EE3004: Assignment 5 Report

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1 Install Instructions

We use *ipython* notebooks for the following assignment. The repository includes the relevant pip requirements.

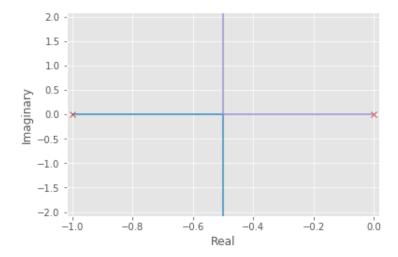
Ensure python 3.6 is installed, and run:

```
pip3 install -r requirements.txt
jupyter lab
```

This should open a browser window with the code snippets.

2 Question 1

```
In []: import numpy as np
      import math
      import scipy
      from control.matlab import *
      import matplotlib.pyplot as plt
      plt.style.use('ggplot')
      pie=np.pi
  We design for M_p = 17, T_s = 3
In []: real=-(4.0/Ts)
       imag=4*pie/(Ts*np.log(100/Mp))
       pole1=complex(real,imag)
       pole2=complex(real,imag)
       theta=180-np.angle(pole1,deg=True)
       wn=np.abs(pole1)
       print (pole1,pole2)
       print(theta,wn)
60.57563766283309 2.714031112851792
```



2.1 Part 3

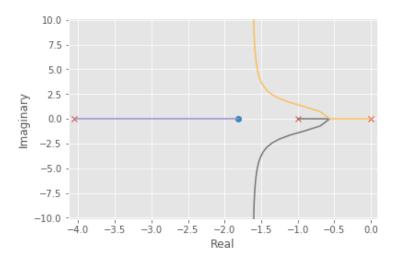
Clearly this doesnot lie on the root locus, hence we need a lead compensator. We calculate ϕ , the necessary angle difference. In addition we calculate θ the positive angle made by the dominant pole, by either $arccos(\zeta)$ or from the angle directly.

```
phi=-180+th
print(f"Phi {phi}")
theta=180-np.angle(pole1,deg=True)
print(f"Theta {theta}")
gamma=0.5*(180-theta-phi)
print(f"Gamma {gamma}")
alpha=np.sin(gamma*pie/180)*np.sin((theta+gamma+phi)*pie/180)/np.sin((theta+gamma)*pie/
print(f"Alpha {alpha}")
zc=wn*np.sin(gamma*pie/180)/np.sin((theta+gamma)*pie/180)
pc=zc/alpha
print (f"zc {zc} pc {pc}")
```

```
Phi 37.45060010083418
Theta 60.57563766283309
Gamma 40.986881118166366
Alpha 0.44819287447848166
zc 1.8169680716950904 pc 4.053986966681072
```

Out [49]:

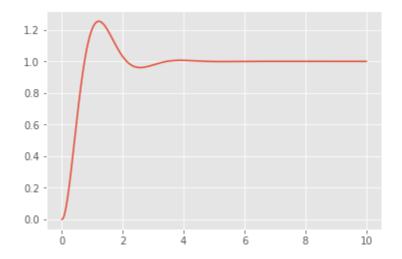
In []: rlocus(Hnew);



4.337684762035719

2.2 Plot with only lead compensator

Out[54]: [<matplotlib.lines.Line2D at 0x1c15f30b00>]



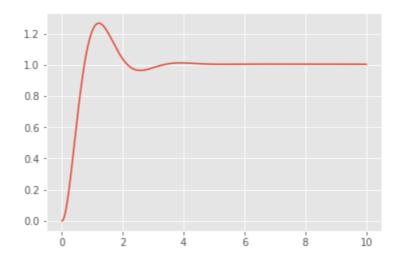
2.3
$$\alpha * \beta = 1$$

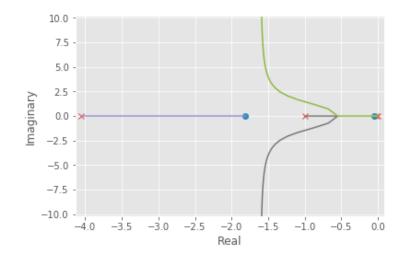
Out[56]:

```
9.678 s<sup>2</sup> + 18.07 s + 0.8792
-----s<sup>4</sup> + 5.076 s<sup>3</sup> + 4.167 s<sup>2</sup> + 0.09085 s
```


OS 26.16900489320202 % Tr 0.6706706706706707 Ts 2.042042042042

Kv: 20.0

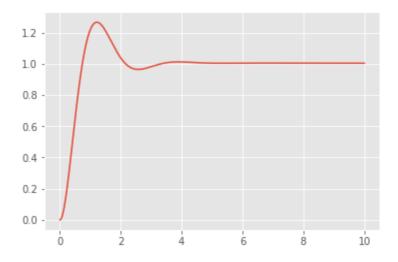




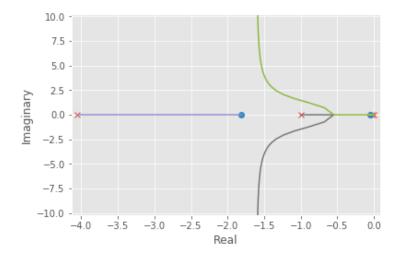
2.4 Part 4, 6 Independent α , β

We set $\beta = Kv_{desired}/Kv_{obtained}$, when obtained from the lead compensator.

