



## BigCat Wireless - EC401 Assignment 3

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1. From the constellation plot shown in figure below, answer the following questions?

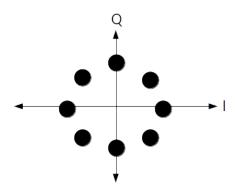


Figure 1: Constellation plot which is to be identified

- 1. What type of modulation does this represent? The modulation is 8-PSK
- 2. How many symbols (M) are represented? 8 symbols
- 3. How many bits per symbol (N) are used? 3 bits
- 4. What's the bit rate when band rate is 10,000 symbols/second? 30,000 bits/second (30kbps)
- 5. Compare the noise performance of this modulation with 16-QAM, which is better?
  - 8-PSK should in general be less susceptible to noise because the symbols would be spread further apart. Approximately, the minimum distance for constellation points for 8-PSK is  $1.325\sqrt{E_b}$  which is slightly greater than for 16-QAM which is  $1.265\sqrt{E_b}$ .





- $6. \ \ Compare \ the \ spectral \ efficiency \ of \ this \ modulation \ with \ QPSK, \ which \ is \ better?$ 
  - 8-PSK has better spectral efficiency since it uses  $2R_b/3$  bandwidth which is less than the bandwidth  $R_b$  used by QPSK
- 2. Solve the IQ demodulator equation showing how I and Q components are received back.

Composite Input signal:

$$X(t) = I * cos(2\pi f_c t) - Q * sin(2\pi f_c t)$$

Inphase Component (multipled with LO signal):

$$\begin{split} X_i(t) &= (I*\cos(2\pi f_c t) - Q*\sin(2\pi f_c t))*\cos(2\pi f_c t) \\ &= \frac{I}{2} + \frac{I}{2}*\cos(2\pi f_c t) - \frac{Q}{2}*\sin(2\pi f_c t) \end{split}$$

Quadrature Component (multipled with  $\pi/2$  shifted LO signal):

$$\begin{split} X_q(t) &= (I * cos(2\pi f_c t) - Q * sin(2\pi f_c t)) * - sin(2\pi f_c t) \\ &= \frac{Q}{2} - \frac{Q}{2} * cos(2\pi f_c t) - \frac{I}{2} * sin(2\pi f_c t) \end{split}$$

After LPF (sin,cos terms with  $f_c$  are cancelled ):

$$X_i(t) = \frac{I}{2}$$
$$X_1(t) = \frac{Q}{2}$$

- 3. Calculate maximum theoretical spectral efficiency for:
  - 1.  $\pi/4$  QPSK:

Maximum theoretical spectral efficiency = 2 bits/s/Hz.

2. 256-QAM:

Maximum theoretical spectral efficiency = 8 bits/s/Hz.



3



4. Create a transmitter and receiver architecture.
Use the following parameters:
Input bit rate 1 Mbps
Modulation method: QPSK or 16-QAM
RRC roll-off factor 0.2
RRC output samples per symbol: 8
Carrier frequency 2.5MHz
Use AWGN channel with SNR =10dB
Equivalent settings in receiver side

```
% passband_modulation.m
    close all;
    bitrate = 1e06;
   Fc = 2.5e06;
   Fs = 8 * bitrate;
    M = 4; % Modulation order
    k = log2(M); % Bits/symbol
    n = 500; % Transmitted bits
    sps = 8; % Samples per symbol
10
    EbNo = 10; \% Eb/No (dB)
11
12
    span = 8; % Filter span in symbols
13
    rolloff = 0.20; % Rolloff factor
14
15
    txfilter = comm.RaisedCosineTransmitFilter('RolloffFactor',
    rolloff, 'FilterSpanInSymbols', span, ...
17
      'OutputSamplesPerSymbol', sps);
18
19
    rxfilter = comm.RaisedCosineReceiveFilter('RolloffFactor',
    rolloff, 'FilterSpanInSymbols', span, ...
      'InputSamplesPerSymbol', sps, ...
22
      'DecimationFactor', sps, 'Gain', 2);
23
25
    ri = comm.internal.RandomIntegerGenerator('SetSize', M, ...
      'SampleTime', 1 / bitrate, 'SamplesPerFrame', n);
26
27
    tx_lo = dsp.SineWave(1, Fc, 0, 'ComplexOutput', true, ...
      'SampleRate', Fs, 'SamplesPerFrame', sps * n);
29
30
    rx_lo = dsp.SineWave(1, Fc, 0, 'ComplexOutput', true, ...
31
      'SampleRate', Fs, 'SamplesPerFrame', sps * n);
32
33
    filtDelay = (txfilter.FilterSpanInSymbols + ...
34
      rxfilter.FilterSpanInSymbols) / 2;
35
36
    errorRate = comm.ErrorRate('ReceiveDelay', filtDelay);
37
    delay = dsp.Delay(8);
    evm = comm.EVM();
```





```
40
    biterr = 0;
    totalbits = 0;
42
    rmsEVM = 0;
43
    tx_scope = dsp.SpectrumAnalyzer('SpectrumType', "Power Density",
      'SampleRate', Fs, 'FrequencyResolutionMethod', ...
      "WindowLength", 'WindowLength', 800, ...
46
      'PlotAsTwoSidedSpectrum', false);
47
48
    rx_scope = dsp.SpectrumAnalyzer('SpectrumType', "Power", ...
49
      'SampleRate', Fs, 'FrequencyResolutionMethod', ...
50
      "WindowLength", 'WindowLength', 1024);
51
    txSigall = [];
53
    rxSigall = [];
54
55
    tx_const_diag = scatterplot(ri());
    rx_const_diag = scatterplot(ri());
57
58
    for idx = 1:20
59
      dataIn = ri();
      modSig = pskmod(dataIn, 4, pi / 4);
61
      txFilterSig = txfilter(modSig);
62
      tx_carrier = tx_lo();
      txSig = real(txFilterSig .* tx_carrier);
      txSigall = txSig;
65
66
      SNR = EbNo + 10 * log10(k) - 10 * log10(sps);
      noisySig = awgn(txSig, SNR, 'measured');
69
      rx_wave = conj(rx_lo());
70
      rxSig = noisySig .* rx_wave;
71
      rxSigall = rxSig;
72
73
      rxFilterSig = rxfilter(rxSig);
74
      dataOut = pskdemod(rxFilterSig, 4, pi / 4);
76
77
      errStat = errorRate(dataIn, dataOut);
78
      biterr = biterr + errStat(2);
      totalbits = totalbits + errStat(3);
80
      rmsEVM = (rmsEVM + evm(delay(modSig), rxFilterSig)) / 2;
      scatterplot(modSig, 1, 0, 'y.', tx_const_diag);
      scatterplot(rxFilterSig, 1, 0, 'y.', rx_const_diag);
85
      tx_scope(txSigall);
86
      rx_scope(rxSigall);
    end
```



