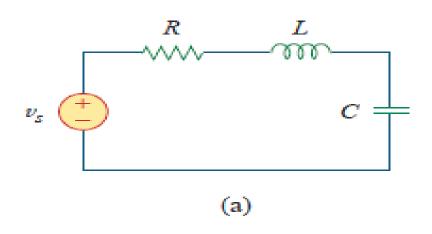
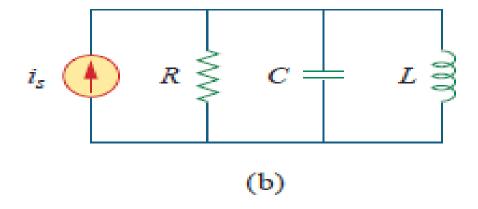
## **Second-Order Circuits**

- > Circuits containing two storage elements.
- > Second-order circuits because their responses are described by differential equations that contain second derivatives.

A **second-order circuit** is characterized by a second-order differential equation. It consists of resistors and the equivalent of two energy storage elements.

# Typical examples of second-order circuits:



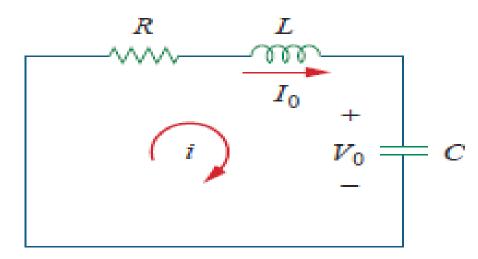


(a) series RLC circuit,

(b) parallel RLC

### The Source-Free Series *RLC* Circuit

The circuit is being excited by the energy initially stored in the capacitor and inductor

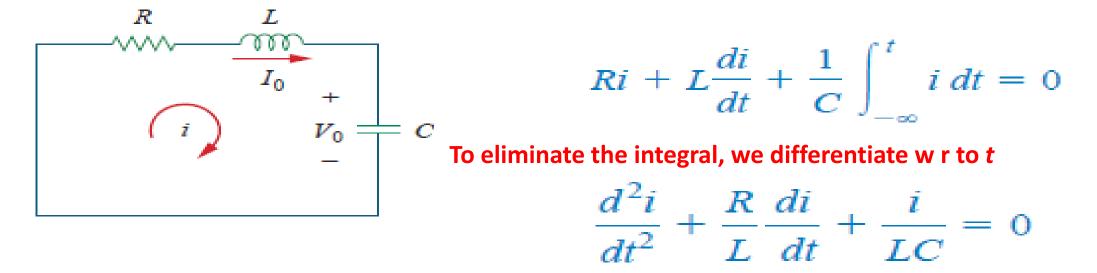


The energy is represented by the initial capacitor voltage  $V_0$  and initial inductor current  $I_0$ . Thus, at t=0

$$v(0) = \frac{1}{C} \int_{-\infty}^{0} i \, dt = V_0$$
$$i(0) = I_0$$

Applying KVL around the loop in Fig. 8.8,

$$Ri + L\frac{di}{dt} + \frac{1}{C} \int_{-\infty}^{t} i \, dt = 0$$



This is a second-order differential equation and to solve this Eqn

To solve such a second-order differential equation requires that we have two initial conditions, such as the initial value of i and its first derivative or initial values of some i and v

$$Ri(0) + L\frac{di(0)}{dt} + V_0 = 0$$
 or  $\frac{di(0)}{dt} = -\frac{1}{L}(RI_0 + V_0)$   
 $i(0) = I_0$   $i = Ae^{st}$ 

where A and s are constants to be determined.

$$\frac{d^2i}{dt^2} + \frac{R}{L}\frac{di}{dt} + \frac{i}{LC} = 0$$

$$As^{2}e^{st} + \frac{AR}{L}se^{st} + \frac{A}{LC}e^{st} = 0$$

$$Ae^{st}\left(s^{2} + \frac{R}{L}s + \frac{1}{LC}\right) = 0$$

$$s_{1} = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^{2} - \frac{1}{LC}}$$

$$s^{2} + \frac{R}{L}s + \frac{1}{LC} = 0$$

$$s_{2} = -\frac{R}{2L} - \sqrt{\left(\frac{R}{2L}\right)^{2} - \frac{1}{LC}}$$

This quadratic equation is known as the *characteristic equation* of the differential Eq. since the roots of the equation dictate the character of *i*.

$$s_{1} = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^{2} - \frac{1}{LC}}$$

$$s_{2} = -\frac{R}{2L} - \sqrt{\left(\frac{R}{2L}\right)^{2} - \frac{1}{LC}}$$

A more compact way of expressing the roots is

$$s_1 = -\alpha + \sqrt{\alpha^2 - \omega_0^2}, \qquad s_2 = -\alpha - \sqrt{\alpha^2 - \omega_0^2}$$

where 
$$\alpha = \frac{R}{2L}$$
,  $\omega_0 = \frac{1}{\sqrt{LC}}$ 

The roots  $s_1$  and  $s_2$  are called *natural frequencies*, measured in nepers per second (Np/s)

