

## Noise Analysis - AM, FM

The following assumptions are made:

- Channel model
  - distortionless
  - Additive White Gaussian Noise (AWGN)
- Receiver Model (see Figure 1)
  - ideal bandpass filter
  - ideal demodulator

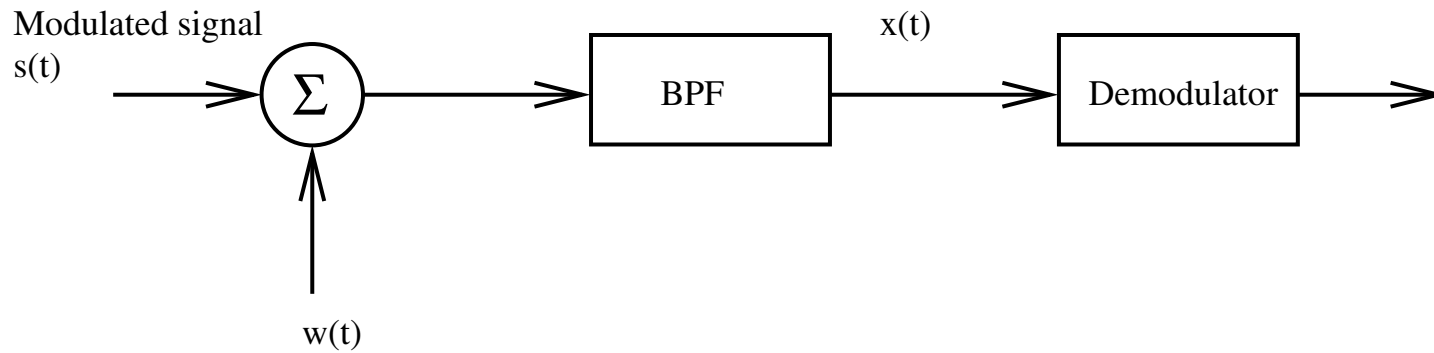


Figure 1: The Receiver Model

- BPF (Bandpass filter) - bandwidth is equal to the message bandwidth  $B$
- midband frequency is  $\omega_c$ .

### Power Spectral Density of Noise

- $\frac{N_0}{2}$ , and is defined for both positive and negative frequency (see Figure 2).
- $N_0$  is the average power/(unit BW) at the front-end of the

receiver in AM and DSB-SC.

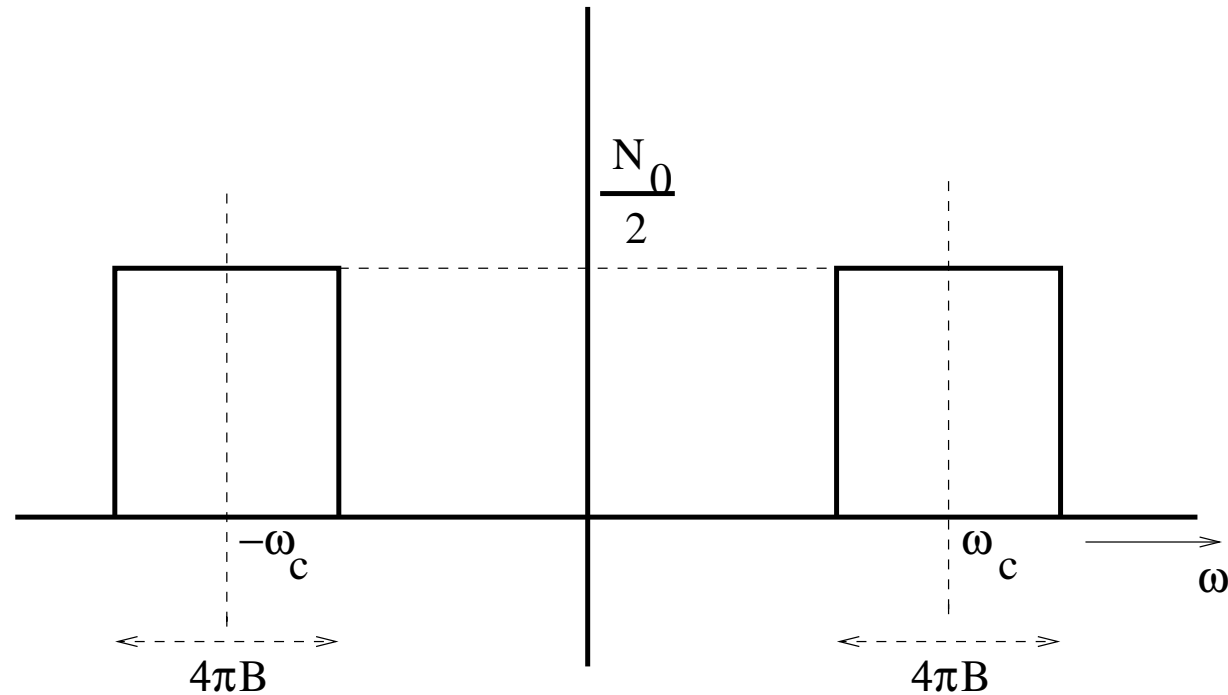


Figure 2: Bandlimited noise spectrum

The filtered signal available for demodulation is given by:

$$\begin{aligned}
 x(t) &= s(t) + n(t) \\
 n(t) &= n_I(t) \cos \omega_c t \\
 &\quad - n_Q(t) \sin \omega_c t
 \end{aligned}$$

