

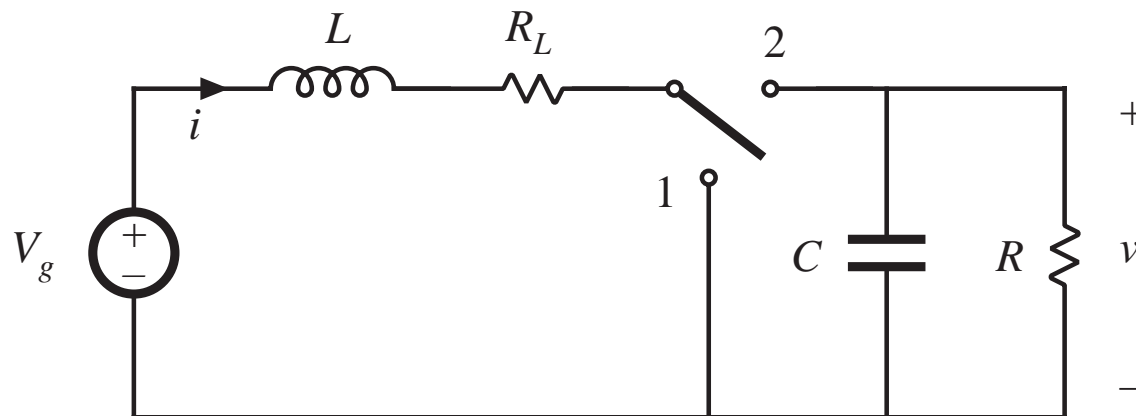
3.2. Inclusion of inductor copper loss

Dc transformer model can be extended, to include converter nonidealities.

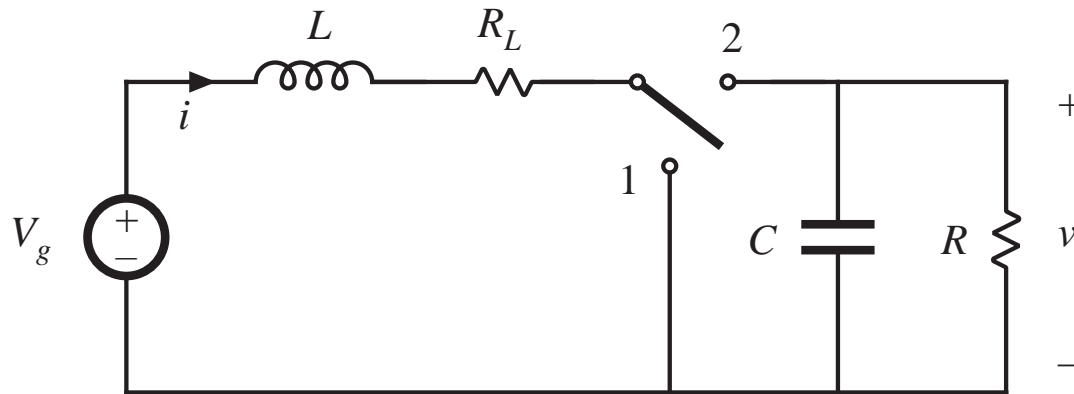
Example: inductor copper loss (resistance of winding):



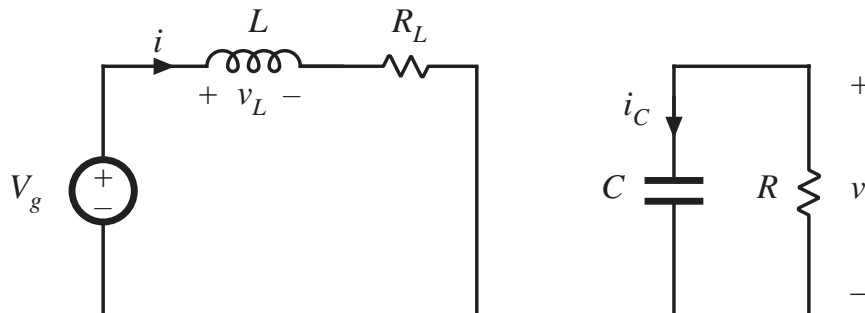
Insert this inductor model into boost converter circuit:



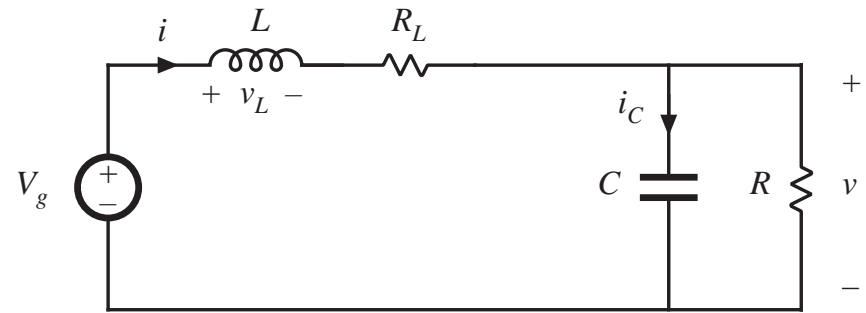
Analysis of nonideal boost converter



switch in position 1



switch in position 2

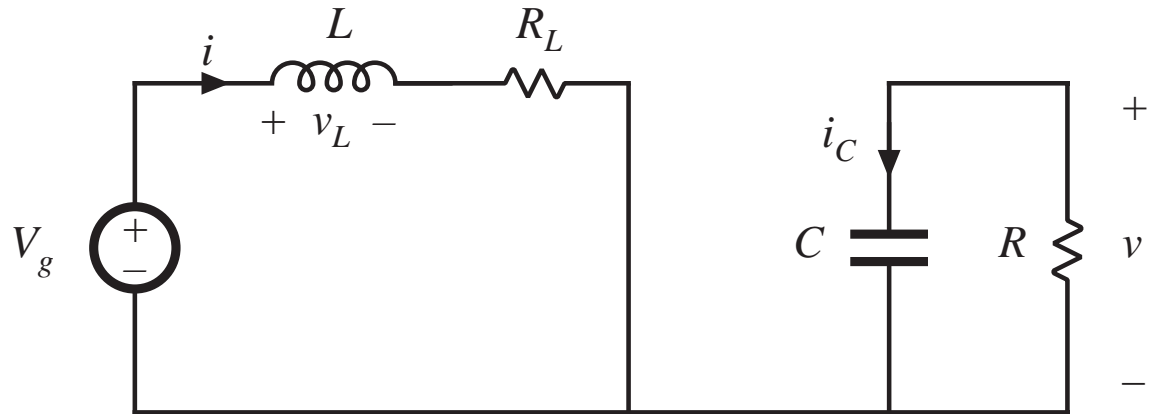


Circuit equations, switch in position 1

Inductor current and capacitor voltage:

$$v_L(t) = V_g - i(t) R_L$$

$$i_C(t) = -v(t) / R$$

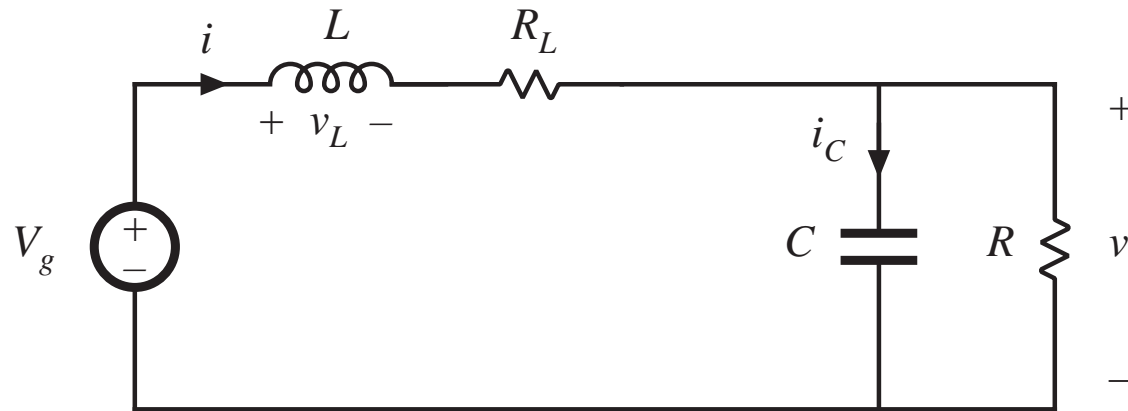


Small ripple approximation:

