

09/01/2020

Signals And Systems:-

- > A Signal is any physical phenomena which conveys information.

Example:-

- > position (Displacement/distance) $x(t)$
- > Velocity $v(t)$
- > Acceleration $a(t)$
- > Force
- > Voltage $V(t)$
- > Temperature of Ocean as a function of its depth $f(x)$

Note:- From the above example we note that signals can be mathematically modeled using functions

Ex- $f(x)$
 dependent \longleftrightarrow independent

- > An image is a multi-dimensional signal with two independent variables characterized as position. $f(x, y)$
- > A video signal is a function of 3 independent variables. $f(x, y, t)$

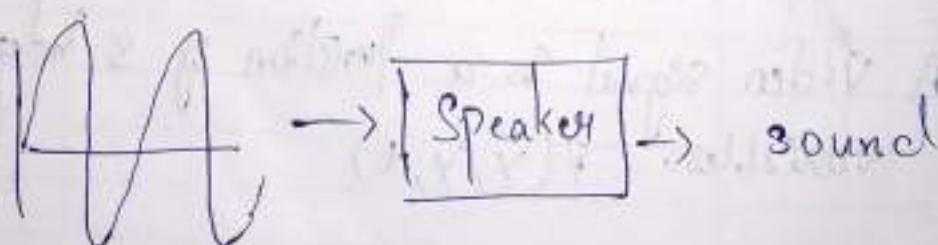
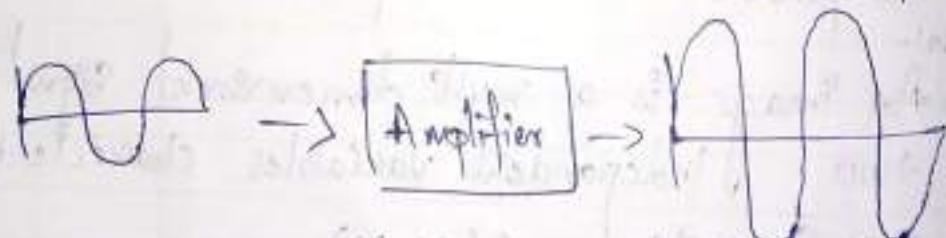
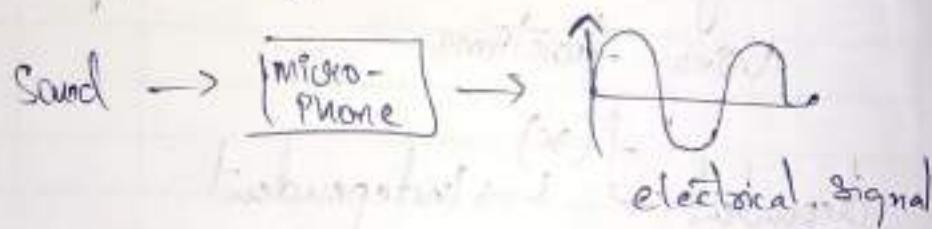
* Signal:-

> A Signal is a function of one or more variables that conveys information about some (real physical) phenomena.

* System:-

> A System is an entity that processes signal in order to produce a more useful signal, may be of the same form or the other form.

i) Eg:- Amplifier system.



2) Integrator, differentiator, clipper, modulator, delay, filter.

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* Classification of Signals :-

→ Signals can be broadly classified into two types.

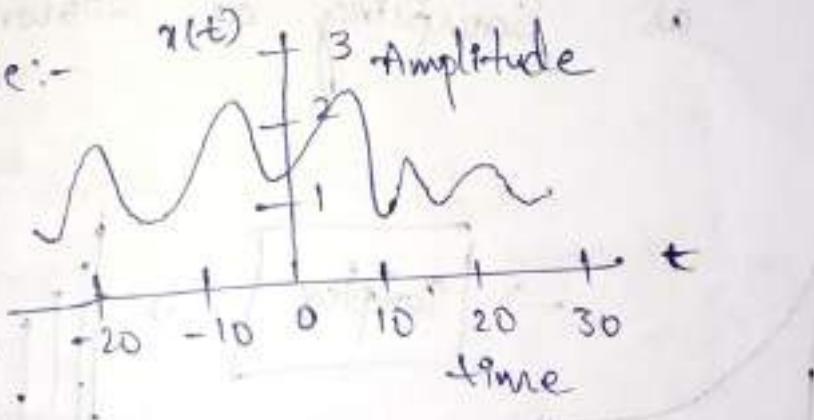
i) Continuous time signal (CTS)

ii) discrete time signal (DTS)

> CTS:- A signal that is defined for every real value of time is called C.T.S.

Here time is denoted as 't'

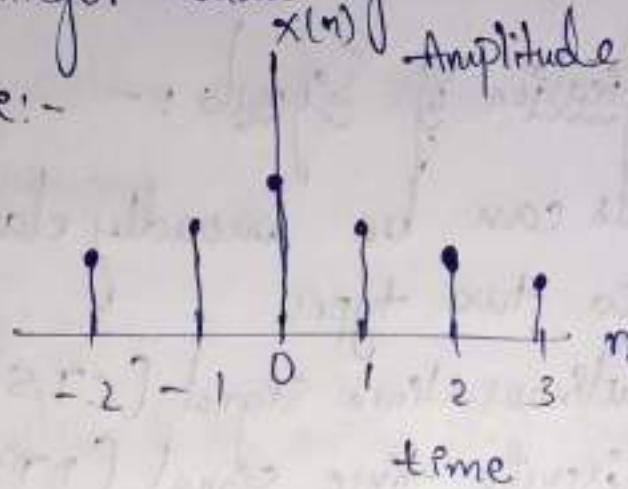
Example:- $x(t)$



Practical Example includes camera, video, voice, voltage waveform.

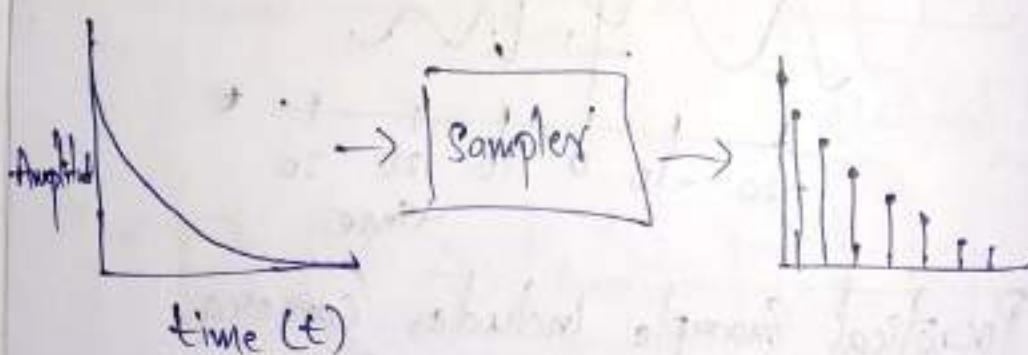
> DTS:- It is a signal that is defined only for discrete instance of time [integer values of time]

Example:-



Note:- time is denoted by 'n'.

Note:- A DIS can be generated by a CTS Using an operation called as Sampling, as shown below



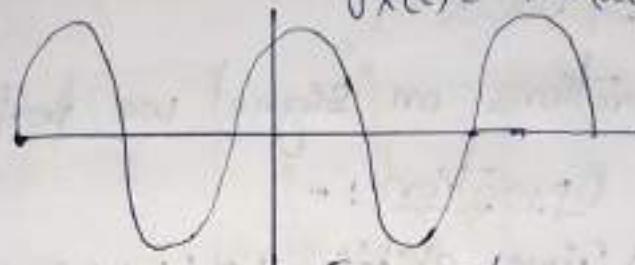
Sampler is a circuit that performs Sampling.

* Signal representation :-

1) CT S :-

- > It can either be represented graphically or mathematically.

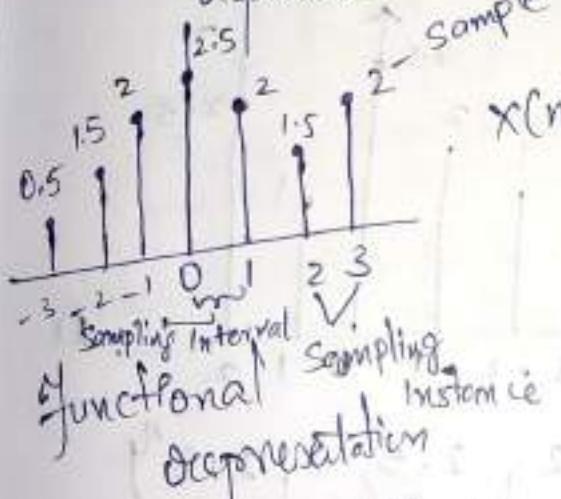
Eg:-



$$T_0 = 2\pi/\omega_0$$

2) DT S :-

Graphical representation



Sequence representation

$$x(n) = \{0.5, 1, 1.5, 2, 2.5, 2, 1.5, 1\}$$

Tabular representation

n	x(n)
-3	0.5
-2	1.5
-1	2
0	2.5
1	2
2	1.5
3	2

$$x(n) = \begin{cases} 0.5, & n = -3 \\ 1.5, & n = -2 \\ 2, & n = -1 \\ 2.5, & n = 0 \\ 2, & n = 1 \\ 1.5, & n = 2 \\ 2, & n = 3 \end{cases}$$

Note :- The first sample of a sequence representation will correspond to origin by default, if an arrow mark is not given

* Signal Operations :-

→ In operations on signal we perform
Time Operation :-

- Time shifting; $x(t \pm t_0)$, $x(n \pm n_0)$
- Time scaling; $x(at)$, $x(t/a)$, $x(an)$, $x(n/b)$
- Time reversal / folding; $x(-t)$, $x(-n)$

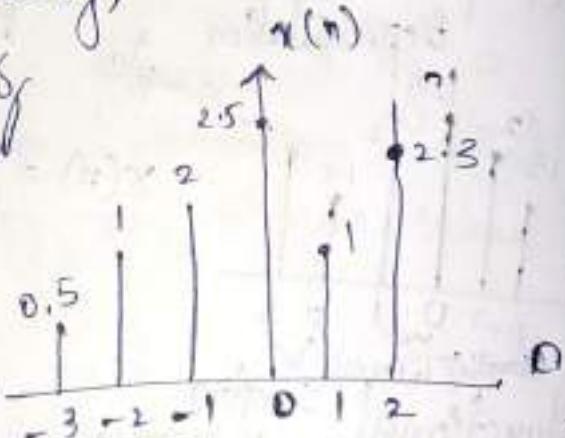
Eg:- Time shifting

Plot (i) $x(n-2)$

(ii) $x(n+2)$

(iii) Sequence

representation

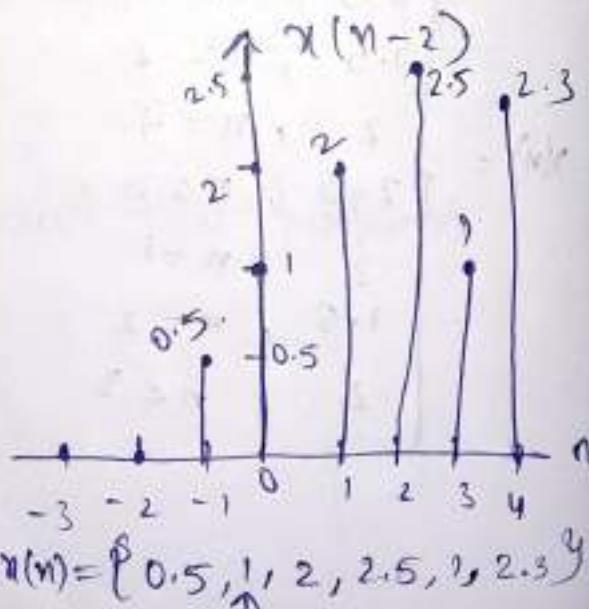


iii) $X(n) = \{ 0.5, 1, 2, 2.5, 1, 2.3 \}$

(i)

n	$x(n-2)$
-3	0
-2	0
-1	0.5
0	1
1	2
2	2.5
3	1
4	2.3

$x(n-2)$

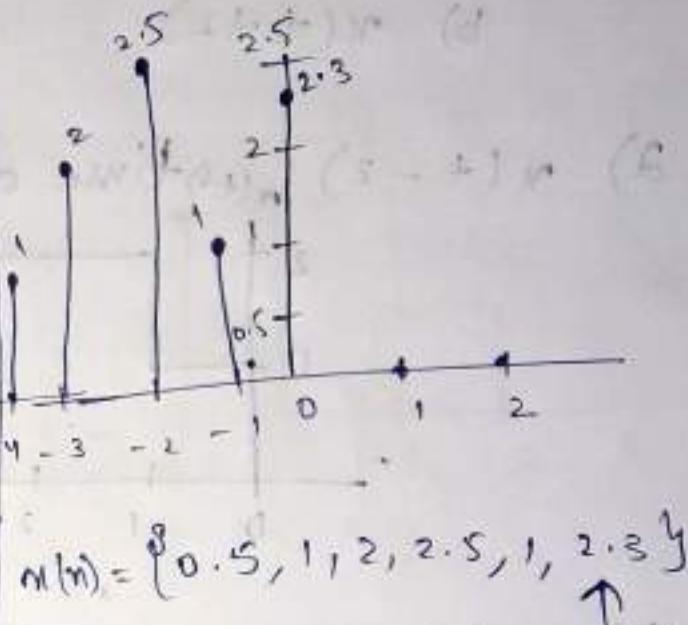


$x(n) = \{ 0.5, 1, 2, 2.5, 1, 2.3 \}$

Note:- $x(n - n_0)$ indicates the signal should be shifted by n_0 seconds to the right. This is called as time delay operation.

(ii) $x(n+2)$

n	$x(n+2)$
-3	2
-2	2.5
-1	1.05
0	2.3
1	0
2	0
-4	1
-5	0.5
-6	0



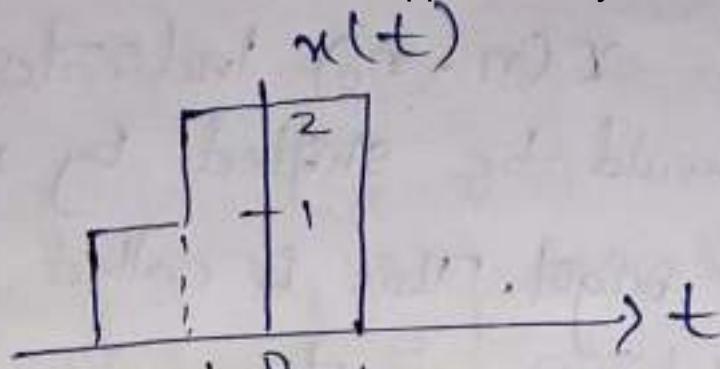
$$x(n) = \{0.5, 1, 2, 2.5, 1, 2.3\}$$

Note:- $x(n+n_0)$ indicates the signal is shifted towards left.

$x(n+n_0)$ is called shifts the signal to the left by n_0 units and is called as time advance operation.

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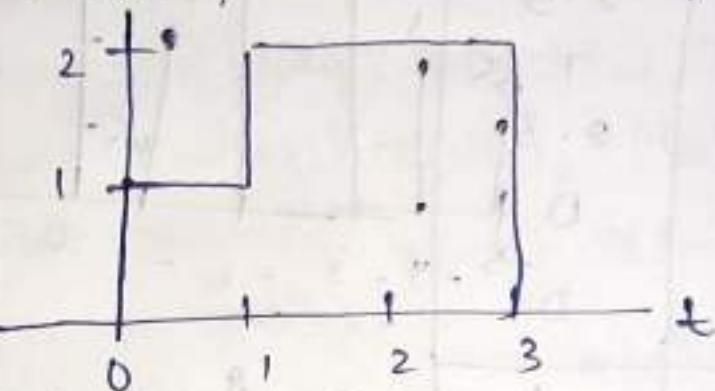
Problem)



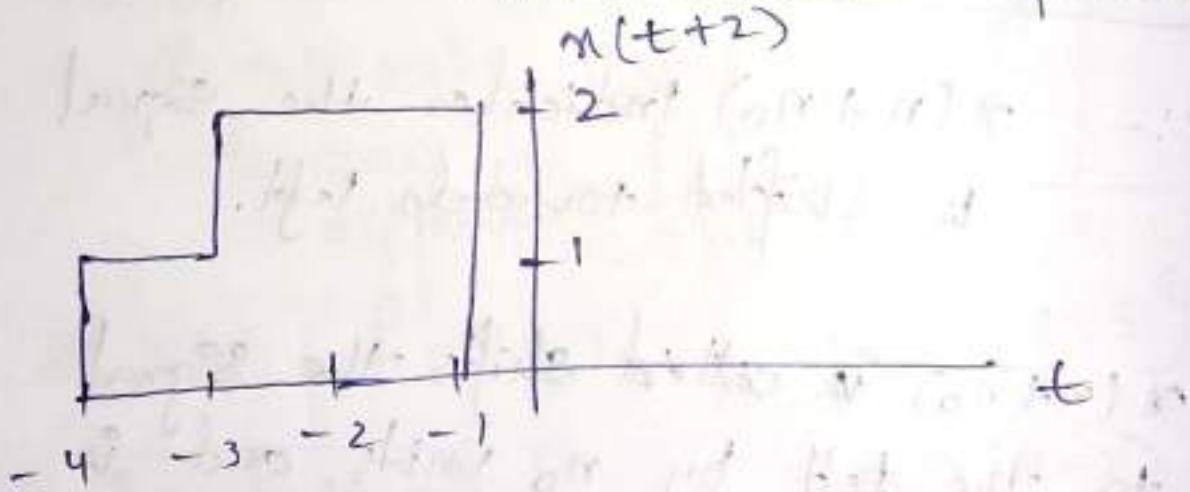
$$\begin{aligned} 1) & g(t) \leftarrow \begin{cases} (+2) & (-2) \\ (-2) & (+2) \end{cases} \xrightarrow{\text{digit}} \\ 2) & x(t-2) \end{aligned}$$

b) $x(t+2)$

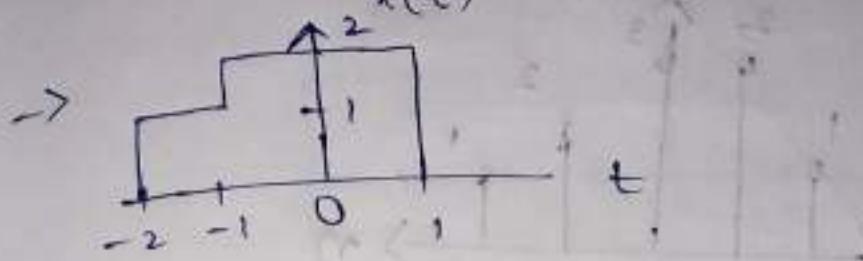
a) $x(t-2)$ $\xrightarrow{x(t+2)}$ time delay operation



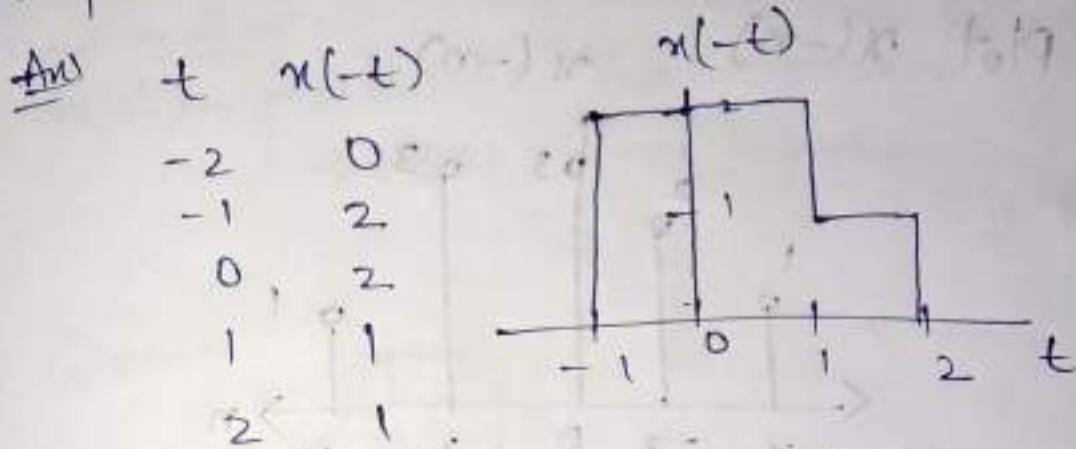
b) $x(t+2)$ time advance operation



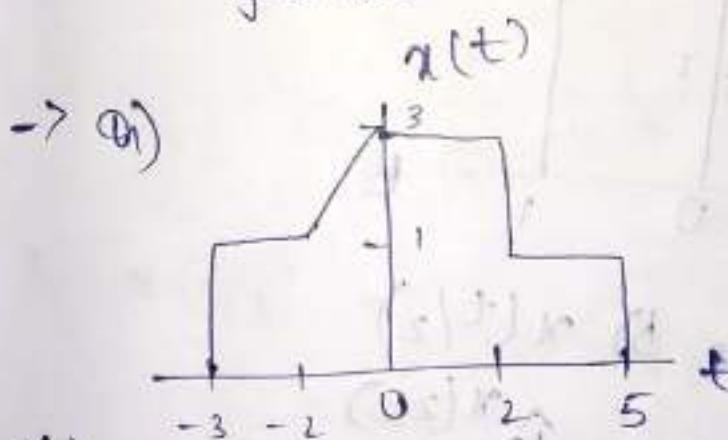
* 2) Time reversal folding :-



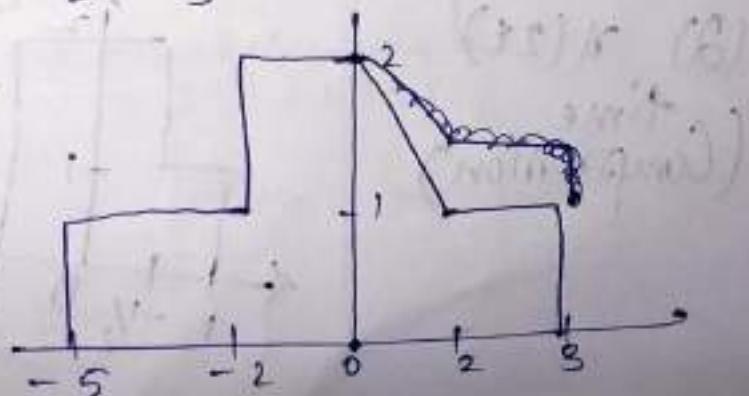
→ plot $x(-t)$



Note:- In time reversal operation, the signals mirror image is taken w.r.t. y-axis

