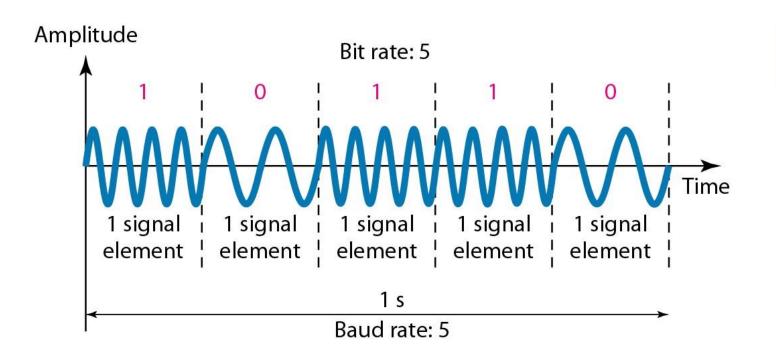
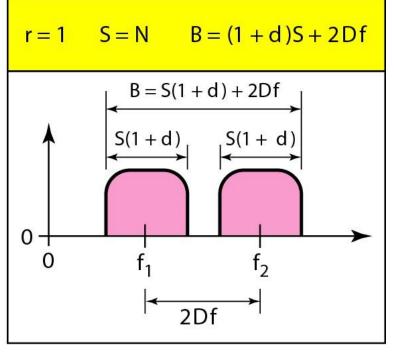


• If the difference between the two frequencies  $(f_1 \text{ and } f_2)$  is  $2\Delta f$ , then the required BW B will be:

$$B = (1+d)S + 2\Delta f$$





### Signal representation of FSK

Binary FASK is represented by

$$s_{i}(t) = \begin{cases} \sqrt{\frac{2E_{b}}{T_{b}}}\cos(2\pi f_{i}t), & 0 \leq t \leq T_{b} \\ 0 & elsewhere \end{cases}$$

• Where, i = 1.2 and

 $E_b$  = transmitted signal energy per bit

- Transmitted frequency is  $f_i = \frac{n_c + i}{T_b}$  for some fixed integer  $n_c$  and i = 1,2
- Symbol "1" is represented by mark frequency,

$$s_1(t) = \sqrt{2\frac{E_b}{T_b}} \cos 2\pi f_1 t$$
 Where,  $f_1 = f_c + \frac{\Omega}{2\pi}$ 

Symbol "0" is represented by space frequency,

$$s_2(t) = \sqrt{2\frac{E_b}{T_b}} \cos 2\pi f_2 t$$
 Where,  $f_2 = f_c - \frac{\Omega}{2\pi}$ 

- This FSK signal is also called as Sunde's FSK
- The two carriers are represented in terms of orthonormal basis functions as

$$\varphi_{i}(t) = \begin{cases} \sqrt{\frac{2}{T_{b}}} \cos 2\pi f_{i}t, & 0 \leq t \leq T_{b} \\ 0, & elsewhere \end{cases}$$

