, $U(s) = R(s)Y_d(s)$

$$Y_d(s) = \frac{1}{s} \Rightarrow U(s) = \frac{R(s)}{s} = \frac{H(s)}{s(1+L(s))} = \frac{2.56(s-6)}{s(s+2)^2}$$

$$u(t) = \frac{96}{25} \left(e^{-2t} + \frac{8}{3} t e^{-2t} - 1 \right)$$

$u(t)$ is bounded, as all poles of $R(s)$ are in LHP



```
R_info = struct with fields:
    RiseTime: 1.6515
    TransientTime: 3.0421
    SettlingTime: 3.0640
    SettlingMin: -3.8400
    SettlingMax: -3.4571
    Overshoot: 0
    Undershoot: 3.8196
    Peak: 3.8400
    PeakTime: 7.8058
```

d)

```
H_zoh = struct with fields:
```

```
Ad: 1.1185
Bd: 0.0847
Cd: 1.2800
Dd: 0
```

```
H_tustin = struct with fields:
```

```
Ad: 1.1186
Bd: 0.3639
Cd: 0.3158
Dd: 0.0542
```


$$\underline{4} \quad G(s) = \frac{2}{s^2(s^2+3)}$$

$$a) \quad H(s) = K ?$$

$$L(s) = \frac{2K}{s^2(s^2+3)}, \quad L'(s) = [(s-p)L(s)]$$

$$n-m=4 \Rightarrow \alpha_\ell = \pm 45^\circ, \pm 135^\circ \text{ \& } \sigma_a = 0$$

$$\angle L(s) = (1+2\ell)180^\circ = \angle L'(s) - \angle(s-p)$$

$$\delta = \angle L'(p) - (1+2\ell)180^\circ$$

$$\angle L'(s) \Big|_{\epsilon\text{-circle}} = \angle L'(p)$$

for pole at $s = +\sqrt{3}j$:

$$L'(s) = \frac{2K}{s^2(s+\sqrt{3}j)} \Rightarrow \angle L'(p) = -180^\circ - 90^\circ$$

$$\delta = -90^\circ$$

by symmetry, for $s = -\sqrt{3}j$: $\delta = 90^\circ$

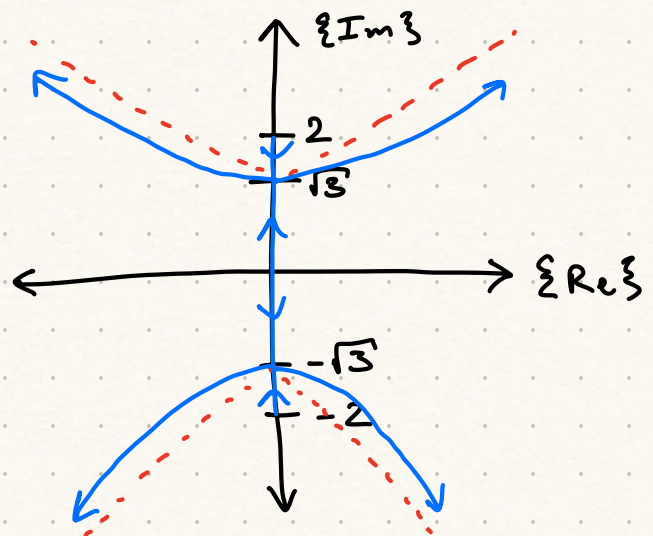
for pole at $s = 0$ (repeated):

$$L'(s) = \frac{2K}{s^2+3} \Rightarrow \angle L'(p) = -180^\circ$$

$$\delta = \pm 180^\circ$$

$$\hookrightarrow \delta = \pm 90^\circ$$

Will be unstable
at all poles for
sufficiently large K



b) Want: CL @ $-1 \pm j$ & all else CL @ -4

$$H(s) = \frac{b_3 s^3 + b_2 s^2 + b_1 s + b_0}{s^3 + a_2 s^2 + a_1 s + a_0}$$

find: $\text{zpk}(H(s))$
& $\text{zpk}(T(s))$

(now Polys

$$CL: (s^3 + a_2 s^2 + a_1 s + a_0)(s^2(s^2 + 3)) + (b_3 s^3 + b_2 s^2 + b_1 s + b_0)(2)$$