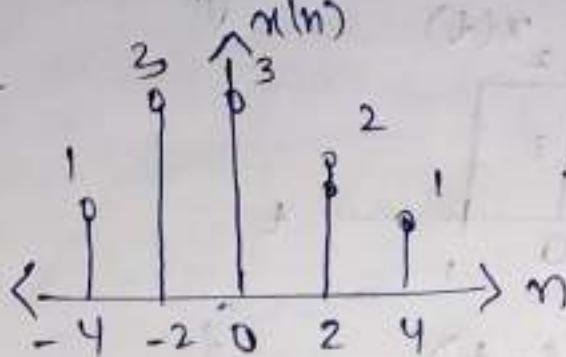


Plot $x(-t)$ →

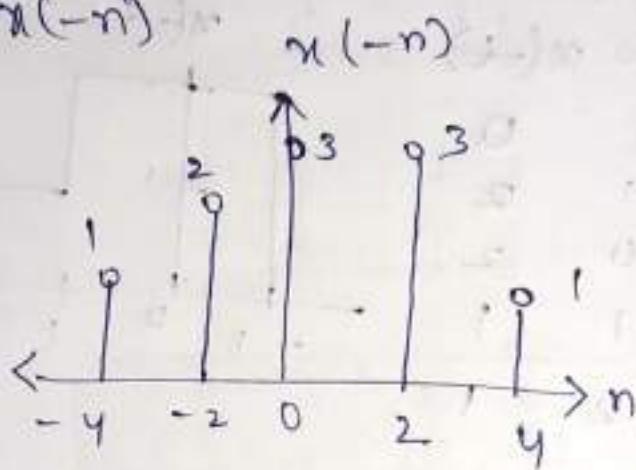


Q) discrete time problem:

Given:-

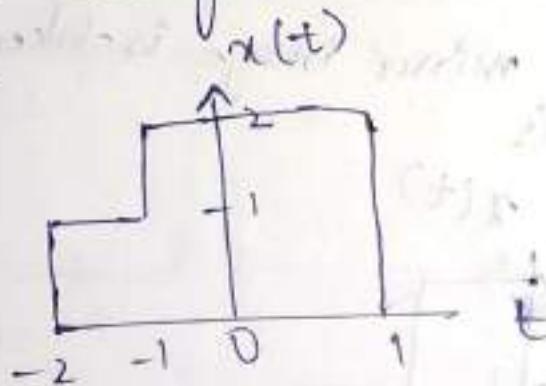


Plot $x(-n)$



3) Time Scaling: - $x(at)$, $x(t/a)$

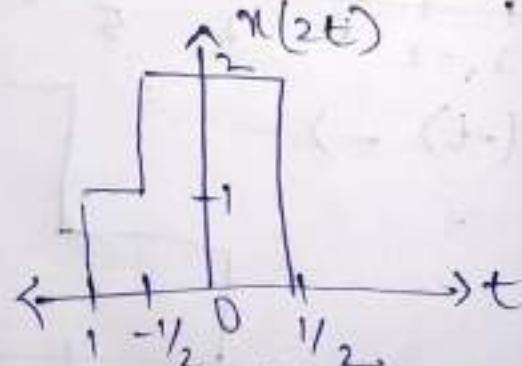
Given:-

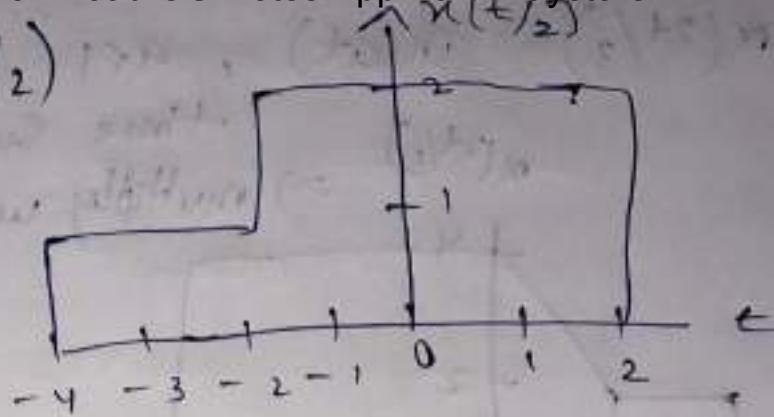


Plot a) $x(2t)$, b) $x(t/2)$

(a) $x(2t)$

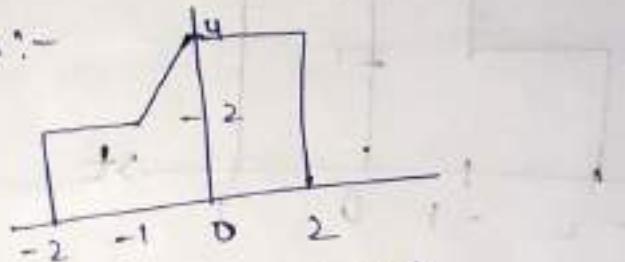
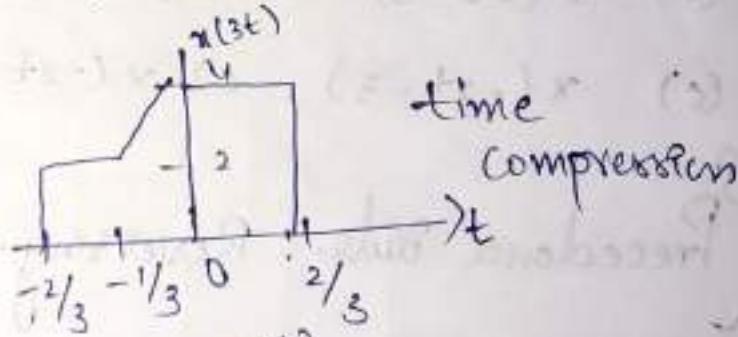
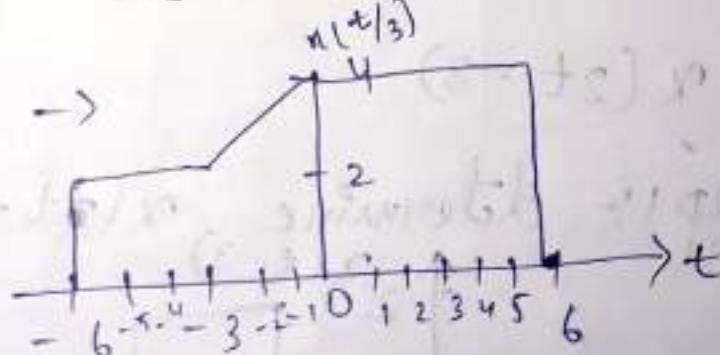
(time compression)



(b) $x(t/2)$ 

(time expansion)
 \rightarrow time scaling:-
 $x(at)$ $\rightarrow |a| > 1 \rightarrow$ time compression
 $\rightarrow |a| < 1 \rightarrow$ time expansion

(a) Given:-

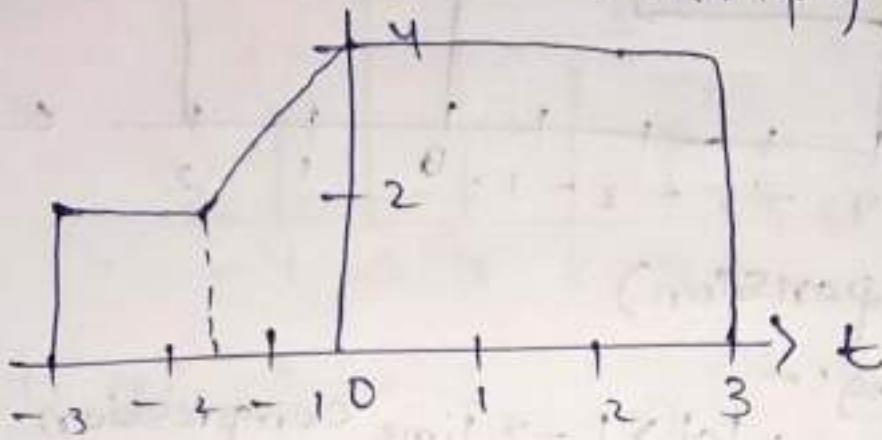
a) $x(3t)$ b) $x(t/3)$ (a) $x(3t) \rightarrow$ (b) $x(t/3) \rightarrow$ 

time expansion

$$(c) \alpha(2t/3)$$

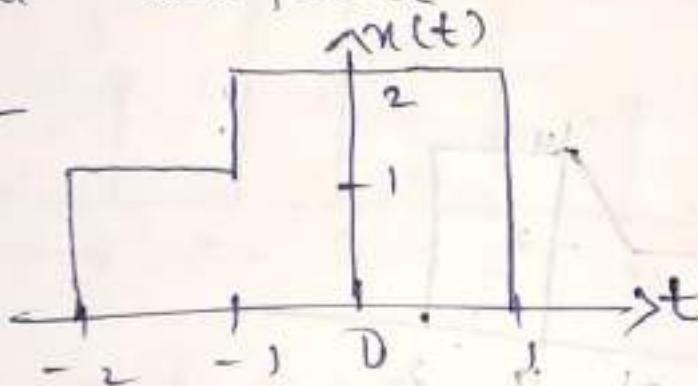
$m(at)$, $a < 1$

$\alpha(2t/3) \Rightarrow$ time expansion
multiply with $3/2$



* multiple transformation:—

Given: —



(a) $\alpha(2t-3)$ b) $\alpha(2t+3)$

(c) $\alpha(-2t-3)$ d) $\alpha(-2t+3)$

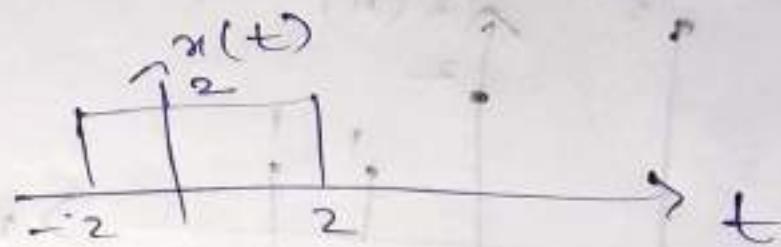
Precedence rule: Reversing \rightarrow shifting \rightarrow scaling

Shifting \rightarrow Scaling \rightarrow reversing

(d) $\alpha(2t-3)$

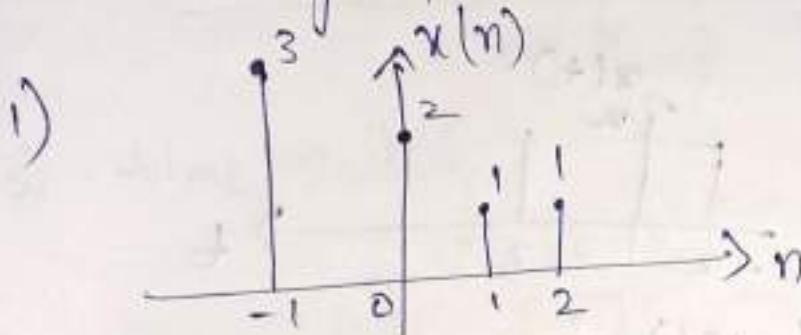
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i) Given:-



② $x(-t/3 + 3/2)$

Note:- The precedence rule ('blood cut') method is not valid for signals discrete time signals involving scaling operation

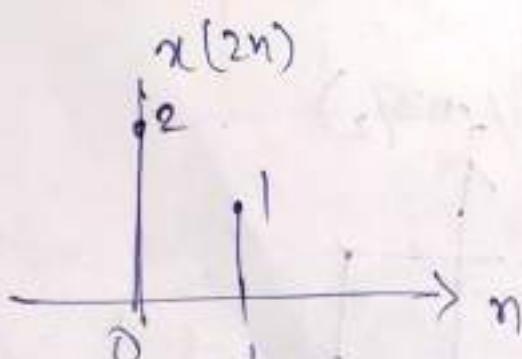


- a) $x(2n)$
- b) $x(n/2)$
- c) $x(\frac{n}{2} - 1)$
- d) $x(-\frac{n}{2} - \frac{1}{2})$
- e) $x(-n-1)$

a) $x(2n)$

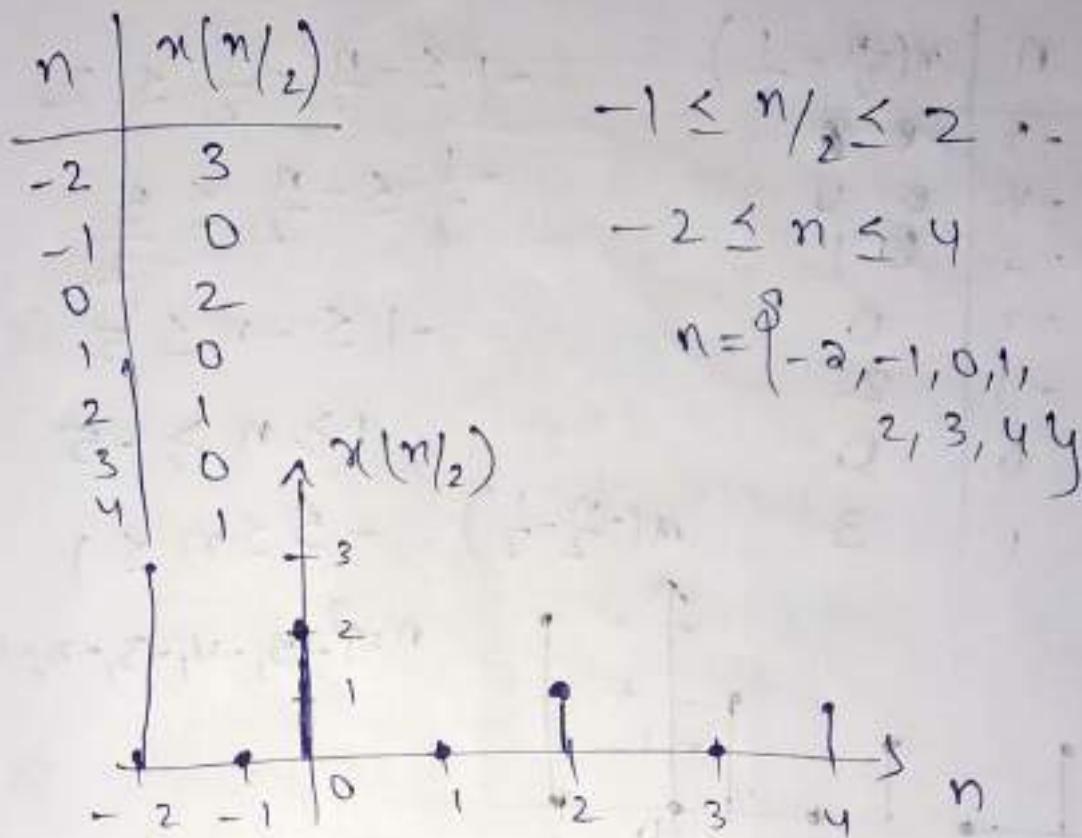
n	$x(2n)$	$-1 \leq 2n \leq 2$
0	2	$\therefore -\frac{1}{2} \leq n \leq 1$
1	1	

$n = \{0, 1\}$ \because Integer

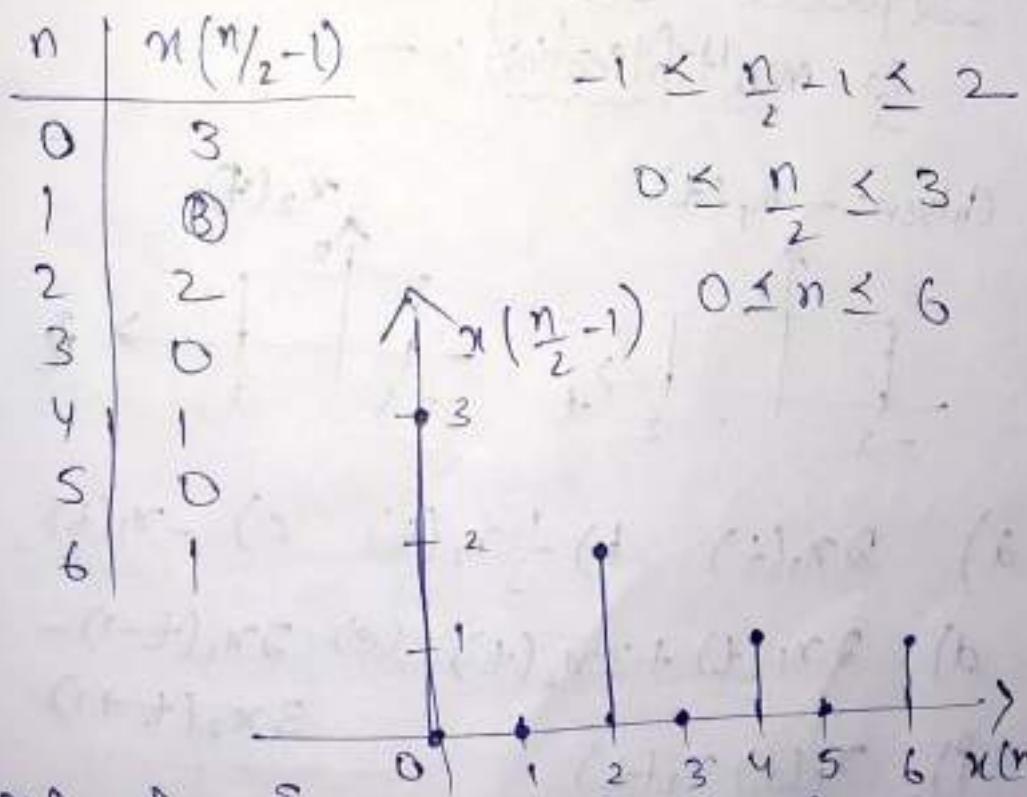


Sequence representation

$$x(2n) = \begin{cases} 2, & n=0 \\ 1, & n=1 \end{cases}$$

b) $x(n/2)$.

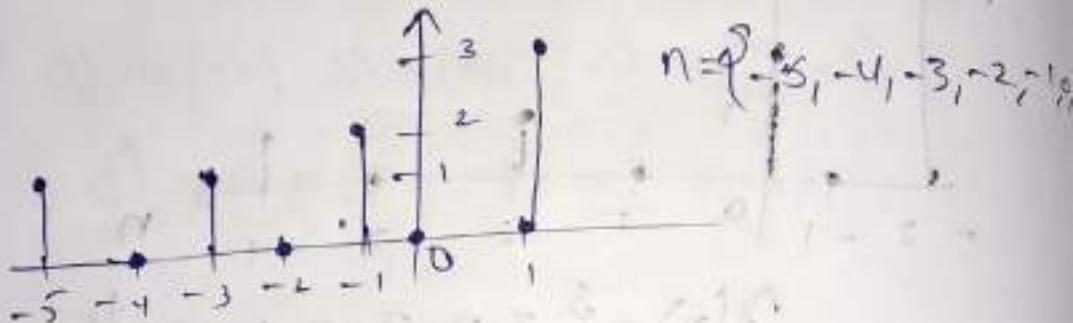
$$x(n) = \{3, 0, 2, 0, 1, 0, 1\}$$

(c) $x(\frac{n}{2}-1)$ 

$$x(\frac{n}{2}-1) = \{3, 0, 2, 0, 1, 0, 1\}$$

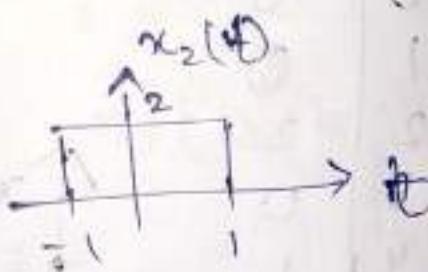
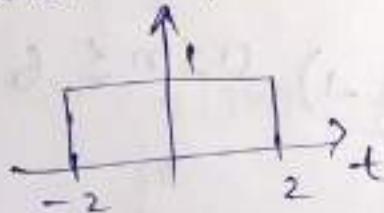
$$d) \mathbf{z}\left(-\frac{n}{2} - \frac{1}{2}\right)$$

n	$\mathbf{z}\left(-\frac{n}{2} - \frac{1}{2}\right)$	$-1 \leq -\frac{n}{2} - \frac{1}{2} \leq 2$
-5	0	$-\frac{1}{2} \leq -\frac{n}{2} \leq \frac{5}{2}$
-4	0	$-\frac{1}{2} \leq -\frac{n}{2} \leq \frac{5}{2}$
-3	1	$-1 \leq -n \leq 5$
-2	0	$1 \geq n \geq -5$
-1	2	$-5 \leq n \leq 1$
0	0	
1	3	



