

3.103

題目：Use MATLAB's f f t command to repeat Problem 3.48.

Problem 3.48:

(a)

$$x[n] = \cos\left(\frac{6\pi}{17}n + \frac{\pi}{3}\right)$$

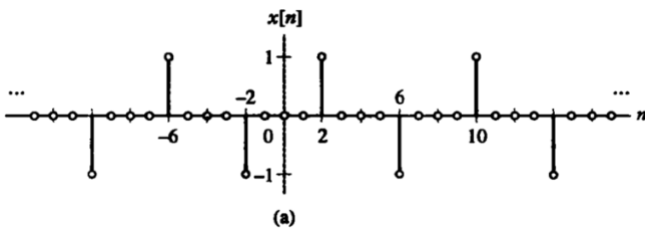
(b)

$$x[n] = 2 \sin\left(\frac{14\pi}{19}n\right) + \cos\left(\frac{10\pi}{19}n\right) + 1$$

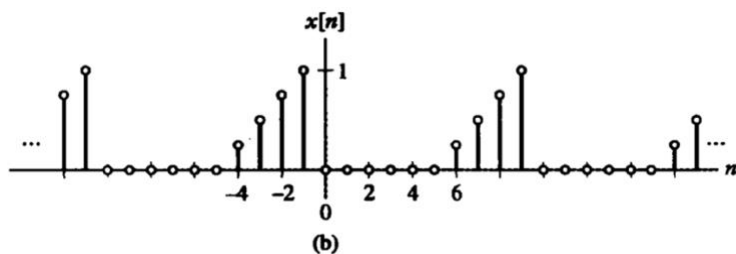
(c)

$$x[n] = \sum_{m=-\infty}^{\infty} (-1)^m (\delta[n - 2m] + \delta[n + 3m])$$

(d)



(e)



作法：

(a)

%3.103a

```
x = cos([0:16]*6*pi/17 + pi/3); %x[n]
```

```
X = fft(x)/17; %對 x[n]做 fourier 轉換，N=17（一個週期）
```

```
stem(0:16,abs(X));
```

```
ylabel('|Xa[k]|');
```

```
stem(0:16,rad2deg(angle(X)));
```

```
ylabel('Xa[k] deg');
```

(b)

%3.103b

```
x = 2*sin([0:18]*14*pi/19)+cos([0:18]*10*pi/19)+ones(1,19); %x[n]
```

```
X = fft(x)/19; %對 x[n]做 fourier 轉換，N=19（一個週期）
```

```
stem(0:18,abs(X));
```

```
ylabel('|Xb[k]|');
```

```
stem(0:18,rad2deg(angle(X)));
```

```
ylabel('Xb[k] deg');
```

(c)

%3.103c

```
x=[2,0,-1,-1,1,0,0,0,1,-1,-1,0]; %x[n]
```

```
X = fft(x)/12; %對 x[n]做 fourier 轉換，N=12（一個週期）
```

```
stem(0:11,abs(X));
```

```
ylabel('|Xc[k]|');
```

```
stem(0:11,rad2deg(angle(X)));
```

```
ylabel('Xc[k] deg');
```

(d)

%3.103d

```
x=[0,0,1,0,0,0,-1,0]; %x[n]
```

```
X = fft(x)/8; %對 x[n]做 fourier 轉換，N=8（一個週期）
```

```
stem(0:7,abs(X));
```

```
ylabel('|Xd[k]|');
```

```
stem(0:7,rad2deg(angle(X)));
```

```
ylabel('Xd[k] deg');
```

(e)

%3.103e

```
x=[0,0,0,0,0,0,1/4,1/2,3/4,1]; %x[n]
```

```
X = fft(x)/10; %對 x[n]做 fourier 轉換，N=10（一個週期）
```

```
stem(0:9,abs(X));
```

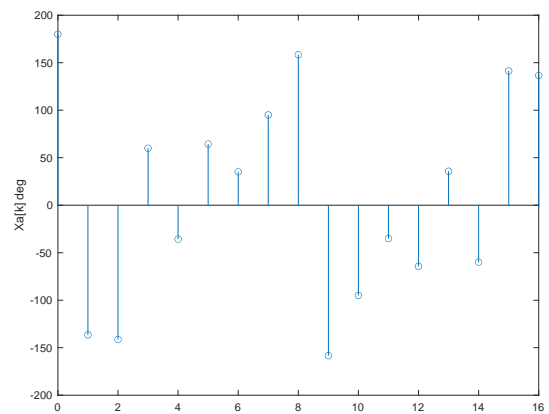
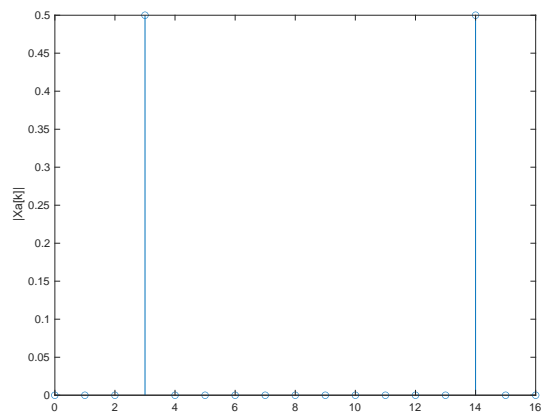
```
ylabel('|Xe[k]|');
```

```
stem(0:9,rad2deg(angle(X)));
```

