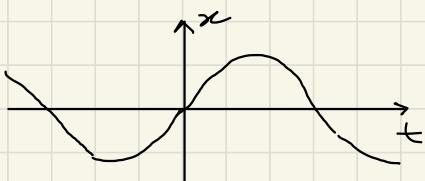
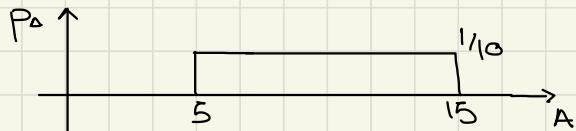
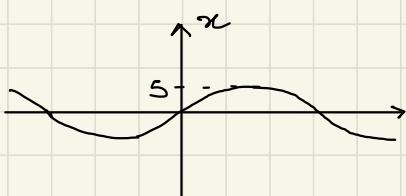


# PROCESS!

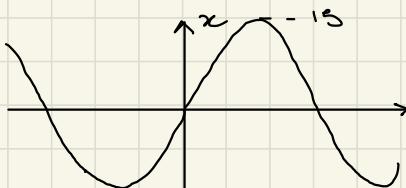


## Esercizi:

- $x(t) = A \sin(2\pi f_0 t)$   $A$  v.c.  $P(A) = \frac{1}{10} \text{rect}\left(\frac{A}{10} - 1\right)$



3 diverse realizzazioni di  $x$



Dove ricavare  $P_{x(t)}(x)$

tempo  
t fisso

$$x(t) = A \underbrace{\sin(2\pi f_0 t)}_{\text{tempo t fisso}}$$

$$x = A S \rightarrow A = \frac{x}{S}, \quad \frac{dx}{dA} = S$$

$$P_x(x) = \frac{P_A(x/S)}{\left|\frac{dx}{dA}\right|} = \frac{1}{S} \cdot \frac{1}{10} \text{rect}\left(\frac{x}{10S} - 1\right)$$

$$= \frac{1}{10 \sin(2\pi f_0 t)} \text{rect}\left(\frac{x - 10 \sin(2\pi f_0 t)}{10 \sin(2\pi f_0 t)}\right)$$

STAZIONARITÀ  
IN SENSO LATO

$$E[x(t)] = \mu_x$$

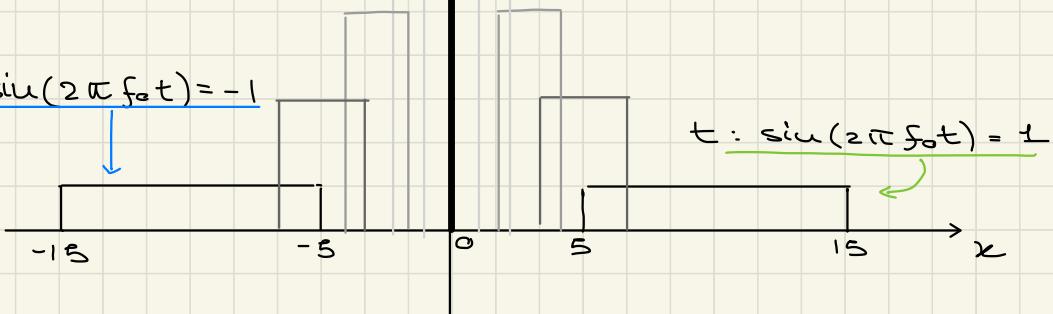
$$E[x(t+\tau)x^*(t)] = R_x(\tau)$$

$$E[|x(t)|^2] = P_x$$

per  $t$ :  $\sin(2\pi f_0 t) = 0 \Rightarrow x = 0$

non è più casuale!  
 $P_x = \delta(x)$

$t$ :  $\sin(2\pi f_0 t) = -1$



$t$ :  $\sin(2\pi f_0 t) = 1$

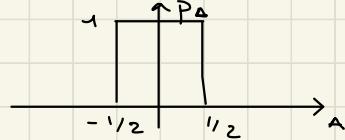
$$\begin{aligned} E[x(t)] &= E[A \sin(2\pi f_0 t)] = E[A] \cdot \sin(2\pi f_0 t) \\ &= 40 \cdot \sin(2\pi f_0 t) \end{aligned}$$

$$\begin{aligned} \text{var}[x(t)] &= \text{var}[A \sin(2\pi f_0 t)] = \text{var}[A] \sin^2(2\pi f_0 t) \\ &= \frac{10^2}{12} \sin^2(2\pi f_0 t) \end{aligned}$$

- $x(t) = A + \cos(2\pi f_0 t)$  A. u.c.  $P_A(A) = \text{rect}(A)$

$P_x$ ?

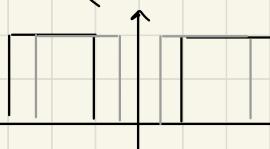
$$x = A + c \rightarrow A = x - c$$



$$\text{e } \frac{dx}{dA} = 1$$

$$P_x(x) = \frac{P_A(x-c)}{\left| \frac{dx}{dA} \right|} =$$

$$= \text{rect}(x - \cos(2\pi f_0 t))$$



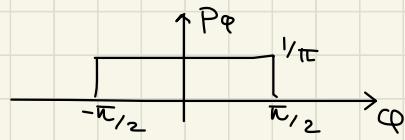
$$E[x(t)] = E[A] + \cos(2\pi f_0 t) = \cos(2\pi f_0 t)$$

$$\text{Var}[x(t)] = \text{Var}[A] = \frac{1}{12}$$

$\mu_x = \mu_x(t)$  quindi  
x NON è  
STAZIONARIO

$$\bullet x_n = \cos(2\pi f_0 n + \varphi) \quad \varphi \text{ u.c.} \quad P_\varphi(\varphi) = \frac{1}{\pi} \operatorname{rect}\left(\frac{\varphi}{\pi}\right)$$

$$\begin{aligned} E[x_n] &= \int_{-\infty}^{+\infty} x_n(\varphi) P_\varphi(\varphi) d\varphi \\ &= \int_{-\pi/2}^{\pi/2} \cos(2\pi f_0 n + \varphi) \frac{1}{\pi} d\varphi \\ &= \frac{1}{\pi} \left[ \sin(2\pi f_0 n + \frac{\pi}{2}) - \sin(2\pi f_0 n - \frac{\pi}{2}) \right] \\ &= \frac{1}{\pi} [ \cos(2\pi f_0 n) - (-\cos(2\pi f_0 n))] \\ &= \frac{2}{\pi} \cos(2\pi f_0 n) \rightarrow \text{NON È STAZIONARIO} \end{aligned}$$



$$\text{Var}[x_n] = E[x_n^2] - E^2[x] =$$

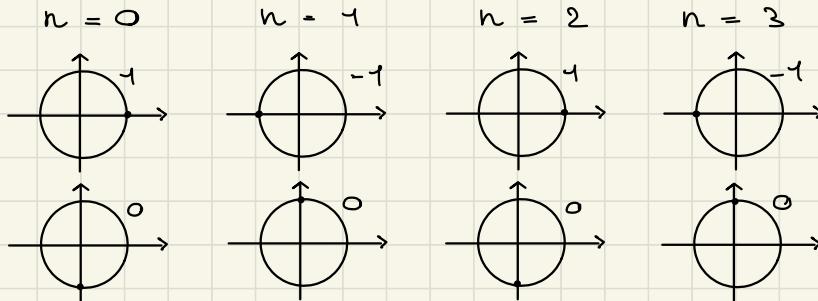
$$\begin{aligned} E[x_n^2] &= \int_{-\pi/2}^{\pi/2} \cos^2(2\pi f_0 n + \varphi) \frac{d\varphi}{\pi} = \\ &= \frac{1}{\pi} \int_{-\pi/2}^{\pi/2} \left( \frac{\cos(4\pi f_0 n + 2\varphi) + 1}{2} \right) d\varphi \\ &= \frac{1}{2\pi} \cdot \left[ \frac{\pi}{2} - \left( -\frac{\pi}{2} \right) \right] = \frac{1}{2} \end{aligned}$$

integrale la  
simeside sul  
suo periodo  
fa 0

$$\rightarrow \text{Var}[x_n] = \frac{1}{2} - \frac{4}{\pi^2} \cos^2(2\pi f_0 n)$$

$$\bullet \quad x_n = \cos(\pi n + \varphi) \quad P_\varphi(\varphi) = \frac{1}{2} \delta(\varphi) + \frac{1}{2} \delta(\varphi + \frac{\pi}{2})$$

$$x_n = \begin{cases} \cos(\pi n) & \text{P} = 1/2 \\ \cos(\pi n - \frac{\pi}{2}) & \text{P} = 1/2 \end{cases}$$



$$\Rightarrow P_x(x_n) = \begin{cases} \frac{1}{2} \delta(x) + \frac{1}{2} \delta(x-1) & n \text{ pair} \\ \frac{1}{2} \delta(x) + \frac{1}{2} \delta(x+1) & n \text{ dispari} \end{cases}$$

$$\begin{aligned} E[x_n] &= \frac{1}{2} \cdot 0 + \frac{1}{2} \cdot 1 = \frac{1}{2} \quad n \text{ pair} \\ &= \frac{1}{2} \cdot 0 + \frac{1}{2} \cdot (-1) = -\frac{1}{2} \quad n \text{ dispari} \end{aligned}$$

$$\begin{aligned} E[x_n^2] &= \frac{1}{2} \cdot 0^2 + \frac{1}{2} \cdot 1^2 = \frac{1}{2} \quad n \text{ pair} \\ &= \frac{1}{2} \cdot 0^2 + \frac{1}{2} \cdot (-1)^2 = \frac{1}{2} \quad n \text{ dispari} \end{aligned}$$

$$\text{Var}[x_n] = \frac{1}{2} - \left(\pm \frac{1}{2}\right)^2 = \frac{1}{2} - \frac{1}{4} = \frac{1}{4}$$

$$\bullet \quad x(t) = 5 \cos(2\pi f_0 t + \varphi)$$

$$P_\varphi(\varphi) = \frac{1}{2\pi} \text{rect}\left(\frac{\varphi - 4\pi}{2\pi}\right) = \frac{1}{2\pi} \text{rect}\left(\frac{\varphi}{2\pi}\right)$$

da  $3\pi$  a  $5\pi$  è come  
dice da  $-\pi$  a  $\pi$

$$E[x(t)] = \int x(t) p_\varphi(\varphi) d\varphi = \int_{-\pi}^{\pi} 5 \cos(2\pi f_0 t + \varphi) \frac{d\varphi}{2\pi}$$

