

3.3 Frequency independent antennas - spiral and log periodic antenna

Module:3 HF, UHF and Microwave Antennas

Course: BECE305L – Antenna and Microwave Engineering

-Dr Richards Joe Stanislaus

Assistant Professor - SENSE

Email: richards.stanislaus@vit.ac.in



VIT[®]

Vellore Institute of Technology
(Deemed to be University under section 3 of UGC Act, 1956)
CHENNAI

Module:3 HF, UHF and Microwave Antennas

7 hours

Wire Antennas - long wire, loop antenna - helical antenna. Yagi-Uda antenna, Frequency independent antennas - spiral and log periodic antenna - Aperture antennas – Horn antenna, Parabolic reflector antenna - Microstrip antenna

- Source of the contents: Balanis Antenna Theory and Antennas from theory to practice (Yi Huang)

1. Frequency independent antennas - Introduction

- The use of **simple, small, lightweight, and economical antennas**, designed to operate over **the entire frequency band** of a given system, would be **most desirable**.
- antennas with **broadband pattern** and impedance characteristics having **bandwidths** to as great as **40:1** or more were referred to as ***frequency independent***.
- **geometries were specified by angles.**
- are primarily used in the **10–10,000 MHz region** in a variety of practical applications such as **TV, point-to-point communication, feeds for reflectors and lenses**, and so forth

1. Frequency independent antennas - Introduction

- **Scale model measurements** :
if *all* the physical dimensions are *reduced* by a factor of two, the performance of the antenna will remain unchanged if the operating frequency is *increased* by a factor of two.
- Also: if the **shape of the antenna were completely specified by angles**, its performance would have to be **independent of frequency**
- **infinite biconical dipole**
To make infinite
- **structures more practical**, the designs usually require that the **current on the structure decrease with distance away from the input terminals**.
After a certain point the current is negligible, and the structure beyond that point to infinity can be truncated and removed.

1. Frequency independent antennas - Introduction

- Practical bandwidths are on the order of about 40:1. Even higher ratios (i.e., 1,000:1) can be achieved in antenna design but they are not necessary, since they would far exceed the bandwidths of receivers and transmitters
- a general solution for the surface $r = F(\theta, \phi)$

$$r = F(\theta, \phi) = e^{a\phi} f(\theta)$$

where

$$a = \frac{1}{K} \frac{dK}{dC}$$

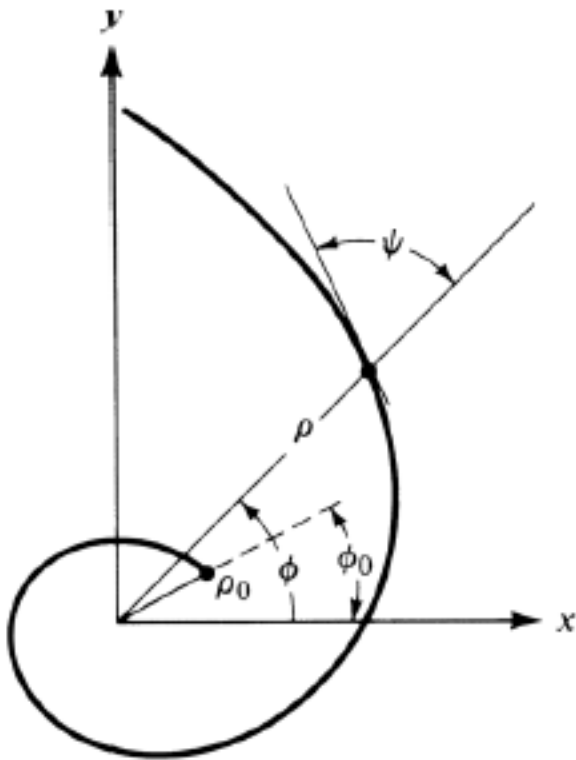
If the antenna is to be scaled to a frequency that is **K times lower than the original frequency**, the antenna's physical surface must be made K times greater to maintain the **same electrical dimensions**. second antenna to achieve congruence with the first, it must be rotated by an angle C so that

$$KF(\theta, \phi) = F(\theta, \phi + C)$$

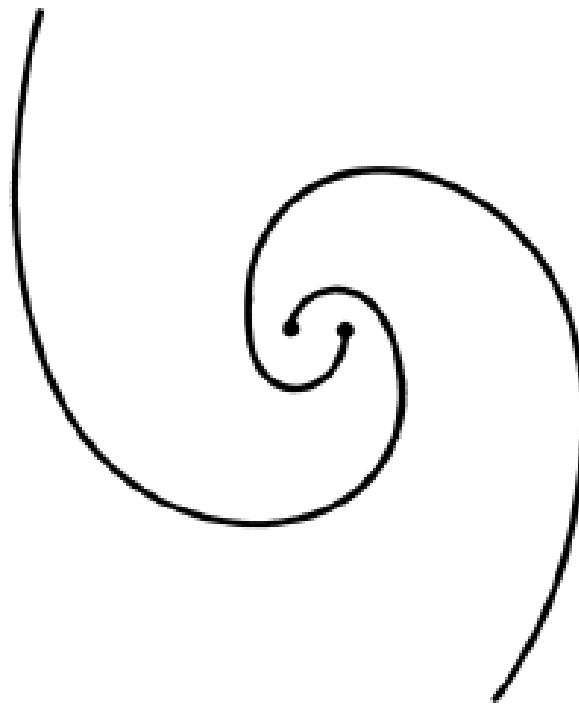
angle of rotation C depends on K , *radiation* pattern will be rotated azimuthally through an angle C .

2. Planar Spiral Antenna

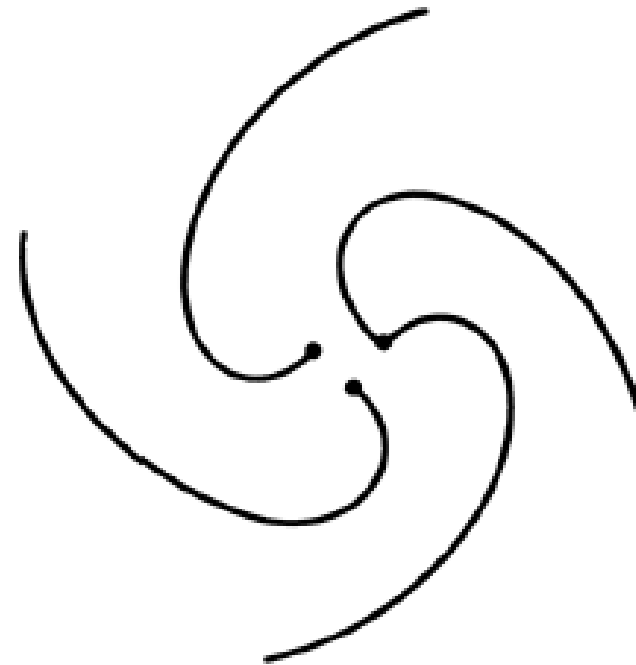
- Spiral Wire antennas



(a) Single spiral



(b) Two spiral ($\phi_0 = 0, \pi$)



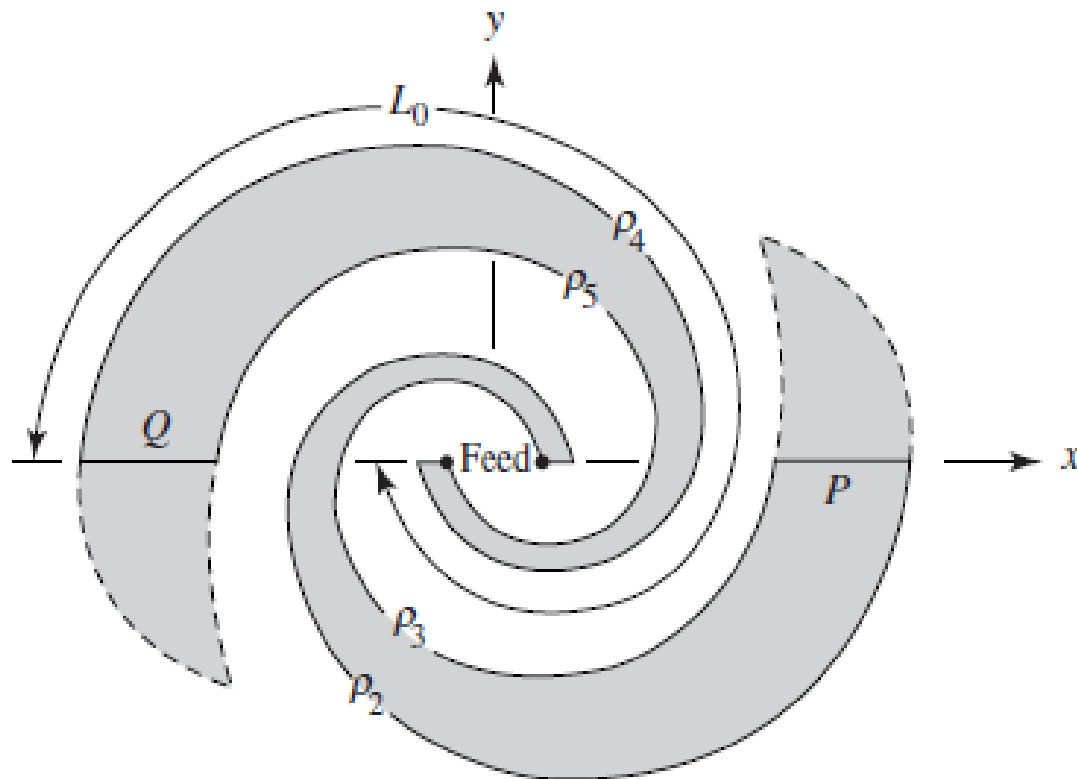
(c) Multiple spiral
($\phi_0 = 0, \pi/2, \pi, 3\pi/2$)