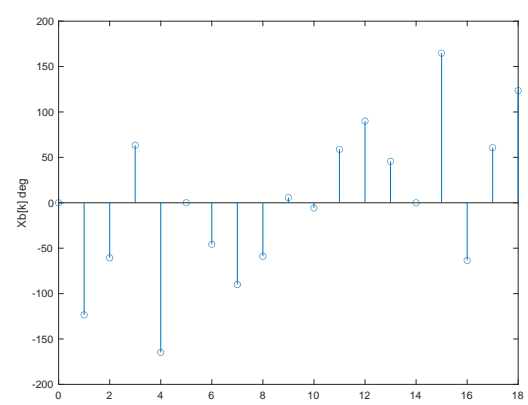
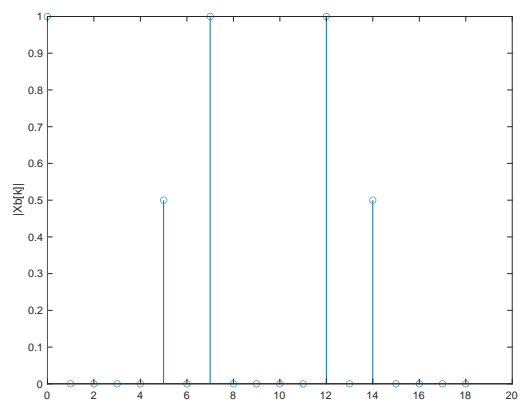
```
ylabel('Xe[k] deg');
```

結果：

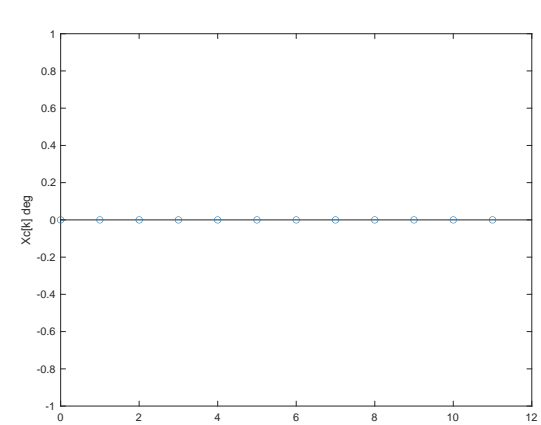
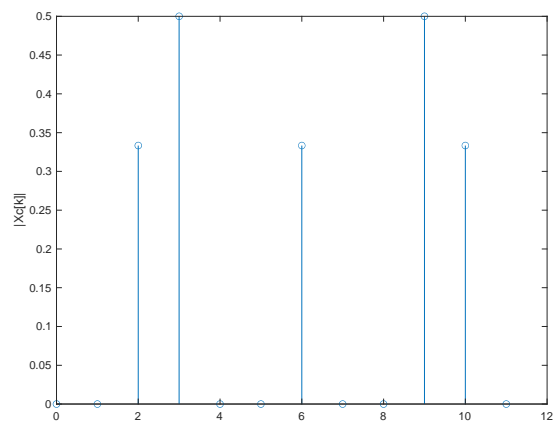
(a)



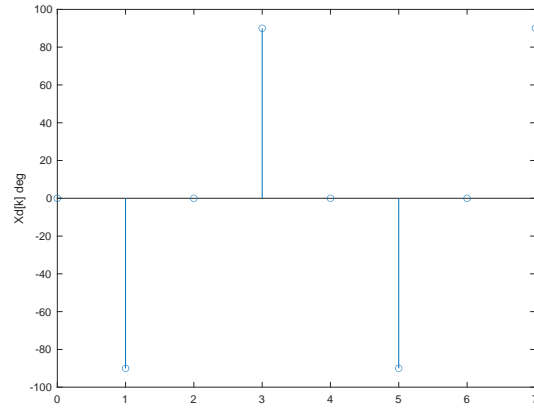
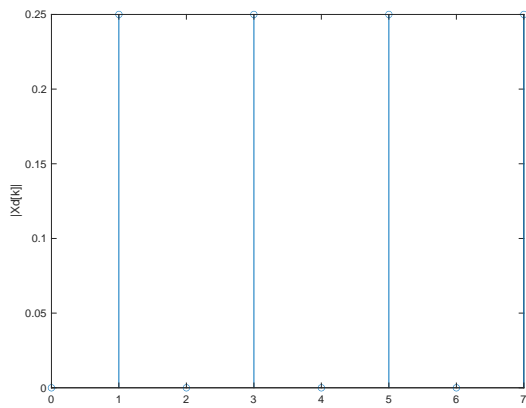
(b)