 $\omega_0$  is known as the resonant frequency or the undamped natural frequency (rad/s)

 $\alpha$  is the neper frequency or the damping factor

$$s^2 + \frac{R}{L}s + \frac{1}{LC} = 0$$

$$s^2 + 2\alpha s + \omega_0^2 = 0$$

### The two values of s in indicate that there are two possible solutions for i

$$i_1 = A_1 e^{s_1 t}, i_2 = A_2 e^{s_2 t}$$
  
 $i(t) = A_1 e^{s_1 t} + A_2 e^{s_2 t}$ 

where the constants  $A_1$  and  $A_2$  are determined from the initial values i(0) and di(0)/dt

### From following, we can infer that there are three types of solutions possible

$$s_1 = -\alpha + \sqrt{\alpha^2 - \omega_0^2}, \qquad s_2 = -\alpha - \sqrt{\alpha^2 - \omega_0^2}$$

- 1. If  $\alpha > \omega_0$ , we have the *overdamped* case.
- 2. If  $\alpha = \omega_0$ , we have the *critically damped* case.
- 3. If  $\alpha < \omega_0$ , we have the *underdamped* case.

# Overdamped Case ( $\alpha > \omega_0$ )

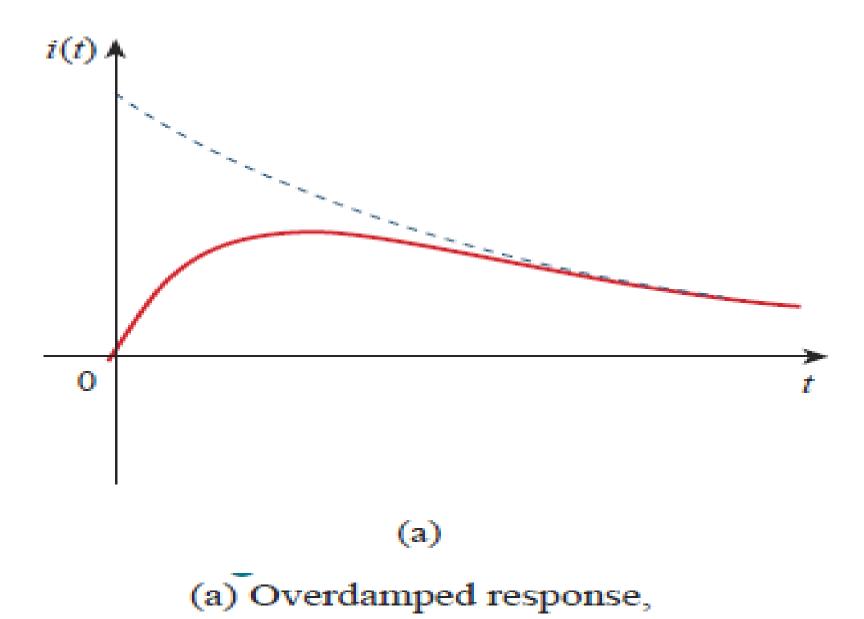
$$s_1 = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}}$$
$$s_2 = -\frac{R}{2L} - \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}}$$

$$s_1 = -\alpha + \sqrt{\alpha^2 - \omega_0^2}, \qquad s_2 = -\alpha - \sqrt{\alpha^2 - \omega_0^2}$$

 $\alpha > \omega_0$  implies  $C > 4L/R^2$ . When this happens, both roots  $s_1$  and  $s_2$  are negative and real.

$$i(t) = A_1 e^{s_1 t} + A_2 e^{s_2 t}$$

which decays and approaches zero as t increases.



## Critically Damped Case ( $\alpha = \omega_0$ )

When  $\alpha = \omega_0$ ,  $C = 4L/R^2$  and

$$s_1 = s_2 = -\alpha = -\frac{R}{2L}$$
  
 $i(t) = A_1 e^{-\alpha t} + A_2 e^{-\alpha t} = A_3 e^{-\alpha t}$ 

where  $A_3 = A_1 + A_2$ . This cannot be the solution, because the two initial conditions cannot be satisfied with the single constant  $A_3$ .

#### So let's begin with

$$\frac{d^2i}{dt^2} + \frac{R}{L}\frac{di}{dt} + \frac{i}{LC} = 0$$

$$\frac{d^2i}{dt^2} + \frac{R}{L}\frac{di}{dt} + \frac{i}{LC} = 0$$

When  $\alpha = \omega_0 = R/2L$ 

$$\frac{d^2i}{dt^2} + 2\alpha \frac{di}{dt} + \alpha^2 i = 0$$

$$\frac{d}{dt}\left(\frac{di}{dt} + \alpha i\right) + \alpha\left(\frac{di}{dt} + \alpha i\right) = 0$$

$$\frac{df}{dt} + \alpha f = 0$$

 $\frac{df}{dt} + \alpha f = 0 If we let <math>f = \frac{di}{dt} + \alpha i$ 

which is a first-order differential equation with solution  $f = A_1 e^{-\alpha t}$ , where  $A_1$  is a constant.