### Signal Space diagram of FSK

• The coefficients  $s_{ij}$ 

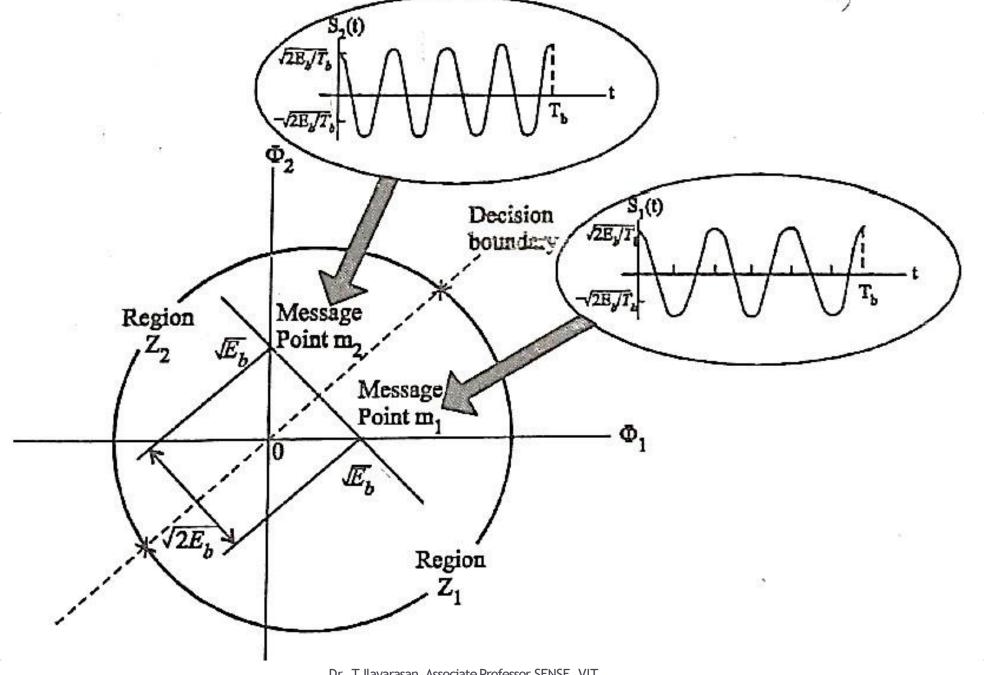
for i = 1,2 and j = 1,2 is given by,

$$S_{ij} = \int_{0}^{T_b} s_i(t) \phi_j(t) dt$$

$$= \int_{0}^{T_b} \sqrt{\frac{2E_b}{T_b}} \cos(2\pi f_i t) \sqrt{2/T_b} \cos(2\pi f_j t) dt$$

$$= \sqrt{E_b} \int_{0}^{T_b} \sqrt{\frac{2}{T_b}} \cos(2\pi f_i t) \sqrt{2/T_b} \cos(2\pi f_i t) dt$$

$$= \begin{cases} \sqrt{E_b} & i = j \\ 0 & i \neq i \end{cases}$$



• A coherent BFSK system is characterised by two dimensional (N=2) signal space with two message points, M=2 and it is defined as,

$$s_1 = \begin{bmatrix} \sqrt{E_b} \\ 0 \end{bmatrix}$$

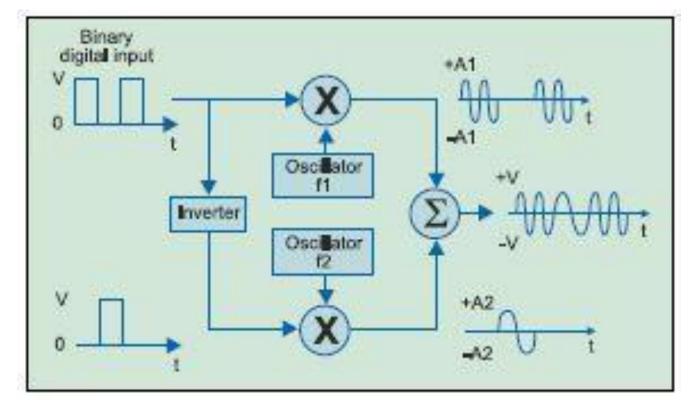
And

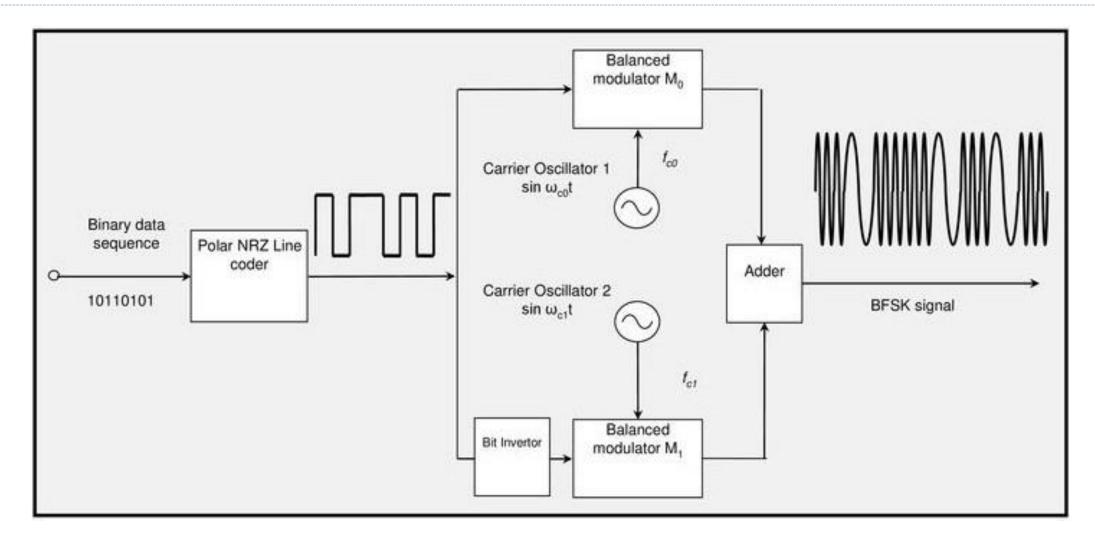
$$s_2 = \begin{bmatrix} 0 \\ \sqrt{E_b} \end{bmatrix}$$

