

# Lecture 3: Experiment 2

## EE380 (Control Systems)

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August 12, 2011



Back

Forward

Close

# Contents

<b>I</b>	<b>Procedure: Discussed in Lecture 3</b>	<b>4</b>
1	Outline of the experiment	5
2	Tasks common to all 6 experiments	6
3	Homework (HW) vs. Lab work (LW)	7
4	Dead zone and how to overcome it	8
5	Least squares sys-id	9
6	Loop-shaping (1/5): Typical $G_{\text{des}}$	10
7	Loop-shaping (2/5): Example	11
8	Loop-shaping (3/5): $\zeta \longleftrightarrow M_p \longleftrightarrow \text{PM}$	12
9	Loop-shaping (4/5): $M_p \longleftrightarrow \text{DD} \longleftrightarrow \text{PM}$	13
10	Loop-shaping (5/5): Determination of $\omega_g$	14


[Back](#)
[Forward](#)
[Close](#)

11 Discretization	15
12 Simulate; LW: C code, Implement, Analyze	16
13 Work involved	17
<b>II Least squares sys-id theory: Part of Lecture 4</b>	<b>18</b>
14 Bilinear transform and Z-transform	19
15 $G(s) \longleftrightarrow G(z)$	20
16 How Z-transform used in our sys-id	21
17 What is least squares sys-id? (1/2)	22
18 What is least squares sys-id? (2/2)	23



Back

Forward

Close

# Procedure: Discussed in Lecture 3



Back

Forward

Close

# Outline of the experiment

Want speed of motor to track a sinusoid. Steps:

- Least squares system identification (sys-id).
- Recognition and compensation of plant's dead zone.
- Design controller using loop-shaping.
- Simulation on PC.
- Deployment on experimental setup.



Back

Forward

Close

# Tasks common to all 6 experiments

## Simulation

- Perform PC-based simulation of CL system using GNU Octave.
- Perform PC-based simulation of digital control of a continuous-time system using GNU Octave.

## Realization on hardware

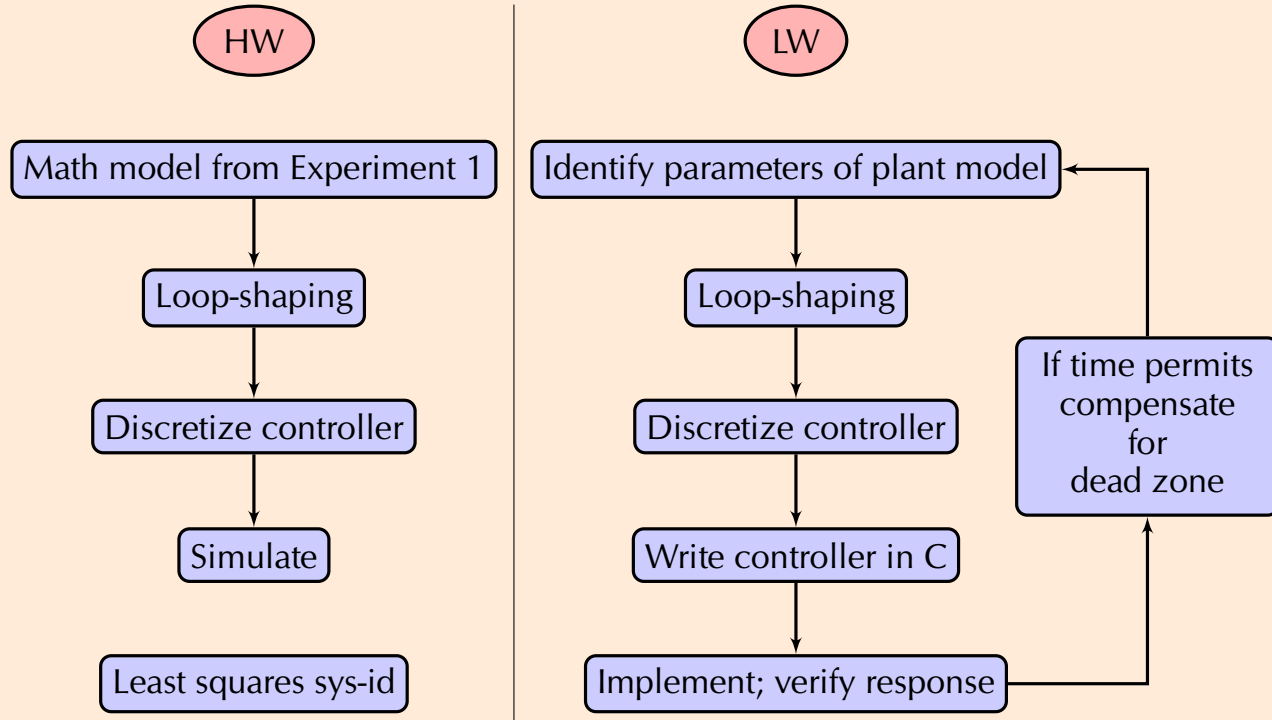
- Utilize the various components of an integrated development environment (IDE): editor, compiler, linker, debugger, and programmer to program a  $\mu$ C.
- Program controller using C language into  $\mu$ C.
- Monitoring: read data into PC from  $\mu$ C using UART modules.

