

# EE910: Digital Communication Systems-I

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## Lecture #7B: Error probability of orthogonal signaling with noncoherent detection



## Error probability of orthogonal signaling with noncoherent detection

- Let us assume  $M$  equiprobable, equal-energy, carrier modulated orthogonal signals are transmitted over an AWGN channel.
- These signals are noncoherently demodulated at the receiver and then optimally detected.
- The lowpass equivalent of the signals can be written as  $M$   $N$ -dimensional vectors ( $N = M$ )

$$\begin{aligned}s_1 &= (\sqrt{2\mathcal{E}_s}, 0, 0, \dots, 0) \\s_2 &= (0, \sqrt{2\mathcal{E}_s}, 0, \dots, 0) \\&\vdots \\s_N &= (0, 0, \dots, 0, \sqrt{2\mathcal{E}_s})\end{aligned}\tag{1}$$

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## Error probability of orthogonal signaling with noncoherent detection

- Without loss of generality we can assume that  $s_{1I}$  is transmitted.
- Therefore the received vector will be

$$\mathbf{r}_I = e^{j\phi} \mathbf{s}_{1I} + \mathbf{n}_I\tag{2}$$

where  $\mathbf{n}_I$  is a complex circular zero-mean Gaussian random vector with variance of each complex component equal to  $2N_0$ .

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## Error probability of orthogonal signaling with noncoherent detection

- The optimal receiver computes and compares  $|\mathbf{r}_l \cdot \mathbf{s}_{ml}|$ , for all  $1 \leq m \leq M$ . This results in

$$\begin{aligned} |\mathbf{r}_l \cdot \mathbf{s}_{1l}| &= |2\mathcal{E}_s e^{j\phi} + \mathbf{n}_l \cdot \mathbf{s}_{1l}| \\ |\mathbf{r}_l \cdot \mathbf{s}_{ml}| &= |\mathbf{n}_l \cdot \mathbf{s}_{ml}| \quad 2 \leq m \leq M \end{aligned} \quad (3)$$

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## Error probability of orthogonal signaling with noncoherent detection

- For  $1 \leq m \leq M$ ,  $\mathbf{n}_l \cdot \mathbf{s}_{ml}$  is a circular zero-mean complex Gaussian random variable with variance  $4\mathcal{E}_s N_0$  ( $2\mathcal{E}_s N_0$  per real and imaginary parts).
- From equation (3) we have

$$\begin{aligned} \text{Re}[\mathbf{r}_l \cdot \mathbf{s}_{1l}] &\sim N(2\mathcal{E}_s \cos\phi, 2\mathcal{E}_s N_0) \\ \text{Im}[\mathbf{r}_l \cdot \mathbf{s}_{1l}] &\sim N(2\mathcal{E}_s \sin\phi, 2\mathcal{E}_s N_0) \\ \text{Re}[\mathbf{r}_l \cdot \mathbf{s}_{ml}] &\sim N(0, 2\mathcal{E}_s N_0) \quad 2 \leq m \leq M \\ \text{Im}[\mathbf{r}_l \cdot \mathbf{s}_{ml}] &\sim N(0, 2\mathcal{E}_s N_0) \quad 2 \leq m \leq M \end{aligned} \quad (4)$$

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## Error probability of orthogonal signaling with noncoherent detection

- From the definition of Rayleigh and Ricean random variables, we conclude that random variables  $R_m$ ,  $1 \leq m \leq M$  defined as

$$R_m = |\mathbf{r}_l \cdot \mathbf{s}_{ml}| \quad 1 \leq m \leq M \quad (5)$$

are independent random variables.

- $R_1$  has a Ricean distribution with parameters  $s = 2\mathcal{E}_s$  and  $\sigma^2 = 2\mathcal{E}_s N_0$ , and
- $R_m, 2 \leq m \leq M$ , are Rayleigh random variables with parameter  $\sigma^2 = 2\mathcal{E}_s N_0$ .



## Error probability of orthogonal signaling with noncoherent detection

- In other words,

$$p_{R_1}(r_1) = \begin{cases} \frac{r_1}{\sigma^2} I_0\left(\frac{sr_1}{\sigma^2}\right) e^{-\frac{r_1^2 + s^2}{2\sigma^2}}, & r_1 > 0 \\ 0, & \text{otherwise} \end{cases} \quad (6)$$

and

$$p_{R_m}(r_m) = \begin{cases} \frac{r_m}{\sigma^2} e^{-\frac{r_m^2}{2\sigma^2}}, & r_m > 0 \\ 0, & \text{otherwise} \end{cases} \quad (7)$$

for  $2 \leq m \leq M$ .



## Error probability of orthogonal signaling with noncoherent detection

- Since by assumption  $\mathbf{s}_{1I}$  is transmitted, a correct decision is made at the receiver if  $R_1 > R_m$  for  $2 \leq m \leq M$ .
- Although random variables  $R_m$  for  $1 \leq m \leq M$  are statistically independent, the events  $R_1 > R_2, R_1 > R_3, \dots, R_1 > R_M$  are not independent due to the existence of the common  $R_1$
- To make them independent, we need to condition on  $R_1 = r_1$  and then average over all values of  $r_1$ .