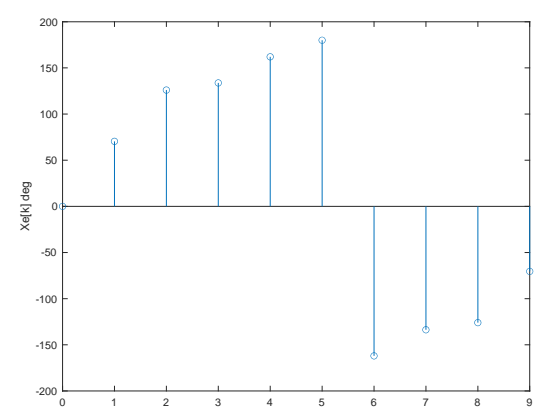
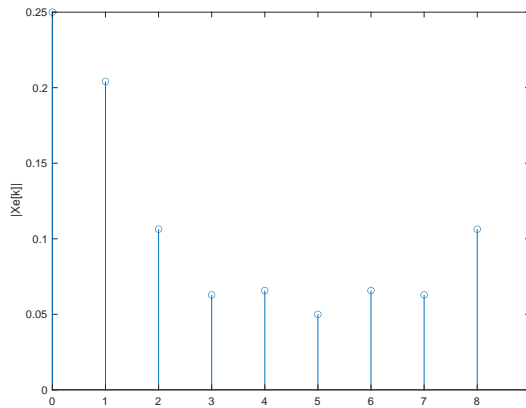
(c)



(d)



(e)



3.107

題目：Let $x(t)$ be the triangular wave depicted in Fig. P3.107.

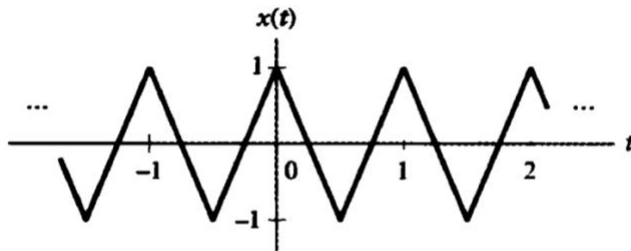


FIGURE P3.107

(a) Find the FS coefficients $X[k]$.

(b) Show that the FS representation for $x(t)$ can be expressed in the form

$$x(t) = \sum_{k=0}^{\infty} B[k] \cos(k\omega_0 t).$$

(c) Define the J -term partial-sum approximation to $x(t)$ as

$$\hat{x}_J(t) = \sum_{k=0}^J B[k] \cos(k\omega_0 t).$$

Use MATLAB to evaluate and plot one period of the j th term in this sum and $\hat{x}_j(t)$ for $j=1,3,7,29$, and 99.

作法：

前面的(a)、(b)兩題如下：

(a)

FS coefficient:

$$X[k] = \frac{1}{T} \int_0^T x(t) e^{-j\frac{2\pi}{T}kt} dt, T=1$$

$$x(t) = 2y(t + \frac{1}{4}) - 1$$

$$Y[k] = \begin{cases} \frac{1}{2}, & k=0 \\ \frac{2\sin(\frac{k\pi}{2})}{jk^2\pi^2}, & k \neq 0 \end{cases}$$

$$x(t-t_0) \xleftrightarrow{\text{FS } \omega_0} e^{-jk\omega_0 t} X[k]$$

$$X[k] = 2e^{jk2\pi\frac{1}{4}} Y[k] - \delta[k]$$

$$X[k] = \begin{cases} -\frac{1}{2}, & k=0 \\ \frac{4\sin(\frac{k\pi}{2})}{k^2\pi^2} e^{j\frac{\pi}{2}(k-1)}, & k \neq 0 \end{cases}$$

(b) $\therefore x(t)$ is real and even

$\therefore X[k]$ also real and even with $X[k] = X[-k]$

$$\begin{aligned} x(t) &= \sum_{k=-\infty}^{\infty} X[k] e^{jk\omega_0 t} + \sum_{k=-\infty}^{-1} X[k] e^{jk\omega_0 t} + X[0] \\ &= \sum_{k=1}^{\infty} X[k] e^{jk\omega_0 t} + \sum_{k=1}^{\infty} X[k] e^{-jk\omega_0 t} + X[0] \end{aligned}$$