We plot the step response and the root locus of the above system.



In []: rlocus(Hleadlag);



3 Question 3

We define a few helper functions ...

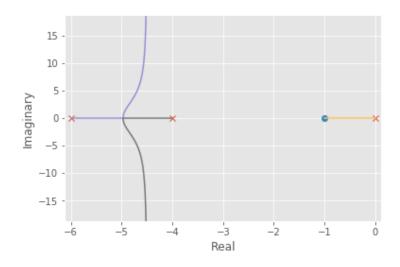
```
def polar(z):
              111
              Get polar representation
              a= z.real
              b= z.imag
              r = math.hypot(a,b)
              theta = math.atan2(b,a)
              return r, theta # use return instead of print.
          def get_poles(zeta,wn):
              Get poles f4om
              111
              return (complex(-zeta*wn, wn*np.sqrt(1-zeta**2)), complex(-zeta*wn, - wn*np.sqrt(1
          def get_alpha(gamma, theta, phi):
              Returns alpha from gamma.
              All angles in radians
              num = np.sin(gamma) * np.sin(theta + gamma + phi)
              denom = np.sin(theta + gamma) * np.sin(gamma + phi)
              return num/denom
          def get_zc(wn, gamma, theta):
              Get zc
              111
              num = wn* np.sin(gamma)
              denom = np.sin(theta + gamma)
              return num/denom
   We design for M_p = 5 %, and T_s is 0.02s, K_v is 1%. And we shall test if the design satisfies for
T_r.
In []: Mp, ts = 5, 0.02
          zeta,wn = get_params(Mp,ts)
          p1,p2 = get_poles(zeta,wn)
Zeta 0.6901067305598216 wn 289.8102440440741
Required dominant poles (-200.00000000000003+209.73787820249777j), (-200.0000000000003-209.7378
```

return zeta, wn

3.1 Part a

This is feasible if (-200.00 + / - 209.737j) lies on the root locus, as dominant poles.

In []: control.root_locus(sys);



Hence choice 1 is not feasible.

3.2 Part b

We first analyse the existent system, understand if the transient parameters need improvement or the steady state. We then desgin appropriately, with the dominant pole assumption.

```
With [U+FFFD] [U+FFFD] G_c(s) = K_c \alpha * (1+Ts)/(1+\alpha * Ts). Needed K_v is 100 (1% error). Current K_v is 10. Hence, K_c * \alpha = 10. We determine \theta, angle of compensation needed at (-200, +/- 209.73)
```

```
In []: polar(control.evalfr(sys, complex(200,-209.73)))[1]
```

Out[179]: 1.596108226418121
In []: np.pi - 1.596108226418121
Out[180]: 1.5454844271716721

We need -88.54 deg as compensation. Besides this, we need to meet:

- Angle sum = π
- Magnitude at roots = 1
- $K_c * \alpha = 10$

We treat this as a lead compensation case.

```
In []: phi = 85.97404456267036 * np.pi/180
          theta = np.arccos(zeta)
          gamma = (1.596108226418121 - theta)/2
          print(f"Gamma {gamma * 180/np.pi}")
          alpha = get_alpha(gamma = gamma, phi = phi, theta = theta)
          print(f"Alpha {alpha}")
          temp_sys = alpha * control.TransferFunction([T,1],[T*alpha, 1]) * sys
          Kc = 1/polar(control.evalfr(temp_sys,complex(-200,209.73)))[0]
          print(f"Kc {Kc}")
Gamma 22.54441156486525
Alpha 0.183975416417848
Kc 752.8216104925237
   To determine T, we use the magnitude condition.
In []: cur_mag = polar(control.evalfr(sys, complex(-200,209.73)))[0]
```

```
print(f"Current magnitude without compensator at requisite roots {cur_mag}")
Current magnitude without compensator at requisite roots 0.002919450489681809
Z_c 119.09350502020902
T 0.00839676353324482
Solved parameters :
Alpha 0.183975416417848
Kc 54.355088275967454
```

3.3 Part 3

We perform a grid search, while looking for the following:

• Angle sum = π

T 0.00839676353324482

- Magnitude at roots = 1
- $K_c * \alpha = 10$

We keep varying α and T. We try:

```
• grid search
```

• gradient descent

```
In []: aa = np.arange(0.1, 2.0, (0.2-0.1)/1000)
          tt = np.arange(0, 0.01, 0.01/100)
          11 = [] # loss array
          def loss1(a,t, pole = complex(-200,209.73)):
              Angle compensation
              diff = polar(1 + t*pole)[1] - polar(1 + a*t*pole)[1] - 1.5454844271716721
              diff = np.abs(diff)
              return diff
          def loss2(a,t, pole = complex(-200,209.73)):
              Magnitude compensation
              diff = polar(1 + t*pole)[0] / polar(1 + a*t*pole)[0] - 1/(10*cur_mag)
              return np.abs(diff)
          def loss(a,t, pole = complex(-200,209.73)):
              return loss1(a,t, pole) + 0*loss2(a,t,pole)
          1_min = None
          a_min = None
          t_min = None
          for a in aa:
              for t in tt:
                  l = loss(a,t)
                  if not l_min or (l_min > 1):
                      1_{min} = 1
                      a_min = a
                      t_min = t
                  11.append(loss)
          print(f"Loss {l_min} a_min {a_min} t_min {t_min}")
          print(f"Angle loss {loss1(a_min,t_min)}")
          #print(f"Mag loss {loss2(a_min, t_min)}")
Loss 1.884514472694221e-06 a_min 0.1669000000000194 t_min 0.0077
Angle loss 1.884514472694221e-06
Mag loss 32.09852385559125
```

3.4 Part 4,5 with Part 2

Clearly, we have used a lead compensator.

The compensated transfer function is:

```
Compensator

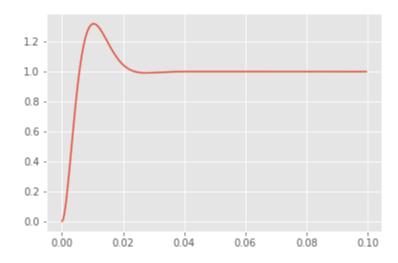
1.163 s + 138.5
-----

0.001545 s + 1

Final Sys
279.1 s^2 + 3.352e+04 s + 3.324e+04
------

0.001545 s^4 + 1.015 s^3 + 10.04 s^2 + 24 s
```

The transient responses associated with this are:



3.5 Part 4,5 with Part 3

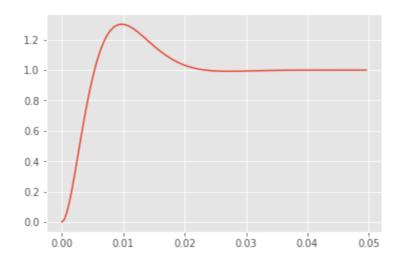
Clearly, we have used a lead compensator. The compensated transfer function is:

```
Compensator
1.224 s + 159
-----
0.001285 s + 1

Final Sys
293.8 s^2 + 3.845e+04 s + 3.816e+04
```

```
0.001285 s<sup>4</sup> + 1.013 s<sup>3</sup> + 10.03 s<sup>2</sup> + 24 s
```

In []: tt, yy = control.step_response(control.feedback(comp_sys2), T = np.arange(0,0.05,0.04/100) and the control of the cont



Which has a much smaller T_s . We also note that in the above two cases, no lag compensation is needed, since the K_v obtained is good enough.

4 Question 4

```
In [2]: import numpy as np
    import math
    import scipy
    from control.matlab import *
    import control as cont
    import matplotlib.pyplot as plt
    pie=np.pi
```

Stepinfo function which gives all the useful information like overshoot, rise time and settle time from the response function

Calculating tau and wn from formulas

$$M = 100e^{\frac{-\zeta\pi}{\sqrt{1-\zeta^2}}}$$
$$ts = \frac{4}{\zeta\omega_n}$$

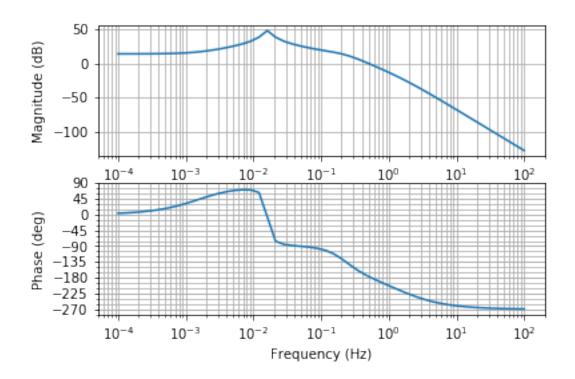
Calculating gamma and omega from the formulas