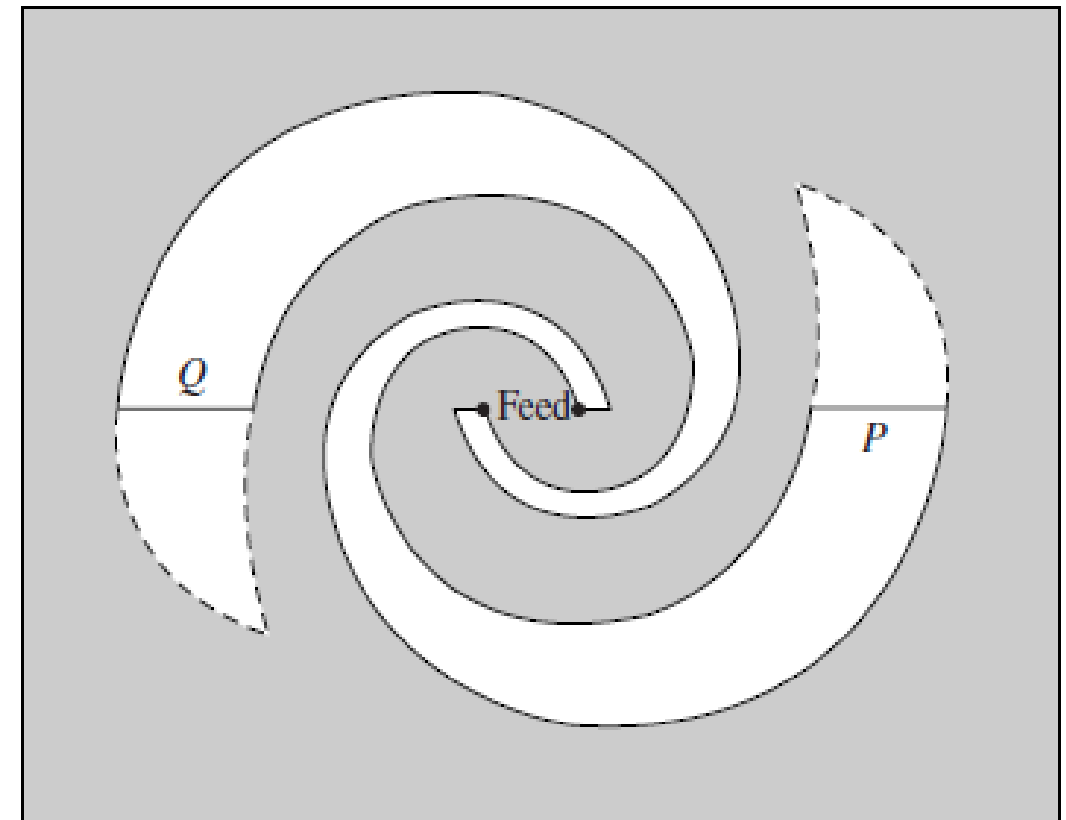
(d) Multiple spiral
($\phi_0 = 0, \pi/2, \pi, 3\pi/2$)

2. Planar Spiral Antenna

- Planar spiral: One antenna would consist of two metallic arms suspended in free-space

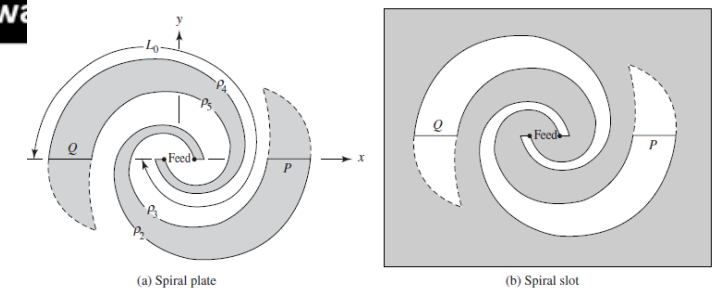


(a) Spiral plate



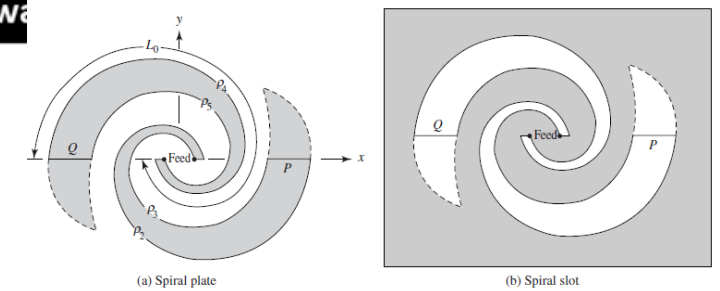
(b) Spiral slot

2. Planar Spiral Antenna



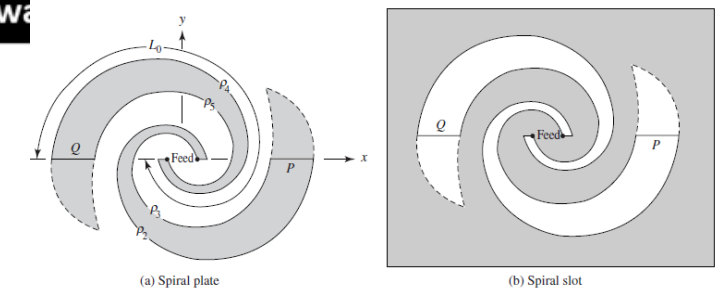
- The **slot antenna is the most practical**, because it can be conveniently **fed by a balanced coaxial arrangement** to maintain its overall balancing
- **input impedance for an infinite structure** should be $Z_s = Z_c = 188.5 \simeq 60\pi$ ohms
- **Spiral slot antennas**, with good radiation characteristics, can be built with **one-half to three turns**. The most optimum design seems to be that with **1.25 to 1.5 turns with an overall length equal to or greater than one wavelength**. The rate of expansion should not exceed about 10 per turn.

2. Planar Spiral Antenna



- The patterns are bidirectional, **single lobed**, **broadside** (**maximum normal to the plane**), and must vanish along the directions occupied by the infinite structure.
- The wave is circularly polarized near the axis of the main lobe over the usable part of the bandwidth.
For a fixed cut, the **beamwidth** will vary with frequency since the **pattern rotates**. Typical variations are on the order of 10° .
- however, slot antennas with **more broad arms** and/or **more tightly wound spirals** exhibit **smoother and more uniform patterns with smaller variations** in beamwidth with frequency.
- For symmetrical structures, the pattern is also symmetrical with no tilt to the lobe structure.

2. Planar Spiral Antenna



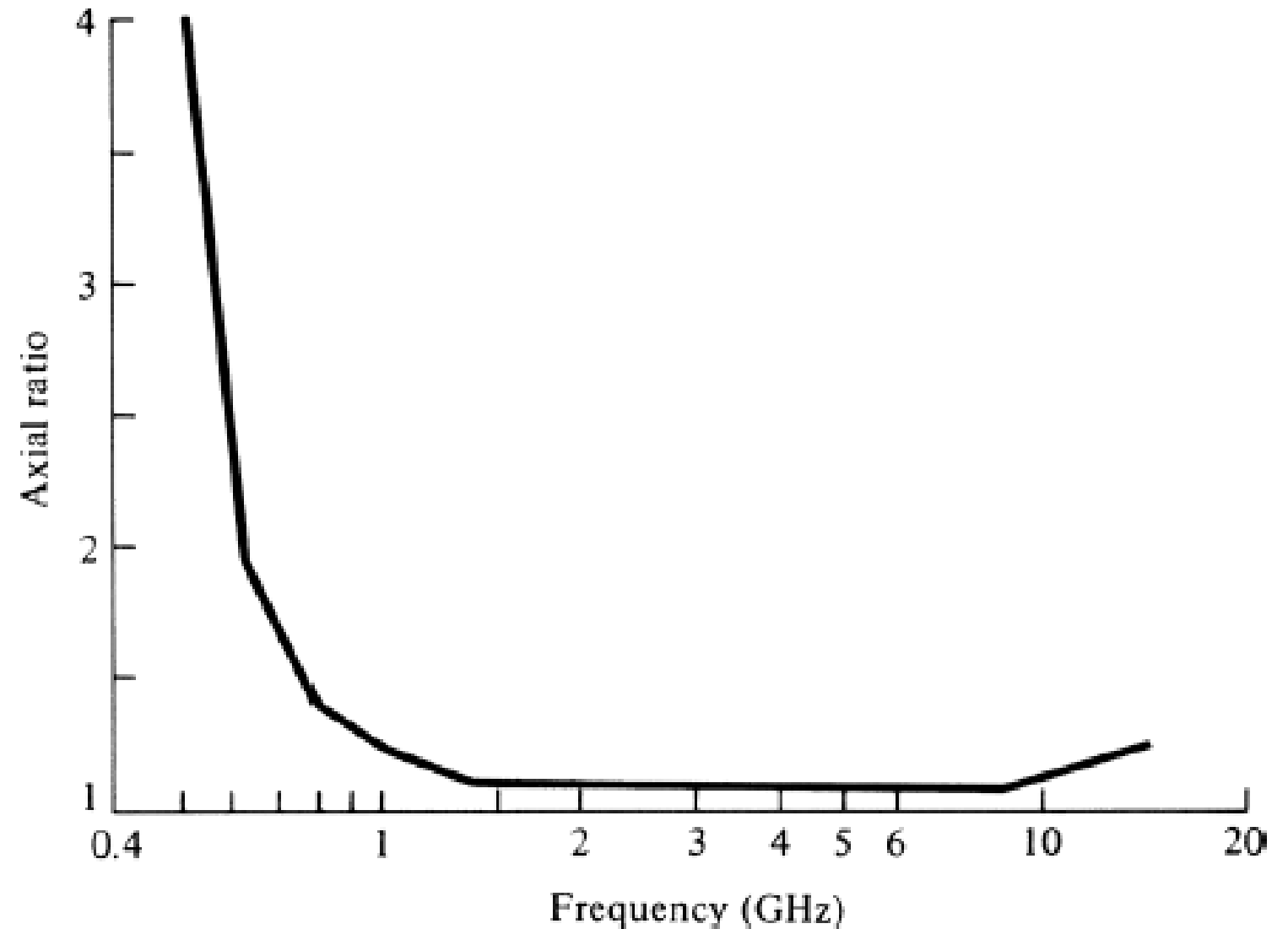
- The **polarization of the radiated wave is controlled by the length of the arms.**