* Amplitude Scaling, Addition, Subtraction & multiplication: —

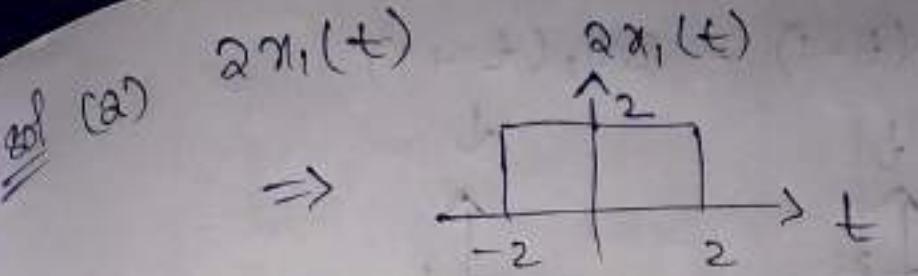
i) Given: $\mathbf{x}_1(t)$



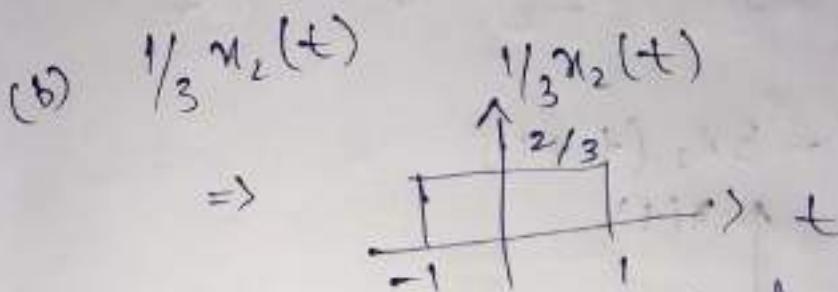
a) $2x_1(t)$ b) $\frac{1}{3}x_2(t)$ c) $-x_1(t)$

d) $2x_1(t) + 3x_2(t)$ e) $2x_1(t-1) - 3x_2(t+1)$

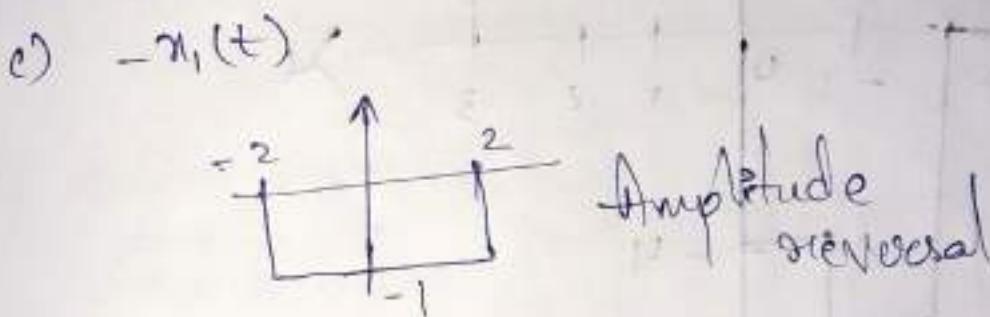
f) $x_1(t)x_2(t)$



(\because This process is called Amplification)

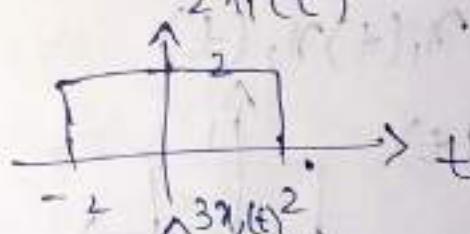


(\because This process is called

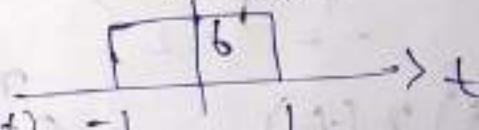


d) $2x_1(t) + 3x_2(t)$

$2x_1(t) \Rightarrow$

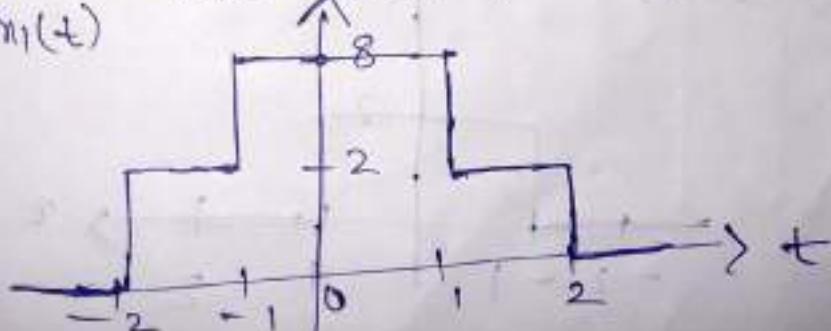


$3x_2(t) \Rightarrow$

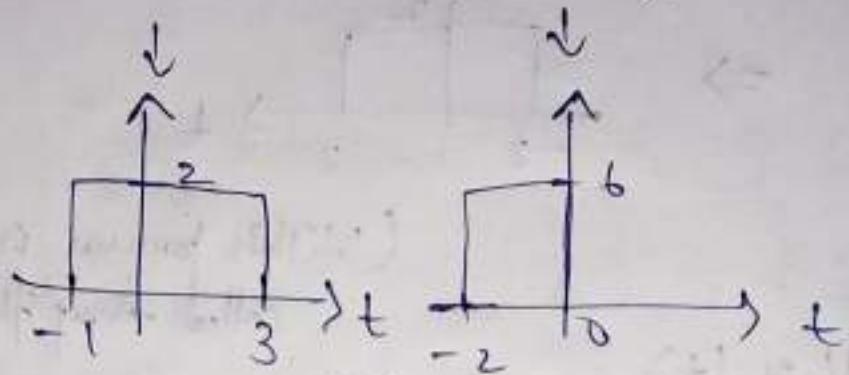


$2x_1(t)$

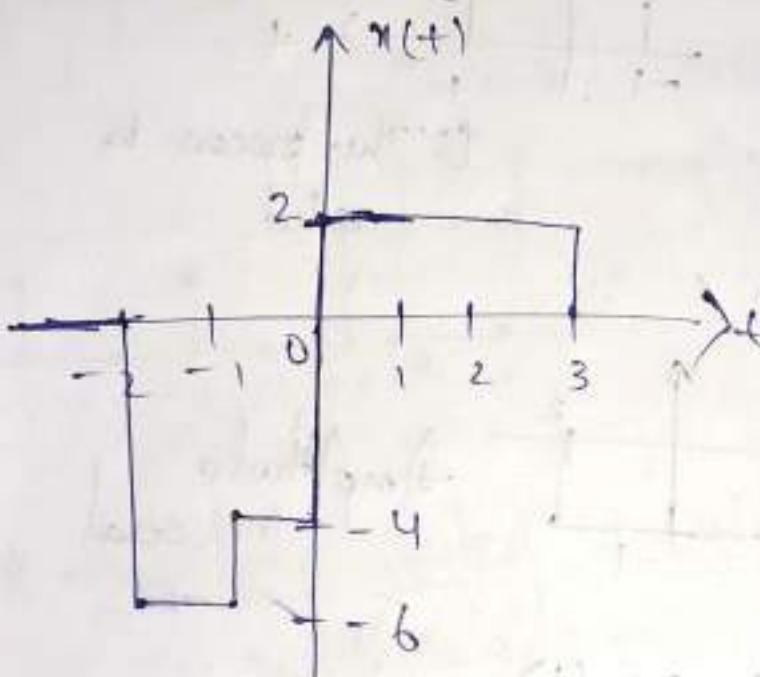
$2x_1(t) + 3x_2(t)$



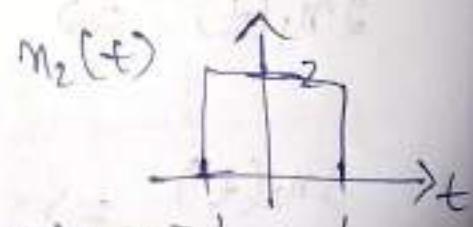
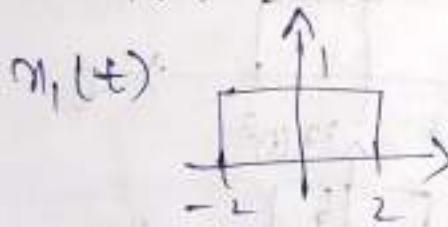
$$(e) 2x_1(t-1) - 3x_2(t+1)$$



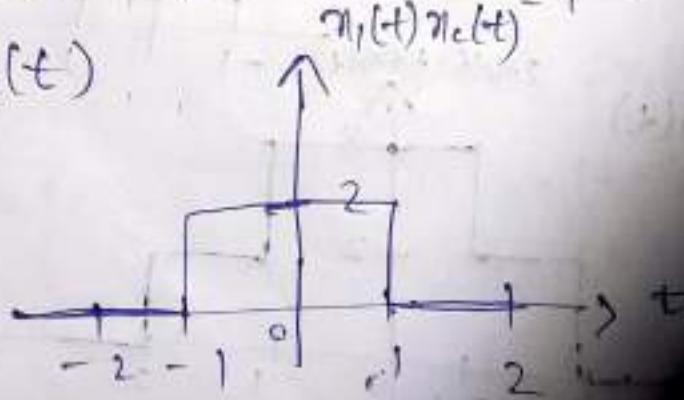
$$2x_1(t-1) - 3x_2(t+1)$$

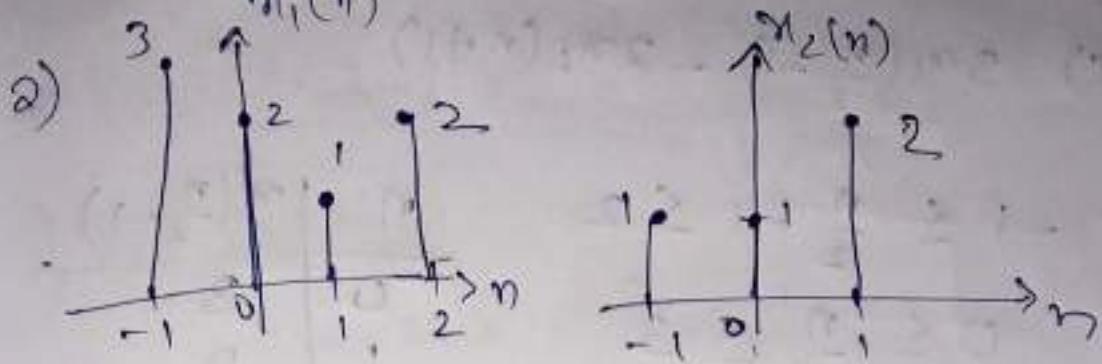


$$(f) x_1(t)x_2(t)$$

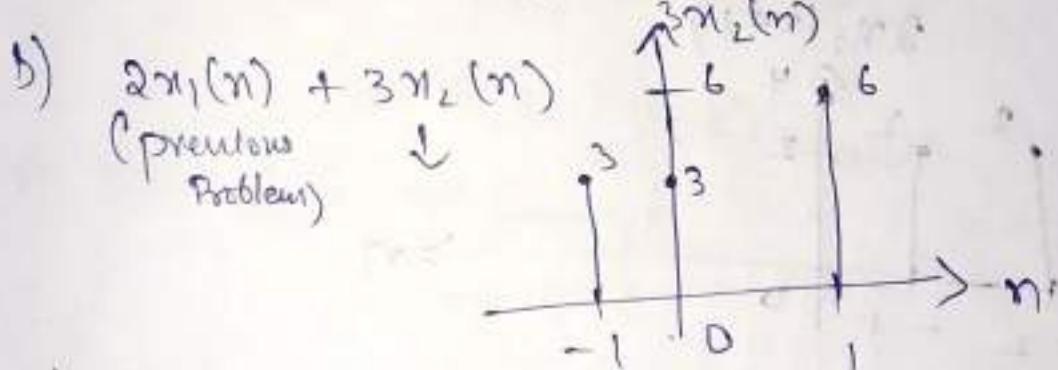
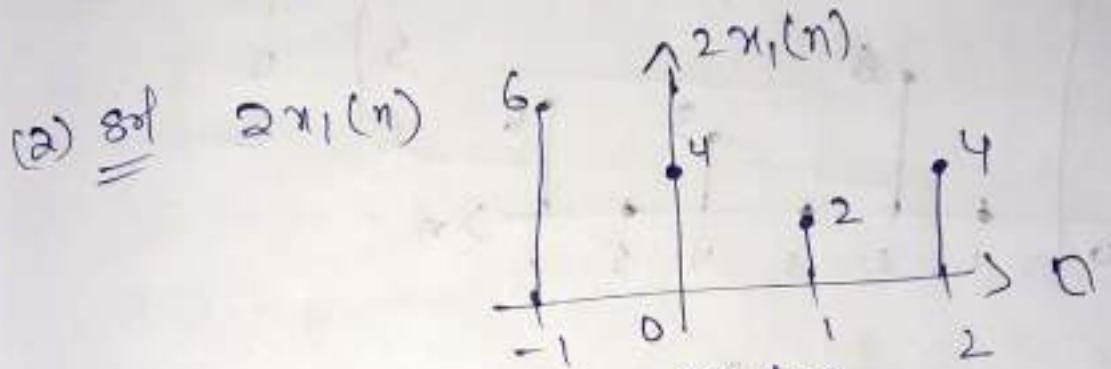


$$x_1(t)x_2(t)$$

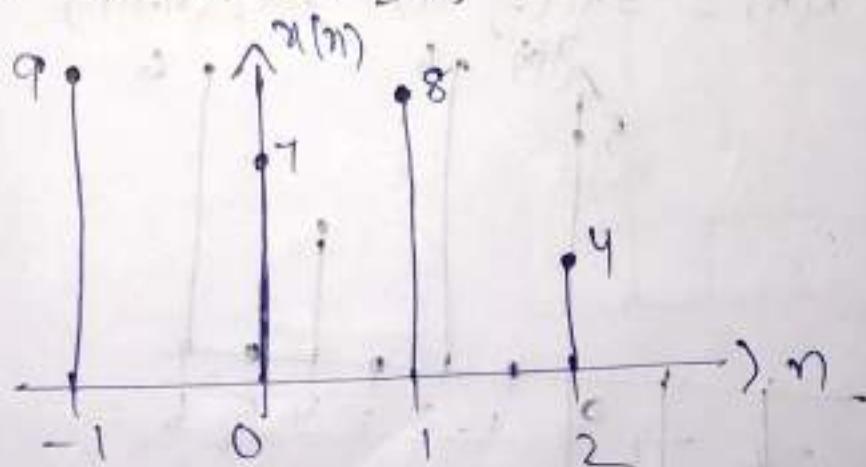




- (a) $2x_1(n)$ (b) $2x_1(n) + 3x_2(n)$
 (c) $3x_1\left(\frac{n}{2} - 1\right) - 2x_2(n+1)$ (d) $x_1(n)x_2(n)$



let $x(n) = 2x_1(n) + 3x_2(n)$



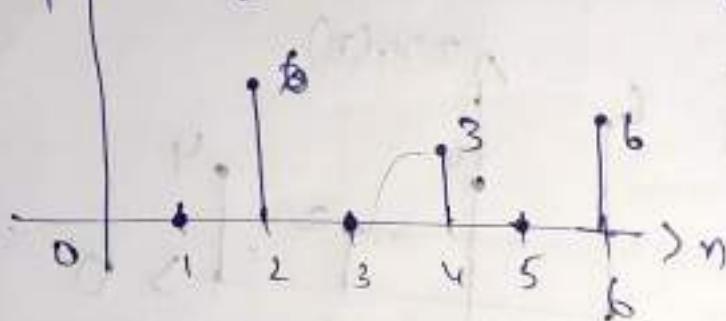
$$(C) \quad 3x_1\left(\frac{n}{2} - 1\right) - 2x_2(n+1)$$

$$-1 \leq \frac{n}{2} - 1 \leq 2$$

$$0 \leq \frac{n}{2} \leq 3$$

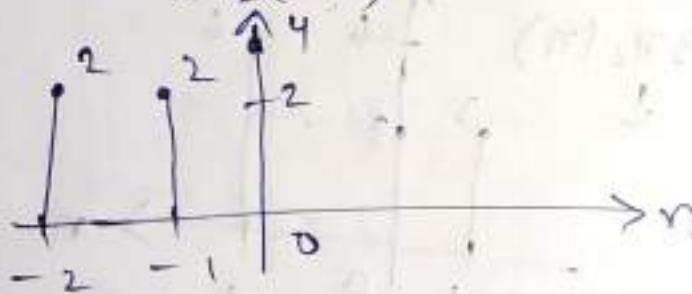
$$0 \leq n \leq 6$$

$$y = 3x_1\left(\frac{n}{2} - 1\right)$$

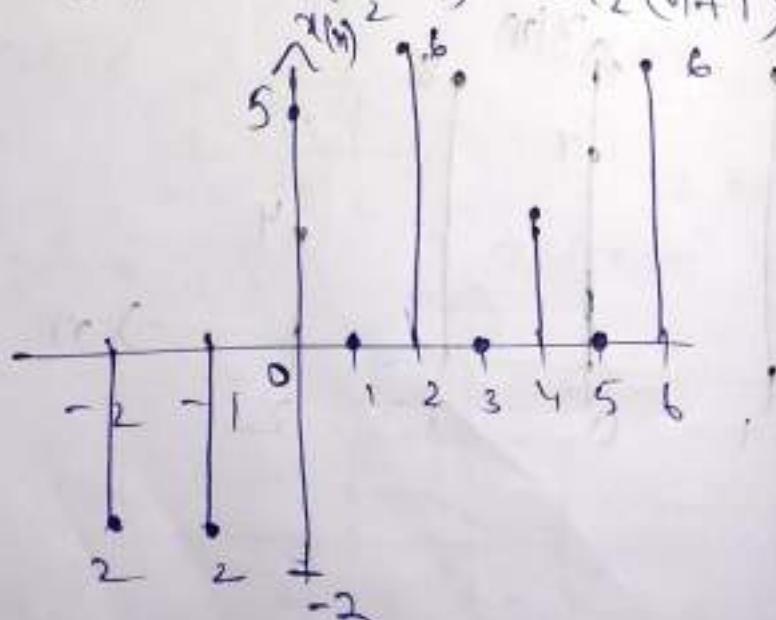


n	$x_1\left(\frac{n}{2} - 1\right)$
0	0
1	3
2	0
3	2
4	3
5	0
6	1
7	0
8	2

$$2x_2(n+1)$$



$$\text{Let } x(n) = 3x_1\left(\frac{n}{2} - 1\right) - 2x_2(n+1)$$



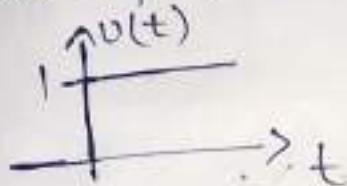
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* Basic / Standard / Elementary Signal :-

1) Unit Step Signal

Heaviside step signal

Continuous time
unit step signal

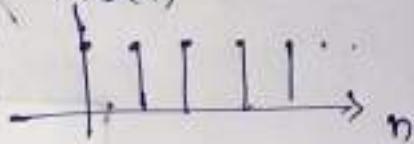


$$v(t) = 1, t \geq 0$$

$$v(t) = 0, t < 0$$

Discrete time

Unit step signal

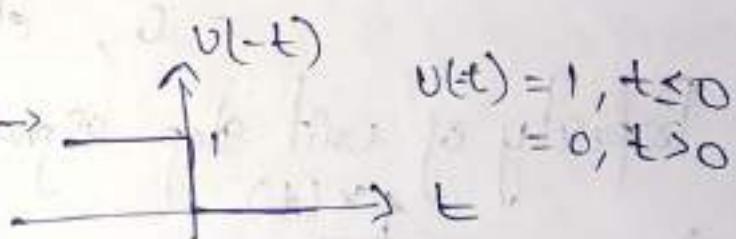


$$v(n) = 1, n \geq 0$$

$$v(n) = 0, n < 0$$

$$v(n) = \{1, 1, \dots\}$$

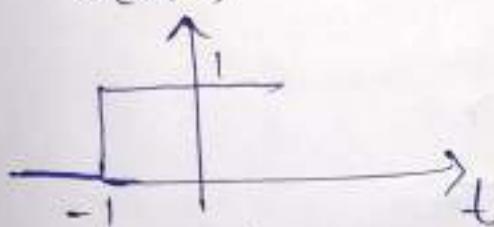
i) Plot $v(-t)$



$$\begin{aligned} v(-t) &= 1, t \leq 0 \\ &= 0, t > 0 \end{aligned}$$

a) Plot $v(t+1)$ & $v(t-1)$

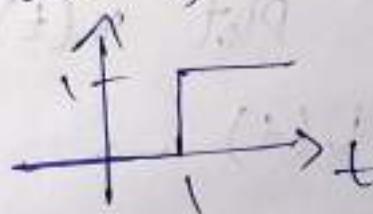
$v(t+1)$



$$v(t+1) = 1, t \geq -1$$

$$v(t+1) = 0, t < -1$$

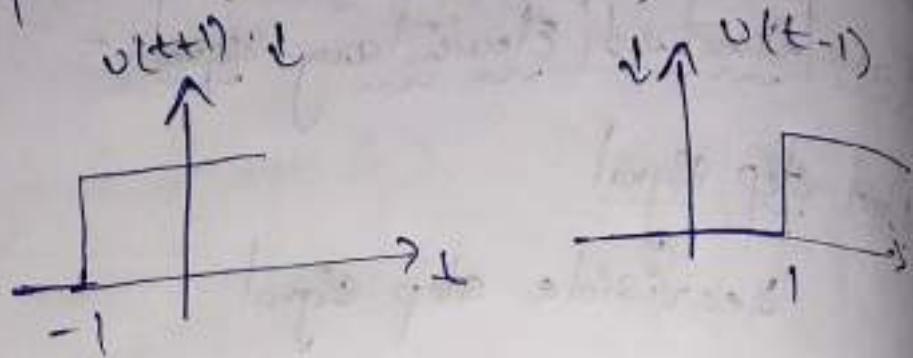
$v(t-1)$



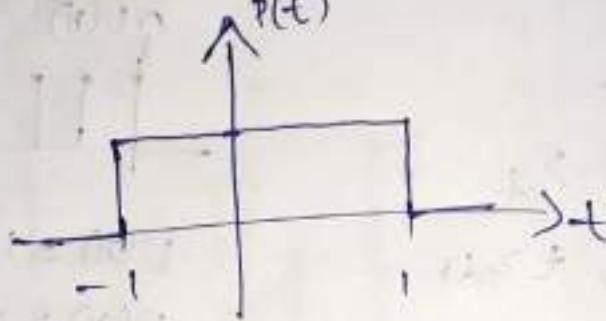
$$v(t-1) = 1, t \geq 1$$

$$v(t-1) = 0, t < 1$$

3) plot $U(t+1) - U(t-1)$



$$p(t) = U(t+1) - U(t-1)$$

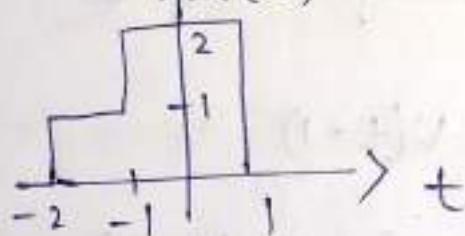


$$P(t) = 1, -1 \leq t \leq 1$$

$$= 0, \text{ else}$$

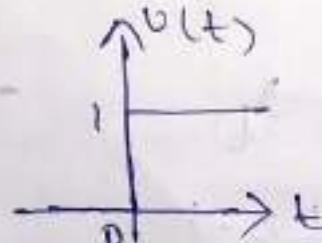
Property of unit step signal:-

1)



