

## EE910: Digital Communication Systems-I

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## Lecture #2F: A brief introduction to random processes



## Random processes

- The mean and the autocorrelation of a random  $X(t)$  are defined as

$$m_x(t) = E[X(t)] \quad (1)$$

$$R_x(t_1, t_2) = E[X(t_1)X^*(t_2)] \quad (2)$$

- The cross-correlation of two random process  $X(t)$  and  $Y(t)$  is defined as

$$R_{XY}(t_1, t_2) = E[X(t_1)Y^*(t_2)] \quad (3)$$

- We have  $R_X(t_2, t_1) = R_X^*(t_1, t_2)$ , i.e.,  $R_X(t_1, t_2)$  is Hermitian.
- For cross-correlation  $R_{XY}(t_1, t_2) = R_{YX}^*(t_1, t_2)$

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## Wide-Sense Stationary Random Processes

- Random process  $X(t)$  is wide-sense stationary (WSS) if its mean is constant and autocorrelation is a function of time difference i.e

$$R_X(t_1, t_2) = R_X(\tau) \quad (4)$$

where  $\tau = t_1 - t_2$

- Two processes  $X(t)$  and  $Y(t)$  are jointly WSS if both  $X(t)$  and  $Y(t)$  are WSS and  $R_{XY}(t_1, t_2) = R_{XY}(\tau)$
- A complex process is WSS if its real and imaginary parts are WSS

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## WSS Processes

- For a WSS process, the power spectrum is the Fourier transform of the autocorrelation function  $R_X(\tau)$

$$S_X(f) = \mathfrak{F}[R_X(\tau)] \quad (5)$$

- Cross spectral density of two jointly WSS processes is defined as

$$S_{XY}(f) = \mathfrak{F}[R_{XY}(\tau)] \quad (6)$$

- The CSD satisfies the following property:

$$S_{XY}(f) = S_{X^*Y}^*(f) \quad (7)$$

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## WSS Processes

- If  $X(t)$  and  $Y(t)$  are jointly WSS random processes, then  $Z(t) = aX(t) + bY(t)$  is a WSS process with

$$R_Z(\tau) = |a|^2 R_X(\tau) + |b|^2 R_Y(\tau) + ab^* R_{XY}(\tau) + ba^* R_{YX}(\tau) \quad (8)$$

$$S_Z(f) = |a|^2 S_X(f) + |b|^2 S_Y(f) + 2\text{Re}[ab^* S_{XY}(f)] \quad (9)$$

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## WSS Processes

- When a WSS process  $X(t)$  passes through an LTI system with impulse response  $h(t)$  and transfer function  $H(f) = \mathcal{F}[h(t)]$ , the output processes  $Y(t)$  and  $X(t)$  are jointly WSS and the following relations hold

$$m_Y = m_X \int_{-\infty}^{\infty} h(t) dt \quad (10)$$

$$R_{XY}(\tau) = R_X(t) * h^*(-\tau) \quad (11)$$

$$R_Y(\tau) = R_X(t) * h(\tau) * h^*(-\tau) \quad (12)$$

$$m_Y = m_X H(0) \quad (13)$$

$$S_{XY}(f) = S_X H^*(f) \quad (14)$$

$$S_Y(f) = S_X(f) |H(f)|^2 \quad (15)$$

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## WSS Processes

- The power in a WSS process  $X(t)$  is the sum of the power at all frequencies

$$P_X = E[|X(t)|^2] = R_X(0) = \int_{-\infty}^{\infty} S_X(f) df \quad (16)$$

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## Gaussian Random Processes

- A real random process  $X(t)$  is Gaussian if for all positive integers  $n$  and for all  $(t_1, t_2, \dots, t_n)$ , the random vector  $(X(t_1), X(t_2), \dots, X(t_n))^t$  is Gaussian random vector.
- Two real random processes  $X(t)$  and  $Y(t)$  are jointly Gaussian if for all positive integers  $n, m$  and all  $(t_1, t_2, \dots, t_n)$ , and  $(t'_1, t'_2, \dots, t'_m)$  the random vector

$$(X(t_1), X(t_2), \dots, X(t_n), Y(t'_1), Y(t'_2), \dots, Y(t'_m))^t \quad (17)$$

is a Gaussian vector.

## White Processes

- A process is called a white process if its power spectral density is constant for all frequencies

$$S_X(f) = \frac{N_o}{2} \quad (18)$$

- Power in a white process is infinite, indicating that white process cannot exist as a physical process

# Discrete-Time Random Processes

- PSD of a WSS discrete time random process is defined as the discrete-time Fourier transform of its autocorrelation function

$$S_X(f) = \sum_{m=-\infty}^{\infty} R_X(m) \exp^{-j2\pi fm} \quad (19)$$

- The power in discrete time random process is given by

$$P = E[|X(n)|^2] = R_X(0) = \int_{-\frac{1}{2}}^{\frac{1}{2}} S_X(f) df \quad (20)$$