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## Error probability of orthogonal signaling with noncoherent detection

- Therefore

$$\begin{aligned}
 P_c &= P[R_2 < R_1, R_3 < R_1, \dots, R_M < R_1] \\
 &= \int_0^\infty P[R_2 < r_1, R_3 < r_1, \dots, R_M < r_1 | R_1 = r_1] p_{R_1}(r_1) dr_1 \quad (8) \\
 &= \int_0^\infty (P[R_2 < r_1])^{M-1} p_{R_1}(r_1) dr_1
 \end{aligned}$$

- But

$$\begin{aligned}
 P[R_2 < r_1] &= \int_0^{r_1} p_{R_2}(r_2) dr_2 \\
 &= 1 - e^{-\frac{r_1^2}{2\sigma^2}} \quad (9)
 \end{aligned}$$

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## Error probability of orthogonal signaling with noncoherent detection

- Using the binomial expansion, we have

$$\left(1 - e^{-\frac{r_1^2}{2\sigma^2}}\right)^{M-1} = \sum_{n=0}^{M-1} (-1)^n \binom{M-1}{n} e^{-\frac{nr_1^2}{2\sigma^2}} \quad (10)$$

- Substituting into equation (8), we get

$$\begin{aligned} P_c &= \sum_{n=0}^{M-1} (-1)^n \binom{M-1}{n} \int_0^\infty e^{-\frac{nr_1^2}{2\sigma^2}} \frac{r_1}{\sigma^2} I_0\left(\frac{sr_1}{\sigma^2}\right) e^{-\frac{r_1^2+s^2}{2\sigma^2}} dr_1 \\ &= \sum_{n=0}^{M-1} (-1)^n \binom{M-1}{n} \int_0^\infty \frac{r_1}{\sigma^2} I_0\left(\frac{sr_1}{\sigma^2}\right) e^{-\frac{(n+1)r_1^2+s^2}{2\sigma^2}} dr_1 \\ &= \sum_{n=0}^{M-1} (-1)^n \binom{M-1}{n} e^{-\frac{ns^2}{2(n+1)\sigma^2}} \int_0^\infty \frac{r_1}{\sigma^2} I_0\left(\frac{sr_1}{\sigma^2}\right) e^{-\frac{(n+1)r_1^2+\frac{s^2}{n+1}}{2\sigma^2}} dr_1 \end{aligned} \quad (11)$$

## Error probability of orthogonal signaling with noncoherent detection

- By introducing a change of variables

$$s' = \frac{s}{\sqrt{n+1}} \quad (12)$$

$$r' = r_1 \sqrt{n+1} \quad (13)$$

the integral in equation (11) becomes

$$\begin{aligned} \int_0^\infty \frac{r_1}{\sigma^2} I_0\left(\frac{sr_1}{\sigma^2}\right) e^{-\frac{(n+1)r_1^2+\frac{s^2}{n+1}}{2\sigma^2}} dr_1 &= \frac{1}{n+1} \int_0^\infty \frac{r'}{\sigma^2} I_0\left(\frac{r's'}{\sigma^2}\right) e^{-\frac{r'^2+s'^2}{2\sigma^2}} dr' \\ &= \frac{1}{n+1} \end{aligned} \quad (14)$$

where we have used the fact that the area under a Ricean pdf is equal to 1.

## Error probability of orthogonal signaling with noncoherent detection

- Substituting Equation (14) into Equation (11) and noting that  $\frac{s^2}{2\sigma^2} = \frac{4\mathcal{E}_s^2}{4\mathcal{E}_s N_0} = \frac{\mathcal{E}_s}{N_0}$  we obtain

$$P_c = \sum_{n=0}^{M-1} \frac{(-1)^n}{n+1} \binom{M-1}{n} e^{-\frac{n}{n+1} \frac{\mathcal{E}_s}{N_0}} \quad (15)$$

- Then the probability of a symbol error becomes

$$P_e = \sum_{n=0}^{M-1} \frac{(-1)^{n+1}}{n+1} \binom{M-1}{n} e^{-\frac{n \log_2 M}{n+1} \frac{\mathcal{E}_b}{N_0}} \quad (16)$$

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## Error probability of orthogonal signaling with noncoherent detection

- For binary orthogonal signalling, including binary orthogonal FSK with noncoherent detection, Equation (16) simplifies to

$$P_b = \frac{1}{2} e^{-\frac{\mathcal{E}_b}{2N_0}} \quad (17)$$

- Comparing this result with coherent detection of binary orthogonal signals for which the error probability is given by

$$P_b = Q\left(\sqrt{\frac{\mathcal{E}_b}{N_0}}\right) \quad (18)$$

- Using the inequality  $Q(x) \leq \frac{1}{2} e^{-x^2/2}$ , we conclude that  $P_b(\text{noncoherent}) \geq P_b(\text{coherent})$ , as expected.

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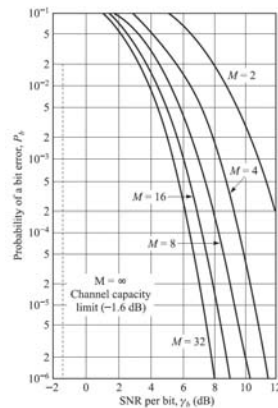
## Error probability of orthogonal signaling with noncoherent detection

- For error probabilities less than  $10^{-4}$ , the difference between the performance of coherent and noncoherent detection of binary orthogonal is less than 0.8 dB
- For  $M > 2$ , we may compute the probability of a bit error by making use of the relationship

$$P_b = \frac{2^{k-1}}{2^k - 1} P_e \quad (19)$$

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## Error probability of orthogonal signaling with noncoherent detection



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