

Module:7 Special Function ICs

Course: BECE206L – Analog Circuits



VIT[®]

Vellore Institute of Technology

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CHENNAI

Module:7 Special Function ICs

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- **IC 555 timer**, Astable and **Monostable operations**, and **applications**. IC voltage regulator - LM317.

1. Introduction

- IC 555 timer – highly stable device for generating accurate time delay and oscillations
- Either as 14 pin DIP with two 555 timers or as single 8 pin DIP with one 555 timer
- Supply voltage: +5V to +18V
- Can drive a load upto 220mA
- Compatible with both Transistor Transistor Logic (TTL) or CMOS(complementary MOSFET) logic circuits
- Applications: oscillator, pulse generator, ramp and square wave generator, mono-shot multivibrator, burglar alarm, traffic light control and voltage monitor, etc

2. PIN Diagram and Functional diagram

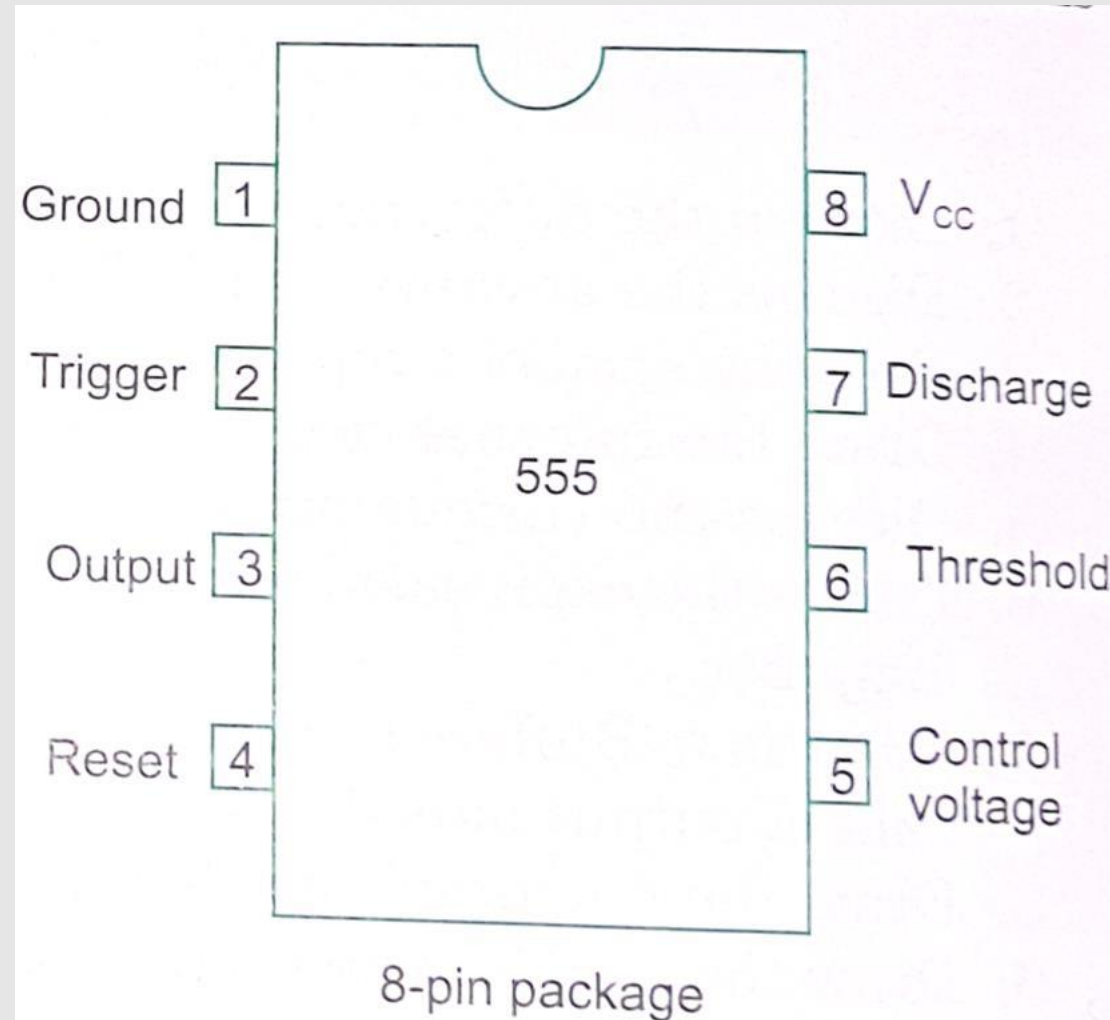
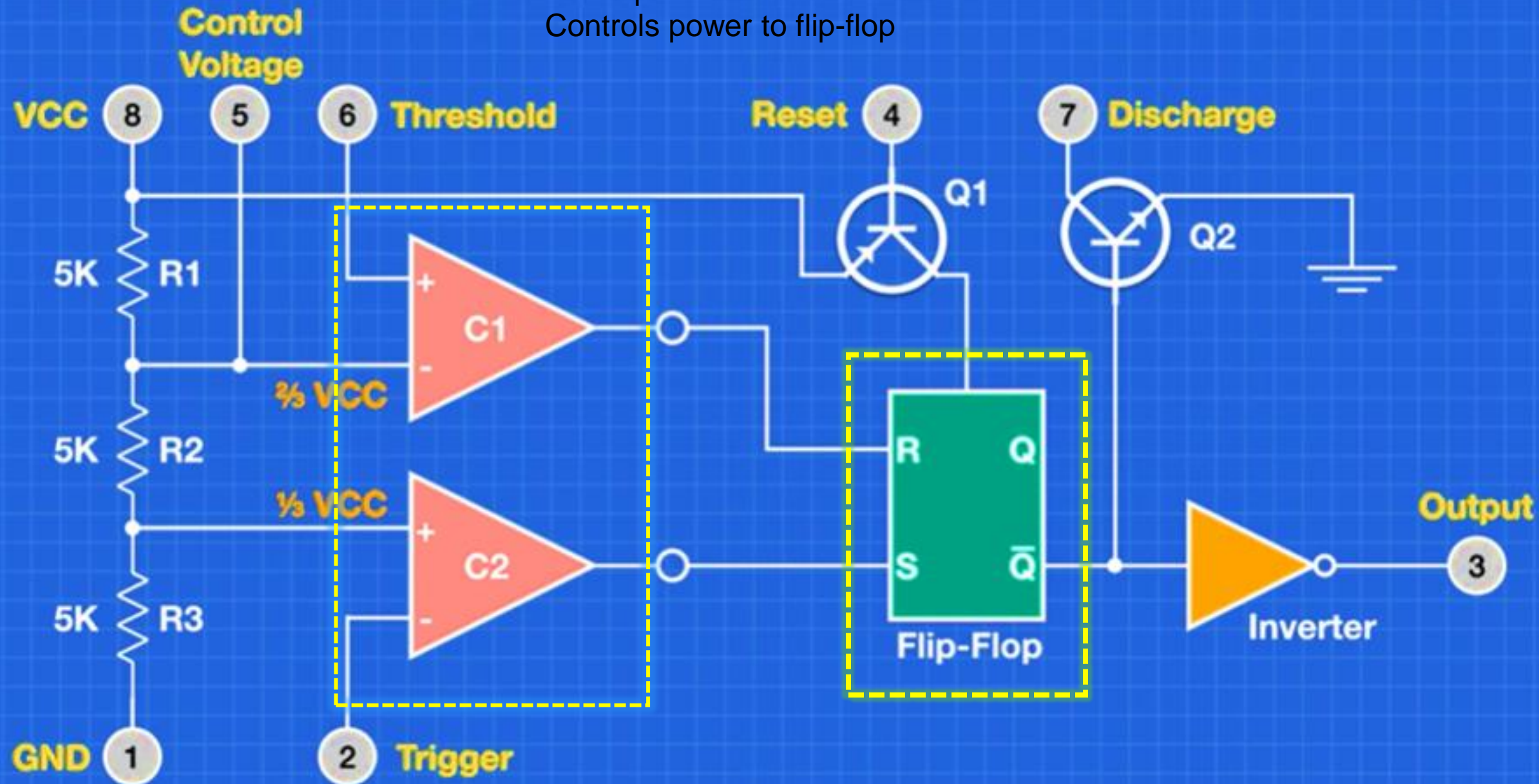


Fig. 9.1 Pin diagram

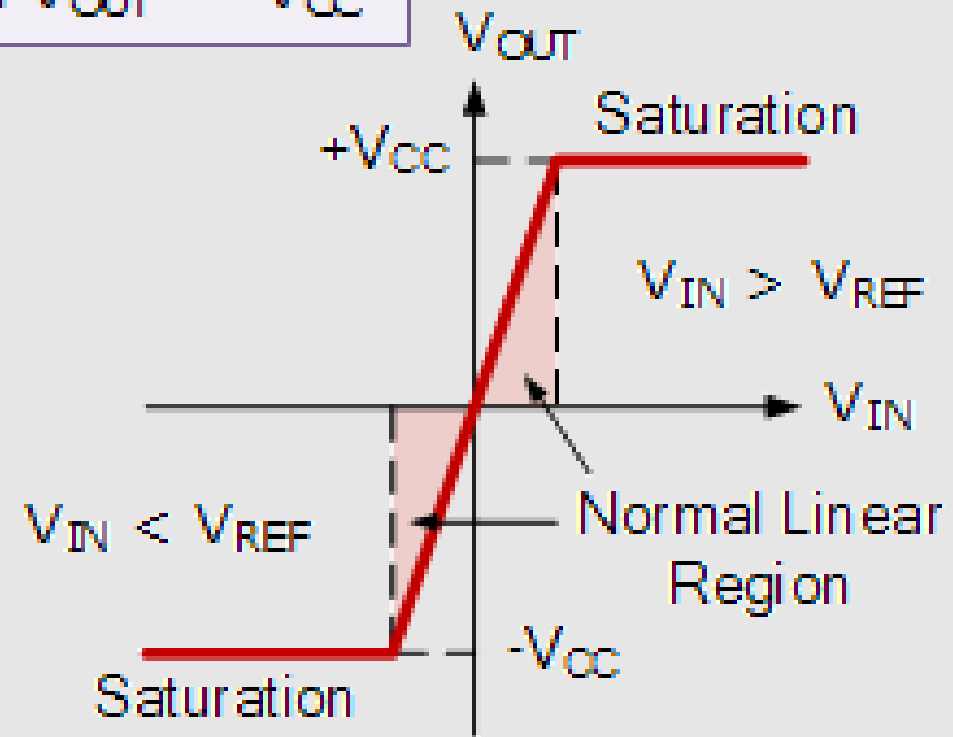
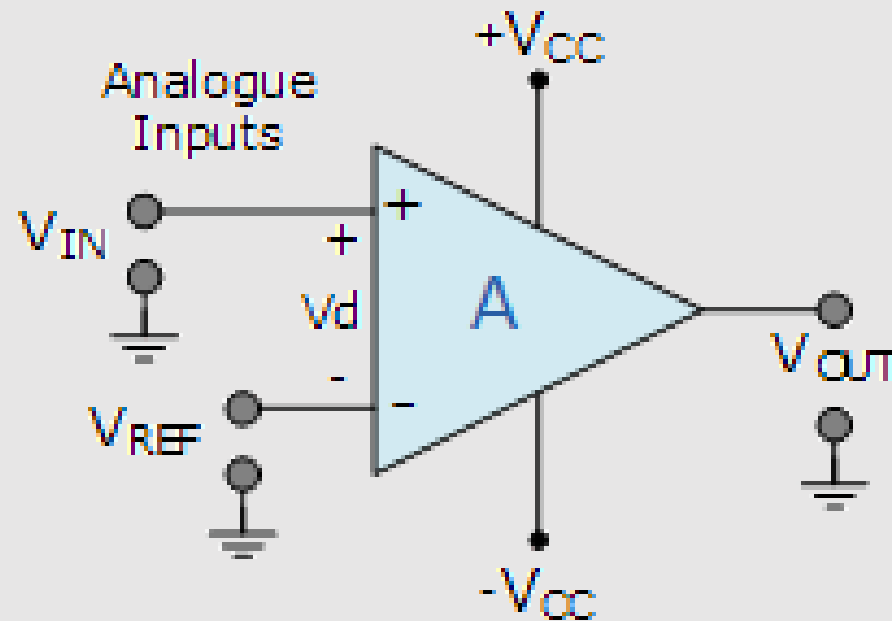
Reset pin connected to TRANSISTOR
Controls power to flip-flop



Comparator

As v_1 and v_2 are applied at terminals:
non-inverting(+) and inverting(-)
For infinite gain: v_o will be in
positive saturation ($+V_{sat}$) if $v_1 > v_2$
negative saturation ($-V_{sat}$) if $v_2 > v_1$

If $V_{IN} > V_{REF}$ then $V_{OUT} = +V_{CC}$
If $V_{IN} < V_{REF}$ then $V_{OUT} = -V_{CC}$



Flip Flop

- Set/Reset with Inverted Inputs
- Set and Reset
- Q and Not Q
- Flip-flops and latches are used as data storage elements. A flip-flop is a device which stores a single bit (binary digit) of data; one of its two states represents a "one" and the other represents a "zero".

Threshold	Trigger	R	S	NOT Q
LOW	LOW	LOW	HIGH	LOW
LOW	HIGH	LOW	LOW	LOW
HIGH	HIGH	HIGH	LOW	HIGH
HIGH	LOW	HIGH	HIGH	INVALID

INPUTS			OUTPUT	STATE
CLK	S	R	Q	
X	0	0	No Change	Previous
↑	0	1	0	Reset
↑	1	0	1	Set
↑	1	1	-	Forbidden

2. PIN Diagram and Functional diagram

- **$5k\Omega$ internal resistors**: act as voltage dividers
Provides bias of $2V_{cc}/3$ to **upper comparator** (UC) and Bias of $1V_{cc}/3$ to **lower comparator** (LC)
- These voltage levels are thresholds for comparator and fix timing interval
- **By providing modulation to control voltage at pin 5, time can be varied electronically**
- When no control is needed, connect $0.01\mu F$ capacitor between 5 and ground, to bypass noise, ripples.