↑  
Integro sul periodo

$$= 0 \quad \checkmark$$

$$\begin{aligned} \text{Var}[x(t)] &= E[x^2(t)] = \frac{25}{2\pi} \int_{-\pi}^{\pi} \cos^2(2\pi f_0 t + \varphi) d\varphi \\ &= \frac{25}{4\pi} \int_{-\pi}^{\pi} (1 + \cos(4\pi f_0 t + 2\varphi)) d\varphi \\ &= \frac{25}{4\pi} \cdot 2\pi = \frac{25}{2} \quad \checkmark \end{aligned}$$

↓ Integro 2 volte  
sul periodo

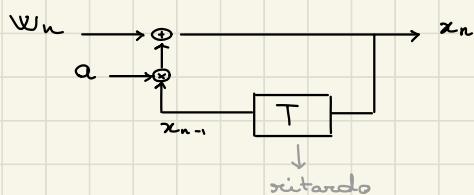
$$\begin{aligned} R_x(t_1, t_2) &= E[\underbrace{x(t_1) \cdot x(t_2)}_{g(\varphi)}] = \int g(\varphi) p_\varphi(\varphi) d\varphi \\ &= \frac{25}{2\pi} \int_{-\pi}^{\pi} \cos(2\pi f_0 t_1 + \varphi) \cos(2\pi f_0 t_2 + \varphi) d\varphi \\ &= \frac{25}{4\pi} \int_{-\pi}^{\pi} [\cos(2\pi f_0(t_1 + t_2) + 2\varphi) + \cos(2\pi f_0(t_1 - t_2))] d\varphi \\ &= \frac{25}{4\pi} \cdot 2\pi \cos(2\pi f_0(t_1 - t_2)) = \frac{25}{2} \cos(2\pi f_0(t_1 - t_2)) \end{aligned}$$

↓

$$R_x = R_x(\tau) \quad \checkmark$$

→ Il processo  $x(t)$  è STAZIONARIO

- Processo autoregressivo



a t.c.  $x_n$  stazionario?

$$\begin{aligned} x_n &= a x_{n-1} + w_n \\ E[w_n] &= 0 \\ E[w_n^2] &= \sigma_w^2 \\ E[w_n w_m] &= 0, \quad n \neq m \\ R_w(n-m) &= \sigma_w^2 \delta_{n-m} \end{aligned}$$

$$E[x_n] = E[\alpha x_{n-1} + w_n] = \alpha E[x_{n-1}]$$

Se  $x_n$  è stazionario, il valore atteso non deve dipendere dal tempo, cioè  $E[x_n] = E[x_{n-1}] = \mu_x$

$$\hookrightarrow \mu_x = \alpha \mu_x \rightarrow \begin{cases} \alpha = 1, \mu_x \neq 0 \\ \text{a qualsiasi, } \mu_x = 0 \end{cases}$$

- $\mu_x = 0$

$$\text{Var}[x_n^2] = \text{Var}[\alpha^2 x_{n-1}^2 + w_n^2 + 2\alpha x_{n-1} w_n]$$

↓ poiché  $x_n$  stazionario

$$\sigma_x^2 = \alpha^2 \sigma_x^2 + \sigma_w^2 + 2\alpha \text{Var}[x_{n-1} w_n]$$

|

$$\sigma_x^2 = \alpha^2 \sigma_x^2 + \sigma_w^2$$

o, poiché  $w$  al campione  $n$  non può influenzare  $x$  al campione  $n-1$ , che è avvenuto all'istante prima

$$\hookrightarrow \alpha^2 = 1 - \frac{\sigma_w^2}{\sigma_x^2} \rightarrow 0 \leq \alpha^2 \leq 1 \quad 0 \leq |\alpha| \leq 1$$

- $\mu_x \neq 0 \rightarrow \alpha = 1 \iff \sigma_w^2 = 0$  poiché  $\alpha^2 = 1 - \frac{\sigma_w^2}{\sigma_x^2}$

$n-m$

$$\begin{aligned} R_x(k) &= E[x_{n+k} x_n] = E[(\alpha x_{n-1+k} + w_{n+k}) x_n] \\ &= \alpha R_x(k-1) + E[w_{n+k} x_n] \end{aligned}$$

0

$$R_x(k) = \alpha R_x(k-1)$$

$$R_x(0) = \sigma_x^2, \quad R_x(1) = \alpha \sigma_x^2 \quad R_x(2) = \alpha^2 \sigma_x^2$$

$$\hookrightarrow R_x(k) = \alpha^{(k)} \sigma_x^2$$

se  $x$  stazionario,  $R_x(k) = R_x(-k)$

## Processi Stazionari

$$R_x(t_1, t_2) = E[x(t_1)x^*(t_2)] = R_x(\underbrace{t_1 - t_2}_{\tau})$$

$$R_x(\tau) = E[x(t+\tau)x^*(t)] = R_x(t+\tau-t) = R_x(\tau)$$

$$R_x(0) = E[|x(t)|^2] = P_x \geq 0$$

$$\begin{aligned} R_x(-\tau) &= E[x(t-\tau)x^*(t)] = (E[x(t)x^*(t-\tau)])^* \\ &= (R_x(t-t+\tau))^* = R_x^*(\tau) \end{aligned}$$

$$|R_x(\tau)| \leq R_x(0)$$

$$E[x(t)] = 0 \Rightarrow R_x(\tau) = \text{cov}(x(t+\tau), x(t))$$

$$\downarrow \\ u, v \in \mathbb{C}$$

$$\text{cov}(u, v) = E[(u - E[u])(v - E[v])^*]$$

$$\text{Cross-Correlazione: } R_{yx}(\tau) = E[y(t+\tau)x^*(t)]$$

$$\underline{\text{Es}}: y(t) = a x(t - \tau_a)$$

$$\begin{aligned} R_{yx}(\tau) &= E[y(t+\tau)x^*(\tau)] = a E[x(t+\tau-\tau_a)x^*(t)] \\ &= a R_x(\tau - \tau_a) \end{aligned}$$

$$\begin{aligned} R_y(\tau) &= E[y(t+\tau)y^*(\tau)] = |a|^2 E[x(t+\tau-\tau_a)x^*(t-\tau_a)] \\ &= |a|^2 R_x(\tau) \end{aligned}$$

$$y(t) = a x(t - \tau_a) + b x(t - \tau_b)$$

$$\begin{aligned} R_{yx}(\tau) &= E[(a x(t+\tau-\tau_a) + b x(t+\tau-\tau_b)) \cdot x^*(t)] = \\ &= a E[x(t+\tau-\tau_a)x^*(t)] + b E[x(t+\tau-\tau_b)x^*(t)] = a R_x(\tau - \tau_a) + b R_x(\tau - \tau_b) \end{aligned}$$

$$R_y(\tau) = E[(\alpha x(t+\tau-\tau_a) + b x(t+\tau-\tau_b)) (\alpha x(t+\tau_a) + b x(t+\tau_b))^*] \\ = |\alpha|^2 R_x(\tau) + |b|^2 R_x(\tau) + \alpha b^* R_x(\tau - (\tau_a - \tau_b)) + \alpha^* b R_x(\tau - (\tau_b - \tau_a))$$

## Stima Autocorrelazione

$$R_x(\tau) = E[x(t+\tau)x^*(t)]$$

$$\hat{R}_x(\tau) = \frac{1}{T_0} \int_{T_0} x(t+\tau)x^*(t) dt$$

è una variabile casuale

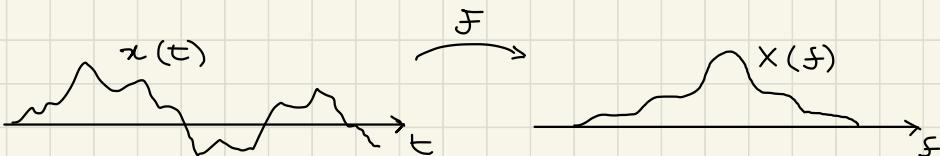
To tempo di osservazione

$$E[\hat{R}_x(\tau)] = \frac{1}{T_0} \int_{T_0} E[x(t+\tau)x^*(t)] dt = \\ = \frac{1}{T_0} \int_{T_0} R_x(\tau) dt = R_x(\tau) \frac{1}{T_0} \int_{T_0} dt = \\ = R_x(\tau) \frac{T_0 - |\tau|}{T_0}$$

fattore di polarizzazione

$$E[\hat{R}_x(\tau)] = \frac{1}{T_0} |X_{T_0}(f)|^2$$

## Caratterizzazione in Frequenza (di proc. staz.)



Valore atteso

$$E[X(f)] = \int E[x(t)] e^{-j2\pi ft} dt = \int \mu_x e^{-j2\pi ft} dt = \mu_x \delta(f)$$

X(f) non è staz!

