Design of Linear Phase FIR Filters using Windowing Technique

Presented by

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Design Procedure of Linear Phase FIR Filters by Windowing Technique

Given, $H_d(e^{j\omega}) \rightarrow$ Desired Frequency Response Function of the Filter

Step1: From the desired frequency response specification $H_d(e^{j\omega})$, find the corresponding unit impulse response.

$$h_d[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(e^{j\omega}) e^{j\omega n} d\omega$$

Step 2: In most of the practical cases, the obtained impulse response $h_d[n]$ is of infinite duration and is truncated to get an FIR filter of length M by multiplying $h_d[n]$ with window function w[n].

$$h[n] = h_d[n]w[n]$$

where, $w[n] \rightarrow \text{Window Function of length } M$

 $h[n] \rightarrow$ Impulse Response of FIR Filter having length M

Step 3: The transfer function H(z) is obtained from h[n] by taking its z transform.

$$H(z) = \sum_{n=-\infty}^{\infty} h[n]z^{-n} = \sum_{n=0}^{M-1} h[n]z^{-n}$$

Step 4: The magnitude response $|H(e^{j\omega})|$ is obtained from H(z) by substituting $z = e^{j\omega}$.

Summary of Symmetric Window Functions

Rectangular Window,
$$w[n] = \begin{cases} 1 & 0 \le n \le M - 1 \\ 0 & otherwise \end{cases}$$

Bartlett Window, $w[n] = \begin{cases} 1 - \frac{M-1}{2} & 0 \le n \le M - 1 \\ \frac{M-1}{2} & 0 \le n \le M - 1 \end{cases}$

Hanning Window, $w[n] = \begin{cases} 0.5 - 0.5 \cos \frac{2\pi n}{M-1} & 0 \le n \le M - 1 \\ 0 & otherwise \end{cases}$

Hamming Window, $w[n] = \begin{cases} 0.54 - 0.46 \cos \frac{2\pi n}{M-1} & 0 \le n \le M - 1 \\ 0 & otherwise \end{cases}$

Blackman Window, $w[n] = \begin{cases} 0.42 - 0.5 \cos \frac{2\pi n}{M-1} + 0.08 \cos \frac{4\pi n}{M-1} & 0 \le n \le M - 1 \\ 0 & otherwise \end{cases}$

Summary of Frequency Response of Linear Phase FIR Filters

Low Pass Filter

$$H_{d}\left(e^{j\omega}\right) = \begin{cases} e^{-j\omega\tau} & 0 \le |\omega| < \omega_{c} \\ 0 & \omega_{c} \le |\omega| \le \pi \end{cases}$$

High Pass Filter

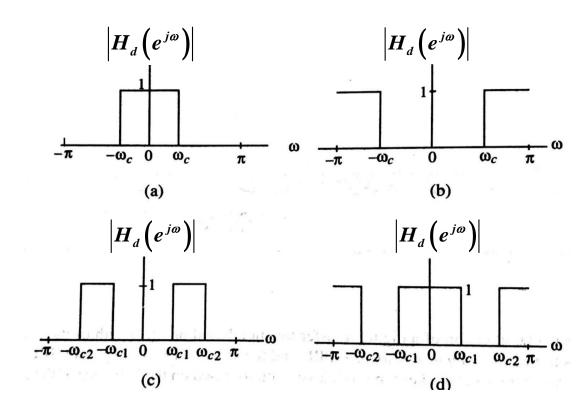
$$H_d\left(e^{j\omega}\right) = \begin{cases} e^{-j\omega\tau} & \omega_c \le |\omega| < \pi \\ 0 & 0 \le |\omega| \le \omega_c \end{cases}$$

Band Pass Filter

$$H_{d}\left(e^{j\omega}\right) = \begin{cases} e^{-j\omega\tau} & \omega_{c1} \leq |\omega| < \omega_{c2} \\ 0 & 0 \leq |\omega| \leq \omega_{c1} \text{ and } \omega_{c2} \leq |\omega| \leq \pi \end{cases}$$

Band Stop Filter

$$H_{d}\left(e^{j\omega}\right) = \begin{cases} e^{-j\omega\tau} & 0 \le |\omega| \le \omega_{c1} \text{ and } \omega_{c2} \le |\omega| \le \pi \\ 0 & \omega_{c1} \le |\omega| < \omega_{c2} \end{cases}$$



Magnitude Response of Ideal Digital Filters
(a) LPF (b) HPF (c) BPF (d) BSF

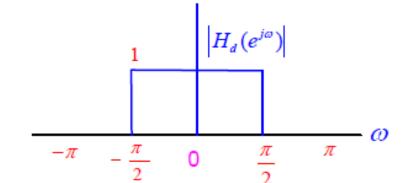
Question: For the specifications given below, design a Symmetric FIR Low Pass Filter using Rectangular Window. Also plot the magnitude response of the designed FIR filter.