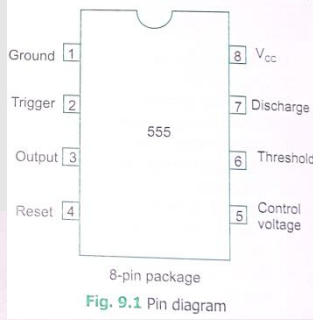
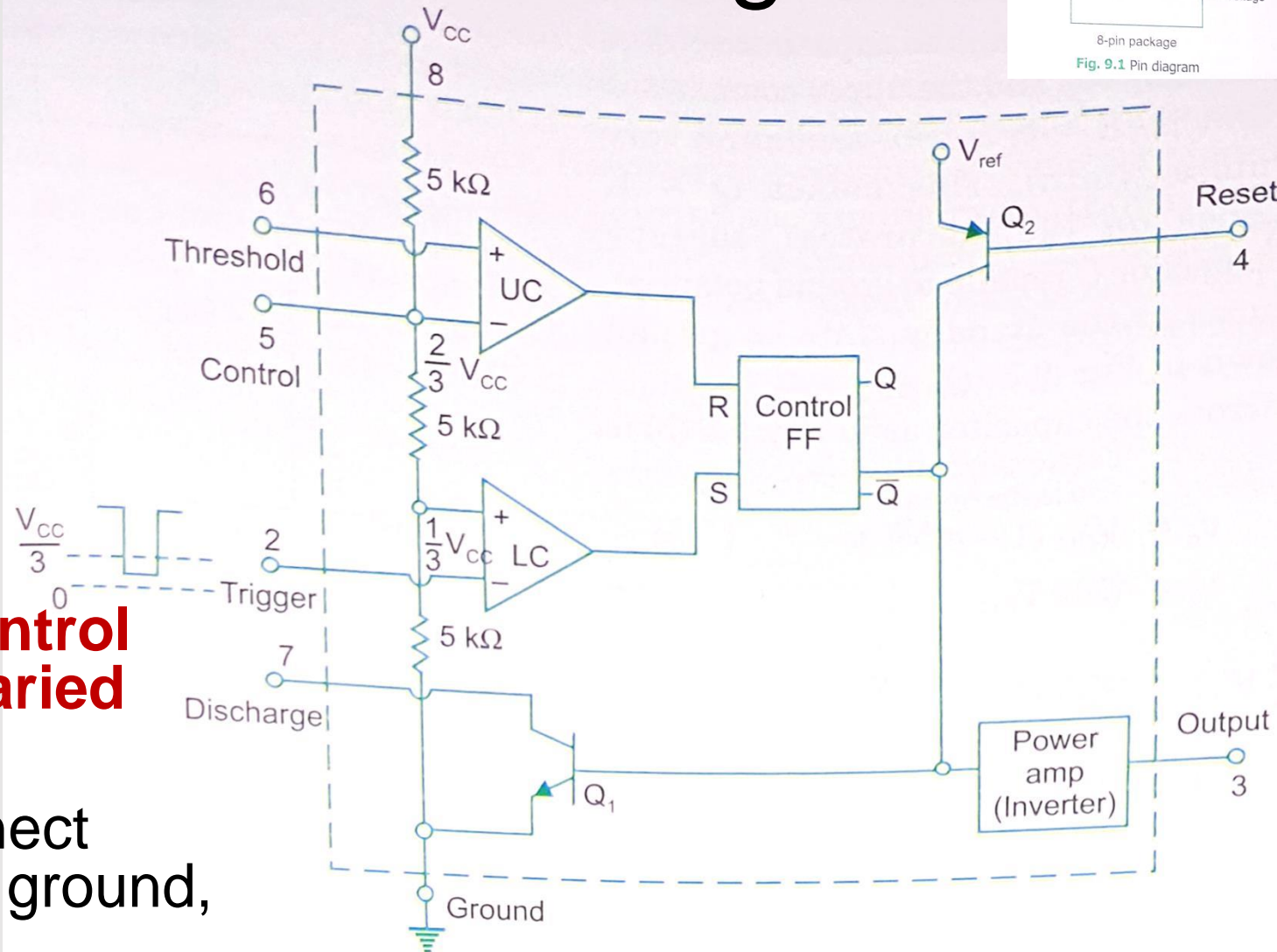
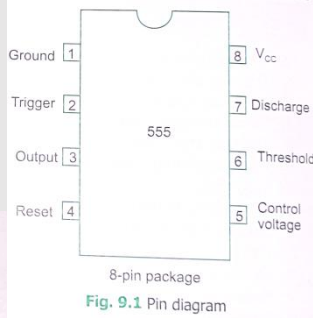


Fig. 9.2 Functional diagram of 555 timer

2. PIN Diagram and Functional diagram



- In standby(stable state), output \bar{Q} of flipflop (FF) is HIGH, which makes power amp(inverter) output to LOW.
- When negative going trigger (which goes from value above $V_{cc}/3$ to below $V_{cc}/3$) is given to pin 2, output of the comparator LC goes HIGH, Sets flipflop FF: **output Q goes HIGH, \bar{Q} goes LOW,**
- When positive going signal at threshold (pin 6) goes above $2V_{cc}/3$, Output of UC comparator goes HIGH, Resets the flipflop, **$Q \rightarrow LOW$, $\bar{Q} \rightarrow HIGH$**

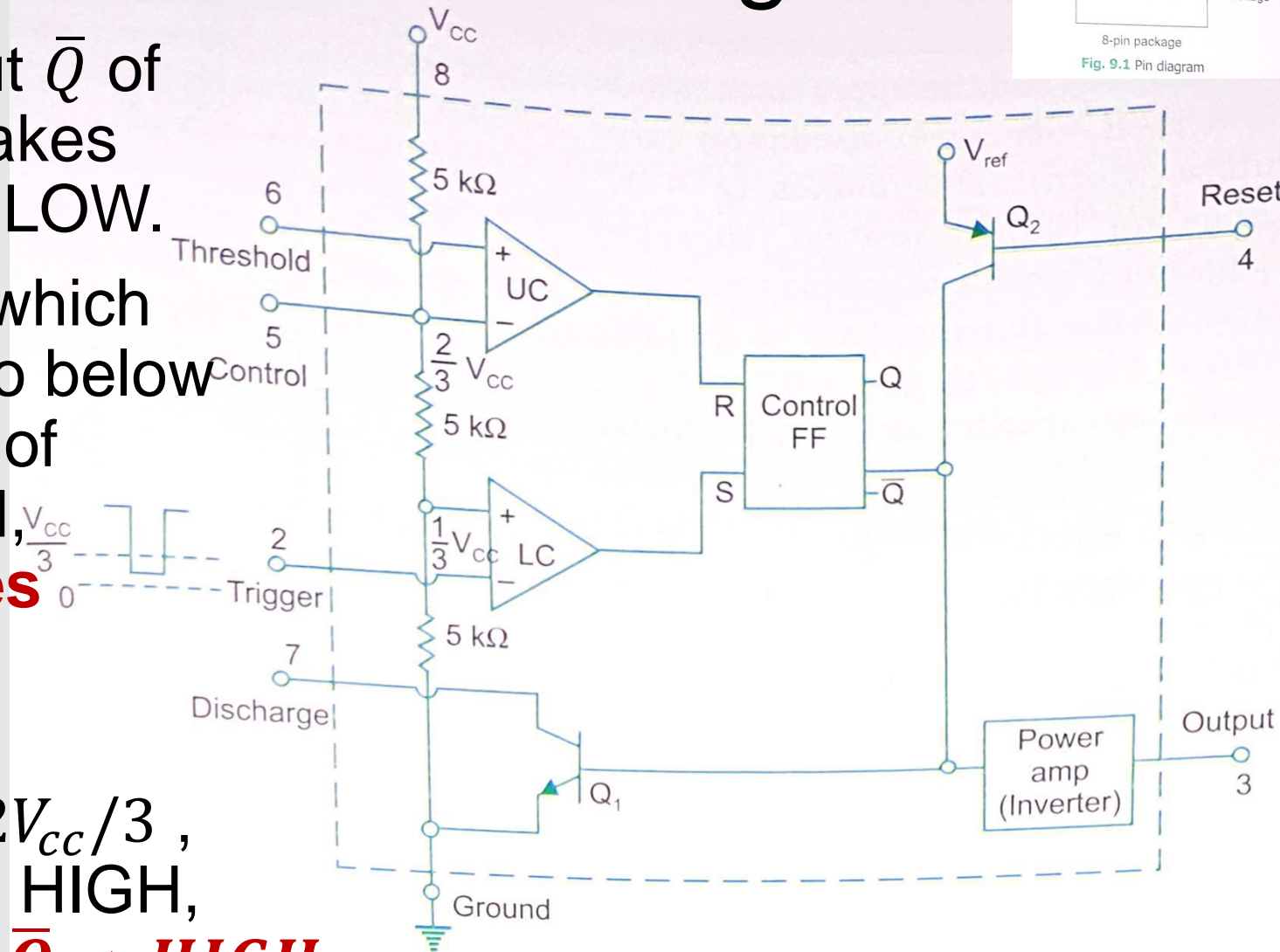


Fig. 9.2 Functional diagram of 555 timer

2. PIN Diagram and Functional diagram

- RESET pin (4) when enabled (LOW or less than 0.4V for Q2's PN voltage), this overrides any instruction coming from lower comparator to FF.
- RESET pin(4) when unused, is returned to V_{CC} .
- Q_2 serves as buffer to isolate reset input from FF and transistor Q_1
- Q_2 transistor is driven by V_{ref} (internal reference voltage) which is derived from V_{CC}

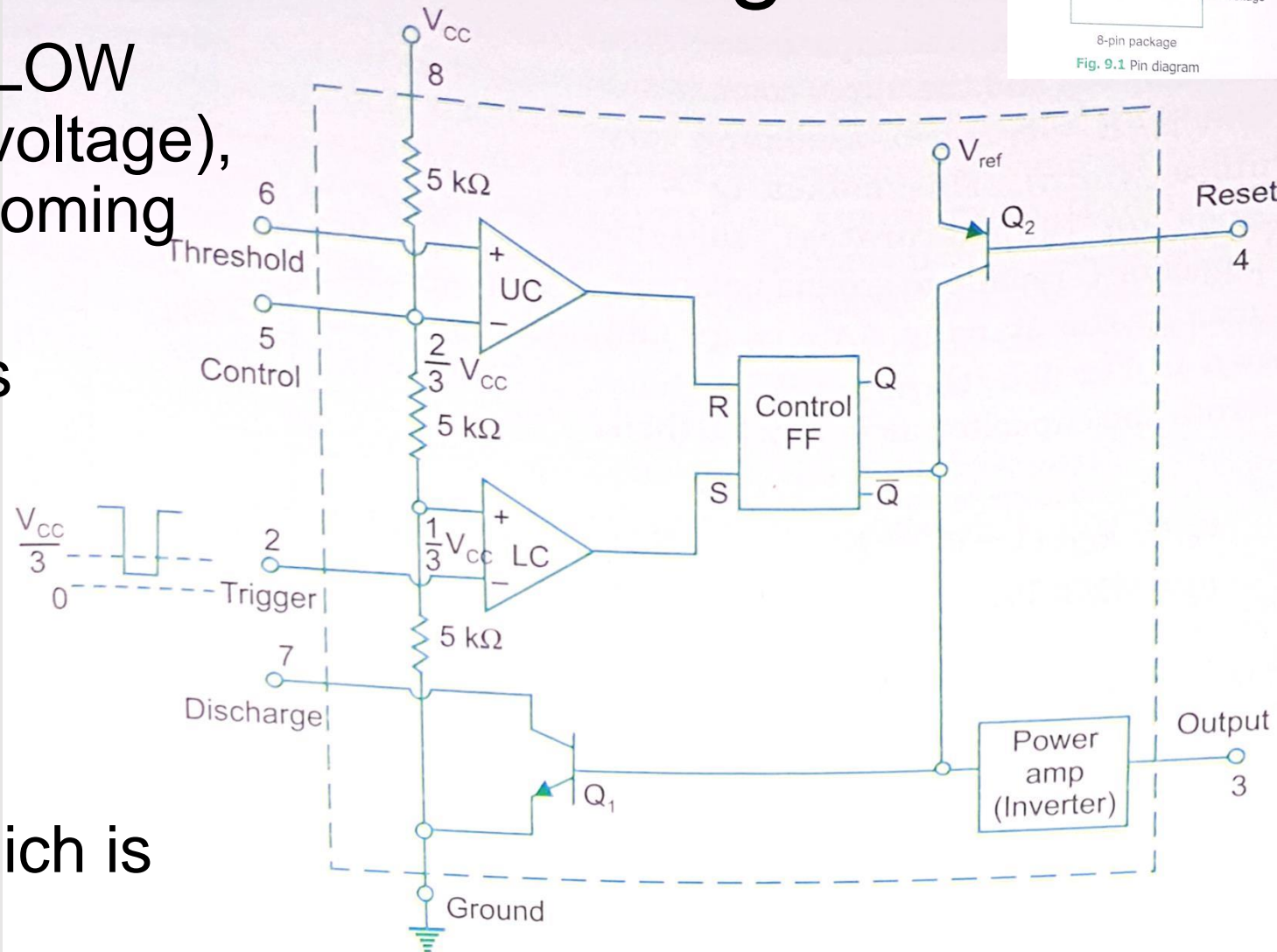
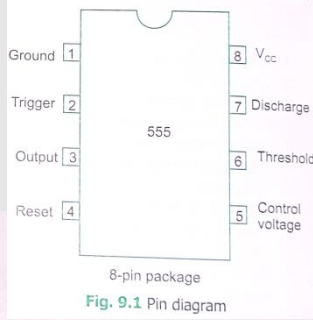