• Euclidean distance between  $s_1$  and  $s_2$  is  $\sqrt{2E_b}$ 

### **Generation of FSK (FSK Modulator)**

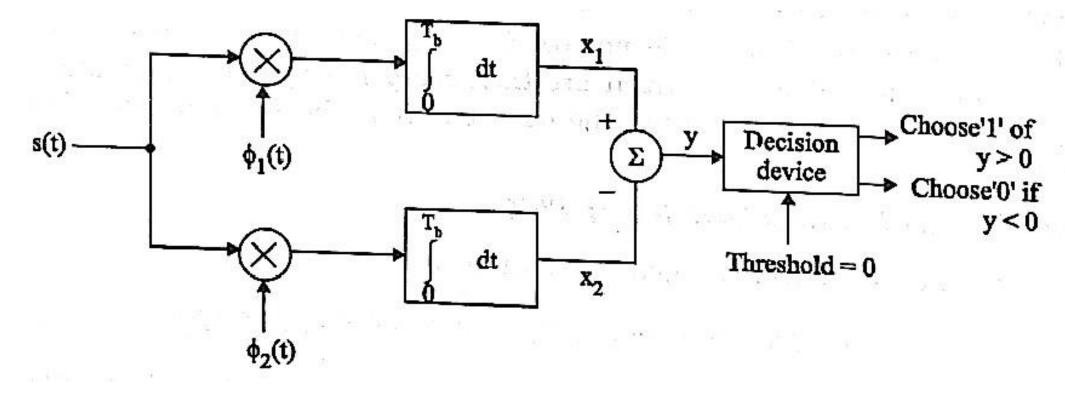
• The FSK transmitter (Modulator) consists of an On-Off Level encoder, product modulator and summer.





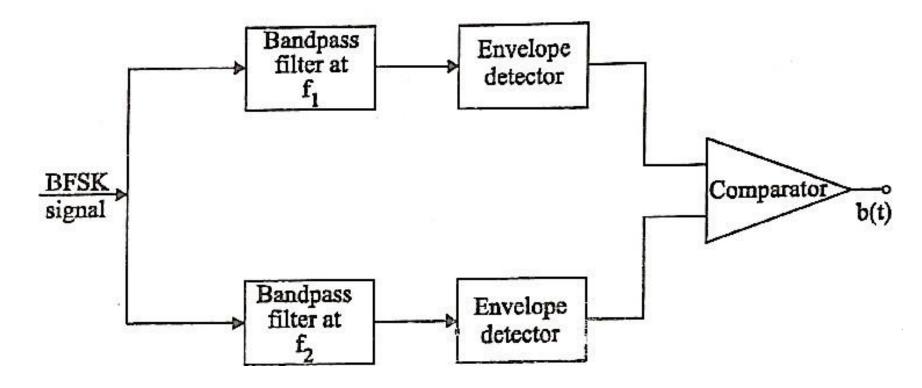
#### **Coherent BFSK Receiver**

• The block diagram of (coherent) Synchronous FSK detector consists of two mixers with local oscillator circuits, two band pass filters and a decision circuit.



#### Non-Coherent BFSK receiver

• The block diagram of (non-coherent) Asynchronous FSK detector consists of two band pass filters, two envelope detectors, and a decision circuit.



### Probability of error in BFSK

$$P_{e} = \frac{1}{2} \operatorname{erfc} \left( \frac{s_{1}(t) - s_{2}(t)}{2\sqrt{2} \sigma} \right)$$

$$\left(\frac{s_1(t)-s_2(t)}{\sigma}\right)^2 = \int_0^{T_b} \frac{s^2(t) dt}{N_o/2}$$

$$s_1(t)-s_2(t) = \sqrt{2P} \left[\cos(2\pi f_c + \Omega) t - \cos(2\pi f_c - \Omega) t\right]$$
By formula,

$$s\left( t\right) =s_{1}\left( t\right) -s_{2}\left( t\right)$$

$$s_1(t) = \sqrt{2P} \cos(2\pi f_c + \Omega) t$$

$$s_1(t) = \sqrt{2P} \cos(2\pi f_c + \Omega) t$$
  
$$s_2(t) = \sqrt{2P} \cos(2\pi f_c - \Omega) t$$

$$s_1(t) - s_2(t) = \sqrt{2P} \left[ \cos(2\pi f_c + \Omega) t - \cos(2\pi f_c - \Omega) t \right]$$