$$\gamma = \tan^{-1} \frac{2\zeta}{\sqrt{\sqrt{1 + 4\zeta^4} - 2\zeta^2}}$$

$$\omega_{gc} = \omega_n \sqrt{\sqrt{1 + 4\zeta^4} - 2\zeta^2}$$

4.1 Bode Plot of H

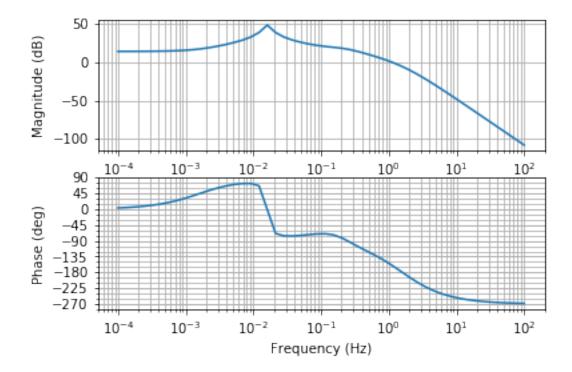
In [27]: bode(H);



```
In [28]: [gm,pm,wg,wm]=margin(H)
         print (f"gain-margin {gm} phase-margin {pm} \ngain cross over freq {wg} phase cross over
         phi=gamma-pm
         wm=wm
         print(f"\nPhi {phi} w_m {wm}")
         alpha=(1-np.sin(phi*pie/180))/(1+np.sin(phi*pie/180))
         print(f"\nAlpha {alpha}")
         T=1/np.sqrt(alpha)/wm
         print(f"T {T}")
         Gc=tf([T,1],[alpha*T,1])
         Hnew=H*Gc
         print(f"\nHnew {Hnew}")
gain-margin 1.214662566961802 phase-margin 3.7857102810716015
gain cross over freq 3.4359780521729917 phase cross over freq 3.1344441272838512
Phi 54.80735798389206 w_m 3.1344441272838512
Alpha 0.1005828635112053
T 1.0059524889079905
```

```
100.6 \text{ s}^3 + 201.6 \text{ s}^2 + 102 \text{ s} + 1
```

 $0.1012 \text{ s}^{\circ}6 + 2.216 \text{ s}^{\circ}5 + 14.27 \text{ s}^{\circ}4 + 24.33 \text{ s}^{\circ}3 + 20.62 \text{ s}^{\circ}2 + 0.6426 \text{ s} + 0.202$



Phase margin 20.823595702607207 Gain margin 6.246114286537816 Phi new 54.80735798389206

Since there is a huge difference between the phase margin and the desired value, we increment phi by 60 degrees

```
print(f"Alpha {alpha_one}")

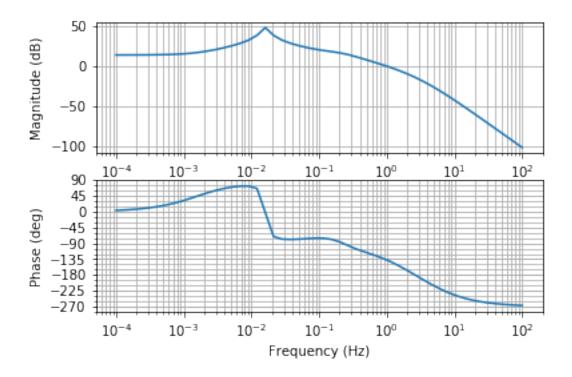
mag10 = 20*np.log10(1/np.sqrt(alpha_one))
print(f"decibels magnitude {mag10}")

mag_one=np.sqrt(alpha_one)
print(f"Magnitude descent needed to offset {mag_one}")

Phi altered 114.80735798389206
Alpha 0.04836989761311832
decibels magnitude 13.154248319025266
Magnitude descent needed to offset 0.21993157484344608
```

To calculate the frequency at which magnitude is square root of alpha, we find the gain crossover frequency of the transfer function H/square root of alpha

```
In [40]: [gm1,pm1,wm1,wg1]=margin(H/mag_one)
         print(f"Cross over {wg1}")
         T_1=1/np.sqrt(alpha_one)/wg1
         print(f"T_1 {T_1}")
         Gc_{one=tf([T_1,1],[alpha_one*T_1,1])}
         Hnew_one=Gc_one*H
         print(Hnew_one)
Cross over 6.246114286537809
T_1 0.7279515743095443
                      72.8 \text{ s}^3 + 173.5 \text{ s}^2 + 101.7 \text{ s} + 1
______
0.03521 \text{ s}^6 + 1.423 \text{ s}^5 + 12.8 \text{ s}^4 + 22.97 \text{ s}^3 + 20.58 \text{ s}^2 + 0.6293 \text{ s} + 0.202
In [41]: bode(Hnew_one)
         plt.show()
         [gm2,pm2,wm2,wg2]=margin(Hnew_one)
         print(f"Phase margin {pm2} Gain margin {wg2}")
```



Phase margin 42.84878119640814 Gain margin 6.246114286537816

4.2 Step Response of Closed Loop System

Overshoot 29.495260712314074 % Time rise 0.266026602660 Time settle 8.58685868586