Length of the Filter = 11

Cutoff Frequency = $\pi/2$ rad/sample.

Solution:

Given Specifications;
$$M = 11, \omega_c = \frac{\pi}{2} rad / sample, \tau = \frac{M-1}{2} = 5$$



Step 1:

$$H_d\left(e^{j\omega}\right) = \begin{cases} e^{-j\omega\tau} & 0 \le |\omega| < \omega_c \\ 0 & \omega_c \le |\omega| \le \pi \end{cases}$$

$$h_d[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega$$

$$=\frac{1}{2\pi}\int_{-\omega_{c}}^{\omega_{c}}e^{j\omega(n-\tau)}d\omega=\frac{1}{2\pi}\left[\frac{e^{j\omega(n-\tau)}}{j(n-\tau)}\bigg|_{-\omega_{c}}^{\omega_{c}}\right]$$

$$=\frac{1}{(n-\tau)\pi}\left[\frac{e^{j\omega_c(n-\tau)}-e^{-j\omega_c(n-\tau)}}{2j}\right]$$

$$\Rightarrow h_d[n] = \frac{\sin \omega_c(n-\tau)}{\pi(n-\tau)} \quad for \ n \neq \tau$$

for
$$n = \tau$$
; $h_d \left[\tau\right] = \frac{1}{2\pi} \int_{-\omega_c}^{\omega_c} 1 d\omega = \frac{\omega_c}{\pi}$

$$h_d[n] = \frac{\sin(0.5\pi(n-5))}{\pi(n-5)}$$
 for $0 \le n \le 10$ and $n \ne 5$
 $h_d[\tau] = \frac{0.5\pi}{n} = 0.5$ for $n = \tau = 5$

$$h_{d}[0] = \frac{\sin(0.5\pi(0-5))}{\pi(0-5)} = 0.0637 \qquad h_{d}[10] = \frac{\sin(0.5\pi(10-5))}{\pi(10-5)} = 0.0637$$

$$h_{d}[1] = \frac{\sin(0.5\pi(1-5))}{\pi(1-5)} = 0 \qquad h_{d}[9] = \frac{\sin(0.5\pi(9-5))}{\pi(9-5)} = 0$$

$$h_{d}[2] = \frac{\sin(0.5\pi(2-5))}{\pi(2-5)} = -0.1061 \qquad h_{d}[8] = \frac{\sin(0.5\pi(8-5))}{\pi(8-5)} = -0.1061$$

$$h_{d}[3] = \frac{\sin(0.5\pi(3-5))}{\pi(3-5)} = 0 \qquad h_{d}[7] = \frac{\sin(0.5\pi(7-5))}{\pi(7-5)} = 0$$

$$h_{d}[4] = \frac{\sin(0.5\pi(4-5))}{\pi(4-5)} = 0.3183 \qquad h_{d}[6] = \frac{\sin(0.5\pi(6-5))}{\pi(6-5)} = 0.3183$$

$$h_d[5] = \frac{\omega_c}{\pi} = \frac{0.5\pi}{\pi} = 0.5$$

Step 2: Rectangular Window w[n]=1 for $0 \le n \le 10$

$$h[n] = h_d[n]w[n]$$

n	$h_d[n]$	w[n]	h[n]
0	0.0637	1	0.0637
1	0	1	0
2	-0.1061	1	-0.1061
3	0	1	0
4	0.3183	1	0.3183
5	0.5	1	0.5
6	0.3183	1	0.3183
7	0	1	0
8	-0.1061	1	-0.1061
9	0	1	0
10	0.0637	1	0.0637

Step 3:

$$H(z) = \sum_{n=0}^{10} h[n]z^{-n}$$

$$= h[0] + h[1]z^{-1} + h[2]z^{-2} + h[3]z^{-3} + h[4]z^{-4} + h[5]z^{-5} + h[6]z^{-6} + h[7]z^{-7} + h[8]z^{-8} + h[9]z^{-9} + h[10]z^{-1}$$

$$= h[0](1+z^{-10}) + h[1](z^{-1}+z^{-9}) + h[2](z^{-2}+z^{-8}) + h[3](z^{-3}+z^{-7}) + h[4](z^{-4}+z^{-6}) + h[5]z^{-5}$$

$$= z^{-5} \Big[h[0](z^{5}+z^{-5}) + h[1](z^{4}+z^{-4}) + h[2](z^{3}+z^{-3}) + h[3](z^{2}+z^{-2}) + h[4](z^{1}+z^{-1}) + h[5] \Big]$$

$$\Rightarrow H(z) = z^{-5} \Big[0.5 + 0.0637(z^{5}+z^{-5}) - 0.1061(z^{3}+z^{-3}) + 0.3183(z^{1}+z^{-1}) \Big]$$