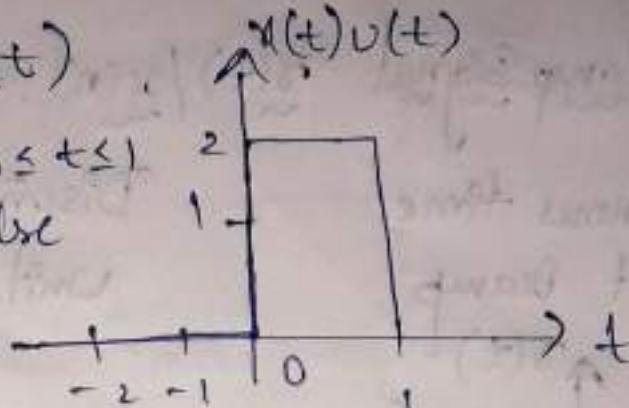
Plot $x(t) U(t)$

$U(t) \rightarrow$



$$(a) \quad u(t) v(t)$$

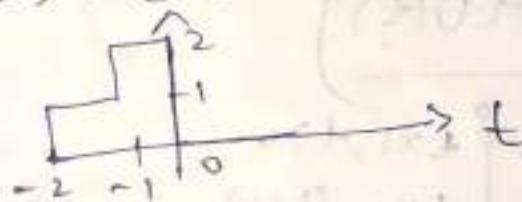
$$u(t)v(t) = 1, \quad 0 \leq t \leq 1 \\ = 0, \quad \text{else}$$



Note:- Step signal truncates the positive portion of $u(t)$ when multiplied by $v(t)$
Hence $v(-t)$ truncates positive portion of $u(t)$

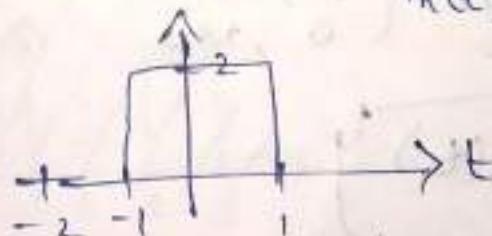
$$(b) \quad u(t) v(-t)$$

$$u(t)v(-t) = 1, \quad -2 \leq t \leq -1 \\ = 2, \quad -1 \leq t \leq 0 \\ = 0, \quad \text{else}$$



$$(c) \quad u(t)v(t+1)$$

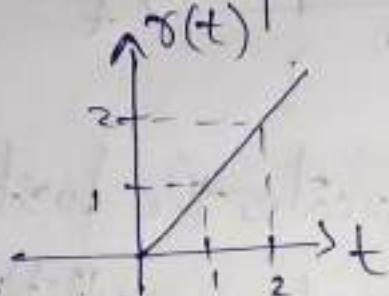
$$u(t)v(t+1) = 2, \quad -1 \leq t \leq 0 \\ = 0, \quad \text{else}$$



∴ Step signal is used to truncate (cutting off) the portion of a signal

2) Ramp signal $\underline{s(t)} / \underline{s(n)} :-$

* Continuous time
Unit ramp



$$y = mx$$

$$m = 1$$

$$y = x$$

$$\Rightarrow s(t) = t, t \geq 0$$

$$= 0, t < 0$$

$$\Rightarrow \boxed{s(t) = t u(t)}$$

$$= s(t) = \begin{cases} t, & t \geq 0 \\ 0, & t < 0 \end{cases}$$

$$= r(t) = \begin{cases} t, & t \geq 0 \\ 0, & t < 0 \end{cases}$$

$$\boxed{u(t) = \frac{d}{dt} s(t)}$$

$$= \frac{d}{dt} \begin{cases} t, & t \geq 0 \\ 0, & t < 0 \end{cases}$$

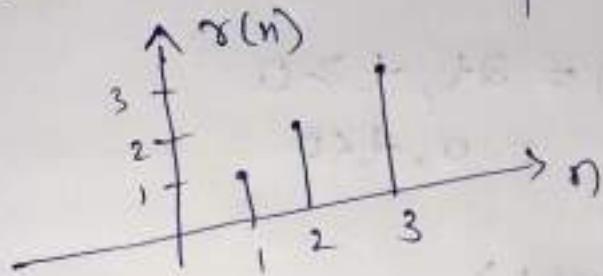
$$= \begin{cases} 1, & t \geq 0 \\ 0, & t < 0 \end{cases}$$

$$= v(t)$$

Discrete time
Unit ramp

$$\left\{ \gamma(t) = \int_{-\infty}^t v(t) dt \right\}$$

* Discrete time unit samp:-

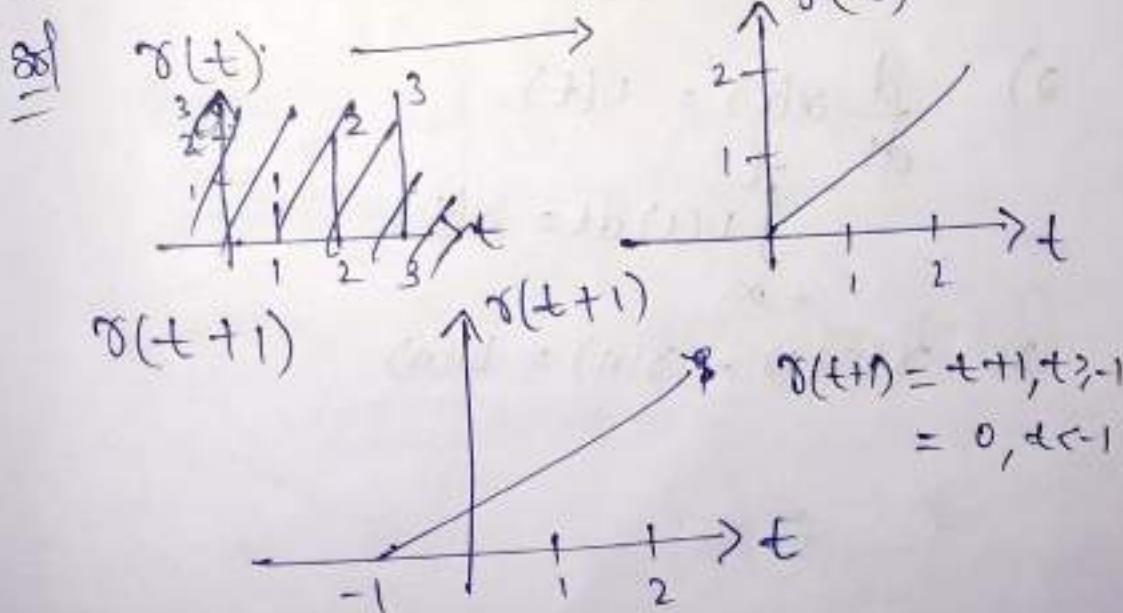


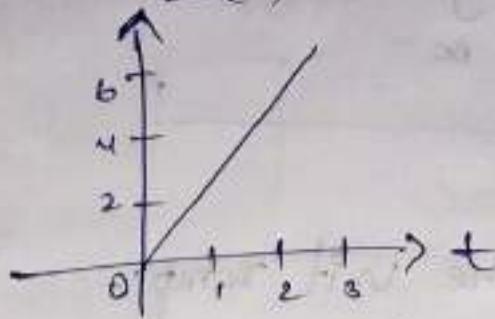
$$\begin{cases} \gamma(n) = n, & n \geq 0 \\ 0, & n < 0 \end{cases}$$

$$\left\{ \gamma(n) = n \cdot v(n) \right\}$$

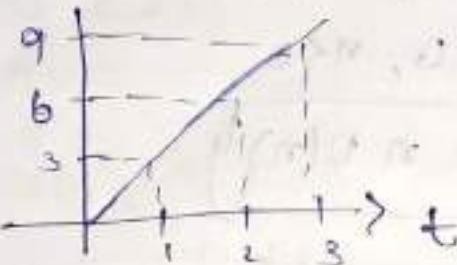
$$\Rightarrow \left\{ \gamma(n+1) - \gamma(n) = v(n) \right\}$$

i) plot $\gamma(t+1)$



2) plot $\gamma_{2v}(t)$ 

$$\gamma_{2v}(t) = \begin{cases} 2t, & t \geq 0 \\ 0, & t < 0 \end{cases}$$

3) plot $\gamma_{3v}(t)$ 

Summary:-

1) $\gamma_v(t) = t v(t)$

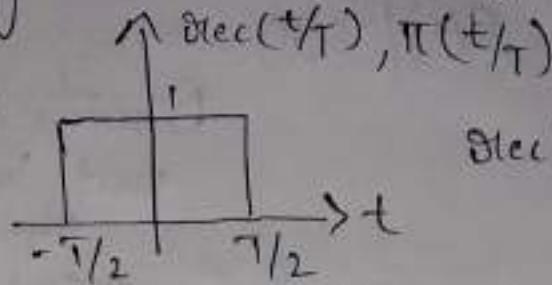
$v(n) = n v(n)$

2) $\frac{d}{dt} \gamma_v(t) = v(t)$

$$\int_{-\infty}^t v(t) dt = \gamma_v(t)$$

3) $\gamma_{(n+1)v} - \gamma_{nv} = V(n)$

3) Rectangular Gate (π) pulse signal

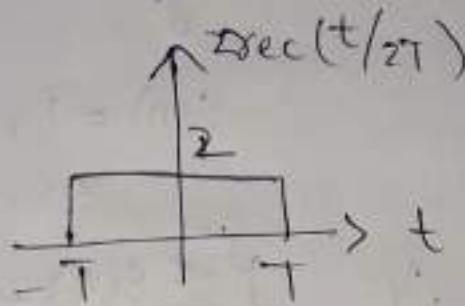


$$\text{sec}(t/T) = 1, -\frac{T}{2} \leq t \leq \frac{T}{2}$$

$$= 0, \text{else}$$

$T \rightarrow$ width of the pulse

* plot $2\text{sec}(t/2T)$

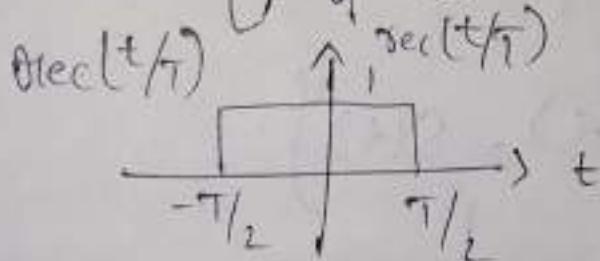
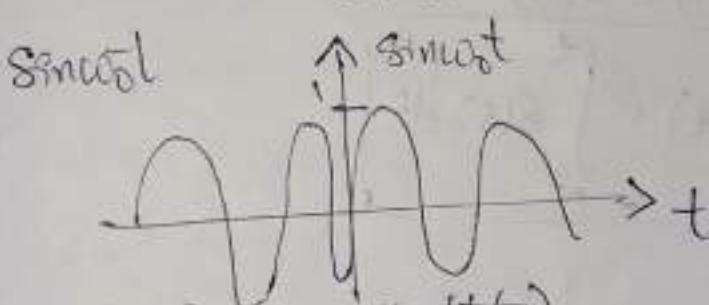


$$\text{sec}\left(\frac{t}{2T}\right) = 2$$

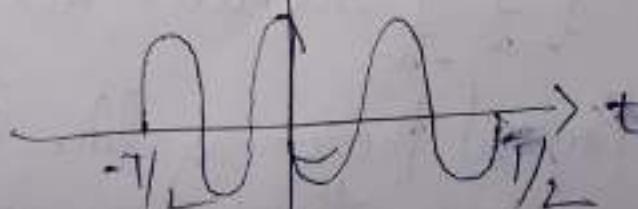
$$-T \leq t \leq T$$

$$= 0, \text{else}$$

1) Plot $\sin\omega_0 t \text{sec}(t/T)$

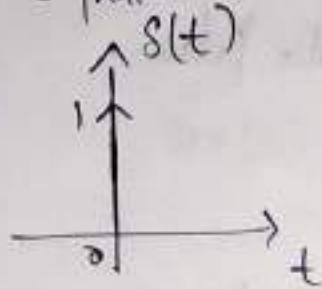


$\sin\omega_0 t (\text{sec}(t/T))$



3) 4) Impulse Signal:-
 $s(t)/s(n)$

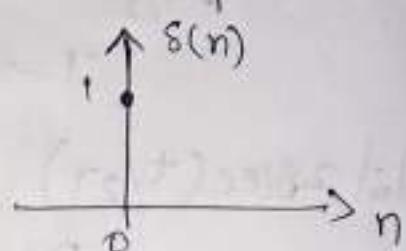
Continuous time
Unit impulse
signal



$$\begin{aligned}s(t) &= 1, t=0 \\ &= 0, t \neq 0\end{aligned}$$

$$\left\{ \text{Area} = \int_{-\infty}^{\infty} s(t) dt = 1 \right\}$$

Discrete time
Unit impulse
signal



$$\begin{aligned}s(n) &= 1, n=0 \\ &= 0, n \neq 0\end{aligned}$$

$$\left\{ \sum_{n=-\infty}^{\infty} s(n) = 1 \right\}$$

$$\begin{aligned}\frac{d}{dt}(u(t)) &= s(t) \\ \Rightarrow \left[u(t) = \int_{-\infty}^t s(t) dt \right]\end{aligned}$$

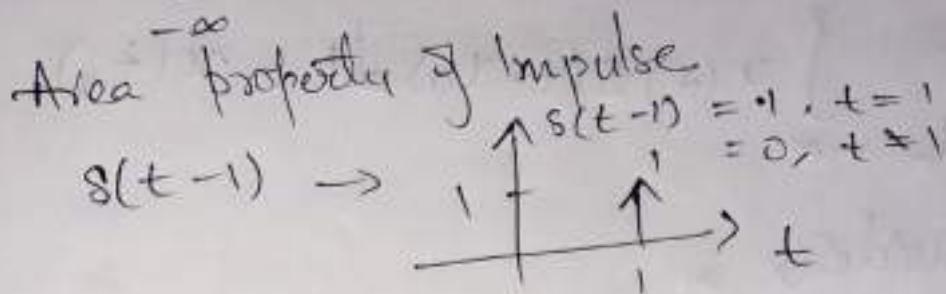
$$\left\{ \frac{d^2 s(t)}{dt^2} = s(t) \right\}$$

$$\left\{ \int_{-\infty}^t \int_{-\infty}^t s(t) dt = r(t) \right\}$$

Eg:- Physical phenomena like hammering of nail & thunderstorm are examples of impulse signal.

* Properties of Impulse Function:-

1) $\int_{-\infty}^{\infty} s(t) dt = 1$ Area is always one
 $\sum_{n=-\infty}^{\infty} s(n) = 1$



2) Product property

$$\alpha(n)s(n-n_0) = \alpha(n_0)s(n-n_0)$$

$$\alpha(t)s(t-t_0) = \alpha(t_0)s(t-t_0)$$

Evaluate

$$1) (n-3)s(n-1)$$

So we have Product property
 $\alpha(n)s(n-n_0) = \alpha(n_0)s(n-n_0)$

$$(n-3)s(n-1) = \alpha(1)s(n-1) - 2s(n-1)$$

$$2) (t^2+1)s(t)$$

and $\alpha(t)s(t-t_0) = \alpha(t_0)s(t-t_0)$

$$(t^2+1)s(t) = \alpha(0)s(t) = 1s(t) = s(t)$$

(3) Sifting property

$$\sum_{n=-\infty}^{\infty} x(n) s(n-n_0) = x(n_0)$$

$$\int_{-\infty}^{\infty} x(t) s(t-t_0) dt = x(t_0)$$

Consider,

$$\int_{-\infty}^{\infty} x(t) s(t-t_0) dt$$

$$= \int_{-\infty}^{\infty} x(t_0) s(t-t_0) dt \quad \text{from product property}$$

$$= x(t_0) \int_{-\infty}^{\infty} s(t-t_0) dt$$

$$\therefore x(t_0)$$

$$\int_{t_1}^{t_2} x(t) s(t-t_0) dt = x(t_0), t_1 \leq t_0 \leq t_2$$

$$= 0, \text{ else}$$

$$\sum_{n_1}^{n_2} x(n) s(n-n_0) = x(n_0), n_1 \leq n_0 \leq n_2$$

$$= 0, \text{ else}$$

$$\left\{ \begin{array}{l} \int_{-\infty}^{\infty} x(t) s(t) dt = x(0) \\ x(t) s(t) = x(0) \delta(t) \end{array} \right.$$

$$1) \int_0^2 (t^2+1) (\delta(t-1/2)) dt$$

$$\begin{aligned} \gamma(t_0) &= \gamma(1/2) \\ &= (1/2)^2 + 1 = 5/4 \end{aligned}$$

$$2) \int_1^2 (t^2+1) \delta(t-3) = 0$$

$$3) \text{ Evaluate } \int_{-\infty}^{\infty} (t^2+1) \delta(t-2) dt$$

$$\text{Sol} \quad \int_{-\infty}^{\infty} (t^2+1) \delta(t-2) dt$$

$$= \gamma(t_0)$$

$$= \gamma(2)$$

$$= 2^2 + 1/2$$

$$= 4 + 1/2 = 9/2$$

$$4) \int_{-1000}^1 (t^2+1) \delta(t-1) dt = 0$$

$$5) \int_{-\infty}^{\infty} \cos \pi(t-1) \delta(t-2) dt$$

$$\text{Sol} \Rightarrow \gamma(t) = \gamma(2)$$

$$= \cos \pi(2-1)$$

$$= \cos \pi$$

$$= -1$$

$$\begin{aligned}
 (6) \quad & \int_{-\infty}^{\infty} \cos \frac{\pi}{2}(t-5) s(at-3) dt \\
 & = \int_{-\infty}^{\infty} \cos \left(\frac{\pi}{2}(t-5) \right) s(at-3) dt \\
 & = \alpha(3) \\
 & = \cos \frac{\pi}{2}(3-5) \\
 & = \cos -\pi \\
 & = \cos \pi = -1
 \end{aligned}$$

$$\begin{aligned}
 7) \quad & \int_{-\infty}^{\infty} (t + 8 \sin 2\pi t) s(t-1) dt \\
 & = \alpha(1) \\
 & = 1 + \sin 2\pi \\
 & = 1 + 0 \\
 & = 1,
 \end{aligned}$$

(4) Scaling property:-

$$\left\{ s(at-b) = \frac{1}{|a|} s(t-b/a) \right\}$$

Consider $\int_{-\infty}^{\infty} \gamma(t) s(at-b)$

Put $at-b=u$
 $dt = \frac{du}{a}$ $t = \frac{u+b}{a}$

$$\begin{aligned}
 & \int_{-\infty}^{\infty} x\left(\frac{b+u}{a}\right) \delta(u) du \\
 &= \frac{1}{a} \int_{-\infty}^{\infty} x\left(\frac{b+u}{a}\right) du = \frac{1}{a} x(0) \\
 &= \frac{1}{a} x(b/a)
 \end{aligned}$$

$$\begin{aligned}
 \int_{-\infty}^{\infty} x(t) \delta(at-b) &= \frac{1}{a} x(b/a) \\
 &= \frac{1}{a} \int_{-\infty}^{\infty} x(t) \delta(t-b/a) dt \\
 \boxed{\delta(at-b) = \frac{1}{|a|} \delta(t-b/a)}
 \end{aligned}$$

$$\left\{ \delta(at) = \frac{1}{|a|} \delta(t) \right\} \quad \left\{ \delta(-at-b) = \frac{1}{|a|} \delta(t-b/a) \right\}$$

i) $\int_{-\infty}^{\infty} \cos \frac{\pi}{2}(t-3) \delta(-2t-3)$

$$\begin{aligned}
 & \stackrel{S}{=} \int_{-\infty}^{\infty} \cos \frac{\pi}{2}(t-5) \frac{1}{|-2|} \delta(t-3/2) dt
 \end{aligned}$$

$$\frac{1}{2} \int_{-\infty}^{\infty} \cos \frac{\pi}{2}(t-5) \delta(t-3/2) dt$$

$$\begin{aligned}
 & \frac{1}{2} \left[\cos \frac{\pi}{2} \left(\frac{3}{2} - 5 \right) \right] \\
 &= \frac{1}{2} \cos \left(\frac{\pi}{2} \right) \left(\frac{7}{2} \right)
 \end{aligned}$$

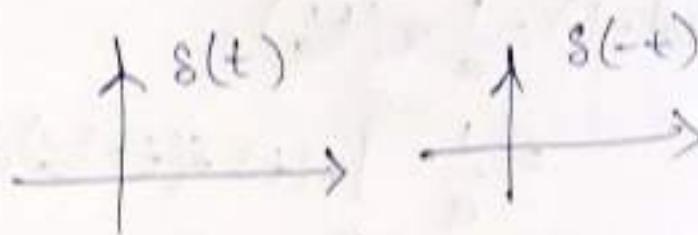
$$= \frac{1}{2} \cos \frac{7\pi}{4} = 0.497$$