Fig. 9.2 Functional diagram of 555 timer



3. Monostable Operation

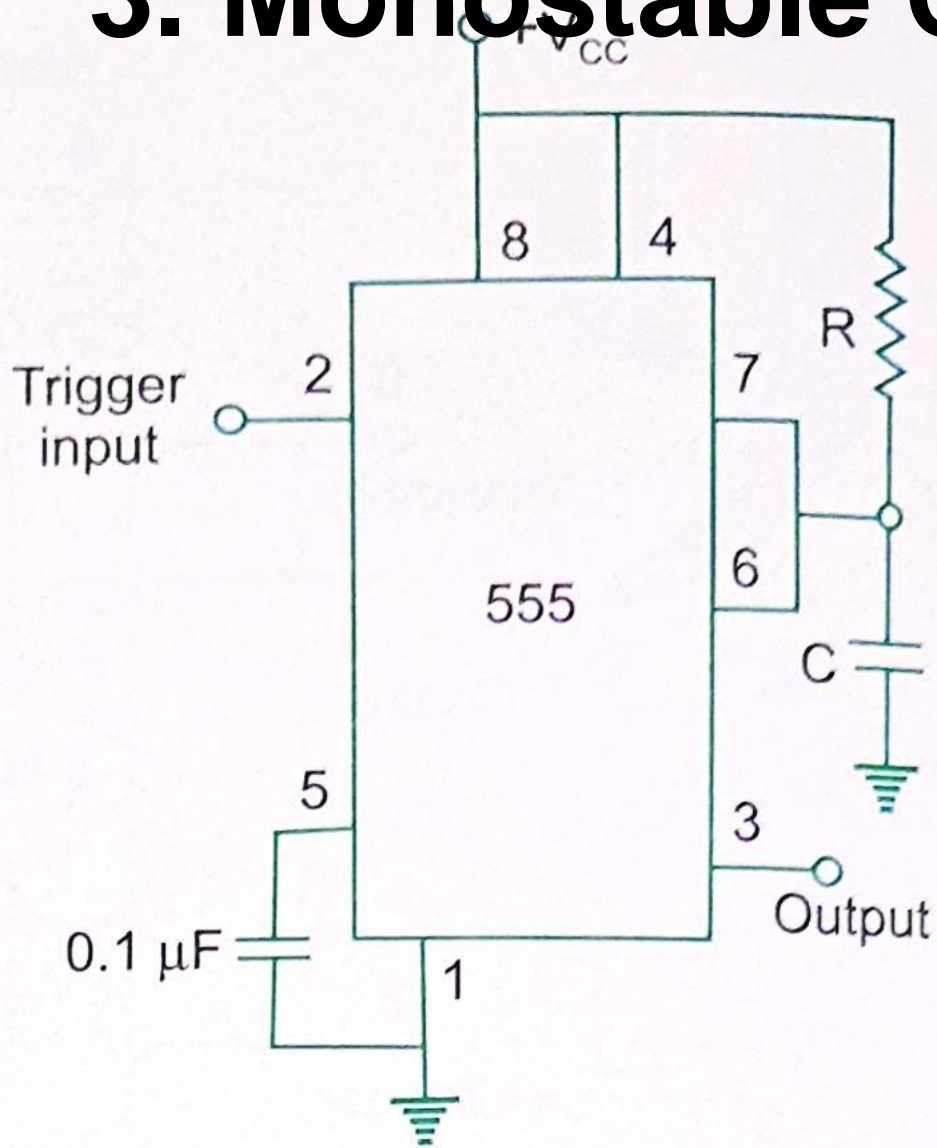
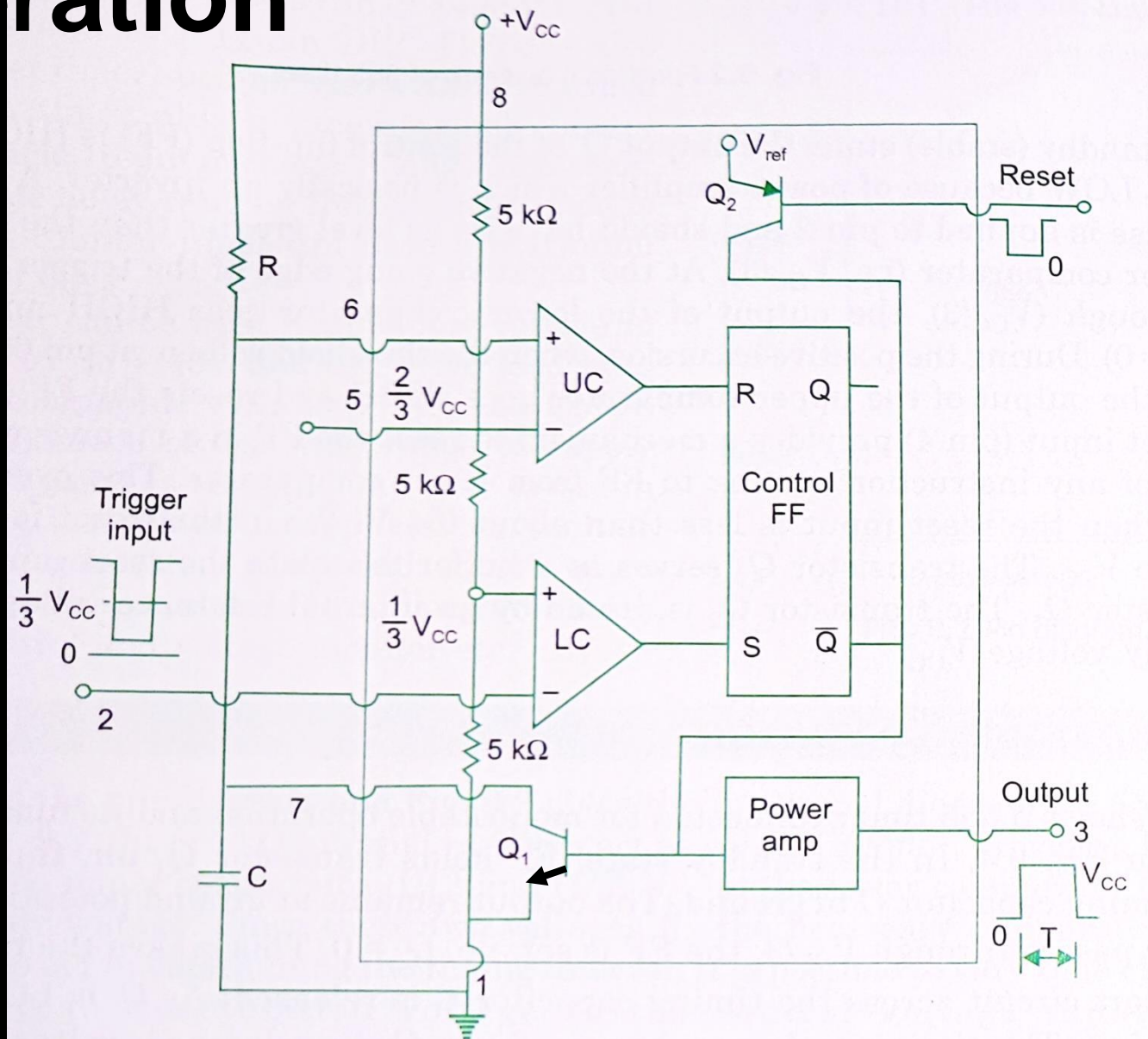


Fig. 9.3 Monostable multivibrator



3. Monostable Operation

- 1. In standby state:

FF holds transistor Q_1 on.

Thus clamping the external timing capacitor C to ground. Output remains at ground potential (LOW)

- 2. As trigger passes through

$V_{CC}/3$, FF is set, $Q = 1$; $\bar{Q} = 0$:

This makes transistor Q_1 off, **previous short circuit across timing capacitor C is removed.** C is unclamped, voltage across it rises exponentially **through R towards V_{CC} with time constant RC .**

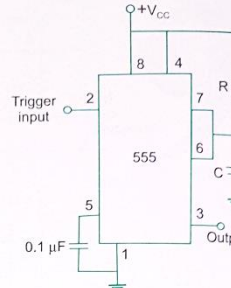
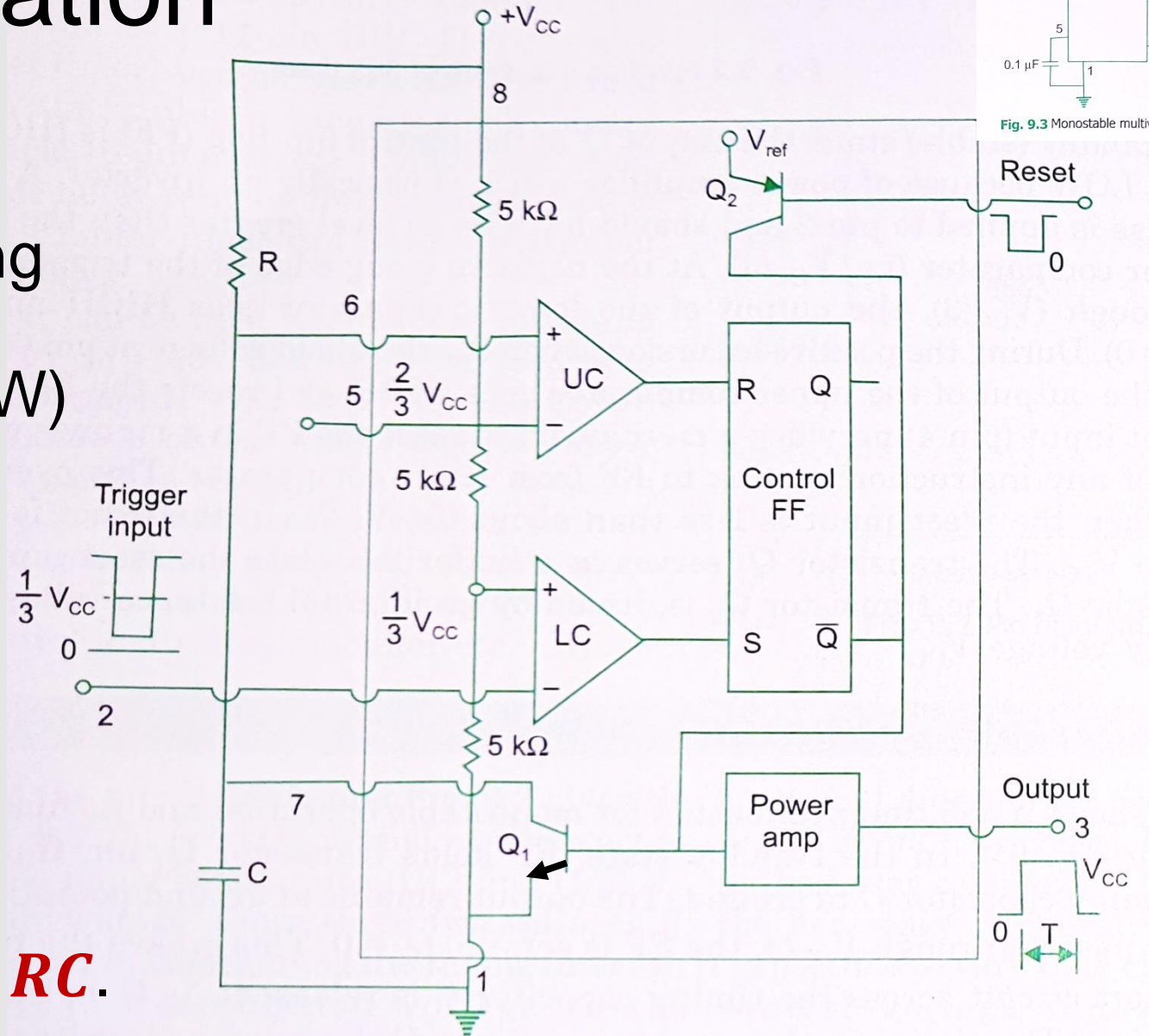
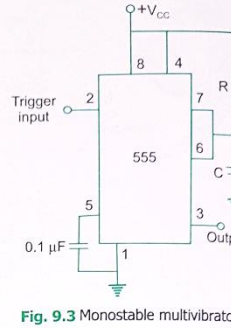
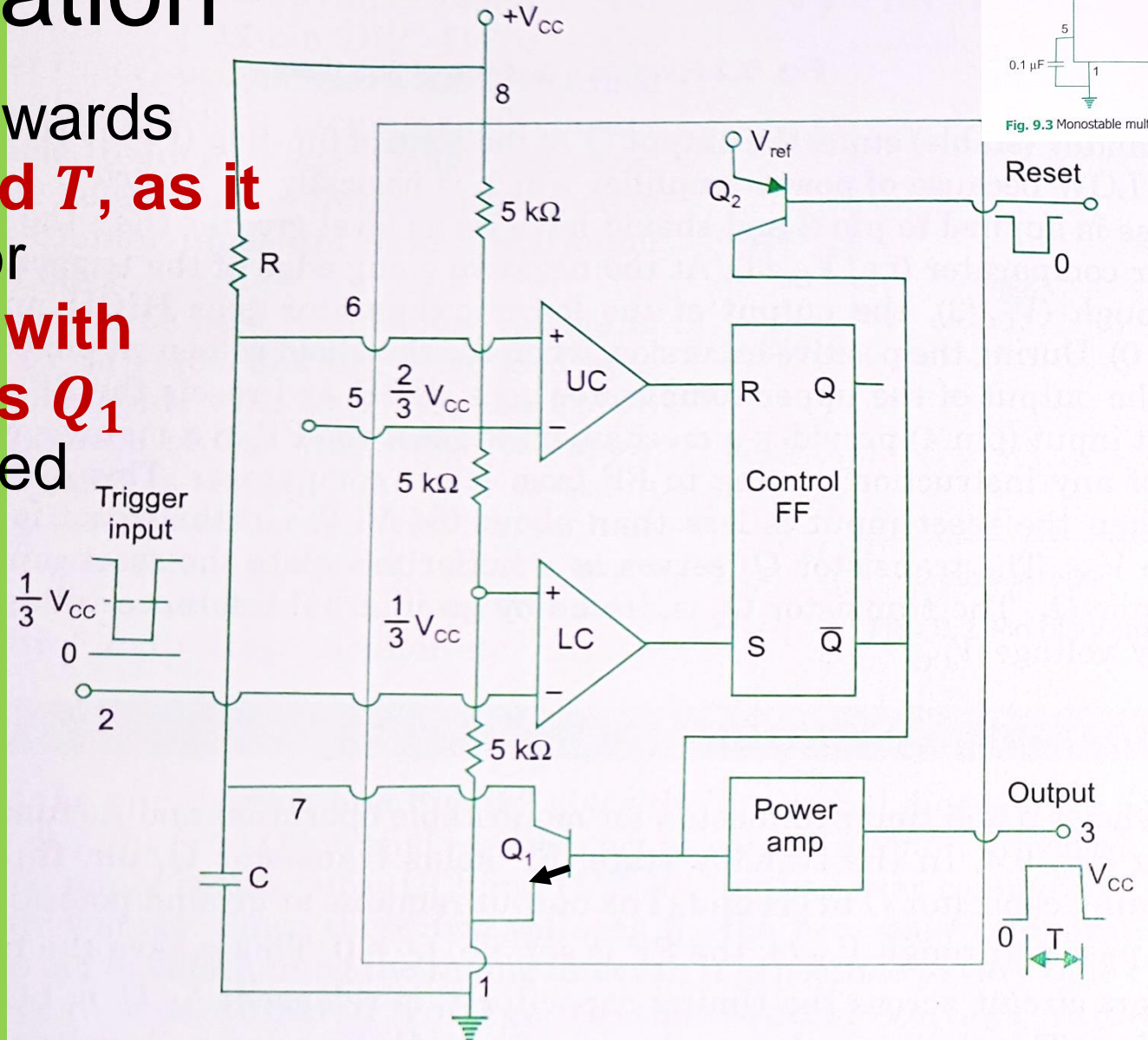