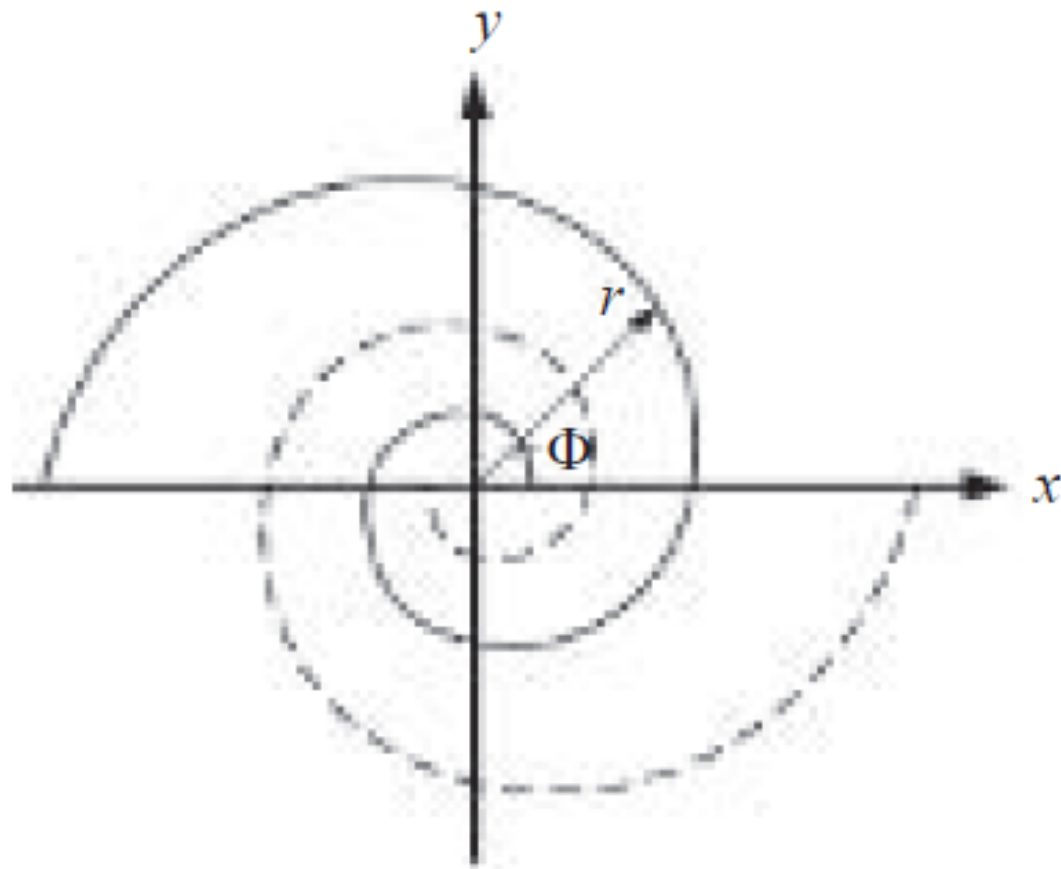
For **very low frequencies**, such that the **total arm length is small compared to the wavelength**, the **radiated field is linearly polarized**.

As the **frequency increases**, the **wave becomes elliptically polarized** and eventually **achieves circular polarization**.

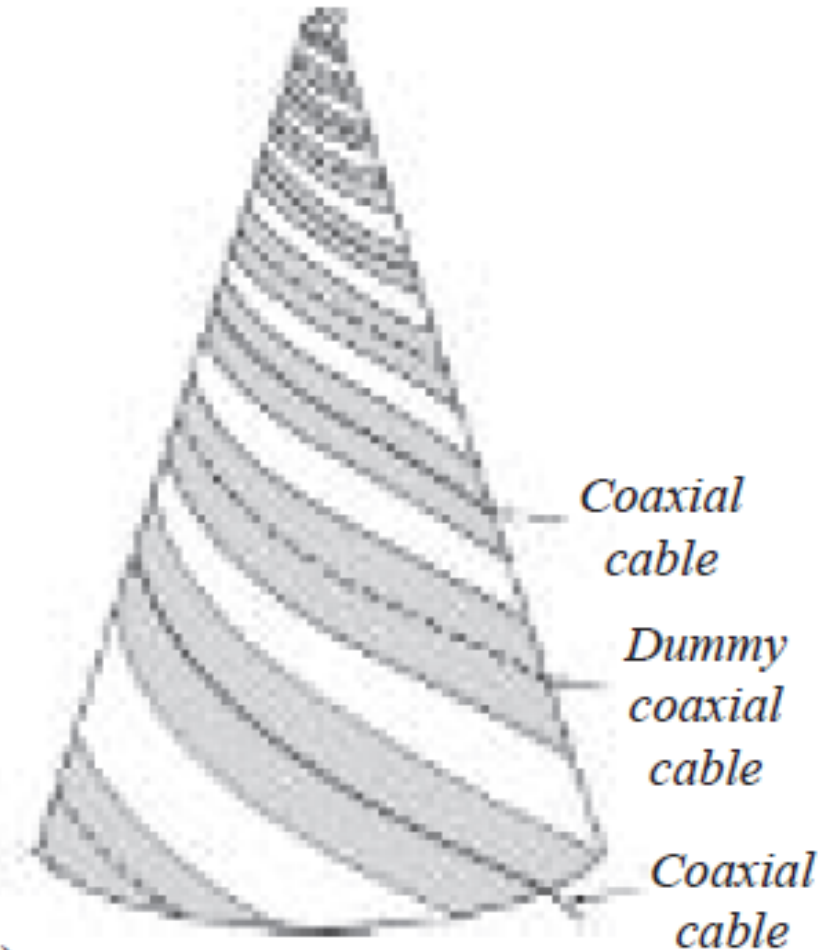
- Pattern is essentially unaltered through this frequency
- the **polarization change with frequency can be used as a convenient criterion** to select the **lower cutoff frequency of the usable bandwidth** chosen to be the point where the **axial ratio is equal or less than 2 to 1**, when the overall arm length is about one wavelength.

On-axis polarization as a function of frequency for one-turn spiral slot.





(a)



(b)