$$\therefore X[k] = X[-k]$$

$$= \sum_{k=0}^{\infty} X[k] (e^{jk\omega_0 t} + e^{-jk\omega_0 t}) + X[0]$$

$$= \sum_{k=0}^{\infty} B[k] \cos(k\omega_0 t)$$

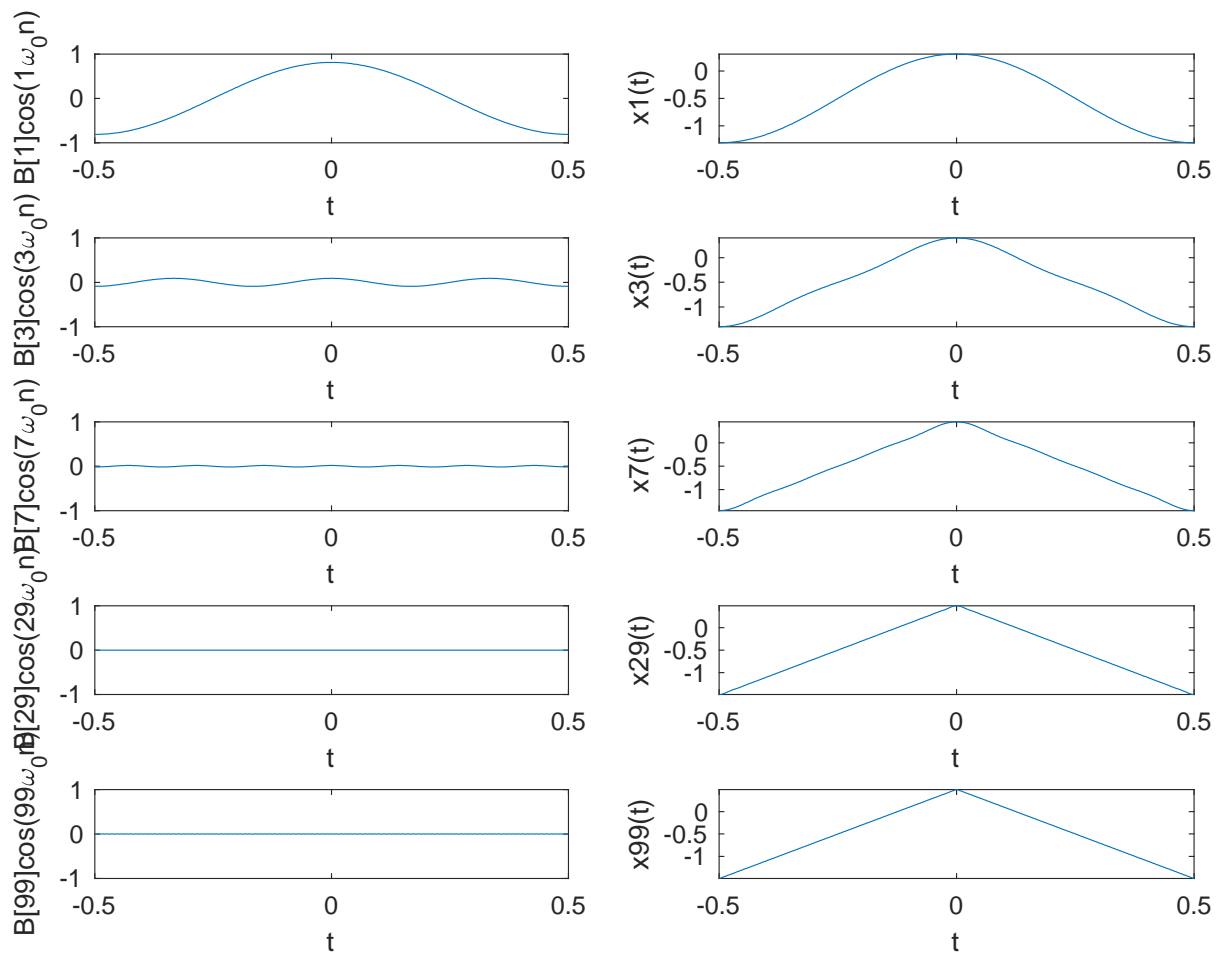
$$\text{where } B[k] = \begin{cases} X[0], & k=0 \\ 2X[k], & k \neq 0 \end{cases}$$

(c) 將(a)、(b)的 $X[k]$ 、 $x[t]$ 式子寫入 MATLAB，並使用迴圈求出 partial sum。

%3.107

```
k = 1:1:99;
t = linspace(-0.5,0.5,400);
B = zeros(1,100);
B(1) = -0.5; % coefficient for k=0
B(2:1:100) = 8*sin(k*pi/2).exp(1i*pi*(k-1)/2)./(pi*k).^2; % coefficient for k!=0
xJhat(1,:) = B(1)*cos(0*2*pi*t); % term in sum for k=0
% accumulate partial sum
for k = 2:1:100
    Bcos(k,:) = B(k)*cos((k-1)*2*pi*t);
    xJhat(k,:) = xJhat(k-1,:)+B(k)*cos((k-1)*2*pi*t);
end
j = [1 3 7 29 99]+1;
for i = 1:1:5
    subplot(5,2,2*i-1);plot(t,Bcos(j(i),:));
    ylabel(['B' num2str(j(i)-1) 'cos(' num2str(j(i)-1) '\omega_0n)']);
    xlabel('t');axis([-0.5 0.5 -1 1]);
    subplot(5,2,2*i);plot(t,xJhat(j(i),:));
    ylabel(['x' num2str(j(i)-1) '(t)']);
    xlabel
```

結果：



3.109

題目：Use MATLAB to repeat Example 3.15, with the following values for the time constant:

- (a) $RC = 0.01$ s.
- (b) $RC = 0.1$ s.
- (c) $RC = 1$ s.

Example 3.15:

Let us find the FS representation for the output $y(t)$ of the RC circuit depicted in Fig. 3.2 in response to the square-wave input depicted in Fig. 3.21, assuming that $T_0/T = 1/4$, $T = 1$ s, and $RC = 0.1$ s.

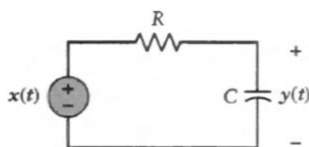


FIGURE 3.2 RC circuit for Example 3.1.