Fig. 9.3 Monostable multivibrator

3. Monostable Operation

- As capacitor tries to charge towards V_{CC} , through R with time period T , as it reaches $2V_{CC}/3$, UC comparator output is HIGH, and **FF is reset with $Q = 0$ and $\bar{Q} = 1$** . This **saturates Q_1 to be ON**, and capacitor is shorted to ground (discharges) through transistor.
- **The output returns to standby**



3. Monostable Operation

- Voltage across capacitor:

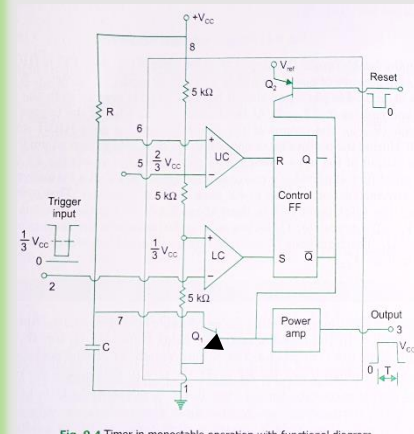
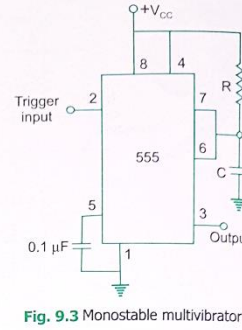
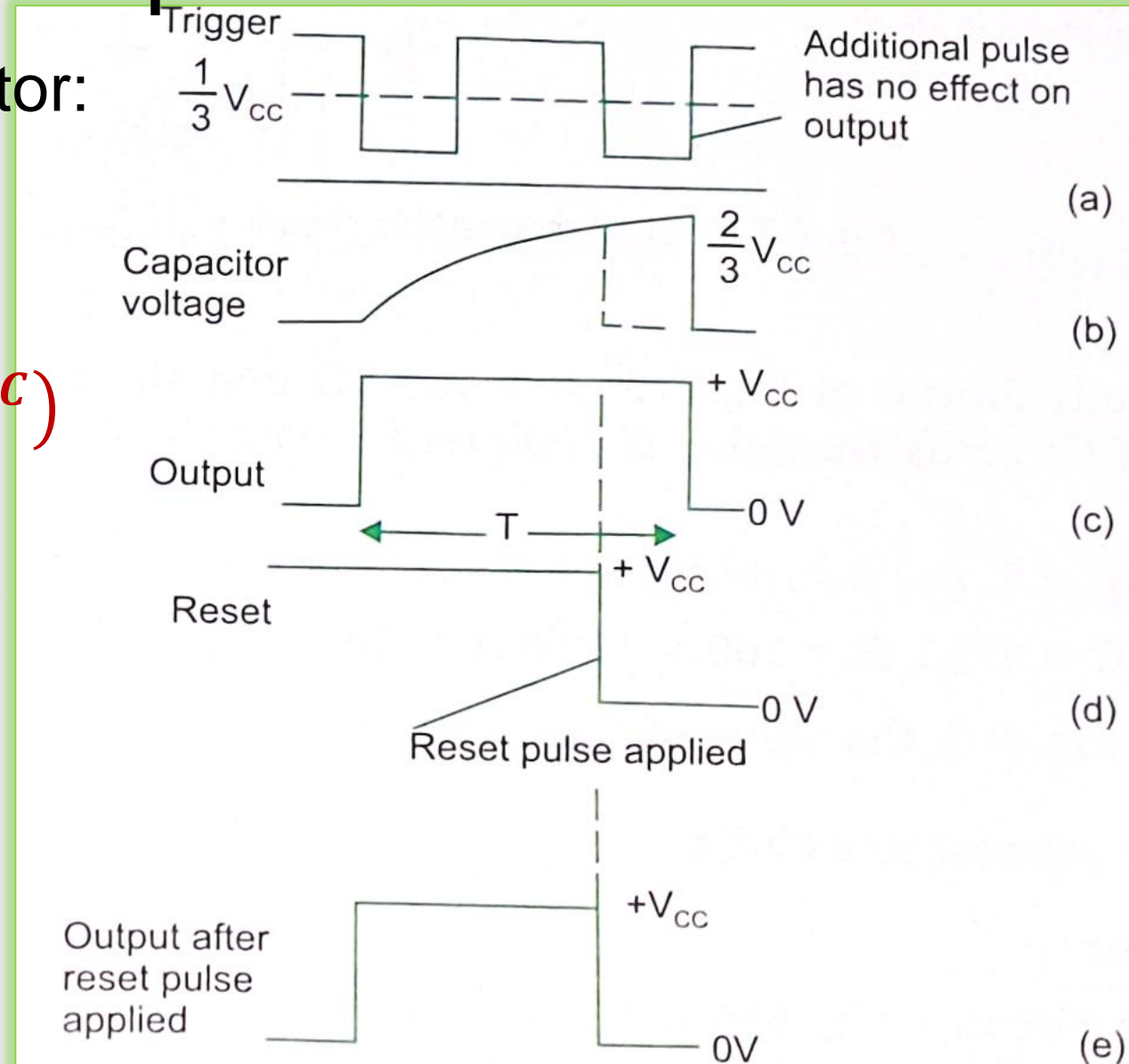
$$v_C = V_{CC}(1 - e^{-t/RC})$$

- At $t = T$, $v_C = \frac{2}{3}V_{CC}$

$$\frac{2}{3}V_{CC} = V_{CC}(1 - e^{-t/RC})$$

$$T = RC \ln(1/3)$$

$$T = 1.1 RC \text{ (seconds)}$$



3. Monostable Operation

- But, **when negative RESET input is given, Q_2 is ON,**

This saturates Q_1 and output will remain low until RESET is removed.

- When **RESET input goes high, until a negative trigger is initiated, Capacitor will not initiate charging and output remains low until the negative trigger input is given at pin 2.**

- then output will go to LOW

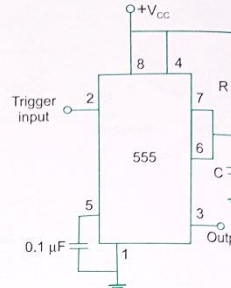
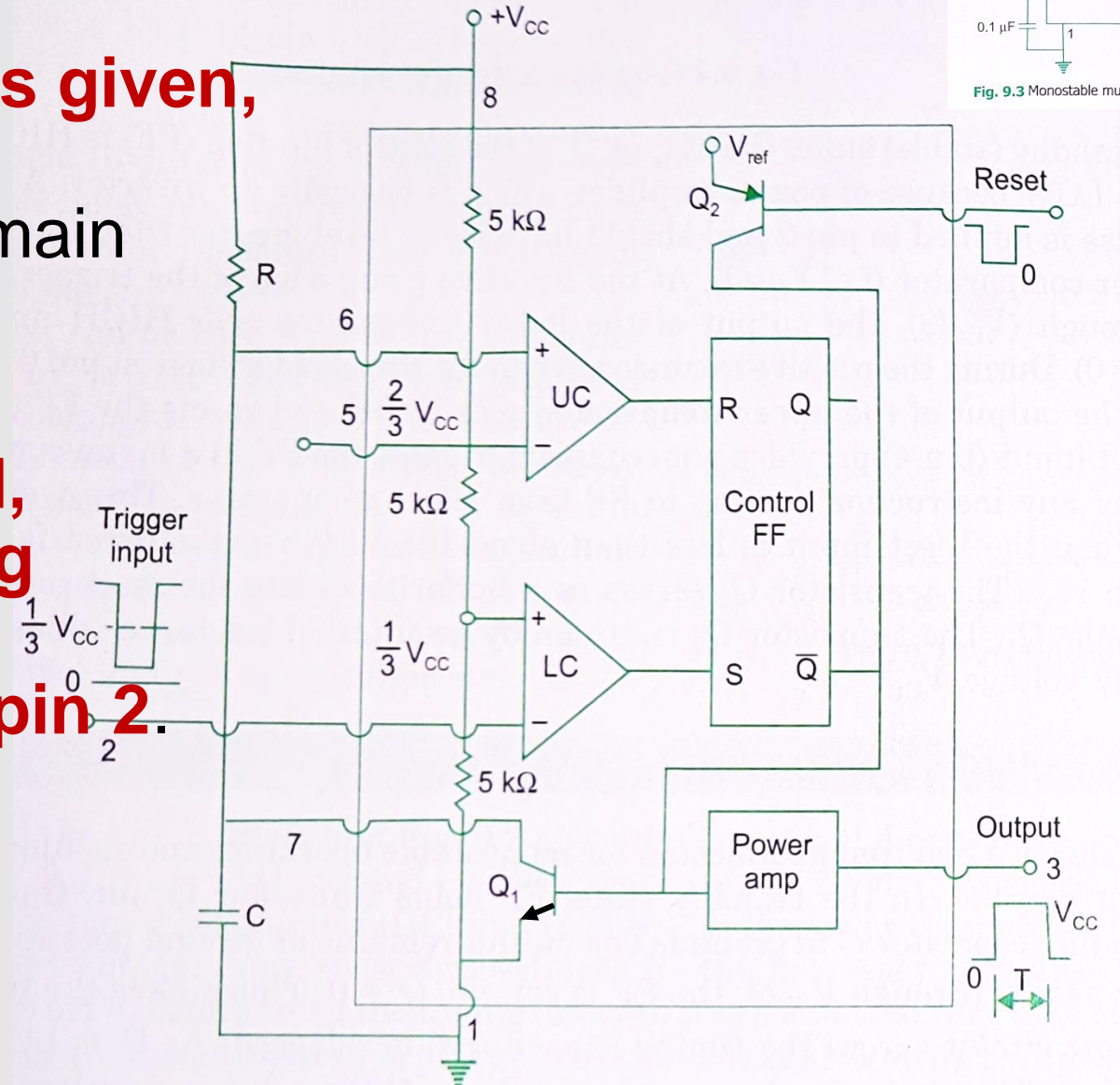


Fig. 9.3 Monostable multivibrator

3.2 Problem

- In monostable multivibrator, $R = 100k\Omega$, $T = 100\text{ ms}$. Calculate the value of C.
- $T = 1.1 RC$

$$C = \frac{T}{1.1R} = \frac{100 \times 10^{-3}}{1.1 \times 100 \times 10^3} = 0.9\mu F$$

4. Applications of Monostable mode:

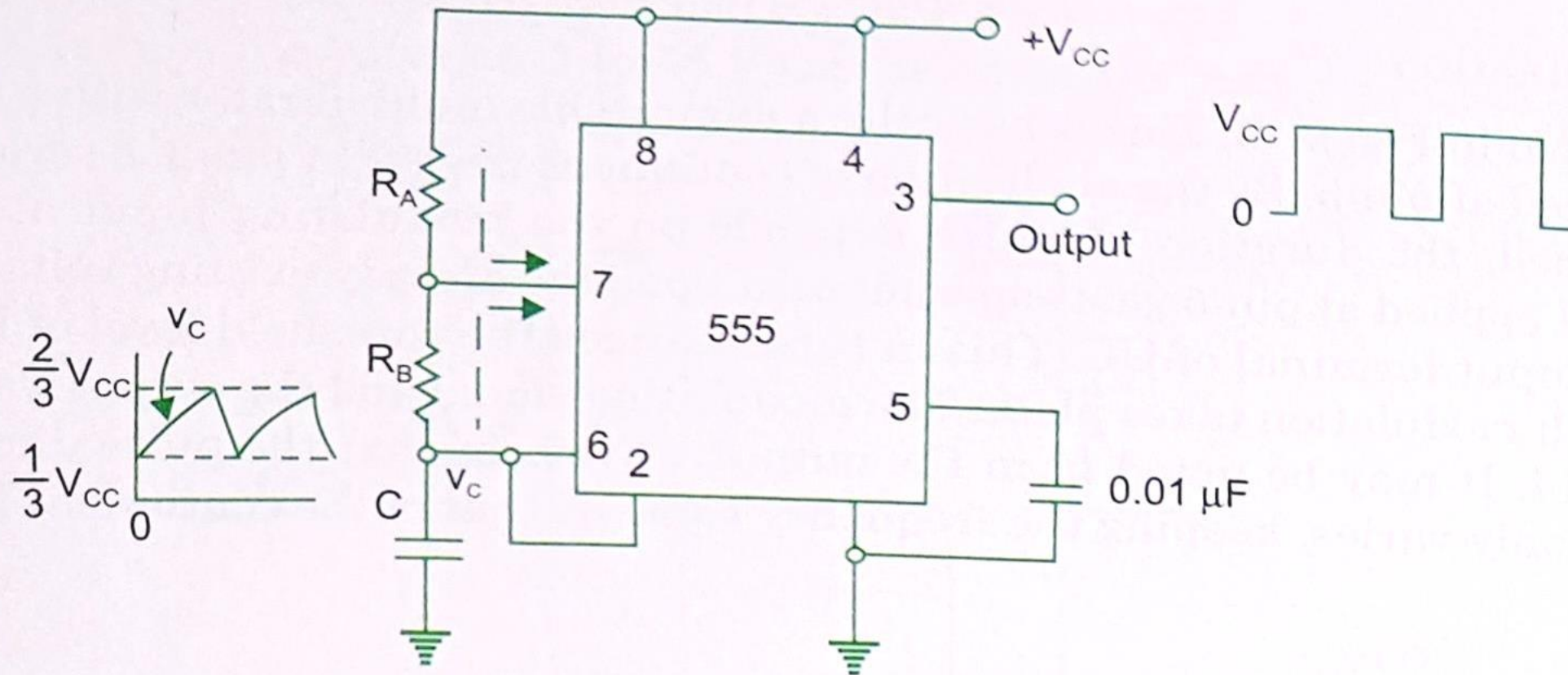
- Missing pulse detector
- Linear ramp generator
- Frequency divider
- Pulse width modulator