## Analysis

- Compare actual performance with predicted performance.

[Back](#)[Forward](#)[Close](#)

# Homework (HW) vs. Lab work (LW)



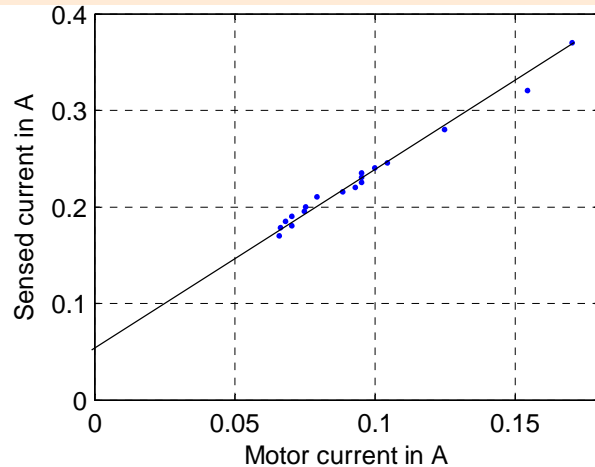
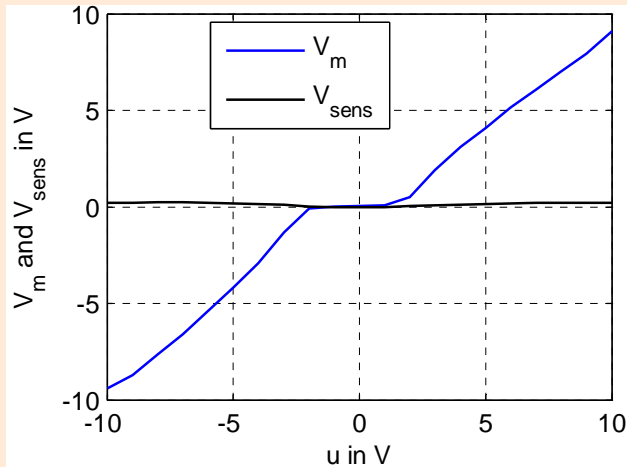
Back

Forward

Close

# Dead zone and how to overcome it

If  $u < 2$  V when  $V_s = 12$  V, then  $V_m$  will be zero.



Overcoming dead zone

```
if(u<0&&u>-2)
    u = u - 2;
else if(u>0&&u<2)
    u = u + 2;
```



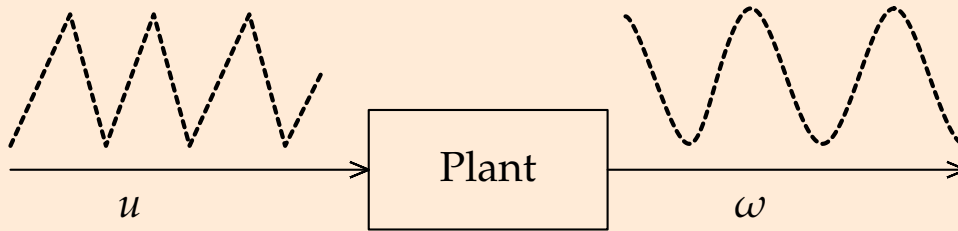
Back

Forward

Close



# Least squares sys-id



- Place C code of triangular waveform generator in `main-prog.c`.
- Prepare setup to apply  $u$  in OL and collect  $\omega$  from motor.
- Input  $u_1, u_2, u_3, \dots$  forming the triangular waveform.
- Collect  $u_1, u_2, u_3, \dots$  and  $\omega_1, \omega_2, \omega_3, \dots$  into `terminal.log`.
- Form `readSID.m` by replacing plots-related section of `readplot.m` with sys-id code from `sysid.m`.
- Execute `readSID.m` to obtain  $K, a, b$  of  $\frac{K}{s^2 + as + b}$ .