- Potenza

$$X(f) = \int_{-\infty}^{+\infty} x(t) e^{-j2\pi f t} dt \approx \sum_{n=-\infty}^{\infty} x_n e^{-j2\pi f n t}$$

$\downarrow$   
somma di variabili indipendenti

$$\mathbb{E}[|X(f)|^2] \rightarrow \infty$$

- Densità spettrale di potenza

$$S_x(f) = \lim_{T_0 \rightarrow \infty} \frac{1}{T_0} \mathbb{E}[|X_{T_0}(f)|^2]$$

??  
potenza per  
unità di frequenza

$$X_{T_0}(f) = \int_{-T_0/2}^{T_0/2} x(t) e^{-j2\pi f t} dt$$

### Teatrino di Wiener

$$S_x(f) = \mathcal{F}[R_x(\tau)]$$

Dim.:  $\mathcal{F}^{-1}[S_x(f)] = \lim_{T_0 \rightarrow \infty} \mathbb{E}\left[\frac{\mathcal{F}^{-1}[|X_{T_0}(f)|^2]}{T_0}\right] =$

$$= \lim_{T_0 \rightarrow \infty} \mathbb{E}[\hat{R}(\tau)] = \lim_{T_0 \rightarrow \infty} \frac{T_0 - |\tau|}{T_0} R_x(\tau) =$$

$$= R_x(\tau)$$

$$S_x(f) = \mathcal{F}[R_x(\tau)] = \int R_x(\tau) e^{-j2\pi f \tau} d\tau$$

$$R_x(\tau) = \mathcal{F}^{-1}[S_x(f)] = \int S_x(f) e^{j2\pi f \tau} df$$

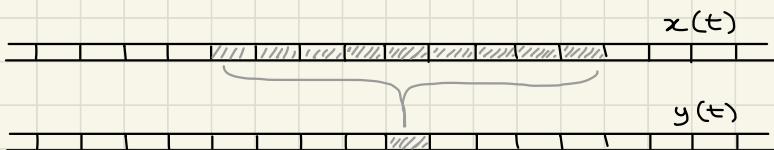
$$P_x = R_x(0) = \int S_x(f) df$$

$P_x(f_0, \Delta f)$  potenza attorno alla frequenza  $f_0$

$\frac{P_x(f_0, \Delta f)}{\Delta f}$  densità di potenza attorno a  $f_0$

$$x(t) \rightarrow \boxed{h(t)} \rightarrow y(t) \quad P_y = E[|y(t)|^2] = P_x(f_0, \Delta f)$$

$y(t)$  è stazionario? ( $x(t)$  sempre staz.)



$y(t) = \int h(\tau) x(t - \tau) d\tau$  ogni valore di  $y(t)$  è la somma pesata di tanti valori di  $x$ , ma poiché il peso è lo stesso per tutti (cioè  $h(t)$ ) se prima istanti di tempo diversi erano equivalenti (stazionarietà) lo sareanno ancora

$$P_y = \int S_y(f) df \text{ per la stazionarietà di } y$$

$$S_y(f) = \left\{ \lim_{T_0 \rightarrow \infty} \frac{1}{T_0} E[X(f)^2] \right\} |H(f)|^2$$

$$\begin{aligned} Y(f) &= X(f) H(f) & S_y(f) &= S_x(f) |H(f)|^2 \\ |Y(f)|^2 &= |X(f)|^2 |H(f)|^2 \end{aligned}$$

$$P_x(f_0, \Delta f) = P_y = \int |H(f)|^2 S_x(f) df = \int_{\Delta f} S_x(f) df = S_x(f_0) \cdot \Delta f$$

$$\left[ S_x(f) = \frac{P_x(f, \Delta f)}{\Delta f} \right]$$

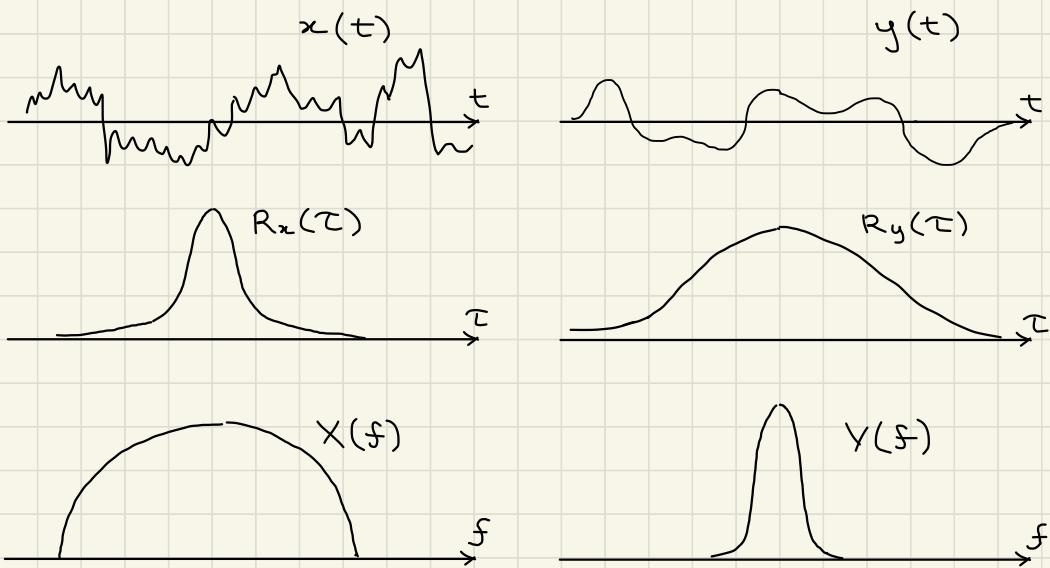
$$S_x(f) = \mathcal{F}[R_x(\tau)] \geq 0 \quad \text{sempre}$$

$$R_x(\tau) = \mathcal{F}^{-1}[S_x(f)]$$

↓

simmetria hermitiana

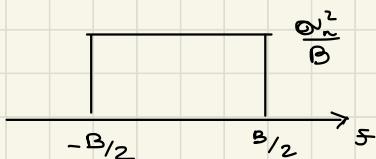
$$R_x(-\tau) = R_x^*(\tau)$$



$$\text{Se } E[x(t)] = \mu_x \rightarrow R_x(\tau) = C_x(\tau) + |\mu_x|^2$$

$$S_x(f) = \mathcal{F}[C_x(\tau)] + |\mu_x|^2 S(f)$$

Rumore bianco



$$n(t) \quad E[|n(t)|^2] = \sigma_n^2$$

$$R_n(\tau) = \sigma_n^2 \frac{\sin(\pi \tau B)}{\pi \tau B}$$

## Processi filtrati (LT I)

$$x(t) \quad y(t) = x(t) * h(t)$$

$$X(f) \quad Y(f) = X(f) \cdot H(f)$$

$$S_x(f) \quad S_y(f) = S_x(f) |H(f)|^2$$

$$R_x(\tau) \quad R_y(\tau) = R_x(\tau) * \mathcal{F}^{-1}[|H(f)|^2] = \\ = R_x(\tau) * \underbrace{h(\tau) * h^*(-\tau)}_{R_h(\tau)}$$

Densità Spettrale di Potenza

$$S_x(f) = \lim_{T_0 \rightarrow \infty} \frac{1}{T_0} E[|X_{T_0}(f)|^2] = \mathcal{F}[R_x(\tau)]$$

Cross-Spettro

$$S_{yx}(f) = \lim_{T_0 \rightarrow \infty} \frac{1}{T_0} E[Y_{T_0}(f) \cdot X_{T_0}^*(f)] = \mathcal{F}[R_{yx}(\tau)]$$

$$S_{yx}(f) = S_x(f) \cdot H(f) \quad \left\{ \begin{array}{l} S_{xy}(f) = S_x(f) H^*(f) \\ R_{xy}(\tau) = R_x(\tau) * h(\tau) \end{array} \right. \quad \left\{ \begin{array}{l} R_{xy}(\tau) = R_x(\tau) * h^*(-\tau) \end{array} \right.$$

Esercizi:

- $y(t) = \alpha x(t - \tau_0)$

Dimostrare che  $R_{yx}(\tau) = \alpha R_x(\tau - \tau_0)$   
 $R_y(\tau) = |\alpha|^2 R_x(\tau)$ .

$$h(t) = \alpha \delta(t - \tau_0)$$

$$y(t) = x(t) * h(t)$$

$$H(f) = \alpha e^{-j2\pi f \tau_0} \quad |H(f)|^2 = |\alpha|^2$$

$$S_{yx}(f) = S_x(f) \cdot \alpha e^{-j2\pi f \tau_0}$$

$$S_y(f) = |\alpha|^2 S_x(f)$$

$$\begin{aligned} R_{xy}(\tau) &= R_x(\tau) * h(\tau) = \\ &= R_x(\tau) * \alpha \delta(\tau - \tau_0) = \alpha R_x(\tau - \tau_0) \end{aligned}$$

$$R_y(\tau) = \mathcal{F}^{-1}[S_y(f)] = |\alpha|^2 \mathcal{F}^{-1}[S_x(f)] = |\alpha|^2 R_x(\tau)$$

