#### 2.5 Circular Array

# **Module:2 Linear and Planar Arrays**Course: BECE305L – Antenna and Microwave Engineering

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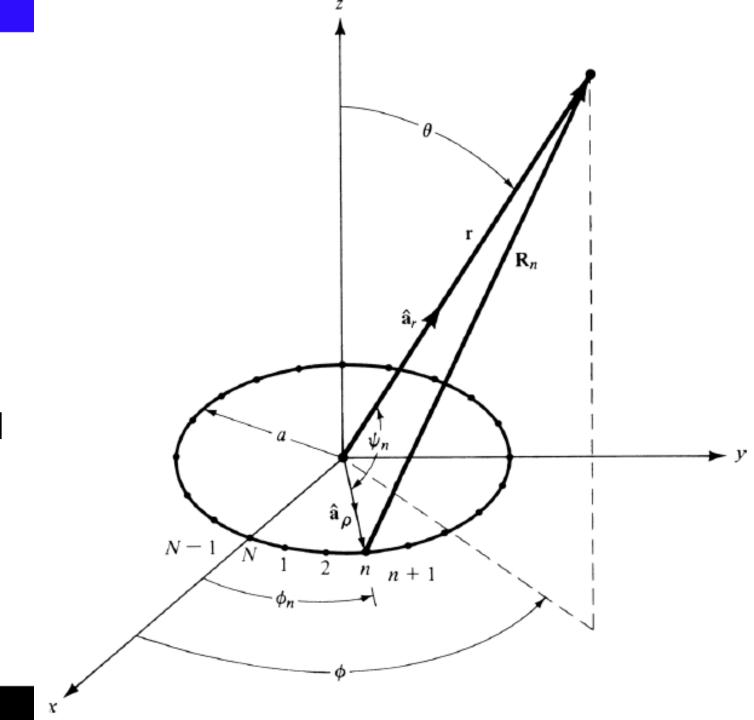
## **Module:2** Linear and Planar Arrays

• Two element array, N-element linear array - broadside array, End fire array - Directivity, radiation pattern, pattern multiplication. Non-uniform excitation - Binomial, Chebyshev distribution, Arrays: Planar array, circular array, Phased Array antenna (Qualitative study).

 Source of the contents: Constantine A. Balanis - Antenna theory analysis and design (2016)

### 1. Circular array

- The circular array, in which the elements are placed in a circular ring.
- Applications: radio direction finding, air and space navigation, underground propagation, radar, sonar, wireless communication, and in particular for smart antennas
- N isotropic elements are equally spaced on the x-y plane along a circular ring of the radius a.



#### 2.5 Circular Array

 normalized field of the array Rn is the distance from the nth element to the observation point  $E_n(r, \theta, \phi) = \sum_{n=1}^{N} a_n \frac{e^{-jkR_n}}{R_n}$ 

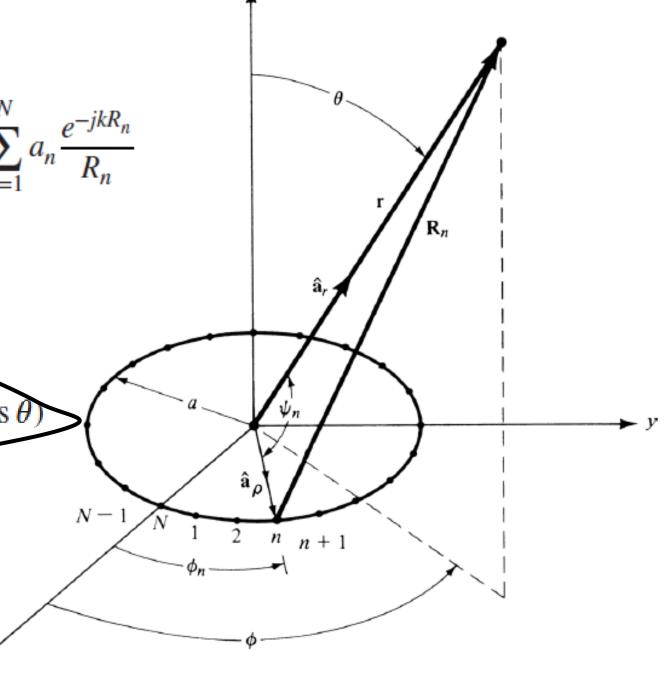
$$E_n(r,\theta,\phi) = \sum_{n=1}^{N} a_n \frac{e^{-jkR_n}}{R_n}$$

• for  $r \gg a$ 

$$R_n \simeq r - a\cos\psi_n = r - a(\hat{\mathbf{a}}_\rho \cdot \hat{\mathbf{a}}_r)$$
$$= r - a\sin\theta\cos(\phi - \phi_n)$$

$$\hat{\mathbf{a}}_{\rho} \cdot \hat{\mathbf{a}}_{r} = (\hat{\mathbf{a}}_{x} \cos \phi_{n} + \hat{\mathbf{a}}_{y} \sin \phi_{n}) \cdot (\hat{\mathbf{a}}_{x} \sin \theta \cos \phi + \hat{\mathbf{a}}_{y} \sin \theta \sin \phi + \hat{\mathbf{a}}_{z} \cos \theta)$$
$$= \sin \theta \cos(\phi - \phi_{n})$$

$$E_n(r,\theta,\phi) = \frac{e^{-jkr}}{r} \sum_{n=1}^{N} a_n e^{+jka\sin\theta\cos(\phi-\phi_n)}$$



$$E_n(r,\theta,\phi) = \frac{e^{-jkr}}{r} \sum_{n=1}^{N} a_n e^{+jka\sin\theta\cos(\phi - \phi_n)}$$

