

Physical Communication

Implementaion of OFDM Transceiver

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CONTENT

1. Introduction
2. Transmitter
3. Channel
4. Receiver
5. Results
6. Conclusions and Discussion

1. Introduction
2. Transmitter
3. Channel
4. Receiver
5. Results
6. Conclusions and Discussion

INTRODUCTION

1. **Goal of the project:** Make a transmit/receive simulation model for a generic OFDM simulation.
2. **Future work:** Compliant with the IEEE 802.11p standard.

MODEL STRUCTURE

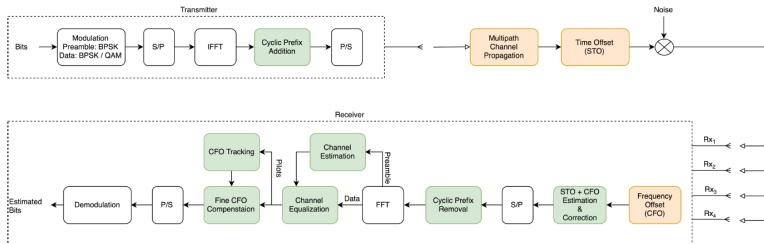


Figure: OFDM Model Structure.

1. Introduction
2. Transmitter
3. Channel
4. Receiver
5. Results
6. Conclusions and Discussion

TRANSMITTER OVERVIEW

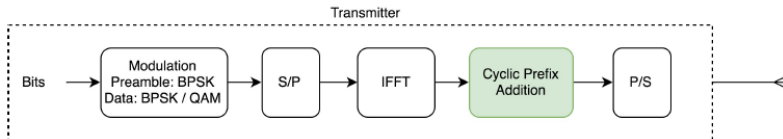


Figure: Transmitter Structure.

MODULATOR

QAM: amplitude modulation with two orthogonal carriers.

Gray Code: two successive values differ in only one bit.

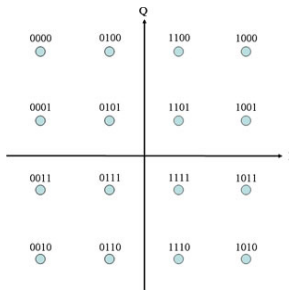


Figure: 16-QAM Signal Constellation with Gray Coding.

S-P/P-S CONVERTER

Serial to Parallel during modulation: The input data is formatted into 4 bits/symbol(16-QAM), 10 symbols/subcarrier and with 90 subcarriers. The goal is to perform IFFT calculation on each subcarrier at the same time.

Parallel to Serial after adding cyclic prefix and before transmission: Transmit the serial data with one antenna.

IFFT on transmitter: multi-carrier modulation.

FFT on receiver: multi-carrier demodulation.

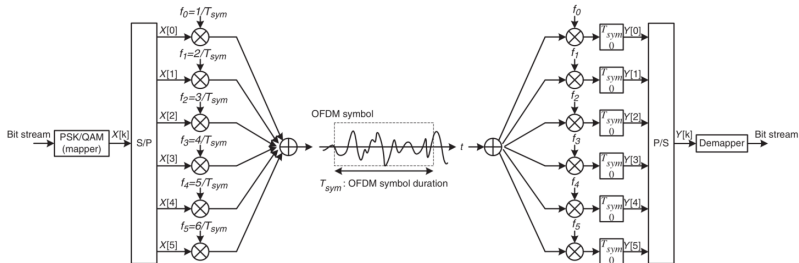
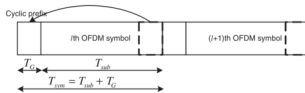


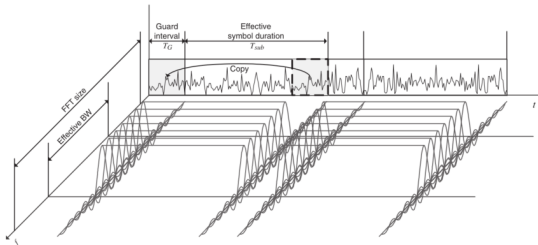
Figure: IFFT and FFT at transmitter and receiver side.