Figure 5.21 Spiral antennas (a) the wire type; (b) the conical type

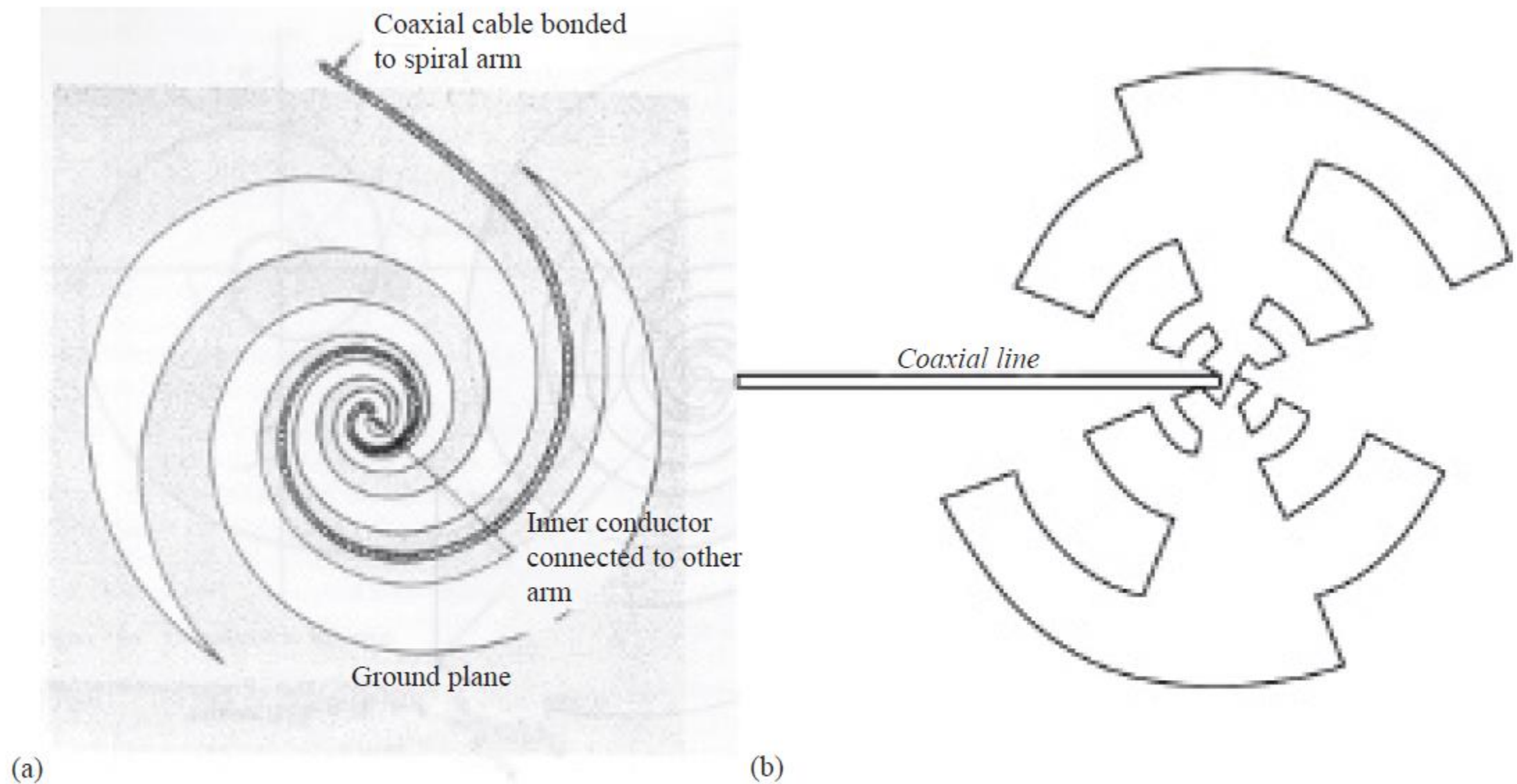
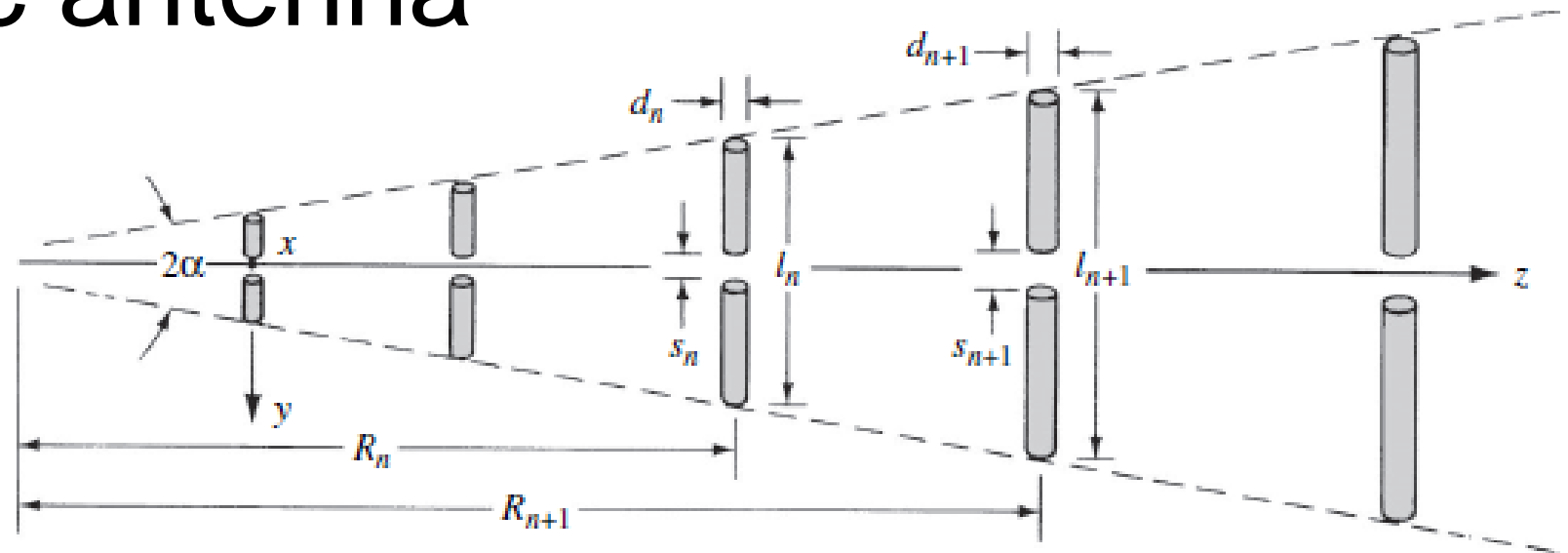
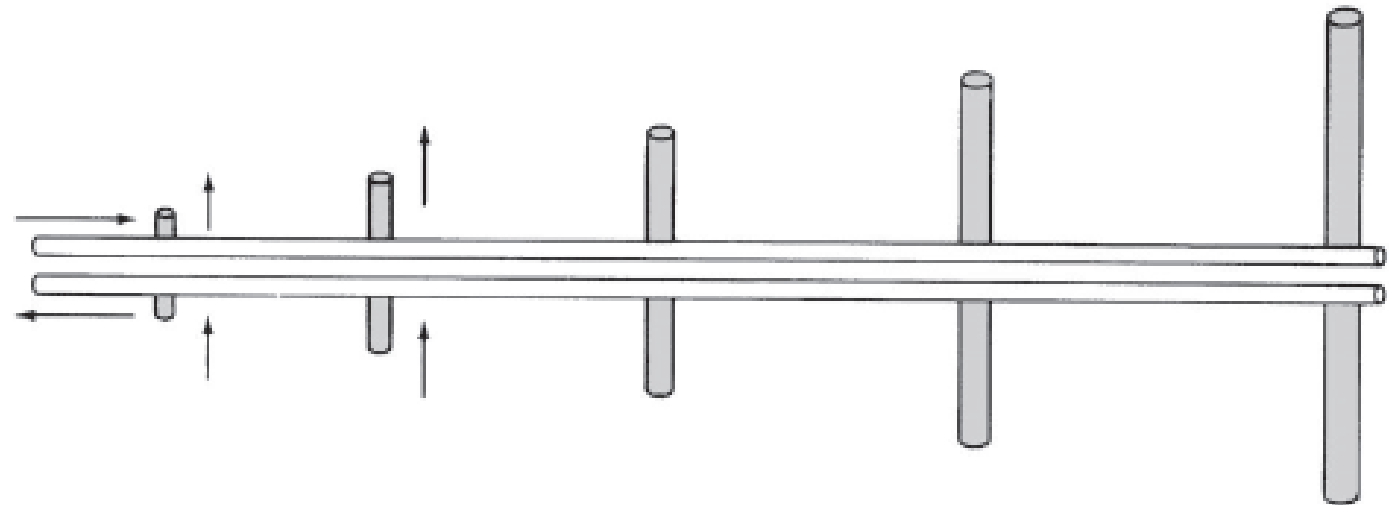


Figure 5.22 Two examples of self-complementary antennas (a) planar spiral; (b) log-periodic toothed planar antenna. (Reproduced by permission of John Wiley & Sons, Inc.)