$n_I(t) \cos \omega_c t$  is the in-phase component and

$n_Q(t) \sin \omega_c t$  is the quadrature component.

$n(t)$  is the representation for narrowband noise.

There are different measures that are used to define the Figure of Merit of different modulators:

- Input SNR:

$$(SNR)_I = \frac{\text{Average power of modulated signal } s(t)}{\text{Average power of noise}}$$

- Output SNR:

$$(SNR)_O = \frac{\text{Average power of demodulated signal } s(t)}{\text{Average power of noise}}$$

The Output SNR is measured at the receiver.

- Channel SNR:

$$(SNR)_C = \frac{\text{Average power of modulated signal } s(t)}{\text{Average power of noise in message bandwidth}}$$

- Figure of Merit (FoM) of Receiver:

$$FoM = \frac{(SNR)_O}{(SNR)_C}$$

To compare across different modulators, we assume that (see Figure 3):

- The modulated signal  $s(t)$  of each system has the same average power
- Channel noise  $w(t)$  has the same average power in the message bandwidth  $B$ .

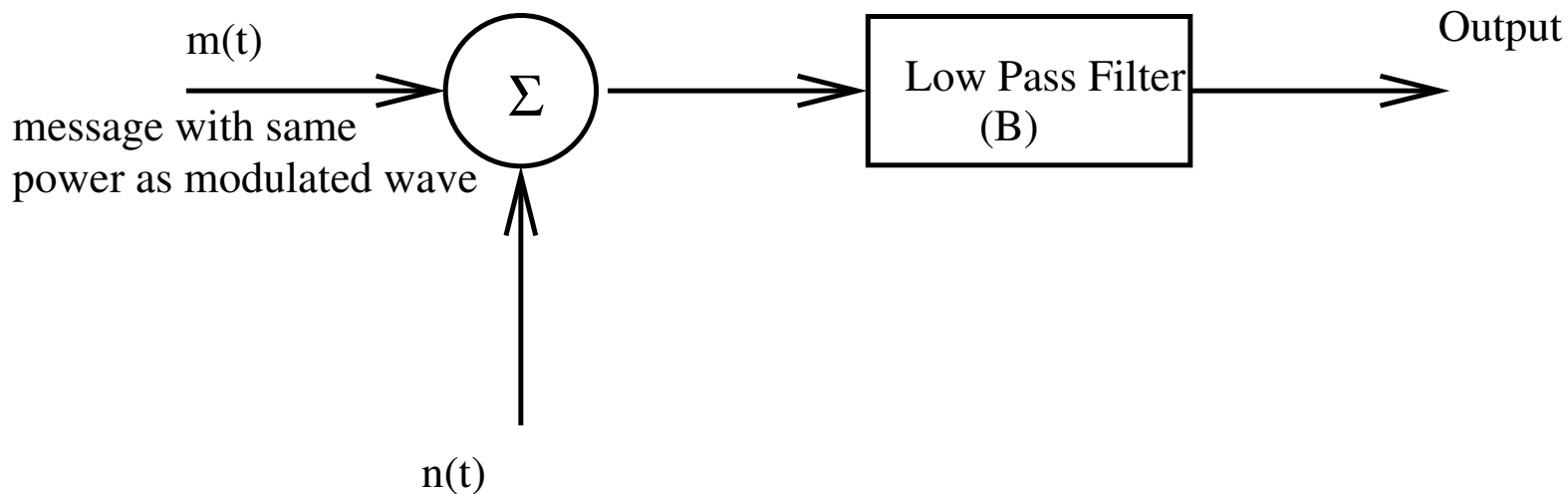


Figure 3: Basic Channel Model

## Figure of Merit (FoM) Analysis

- DSB-SC (see Figure 4)

$$s(t) = CA_c \cos(\omega_c t) m(t)$$

$$(SNR)_C = \frac{A_c^2 C^2 P}{2BN_0}$$

$$P = \int_{-2\pi B}^{+2\pi B} S_M(\omega) d\omega$$

$$\begin{aligned} x(t) &= s(t) + n(t) \\ &= CA_c \cos(\omega_c t) m(t) \\ &\quad + n_I(t) \cos \omega_c t + n_Q(t) \sin \omega_c t \end{aligned}$$