A pilot symbol is a complete OFDM symbol where the value of each subcarrier is predefined and known in transmitter and receiver. It is repeated with a certain rate that depends on how fast the channel changes. The received signal is correlated with the pilot symbol to detect the OFDM symbol start. It can also be used for channel estimation. In our experiment, the pilot frequency representation is $5 + 5i$ and there are 6 groups of pilot tones.

Cyclic Prefix: OFDM guard interval.



(a) OFDM symbols with CP



(b) Time/frequency-domain description of OFDM symbols with CP

Figure: Cyclic Prefix Illustration.

1. Introduction
2. Transmitter
3. Channel
4. Receiver
5. Results
6. Conclusions and Discussion

Channel

Multipath fading

MULTIPATH FADING

THE SIGNAL ARRIVES FROM DIFFERENT PATHS WITH DIFFERENT DELAYS

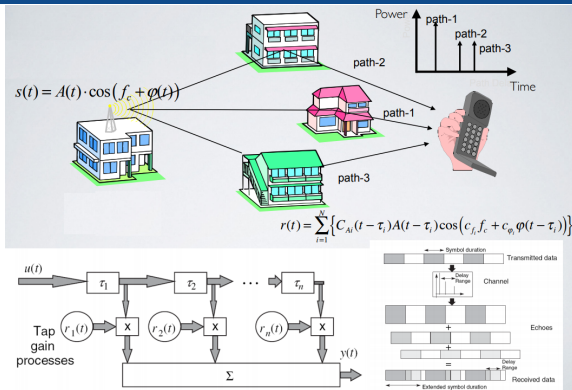


Figure: Multipath Transmission

Leads to:

- ▶ Signal Fading
- ▶ Signal dispersion and distortion

Channel

General Idea (Transfer function)

TRANSFER FUNCTION

THE INTERPOLATED SIGNAL IS CALCULATED

- ▶ Based on the Channel estimate.
- ▶ Difference between estimators' theoretical values and true value estimated quantity.
- ▶ Use of Location of pilot sequence
- ▶ Calculation of Means square Error of Estimator

1. Introduction
2. Transmitter
3. Channel
4. Receiver
5. Results
6. Conclusions and Discussion

SYNCHRONIZATION FOR OFDM

HOW CAN WE CORRELATE THE SIGNAL TO THE REAL ONE

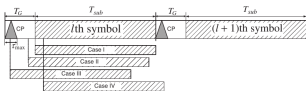
1. OFDM: Orthogonal subcarriers for parallel transmission
2. Advantage only when orthogonality is maintained
3. If not orthogonal, the performance may be degraded
4. Interference (Inter-symbol interference, Inter-carrier interference)

Receiver

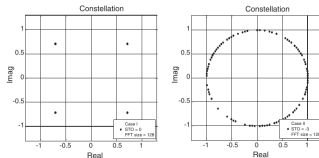
Symbol Time Offset (STO)

THE SYMBOL HAS TO BE ALLIGNED

1. Symbol-timing synchronization detects the starting point per OFDM symbol
2. STO can be calculated in both domains
3. Time is simpler to calculate
4. Phase offset may be created



(a) Cases of STO



(b) Comparison of Signal with and without STO

STO ESTIMATION TECHNIQUES

MORE COMPLEX TECHNIQUES CAN LEAD TO BETTER ESTIMATION

Other Estimation techniques exist in the Time domain:

1. Previous technique can calculate the maximum likelihood instead
2. Training symbol can be used (Symbol division in two parts and correlation)
3. Minimum Squared Difference can be calculated for the maximum correlation
4. Cross-Correlation (Between Training Symbol and received signal)

STO can be calculated in the Frequency domain as well estimating the phase difference.