- $y(t) = \alpha x(t - \tau_a) + b(t - \tau_b)$

Dimostrare che

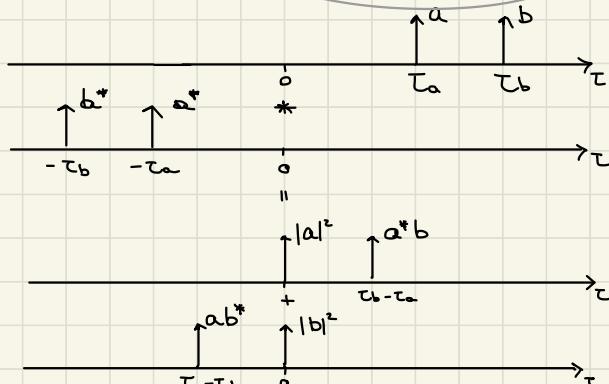
$$R_y(\tau) = \alpha R_x(\tau - \tau_a) + b R_x(\tau - \tau_b)$$

$$R_{xy}(\tau) = (|\alpha|^2 + |b|^2) R_x(\tau) + ab^* R_x(\tau - (\tau_a - \tau_b)) + a^* b R_x(\tau - (\tau_b - \tau_a))$$

$$h(t) = \alpha \delta(t - \tau_a) + b \delta(t - \tau_b)$$

$$R_{yx}(\tau) = R_x(\tau) * h(\tau) = \alpha R_x(\tau - \tau_a) + b R_x(\tau - \tau_b) \quad \checkmark$$

$$R_y(\tau) = R_x(\tau) * h(\tau) * h^*(-\tau)$$



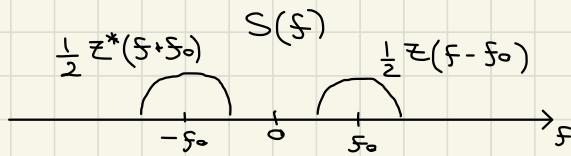
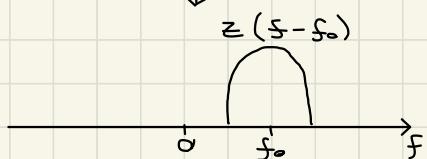
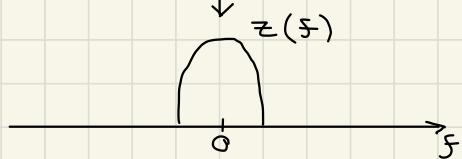
$$= R_x(\tau) * [(|\alpha|^2 + |b|^2) + ab^* \delta(\tau - (\tau_a - \tau_b)) + a^* b \delta(\tau - (\tau_b - \tau_a))] \quad \checkmark$$

componente in fase componenti in quadratura

- $$s(t) = x(t) \cos(2\pi f_0 t) - y(t) \sin(2\pi f_0 t)$$

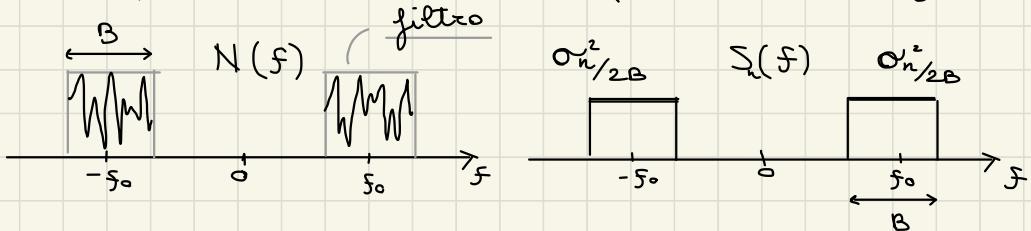
$$= \operatorname{Re} [z(t) e^{+j2\pi f_0 t}]$$

$$z(t) = x(t) + jy(t)$$



$$E_s = \frac{1}{2} E_z \quad P_s = \frac{1}{2} P_z$$

$n(t)$  processo casuale (rumore o segnale)



$$\left[ P_n = E[n^2(t)] \right] = B k T$$

$$n(t) = w_I(t) \cos(2\pi f_0 t) - w_Q(t) \sin(2\pi f_0 t)$$

$$= \operatorname{Re} [w(t) e^{+j2\pi f_0 t}]$$

$$w(t) = w_I(t) + jw_Q(t)$$

$$E[n(t)] = 0 \implies E[w_I] \cdot \cos - E[w_Q] \sin = 0$$

media nulla

$$\mathbb{E}[n^2(t)] = \sigma_n^2 = \mathbb{E}[w_i^2(t)] \cos^2(2\pi f_0 t) + \mathbb{E}[w_o^2(t)] \sin^2(2\pi f_0 t)$$

$$- 2 \mathbb{E}[w_i(t) w_o(t)] \sin(2\pi f_0 t) \cos(2\pi f_0 t)$$

$$\Rightarrow \mathbb{E}[w(t)] = 0$$

$$\Rightarrow \mathbb{E}[|w(t)|^2] = 2\sigma_n^2$$

$$(E_z = 2E_s, P_z = 2P_s)$$

per avere  $\sigma_w^2 = \sigma_n^2$

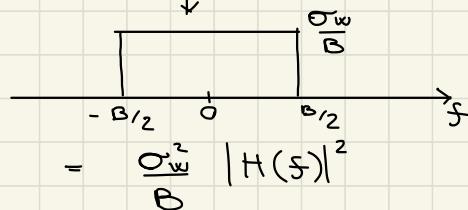
basterebbe porre  $w(t) = \frac{1}{\sqrt{2}} (w_i(t) + j w_o(t))$

- $\varepsilon_{rx}(t) = g(t) + w(t)$

$\varepsilon_{rx}(t) = \varepsilon(t) * h(t) = g(t) * h(t) + w_r(t)$

$$S_{w_r}(f) = S_w(f) \cdot |H(f)|^2$$

$$w(t) * h(t)$$



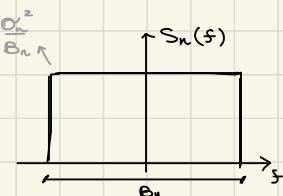
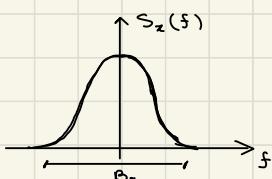
$$\sigma_{w_r}^2 = \int S_{w_r}(f) df = \frac{\sigma_w^2}{B} \int |H(f)|^2 df = \frac{\sigma_w^2}{B} \int |h(t)|^2 dt$$

- $x(t)$  processo stazionario  
 $n(t)$

$$\mathbb{E}[x(t)] = \mathbb{E}[n(t)] = 0$$

$$\sigma_x^2 = \mathbb{E}[|x(t)|^2]$$

$$\sigma_n^2 = \mathbb{E}[|n(t)|^2]$$



$$\mathbb{E}[x(t+\tau) n^*(t)] = 0 \quad \forall \tau$$

$$B_n > B_x$$

scorrelati

$$y(t) = x(t) + n(t)$$

sequale  
rumore

a)  $E[y(t)] = ?$

b)  $P_y = E[|y(t)|^2] = ?$

c)  $R_y(\tau) = ?$

d)  $S_y(f) = ?$

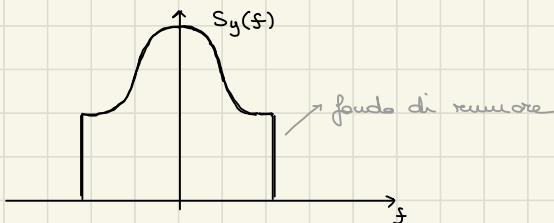
e)  $SNR = ?$

Q)  $E[y(t)] = E[x(t)] + E[n(t)] = 0$

c)  $R_y(\tau) = E[y(t+\tau)y^*(t)] = E[(x(t+\tau)+n(t+\tau)) \cdot (x(t)+n(t))^*] =$   
 $= E[x(t+\tau)x^*(t) + n(t+\tau)x^*(t) + x(t+\tau)n^*(t) + n(t+\tau)n^*(t)] =$   
 $\quad \quad \quad \downarrow \quad \quad \quad \downarrow \quad \quad \quad \downarrow \quad \quad \quad \downarrow$   
 $\quad R_x(\tau) \quad \quad \quad R_{nx}(\tau) = 0 \quad \quad R_{xn}(\tau) = 0 \quad \quad R_n(\tau)$   
 $= R_x(\tau) + R_y(\tau)$   
 $\quad \quad \quad \downarrow \quad \quad \quad \downarrow$   
 $\quad \quad \quad \mathcal{F}^{-1}[S_y(f)] \quad \alpha_x^2 \operatorname{sinc}(tB_x)$

b)  $P_y = R_y(0) = R_x(0) + R_n(0) = \alpha_x^2 + \alpha_n^2$

d)  $S_y(f) = \mathcal{F}[R_y(\tau)] = S_x(f) + S_n(f)$



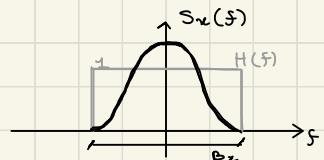
e)  $SNR = \frac{P_x}{P_n} = \frac{\alpha_x^2}{\alpha_n^2}$

$y(t) \rightarrow \boxed{h(t)} \rightarrow v(t) \quad \text{con } h(t): SNR_h > SNR$

$\Rightarrow h(t)$  deve essere un filtro con la banda di  $x$  (e ampl. 1)