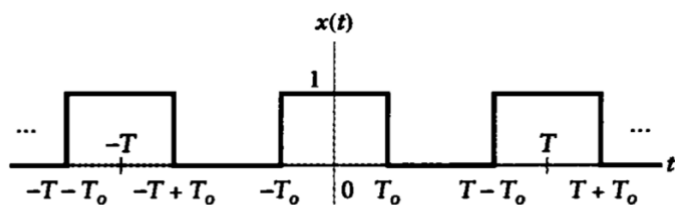


FIGURE 3.21 Square wave for Example 3.13.

作法：

Solution: If the input to an LTI system is expressed as a weighted sum of sinusoids, then the output is also a weighted sum of sinusoids. As shown in Section 3.2, the k th weight in the output sum is given by the product of the k th weight in the input sum and the system frequency response evaluated at the k th sinusoid's frequency. Hence, if

$$x(t) = \sum_{k=-\infty}^{\infty} X[k]e^{jk\omega_o t},$$

then the output is

$$y(t) = \sum_{k=-\infty}^{\infty} H(jk\omega_o)X[k]e^{jk\omega_o t},$$

where $H(j\omega)$ is the frequency response of the system. Thus,

$$y(t) \xleftrightarrow{FS; \omega_o} Y[k] = H(jk\omega_o)X[k].$$

In Example 3.1, the frequency response of the RC circuit was determined to be

$$H(j\omega) = \frac{1/RC}{j\omega + 1/RC}.$$

The FS coefficients of the square wave were given in Eq. (3.23). Substituting for $H(jk\omega_o)$ with $RC = 0.1$ s and $\omega_o = 2\pi$, and using $T_o/T = 1/4$, gives

$$Y[k] = \frac{10}{j2\pi k + 10} \frac{\sin(k\pi/2)}{k\pi}.$$

The $Y[k]$ go to zero in proportion to $1/k^2$ as k increases, so a reasonably accurate representation of $y(t)$ may be determined with the use of a modest number of terms in the FS. We plot the magnitude and phase spectra of $X[k]$ and $Y[k]$, and we determine $y(t)$ using the approximation

$$y(t) \approx \sum_{k=-100}^{100} Y[k]e^{jk\omega_o t}. \quad (3.30)$$

由以上課本的 Example 3.15 的 solution，可知

$$Y[k] = \frac{\frac{1}{RC}}{j2\pi k + \frac{1}{RC}} \frac{\sin(\frac{k\pi}{2})}{k\pi}$$

$$y(t) = \sum_{k=-100}^{100} Y[k]e^{j2\pi kt}$$

將以上兩式子代入 MATLAB，並使用迴圈求出 partial sum。

(a)

```
%3.109a
k = 1:1:100;
t = linspace(-0.5,0.5,400);
Y = zeros(1,100);
RC=0.01;
Y(1)=0;
Y(2:1:101) = (1/RC)./(1i*pi*k*2+(1/RC)).*sin(k*pi/2)./(k*pi); % % coefficient for k
yJhat(1,:) = Y(1).*exp(1i*pi*2*0*t); % term in sum for k=0
% accumulate partial sum
for k = 2:1:100
    Yexp(k,:) = Y(k).*exp(1i*pi*(k-1)*2*t);
    yJhat(k,:) = yJhat(k-1,:)+Y(k).*exp(1i*pi*(k-1)*2*t);
end
j = 100;
plot(t,yJhat(j,:));
ylabel(['y(t)']);
xlabel('t');
```

(b)

```
%3.109b
k = 1:1:100;
t = linspace(-0.5,0.5,400);
Y = zeros(1,100);
RC=0.1;
Y(1)=0;
```

```

Y(2:1:101) = (1/RC)./(1i*pi*k*2+(1/RC)).*sin(k*pi/2)./(k*pi); % % coefficient for k
yJhat(1,:) = Y(1).*exp(1i*pi*2*0*t); % term in sum for k=0
% accumulate partial sum
for k = 2:1:100
    Yexp(k,:) = Y(k).*exp(1i*pi*(k-1)*2*t);
    yJhat(k,:) = yJhat(k-1,:)+Y(k).*exp(1i*pi*(k-1)*2*t);
end
j = 100;
plot(t,yJhat(j,:));
ylabel(['y(t)']);
xlabel('t');

```

(c)