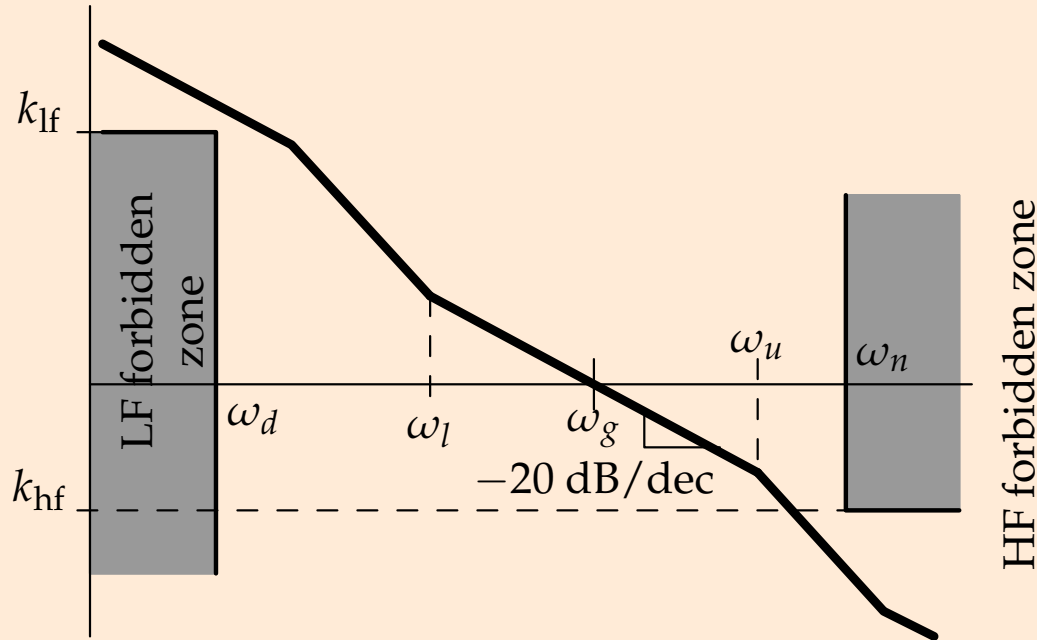


Back

Forward

Close

# Loop-shaping (1/5): Typical $G_{\text{des}}$



# Loop-shaping (2/5): Example

2. The uncompensated unity feedback CL TF of a system is  $M(s) = \frac{1}{s^2+1}$ . Design a compensator (decide between lead and lag) that will provide the CL system's step response a damping coefficient of  $\zeta \geq 0.55$ .

step 1:

The OL TF is

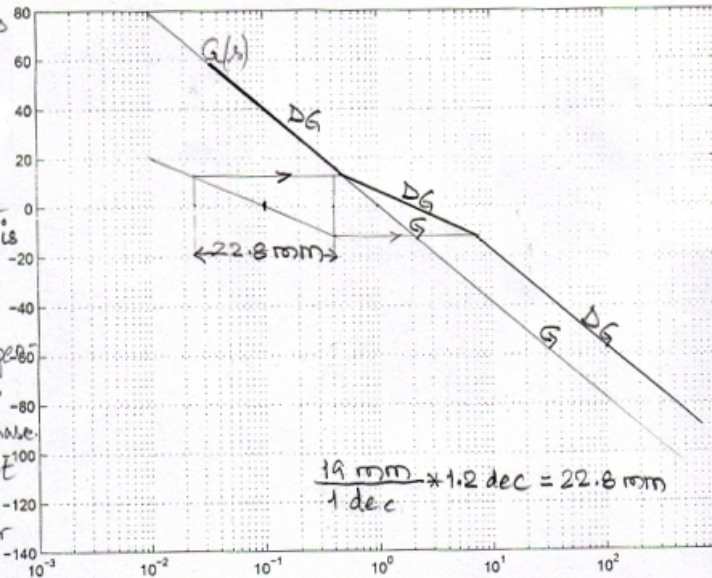
$$G(s) = \frac{1}{s^2}$$

step 2:

The phase of  $G(j\omega)$  is  $-180^\circ$  across all  $\omega$ .