Receiver

Carrier Frequency Offset (CFO)

FREQUENCY SHIFT (ϵ) LEADS TO ICI

- ▶ f_c equals to transmitter carrier frequency
- ▶ f'_c equals to receiver carrier frequency
- ▶ δf equals to the subcarrier spacing
- ▶ $f_d = \frac{v \cdot f_c}{c}$
- ▶ $\epsilon = \frac{f_c - f'_c}{f_d}$

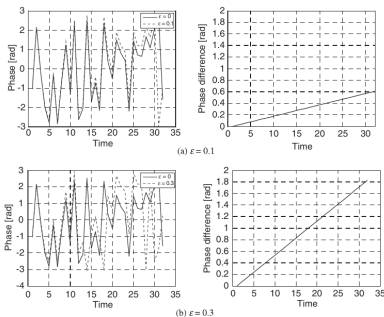


Figure: Comparison of Signal with different CFO's

CFO ESTIMATION TECHNIQUES

CAN BE DONE ON TIME AND FREQUENCY DOMAIN

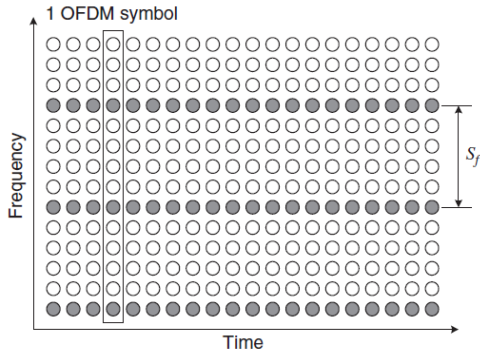
- ▶ The same techniques as in STO can be applied (CP, Training Symbol)
- ▶ Conversion of the phase rotation has to be made
- ▶ The frequency domain is simpler
- ▶ Pilot tones can be used to track CFO

Receiver

Channel estimation

RECEIVER

CHANNEL ESTIMATION



$$\blacktriangleright S_f \leq \frac{1}{\sigma_{\max}}$$

Figure: Comb-type pilot arrangement

- Suitable for fast-fading channels, but not for frequency-selective channels

► LS Channel Estimation

The least-square (LS) channel estimation method finds the channel estimate $\hat{\mathbf{H}}$ in such a way that the following cost function is minimized:

$$\begin{aligned} J(\hat{\mathbf{H}}) &= \|\mathbf{Y} - \mathbf{X}\hat{\mathbf{H}}\|^2 \\ &= (\mathbf{Y} - \mathbf{X}\hat{\mathbf{H}})^H (\mathbf{Y} - \mathbf{X}\hat{\mathbf{H}}) \\ &= \mathbf{Y}^H \mathbf{Y} - \mathbf{Y}^H \mathbf{X} \hat{\mathbf{H}} - \hat{\mathbf{H}}^H \mathbf{X}^H \mathbf{Y} + \hat{\mathbf{H}}^H \mathbf{X}^H \mathbf{X} \hat{\mathbf{H}} \end{aligned}$$

$$\hat{\mathbf{H}}_{LS} = (\mathbf{X}^H \mathbf{X})^{-1} \mathbf{X}^H \mathbf{Y} = \mathbf{X}^{-1} \mathbf{Y}$$

RECEIVER

CHANNEL ESTIMATION

► Data interpolation

```
if pilot_loc(1)>1
    slope = (H(2)-H_est(1))/(pilot_loc(2)-pilot_loc(1));
    H = [H(1)-slope*(pilot_loc(1)-1) H];
    pilot_loc = [1 pilot_loc];
end

if pilot_loc(end) <n_subcarriers
    slope = (H(end)-H(end-1))/(pilot_loc(end)-pilot_loc(end-1));
    H = [H H(end)+slope*(n_subcarriers-pilot_loc(end))];
    pilot_loc = [pilot_loc n_subcarriers];
end

H_interpolated=interp1(pilot_loc,H,(1:n_subcarriers));
```

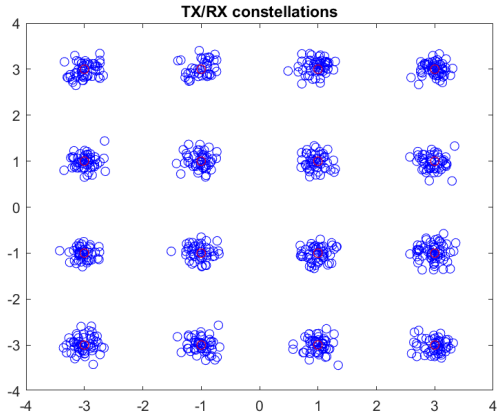
1. Introduction
2. Transmitter
3. Channel
4. Receiver
5. Results
6. Conclusions and Discussion