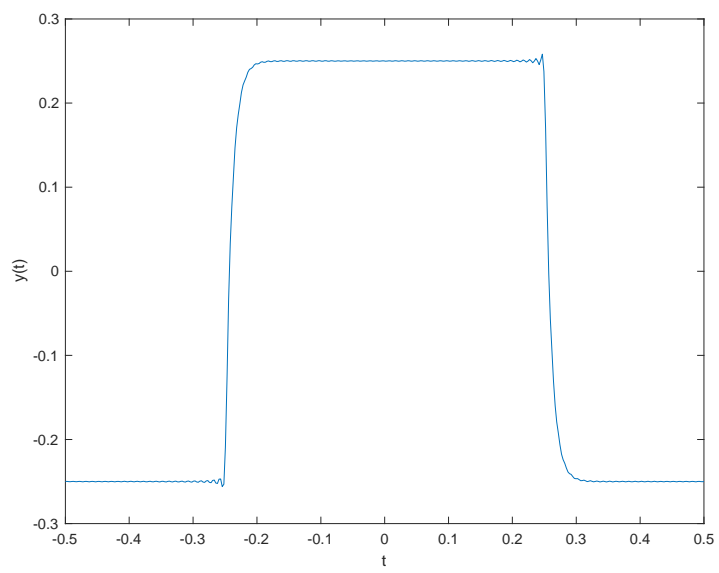
```

%3.109c
k = 1:1:100;
t = linspace(-0.5,0.5,400);
Y = zeros(1,100);
RC=1;
Y(1)=0;
Y(2:1:101) = (1/RC)./(1i*pi*k*2+(1/RC)).*sin(k*pi/2)./(k*pi); % % coefficient for k
yJhat(1,:) = Y(1).*exp(1i*pi*2*0*t); % term in sum for k=0
% accumulate partial sum
for k = 2:1:100
    Yexp(k,:) = Y(k).*exp(1i*pi*(k-1)*2*t);
    yJhat(k,:) = yJhat(k-1,:)+Y(k).*exp(1i*pi*(k-1)*2*t);
end
j = 100;
plot(t,yJhat(j,:));
ylabel(['y(t)']);
xlabel('t');

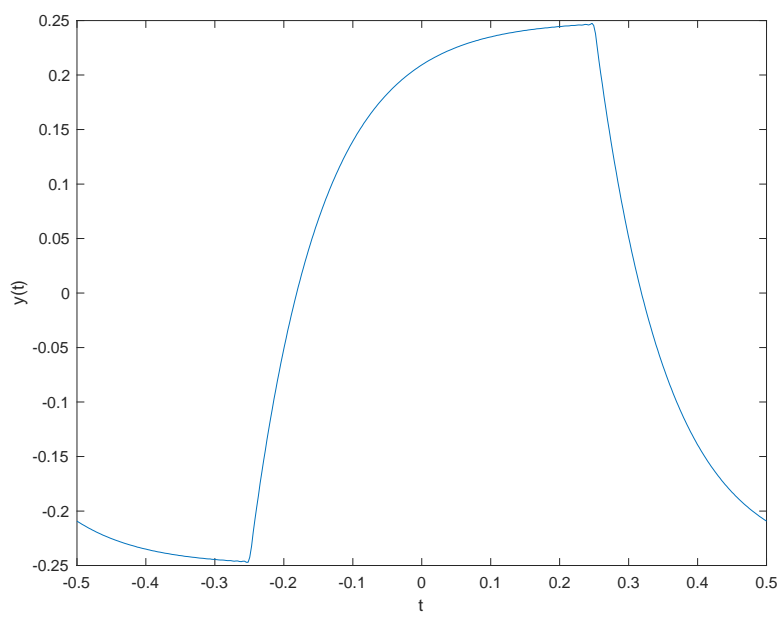
```

結果：

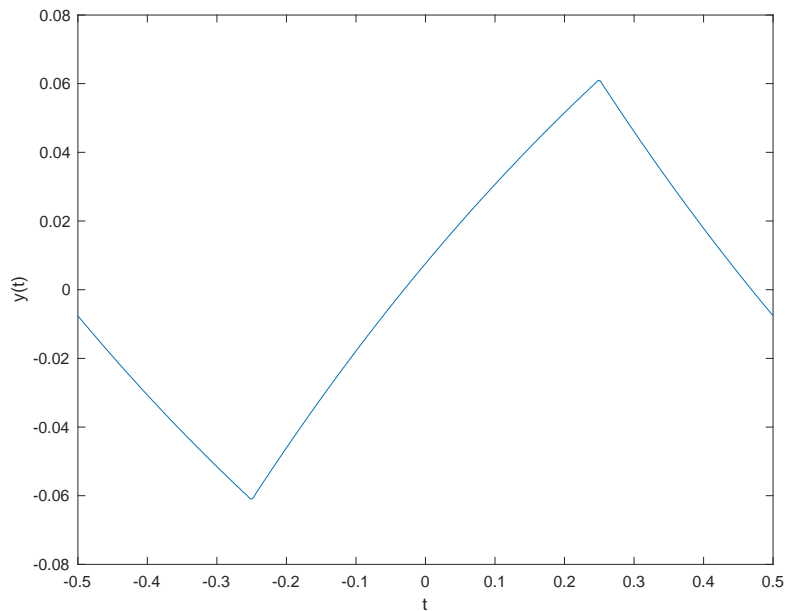
(a)



(b)



(c)



3.111

題目：This experiment builds on Problem 3.72.

(a) Graph the magnitude response of the circuit depicted in Fig. P3.72. Use 501 logarithmically spaced frequencies from 1 rad/s to 10^5 rad/s. You can generate N logarithmically spaced values between 10^{d1} and 10^{d2} by using the MATLAB command `logspace(d1,d2,N)`.

(i) Assume that $L = 10$ mH.

(ii) Assume that $L = 4$ mH.

(b) Determine and plot the output, using at least 99 harmonics in a truncated FS expansion, if the input is the square wave depicted in Fig. 3.21 with $T = 2\pi \times 10^{-3}$ and $T_0 = (\pi/2) \times 10^{-3}$

(i) Assume that $L = 10$ mH.

(ii) Assume that $L = 4$ mH.