Step 4:

We have,

$$H(z) = z^{-5} \Big[0.5 + 0.0637 (z^{5} + z^{-5}) - 0.1061 (z^{3} + z^{-3}) + 0.3183 (z^{1} + z^{-1}) \Big]$$

$$H(e^{j\omega}) = H(z) \Big|_{z=e^{j\omega}} = e^{-j5\omega} \Big[0.5 + 0.0637 (e^{j5\omega} + e^{-j5\omega}) - 0.1061 (e^{j3\omega} + e^{-j3\omega}) + 0.3183 (e^{j\omega} + e^{-j\omega}) \Big]$$

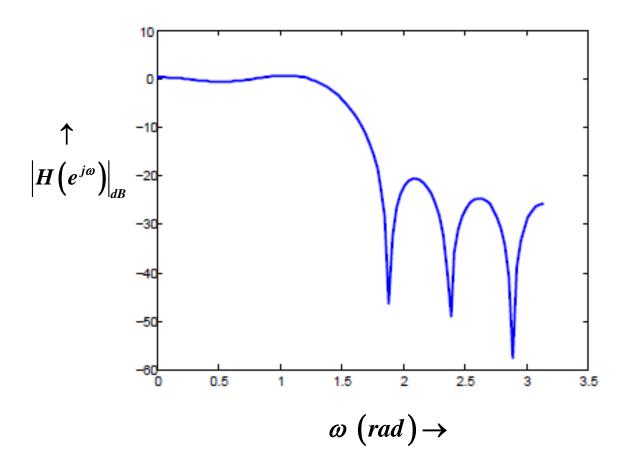
$$\Rightarrow H(e^{j\omega}) = e^{-j5\omega} \Big[0.5 + 0.0637 \times 2\cos 5\omega - 0.1061 \times 2\cos 3\omega + 0.3183 \times 2\cos \omega \Big]$$

$$\Rightarrow H(e^{j\omega}) = e^{-j\omega 5} \Big[0.5 + 0.1273\cos 5\omega - 0.2122\cos 3\omega + 0.6366\cos \omega \Big]$$

$$H(e^{j\omega}) = e^{-j\omega 5} [0.5 + 0.1273\cos 5\omega - 0.2122\cos 3\omega + 0.6366\cos \omega]$$

$$\Rightarrow |H(e^{j\omega})| = |0.5 + 0.1273\cos 5\omega - 0.2122\cos 3\omega + 0.6366\cos \omega|$$

$$|H(e^{j\omega})|_{dB} = 20\log_{10}|H(e^{j\omega})|$$



Question: For the given frequency response specifications, design a **Symmetric FIR High Pass Filter** using **Hamming Window**. Also find the magnitude response of the designed FIR filter.

$$H_{d}(e^{j\omega}) = \begin{cases} e^{-j3\omega} & 1 \le |\omega| < \pi \\ 0 & Otherwise \end{cases}$$

Solution:

Given Specifications; $\tau = 3$, $\omega_c = 1 \text{rad} / \text{sample}$, $M = 2\tau + 1 = 7$

Step 1:

$$h_{d}[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega$$

$$= \frac{1}{2\pi} \left[\int_{-\pi}^{-\omega_{c}} e^{j\omega(n-\tau)} d\omega + \int_{\omega_{c}}^{\pi} e^{j\omega(n-\tau)} d\omega \right]$$

$$= \frac{1}{2\pi} \left[\frac{e^{j\omega(n-\tau)}}{j(n-\tau)} \Big|_{-\pi}^{-\omega_{c}} + \frac{e^{j\omega(n-\tau)}}{j(n-\tau)} \Big|_{\omega_{c}}^{\pi} \right]$$

$$= \frac{1}{(n-\tau)\pi} \left[\frac{e^{-j\omega_{c}(n-\tau)} - e^{-j\pi(n-\tau)}}{2j} + \frac{e^{j\pi(n-\tau)} - e^{j\omega_{c}(n-\tau)}}{2j} \right]$$

$$= \frac{1}{(n-\tau)\pi} \left[\frac{(e^{j\pi(n-\tau)} - e^{-j\pi(n-\tau)})}{2j} - \frac{(e^{j\omega_{c}(n-\tau)} - e^{-j\omega_{c}(n-\tau)})}{2j} \right]$$