Results

Transmitter side

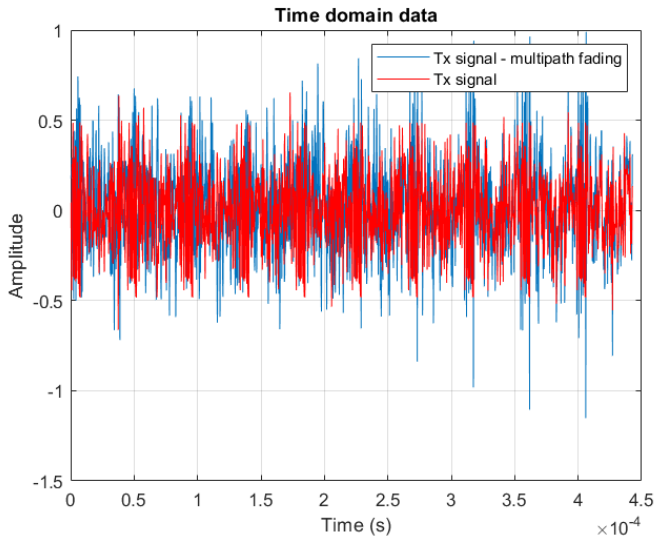
RESULTS

TRANSMITTER SIDE - TX CONSTELLATIONS



RESULTS

TRANSMITTER SIDE - TIME DOMAIN SIGNAL

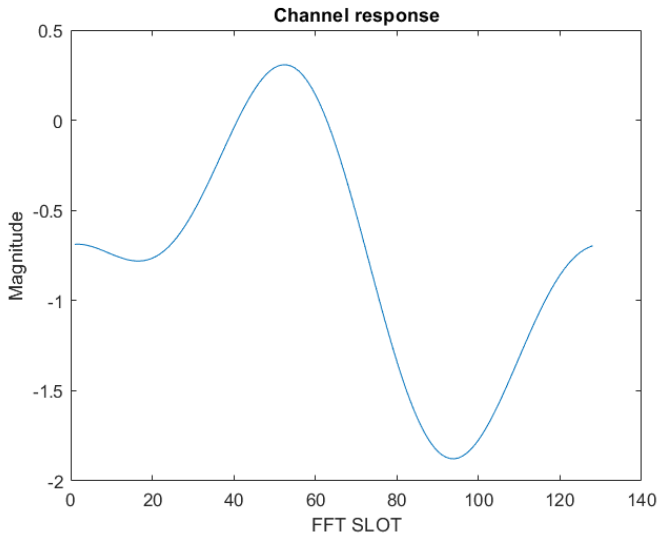


Results

Channel transfer function

RESULTS

CHANNEL TRANSFER FUNCTION



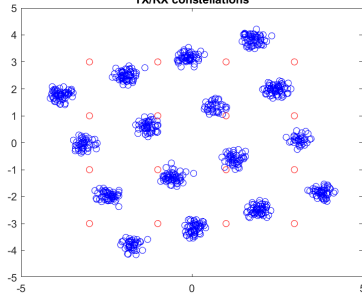
Results Compensation

RESULTS

PHASE COMPENSATION

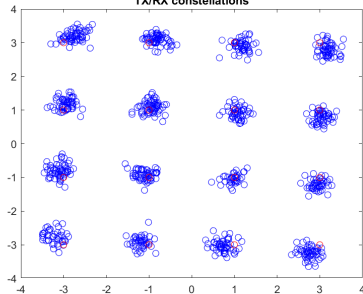
FREQUENCY OFFSET = 0; PHASE OFFSET = 20

WITHOUT PHASE TRACKING
TX/RX constellations



```
tx data != rx data  
>> main
```