5) Symmetry Property:-

Impulse function exhibits even symmetry

$$\delta(-t) = \delta(t), \quad \delta(-n) = \delta(n)$$

$$\pi(-t) = \pi(t)$$



i) Express mathematically the signal

$$v(2t+1)$$

Given:- $v(2t+1)$

We know that

$$v(t) = 1, \quad t \geq 0$$

$$= 0, \quad t < 0$$

$$v(2t+1) = 1, \quad 2t+1 \geq 0$$

$$t \geq -\frac{1}{2}$$

$$= 0, \quad 2t+1 < 0$$

$$t < -\frac{1}{2}$$

2) $v(2t-3)$

Ans. i.e. T $v(t) = 1, t \geq 0$
 $= 0, t < 0$

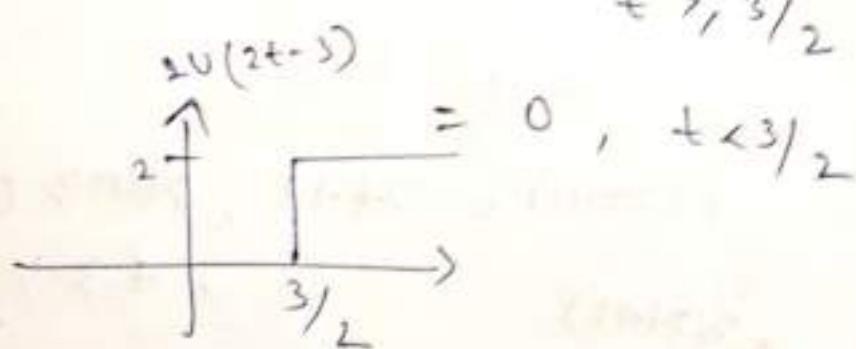
$$\begin{aligned} v(2t-3) &= 1, 2t-3 \geq 0 \\ &\quad t \geq 3/2 \\ &= 0, 2t-3 < 0 \\ &\quad t < 3/2 \end{aligned}$$

3) $2v(2t-3)$

$$\begin{aligned} v(t) &= 1, t \geq 0 \\ &= 0, t < 0 \end{aligned}$$

$$2v(2t-3) = 2, 2t-3 \geq 0$$

$$t \geq 3/2$$



4) $\gamma(2t)$, ~~$\gamma_0(2t)$~~

Ans. i.e. T $\gamma(t) = t, t \geq 0$
 $= 0, t < 0$

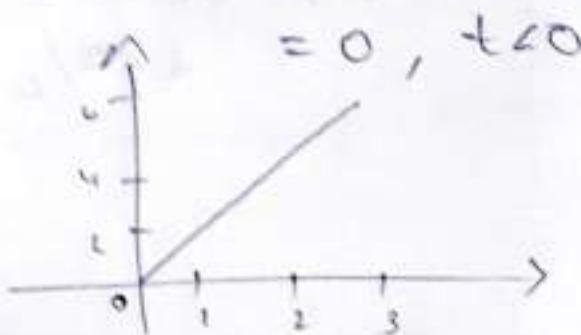
$$\begin{aligned} \gamma(2t) &= 2t, 2t \geq 0 \\ &= 0, 2t < 0 \end{aligned}$$

5) $v(t)$

$$v(t) = t, \quad t \geq 0$$

$$= 0, \quad t < 0$$

$$2v(t) = 2t, \quad t \geq 0$$

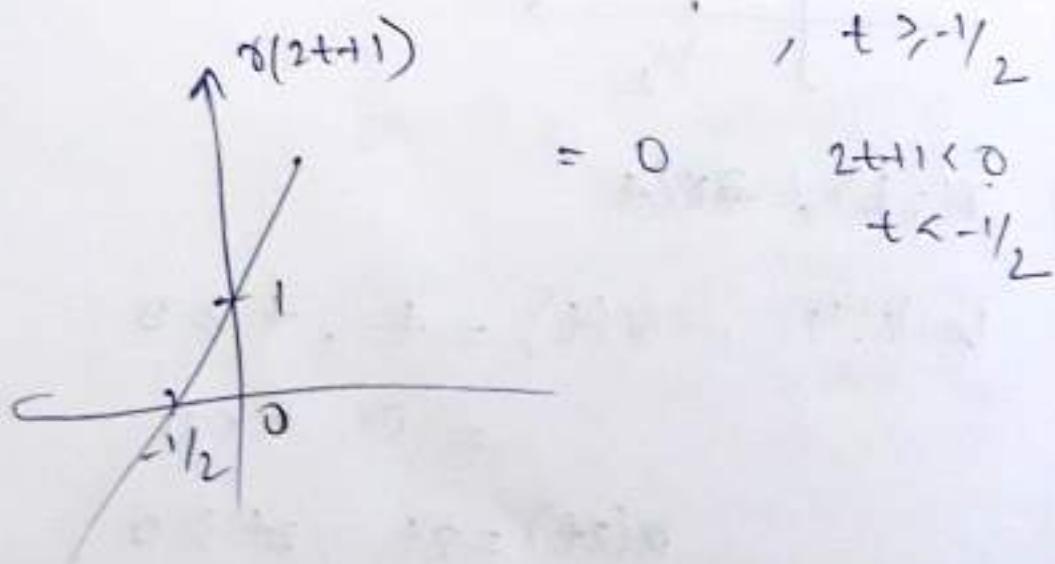


6) $v(2t+1)$

$$v(t) = t, \quad t \geq 0$$

$$= 0, \quad t < 0$$

$$v(2t+1) = 2t+1, \quad 2t+1 \geq 0$$



$$\delta(a-t) = \frac{1}{|a|} \delta(t-a)$$

Pb) Evaluate $\delta(2t-3)$

$$\frac{1}{|2|} \delta(t - 3/2)$$

Pb) Evaluate $\delta(-2t-3)$

$$\frac{1}{2} \delta(t + 3/2)$$

$$\frac{1}{2} \delta(t + 3/2)$$

Precedence rule is not applicable for
impulse function • time scaled

Pb) $\int_{-\infty}^{\infty} (t^2+1) \delta(-2t-3) dt$

$$\int_{-\infty}^{\infty} (t^2+1) \frac{1}{2} \delta(t + 3/2) dt$$

$$\frac{1}{2} \int_{-\infty}^{\infty} (t^2+1) \delta(t + 3/2) dt$$

$$= \frac{1}{2} x(t_0)$$

$$= \frac{1}{2} [x(t^2+1)]_{t=-3/2}$$

$$= \frac{1}{2} \left(\frac{9}{4} + 1\right)$$

$$= 13/8$$

Pb1) Evaluate $\sum_{-\infty}^{\infty} n^2 \delta(n-4)$

$$= (n^2)_{n=4}$$

$$= (4)^2 = 16$$

Pb2) $\sum_{-1}^2 4^n \delta(n+3)$

$$= 0 \quad (\because 3 \text{ does not lie between } -1 \text{ & } 2)$$

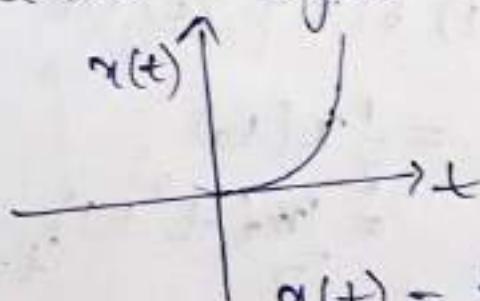
Pb3) $\sum_{-\infty}^{\infty} 4^n \delta(n+3)$

$$= x(n_0)$$

$$= (4^n)_{n=-3}$$

$$= 4^3 = 4^6$$

* Parabolic Signal:-



$$x(t) = t^2/2, t \geq 0$$

Mathematically, $x(t) = 0, t < 0$

$$x(t) = \frac{t^2}{2} u(t)$$

$$\frac{dx}{dt} = \alpha(t)$$

Differentiation of time variable signal is
a ramp

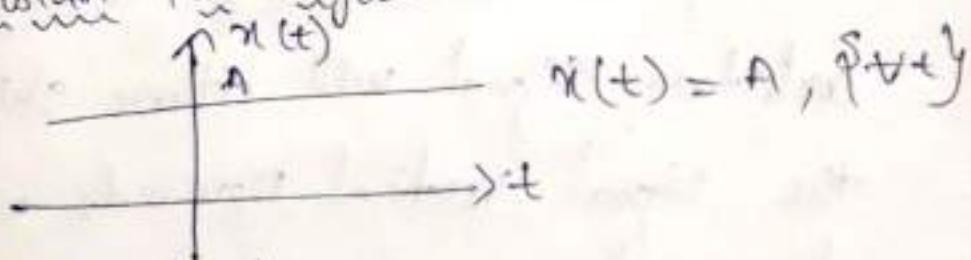
$$v(t) = \frac{d^2}{dt^2} x(t)$$

$$s(t) = \frac{d^3}{dt^3} x(t)$$

$$x(t) \xrightarrow{\frac{d}{dt}} \alpha(t) \xrightarrow{d/dt} v(t) \xrightarrow{d/dt} s(t)$$

$$s(t) \xrightarrow{\int_{-\infty}^t} v(t) \xrightarrow{\int_{-\infty}^t} \alpha(t) \xrightarrow{\int_{-\infty}^t} x(t)$$

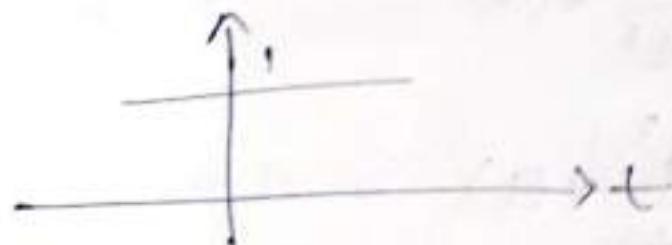
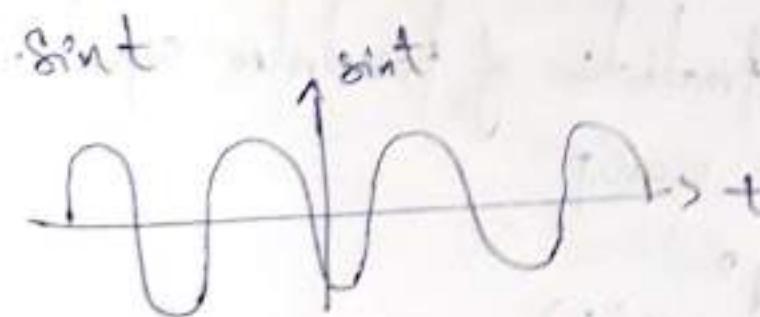
* Constant DC signal:-



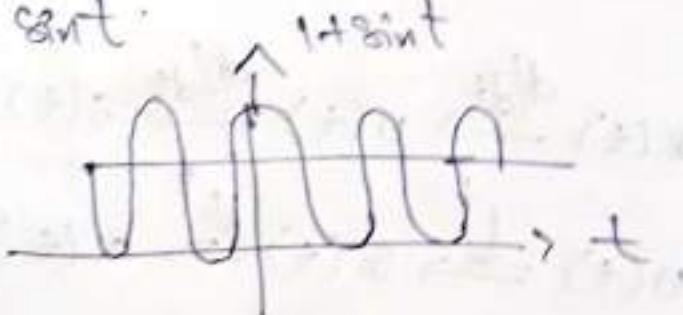
Mathematically $x(t) = A, \forall t \in \mathbb{R}$

$$a(t) = A(v(t) - v(-t))$$

Pb) Plot $1 + \sin t$



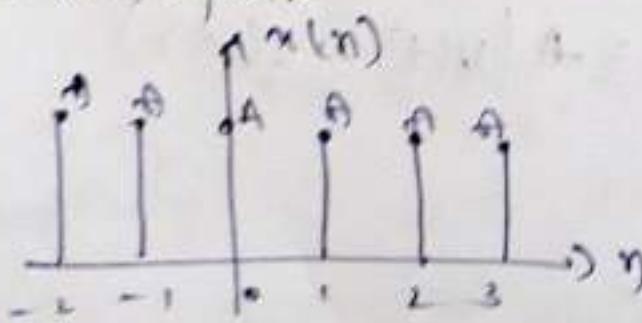
$1 + \sin t$



Note:-

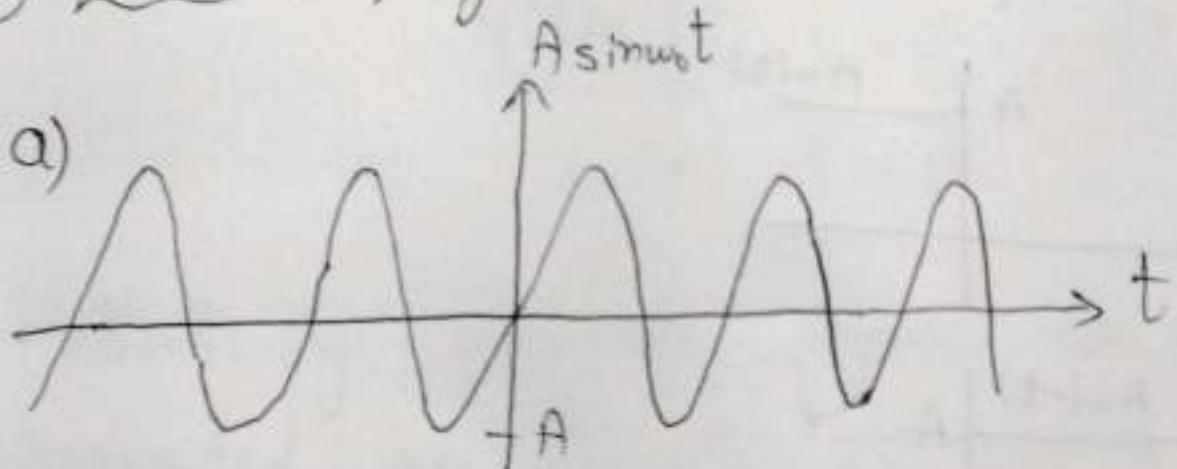
A D.C. Signal when added to any arbitrary signal will either shift the signal vertically upwards or downwards as shown.

Discrete D.C. signal:-



$$x(n) = A$$

7.) Sinusoidal Signal:- $\sin\omega_0 t, \cos\omega_0 t$



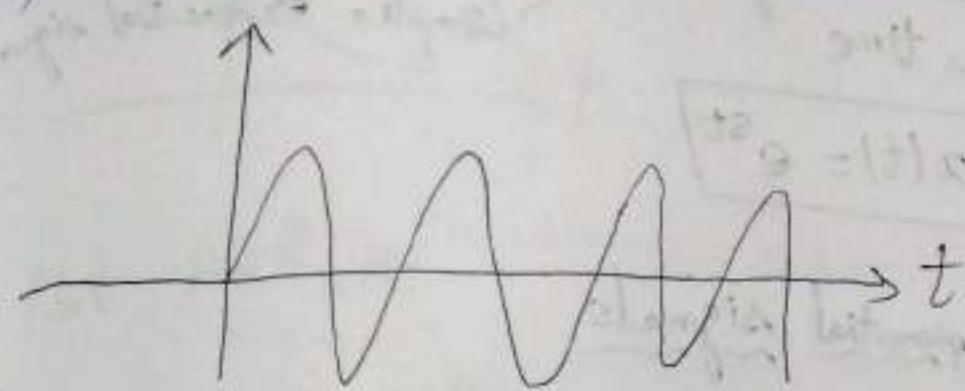
$$x(t) = A \sin \omega_0 t$$

A - maximum amplitude ω_0 - angular frequency

f_0 - frequency $\omega_0 = 2\pi f_0$

$$T_0 = \frac{1}{f_0}$$

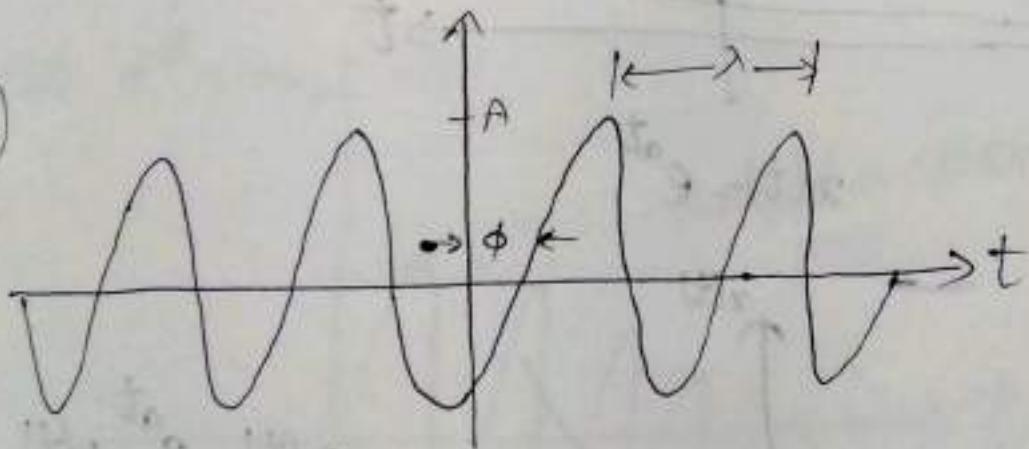
b) Plot $\sin \omega_0 t u(t)$



$$x(t) = A \sin \omega_0 t, \quad t \geq 0$$

$$= 0, \quad t < 0$$

b)



$$x(t) = A \sin(\omega_0 t + \phi)$$

$$\phi = \frac{2\pi}{\lambda} \times d$$

$\phi \rightarrow \text{phase}$

The phase of a signal dictates the offset of the signal from its origin.

8.) Exponential Signal

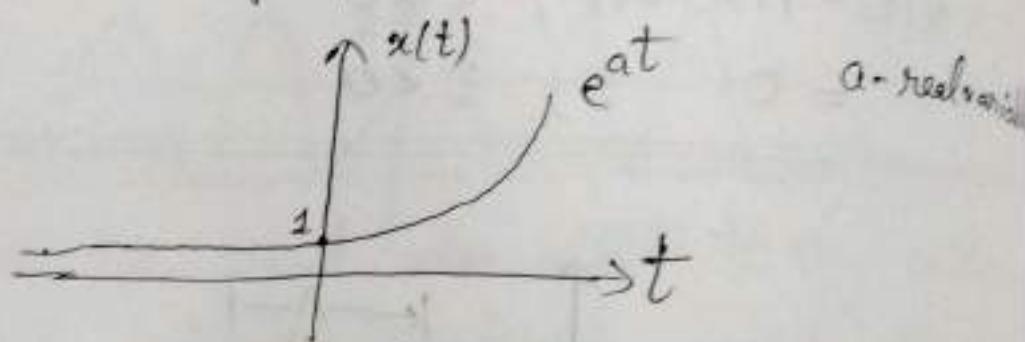
① Continuous time

$x(t) = e^{st}$

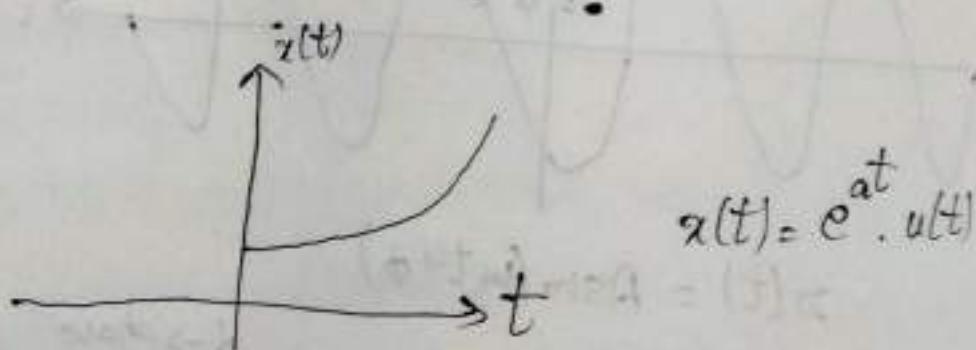
Real exponential signal
Complex exponential sig.