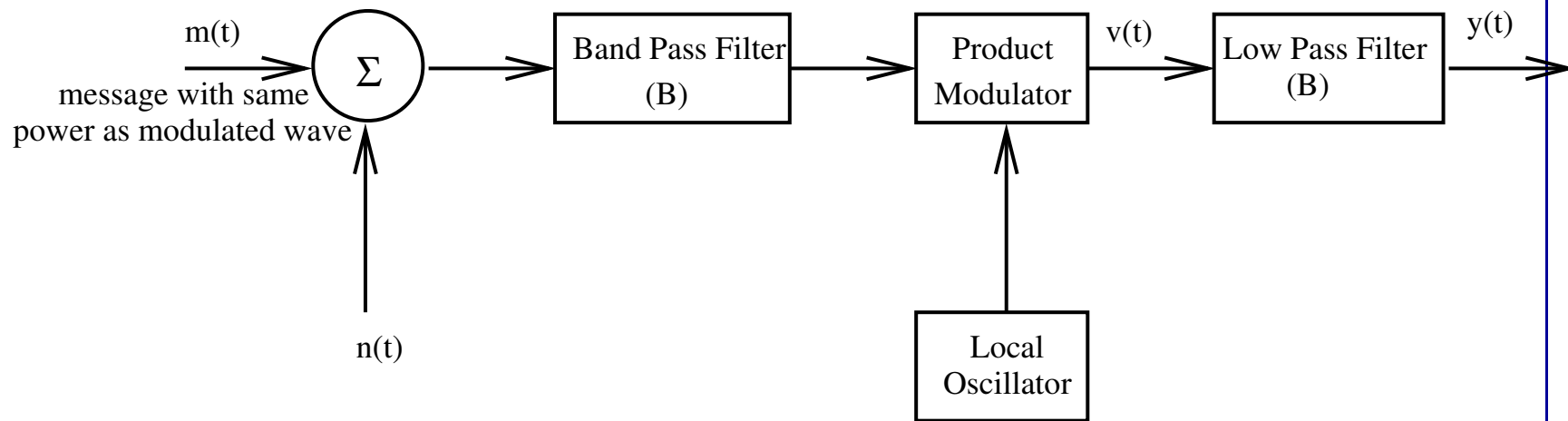


Figure 4: Analysis of DSB-SC System in Noise

The output of the product modulator is



$$\begin{aligned}
v(t) &= x(t) \cos(\omega_c t) \\
&= \frac{1}{2} A_c m(t) + \frac{1}{2} n_I(t) \\
&\quad + \frac{1}{2} [C A_c m(t) + n_I(t)] \cos 2\omega_c t \\
&\quad - \frac{1}{2} n_Q(t) \sin 2\omega_c t
\end{aligned}$$

The Low pass filter output is:

$$= \frac{1}{2} A_c m(t) + \frac{1}{2} n_I(t)$$

- $\implies$  ONLY inphase component of noise  $n_I(t)$  at the output
- $\implies$  Quadrature component of noise  $n_Q(t)$  is filtered at the output
- Band pass filter width =  $2B$