WITH PHASE TRACKING
TX/RX constellations



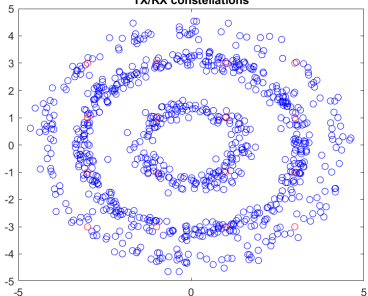
```
tx data = rx data  
f4 >>
```

RESULTS

FREQUENCY COMPENSATION

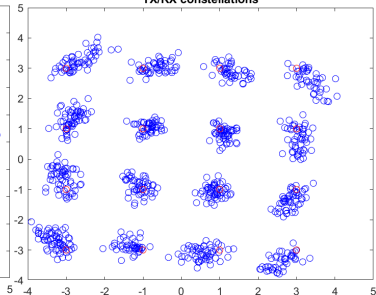
FREQUENCY OFFSET = 7000; PHASE OFFSET = 0;

WITHOUT CFO ESTIMATION
TX/RX constellations



```
tx data := rx data  
>> main
```

WITH CFO ESTIMATION
TX/RX constellations



```
tx data = rx data  
fx >>
```

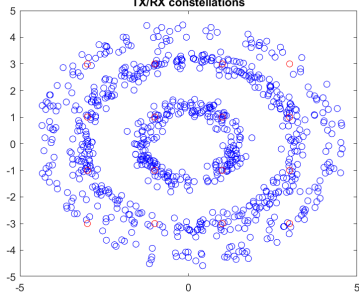
RESULTS

FREQUENCY AND PHASE COMPENSATION

FREQUENCY OFFSET = 7000, PHASE OFFSET = 20

NO PHASE AND CFO ESTIMATION

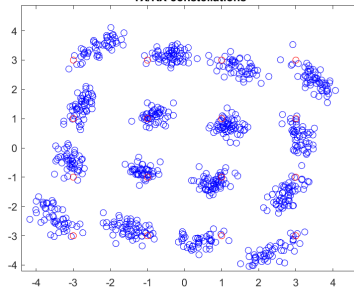
TX/RX constellations



```
tx data != rx data  
>> main
```