→ Real exponential signals

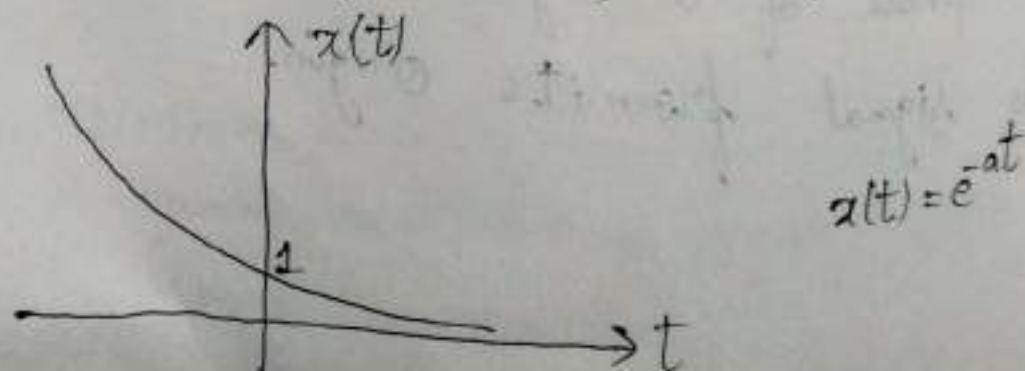
a) Exponentially growing signals



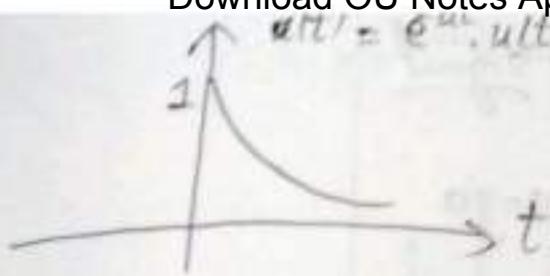
✓ b)
(default)



c) Exponentially decreasing decaying signal

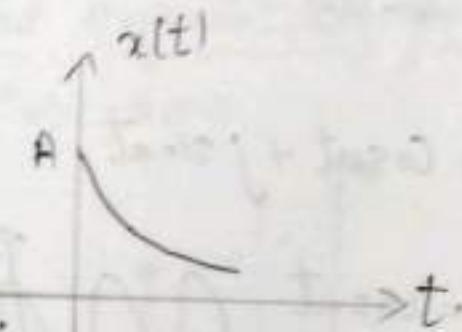


d)



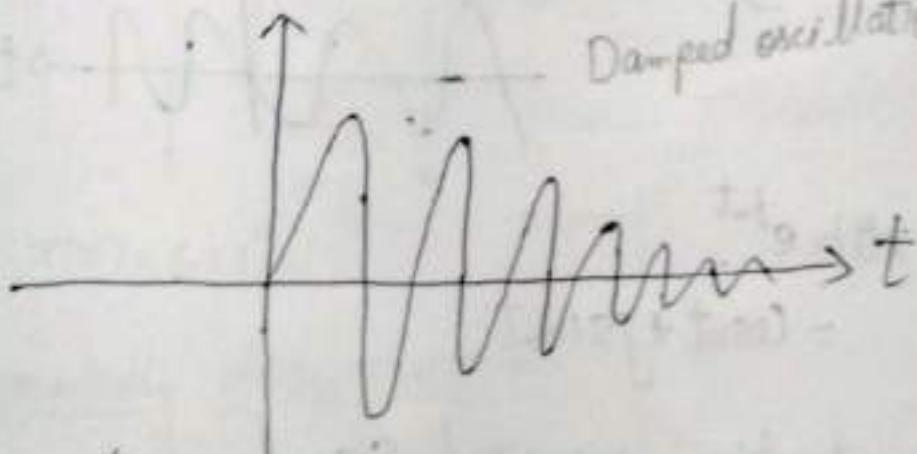
b) Plot $Ae^{at} u(t)$

Sol:-



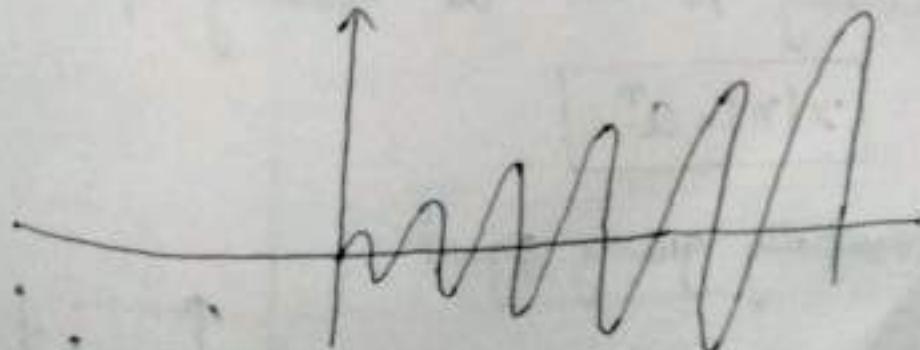
b) Plot $e^{at} \sin \omega_0 t, u(t)$

Sol:-



b) Plot $e^{at} \sin \omega_0 t u(t)$

Sol:-



→ Complex exponential signal

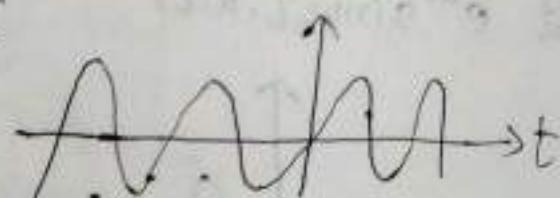
$$x(t) = e^{jat}, \quad s = ja$$

Complex exponential function is also a periodic signal function similar to sinusoidal signal.

$$x(t) = e^{jat} = \cos at + j \sin at$$

$$\operatorname{Re}[e^{jat}] = \cos at$$

$$\operatorname{Im}[e^{jat}] = \sin at$$



$$x(t) = e^{jwt}$$

$$= \cos wt + j \sin wt$$

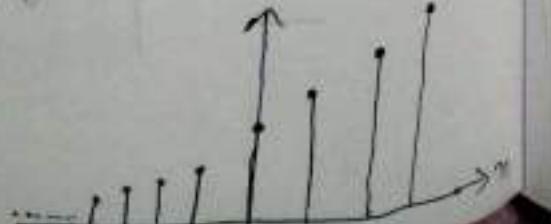
② Discrete time exponential signal

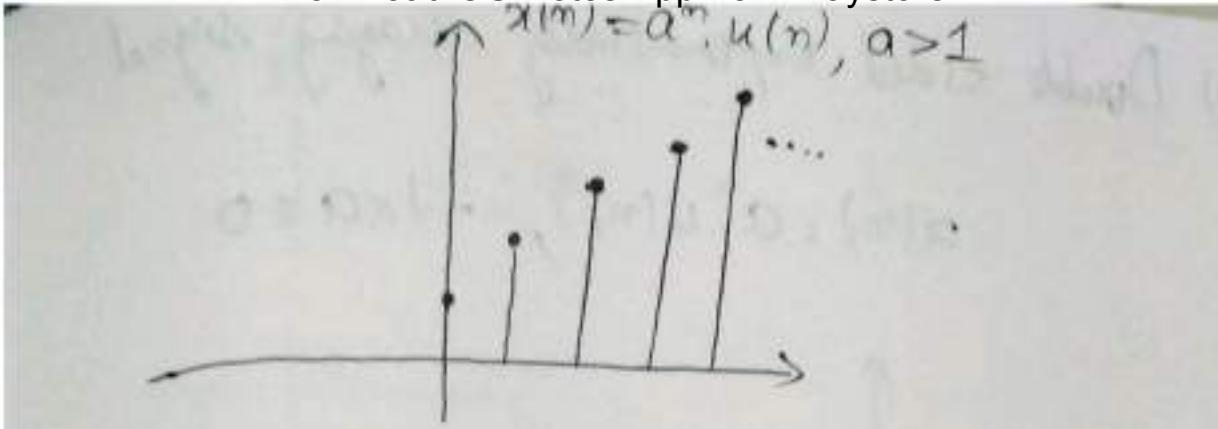
Mathematically it has the following representation

$$x(n) = a^n$$

i) Exponentially growing signal

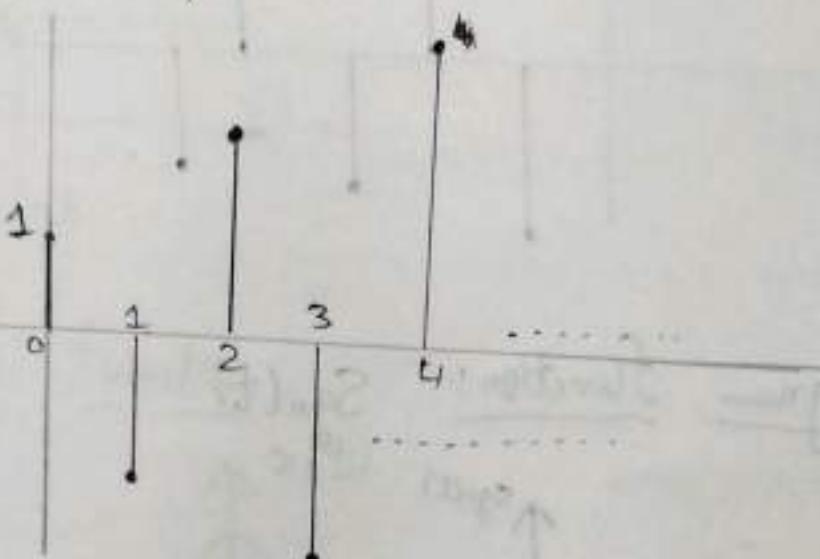
$$x(n) = a^n, \quad a > 1$$





ii) Double sided exponentially growing signal:-

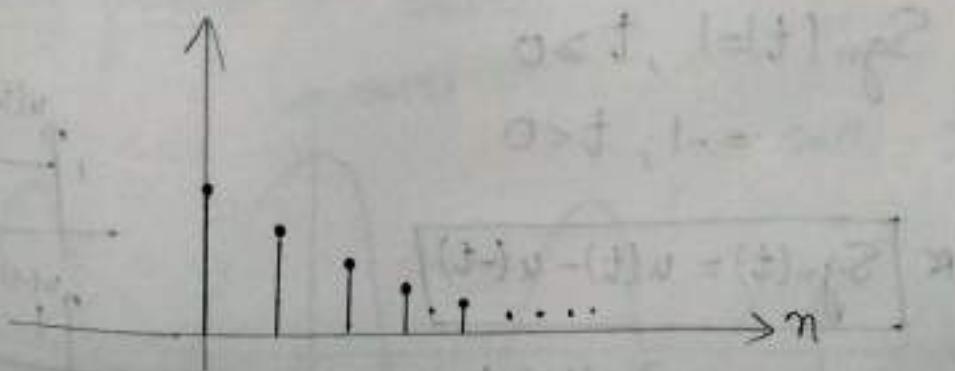
$$x(n) = a^n u(n), a < -1$$



$$x(n) = (-2)^n u(n)$$

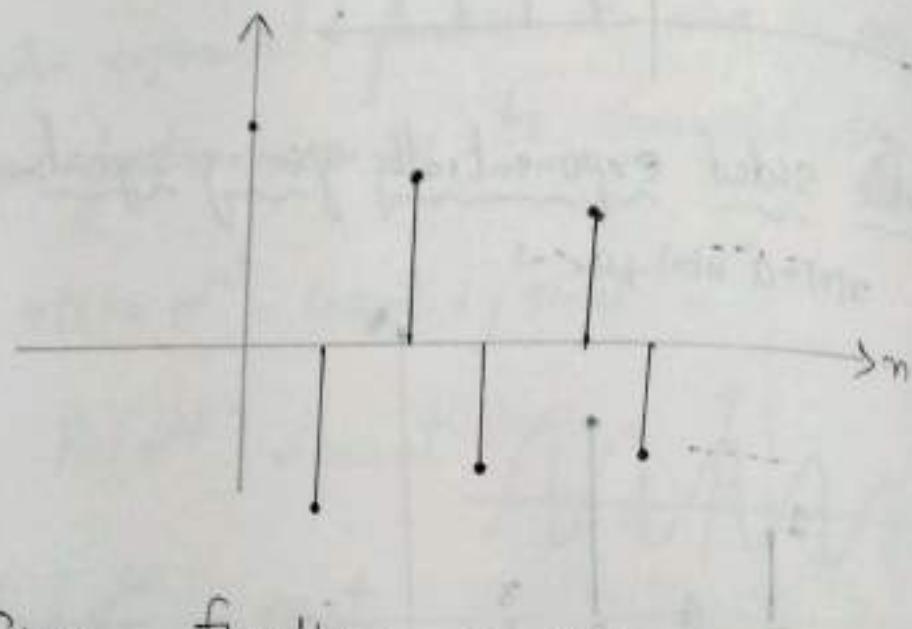
iii) Exponentially decaying signal

$$x(n) = a^n u(n), 0 < a < 1$$

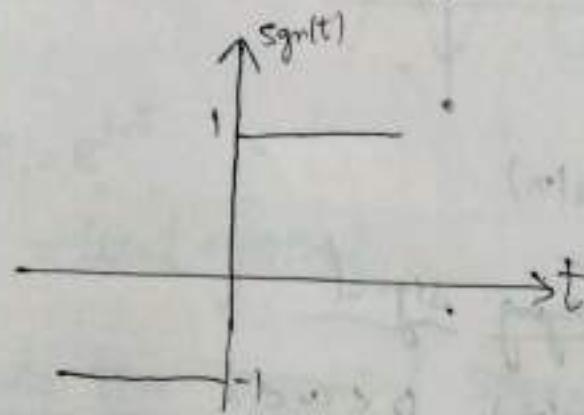


iv) Double sided exponentially decaying signal

$$x(n) = a^n u(n), -1 \leq a < 0$$



⑨ Signum function :- $\text{Sgn}(t)$

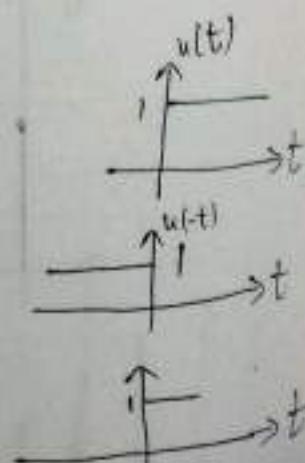


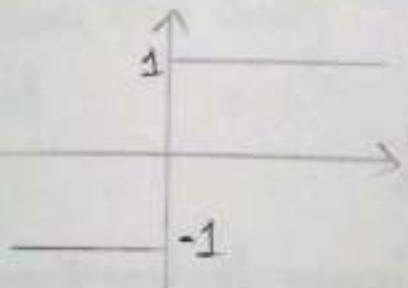
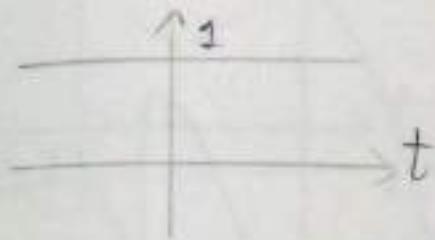
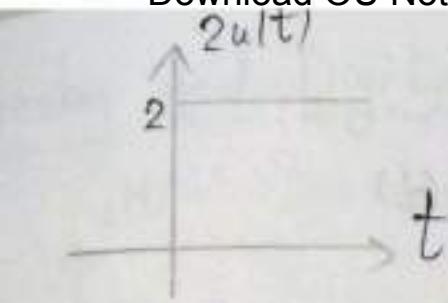
$$\text{Sgn}(t) = 1, t \geq 0$$

$$= -1, t < 0$$

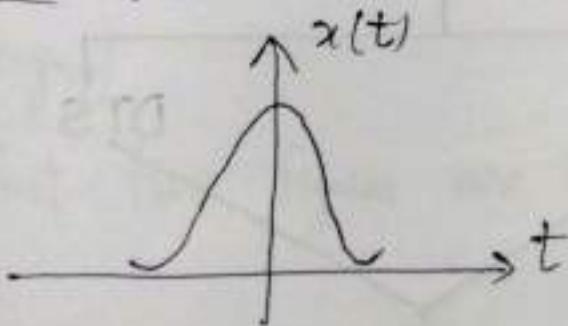
*
$$\boxed{\text{Sgn}(t) = u(t) - u(-t)}$$

$$= 2u(t) - 1$$



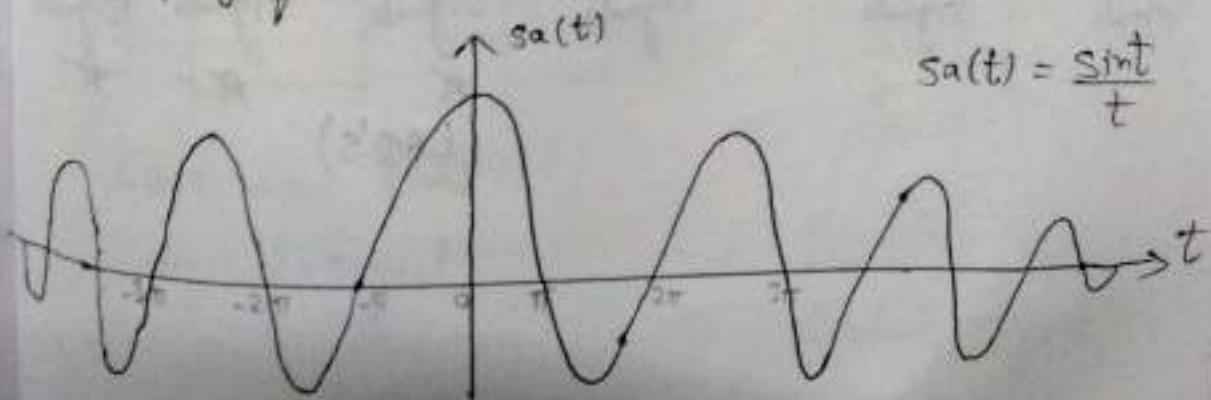


10) Gaussian function:-

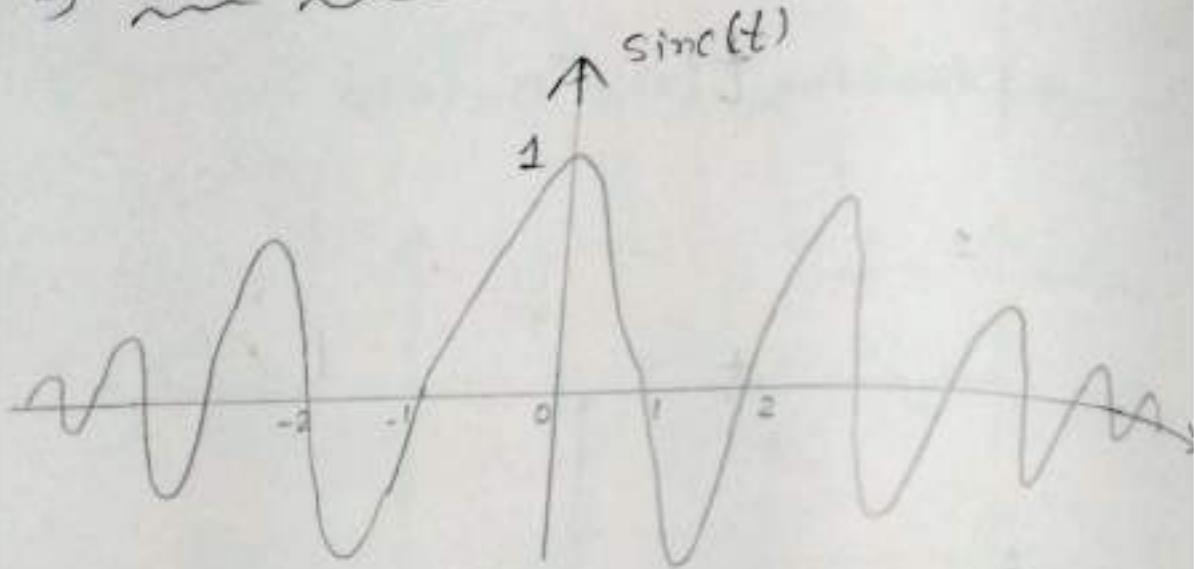


$$x(t) = e^{-at^2}$$

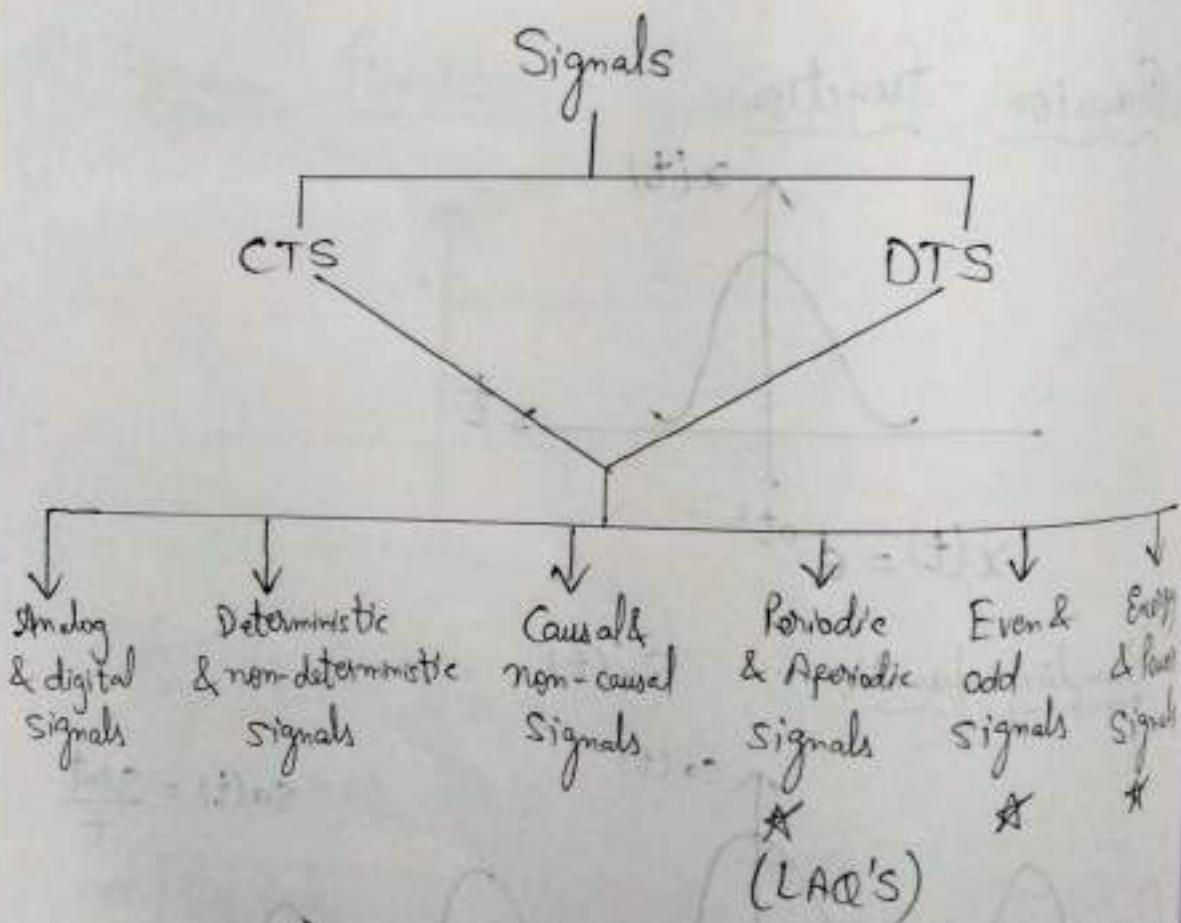
11) Sampling function :- $s_a(t)$



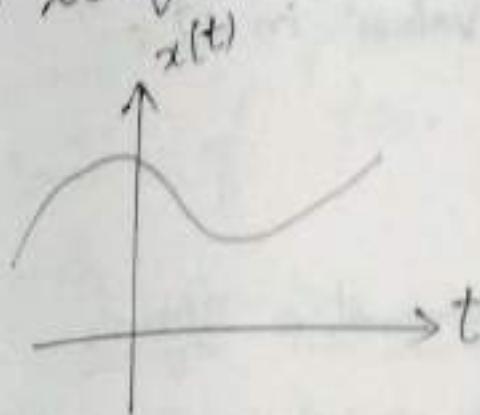
12) Sinc function $\text{sinc}(t)$



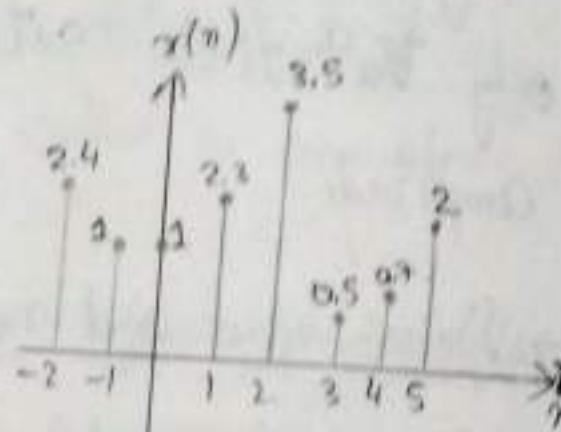
$$\text{sinc}(t) = \frac{\sin \pi t}{\pi t}$$



1) Analog and Digital Signals



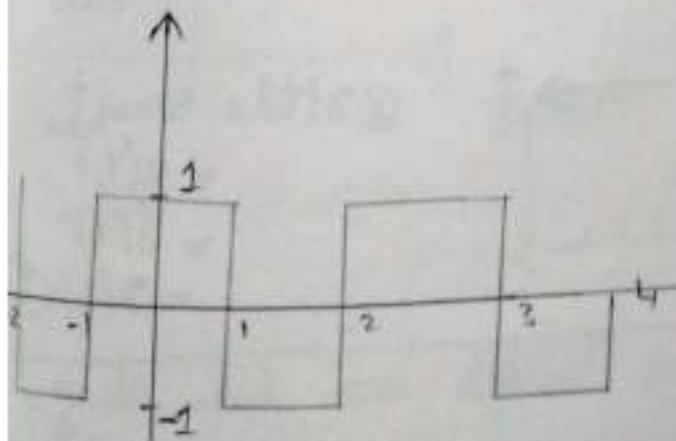
Continuous time analog
signal



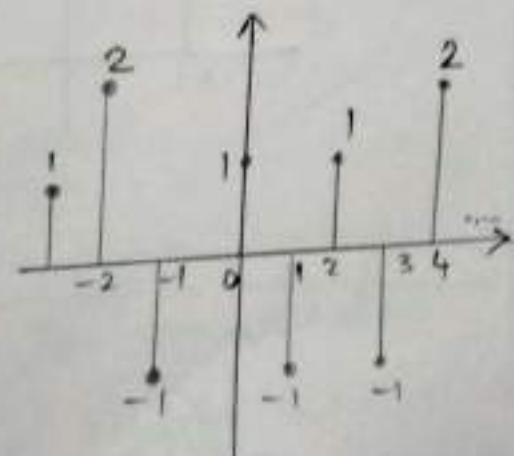
Discrete time
analog signal

Analog signal - A signal which can take a continuum (all real values) range of values is called as an analog signal.