5



```
fprintf('\nBit Errors = %d', biterr);
fprintf('\nBits Transmitted = %d\n', totalbits);
fprintf('\nBER = %5.2e\n', biterr / totalbits);
fprintf('\nRMS EVM = %.2f %% \n', rmsEVM);

release(tx_scope);
release(rx_scope);

Sample output:
Bit Errors = 157 Bits Transmitted = 104840
BER = 1.50e-03
```

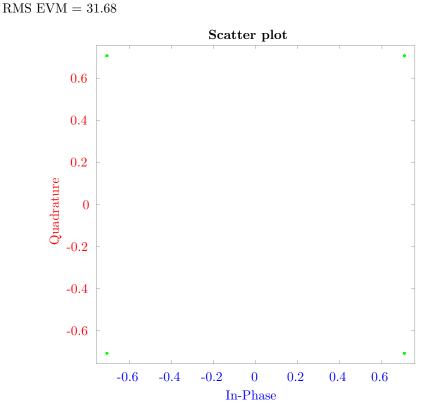


Figure 2: Transmitted constellation for QPSK





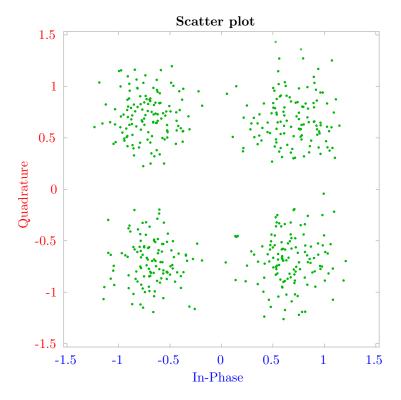


Figure 3: Received constellation for QPSK  $\,$ 

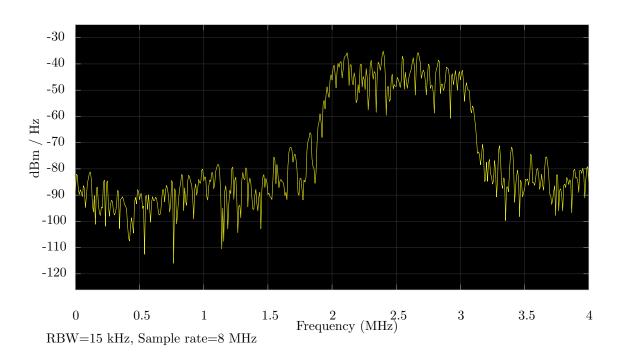


Figure 4: Transmitted spectrum for QPSK  $\,$ 



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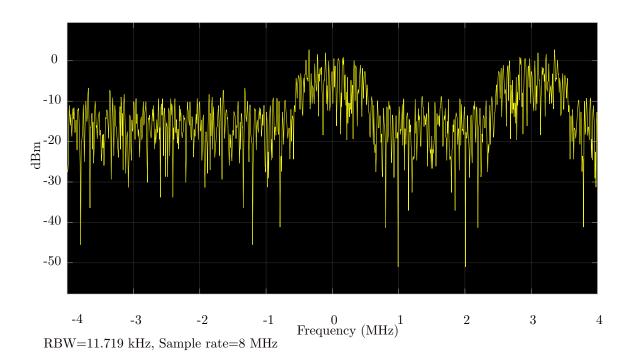


Figure 5: Received spectrum for QPSK  $\,$