A lag compensator adds negative phase. So, it cannot create a (+ve) PM for this  $G(s)$ .

So, we choose a lead compensator.



step 3:

$$\zeta \geq 0.55 \Rightarrow PM \geq 55^\circ$$

$\Rightarrow$  Let's decide to provide a  $PM = 60^\circ$ . For this we will create  $DG$  such that it has a  $-20$  dB/decade section that is  $1.2$  dec. wide and is centered about  $\omega_g$ . Also, since there are no other spec-s, we can retain the present value of  $K_a$ .

Step 4: Construct a  $22.8$  mm wide  $-20$  dB/dec section symmetrically positioned about  $0.1$  rad/s as shown. Use this section to complete the ABMP of  $DG$  as shown.

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Assignment posted on March 20, 2009.

1 of 2

$$DG = G_{des}$$



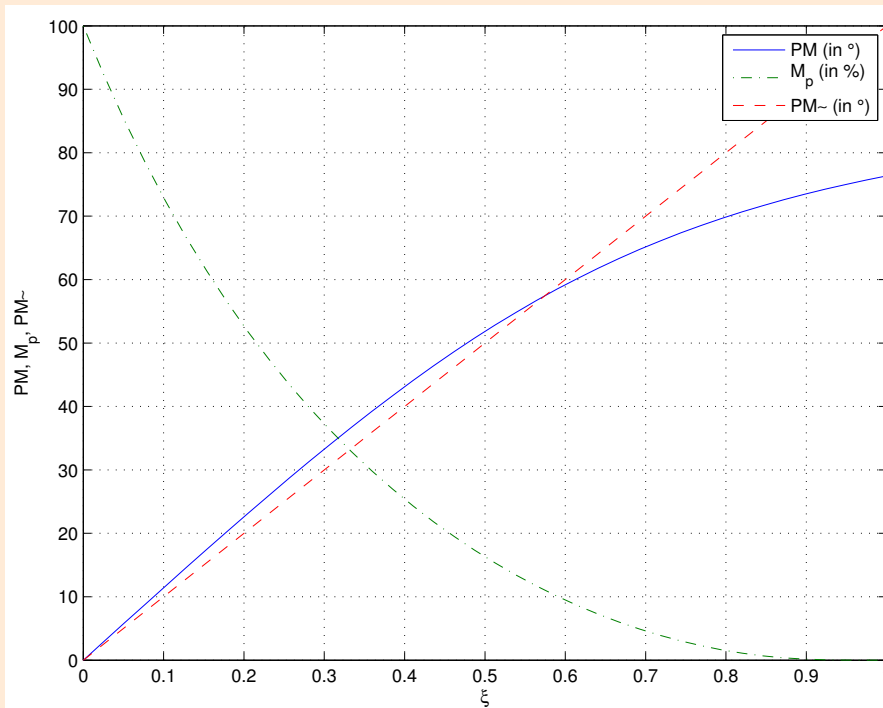
Back

Forward

Close