Problem 3.7:

3.72 Consider the RLC circuit depicted in Fig. P3.72 with input $x(t)$ and output $y(t)$.

- Write a differential-equation description for this system and find the frequency response. Characterize the system as a filter.
- Determine and plot the output if the input is the square wave depicted in Fig. 3.21 with $T = 2\pi \times 10^{-3}$ and $T_o = (\pi/2) \times 10^{-3}$, assuming that $L = 10$ mH.

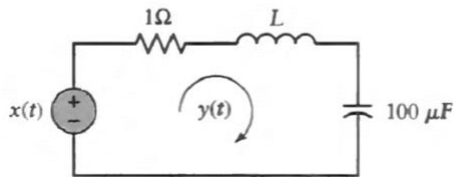


FIGURE P3.72

Fig. 3.21:

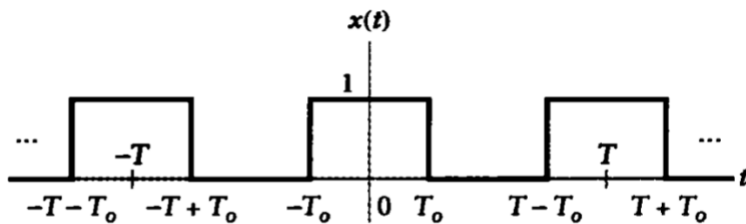


FIGURE 3.21 Square wave for Example 3.13.

作法：

(a) 首先先找出 $H(j\omega)$ 再將其代入 MATLAB，再畫出 magnitude response。

$$\begin{aligned}
 x(t) &= y(t) + L \frac{d}{dt} y(t) + \frac{1}{C} \int_{-\infty}^{\infty} y(\tau) d\tau \\
 \frac{d}{dt} x(t) &= \frac{d}{dt} y(t) + L \frac{d^2}{dt^2} y(t) + \frac{1}{C} y(t) \\
 H(j\omega) &= \frac{j\omega}{\frac{1}{C} + j\omega + L(j\omega)^2}
 \end{aligned}$$

(i) 10mh

%3.111a1

f= logspace(0, 5, 501);

```

C= 100*(10.^(-6));
L= 10*(10.^(-3));
w= 2*pi*f;
s= 1i*w;
H= s./ (s.^2*L + s + 1/C); %transfer function of the bandpass filter
magnitude = 20*log10(abs(H));
plot(w, magnitude);
xlabel('w(rad/s)');
ylabel('20log(|H(jw)|)');
title('L=10mH magnitude response in dB');

```

(ii)4mh

```

%3.111a2
f= logspace(0, 5, 501);
C= 100*(10.^(-6));
L= 4*(10.^(-3));
w= 2*pi*f;
s= 1i*w;
H= s./ (s.^2*L + s + 1/C); %transfer function of the bandpass filter
magnitude = 20*log10(abs(H));
plot(w, magnitude);
xlabel('w(rad/s)');
ylabel('20log(|H(jw)|)');
title('L=4mH magnitude response in dB');

```

(b) 首先先找出 $Y[k]$ ，再找出 $y[t]$ ，再將其代入 MATLAB，再畫出 output。

$$H[k] = \frac{j k \omega_0}{\frac{1}{C} + j k \omega_0 + L(j k \omega_0)^2}$$

$$X[k] = \frac{2 \sin(k \omega_0 T_0)}{T k \cdot \omega_0}$$

$$Y[k] = \frac{2 \sin(k \omega_0 T_0)}{T k \omega_0} \frac{j k \omega_0}{\frac{1}{C} + j k \omega_0 + L(j k \omega_0)^2}$$

$$y(t) = \sum_{k=-k_0}^{k_0} Y[k] e^{j \omega_0 k t}$$

$$T_0 = \pi/2 \times 10^{-3} \quad k_0 \geq 49$$

$$T = 2 \times \pi \times 10^{-3} \quad |t| \leq \pi \times 10^{-3}$$

$$\omega_0 = 2\pi/T$$

(i) 10mh

```
%3.111b1
```

```
k = 1:1:50;
```

```
C = 100*(10.^(-6));
```

```
L = 10*(10.^(-3));
```

```
w = 1000;
```

```
s = 1i*w*k;
```

```
H = s./ (s.^2*L + s + 1/C); %transfer function of the bandpass filter
```

```
T0 = pi./2000;
```

```
T = pi./500;
```

```
t = linspace(-pi*10.^(-3),pi*10.^(-3),400);
```

```
Y = zeros(1,50);
```

```
Y(1)=0;
```



```
Y(2:1:51) = H.*(sin(k*w*T0)*2)./(k*w*T); ; % coefficient for k!=0
```

```
yJhat(1,:) = Y(1).*exp(1i*w*0*t); % term in sum for k=0
```

```
% accumulate partial sum
```

```
for k = 2:1:50
```

```
Yexp(k,:) = Y(k).*exp(1i*(k-1)*w*t);
```

```
yJhat(k,:) = yJhat(k-1,:)+Y(k).*exp(1i*w*(k-1)*t);
```

```
end
```

```
j = 50;
```

```
plot(t,yJhat(j,:));
```

```
ylabel(['y(t)']);
```

```
xlabel('t');
```

```
title('L=10mH output');
```