Receiver output is  $\frac{n_I(t)}{2}$

Average power of  $n_I(t)$  same as that  $n(t)$

$$\text{Average noise power} = \left(\frac{1}{2}\right)^2 2BN_0$$

$$= \frac{1}{2}BN_0$$

$$(SNR)_{O,DSB-SC} = \frac{C^2 A_c^2 P/4}{BN_0/2}$$

$$= \frac{C^2 A_c^2 P}{2BN_0}$$

$$FoM_{DSB-SC} = \left( \frac{(SNR)_O}{(SNR)_C} \right) |_{DSB-SC} = 1$$

- Amplitude Modulation

- The receiver model is as shown in Figure 5

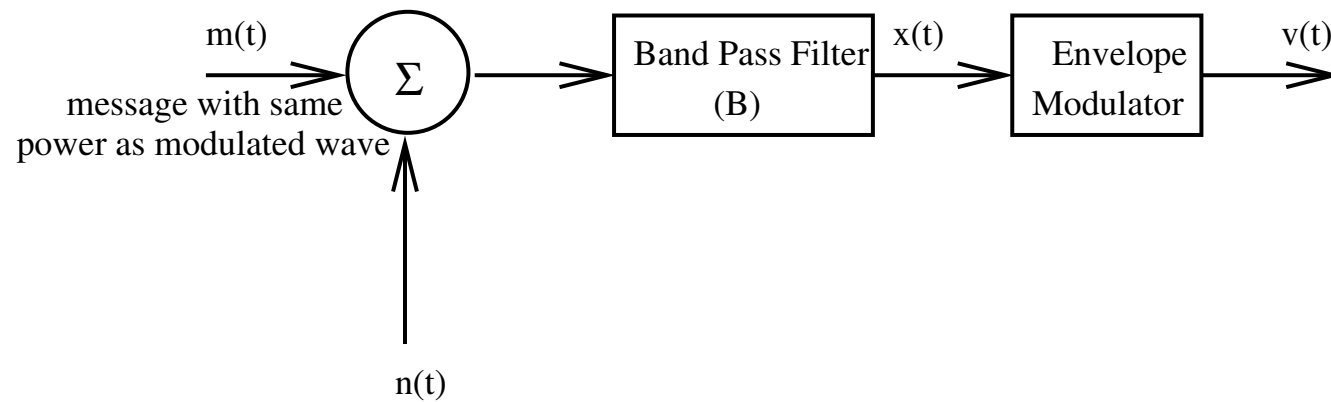


Figure 5: Analysis of AM System in Noise

$$\begin{aligned}
s(t) &= A_c[1 + k_a m(t)] \cos \omega_c t \\
(SNR)_{C,AM} &= \frac{A_c^2(1 + k_a^2 P)}{2BN_0} \\
x(t) &= s(t) + n(t) \\
&= [A_c + A_c k_a m(t) + n_I(t)] \cos \omega_c t \\
&\quad - n_Q(t) \sin \omega_c t \\
y(t) &= \text{envelope of } x(t) \\
&= \left[ [A_c + A_c k_a m(t) + n_I(t)]^2 + n_Q^2(t) \right]^{\frac{1}{2}} \\
&\approx A_c + A_c k_a m(t) + n_I(t) \\
(SNR)_{O,AM} &\approx \frac{A_c^2 k_a^2 P}{2BN_0} \\
FoM_{AM} &= \left( \frac{(SNR)_O}{(SNR)_C} \right) |_{AM} = \frac{k_a^2 P}{1 + k_a^2 P}
\end{aligned}$$

Thus the  $FoM_{AM}$  is always inferior to  $FoM_{DSB-SC}$

– Frequency Modulation

- \* The analysis for FM is rather complex
- \* The receiver model is as shown in Figure 6

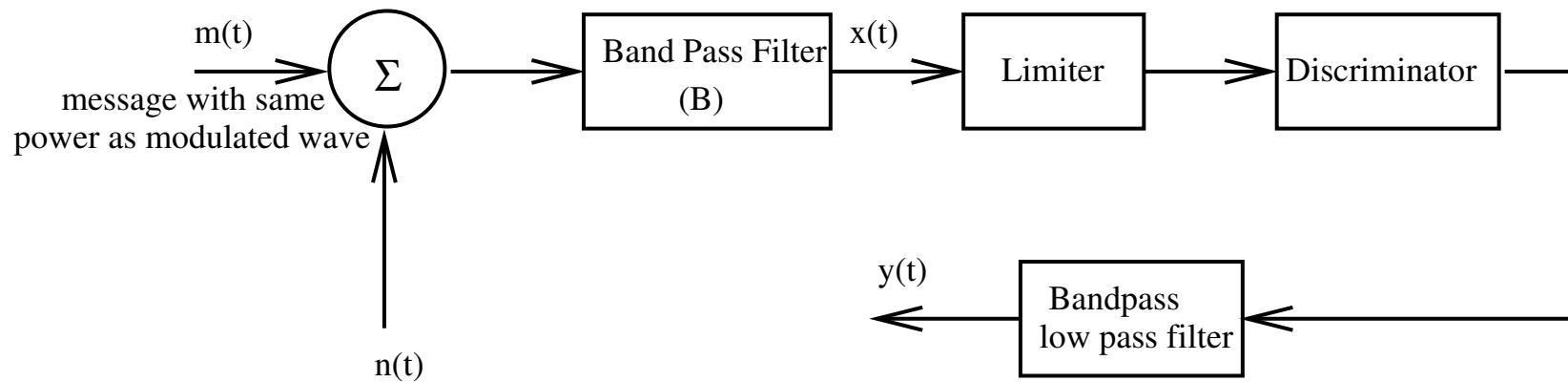


Figure 6: Analysis of FM System in Noise

$$(SNR)_{O,FM} = \frac{3A_c^2 k_f^2 P}{2N_0 B^3}$$

$$(SNR)_{C,FM} = \frac{A_c^2}{2BN_0}$$

$$FoM_{FM} = \left( \frac{(SNR)_O}{(SNR)_C} \right) |_{FM} = \frac{3k_f^2 P}{B^2}$$

The significance of this is that *when the carrier SNR is high, an increase in transmission bandwidth  $B_T$  provides a corresponding quadratic increase in output SNR or  $FoM_{FM}$*