3. Log periodic antenna

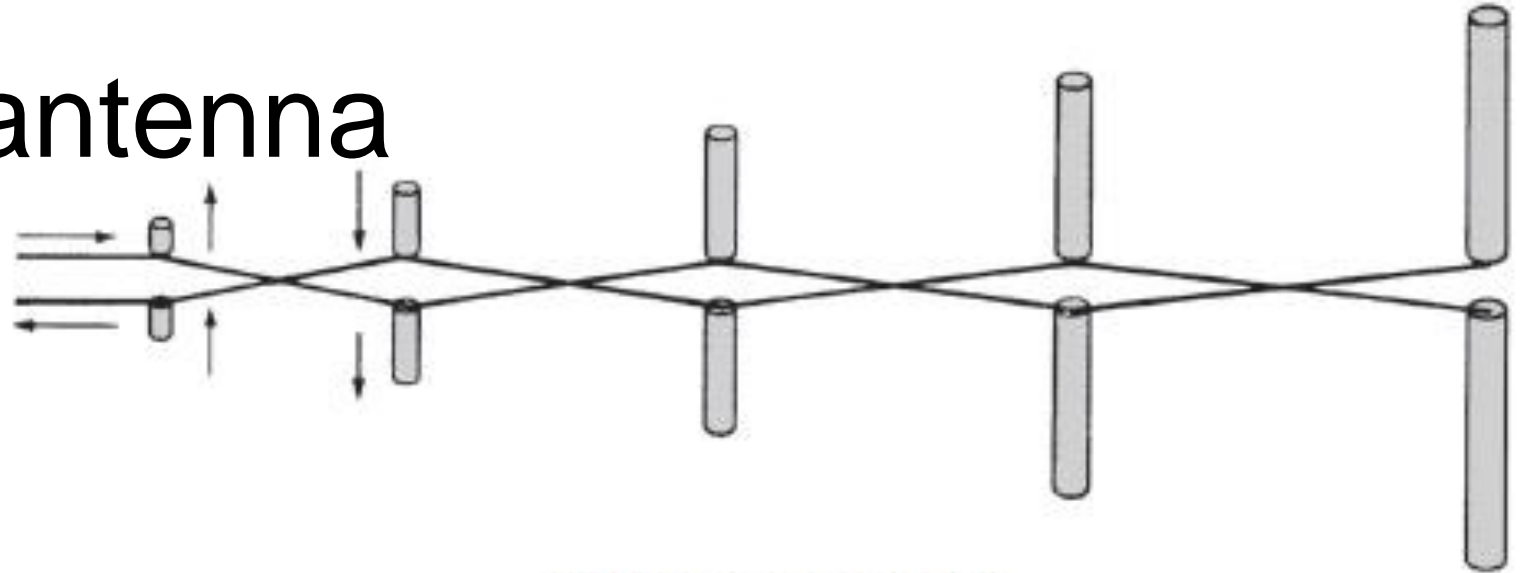


(a) Dipole array

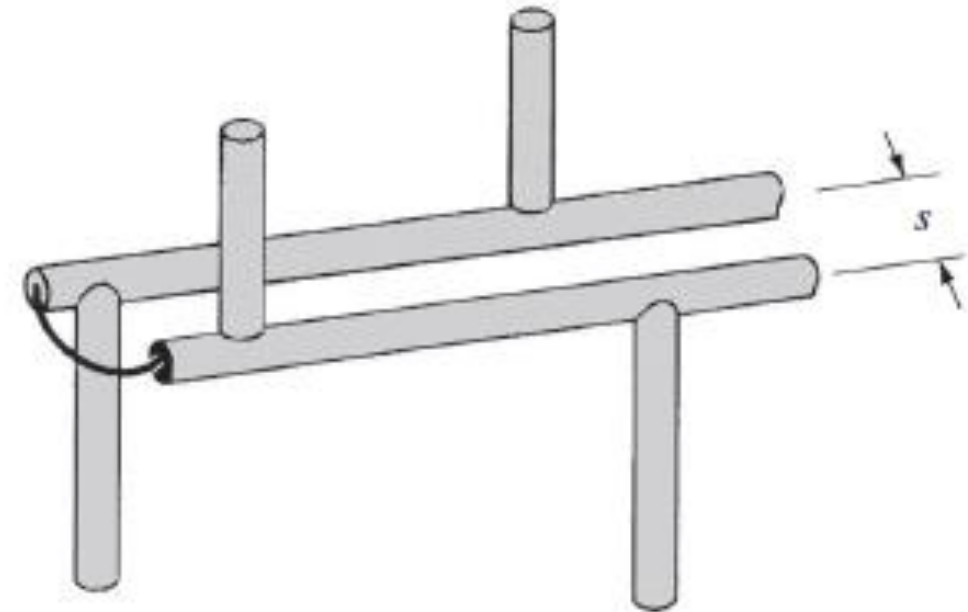


(b) Straight connection

3. Log periodic antenna



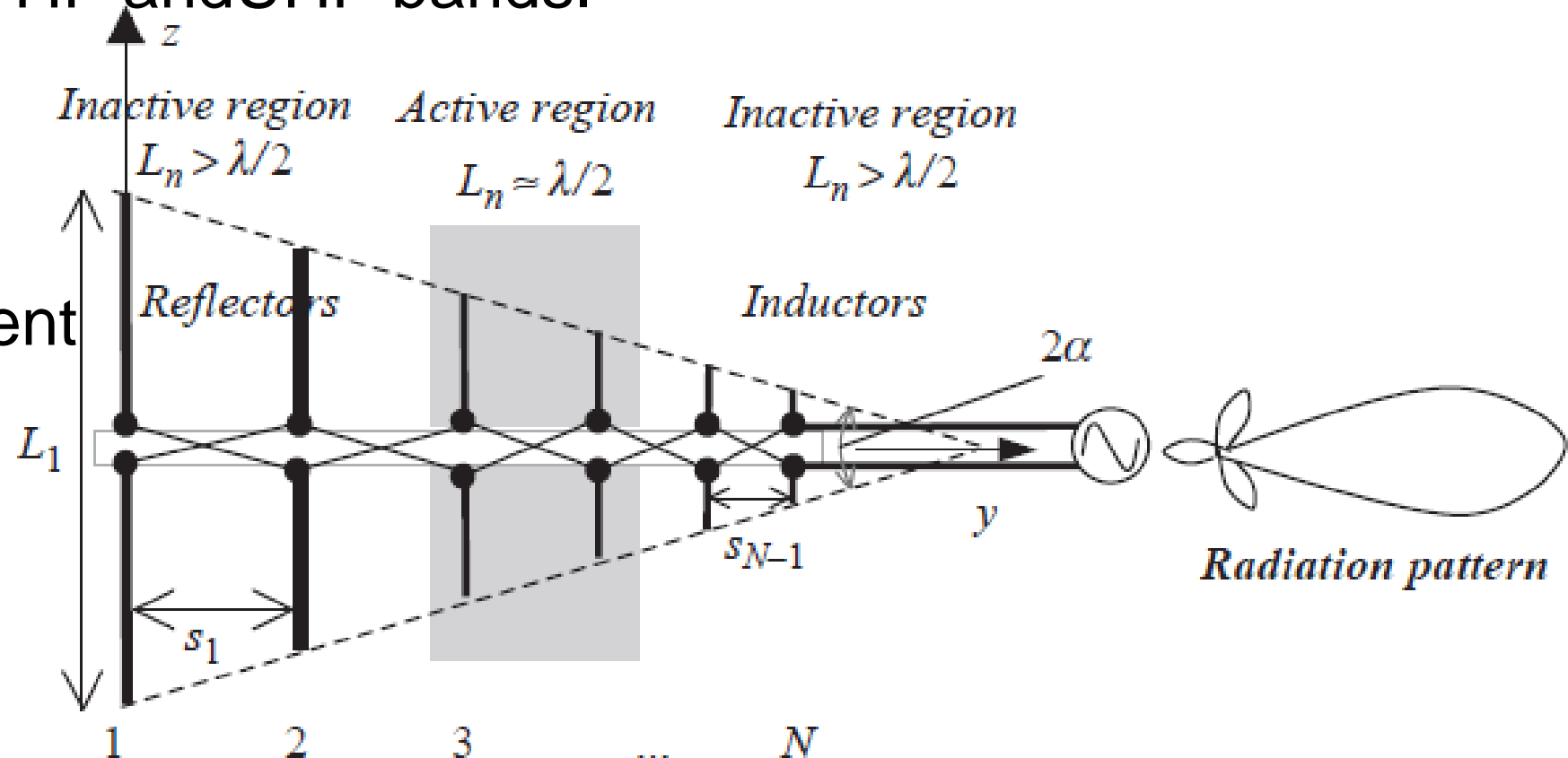
(c) Crisscross connection



(d) Coaxial connection

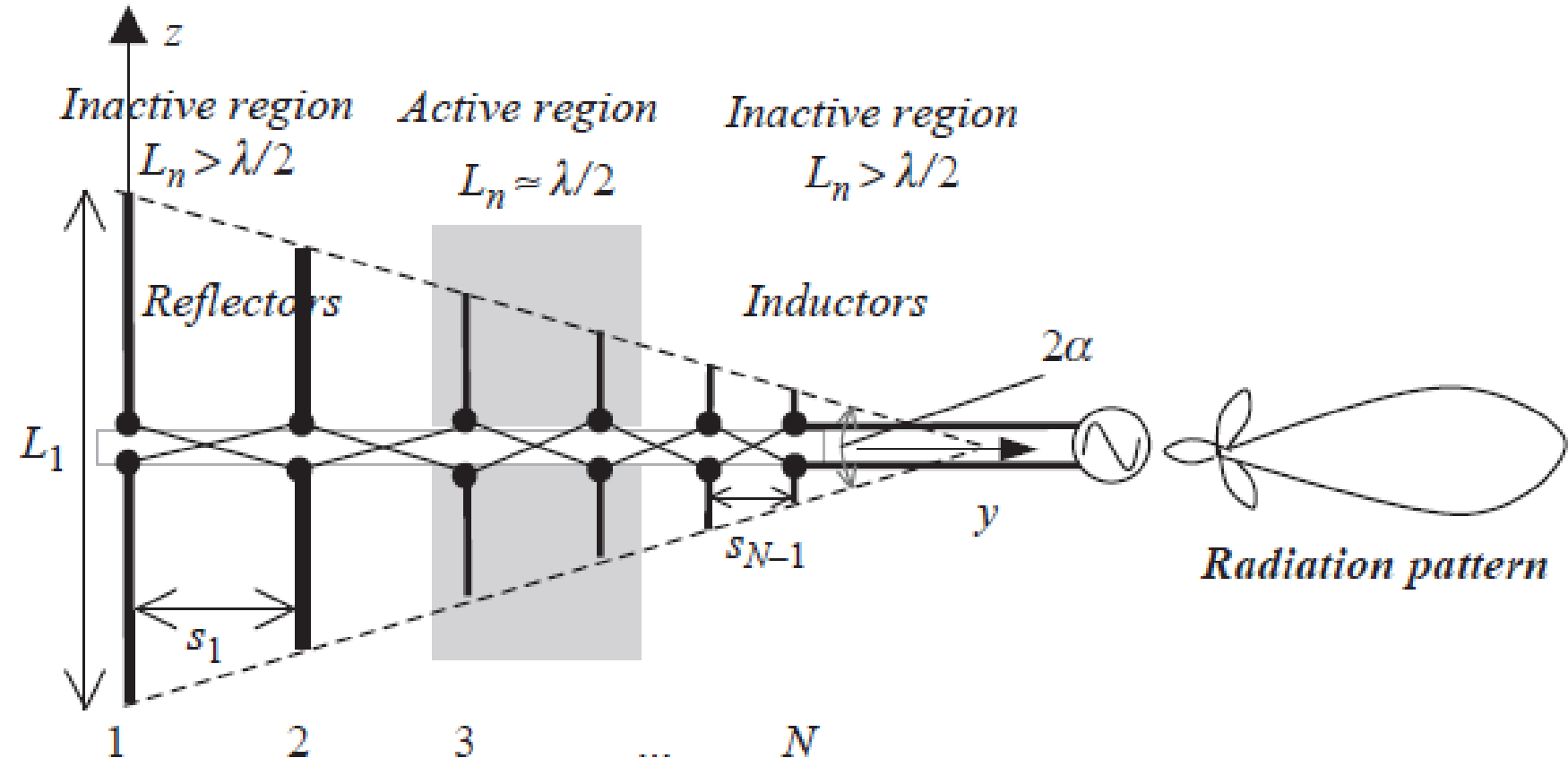
3. Log periodic antenna

- end-fire radiation pattern and directivity (typically between 7 and 15 dBi)
- widely used in the VHF and UHF bands.
- Much wider bandwidth than Yagi-Uda
- Feeder: Each element is connected to source
Each element is active