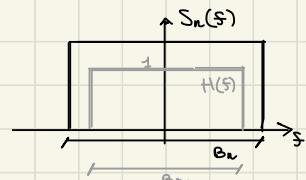
$$H(f) = \operatorname{rect}\left(\frac{f}{B_x}\right) \longrightarrow h(t) = B_x \operatorname{sinc}(tB_x) = \frac{\sin(\pi t B_x)}{\pi t}$$

$$\begin{aligned}
 u(t) &= y(t) * h(t) \\
 &= x(t) * h(t) + n(t) * h(t) \\
 &\downarrow \qquad \qquad \downarrow \\
 x(t) &\qquad\qquad w(t)
 \end{aligned}$$



$$SNR_h = \frac{\sigma_x^2}{\sigma_n^2}$$

$$\begin{aligned}
 S_w(f) &= |H(f)|^2 S_n(f) \\
 P_w &= \int_{-\infty}^{\infty} S_w(f) df = \int_{-\infty}^{\infty} S_n(f) = \frac{\sigma_n^2}{B_n} B_n < \frac{\sigma_x^2}{B_n}
 \end{aligned}$$



$$\Rightarrow SNR_h = \frac{\sigma_x^2}{\sigma_n^2} \cdot \frac{B_n}{B_x}$$

- $x(t)$  processo stazionario con autocorrelazione  $R_x(\tau)$

$$x'(t) = \frac{dx(t)}{dt} \quad R_{x'x}(\tau) = ?$$

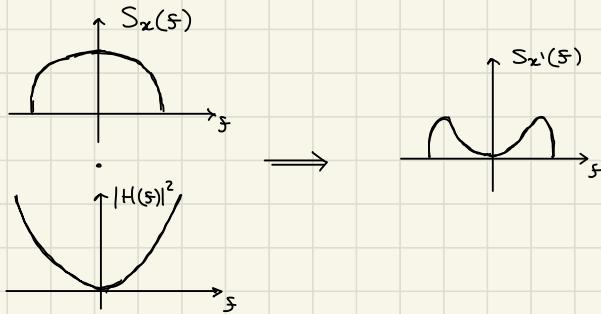
$$S_{x'}(f) = ?$$

$$S_{x'x}(f) = H(f) S_x(f) = j2\pi f S_x(f)$$

$$R_{x'x}(\tau) = \mathcal{F}^{-1}[S_{x'x}(f)] = \frac{d}{d\tau} R_x(\tau)$$

$$S_{x'}(f) = |H(f)|^2 S_x(f) = 4\pi^2 f^2 S_x(f)$$

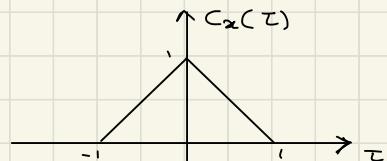
→ filtro passa-basso



- $x(t)$  processo stazionario gaussiano reale  
 $x(t) \sim \mathcal{N}(\mu_x, \sigma_x^2)$

$$\rho_x = 26$$

$$C_x(\tau) = \begin{cases} 1 - |\tau| & |\tau| < 1 \\ 0 & |\tau| > 1 \end{cases} \quad \rightarrow$$



$$\sigma_x^2 = C_x(0) = 1$$

$$\rho_x = \sigma_x^2 + \mu_x^2 \implies \mu_x = \pm 5$$

vanno bene entrambi,  
ne sceglie uno

$$x(t) \sim \mathcal{N}(5, 1)$$

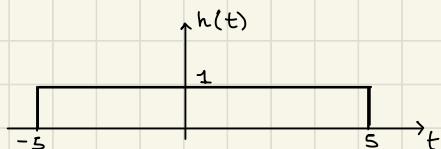
$$R_x(\tau) = C_x(\tau) + \mu_x^2$$

$$S_x(f) = \Im[C_x(\tau)] + 2\delta(f) = \left(\frac{\sin(\pi f)}{\pi f}\right)^2 + 2\delta(f)$$

$$h(t) = \text{rect}\left(\frac{t}{10}\right)$$

↓

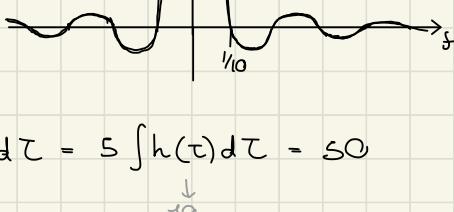
$$H(f) = 10 \sin\left(10f\right)$$



$$y(t) = x(t) * h(t)$$

$$E[y(t)] = \int E[x(t-\tau)] \cdot h(\tau) d\tau = 5 \int h(\tau) d\tau = 50$$

$$E[y(t)] = \mu_x \cdot H(0)$$

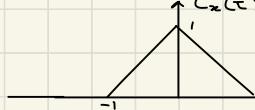


$$C_y(\tau) = R_y(\tau) - \mu_y^2$$

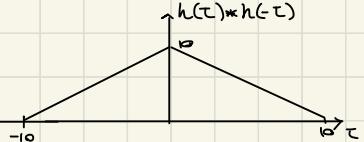
$$R_y(\tau) = R_x(\tau) * h(\tau) * h(-\tau) =$$

$$= \underbrace{C_x(\tau) * h(\tau) * h(-\tau)}_{C_y(\tau)} + \underbrace{\mu_x^2 * h(\tau) * h(-\tau)}_{\mu_y^2}$$

$$\Rightarrow C_y(\tau) = C_x(\tau) * \overbrace{h(\tau) * h(-\tau)}^{R_h(\tau)} =$$



\*



Voglio sapere il tempo di decorrelazione cioè il  $\tau$  oltre il quale la correlazione è sempre nulla (di  $C_y$ )

$$T_{C_y} = 2 + 20 = 22 \text{ s}$$

$$\rightarrow t_{dec} = 11 \text{ s}$$

$$\sigma_y^2 = ?$$

$$\sigma_y^2 = C_y(0) \rightarrow C_y(\tau) = \int C_x(t) R_h(\tau-t) dt$$

$$C_y(0) = \int C_x(t) R_h(-t) dt$$

$$\rightarrow \sigma_y^2 = \int_{-1}^1 (1-|t|) \cdot (10-|t|) dt$$

$$= 2 \int_0^1 (10-t-10t+t^2) dt$$

$$= 2 \left[ \frac{t^3}{3} - \frac{11}{2}t^2 + 10t \right]_0^1 = 2 \left[ \frac{1}{3} - \frac{11}{2} + 10 \right]$$

$$= 2 \cdot \left[ \frac{2-33+60}{6} \right] = \frac{29}{3}$$

## Processi campionati

$$x(t) \xrightarrow[t=nT]{} x_n \quad R_{x_n}(k) = E[x_{n+k} x_n^*] = R_x(\tau=kT)$$

$$x_n = x(t=nT) \longrightarrow R_{x_n}(k) = R_x(t=kT)$$

$$C_{x_n}(k) = C_x(t=kT)$$

$$y_n = x_n * h_n$$

$$R_{yx}(k) = R_x(k) * h_k * h_{-k}^* \quad R_{yx_n}(k) = R_x(k) * h_k$$

$$S_{x_n}(f) = \mathcal{F}[R_{x_n}(k)] = \frac{1}{T} \sum_{k=-\infty}^{\infty} S_x(f - \frac{k}{T})$$

$$S_{y_n}(f) = |H(f)|^2 S_{x_n}(f)$$

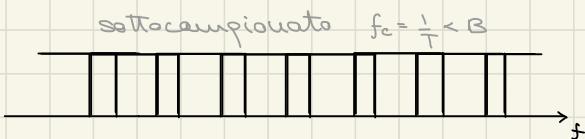
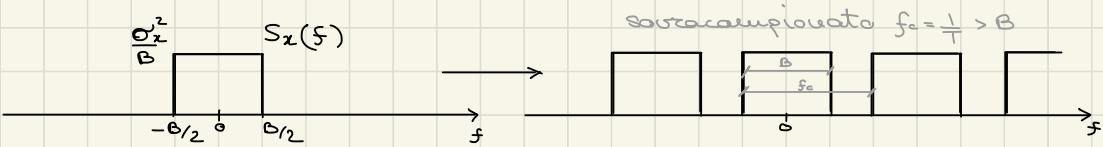
$$S_{y_n}(f) = H(f) S_{x_n}(f)$$

$$P_{x_n} = T \int_{-\frac{B}{2T}}^{\frac{B}{2T}} S_{x_n}(f) df$$

Esercizi:

- $x(t)$  processo  $E[x(t)] = 0$ ,  $E[|x(t)|^2] = \sigma_x^2$  bianco nella banda  $B$ .

$$R_x(\tau) = \sigma_x^2 \frac{\sin(\pi B \tau)}{\pi B \tau} \longrightarrow R_{x_n}(k) = \sigma_x^2 \frac{\sin(\pi B k T)}{\pi B k T}$$



- $$\bullet \quad x(t) \text{ processo } \mu_x = 1, \sigma_x^2 = 2, C_x(\tau) = 0 \quad \forall \tau > 5ms$$

$$x_n = x(t = nT) \text{ con } T = 17ms \quad \uparrow c_x$$

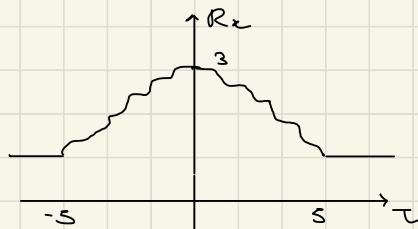
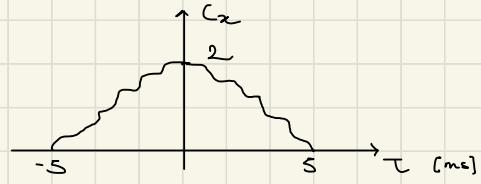
$$S_{x_n}(s) = ?$$

$$R_x(\tau) = C_x(\tau) + |M_x|^2$$

$$P_x = \sigma_x^2 + |M_x|^2 = 3$$

$$C_{x_m}(k) = C_x(T = kT) = \sigma_x^2 \delta_k$$

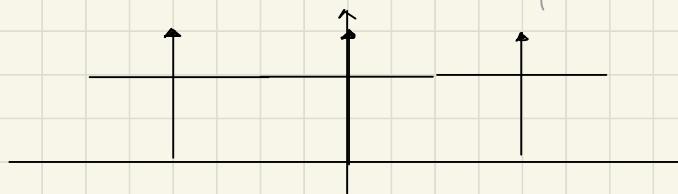
Poiché  $T > 5\text{ms}$ ,  $C_{x_n} \neq 0$   
solo per  $K = 0$ .