Astable Multivibrator

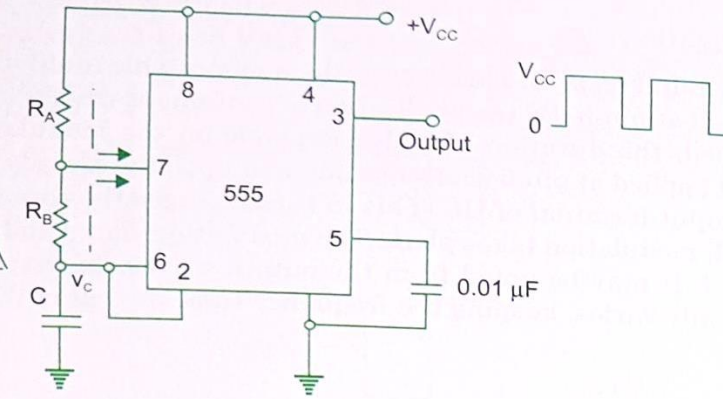
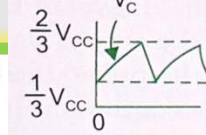
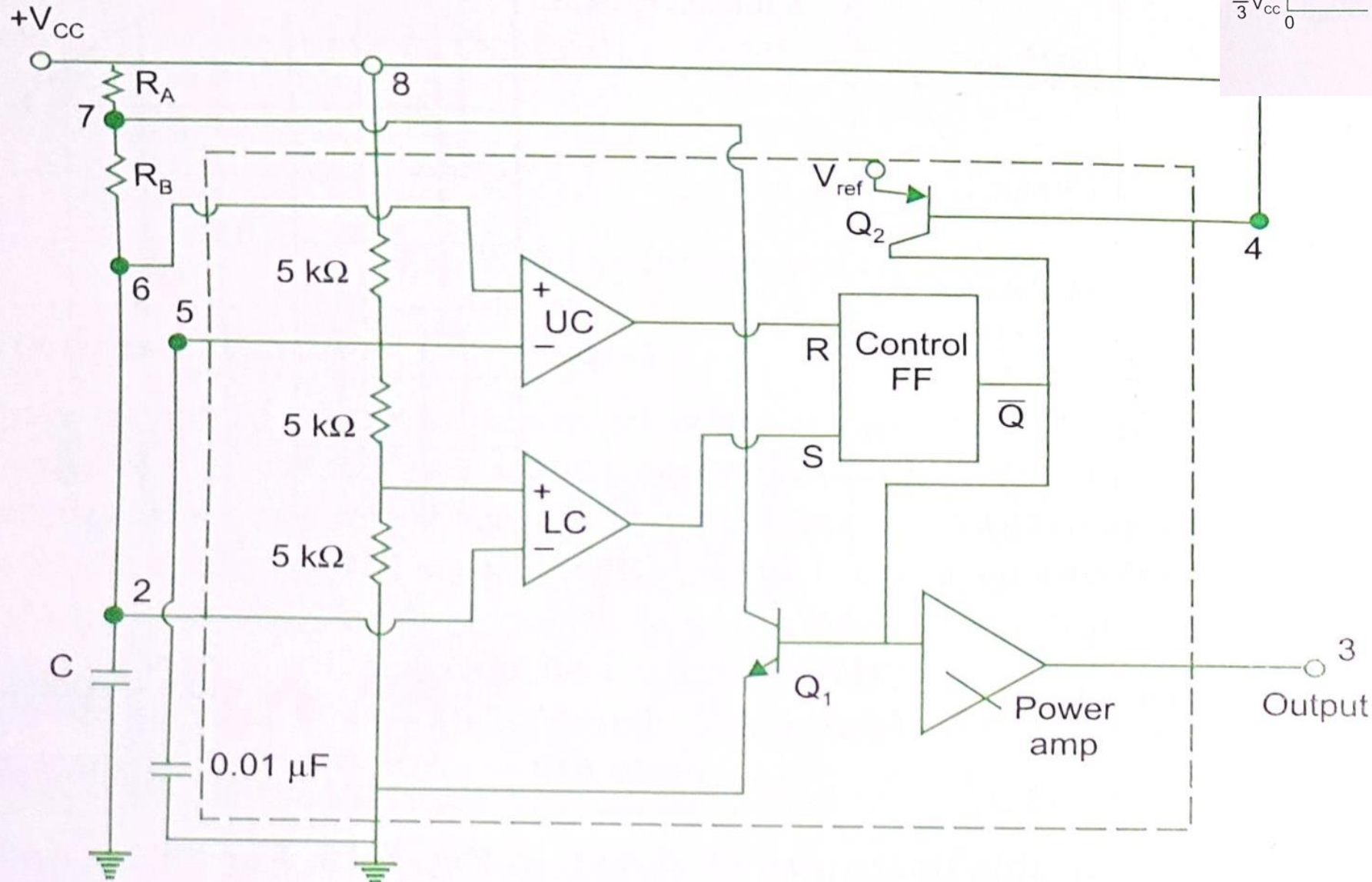
- **Timing resistor is split into two (R_A and R_B)**

Pin 7 (discharging transistor path) is connected between R_A and R_B

- When V_{CC} is connected, **Capacitor charges towards V_{CC} with time constant $(R_A + R_B)C$**

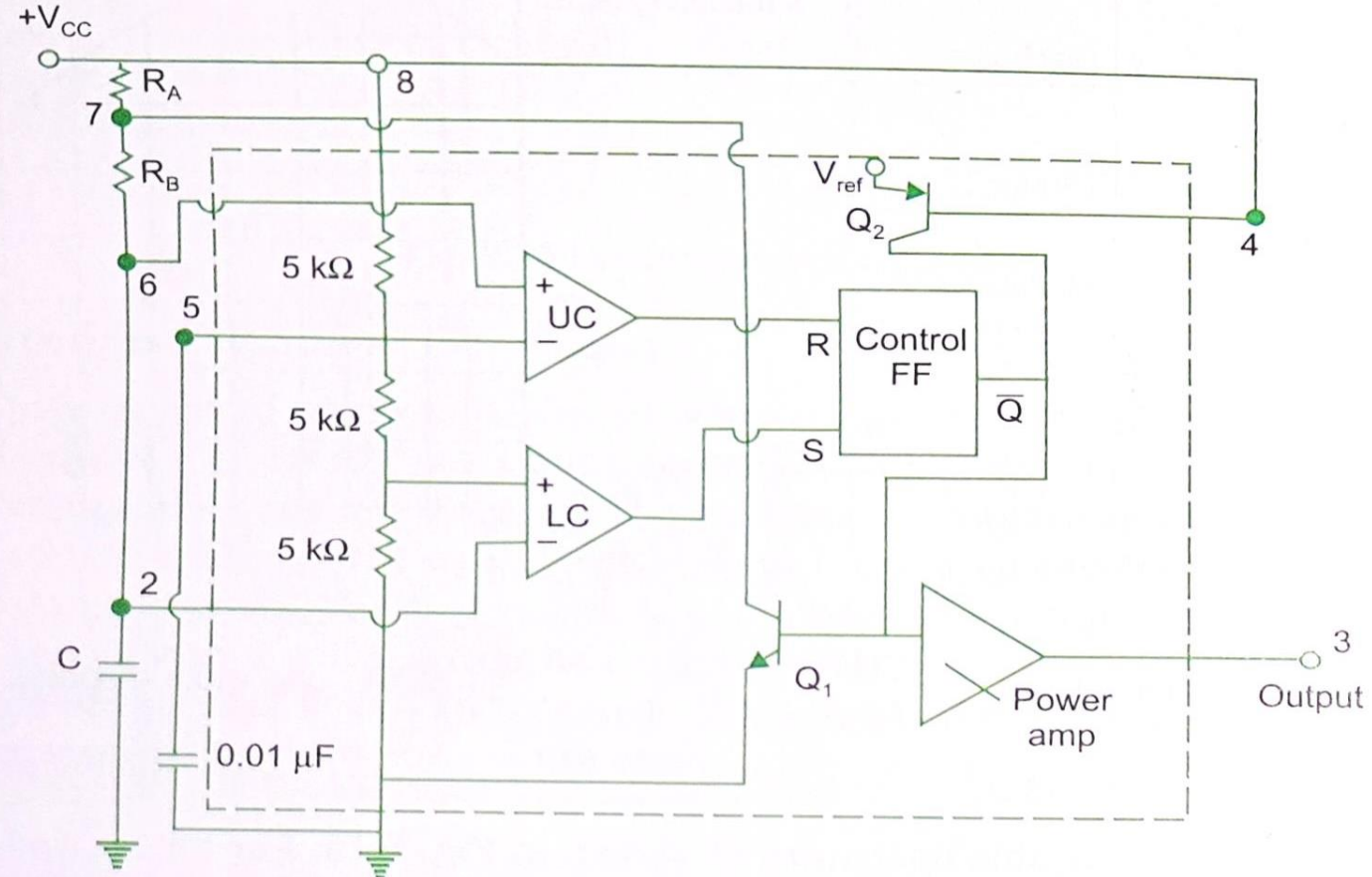
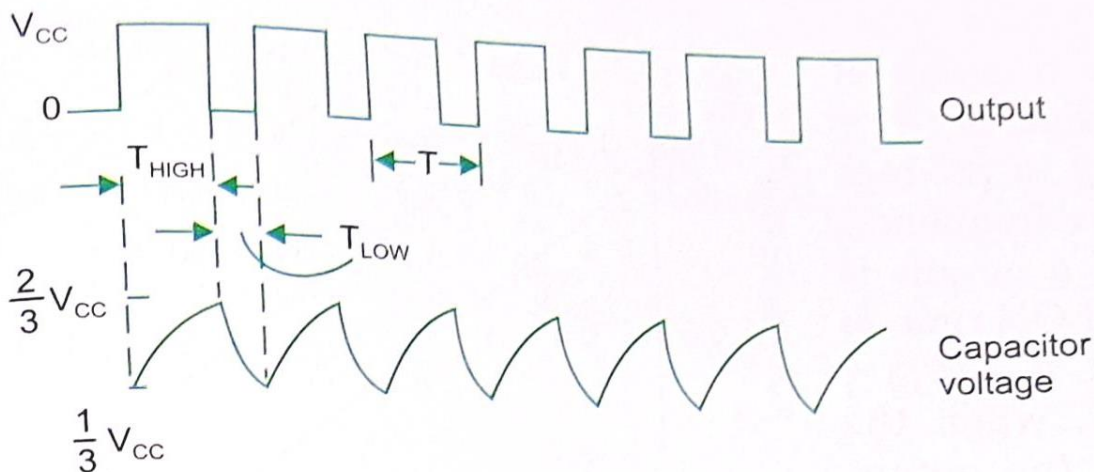
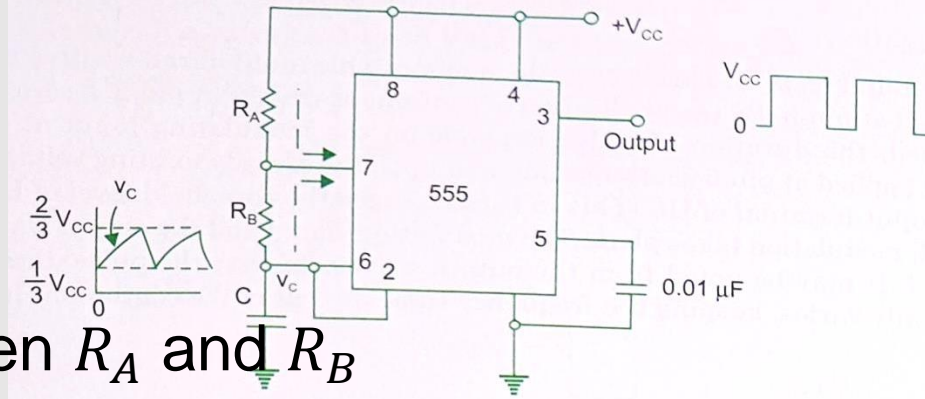