WITH PHASE AND CFO ESTIMATION

TX/RX constellations



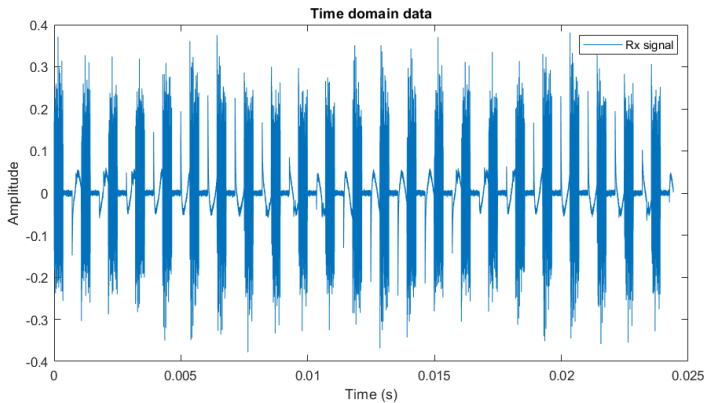
```
tx data = rx data  
fz >>
```


Results

Real world - TX signal

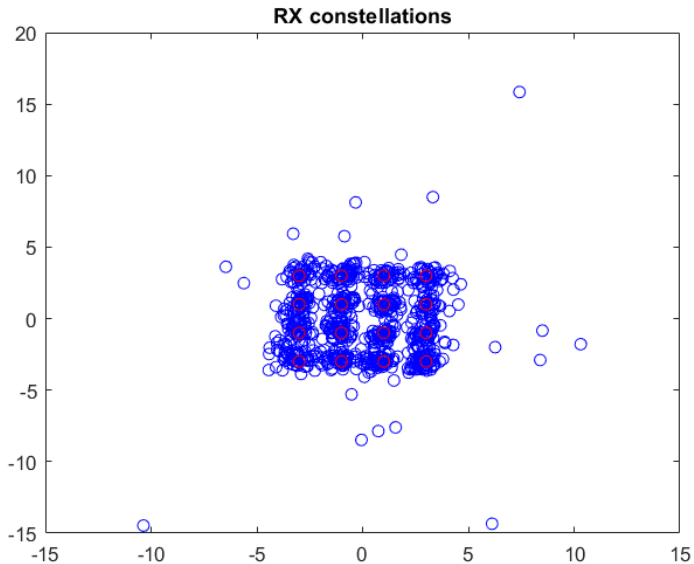
RESULTS

REAL WORLD - TX SIGNAL - SCENARIO 1



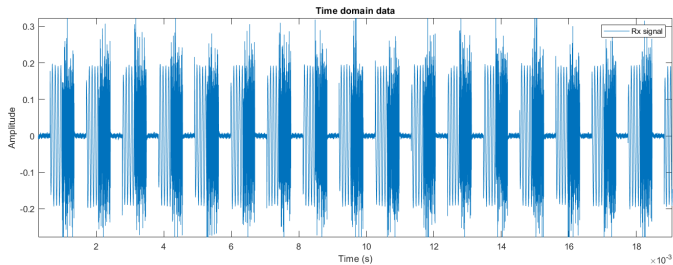
RESULTS

REAL WORLD - RX SIGNAL - SCENARIO 1



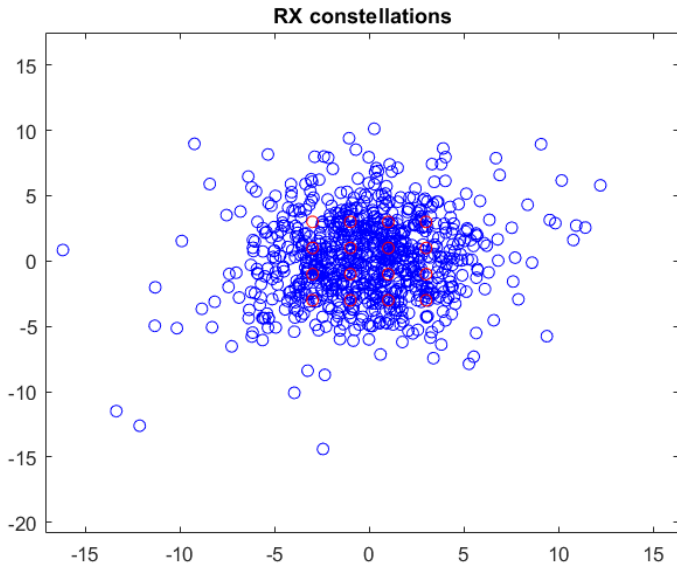
RESULTS

REAL WORLD - TX SIGNAL - SCENARIO 2



RESULTS

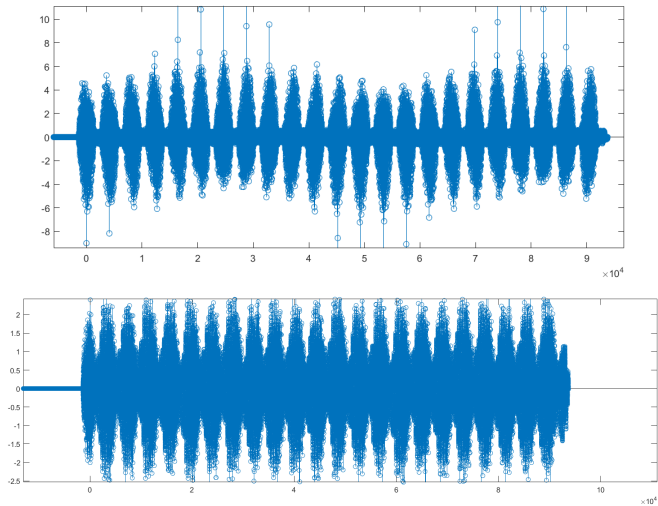
REAL WORLD - TRX SIGNAL - SCENARIO 2



Results Intepretation

RESULTS

INTERPRETATION



Results

Tried but did not worked

MORE

TRIED BUT NOT WORKED

- ▶ Data coding (only TX part implemented but not used)
- ▶ Use preamble for channel estimation (not used in real world scenario)
- ▶ Get the real channel transfer function
- ▶ Use MMSE for channel estimation

1. Introduction
2. Transmitter
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4. Receiver
5. Results
6. Conclusions and Discussion

REPOSITORY LINK

`https:
//github.com/vladBaciu/SW_2020_Matlab_OFDM_simulation`