$$R_{zn}(k) = \sigma_x^2 \delta_k + |\mu_x|^2$$

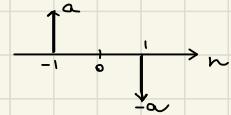
$$S_{X_K}(f) = \mathcal{F} [R_{X_K}(K)] = \sum_{k=-\infty}^{+\infty} R_{X_K}(k) e^{-j2\pi f k T}$$

$$= \mathcal{F}[\alpha_x^2 \delta_k] + \mathcal{F}[|\mu_x|^2] \quad \xrightarrow{\alpha_x^2 \downarrow} \quad \frac{|\mu_x|^2}{T} \delta(f)$$



Progettare un filtro FIR (Finite Impulse Response) affinché il processo filtrato abbia  $\mu_y = 0$  e  $\sigma_y^2 = 16$

$$y_n = x_n * h_n, \quad h_n = a(\delta_{n+1} - \delta_{n-1})$$



$$E[y_n] = E\left[\sum_k x_{n-k} h_k\right] = \sum_k E[x_{n-k}] h_k = \mu_x \sum_k h_k = \mu_x - \mu_y$$

$$E[|y_n|^2] = E[|\alpha|^2 |x_{n+1}|^2 + |\alpha|^2 |x_{n-1}|^2 - 2 |\alpha|^2 \operatorname{Re}(x_n, x_{n-1}^*)]$$

$$\text{Re} [E[x_{n+1} x_{n-1}^*]] \xrightarrow{\text{incorretto}} \text{Re} [E[x_{n+1}] E[x_{n-1}]]$$

$\downarrow \mu_x \quad \downarrow \mu_x^*$

$$\rightarrow P_y = (|a|^2 \cdot 3 + |a|^2 \cdot 3 - 2 \cdot |a|^2 \cdot 1 \cdot 1) = |a|^2 \cdot 4 = \sigma_y^2 = 16$$

$$\Rightarrow a = \pm 2 \quad h_n = 2(\delta_{n+1} - \delta_{n-1})$$

## Trasformate di Fourier in Matlab

$$X(f) = \int x(t) e^{-j2\pi f t} dt$$

tempo di campionamento

$$X(f) = \left( \sum_n x_n e^{-j2\pi f n T} \right) \cdot \Delta t$$

$x_n(f)$

$f_c = \frac{1}{\Delta t}$

$f \in \left( -\frac{f_c}{2}, \frac{f_c}{2} \right)$

$$X = W_{kn} \cdot x_n \cdot \Delta t$$

frequenza di campionamento

$$W_{kn} = \begin{bmatrix} & n \\ & \vdots \\ e^{-j2\pi f_k t_n} & \end{bmatrix}_k$$

$$x_n = \begin{bmatrix} x_1 \\ \vdots \\ x_n \end{bmatrix}_n$$

$$f = \left[ -\frac{f_c}{2} \quad \cdot \quad \underbrace{\cdot}_{\Delta f} \cdots \quad \frac{f_c}{2} \right]^T$$

$$x(t) = \int X(f) e^{j2\pi f t} df = \sum_k X_k e^{j2\pi f_k t_n} \cdot \Delta f \quad \frac{1}{\Delta f} > T_0$$

$$N = \frac{T_0}{\Delta t} \quad N_f = \frac{f_c}{\Delta f} = \frac{1}{\Delta t} \cdot \frac{1}{\Delta f} \rightarrow N_f > N$$

$$x_n = \sum_k X_k e^{j2\pi f_k t_n} \Delta f \cdot \Delta t$$

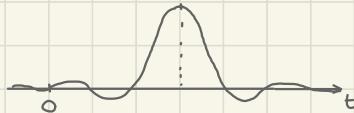
$$\longrightarrow X_k = \sum_n x_n e^{-j2\pi f_k t_n} \iff x = Wx$$

$$x_n = \sum_k X_k e^{j2\pi f_k t_n} \iff x = W^H x$$

inversa & complessa coniugata

$$X = \text{FFT}(x, N_f)$$

$$X_k = \sum_{n=0}^{N_f-1} x_n e^{-j2\pi \frac{k n}{N_f}}$$



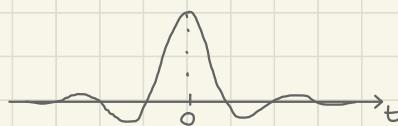
$$f_k = \frac{k}{N_f} \quad f_k = \left[ 0, \frac{1}{N_f}, \dots, \frac{N_f-1}{N_f} \right]$$

mi fa vedere solo metà dell'asse delle frequenze

↓  
FFT shift

$$X = \text{FFTshift}(\text{FFT}(x, N_f))$$

$$f_k = \left[ -\frac{N_f-1}{2}, \dots, \frac{N_f-1}{2} \right] \frac{1}{N_f}$$

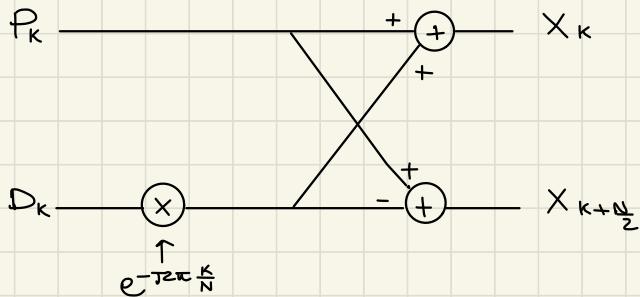


Su Matlab calcolare la trasformata di Fourier con le matrici è un'operazione molto gravosa. Per questo per calcolarla la funzione implementata in Matlab divide la sommatoria in 2 sommatorie di termini pari e dispari

$$X_k = \sum_{n=0}^{N_f-1} x_n e^{-j2\pi \frac{k n}{N_f}} = \underbrace{\sum_{n=0}^{N_f-1} x_{2n} e^{-j2\pi \frac{k 2n}{N_f}}}_{(N)^2} + e^{-j2\pi \frac{k}{N_f}} \underbrace{\sum_{n=0}^{N_f-1} x_{2n+1} e^{-j2\pi \frac{k 2n+1}{N_f}}}_{D_N \left(\frac{N}{2}\right)^2} \quad (N=N_f)$$

$$X_k = P_k + e^{-j2\pi \frac{k}{N_f}} D_k$$

$$X_{k+N_f} = P_k - e^{-j2\pi \frac{k}{N_f}} D_k$$



Se ripeto il processo e divido  $P_k$  e  $D_k$  posso raggiungere un numero di calcoli minimo pari a

$$\boxed{\frac{N}{2} \log_2 N} \rightarrow \text{se } N \text{ è potenza di 2}$$

### Stima

$$\hat{\theta} = \hat{\theta}(\theta, \text{rumore, o.c.})$$

↓  
parametri  
deterministici

$$\hat{\theta}(\theta) = \hat{\theta}$$

↓  
stimatore      stima

$$[E_m = \hat{\theta} - \theta]$$

Valutazione teorica:

$$E[\hat{\theta} - \theta] \begin{cases} \neq 0 & \text{polarizzato (biased)} \\ = 0 & \text{non polarizzato (unbiased)} \end{cases}$$

migliore

Le stimatrici

$$E[(\hat{\theta} - \theta)^2] \quad \text{Mean Root Square (MSR)}$$

migliore se piccole

Se unbiased:  
 $MSR = \text{var}(\hat{\theta})$

Valutazione sperimentale:

$$\begin{aligned}\hat{\theta}_1 &= \hat{\theta}_1 \\ \hat{\theta}_2 &= \hat{\theta}_2 \\ \hat{\theta}_3 &= \hat{\theta}_3 \\ \vdots & \\ \hat{\theta}_n &= \hat{\theta}_n\end{aligned}$$

$$\frac{1}{N} \sum_{n=1}^N (\hat{\theta}_n - \theta)$$

Media

$$\frac{1}{N} \sum_{n=1}^N (\hat{\theta}_n - \theta)^2$$

MSR

$$\text{Es: } \hat{\theta}_{1\dots N} = \theta + w_n \quad N \text{ campioni}$$

$$w_n \sim \mathcal{N}(0, \sigma^2) \quad E[w_n w_m] = 0 \xrightarrow{\text{autocorrelazione imp.}} n \neq m$$

$$\rightarrow \hat{\theta} = \hat{\theta}_1 \quad \varepsilon = \hat{\theta} - \theta = w_n$$

$$E[\hat{\theta} - \theta] = E[\hat{\theta}_1 - \theta] = E[w_n] = 0$$

$$E[(\hat{\theta} - \theta)^2] = E[w_n^2] = \sigma^2 \implies \hat{\theta} = \theta \pm \sigma$$

$$\rightarrow \hat{\theta} = \frac{1}{N} \sum_{n=1}^N \hat{\theta}_n \quad \varepsilon = \frac{1}{N} \sum_{n=1}^N (\theta + w_n) - \theta = \frac{1}{N} \sum_n w_n$$

$$E[\hat{\theta} - \theta] = \frac{1}{N} \sum_{n=1}^N E[\hat{\theta}_n] - \theta = 0$$

$$E[(\hat{\theta} - \theta)^2] = \frac{1}{N} \sigma^2 \implies \hat{\theta} = \theta \pm \frac{\sigma}{\sqrt{N}}$$

$$\begin{aligned}E\left[\left(\frac{1}{N} \sum_{n=1}^N w_n\right)^2\right] &= \frac{1}{N^2} E\left[\sum_{n=1}^N w_n^2\right] = \frac{1}{N^2} \sum_{n=1}^N E[w_n^2] \\ &= \frac{1}{N^2} \cdot N \sigma^2 = \frac{1}{N} \sigma^2\end{aligned}$$

$$\text{Es: } x_n = \mu + w_n \quad R_w(m) = \sigma_w^2 \delta_m$$

$$\hat{\mu} = \frac{1}{N} \sum_{n=1}^N x_n$$

$$\sqrt{E[\varepsilon^2]} = \frac{\sigma_w}{\sqrt{N}} \quad \text{deviazione standard dell'era}$$

non posso + ricavare

Se aumentassi  $N$  di molto (sotto campionam.) posso migliorare la stima?  $\rightarrow$  Non per forza