$$T_{des}: (s - (-1 + j))(s - (-1 - j))(s + 4)^5$$

$$\begin{aligned} \rightarrow a_2 &= 22 & b_3 &= 1141.5 \\ a_1 &= 199 & b_2 &= 1031 \\ a_0 &= 934 & b_1 &= 2304 \\ & & b_0 &= 1024 \end{aligned}$$

```
H =
1141.5 (s+0.4939) (s^2 + 0.4093s + 1.816)
-----
(s+11.86) (s^2 + 10.14s + 78.74)

Continuous-time zero/pole/gain model.
Model Properties

T =
2283 (s+0.4939) (s^2 + 0.4093s + 1.816)
-----
(s+4)^5 (s^2 + 2s + 2)

Continuous-time zero/pole/gain model.
Model Properties

poles_T = 7x1 complex
-4.0075 + 0.0000i
-4.0023 + 0.0071i
-4.0023 - 0.0071i
-3.9939 + 0.0044i
-3.9939 - 0.0044i
-1.0000 + 1.0000i
-1.0000 - 1.0000i

zeros_T = 3x1 complex
-0.2046 + 1.3321i
-0.2046 - 1.3321i
-0.4939 + 0.0000i
```

$$c) H(s) = \frac{b_3 s^3 + b_2 s^2 + b_1 s + b_0}{s^4 + a_3 s^3 + a_2 s^2 + a_1 s + a_0}$$

(now Polys

$$CL: (s^4 + a_3 s^3 + a_2 s^2 + a_1 s + a_0)(s^2(s^2 + 3)) + (b_3 s^3 + b_2 s^2 + b_1 s + b_0)(2)$$

$$T_{des}: (s - (-1 + j))(s - (-1 - j))(s + 4)^6$$

```
H =
5597 (s+0.4116) (s^2 + 0.125s + 1.778)
-----
(s^2 + 20.06s + 115.9) (s^2 + 5.939s + 51.91)

Continuous-time zero/pole/gain model.
Model Properties

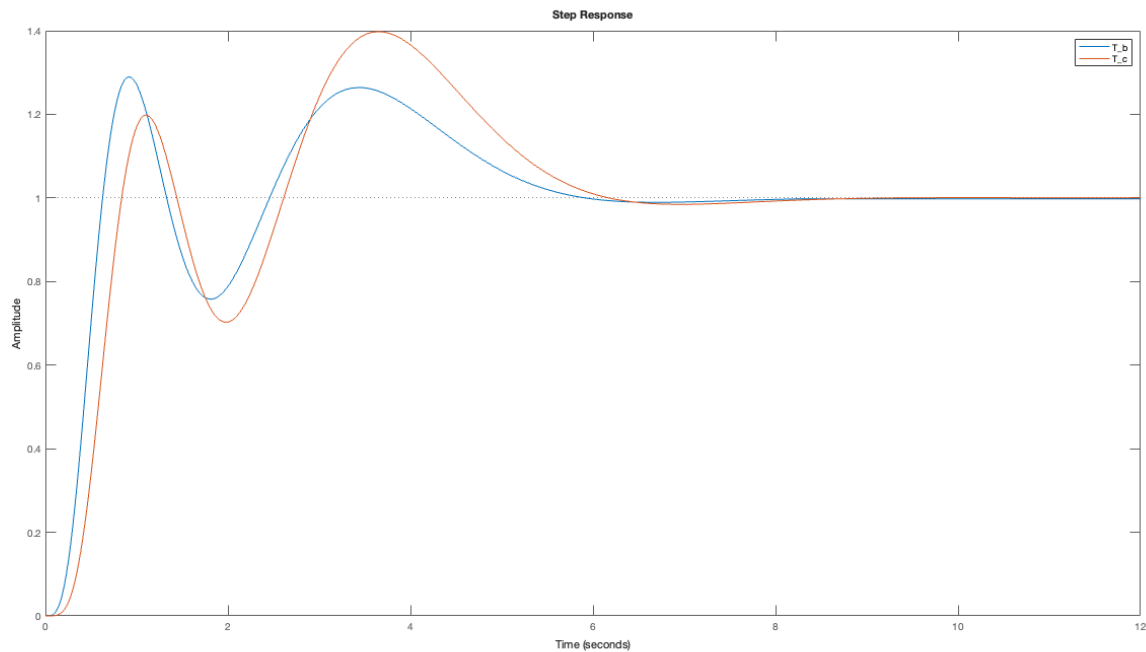
T =
11194 (s+0.4116) (s^2 + 0.125s + 1.778)
-----
(s+4)^6 (s^2 + 2s + 2)

Continuous-time zero/pole/gain model.
Model Properties

poles_T = 8x1 complex
-4.0247 + 0.0000i
-4.0123 + 0.0214i
-4.0123 - 0.0214i
-3.9877 + 0.0213i
-3.9877 - 0.0213i
-3.9753 + 0.0000i
-1.0000 + 1.0000i
-1.0000 - 1.0000i

zeros_T = 3x1 complex
-0.0625 + 1.3320i
-0.0625 - 1.3320i
-0.4116 + 0.0000i
```


d) Neither controllers are good, as they both have about (or over) 5.5s settling times, and significant overshoot. This is due to the introduction of extra zeroes and their compensatory poles "spilling" our intended behavior of the designed poles



e) Otherwise, the benefit of C is that it has an extra pole, ensuring it has noise stability, which B does not