## Cyclostationary Random Processes

- A random process  $X(t)$  is cyclostationary if its mean and autocorrelation are periodic functions with same period  $T_o$

$$m_X(t + T_o) = m_X(t) \quad (21)$$

$$R_X(t_1 + T_o, t_2 + T_o) = R_X(t_1, t_2) \quad (22)$$

- Average autocorrelation function defined as

$$\overline{R_X}(t) = \frac{1}{T_o} \int_0^{T_o} R_X(t + \tau, t) \quad (23)$$

## Proper and Circular Random Processes

- For a Complex Random process  $Z(t) = X(t) + jY(t)$ , the covariance and the pseudo covariance are

$$C_Z(t + \tau, t) = E[Z(t + \tau)Z^*(t)] \quad (24)$$

$$\tilde{C}_Z(t + \tau, t) = E[Z(t + \tau)Z(t)] \quad (25)$$

which can be written as

$$C_Z(t + \tau, t) = C_X(t + \tau, t) + C_Y(t + \tau, t) + j(C_{XY}(t + \tau, t) - C_YX(t + \tau, t)) \quad (26)$$

$$\tilde{C}_Z(t + \tau, t) = C_X(t + \tau, t) - C_Y(t + \tau, t) + j(C_{XY}(t + \tau, t) + C_YX(t + \tau, t)) \quad (27)$$

- A complex random process  $Z(t)$  is proper if its pseudocovariance is zero

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## Bandpass and Lowpass Random Processes

- For a bandpass process the power spectral density is located around frequencies  $\pm f_o$  and for lowpass processes the PSD is located around zero frequency
- The in-phase and quadrature components of a bandpass random process  $X(t)$  is

$$X_i(t) = X(t) \cos 2\pi f_o t + \hat{X}(t) \sin 2\pi f_o t \quad (28)$$

$$X_q(t) = \hat{X}(t) \cos 2\pi f_o t - X(t) \sin 2\pi f_o t \quad (29)$$

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## Bandpass and Lowpass Random Processes

- $X_i(t)$  and  $X_q(t)$  are jointly WSS zero-mean random process
- $X_i(t)$  and  $X_q(t)$  have the same power spectral density
- $X_i(t)$  and  $X_q(t)$  are both lowpass process
- We define lowpass equivalent process  $X_l(t)$  as

$$X_l(t) = X_i(t) + jX_q(t) \quad (30)$$

## Bandpass and Lowpass Random Processes

- Since  $X(t)$  by assumption is zero mean so is  $\hat{X}(t)$ , its Hilbert transform. Therefore  $X_i(t)$  and  $X_q(t)$  are both zero-mean process
- Autocorrelation of  $X_i(t)$  is

$$\begin{aligned} R_{X_i}(t + \tau, t) &= E[(X_i(t + \tau)X_i(t))] \\ &= E[X(t + \tau) \cos 2\pi f_o(t + \tau) + \hat{X}(t + \tau) \sin 2\pi f_o(t + \tau)] \\ &\quad \times [X(t) \cos 2\pi f_o t + \hat{X}(t) \sin 2\pi f_o t] \end{aligned} \quad (31)$$

on solving it we get

$$R_{X_i}(\tau) = R_x(\tau) \cos(2\pi f_o \tau) + \hat{R}_X(\tau) \sin(2\pi f_o \tau) \quad (32)$$



## Bandpass and Lowpass Random Processes

- Similarly we have

$$R_{X_q}(\tau) = R_{X_i}(\tau) = R_x(\tau) \cos(2\pi f_o \tau) + \hat{R}_x(\tau) \sin(2\pi f_o \tau) \quad (33)$$

$$R_{X_i X_q}(\tau) = -R_{X_q X_i}(\tau) = R_x(\tau) \sin(2\pi f_o \tau) - \hat{R}_x(\tau) \cos(2\pi f_o \tau) \quad (34)$$

- Power Spectral densities would be

$$S_{X_i} = S_{X_q} = \begin{cases} S_X(f + f_o) + S_X(f - f_o) & |f| < f_o \\ 0 & \text{otherwise} \end{cases} \quad (35)$$

$$S_{X_i X_q} = -S_{X_i X_q} = \begin{cases} j [S_X(f + f_o) - S_X(f - f_o)] & |f| < f_o \\ 0 & \text{otherwise} \end{cases} \quad (36)$$

## Bandpass and Lowpass Random Processes

- The complex process  $X_I(t) = X_i(t) + jX_q(t)$  as the lowpass equivalent of  $X(t)$ . Since  $X_i(t)$  and  $X_q(t)$  are both lowpass processes, we conclude that  $X_I(t)$  is also a lowpass process

$$\begin{aligned} R_{X_I}(\tau) &= 2R_{X_i}(\tau) + 2jR_{X_q X_i}(\tau) \\ &= 2[R_X(\tau) + j\hat{R}_X(\tau)]e^{-j2\pi f_o t} \end{aligned} \quad (37)$$

- Taking Fourier transform on both sides we have

$$S_{X_I} = \begin{cases} 4S_X(f + f_o) & |f| < f_o \\ 0 & \text{otherwise} \end{cases} \quad (38)$$

$$S_X(f) = \frac{1}{4}[S_{X_l}(f - f_o) + S_{X_l}(f + f_o)] \quad (39)$$