The first two figures are analog signals.



Continuous time
digital signal.



Discrete time
digital signal.

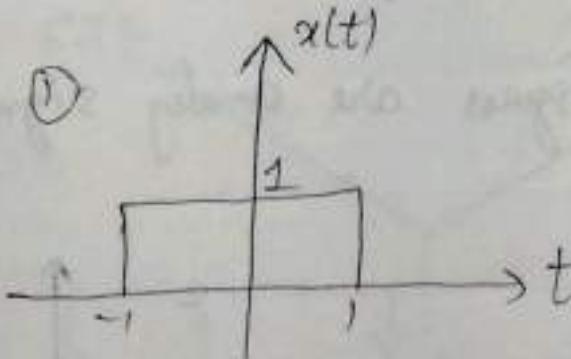
Digital signals are signals which can take only a finite set of values in its amplitude

2) Deterministic and non-deterministic signals

Signal whose value is certain at any instant of time is called deterministic
(or)

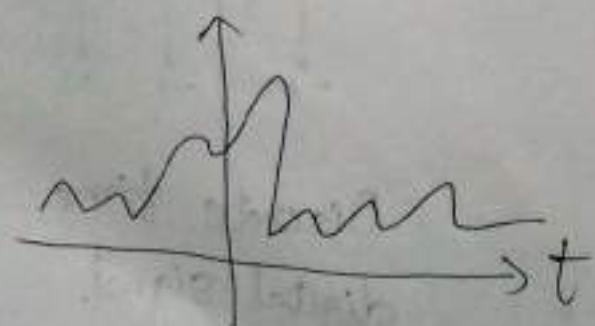
Signal whose value can be determined or predicted at any instant of time are called as deterministic signals

Eg:- ①



$$\begin{aligned} x(t) &= 1, -1 \leq t \leq 1 \\ &= 0, \text{else} \end{aligned}$$

$$\begin{aligned} ② x(t) &= \sin \omega_0 t \\ &= u(t) \\ &= \delta(t) \\ &= g(t) = \frac{1}{2} u(t) \end{aligned}$$



Non-deterministic signal

Signals whose values are uncertain or cannot be predicted or determined at any instant of time are called as non-deterministic signals.

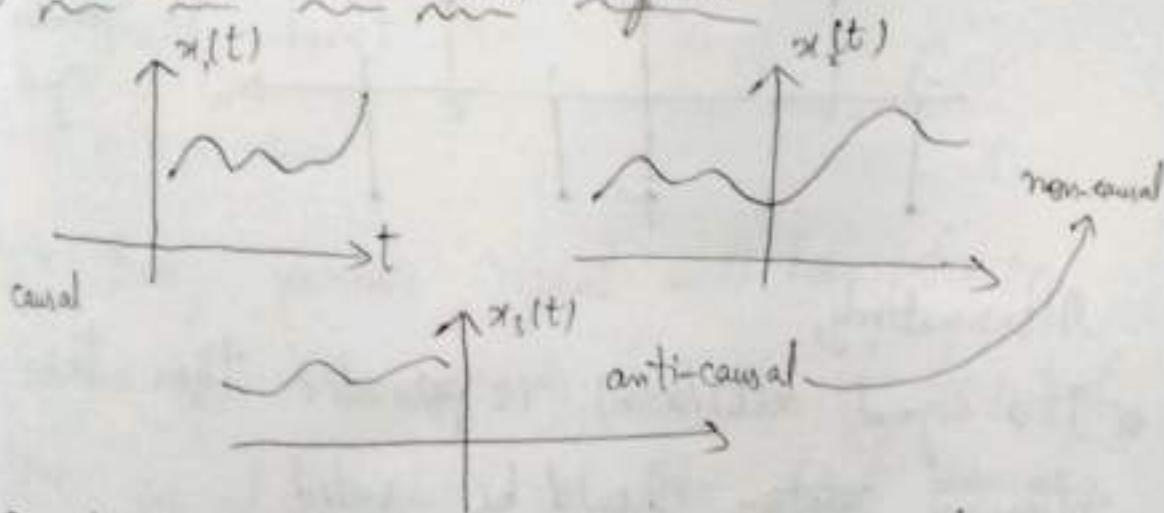
Since the nature of these signals is random, they are also called as random signals.

Eg:- 1) The noise present in T.V and audio receivers

2) ECG and EEG

3)

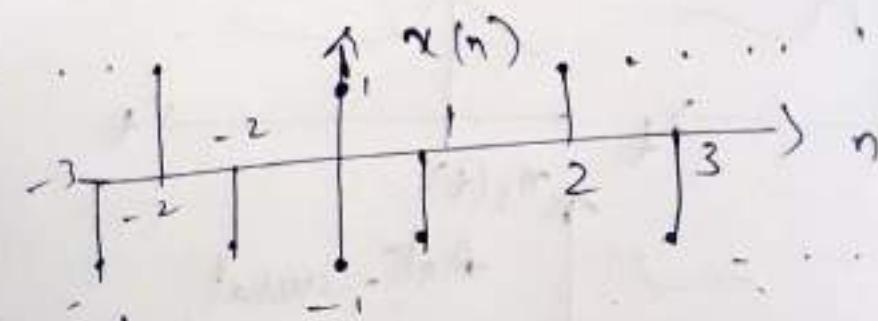
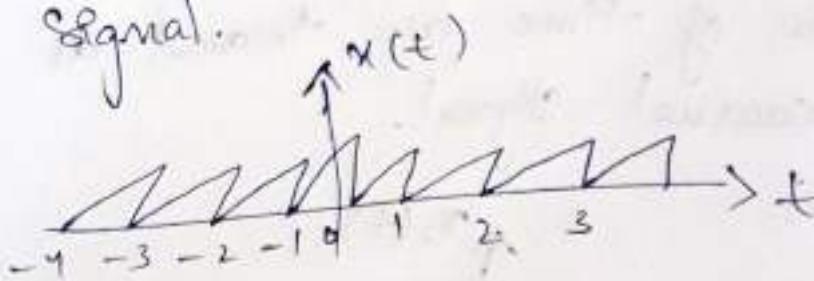
3) Causal and non-causal Signals



Signals which are defined only for the duration of time are called as causal signals whereas signals which have values defined for -ve duration of time are termed as non-causal signals. $u(t), r(t)$ are causal signals. $\sin \omega t, \operatorname{sgn}(t)$ are non-causal.

* Periodic & Aperiodic Signals:-

→ Signals which have a particular pattern repeating at regular intervals of time are called as periodic signals. Otherwise called Aperiodic signal.



Alternatively,

If a signal $x(t)$ / $x(n)$ is periodic
then the following condition should
be valid.

$$x(t \pm T) = x(t)$$

$$x(n \pm N) = x(n)$$

T, N - Time periods of
CIS & DTS
respectively.

Note:- For a signal to be periodic, the necessary condition is, it should be ever lasting. That is it should exists for $-\infty \leq t \leq \infty$

$$-\infty \leq n \leq \infty$$

Note:- Every periodic signal will have infinite number of time periods, but one fundamental time period, which is the least possible time period. Denoted by letter T_0 , ω_0 for CTS, DTS respectively.

* Check the periodicity & determine it if any:-

$$1) x(t) = \sin \omega_0 t, T = \frac{2\pi}{\omega_0}$$

$$x(t+T_0) = x(t)$$

$$\sin \omega_0 (t + T_0) = \sin \omega_0 t$$

$$\sin (\omega_0 t + \omega_0 T_0) = \sin \omega_0 t$$

$$\omega_0 T_0 = m(2\pi)$$

$$m=1, T = T_0$$

$$* * \left[T_0 = \frac{2\pi}{\omega_0} \right] * *$$

$$2) x(t) = \sin 2t$$

$$\stackrel{\text{sol}}{=} x(t + T_0) = x(t)$$

$$\sin 2(t + T_0) = \sin 2t$$

$$\sin 2t + 2\pi$$

$$\omega_0 = 2 \quad T_0 = 2\pi/\omega_0$$

$$T_0 = \frac{2\pi}{2}$$

$$\boxed{T_0 = \pi}$$

$$3) x(t) = \sin \sqrt{2}t$$

$$\stackrel{\text{sol}}{=} \omega_0 = \sqrt{2}$$

$$T_0 = \frac{2\pi}{\sqrt{2}}$$

$$\boxed{T_0 = \sqrt{2}\pi}$$

Note:- All continuous sinusoids are periodic

$$4) x(n) = \sin \omega_0 n$$

$$\stackrel{\text{sol}}{=} x(n+N) = x(n)$$

$$\sin \omega_0(n+N) = \sin \omega_0 n$$

$$\sin(n\omega_0 + N\omega_0) = \sin \omega_0 n$$

$$\omega_0 N = k(2\pi)$$

$$\left[N = \frac{2\pi k}{\omega_0} \right]$$

$$\left[\frac{N}{k} = \frac{2\pi}{\omega_0} \right]$$

→ The above condition states that for a discrete time signal to be periodic, the ratio π/ω_0 must be a rational number.

→ Check the periodicity $x(n) = \sin \omega_0 n$

i) $x(n) = \sin 2n$

Sol $\omega_0 = 2$

$$\frac{2\pi}{\omega_0} = \frac{2\pi}{2}$$

$$= \pi \text{ (rational)}$$

It's Aperiodic
 $N_0 = \infty$

2) $x(n) = \sin 2\pi n$

$$\omega_0 = 2\pi$$

$$\frac{2\pi}{\omega_0} = \frac{2\pi}{2\pi} = 1 \text{ Periodic}$$

$$N = \frac{2\pi k}{\omega_0}, k = 1, 2, 3, \dots$$

$$N = \frac{2\pi}{2\pi} k$$

$$\boxed{\begin{aligned} N &= k \\ N_0 &= 1 \end{aligned}}$$

3) Given: - $\sin \frac{2\pi}{3} n$

$$\omega_0 = \frac{2\pi}{3}$$

$$\Rightarrow \frac{2\pi}{\omega_0} = \sqrt{\frac{6\pi}{3\pi/3}} = \frac{3\sqrt{3}}{2}$$

$$\frac{2\pi}{\omega_0} = \frac{2\pi}{5\pi/3} = \frac{6}{5}$$

Periodic

$$N = \frac{2\pi}{\omega_0} k, \text{ where } k \dots$$

$$N = \frac{2\pi}{5\pi/3} k$$

$$N = \frac{6}{5} k, k = 5, 10, 15 \dots$$

$$N_0 = \frac{6}{5}(5) = 6$$

$$\boxed{N_0 = 6}$$

Note: - The Discrete time signals may or may not be periodic.

$$n(t) = e^{2/3jt}$$

* i) $\omega_0 \cdot k \cdot T \quad n(t) = e^{\omega_0 jt}$

$$\omega_0 = 2/3$$

$$T_0 = \frac{2\pi}{\omega_0} = \frac{2\pi}{2/3} = 3\pi$$

$$(T_0 = 3\pi)$$

(xx)

$$e^{j\omega_0 t} = \cos \omega_0 t + j \sin \omega_0 t$$

$$e^{2/3jt} = \cos \frac{2}{3}t + j \sin \frac{2}{3}t$$

$$= \underset{3\pi}{\downarrow} \dots + \underset{3\pi}{\downarrow}$$

$$\Rightarrow 1.c.m (3\pi, 3\pi)$$

$$(T_0 = 3\pi)$$

Q) $n(n) = e^{2\pi j n}$

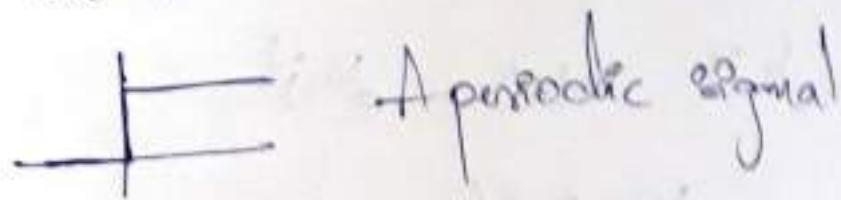
$$\omega_0 = 2\pi$$

$$\frac{2\pi}{\omega_0} = \frac{2\pi}{2\pi} = 1$$

$$N = \frac{2\pi k}{\omega_0} = \frac{2\pi}{2\pi} k$$

$$N = k, k = 1, 2, 3, \dots$$

$$(N_0 = 1),$$

3) $v(t)$ 

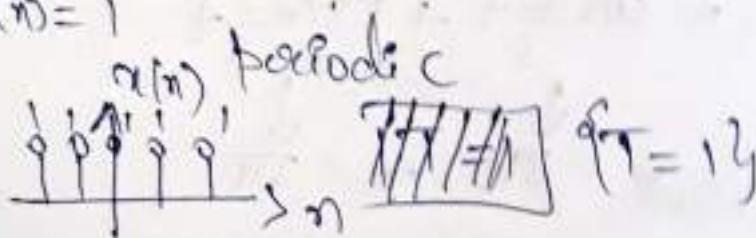
$\theta(t) \rightarrow$ A periodic signal

constant signal

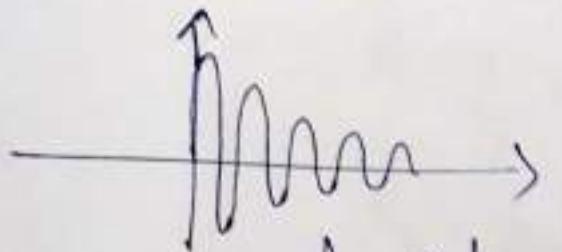


but discrete time constant signal is

$$\rightarrow a(n)=1$$



$$4) v(t) = e^{j\omega_0 t} u(t)$$



Aperiodic
($T_0 = \infty$)

5) $x(t) = e^{-(2+3j)t}$

Sol Given: - $x(t) = e^{-2t} \cdot e^{-3jt}$

Aperiodic signal \swarrow Periodic \searrow

\therefore Aperiodic signal.

6) $x(t) = e^{-2+3j} t$

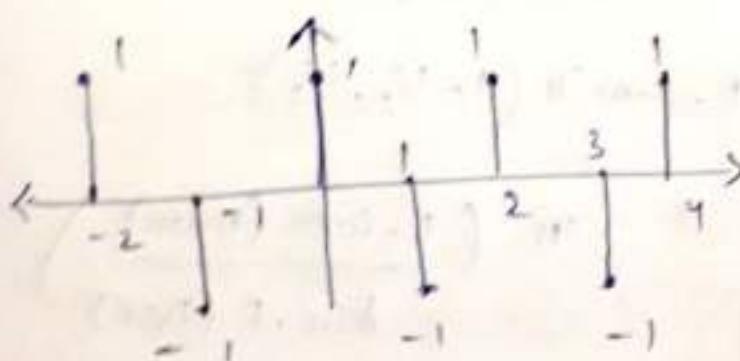
Sol Given: - $x(t) = \underbrace{e^{-2}}_{\text{Amplitude}} \cdot \underbrace{e^{3jt}}_{\text{Periodic}}$

Amplitude Periodic

\therefore 2h periodic

$$T_0 = \frac{2\pi}{\omega_0} = \frac{2\pi}{3}$$

$$\rightarrow x(n) = (-1)^n$$



Periodic $\Rightarrow \boxed{N=2}$

$$\rightarrow x(t) = \sin(2\pi \cdot 128\pi t + 38\pi t)$$

$$\Rightarrow S_1 \tau_1 = \frac{2\pi}{2}, \quad \tau_1 = \frac{2\pi}{1} \quad \tau_3 = \frac{2\pi}{6}$$

$$= \pi \quad = \pi/2 \quad = \pi/3$$

S₂) If the ratios of time-periods is rational then the signal is periodic otherwise Aperiodic.

$$\frac{\tau_2}{\tau_1}, \frac{\tau_3}{\tau_1}$$

$$= 1/2 \quad \text{periodic}$$

$$= 0.5$$

S₃) L.C.M (τ_1, τ_2, τ_3)

$$\text{L.C.M } (\pi, \frac{\pi}{2}, \frac{\pi}{3})$$

$$\text{L.C.M } \pi (1, 1/2, 1/3)$$

$$\pi \left(\frac{\text{L.C.M (num)}}{\text{H.C.F (Den)}} \right)$$

$$\frac{\pi (\text{L.C.M } (1, 1, 1))}{\text{H.C.F } (1, 2, 3)}$$

$$= \frac{\pi \times 1}{1} = \pi.$$

$$x(t) = 28\sin \frac{2\pi}{3}t + 3\sin 2t + 4\sin \pi t$$

(2)

$$s_1) \quad T_1 = \frac{2}{3}, \quad T_2 = \pi, \quad T_3 = \frac{2\pi}{3}$$

$$s_2) \quad \frac{T_1}{T_2} = \frac{2/3}{\pi} = \frac{2}{3\pi} \text{ irrational}$$

$$\frac{T_3}{T_2} = \frac{\pi}{2\pi/3} = \frac{3}{2} = \text{rational}$$

Aperiodic

$$s_3) \quad T = \infty$$

Some of the Continuous-time sinusoid
are may or may not be periodic

$$③ \quad x(n) = 2\cos(6\pi n + 90^\circ) + 3\cos(2\pi n + 15^\circ) \\ + \sin 8\pi n$$

so

$$N_1 = \frac{2\pi}{6\pi} k$$

$$= \frac{2\pi}{6\pi} k$$

$$N_1 = k/3 \quad (k=3, 6, 9, \dots)$$

$$\boxed{N_1 = 1} \quad (k=3)$$

$$N_2 = \frac{2\pi}{2\pi} k$$

$$N_2 = \frac{2\pi}{2\pi} k = k$$

$$\boxed{N_2 = 1} \quad \underbrace{k=1, 2, 3, \dots}_{k \in \mathbb{Z}}$$

$$N_3 = \frac{2\pi}{8\pi} k$$

$$= \frac{k}{4} \quad (k=4, 8, \dots)$$

$$\boxed{N_3 = 1} \quad \boxed{k \in \mathbb{Z}}$$

$$S2) \frac{N_2}{N_1} = 1 \quad \frac{N_3}{N_1} = 1 \\ = \text{rational} \quad = \text{rational}$$

$$S3) N = \text{L.C.M}(N_1, N_2, N_3)$$

$$= \text{L.C.M}(1, 1, 1)$$

$$\boxed{N=1} = 1$$

$$P3) x(t) = 8\sin 2t \cos 3t$$

$$\underline{\text{Sol}} \quad x(t) = \frac{1}{2}[8\sin 2t \cos 3t]$$

$$x(t) = \frac{1}{2} [8\sin 5t - 8\sin t]$$

$$x(t) = \frac{1}{2} \sin 5t - \frac{1}{2} \sin t$$

$$S2) \quad \tau_1 = \frac{2\pi}{5} \quad T_2 = \frac{2\pi}{1}$$

$$S3) \quad \frac{\tau_1}{T_2} = \frac{1}{5} = \text{rational} = 0.2$$

\therefore It's periodic

$$S4) \quad \text{L.C.M}(\tau_1, T_2)$$

$$\text{L.C.M}\left(\frac{2\pi}{5}, 2\pi\right)$$

$$\text{L.C.M. } 2\pi \left(\frac{1}{5}, 1\right)$$

$$\text{Qn. L.C.M} \left(\frac{1}{3}, 1 \right) = 2\pi \frac{\text{L.C.M}(1,1)}{\text{H.C.D}(5,1)}$$

$$= \frac{1}{1} \times 2\pi$$

$\boxed{T = 2\pi}$

(Pb) $x(t) = \cos^2 ut$

$$x(t) = \frac{1 + \cos 8t}{2}$$

$$x(t) = \frac{1}{2} + \frac{1}{2} \cos 8t$$

$$\omega_0 = 8$$

$$\therefore T = \frac{2\pi}{\omega_0}$$

$$T = \frac{2\pi}{8} = \frac{\pi}{4}$$

$\boxed{T = \frac{\pi}{4}}$

(Pb) $x(n) = e^{j6\pi n} + e^{j2\pi n/3}$

$\boxed{N = \frac{2\pi}{\omega_0} K}$

S1) $\omega_1 = 6\pi$

$$N = \frac{2\pi}{6\pi} K$$

S2) $\omega_2 = 2\pi/3$

$$N_2 = \frac{2\pi}{2\pi/3} K$$

$$= 3K.$$

$\boxed{N = 1}$ $\boxed{N_2 = 3}$

S2) $N_1/N_2 = 1/3 = 0.333$

S3) $\omega = \text{L.C.M}(1,3) = 3$

5*) Energy & Power Signal :-

(W.H)

→ Energy of any arbitrary signal can be expressed using the following formula.

$$E = P \times T$$

$$E = \int_{-\infty}^{\infty} |x(t)|^2 dt$$

→ For a DIS the energy can be formulated as follows

$$E = \sum_{-\infty}^{\infty} |x(n)|^2$$

→ The power of a signal can be formulated using the following two.