1. Astable Multivibrator



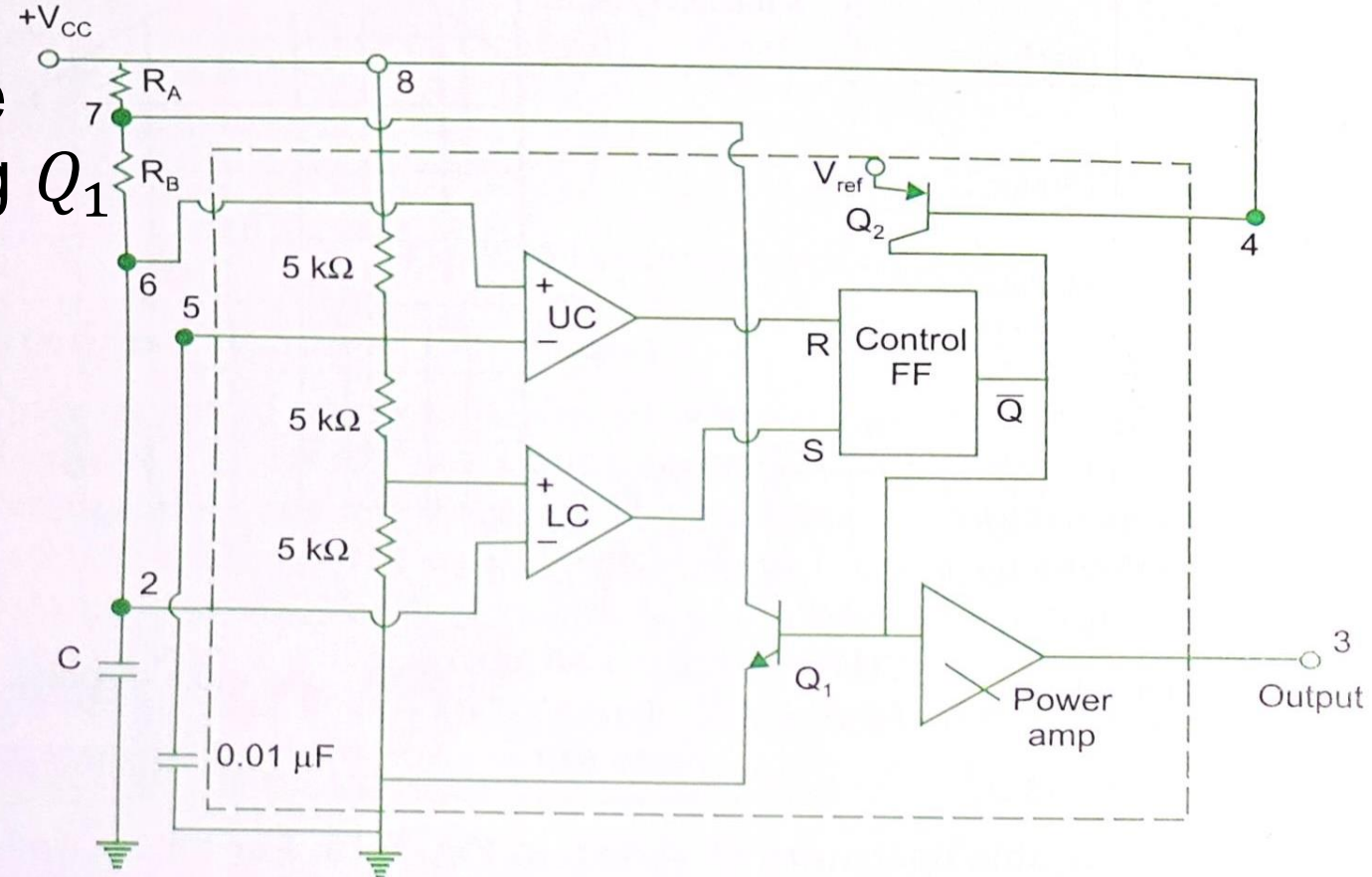
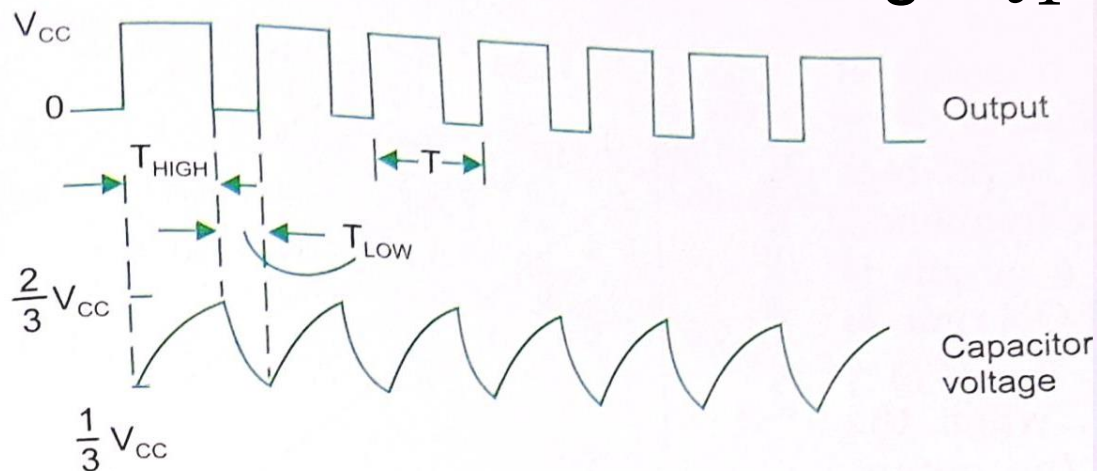
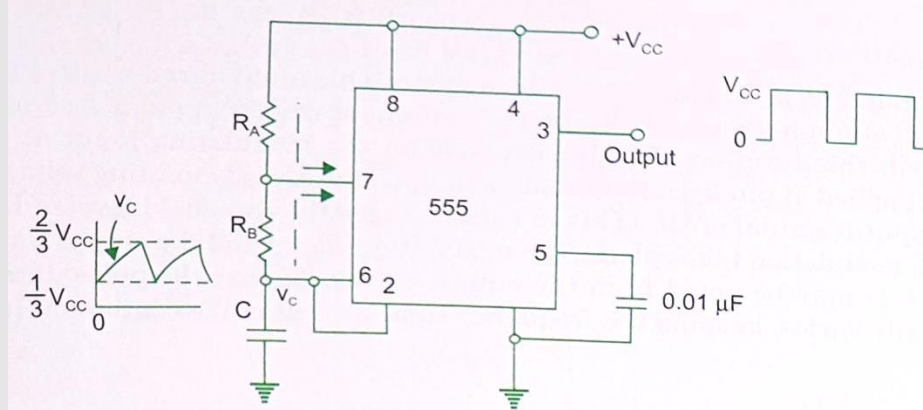
1. Astable Multivibrator

- Timing resistor is split into two (R_A and R_B)
Pin 7 (discharging transistor path) is connected between R_A and R_B
- When V_{CC} is connected, Capacitor charges towards V_{CC} with time constant $(R_A + R_B)C$. While charging, output pin 3 is HIGH, since $R = 0, S = 1, \bar{Q} = 0$, unclamped capacitor C (Q1 transistor not operational)



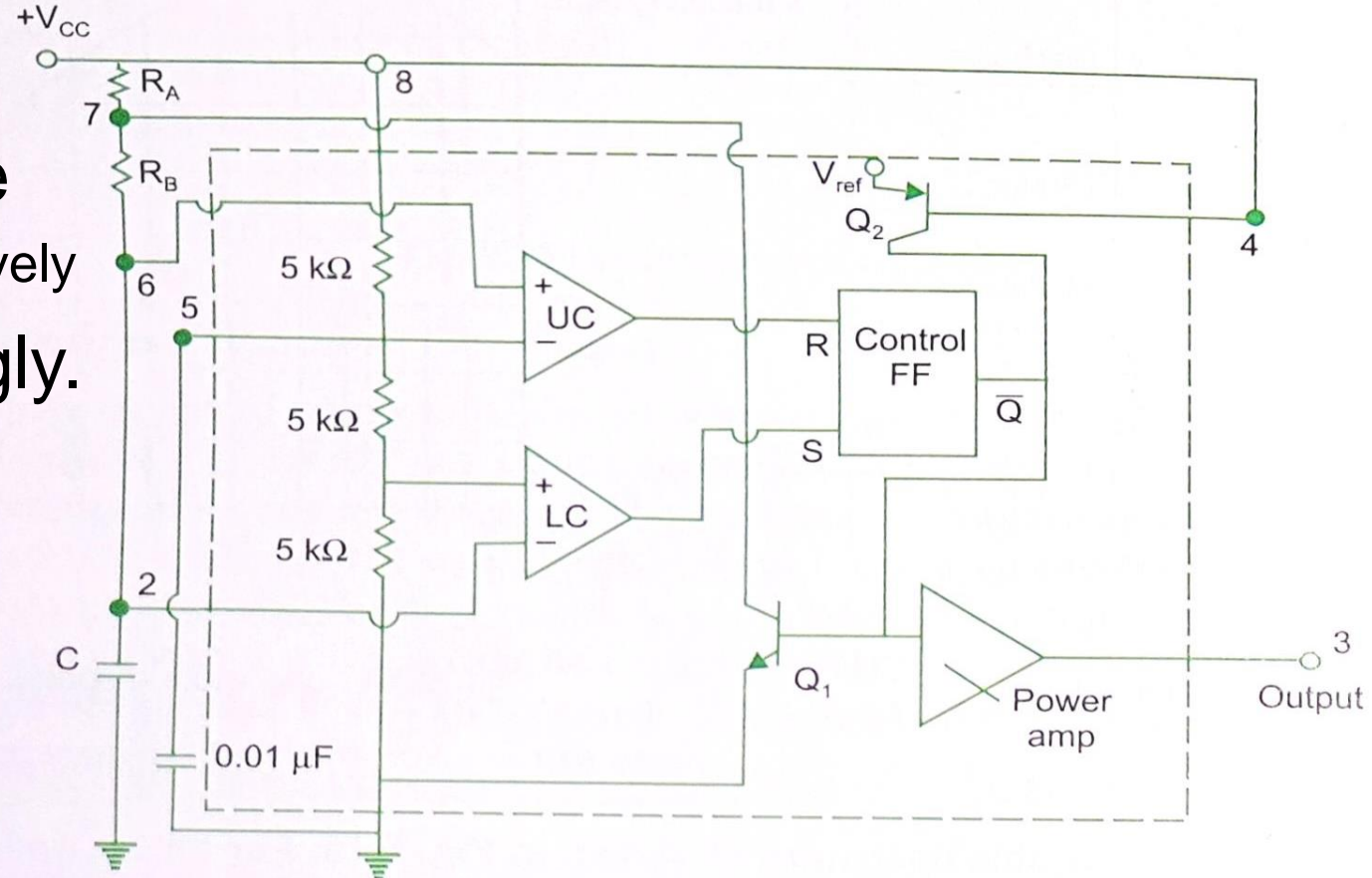
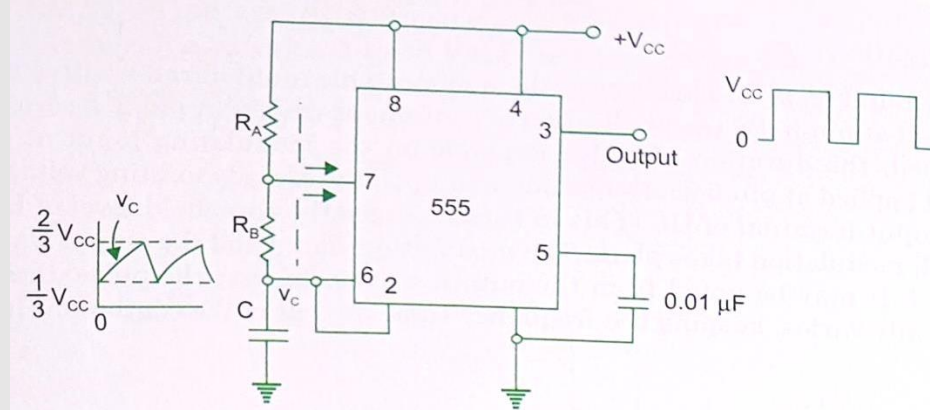
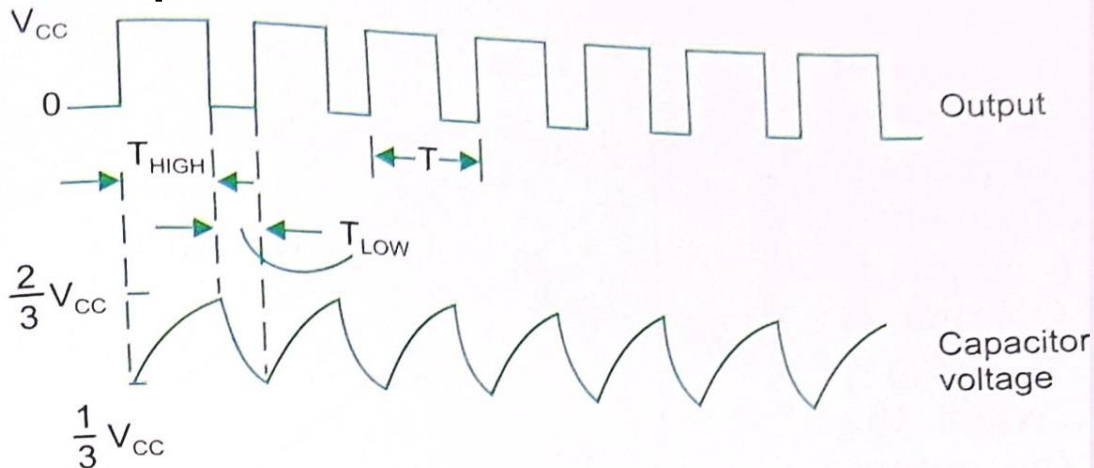
1. Astable Multivibrator

- When Capacitor voltage equals $2V_{CC}/3$, upper comparator triggers control **FF $\bar{Q} = 1$**
Transistor Q_1 will be ON, capacitor will discharge through R_B , Q_1 with time constant $R_B C$
- R_A, R_B are large values to reduce discharge current from damaging Q_1
- Min value of $R_A = V_{CC}/0.2A$
 $0.2A$ is max current through Q_1



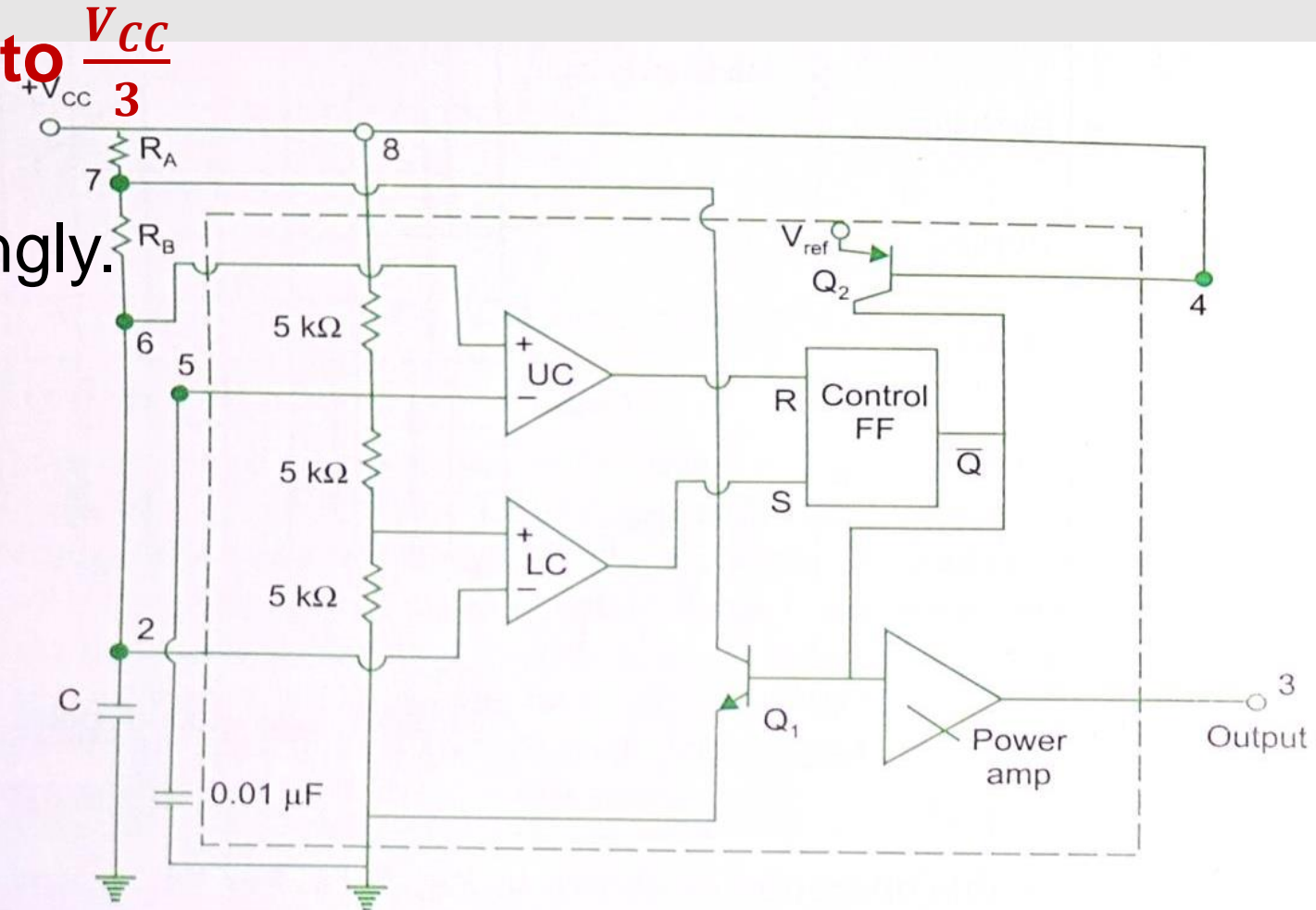
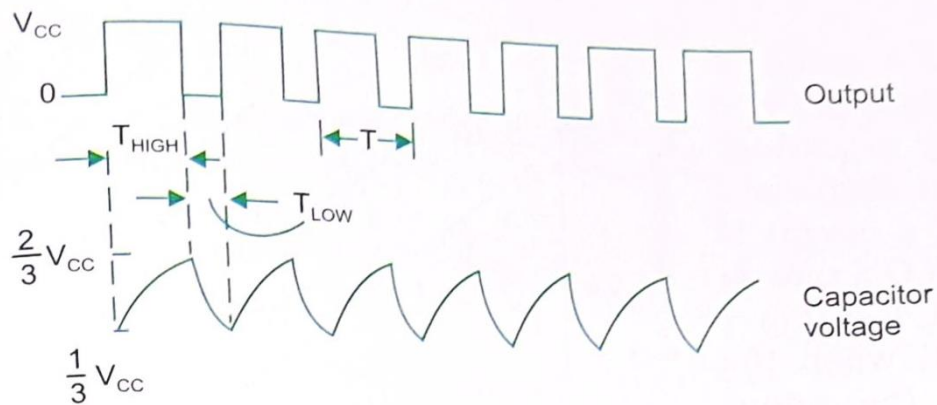
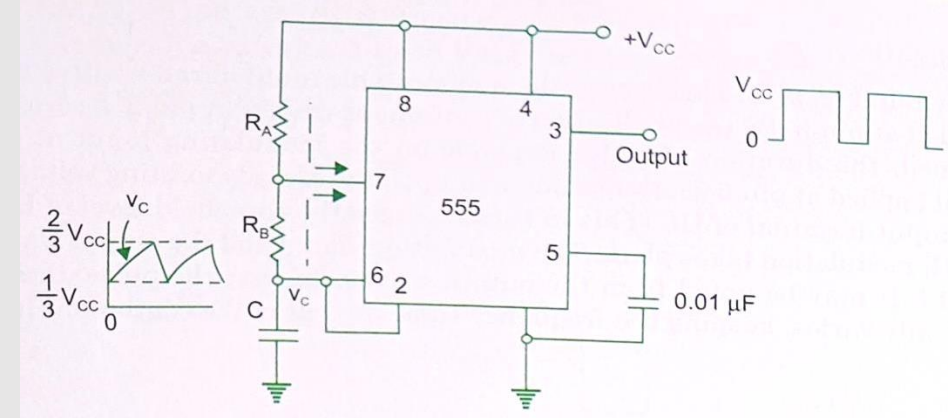
1. Astable Multivibrator