# Loop-shaping (3/5): $\zeta \longleftrightarrow M_p \longleftrightarrow \text{PM}$

From Lecture 26 of EE250-2011



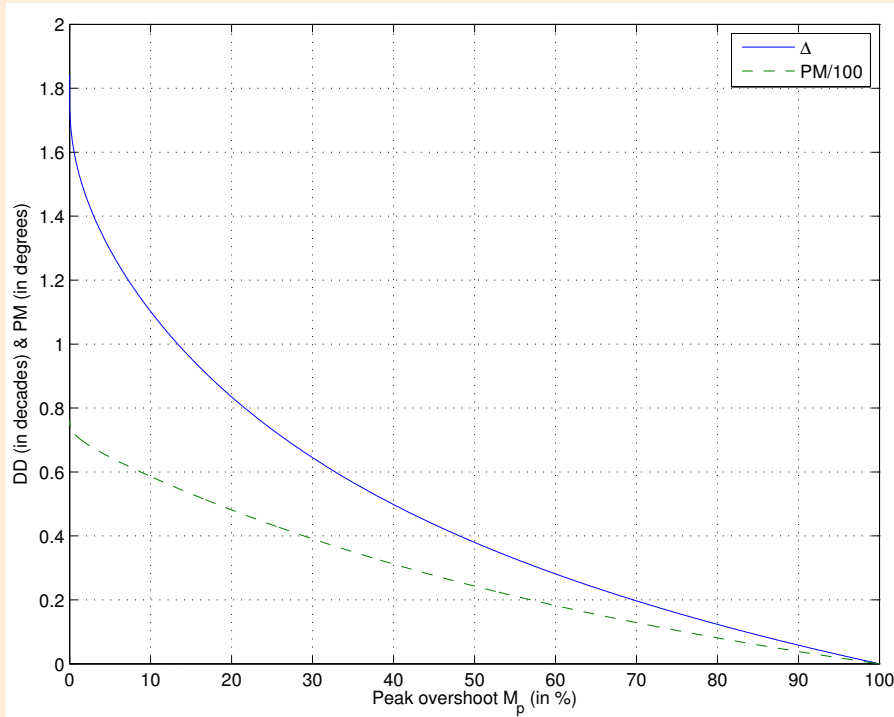
Back

Forward

Close

# Loop-shaping (4/5): $M_p \longleftrightarrow DD \longleftrightarrow PM$

From Lecture 26 of EE250-2011

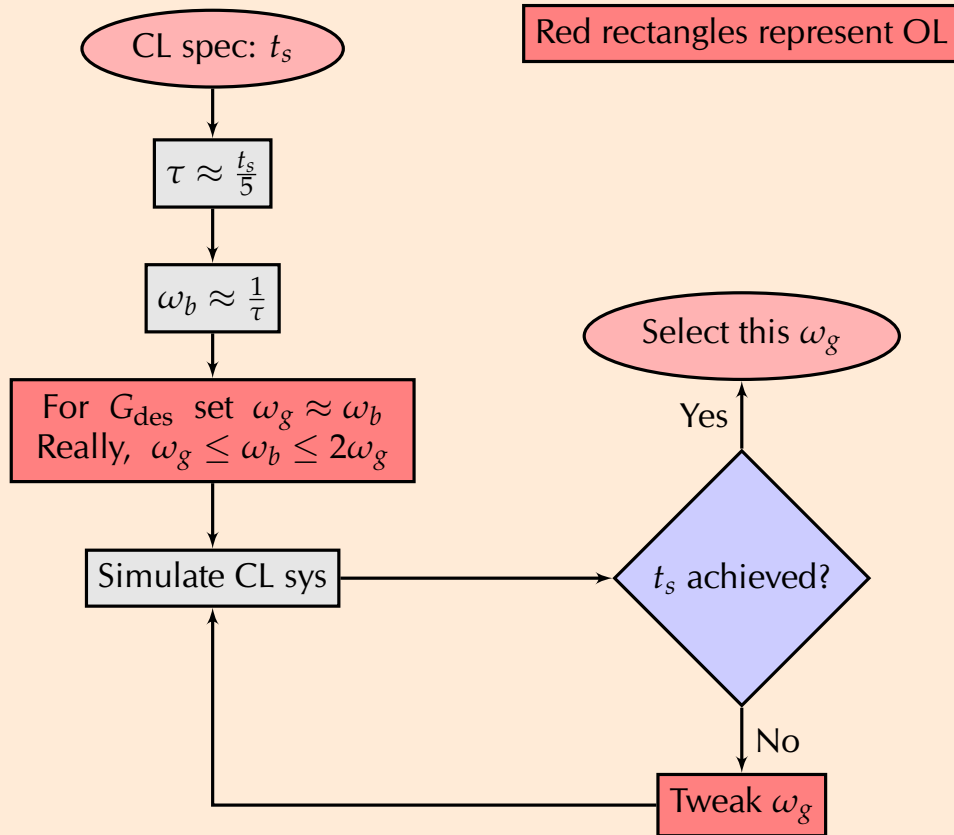


Back

Forward

Close

# Loop-shaping (5/5): Determination of $\omega_g$



# Discretization

$$\frac{a_1 s + a_0}{s^2 + b_1 s + b_0}$$

Simulation  
diagram

$$\begin{aligned}\dot{x}_1 &= x_2 \\ \dot{x}_2 &= -b_0 x_1 - b_1 x_2 + u \\ y &= a_0 x_1 + a_1 x_2\end{aligned}$$

Euler's  
approximation

$$\begin{aligned}x_{1(k+1)} &= x_{1(k)} + \Delta t x_{2(k)} \\ x_{2(k+1)} &= -b_0 \Delta t x_{1(k)} + (1 - b_1 \Delta t) x_{2(k)} + \Delta t u_k \\ y_k &= a_0 x_{1(k)} + a_1 x_{2(k)}\end{aligned}$$