$$v_L(t) = V_g - I R_L$$

$$i_C(t) = -V / R$$

Circuit equations, switch in position 2



$$v_L(t) = V_g - i(t) R_L - v(t) \approx V_g - I R_L - V$$

$$i_C(t) = i(t) - v(t) / R \approx I - V / R$$

Inductor voltage and capacitor current waveforms

Average inductor voltage:

$$\begin{aligned}\langle v_L(t) \rangle &= \frac{1}{T_s} \int_0^{T_s} v_L(t) dt \\ &= D(V_g - I R_L) + D'(V_g - I R_L - V)\end{aligned}$$

Inductor volt-second balance:

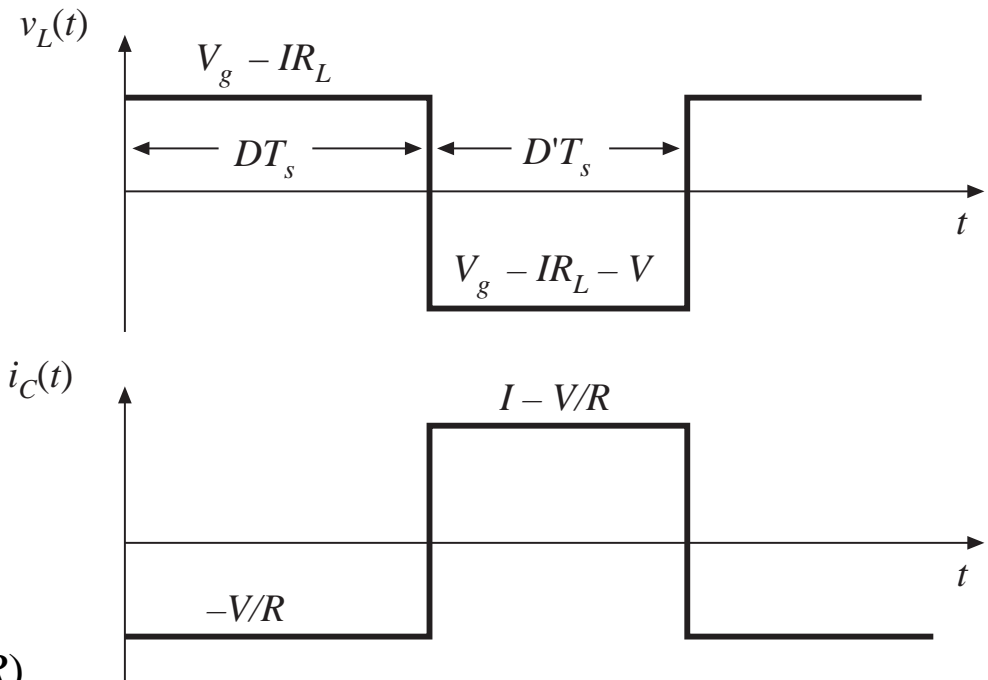
$$0 = V_g - I R_L - D'V$$

Average capacitor current:

$$\langle i_C(t) \rangle = D(-V/R) + D'(I - V/R)$$

Capacitor charge balance:

$$0 = D'I - V/R$$



Solution for output voltage

We now have two equations and two unknowns:

$$0 = V_g - I R_L - D'V$$

$$0 = D'I - V / R$$

Eliminate I and solve for V :

$$\frac{V}{V_g} = \frac{1}{D'} \frac{1}{(1 + R_L / D'^2 R)}$$