$$\Rightarrow h_{d}[\tau] = 1 - \frac{\omega_{c}}{\pi}$$

$$|H_{d}(e^{j\omega})|$$

$$\Rightarrow h_{d}[n] = \frac{\left[\sin \pi (n-\tau) - \sin \omega_{c}(n-\tau)\right]}{(n-\tau)\pi} \quad \text{for } n \neq \tau$$

$$\frac{\omega_{c}(n-\tau)}{2\pi}$$

$$\int \sigma n = \tau; \quad h_{d}[\tau] = \frac{1}{2\pi} \left[\int_{-\pi}^{-\omega_{c}} 1 d\omega + \int_{\omega_{c}}^{\pi} 1 d\omega\right] = \frac{-\omega_{c} + \pi + \pi - \omega_{c}}{2\pi}$$

$$h_d[n] = \frac{\left[\sin \pi (n-3) - \sin (n-3)\right]}{(n-3)\pi} \quad \text{for } 0 \le n \le 6 \text{ and } n \ne 3$$

$$h_d[\tau] = 1 - \frac{\omega_c}{\pi} = 1 - \frac{1}{\pi} \quad \text{for } n = \tau = 3$$

$$h_{d}[0] = \frac{\left[\sin \pi (0-3) - \sin(0-3)\right]}{(0-3)\pi} = -0.0150$$

$$h_{d}[1] = \frac{\left[\sin \pi (1-3) - \sin(1-3)\right]}{(1-3)\pi} = -0.1447$$

$$h_{d}[2] = \frac{\left[\sin \pi (2-3) - \sin(2-3)\right]}{(2-3)\pi} = -0.2678$$

$$h_{d}[4] = \frac{\left[\sin \pi (4-3) - \sin(4-3)\right]}{(4-3)\pi} = -0.2678$$

$$h_d[3] = 1 - \frac{1}{\pi} = 0.6817$$

Step 2:

Hamming Window

$$w[n] = \begin{cases} 0.54 - 0.46\cos\frac{2\pi n}{M-1} & 0 \le n \le M-1 \\ 0 & otherwise \end{cases}$$

$$w[0] = 0.0800$$
 $w[6] = 0.0800$
 $w[1] = 0.3100$ $w[5] = 0.3100$
 $w[2] = 0.7700$ $w[4] = 0.7700$
 $w[3] = 1.0000$

$$h[n] = h_d[n]w[n]$$

n	$h_d[n]$	w[n]	h[n]
0	-0.0150	0.0800	-0.001197
1	-0.1447	0.3100	-0.044862
2	-0.2678	0.7700	-0.206243
3	0.6817	1	0.6817
3 4	0.6817 -0.2678	1 0.7700	0.6817 -0.206243

Step 3:

$$H(z) = \sum_{n=0}^{6} h[n]z^{-n}$$

$$= h[0] + h[1]z^{-1} + h[2]z^{-2} + h[3]z^{-3} + h[4]z^{-4} + h[5]z^{-5} + h[6]z^{-6}$$

$$= z^{-3} \Big[h[3] + h[0](z^{3} + z^{-3}) + h[1](z^{2} + z^{-2}) + h[2](z^{1} + z^{-1}) \Big]$$

$$\Rightarrow H(z) = z^{-3} \Big[0.6817 - 0.001197(z^{3} + z^{-3}) - 0.044862(z^{2} + z^{-2}) - 0.206243(z^{1} + z^{-1}) \Big]$$

Step 4:

We have,

$$H(z) = z^{-3} \left[0.6817 - 0.001197(z^{3} + z^{-3}) - 0.044862(z^{2} + z^{-2}) - 0.206243(z^{1} + z^{-1}) \right]$$

$$H\left(e^{j\omega}\right) = H\left(z\right)\Big|_{z=e^{j\omega}} = e^{-j3\omega}\Big[0.6817 - 0.001197\left(e^{j3\omega} + e^{-j3\omega}\right) - 0.044862\left(e^{j2\omega} + e^{-j2\omega}\right) - 0.206243\left(e^{j\omega} + e^{-j\omega}\right)\Big]$$

$$\Rightarrow H(e^{j\omega}) = e^{-j3\omega} \left[0.6817 - 0.001197 \times 2\cos 3\omega - 0.044862 \times 2\cos 3\omega - 0.044862 \times 2\cos \omega \right]$$

$$\Rightarrow H(e^{j\omega}) = e^{-j3\omega} \left[0.6817 - 0.0023957 \cos 3\omega - 0.0897258 \cos 2\omega - 0.41248674 \cos \omega \right]$$

$$|H(e^{j\omega})| = |0.6817 - 0.0023957\cos 3\omega - 0.0897258\cos 2\omega - 0.41248674\cos \omega|$$