$\downarrow$   
xx l'autocorrelat. del rum.  
non è + impulsiva

$$x(t) = \mu + w(t)$$

$$\hat{\mu} = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) dt \quad \varepsilon = \hat{\mu} - \mu \quad E[\varepsilon] = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} E[x(t)] dt$$

$E[w(t)] = 0$  bianco nella banda  $B$



sottracc. mi  
da un' autocorrelazione del rumore  
non più impulsiva

$$\varepsilon = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} w(t) dt = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} w(\tau) \operatorname{rect}\left(\frac{\tau}{T_0}\right) d\tau$$

$$\varepsilon(t) = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} w(\tau) \operatorname{rect}\left(\frac{t-\tau}{T_0}\right) d\tau = w(t) * \frac{1}{T_0} \operatorname{rect}\left(\frac{t}{T_0}\right)$$

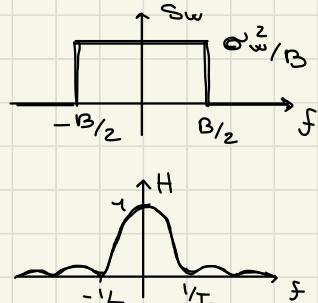
↓

$$\varepsilon = \varepsilon(0) \longrightarrow E[\varepsilon^2] = P_\varepsilon = \int S_\varepsilon(f) df$$

$$S_z(f) = S_w(f) |H(f)|^2$$

$$\frac{\sigma_w^2}{B} \operatorname{rect}\left(\frac{f}{B}\right)$$

$$\operatorname{sinc}^2\left(\pi f T_0\right)$$



$$\Rightarrow P_z = \frac{\sigma_w^2}{BT_0^2} \int_{-B/2}^{B/2} \left( \frac{\sin \pi f T_0}{\pi f} \right)^2 df$$

caso 1.  $T_0 \ll \frac{1}{B}$   $\Rightarrow P_z \approx \sigma_w^2$

tempo di decorrelazione del rumore

caso 2.  $T_0 \gg \frac{1}{B} \Rightarrow P_z \approx \frac{\sigma_w^2}{BT_0^2} \int_{-\infty}^{+\infty} |h(t)|^2 dt = \frac{\sigma_w^2}{BT_0} \operatorname{rect}\left(\frac{t}{T_0}\right)$

$BT_0$  = numero equivalente di campioni indipendenti

$$N = T_0 B$$

$$x(t) \longrightarrow x_n = \mu + w_n$$

$$E[\varepsilon^2] = \frac{\sigma_w^2}{N}$$

$T = \frac{1}{B}$  campionam. critico

autocorrelazione di  $w_n$  impulsiva

### Stima potenza

$$\varepsilon(t) \sim \mathcal{CN}(0, \sigma_z^2)$$

$$\varepsilon = x + jy$$

$$E[x^2] = E[y^2] = \frac{\sigma_z^2}{2}$$

$$E[x] = E[y] = E[xy] = 0$$

$$E[\varepsilon \varepsilon^*] = \sigma_z^2$$

$$* \mathbb{E} [\varepsilon_a \varepsilon_b^* \varepsilon_c^* \varepsilon_d] =$$

$\mathbb{E} [\varepsilon_a \varepsilon_b^*] \in [\varepsilon_d \varepsilon_c^*] + \mathbb{E} [\varepsilon_a \varepsilon_c^*] \in [\varepsilon_d \varepsilon_b^*]$

$$\rho_z = \mathbb{E}[|z|^2] = \sigma_z^2$$

$$\rightarrow \hat{\rho}_z = |z(t)|^2 \text{ con } t \text{ fissato}$$

$$\begin{aligned} \varepsilon &= \hat{\rho}_z - \rho_z \\ &= \hat{\rho}_z - \sigma_z^2 \end{aligned}$$

$$\mathbb{E}[\hat{\rho}_z] = \mathbb{E}[|z(t)|^2] = \sigma_z^2$$

$$\mathbb{E}[\varepsilon] = \mathbb{E}[\rho_z] - \sigma_z^2 = 0 \quad \text{non polarizz.}$$

$$\mathbb{E}[\varepsilon^2] = \mathbb{E}[(\hat{\rho}_z - \sigma_z^2)^2] = \mathbb{E}[\hat{\rho}_z^2] - \sigma_z^4$$

$$\mathbb{E}[\hat{\rho}_z^2] = \mathbb{E}[|z(t)|^4] = \mathbb{E}[\varepsilon \varepsilon^* \varepsilon \varepsilon^*] =$$

$$* = \sigma_z^4 + \sigma_z^4 = 2\sigma_z^4 \quad \text{brutto}$$

$$\Rightarrow \mathbb{E}[\varepsilon^2] = \sigma_z^4 \quad \rightarrow \text{deviazione standard della stima } \sigma_z^2$$

$$\rightarrow \hat{\rho}_z = \frac{1}{T_0} \int_0^{T_0} |z(t)|^2 dt \quad o \quad \hat{\rho}_z = \frac{1}{N} \sum_n |\varepsilon_n|^2$$

$$\mathbb{E}[\hat{\rho}_z] = \sigma_z^2 \quad \mathbb{E}[\varepsilon^2] = \text{var}[\hat{\rho}_z] = \frac{\sigma_z^4}{T_0 B}$$

$$\text{se } |z(t)|^2 = \sigma_z^2 + |w(t)|^2$$

Riassunto: stima di media e stima di potenza

$$x(t) = \mu + \omega(t)$$

$$\mathbb{E}[\omega(t)] = 0 \quad \mathbb{E}[|\omega(t)|^2] = \sigma_\omega^2$$

Tempo discreto:  $\hat{\mu} = \frac{1}{N} \sum_n x_n$

Tempo continuo:  $\hat{\mu} = \frac{1}{T_0} \int_{T_0} x(t) dt$

Errore di stima:  $\varepsilon = \hat{\mu} - \mu$   
 $\mathbb{E}[\varepsilon] = 0 \quad \mathbb{E}[|\varepsilon|^2] = \frac{\sigma_\omega^2}{N_{eq}}$

( $T_0$ )

$$z(t) \sim \mathcal{CN}(0, \sigma_z^2) \quad P_z = \mathbb{E}[|z(t)|^2] = \sigma_z^2$$

$$|z(t)|^2 = P_z + v(t)$$

$$\mathbb{E}[v(t)] = 0 \quad \text{var}[v(t)] = \sigma_v^2$$

Tempo discreto:  $\hat{P}_z = \frac{1}{N} \sum_n |z_n|^2$

Tempo continuo:  $\hat{P}_z = \frac{1}{T_0} \int_{T_0} |z(t)|^2 dt$

Errore di stima:  $\varepsilon = \hat{P}_z - P_z$   
 $\mathbb{E}[\varepsilon] = 0 \quad \mathbb{E}[|\varepsilon|^2] = \frac{\sigma_v^4}{N_{eq}}$

## Stima autocorrelazione

$$x(t) \sim \mathcal{C} \mathcal{N}(0, \sigma_x^2)$$

$$\left[ \hat{R}_x(\tau) = \frac{1}{T_0} \int_{-T_0}^{T_0} x(t+\tau) x^*(t) dt \right] \text{ tempo continuo}$$

$$\left[ \hat{R}_x(m) = \frac{1}{N} \sum_n^N x_{n+m} x_n^* \right] \text{ tempo discreto}$$

Ricordiamo che:

$$\left[ E[\hat{R}_x(\tau)] = R_x(\tau) \cdot \frac{T_0 - |\tau|}{T_0} \right]$$



Tanto più  $T_0$  è piccolo  $\rightarrow$  tanto più il triangolo è stretto  $\rightarrow$  tanto peggiore è la stima

Consideriamo una nuova variabile casuale:

$$\mu(t) = x(t+\tau) x^*(t)$$
  
(oppure  $\mu_n = \mu_{n+m} \mu_n^*$ )

$$\mu(t) = \underbrace{E[\mu(t)]}_{R_x(\tau)} + \mathcal{V}(t)$$

$$E[\mathcal{V}(t)] = 0$$

$$E[|\mathcal{V}(t)|^2] = \text{var}[\mu(t)] = \sigma_\mu^2$$

$$\text{var}[\hat{R}_x(\tau)] = \frac{\sigma_\mu^2}{N_{\text{eq}}}$$

$$\text{var}[\mu(t)] = E[|\mu(t)|^2] - \underbrace{|E[\mu(t)]|^2}_{(R_x(\tau))^2} = \sigma_\mu^2$$

$$\begin{aligned} E[|\mu(t)|^2] &= E[x(t+\tau) x^*(t) x^*(t+\tau) x(t)] = R_x(\tau) \cdot R_x^*(\tau) \\ &= E[x(t+\tau) x^*(t)] E[x^*(t+\tau) x(t)] + \\ &\quad + E[|x(t+\tau)|^2] \cdot E[|x(t)|^2] = |R_x(\tau)|^2 + \sigma_x^4 \end{aligned}$$

$$\sigma_x^2 \cdot \sigma_x^2$$

$$\implies [\text{var}[\mu(t)] = \sigma_x^4 + |R_x(\tau)|^2 - |R_x(\tau)|^2 = \sigma_x^4]$$