(ii)4mh

```
%3.111b2
```

```
k = 1:1:50;
```

```
C= 100*(10.^(-6));
```

```
L= 4*(10.^(-3));
```

```
w= 1000;
```

```
s= 1i*w*k;
```

```
H= s./ (s.^2*L + s + 1/C); %transfer function of the bandpass filter
```

```
T0=pi./2000;
```

```
T=pi./500;
```

```
t = linspace(-pi*10.^(-3),pi*10.^(-3),400);
```

```
Y = zeros(1,50);
```

```
Y(1)=0;
```

```
Y(2:1:51) = H.*(sin(k*w*T0)*2)./(k*w*T); ; % coefficient for k!=0
```

```
yJhat(1,:) = Y(1).*exp(1i*w*0*t); % term in sum for k=0
```

```
% accumulate partial sum
```

```
for k = 2:1:50
```

```
Yexp(k,:) = Y(k).*exp(1i*(k-1)*w*t);
```

```
yJhat(k,:) = yJhat(k-1,:)+Y(k).*exp(1i*w*(k-1)*t);
```

```
end
```

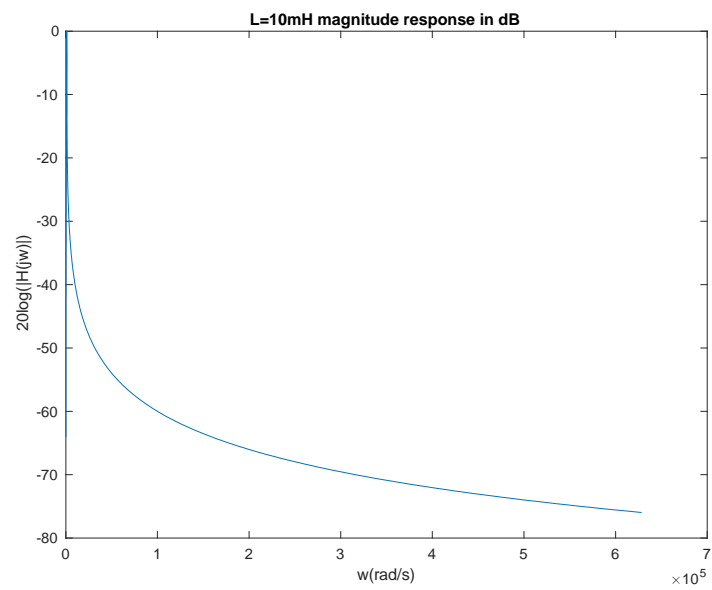
```

j = 50;
plot(t,yJhat(j,:));
ylabel(['y(t)']);
xlabel('t');
title('L=4mH output');

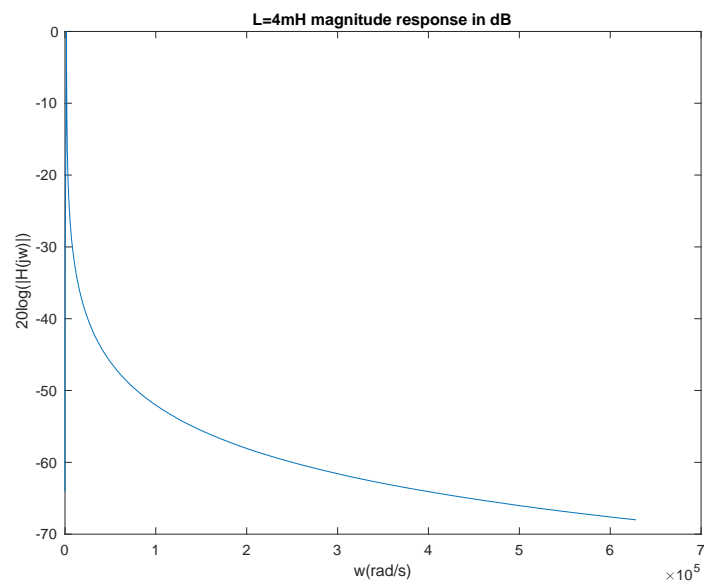
```

結果：

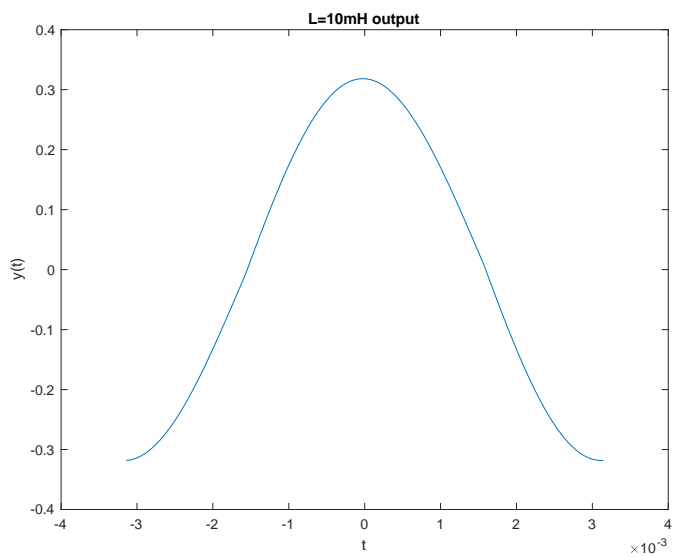
(a)(i)



(a)(ii)



(b)(i)



(b)(ii)