Back

Forward

Close

# Simulate; LW: C code, Implement, Analyze

- Simulation: `easysim.m`
- Discretized controller  
→ C code:
- Implement: As in demo slides
- Analyze: Compare results

$$\begin{aligned}x_1(k+1) &= a_{11}x_1(k) + a_{12}x_2(k) + b_1u(k) \\x_2(k+1) &= a_{21}x_1(k) + a_{22}x_2(k) + b_2u(k) \\y(k) &= c_1x_1(k) + c_2x_2(k) + du(k)\end{aligned}$$

In `main-prog.c` before `main()` insert `float x1[2],x2[2];`  
In `main()` insert `x1[0] = x2[0] = 0;`

```
x1[1] = a11 * x1[0] + a12 * x2[0] + b1 * u;
x2[1] = a21 * x1[0] + a22 * x2[0] + b2 * u;
y = c1 * x1[0] + c2 * x2[0] + d * u;
x1[0] = x1[1];
x2[0] = x2[1];
```



Back

Forward

Close



## Work involved

- Unless order of controller changes in redesign, don't have to use log paper in lab.
- If prepared how to combine `readplot.m` & `sysid.m`, then:

$$Q9 + Q10 \leq 30 \text{ minutes}$$

$$Q11 \leq 30 \text{ minutes}$$

$$Q12 \leq 20 \text{ minutes}$$

$$Q13 + Q15 \leq 10 \text{ minutes}$$

---


$$\text{Total: 1 hour 30 minutes}$$


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- Remove compensation of dead zone and repeat.



Back

Forward

Close

# Least squares sys-id theory: Part of Lecture 4

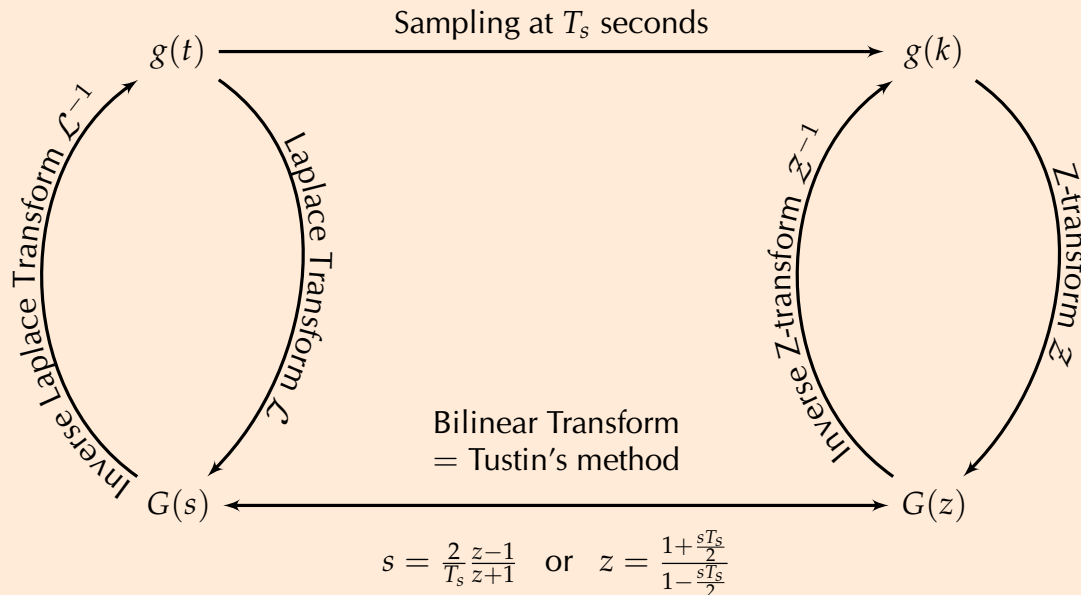


Back

Forward

Close

# Bilinear transform and Z-transform



- Both  $s$ -domain &  $z$ -domain are fictitious domains.
- $s$ -domain simplifies working with differential equations;  $z$ -domain simplifies working with difference equations.
- Bilinear transform is not the only way to go  $G(s) \leftrightarrow G(z)$ .
- $T_s$  is constrained by Nyquist sampling frequency.