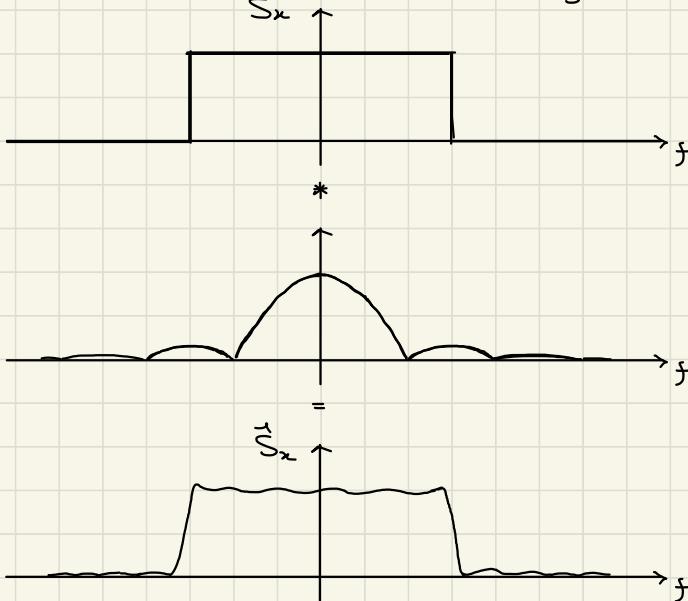
Stima densità spettrale di potenza

$$S_x(f) = \lim_{T_0 \rightarrow \infty} \frac{1}{T_0} E[|X_{T_0}(f)|^2] = \mathcal{F}[R_x(\tau)]$$

$$[\hat{S}_x(f) = \frac{1}{T_0} E[|X_{T_0}(f)|^2] = \mathcal{F}[\hat{R}_x(\tau)] \quad \text{Periodogramma}]$$

$$E[\hat{S}_x(f)] = \mathcal{F}[E[\hat{R}_x(\tau)]] = \mathcal{F}[R_x(\tau) \frac{T_0 - |\tau|}{T_0}]$$

$$\hookrightarrow E[\hat{S}_x(f)] = S_x(f) * \frac{1}{T_0} \left( \frac{\sin \pi f T_0}{\pi f} \right)^2$$



$$\text{var}[\hat{S}_x(f)] \rightarrow \hat{S}_x(f) = \left| \frac{X_{T_0}(f)}{\sqrt{T_0}} \right|^2 = |A(f)|^2$$

$$E[|A(f)|^2] = E[\hat{S}_x(f)] = \sigma_A^2$$

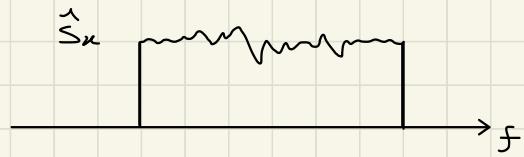
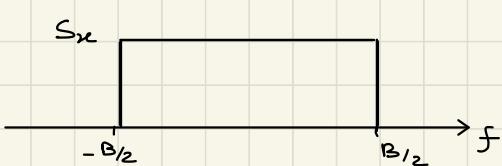
$$\text{Se } A(f) \sim \mathcal{N}(0, \sigma_A^2) \text{ allora } \text{var}[|A(f)|] = \sigma_A^4$$

$$\implies [\text{var}[\hat{S}_x(f)] = (\mathbb{E}[S_x(f)])^2]$$

Questo significa che:  $\sigma_{\hat{S}_x}^2 = \mathbb{E}[\hat{S}_x(f)]$

Non sempre vale che  $A(f) \sim \mathcal{CN}(0, \sigma_A^2)$

↪ allora la  $\text{var}[\hat{S}_x(f)]$  è una stima



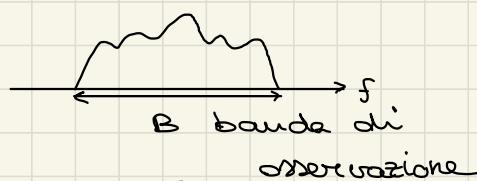
Troppo lungo

$$\mathbb{E}[\hat{S}_x(f)] = \frac{\sigma_x^2}{B} \quad \text{var}[\hat{S}_x(f)] = \left(\frac{\sigma_x^2}{B}\right)^2$$

$$\hat{S}_x(f) = \frac{\sigma_x^2}{B} + \mathcal{V}(f)$$

$$\hat{S}_0 = \frac{1}{B} \int_B \hat{S}_x(f) df$$

$$\mathbb{E}[\hat{S}_0] = S_0 = \frac{\sigma_x^2}{B} \quad \text{var}[\hat{S}_0] = \frac{S_0^2}{N_{eq}}$$



Distanza di correlazione  $\Delta f = \frac{1}{T_0}$

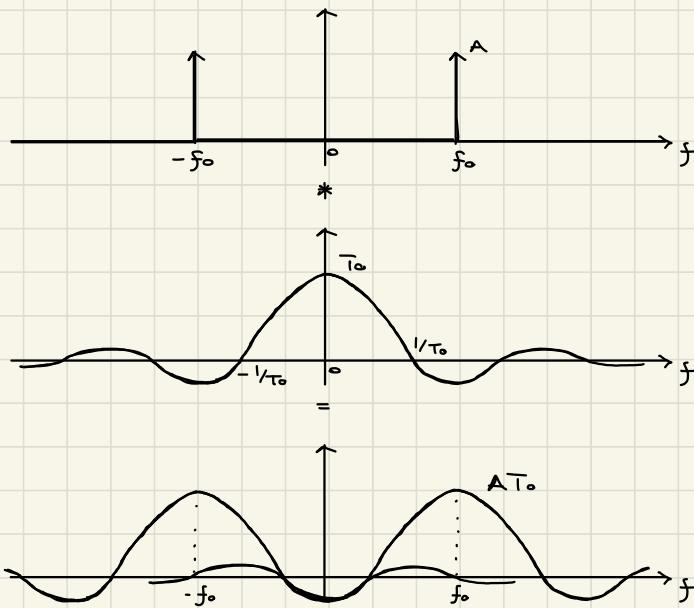
$$X_{T_0}(f) = \int_{T_0} x(t) e^{-j2\pi ft} dt$$

$$X_{T_0}(f) = X(f) * \frac{\sin(\pi f T_0)}{\pi f}$$

→ tutte le trasformate che facciamo sono nella realtà convolute con un

seno cardinale e scalate di un fattore  $\times T_0$ .

Per esempio una sinusoida:



$$E[x^*(f_1) x(f_2)] \begin{cases} = 0 & |f_1 - f_2| > \Delta f \\ \neq 0 & |f_1 - f_2| < \Delta f \end{cases} \rightarrow \text{distanza di correlazione}$$

$$E[x^*(t_1) x(t_2)] \begin{cases} = 0 & |t_1 - t_2| > \tau_{dec} \\ \neq 0 & |t_1 - t_2| < \tau_{dec} \end{cases} \rightarrow \text{tempo di decorrelazione}$$

## Preditzione Lineare

$x(t)$  processo  $E[x(t)] = 0, R_x(\tau)$

Vogliamo stimare  $x(t+\tau)$  noto  $x(t)$

$\hat{x}(t+\tau)$  stima di  $x(t+\tau)$

$$[\hat{x}(t+\tau) = c \cdot x(t)] \text{ (dove } c = \text{const.)}$$

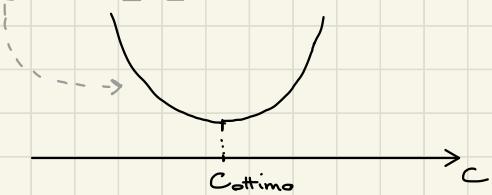
$$[\varepsilon = \hat{x}(t+\tau) - x(t+\tau)]$$

$$[\text{MSE} = E[|\varepsilon|^2]] \text{ (Mean Square Error)}$$

Sceglio  $c$  t.c. MSE sia minimo (m MSE)

$$\varepsilon = c \cdot x(t) - x(t+\tau)$$

$$\begin{aligned} \text{MSE} &= E[|\varepsilon|^2] = E[x^2(t+\tau) - 2c x(t)x(t+\tau) + c^2 x^2(t)] \\ &= R_x(0) - 2c R_x(\tau) + c^2 R_x(0) \end{aligned}$$



Per trovare  $c_0$

è sufficiente derivare:

$$\frac{d \text{MSE}}{dc} = 0 \Rightarrow \left[ c_0 = \frac{R_x(\tau)}{R_x(0)} \right]$$

Con questo valore  $c_0$  si minimizza la MSE

$$m \text{MSE} = R_x(0) - 2 \frac{R_x^2(\tau)}{R_x(0)} + \frac{R_x^2(\tau)}{R_x(0)} = R_x(0) \left( 1 - \frac{R_x^2(\tau)}{R_x^2(0)} \right)$$

Per processi a media nulla:  $\frac{R_x(\tau)}{R_x(0)} = \rho_x(\tau)$

$$m \text{ MSE} = R_x(0) (1 - g_x(\tau))$$