- During discharge of timing capacitor, when voltage of capacitor reaches $V_{CC}/3$, Lower comparator is triggered, $S=1$, $R=0$, $\bar{Q} = 0$, transistor Q_1 OFF. Stops discharge as external timing capacitor is unclamped.
- Capacitor now charges as before and discharges as above, Alternatively
- Output hence switches accordingly.



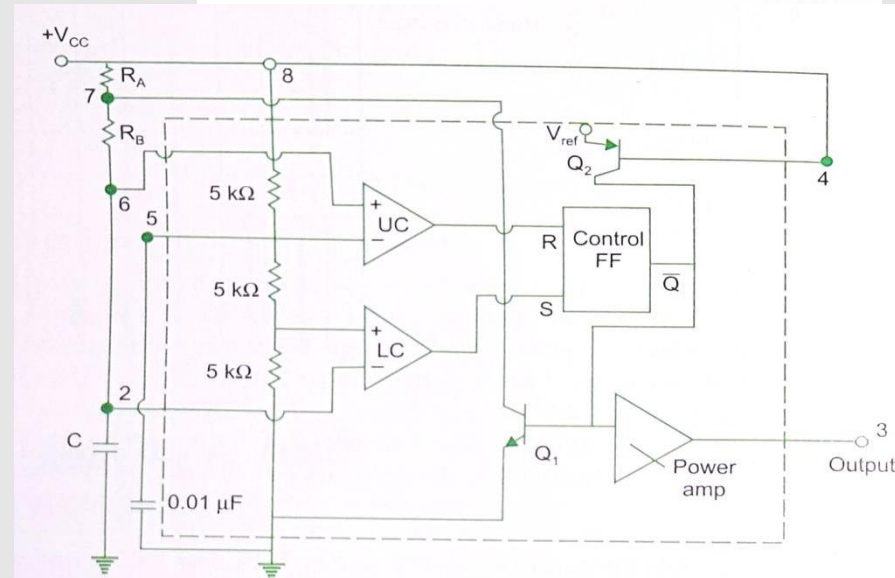
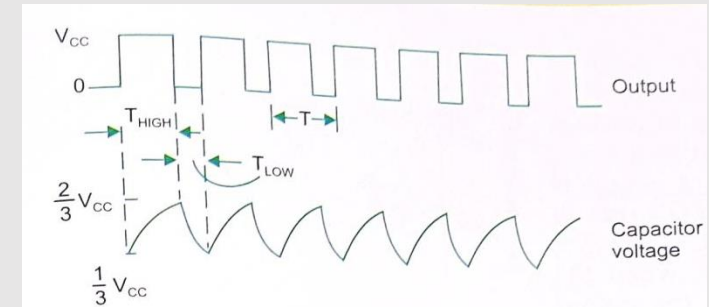
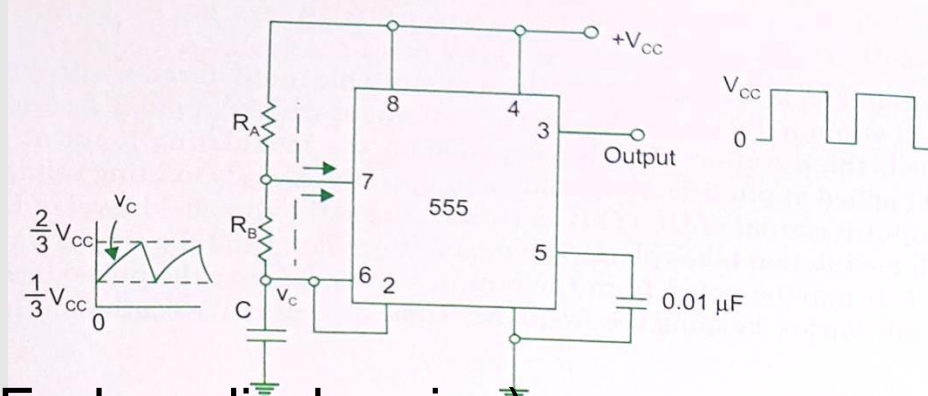
1. Astable Multivibrator

- Capacitor charges from $V_{CC}/3$ to $\frac{2V_{CC}}{3}$ through $R_A + R_B$ and discharges from $2V_{CC}/3$ to $\frac{V_{CC}}{3}$ through R_B
- Output hence switches accordingly. (ON while charging, OFF when discharging)



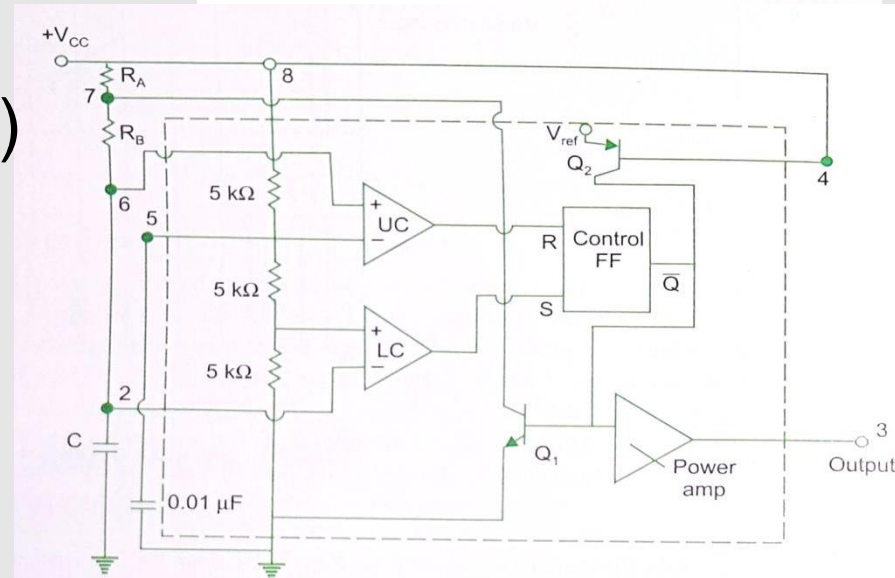
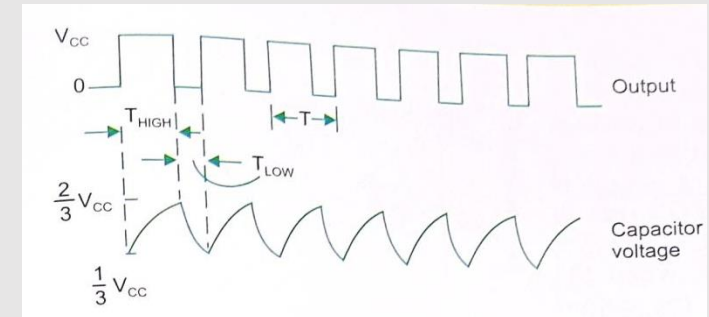
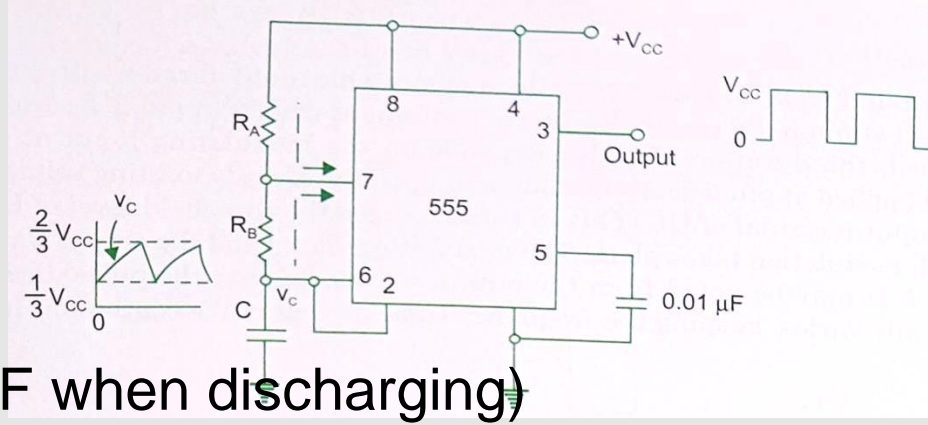
1. Astable Multivibrator

- Capacitor charges from $V_{CC}/3$ to $2V_{CC}/3$ through $R_A + R_B$ and discharges from $2V_{CC}/3$ to $V_{CC}/3$ through R_B
- Output hence switches accordingly. (ON while charging, OFF when discharging)
- Charging capacitor voltage $v_o = V_{CC}(1 - e^{-t/RC})$**
- t_1 : Time to charge from 0 upto $2V_{CC}/3$,
 t_2 : Time to charge from 0 upto $V_{CC}/3$,**
- $v_c(t_1) = \frac{2V_{CC}}{3} = V_{CC}(1 - e^{-t_1/RC}) \quad t_1 = 1.09RC$
 $v_c(t_2) = \frac{V_{CC}}{3} = V_{CC}(1 - e^{-t_2/RC}) \quad t_2 = 0.405RC$**
- $t_{HIGH} = t_1 - t_2$: Time to charge from $\frac{V_{CC}}{3}$ upto $2\frac{V_{CC}}{3}$,
 $= RC(1.09 - 0.405)$
 $= 0.69RC = 0.69(R_A + R_B)C$**



1. Astable Multivibrator

- Capacitor charges from $V_{CC}/3$ to $2V_{CC}/3$ through $R_A + R_B$ and discharges from $2V_{CC}/3$ to $V_{CC}/3$ through R_B
- Output hence switches accordingly. (ON while charging, OFF when discharging)
- $t_{HIGH} = 0.69(R_A + R_B)C$**
- D**ischarging capacitor voltage $v_c = V_{initial}(e^{-t/RC})$
- t_{LOW} : Time to discharge from $2V_{CC}/3$ upto $V_{CC}/3$,
- $v_c(t_{LOW}) = \frac{V_{CC}}{3} = \frac{2V_{CC}}{3}(e^{-t_{LOW}/RC})$
 $t_{LOW} = 0.69RC = 0.69R_B C$
- Time period (one cycle of charging and discharging)
 = Time period of output voltage
- $T = t_{HIGH} + t_{LOW} = 0.69(R_A + 2R_B)C$
- Frequency of oscillation $f = \frac{1}{T} = \frac{1.45}{(R_A + 2R_B)C}$