$$G(s) \longleftrightarrow G(z)$$

- Consider definitions of  $\mathcal{L}$  and  $\mathcal{Z}$

$$Y(s) = \mathcal{L} \{y(t)\} \triangleq \int_{t=0}^{\infty} y(t) e^{-st} dt$$

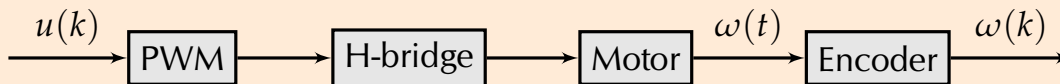
$$Y(z) = \mathcal{Z} \{y(k)\} \triangleq \sum_{k=0}^{\infty} y(k) z^{-k}$$

- Comparison suggests  $z = e^{sT_s}$ .
- To convert  $G(s)$  to  $G(z)$ , can substitute  $s = \frac{\ln z}{T_s}$ .
- Easier to work with an approximation

$$z = e^{sT_s} = e^{\frac{sT_s}{2}} e^{\frac{sT_s}{2}} = \frac{e^{\frac{sT_s}{2}}}{e^{-\frac{sT_s}{2}}} = \frac{1 + \frac{(\frac{sT_s}{2})}{1!} + \frac{(\frac{sT_s}{2})^2}{2!} + \dots}{1 + \frac{(-\frac{sT_s}{2})}{1!} + \frac{(-\frac{sT_s}{2})^2}{2!} + \dots} \approx \frac{1 + \frac{sT_s}{2}}{1 - \frac{sT_s}{2}}$$


[Back](#)
[Forward](#)
[Close](#)

# How Z-transform used in our sys-id



- $u(k)$  denotes sample of  $u(t)$  at sampling instant  $t = kT_s$ .
- Let  $u(k) \rightarrow \omega(k)$  TF be  $G(z)$ .
- Use  $u(k), \omega(k)$  pairs to build  $G(z)$ .
- Use bilinear transform to go from  $G(z)$  to  $G(s)$ .

Important property of Z-transform used:

$$z^{-l}X(z) \leftrightarrow x(k-l) \quad \text{given} \quad X(z) \leftrightarrow x(k).$$



Back

Forward

Close

# What is least squares sys-id? (1/2)

- Let  $G(z) = \frac{b_1z^2 + b_2z + b_3}{z^3 + a_1z^2 + a_2z + a_3} = \frac{Y(z)}{U(z)}$ .

- Cross multiply:

$$b_1z^2U(z) + b_2zU(z) + b_3U(z) = z^3Y(z) + a_1z^2Y(z) + a_2zY(z) + a_3Y(z).$$

- Multiply throughout by  $z^{-3}$ :

$$b_1z^{-1}U(z) + b_2z^{-2}U(z) + b_3z^{-3}U(z) = Y(z) + a_1z^{-1}Y(z) + a_2z^{-2}Y(z) + a_3z^{-3}Y(z).$$

- Take  $\mathcal{Z}^{-1}$  to obtain difference equation

$$b_1u(k-1) + b_2u(k-2) + b_3u(k-3) = y(k) + a_1y(k-1) + a_2y(k-2) + a_3y(k-3).$$



Back

Forward

Close

## What is least squares sys-id? (2/2)

Consider 
$$b_1u(k-1) + b_2u(k-2) + b_3u(k-3) = y(k) + a_1y(k-1) + a_2y(k-2) + a_3y(k-3). \quad (1)$$

- Let  $\sigma = [b_1 \quad b_2 \quad b_3 \quad -a_1 \quad -a_2 \quad -a_3]^\top$ .
- Suppose we have data of  $u(k)$  and  $y(k)$  for  $k = 0, 1, \dots, N$ .
- Problem: Find  $\sigma$  such that (1) holds for this data.  
I.E., find parameters of a TF that fits to input-output data.

- Let error in the fit be

$$\begin{aligned} \varepsilon(k, \sigma) = & b_1u(k-1) + b_2u(k-2) + b_3u(k-3) - y(k) \\ & - a_1y(k-1) - a_2y(k-2) - a_3y(k-3). \end{aligned}$$

- Modified problem: Find  $\sigma$  to minimize  $\mathcal{J}(\sigma) \triangleq \sum_{k=0}^N \varepsilon^2(k, \sigma)$ .
- If  $\mathcal{J}(\sigma = \sigma_0) = 0$ , then find best estimate  $\hat{\sigma}$  of  $\sigma_0$ .