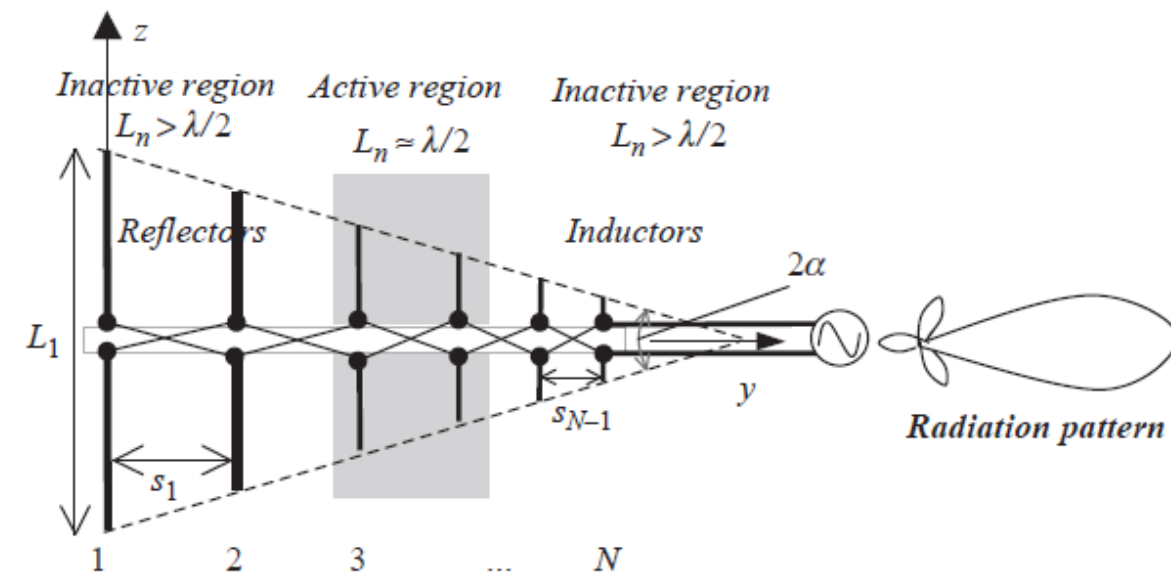
3. Log periodic antenna (log-periodic dipole antenna (LPDA))

- consists of many dipoles of different lengths.
- antenna is divided into the so-called **active region** and **inactive regions**



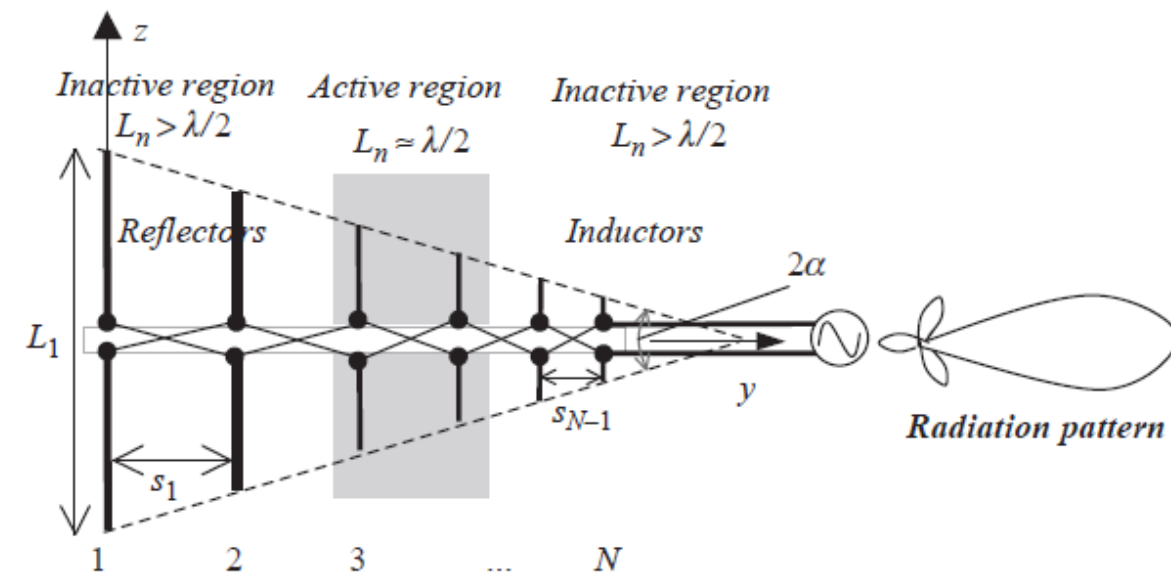
3. Log periodic antenna (log-periodic dipole antenna (LPDA))

- role of a specific dipole element is linked to the operating frequency:
- if its length, L , is around half of the wavelength, it is an active dipole and within the active region;
- if its length is greater than half the wavelength, it is in an inactive region and acts as a reflector;
- if its length is smaller than half the wavelength, it is also in an inactive region but acts as a director, which is very similar to the Yagi-Uda antenna.



3. Log periodic antenna (log-periodic dipole antenna (LPDA))

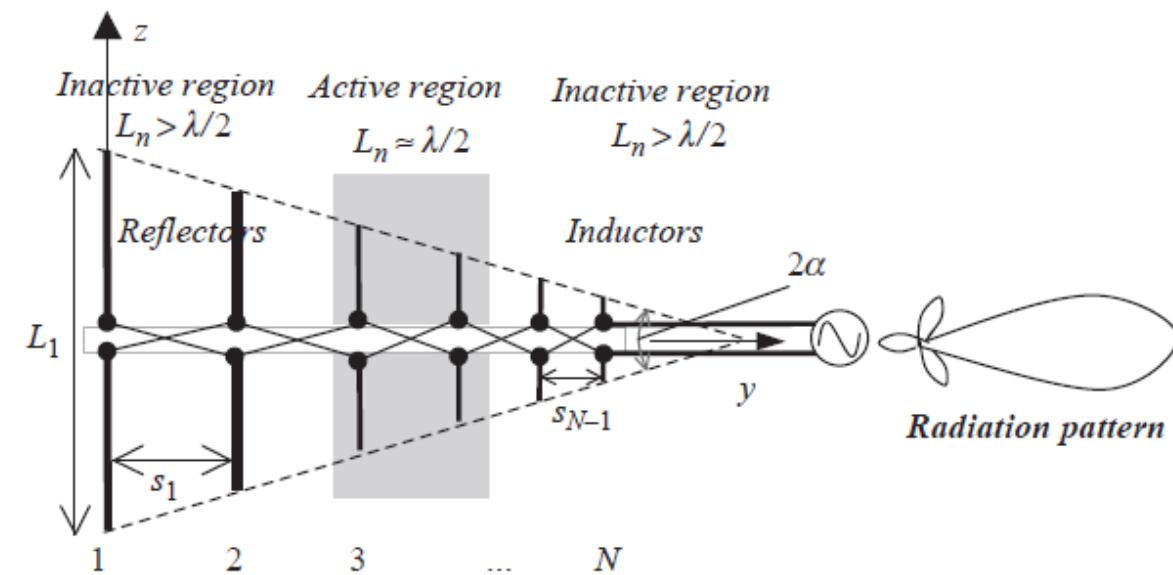
- The difference is that the **driven element shifts with the frequency** – this is why this antenna can offer a **much wider bandwidth** than the Yagi– Uda.
- A traveling wave can also be formed in the antenna.
- The **highest frequency** is basically determined by the **shortest dipole length (L_N)** while the **lowest frequency is determined by the longest dipole length (L_1)**.



3. Log periodic antenna (log-periodic dipole antenna (LPDA))

- The reason for this antenna being called a *log-periodic antenna* is that its **input impedance is a periodic function of the logarithm of the frequency**.

Other parameters that undergo similar variations include the radiation pattern, directivity and beamwidth.



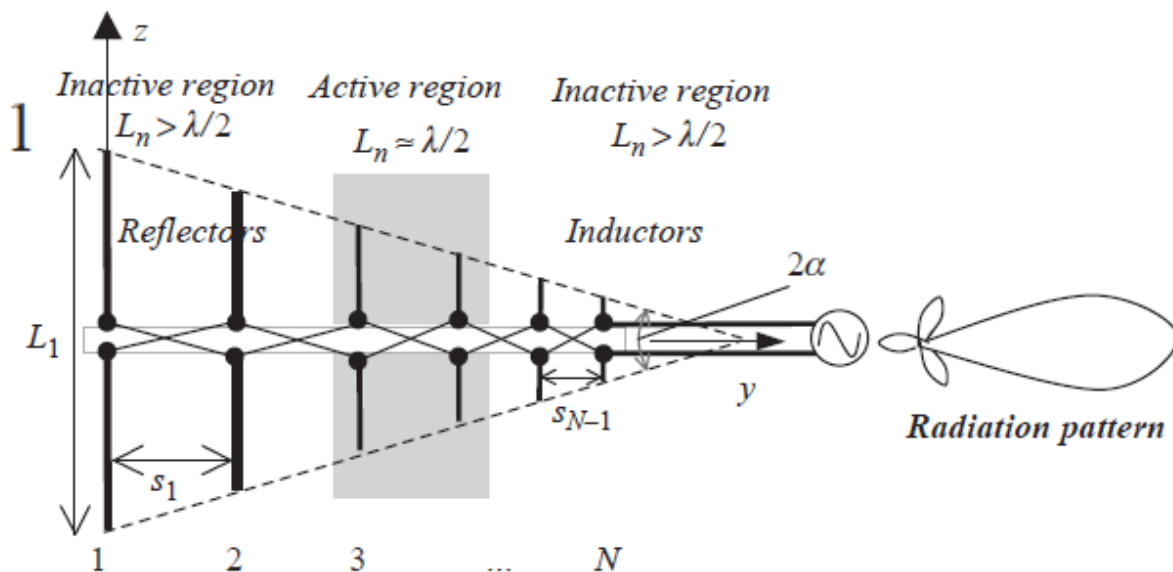
3.1 Log periodic antenna: Design

- L_n = the length of element n , and $n = 1, 2, \dots, N$;
- s_n = the spacing between elements n and $(n + 1)$;
- d_n = the diameter of element n ;
- g_n = the gap between the poles of element n .
- **scaling factor:**

$$\tau = \frac{L_2}{L_1} = \frac{L_{n+1}}{L_n} = \frac{s_{n+1}}{s_n} = \frac{d_{n+1}}{d_n} = \frac{g_{n+1}}{g_n} < 1$$