 $a_n$  = excitation coefficients (amplitude and phase) of nth element

$$\phi_n = 2\pi \left(\frac{n}{N}\right)$$
 = angular position of *n*th element on *x*-*y* plane

• With excitation coefficient  $a_n = I_n e^{j\alpha_n}$ 

 $I_n$  = amplitude excitation of the *n*th element

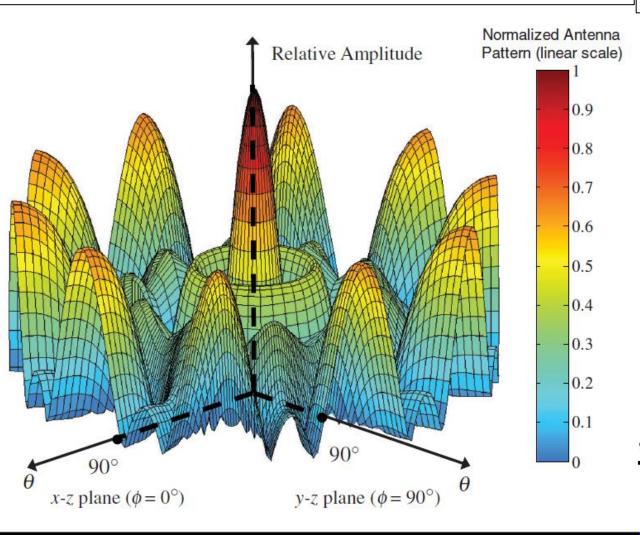
 $\alpha_n$  = phase excitation (relative to the array center) of the *n*th element

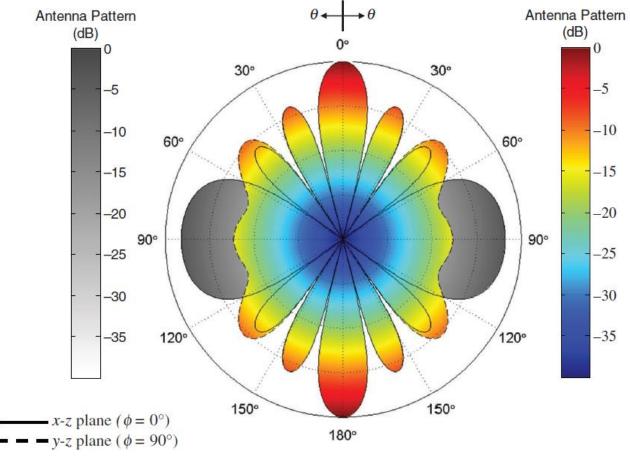
$$E_n(r,\theta,\phi) = \frac{e^{-jkr}}{r} [\text{AF}(\theta,\phi)] \qquad \qquad \xi = \tan^{-1} \left[ \frac{\sin\theta\sin\phi - \sin\theta_0\sin\phi_0}{\sin\theta\cos\phi - \sin\theta_0\cos\phi_0} \right]$$
 
$$\text{AF}(\theta,\phi) = \sum_{n=1}^{N} I_n e^{j[ka\sin\theta\cos(\phi - \phi_n) + \alpha_n]} = \sum_{n=1}^{N} I_n e^{jka(\cos\psi - \cos\psi_0)} = \sum_{n=1}^{N} I_n e^{jk\rho_0\cos(\phi_n - \xi)}$$

 $\rho_0 = a[(\sin\theta\cos\phi - \sin\theta_0\cos\phi_0)^2 + (\sin\theta\sin\phi - \sin\theta_0\sin\phi_0)^2]^{1/2}$ 

Three-dimensional amplitude pattern of the array factor for a uniform circular array of N = 10 elements ( $C\Lambda = ka = 10$ ).

Principal-plane amplitude patterns of the array factor for a uniform circular array of N = 10 elements ( $C\Lambda = ka = 10$ ).





• For a uniform amplitude excitation of each element  $(I_n = I_0)$ ,

$$AF(\theta, \phi) = NI_0 \sum_{m=-\infty}^{+\infty} J_{mN}(k\rho_0) e^{jmN(\pi/2 - \xi)}$$

- $J_p(x)$  is the Bessel function of the first kind
- The part of the array factor associated with the zero order Bessel function  $J_0(k\rho_0)$  is called the *principal term* and the remaining terms are noted as the *residuals*.
- For a circular array with a large number of elements, the term  $J_0(k\rho_0)$  alone can be used to approximate the two-dimensional principal-plane patterns. The remaining terms contribute negligibly because Bessel functions of larger orders are very small