$$\cos (A+B) - \cos (A-B) = 2 \sin A \sin B$$

we get,

$$s_1(t) - s_2(t) = \sqrt{2P} [2 \sin 2 \pi f_c t \sin \Omega t]$$

$$s^{2}(t) = [s_{1}(t) - s_{2}(t)]^{2} = \int_{0}^{T_{b}} (\sqrt{2P} 2 \sin 2\pi f_{c}t \sin \Omega t)^{2} dt$$

$$= \int_{0}^{T_{b}} 2P \times 4 \sin^{2} 2\pi f_{c}t \sin^{2} \Omega t dt$$

$$\therefore \sin^{2} 0 = \frac{1 - \cos 2\theta}{2}$$

$$= \int_{0}^{T_{b}} 2P \times 4 \left(\frac{1 - \cos 4\pi f_{c}t}{2}\right) \left(\frac{1 - \cos 2\Omega t}{2}\right)^{-1} dt$$

$$= 2P \int_{0}^{T_{b}} (1 - \cos 2\Omega t - \cos 4\pi f_{c}t + \cos 4\pi f_{c}t \cos 2\Omega t) dt$$

$$= 2P \int_{0}^{T_{b}} (1 - \cos 2 \Omega t - \cos 4 \pi f_{c} t + \cos 4 \pi f_{c} t \cos 2 \Omega t) dt$$

By Formula,

$$\frac{\cos{(A+B)} + \cos{(A-B)}}{2} = \cos{A}\cos{B}$$

$$=2P\int_{0}^{T_{b}}\left\{1-\cos 2\Omega t-\cos 4\pi f_{c}t+\frac{1}{2}\cos (4\pi f_{c}t+2\Omega t)+\frac{1}{2}\cos (4\pi f_{c}t-2\Omega t)\right\}dt$$

$$= 2P \left[ T_b - \frac{\sin 2\Omega T_b}{2\Omega} \right]$$

$$= 2PT_b \left[ 1 - \frac{\sin 2\Omega T_b}{2\Omega T_b} \right]$$

$$\int_{0}^{T_{b}} s^{2}(t) dt = 2PT_{b} \left[ 1 - \frac{\sin 2\Omega T_{b}}{2\Omega T_{b}} \right]_{\text{polynomial then}}$$

The ratio  $\frac{\sin 2\,\Omega\,T_b}{2\,\Omega\,T_b}$  attains the largest value when it

$$2 \Omega T_{b} \Rightarrow \frac{3 \pi}{2}$$

$$\Rightarrow 2PT_{b} \left[ 1 - \frac{\sin(3 \pi/2)}{3 \pi/2} \right]$$

$$= 2PT_{b} (1.2121)$$

$$\left( \frac{s_{1}(t) - s_{2}(t)}{\sigma} \right)_{\max}^{2} = \frac{\int_{0}^{T_{b}} s^{2}(t) dt}{\frac{N_{o}}{2}}$$

$$= \frac{2.424PT_{b} \times 2}{N_{o}} = \frac{4.848 PT_{b}}{N_{o}}$$

$$\begin{split} s_1\left(t\right) - s_2\left(t\right) &= 2.20 \; \sqrt{\frac{PT_b}{N_o}} \\ P_e &= \frac{1}{2} \, erfc \left(\frac{s_1\left(t\right) - s_2\left(t\right)}{2 \, \sqrt{2} \; \sigma}\right) \\ &= \frac{1}{2} \, erfc \left(\frac{2.20}{2 \, \sqrt{2}} \, \sqrt{\frac{PT_b}{N_o}}\right) \\ P_e &= \frac{1}{2} \, erfc \, \sqrt{\frac{0.6E_b}{N_o}} \end{split}$$

$$= \frac{1}{2} \operatorname{erfc} \sqrt{\frac{0.6E_b}{N_o}}$$

 $P_e$  for Non-coherent BFSK depends on signal energy

$$\therefore P_e = \frac{1}{2}e^{-\frac{E}{2N_o}}$$

### **Advantages**

- Simple circuit is required
- Generation is easier
- Has constant modulated signal envelope

### **Disadvantages**

- Hardware is more complicated when compared to ASK
- Requires large Bandwidth

#### Limitations

- Bandwidth is greater than 4f<sub>b</sub>, which is almost double the bandwidth of BPSK.
- Error rate of BFSK is more compared to BPSK
- Only half of the transmitted energy carries the information signal