HOMEWORK 3. Due: Wednesday, March 22, 2016. in class.

This is an individual assignment!

1. Extracting and simulating the synthesized design.

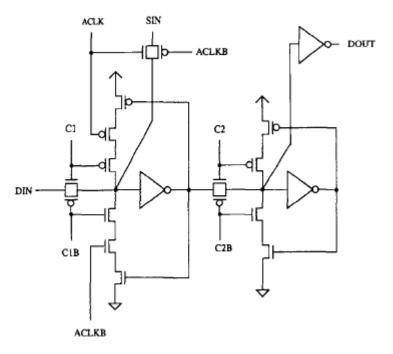
In the first lab, we have created a design by using a strictly digital ('VLSI') design flow. Then in the second lab, we extracted and simulated the synthesized design using custom design tools. In this third lab we will close the loop and build the remaining part of the flow where we will characterize a custom cell for use in the synthesis flow.

To do this, please work through the tutorial entitled: *Introduction to the Custom Design Flow: Building a standard cell*, then turn in:

- a) Report the critical path of the 3-to-8 decoder reported by IC Compiler before and after using your new NAND2 cell. Attach a layout of your cell placed inside IC Compiler (set level=99 to see inside the instance).
- b) Using a Monte Carlo simulation of 300 points, plot the mean and sigma for the inputrising/output-fall delay of INVX0_RVT (10ps input slope, 1fF load) for VDD between 0.2 and 1 on the same plot. Then plot σ/μ vs. VDD on another plot. Lastly, plot the delay of a 6σ cell relative to the mean vs. VDD on a third plot. Explain conceptually why variation has more of an effect at low voltages.
- c) Generate a histogram of the delay at 0.6V and at 1.05V and try to fit a normal curve (use histfit in Matlab). Are either of them Gaussian? If a distribution is asymmetric, explain why one side has the long tail and not the other. (Optional: use the qqplot command to compare two vectors: one vector from your simulations, and the other sampled from a normal distribution with the μ and σ found from the fit using normrnd)
- d) Compute σ_{Vth} for the NMOS in INVX0_RVT. Then, using Ion from the alpha-power-law model (K=1.7e-4, Vth=0.366, alpha=1.36), write out the equation for the delay of an inverter (simply assuming maximum current for a transition from VDD to VDD/2 is close enough: $\Delta T = C\Delta V/I$, where $\Delta V = VDD/2$ and C is approximately 1.5fF). Then use Matlab to run a Monte Carlo simulation on this model for 0.6V and 1.05V by using normrnd with a μ =0 and σ = σ_{Vth} and adding this value to the threshold (which is exactly what HSPICE is doing for Monte Carlo simulation, but HSPICE is solving more complex equations). How do the distributions compare to part c)? Based on the equation you wrote, should the delay be Gaussian?

2. Flip-flops

A sequential circuit is shown in figure.



- a) What is the logic function of this circuit? You can assume that C1 and C2 are opposite phases of the same clock.
- b) Assuming that NMOS and PMOS devices have the same mobility and are all symmetrically sized, order the following timing parameters by magnitude: setup time, hold time, clock-output delay.

3. Timing

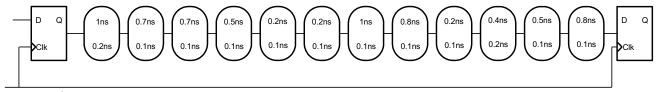


Figure 2.

A pipeline stage is shown in Figure 2. Static combinational logic is placed between two edge-triggered flip-flops. Each cloud of combinational logic has been annotated with its maximum (top) and minimum (bottom) propagation delays.

Your task is to minimize the cycle time by using deeper pipelining. You are allowed to add two more cycles of latency, by adding either latches or flip-flops between combination logic clouds in Figure 2. Available flip-flops have Clk-Q delay of 200ps, setup time of 100ps and zero hold time. Available latches have both Clk-Q and D-Q delays of 150ps, setup time of 100ps, and zero hold time. You can ignore the clock skew. You can add latches or flip-flops in between any two logic clouds, but you cannot break up a logic cloud.

Insert an appropriate number of pipelining elements (latches or flip-flops that minimizes the cycle time with two extra cycles of latency. Clock duty cycle is 50%