1. Astable Multivibrator

- $t_{HIGH} = 0.69(R_A + R_B)C$
- $t_{LOW} = 0.69R_B C$
- $T = t_{HIGH} + t_{LOW} = 0.69(R_A + 2R_B)C$
- $f = \frac{1}{T} = \frac{1.45}{(R_A + 2R_B)C}$
- Wrt transistor Q_1 , when ON, output goes low
- **Duty cycle $D\% = \frac{t_{LOW}}{T} \times 100\%$**

$$D\% = \frac{R_B}{(R_A + 2R_B)} \times 100\%$$

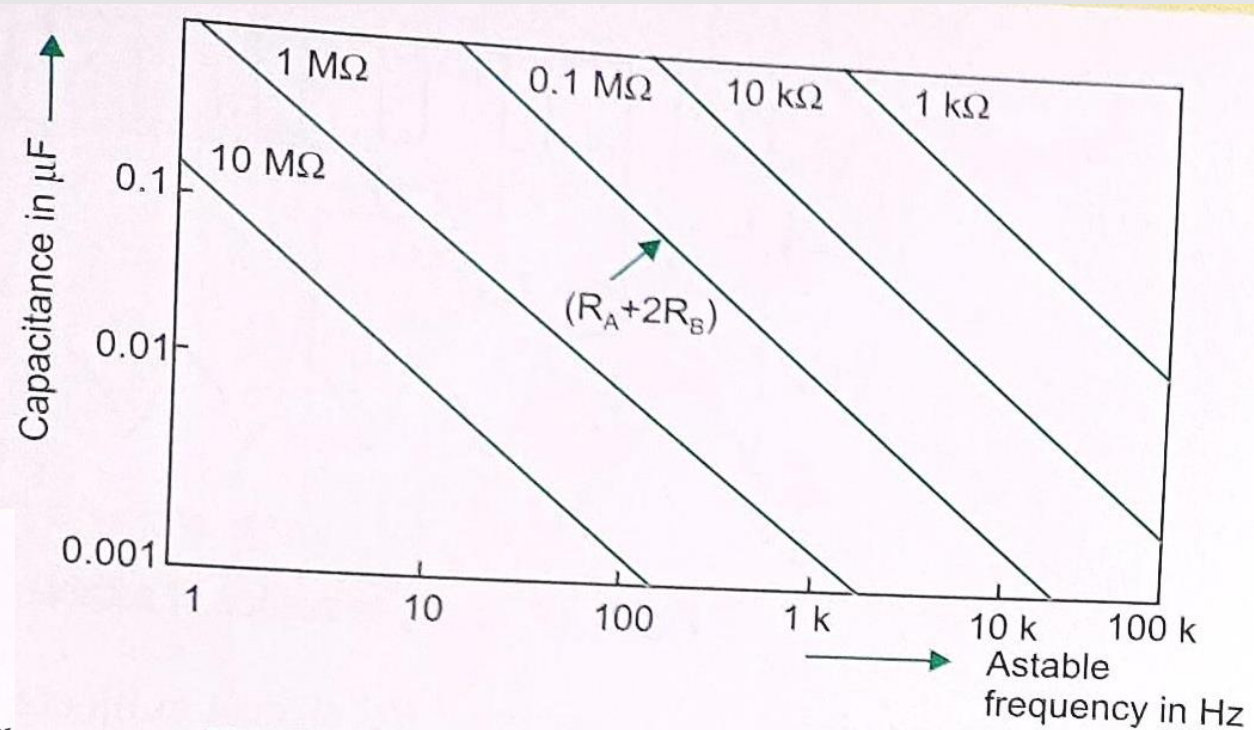
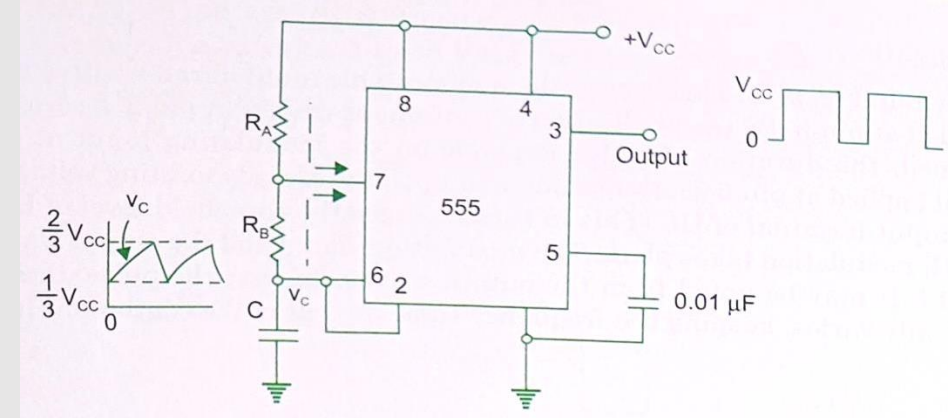
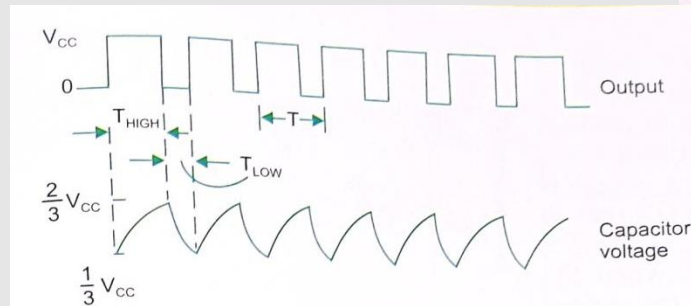


Fig. 9.18 Frequency dependence of R_A , R_B and C

2. Astable with adjustable duty cycle

- With this circuit
$$t_{HIGH} = 0.69R_A C$$
- $t_{LOW} = 0.69R_B C$
- $T = t_{HIGH} + t_{LOW} = 0.69(R_A + R_B)C$
- $f = \frac{1.45}{(R_A + R_B)C}$
- Duty cycle $D\% = \frac{R_B}{R_A + R_B} \times 100\%$

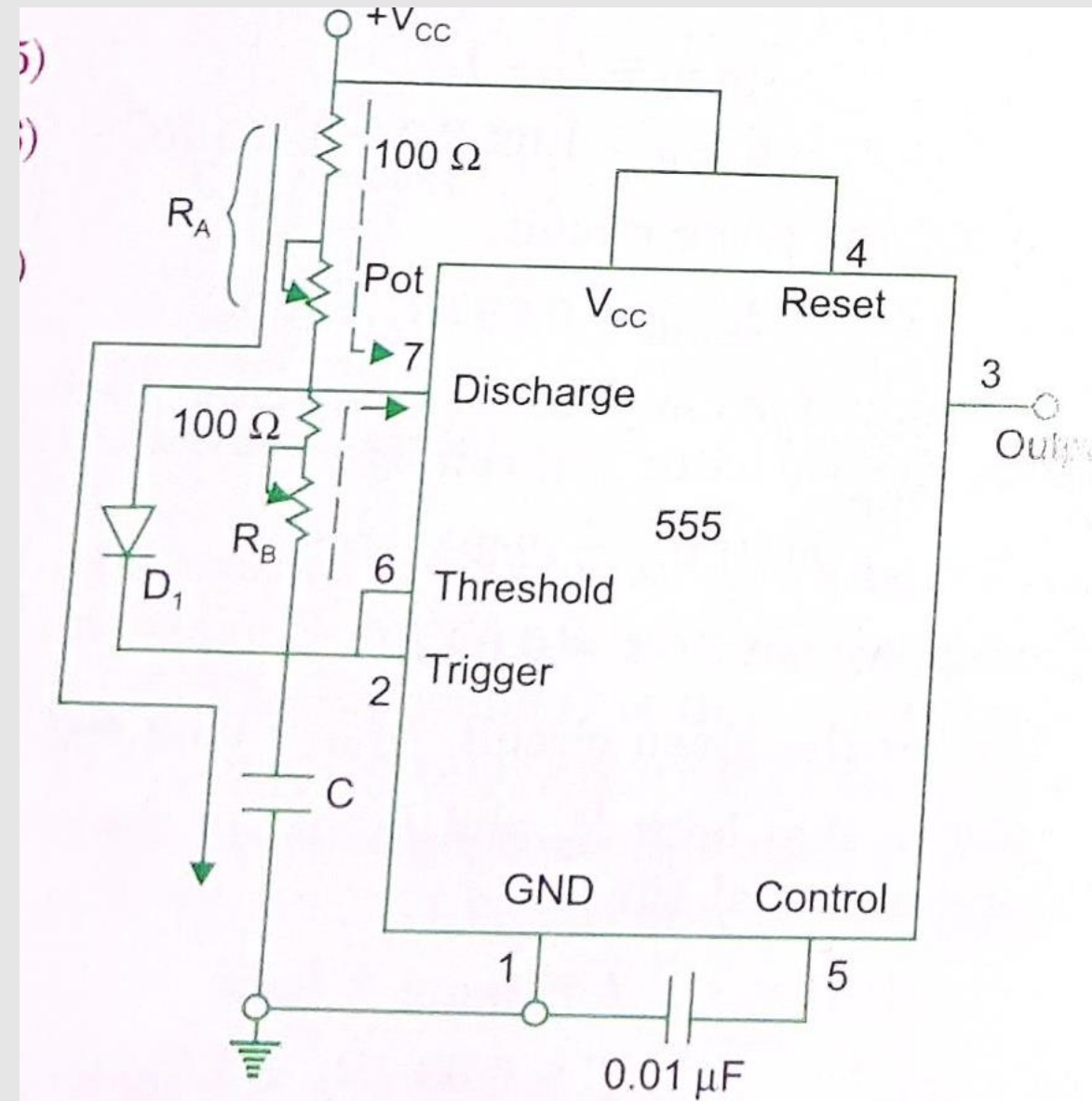


Fig. 9.19 Adjustable duty cycle rectangular wave generator

3.4 Applications of Astable operations

- Schmitt trigger

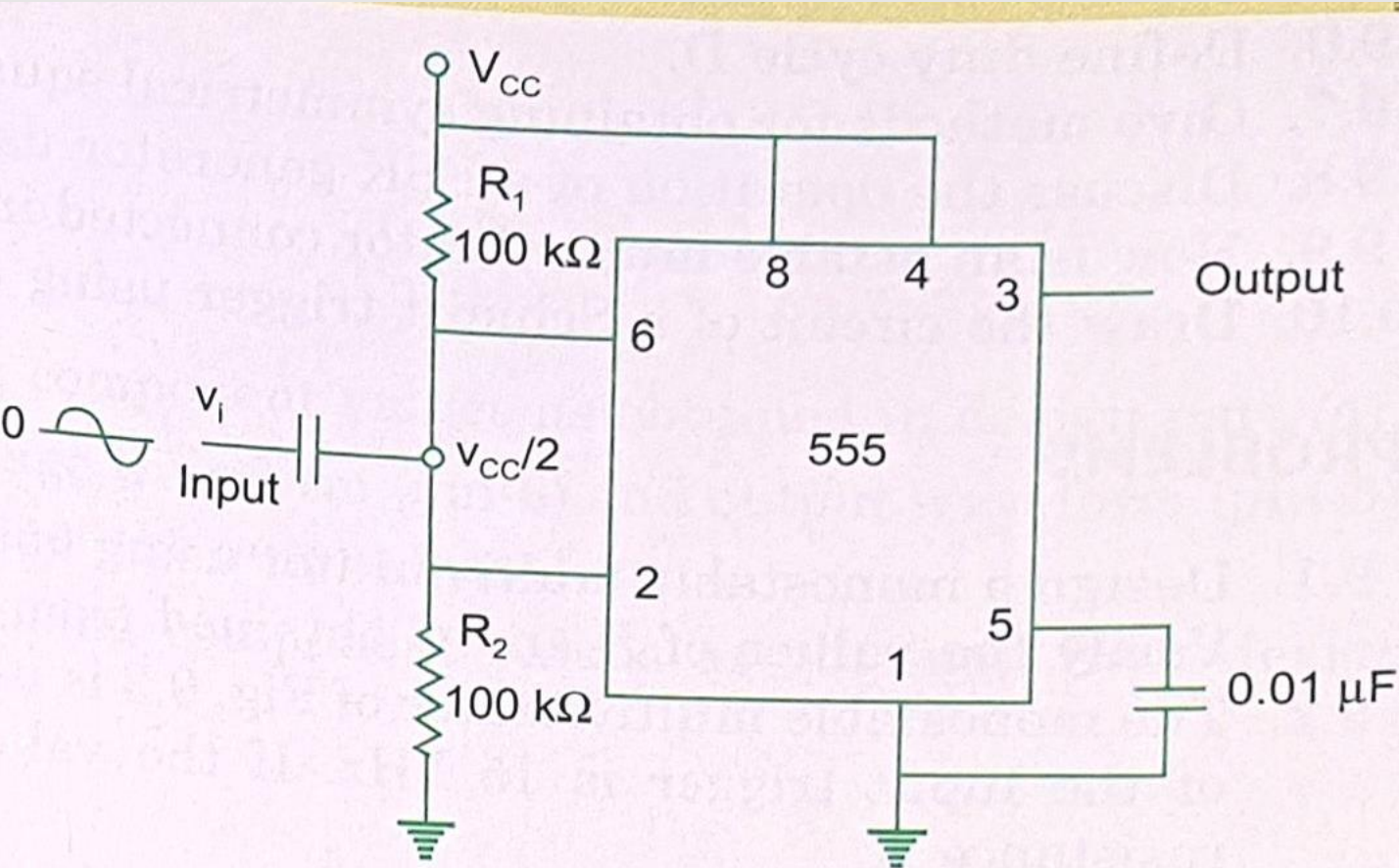


Fig. 9.24 Timer in Schmitt Trigger Operation

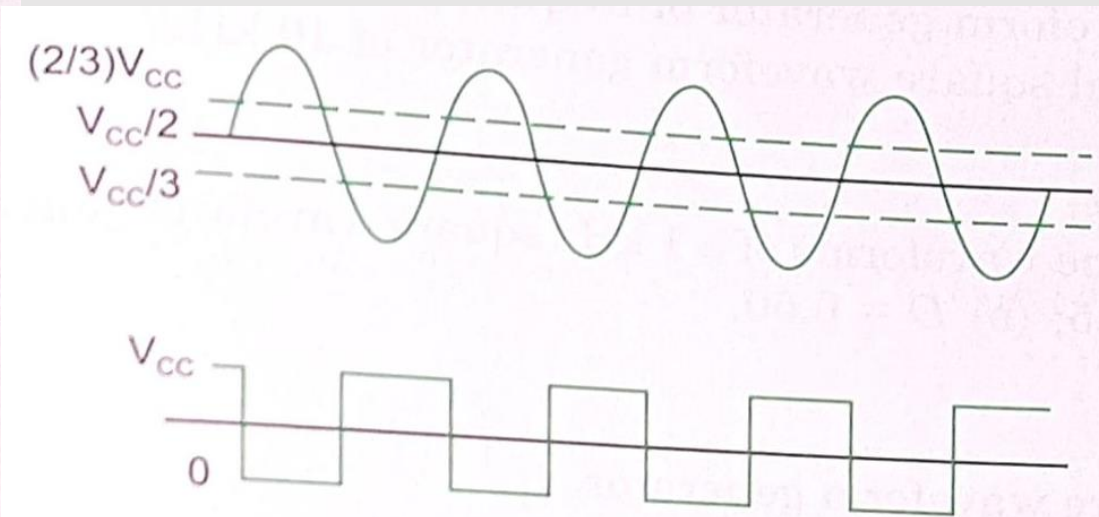


Fig. 9.25 Input output waveforms of Schmitt Trigger