- **spacing factor**

$$\sigma = \frac{s_1}{2L_1} = \frac{s_n}{2L_n} < 1$$



$$\tau = \frac{L_2}{L_1} = \frac{L_{n+1}}{L_n} = \frac{s_{n+1}}{s_n} = \frac{d_{n+1}}{d_n} = \frac{g_{n+1}}{g_n} < 1$$

$$\sigma = \frac{s_1}{2L_1} = \frac{s_n}{2L_n} < 1$$

3.1 Log periodic antenna: [

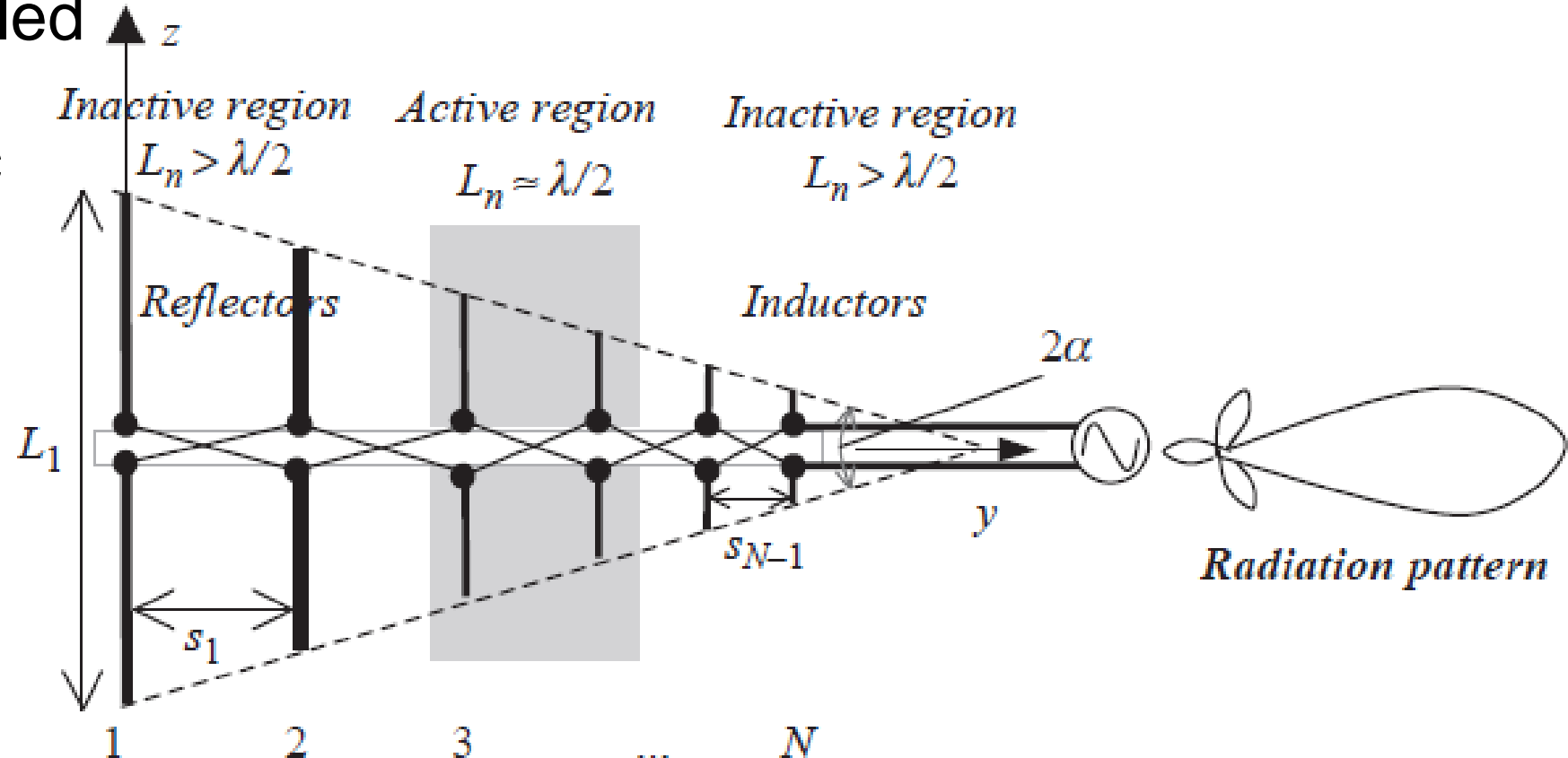
- two straight lines through the dipole ends form an angle 2α , is a characteristic of the frequency-independent structure
- The angle α is called the **apex angle** of the log-periodic antenna

$$\alpha = \tan^{-1} \left(\frac{(L_n - L_{n+1})}{2s_n} \right)$$

$$= \tan^{-1} \left(\frac{L_n(1 - \tau)}{2s_n} \right)$$

$$= \tan^{-1} \left(\frac{(1 - \tau)}{4\sigma} \right)$$

Joe Sta



3.1 Log periodic antenna: [

$$\tau = \frac{L_2}{L_1} = \frac{L_{n+1}}{L_n} = \frac{s_{n+1}}{s_n} = \frac{d_{n+1}}{d_n} = \frac{g_{n+1}}{g_n} < 1$$

$$\sigma = \frac{s_1}{2L_1} = \frac{s_n}{2L_n} < 1$$

- From frequency aspect

$$\tau = \frac{L_{n+1}}{L_n} = \frac{f_n}{f_{n+1}}$$

$$\alpha = \tan^{-1} \left(\frac{(L_n - L_{n+1})}{2s_n} \right)$$

- logarithm of both sides

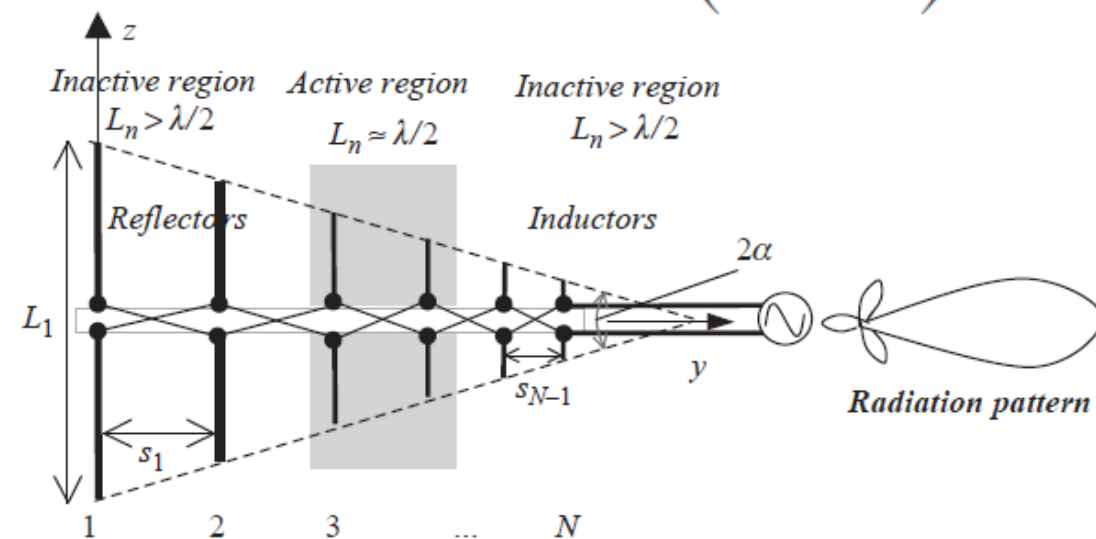
$$\log f_{n+1} = \log f_n - \log \tau$$

$$= \tan^{-1} \left(\frac{L_n(1 - \tau)}{2s_n} \right)$$

- resonant frequency in log scale is increased by **every $|\log \tau|$** .

$$= \tan^{-1} \left(\frac{(1 - \tau)}{4\sigma} \right)$$

- performance of the antenna is periodic in a logarithmic fashion, Hence, **log-periodic antenna**