$$\frac{di}{dt} + \alpha i = A_1 e^{-\alpha t}$$

$$e^{\alpha t} \frac{di}{dt} + e^{\alpha t} \alpha i = A_1$$

$$\frac{d}{dt} (e^{\alpha t} i) = A_1$$

$$e^{\alpha t} i = A_1 t + A_2$$

Integrating both sides yields  $i = (A_1t + A_2)e^{-\alpha t}$ 

where  $A_2$  is another constant. Hence, the natural response of the critically damped circuit is a sum of two terms: a negative exponential and a negative exponential multiplied by a linear term, or

$$i(t) = (A_2 + A_1 t)e^{-\alpha t}$$

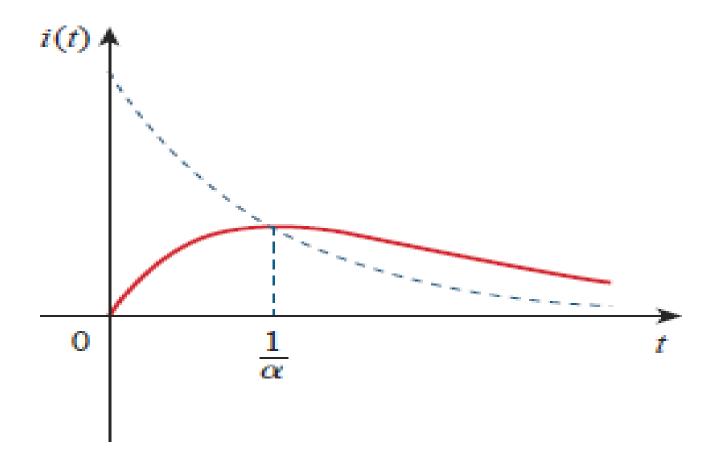


Fig. is a sketch of  $i(t) = te^{-\alpha t}$ , which reaches a maximum value of  $e^{-1}/\alpha$  at  $t = 1/\alpha$ , one time constant, and then decays all the way to zero.

### Underdamped Case ( $\alpha < \omega_0$ )

For  $\alpha < \omega_0$ ,  $C < 4L/R^2$ . The roots may be written as

$$s_1 = -\alpha + \sqrt{-(\omega_0^2 - \alpha^2)} = -\alpha + j\omega_d$$
  
$$s_2 = -\alpha - \sqrt{-(\omega_0^2 - \alpha^2)} = -\alpha - j\omega_d$$

where  $j = \sqrt{-1}$  and  $\omega_d = \sqrt{\omega_0^2 - \alpha^2}$ , which is called the *damping frequency*. Both  $\omega_0$  and  $\omega_d$  are natural frequencies because they help determine the natural response; while  $\omega_0$  is often called the *undamped natural frequency*,  $\omega_d$  is called the *damped natural frequency*. The natural response is

$$i(t) = A_1 e^{-(\alpha - j\omega_d)t} + A_2 e^{-(\alpha + j\omega_d)t}$$
  
=  $e^{-\alpha t} (A_1 e^{j\omega_d t} + A_2 e^{-j\omega_d t})$ 

Using Euler's identities,

$$e^{j\theta} = \cos\theta + j\sin\theta, \qquad e^{-j\theta} = \cos\theta - j\sin\theta$$

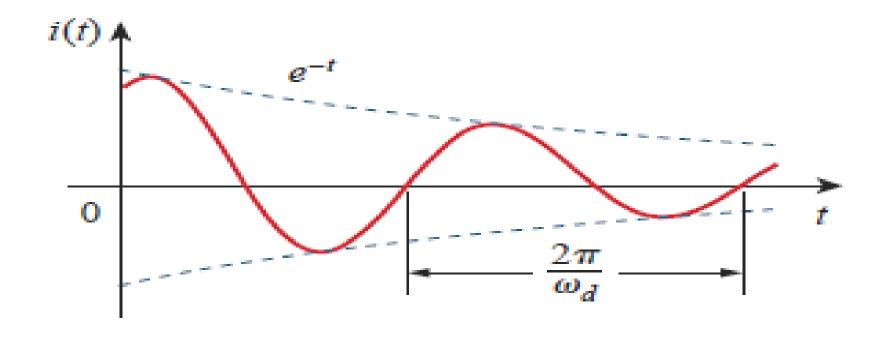
 $i(t) = e^{-\alpha t} [A_1(\cos \omega_d t + j \sin \omega_d t) + A_2(\cos \omega_d t - j \sin \omega_d t)]$ =  $e^{-\alpha t} [(A_1 + A_2) \cos \omega_d t + j(A_1 - A_2) \sin \omega_d t]$ 

Replacing constants  $(A_1 + A_2)$  and  $j(A_1 - A_2)$  with constants  $B_1$  and  $B_2$ ,

we write

we get

$$i(t) = e^{-\alpha t} (B_1 \cos \omega_d t + B_2 \sin \omega_d t)$$



With the presence of sine and cosine functions, it is clear that the natural response for this case is exponentially damped and oscillatory in nature. The response has a time constant of  $1/\alpha$  and a period of  $T = 2\pi/\omega_d$ .