3.1 Log periodic antenna: Design

Table 5.5 Optimum design data for log-periodic antenna

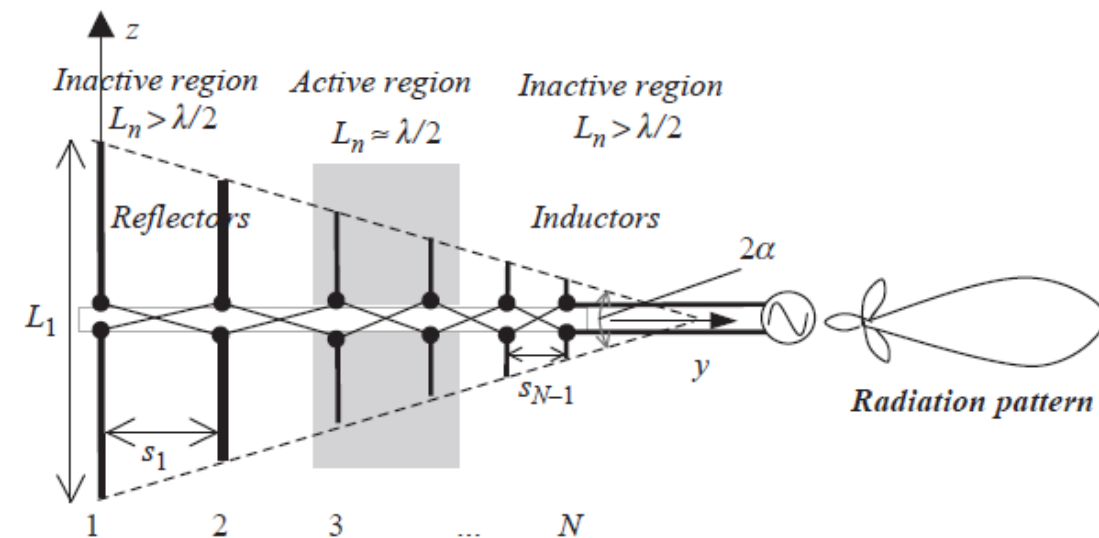
Directivity/dBi	Scaling factor τ	Spacing factor σ	Apex angle α
7	0.782	0.138	21.55°
7.5	0.824	0.146	16.77°
8	0.865	0.157	12.13°
8.5	0.892	0.165	9.29°
9	0.918	0.169	6.91°
9.5	0.935	0.174	5.33°
10	0.943	0.179	4.55°
10.5	0.957	0.182	3.38°
11	0.964	0.185	2.79°

3.1 Log periodic antenna: Design

- directivity, length of the antenna, apex angle, the upper frequency and the lower frequency, should come with the design specifications.
- Another important aspect of the design is the **antenna input impedance**, which can be **tuned by changing the diameter d of the element** and **the feeding gap g between the two poles**

$$g = d \cosh(Z_0/120)$$

- **Z_0 is the characteristic impedance of the feed line to be connected (the desired impedance).**



3.1 Log periodic antenna: Design

- **Practical design:** most likely scenario is that the frequency range is given from

$$L_1 \geq \frac{\lambda_{\max}}{2} = \frac{c}{f_{\min}}; L_N \leq \frac{\lambda_{\min}}{2} = \frac{c}{f_{\max}}$$

$$\frac{f_{\min}}{f_{\max}} = \frac{L_N}{L_1} = \tau \frac{L_{N-1}}{L_1} = \tau^{N-1}$$

- geometrical dimensions are obtained, it is desirable to find the radiation pattern

Log-periodic antenna design. Design a log-periodic dipole antenna to cover all UHF TV channels, which is from 470 MHz for channel 14 to 890 MHz for channel 83. Each channel has a bandwidth of 6 MHz. The desired directivity is 8 dBi

- desired directivity is 8 dBi,

$$f_{\min} = 470 \text{ MHz}, f_{\max} = 890 \text{ MHz}, \text{ and } D = 8 \text{ dBi}$$

- Optimum design from table:
Scaling factor $\tau = 0.865$
Spacing factor $\sigma = 0.157$
Apex angle $\alpha = 12.13^\circ$

Table 5.5 Optimum design data for log-periodic antenna

Directivity/dBi	Scaling factor τ	Spacing factor σ	Apex angle α
7	0.782	0.138	21.55°
7.5	0.824	0.146	16.77°
8	0.865	0.157	12.13°
8.5	0.892	0.165	9.29°
9	0.918	0.169	6.91°
9.5	0.935	0.174	5.33°
10	0.943	0.179	4.55°
10.5	0.957	0.182	3.38°
11	0.964	0.185	2.79°

Log-periodic antenna design. Design a log-periodic dipole antenna to cover all UHF TV channels, which is from 470 MHz for channel 14 to 890 MHz for channel 83. Each channel has a bandwidth of 6 MHz. The desired directivity is 8 dBi

- desired directivity is 8 dBi,

$$f_{\min} = 470 \text{ MHz}, f_{\max} = 890 \text{ MHz}, \text{ and } D = 8 \text{ dBi}$$

- Optimum design from table:

Scaling factor $\tau = 0.865$

Spacing factor $\sigma = 0.157$

Apex angle $\alpha = 12.13^\circ$

Table 5.5 Optimum design data for log-periodic antenna

Directivity/dBi	Scaling factor τ	Spacing factor σ	Apex angle α
7	0.782	0.138	21.55°
7.5	0.824	0.146	16.77°
8	0.865	0.157	12.13°
8.5	0.892	0.165	9.29°
9	0.918	0.169	6.91°
9.5	0.935	0.174	5.33°
10	0.943	0.179	4.55°
10.5	0.957	0.182	3.38°
11	0.964	0.185	2.79°

$$N = \log \left(\frac{f_{\min}}{f_{\max}} \right) / \log(\tau) + 1 = \log \left(\frac{470}{890} \right) / \log(0.865) + 1 = 4.40 + 1 = 5.40 \approx 6$$

- at least six elements are required. Let's take 8 elements for safe side

Log-periodic antenna design. Design a log-periodic dipole antenna to cover all UHF TV channels, which is from 470 MHz for channel 14 to 890 MHz for channel 83. Each channel has a bandwidth of 6 MHz. The desired directivity is 8 dBi

- desired directivity is 8 dBi,

$$f_{\min} = 470 \text{ MHz}, f_{\max} = 890 \text{ MHz}, \text{ and } D = 8 \text{ dBi}$$

- $N = 8$, we can afford to start from a lower frequency, say 400 MHz

Table 5.5 Optimum design data for log-periodic antenna

Directivity/dBi	Scaling factor τ	Spacing factor σ	Apex angle α
7	0.782	0.138	21.55°
7.5	0.824	0.146	16.77°
8	0.865	0.157	12.13°
8.5	0.892	0.165	9.29°
9	0.918	0.169	6.91°
9.5	0.935	0.174	5.33°
10	0.943	0.179	4.55°
10.5	0.957	0.182	3.38°
11	0.964	0.185	2.79°

$$L_1 = \frac{c}{f_{\min}} = \frac{300}{400} = 0.75 \text{ (m)}$$

Log-periodic antenna design. Design a log-periodic dipole antenna to cover all UHF TV channels, which is from 470 MHz for channel 14 to 890 MHz for channel 83. Each channel has a bandwidth of 6 MHz. The desired directivity is 8 dBi

- desired directivity is 8 dBi,

$$f_{\min} = 470 \text{ MHz}, f_{\max} = 890 \text{ MHz}, \text{ and } D = 8 \text{ dBi}$$

- $N = 8$, we can afford to start from a lower frequency, say 400 MHz

Table 5.5 Optimum design data for log-periodic antenna

Directivity/dBi	Scaling factor τ	Spacing factor σ	Apex angle α
7	0.782	0.138	21.55°
7.5	0.824	0.146	16.77°
8	0.865	0.157	12.13°
8.5	0.892	0.165	9.29°
9	0.918	0.169	6.91°
9.5	0.935	0.174	5.33°
10	0.943	0.179	4.55°
10.5	0.957	0.182	3.38°
11	0.964	0.185	2.79°

$$L_1 = \frac{c}{f_{\min}} = \frac{300}{400} = 0.75 \text{ (m)}$$

$$L_2 = \tau L_1 = 0.865 * 0.75 = 0.6487 \text{ (m)}, \dots, L_8 = 0.2718 < \frac{c}{f_{\max}} = \frac{300}{890} = 0.3371 \text{ (m)}$$

$$s_n = 2L_n\sigma = 0.314L_n$$

Log-periodic antenna design. Design a log-periodic dipole antenna to cover all UHF TV channels, which is from 470 MHz for channel 14 to 890 MHz for channel 83. Each channel has a bandwidth of 6 MHz. The desired directivity is 8 dBi

- desired directivity is 8 dBi,

$$f_{\min} = 470 \text{ MHz}, f_{\max} = 890 \text{ MHz}, \text{ and } D = 8 \text{ dBi}$$

- $N = 8$, we can afford to start from a lower frequency, say 400 MHz

Table 5.5 Optimum design data for log-periodic antenna

Directivity/dBi	Scaling factor τ	Spacing factor σ	Apex angle α
7	0.782	0.138	21.55°
7.5	0.824	0.146	16.77°
8	0.865	0.157	12.13°
8.5	0.892	0.165	9.29°
9	0.918	0.169	6.91°
9.5	0.935	0.174	5.33°
10	0.943	0.179	4.55°
10.5	0.957	0.182	3.38°
11	0.964	0.185	2.79°

$$L_1 = \frac{c}{f_{\min}} = \frac{300}{400} = 0.75 \text{ (m)}$$

$$L_2 = \tau L_1 = 0.865 * 0.75 = 0.6487 \text{ (m)}, \dots, L_8 = 0.2718 < \frac{c}{f_{\max}} = \frac{300}{890} = 0.3371 \text{ (m)}$$

$$s_n = 2L_n\sigma = 0.314L_n$$

$$n = 1 \text{ to } 7. (s_1 = 0.2355; \dots, s_7 = 0.0986) \text{ Total length: } L = \sum_n s_n = 1.1142 \text{ (m)}$$