CTS Expression :-

$$P_{avg} = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} |x(t)|^2 dt$$

DIS

$$P_{avg} = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^N |x(n)|^2$$

* Determine whether the following signals are energy or non-energy signals. If the signal is finite, the power.

$$\textcircled{1} \quad x(t) = e^{-at} v(t)$$

sol Given: $x(t) = e^{-at} v(t)$

$$1) \quad E = \int_{-\infty}^{\infty} |x(t)|^2 dt$$

$$E = \int_{-\infty}^{\infty} |e^{-at} \cdot v(t)|^2 dt$$

$$\Rightarrow E = \int_{-\infty}^0 |e^{-at} \cdot 0|^2 dt + \int_0^{\infty} |e^{-at} \cdot 1|^2 dt$$

$$\Rightarrow E = \int_0^{\infty} e^{2at} dt$$

$$\Rightarrow E = \left[\frac{e^{2at}}{-2a} \right]_0^{\infty}$$

$$\Rightarrow E = -\frac{1}{2a} \left[e^{2a(\infty)} - e^{2a(0)} \right]$$

$$\Rightarrow E = -\frac{1}{2a} [0 - 1]$$

$$\Rightarrow \boxed{E = \frac{1}{2a}}$$

γ -A 35

$\gamma/2$

$$2) P_{avg} = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-\gamma/2}^{\gamma/2} |x(t)|^2 dt$$

$$\Rightarrow P = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-\gamma/2}^{\gamma/2} |\bar{e}^{at} v(t)|^2 dt$$

$$\Rightarrow P = \lim_{T \rightarrow \infty} \frac{1}{T} \int_0^{\gamma/2} \bar{e}^{2at} dt$$

$$\Rightarrow P = \lim_{T \rightarrow \infty} \frac{1}{T} \left[\frac{\bar{e}^{2at}}{-2a} \right]_0^{\gamma/2}$$

$$\Rightarrow P = \lim_{T \rightarrow \infty} \frac{1}{T} \left[\frac{1}{-2a} \left[\frac{\bar{e}^{2a\gamma/2}}{2} - \bar{e}^0 \right] \right]$$

$$\Rightarrow P = \lim_{T \rightarrow \infty} \frac{1}{T} \left[\frac{1}{-2a} [\bar{e}^{a\gamma} - 1] \right]$$

$$\Rightarrow P = \lim_{T \rightarrow \infty} \frac{1}{-2aT} [\bar{e}^{aT} - 1]$$

$$\Rightarrow P = \frac{1}{-2a(\infty)} [\bar{e}^\infty - 1]$$

$$\Rightarrow P = -\left(\frac{1}{\infty}\right)$$

$$\Rightarrow \boxed{P = 0}$$

\therefore The signal is an energy signal

Energy Signal :-

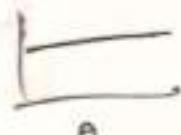
A signal which has a finite Energy is called as an Energy signal.

i.e if $E_{x(t)}, 0 \leq E_{x(t)} \leq \infty$

then $x(t)$ is an Energy signal

Note:- Power of Energy signal is always = 0

$$\text{Pb) } x(t) = u(t)$$



$$\text{Sol Given: } x(t) = u(t)$$

$$E = \int_{-\infty}^{\infty} |x(t)|^2 dt$$

$$\Rightarrow E = \int_{-\infty}^0 0 dt + \int_0^{\infty} 1 \cdot dt$$

$$\Rightarrow E = [t]_0^\infty$$

$$\Rightarrow E = [\infty - 0] = \infty$$

$$\boxed{E = \infty}$$

$$\begin{aligned} P &= \lim_{T \rightarrow \infty} \frac{1}{T} \int_0^T |x(t)|^2 dt = \lim_{T \rightarrow \infty} \frac{1}{T} \int_0^T 1 \cdot dt \\ &= \lim_{T \rightarrow \infty} \frac{1}{T} [t]_0^T \end{aligned}$$

$$\boxed{P=1} \therefore \text{Its Power signal} = \lim_{T \rightarrow \infty} 1 = 1$$

→ Power Signal:-

A signal which has finite power is called power signal i.e.

If $P_{x(t)}, 0 < P_{x(t)} < \infty$, then $x(t)$ is a power signal.

Note: The energy of a power signal is always infinite.

$$E_{x(t)} = \int_{-\infty}^{\infty} |x(t)|^2 dt$$

$$E_{x(n)} = \sum_{n=-\infty}^{\infty} |x(n)|^2$$

$$P_{avg\ x(t)} = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T}^{T} |x(t)|^2 dt$$

$$P_{avg\ x(n)} = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^{N} |x(n)|^2$$

3) $x(t) = A \cos(\omega_0 t + \theta)$

Sol:- $E = \int_{-\infty}^{\infty} |A \cos(\omega_0 t + \theta)|^2 dt = A^2 \int_{-\infty}^{\infty} \cos^2(\omega_0 t + \theta) dt$

$$= A^2 \int_{-\infty}^{\infty} \frac{1 + \cos 2(\omega_0 t + \theta)}{2} dt$$

$$E = A^2 \int_{-\infty}^{\infty} \frac{1}{2} dt + A^2 \int_{-\infty}^{\infty} \cos 2(\omega_0 t + \theta) dt$$

$$= \infty + "$$

$$= \infty$$

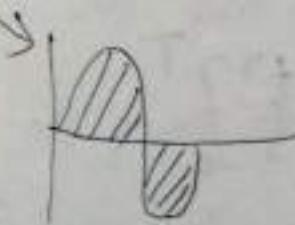
$$P = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T}^{T} |A \cos(\omega_0 t + \theta)|^2 dt$$

$$= \lim_{T \rightarrow \infty} \frac{A^2}{T} \int_{-T}^{T} \frac{1 + \cos 2(\omega_0 t + \theta)}{2} dt$$

$$= \lim_{T \rightarrow \infty} \frac{A^2}{2T} \left[\int_{-T}^{T} 1 dt + \int_{-T}^{T} \cos 2(\omega_0 t + \theta) dt \right]$$

$$= \lim_{T \rightarrow \infty} \frac{A^2}{2T} \left[[t]_{-T}^T + (\sin 2(\omega_0 t + \theta))_{-T}^T \right]$$

$$= \lim_{T \rightarrow \infty} \frac{A^2}{2T} [T + 0] = \frac{A^2}{2} \quad \therefore P = \boxed{\frac{A^2}{2}}$$



$$4.) x(t) = \sin t$$

Sol. - $x(t) = t \cdot u(t)$ or $t, t \geq 0$
 $0, t < 0$

$$\begin{aligned} E &= \int_{-\infty}^{\infty} |x(t)|^2 dt = \int_{-\infty}^{\infty} |t \cdot u(t)|^2 dt \\ &= \int_0^{\infty} |t|^2 dt + \int_{-\infty}^0 |t \cdot 0|^2 dt \\ &= \left[\frac{t^2}{2} \right]_0^{\infty} = \infty \end{aligned}$$

$$\therefore \boxed{E = \infty}$$

$$P = \lim_{T \rightarrow \infty} \frac{1}{T} \int_0^T |x(t)|^2 dt$$

$$= \lim_{T \rightarrow \infty} \frac{1}{T} \int_0^T |t \cdot u(t)|^2 dt$$

$$= \lim_{T \rightarrow \infty} \frac{1}{T} \int_0^T |t \cdot t| dt$$

$$= \lim_{T \rightarrow \infty} \frac{1}{T} \int_0^T |t|^2 dt = \lim_{T \rightarrow \infty} \frac{1}{T} \left[\frac{t^3}{3} \right]_0^T$$

$$= \lim_{T \rightarrow \infty} \frac{T^3}{3T} = \infty$$

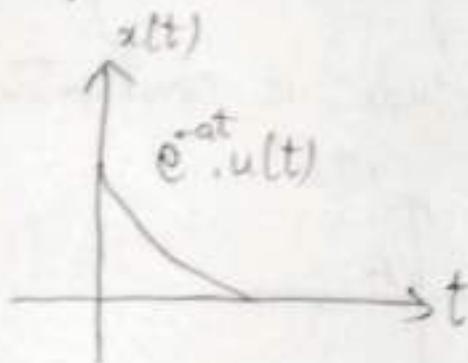
$$\therefore \boxed{P = \infty}$$

Therefore, neither energy nor power signal

Notes

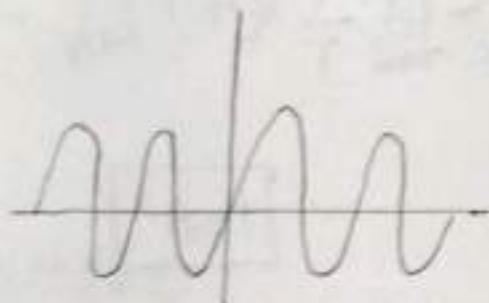
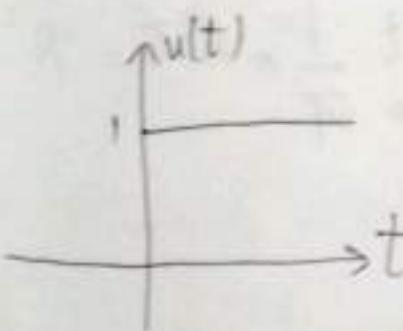
- ① For energy signals as $t \rightarrow \infty$, Amplitude $\rightarrow 0$

$$A \rightarrow 0$$



Same for DTS, i.e. as $n \rightarrow \infty$, $A \rightarrow 0$

- ② As $t \rightarrow \infty$, if amplitude remains constant
then such signals are always power signals

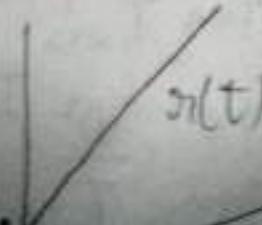


All periodic signals are power signals but
all power signals are not periodic signals.

- ③ If for any signal as $t \rightarrow \infty$, if $A \rightarrow \infty$
then such signals are neither energy nor power

signal

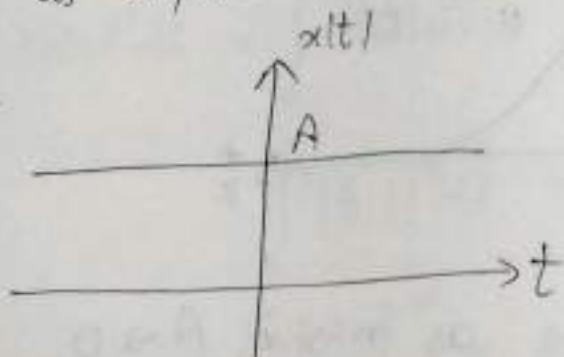
Eg:- Ramp signal
parabolic signal



Pb) $x(t) = A$

Sol:- Just by looking we can say it is a power signal as amplitude is constant.

Method:-



$$P = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-\infty}^{\infty} |x(t)|^2 dt = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} |A|^2 dt$$

$$= \lim_{T \rightarrow \infty} \frac{1}{T} \times [A^2 T]_{-T/2}^{T/2} = \lim_{T \rightarrow \infty} \frac{1}{T} \times A^2 T = A^2$$

$$\therefore \boxed{P = A^2}$$

∴ Power of D.C signal is A^2

Pb) $x(t) = A e^{j(\omega t + \phi)}$

Sol:- It is a power signal

$$P = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-\infty}^{\infty} |A e^{j(\omega t + \phi)}|^2 dt$$

$$= \lim_{T \rightarrow \infty} \frac{A^2}{T} \int_0^T 1 dt = \lim_{T \rightarrow \infty} \frac{A^2}{T} \times T = A^2$$

Q) $x(t) = e^{2t} \cdot u(-t)$

as $t \rightarrow \infty$

$$|x(t)| \rightarrow 0$$

It is an
energy signal

$$\begin{aligned} e^{\omega} u(-\omega) \\ = 0 \cdot e^{\omega} \\ = 0 \end{aligned}$$

$$E = \int_{-\infty}^{\infty} |e^{2t} u(-t)|^2 dt$$

$$\begin{aligned} u(-t) &= 1, t < 0 \\ &= 0, t \geq 0 \end{aligned}$$

$$= \int_{-\infty}^0 |e^{2t} \cdot 1|^2 dt + \int_0^{\infty} |e^{2t} \cdot 0|^2 dt$$

$$= \int_{-\infty}^0 e^{4t} dt = \left[\frac{e^{4t}}{4} \right]_{-\infty}^0 = \frac{1}{4} (1 - 0) = \frac{1}{4}$$

Q) $x(t) = \tilde{e}^{-2t} \cdot u(t-1)$

[$\because u(t-1)$]

Sol:- as $t \rightarrow \infty$, $|x(t)| = 0$

\therefore Energy signal

$$E = \int_{-\infty}^{\infty} |\tilde{e}^{-2t} \cdot u(t-1)|^2 dt \quad u(t-1) = 1, t-1 \geq 0$$

$$= \int_{-\infty}^1 |\tilde{e}^{-2t} \cdot 1|^2 dt \quad t \geq 1$$

$$= 0, t < 1$$

$$\begin{aligned} E &= \int_{-\infty}^1 |\tilde{e}^{-2t} \cdot 0|^2 dt + \int_1^{\infty} |\tilde{e}^{-2t} \cdot 1|^2 dt \\ &= \int_1^{\infty} \tilde{e}^{-4t} dt = \left[\frac{\tilde{e}^{-4t}}{-4} \right]_1^{\infty} \\ &= -\frac{1}{4} (\tilde{e}^{-4} - \tilde{e}^{-4}) = \frac{\tilde{e}^{-4}}{4} = \frac{1}{4e^4} \end{aligned}$$

Pb) $x(t) = j e^{j\pi t}$

Sol:- $j e^{j\pi t} = e^{j\pi_2} \cdot e^{j\pi t}$
 $= e^{j(\pi t + \pi_2)}$

$$\begin{aligned} e^{j\pi t} &= \cos \pi_2 + j \sin \pi_2 \\ &= 0 + j(1) \end{aligned}$$

It's a power signal.

Pb) $x(t) = \begin{cases} t-2 & -2 \leq t < 0 \\ 2-t & 0 \leq t < 2 \\ 0 & \text{else} \end{cases}$

* All finite duration signals are energy signals

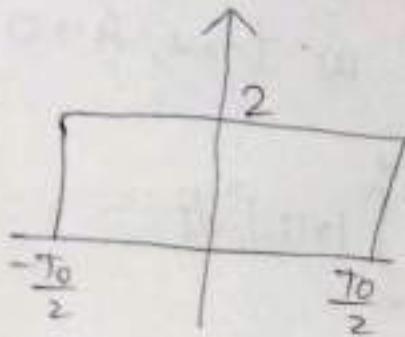
Sol:- $E = \int_{-\infty}^{\infty} |x(t)|^2 dt = \int_{-2}^{0} (t-2)^2 dt + \int_{0}^{2} (2-t)^2 dt$

Pb) $x(t) = 2 \operatorname{rect}\left(\frac{t}{T_0}\right)$

Sol:- $E = \int_{-\infty}^{\infty} |x(t)|^2 dt$

$$= \int_{-T_0/2}^{T_0/2} |2|^2 dt$$

$$= \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} 4 dt = 4 T_0$$



$$P = 0$$

Pb) $x(t) = \cos \omega_0 t \cdot \operatorname{rect}\left(\frac{t}{4}\right)$

Sol:- $E = \int_{-\infty}^{\infty} |x(t)|^2 dt$

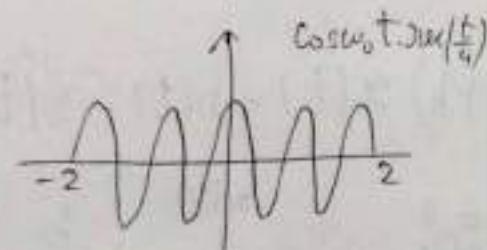
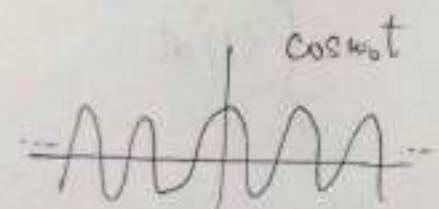
$$= \int_{-2}^2 |\cos \omega_0 t|^2 dt$$

$$= \int_{-2}^2 \frac{1 + \cos 2\omega_0 t}{2} dt$$

$$= \left[\frac{t}{2} \right]_{-2}^2 + \frac{1}{2} \int_{-2}^2 \cos 2\omega_0 t dt$$

$$= \frac{1}{2} + \frac{1}{2} \left[\frac{\sin 2\omega_0 t}{2\omega_0} \right]_{-2}^2 \approx 2 + \frac{1}{4\omega_0} (2 \sin 4\omega_0)$$

$$= 2 + \frac{\sin 4\omega_0}{2\omega_0}$$

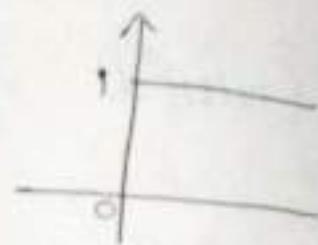


$$\text{Pb) } x(t) = u(t) - u(t-5)$$

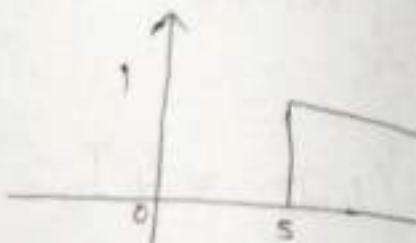
Sol. as $t \rightarrow \infty, A \rightarrow 0$

$$E = \int_{-\infty}^{\infty} |x(t)|^2 dt$$

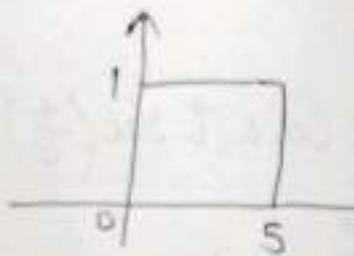
Energy
signal



$$= \int_{-\infty}^{\infty} (u(t) - u(t-5))^2 dt$$



$$= \int_0^5 (1)^2 dt$$

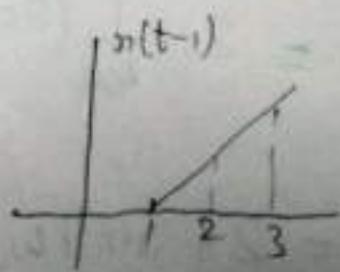
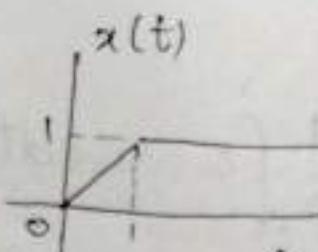
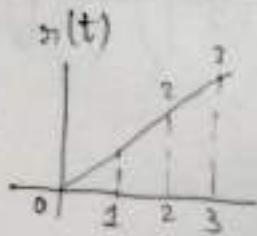


$$= \int_0^5 (1)^2 dt$$

Difference
of unit sig
is energy
signals

$$\text{Pb) } x(t) = r(t) - r(t-1)$$

Sol.



$$\begin{aligned} x(t) &= t, \quad 0 \leq t \leq 1 \\ &= 1, \quad t > 1 \end{aligned}$$

It's a power signal [as amplitude const]

$$P = \frac{1}{T} \int_{-\infty}^{\infty} |x(t)|^2 dt$$

$$= Lt \underset{T \rightarrow \infty}{\frac{1}{T}} \left[\int_0^T t^2 dt + \int_0^T u^2 dt \right]$$

$$= Lt \underset{T \rightarrow \infty}{\frac{1}{T}} \left[\left[\frac{t^3}{3} \right]_0^T + [t^2]_0^T \right]$$

$$= Lt \underset{T \rightarrow \infty}{\frac{1}{T}} \left[\frac{1}{3} T^3 + (T-1)T \right]$$

$$= 0 + Lt \underset{T \rightarrow \infty}{\frac{1}{T}} \frac{T-1}{T} = 0 + Lt \underset{T \rightarrow \infty}{\left(1 - \frac{1}{T}\right)}$$

$$= 1,$$

Pb) $x(n) = A \cdot u(n)$

Sol: as $n \rightarrow \infty$, Amp = A (const)

∴ Power signal

$$P = Lt \underset{N \rightarrow \infty}{\frac{1}{N+1}} \sum_{n=0}^{N-1} |x(n)|^2$$

$$= Lt \underset{N \rightarrow \infty}{\frac{1}{N+1}} \sum_{n=0}^{N-1} |A \cdot u(n)|^2$$

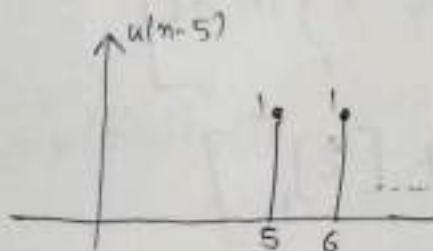
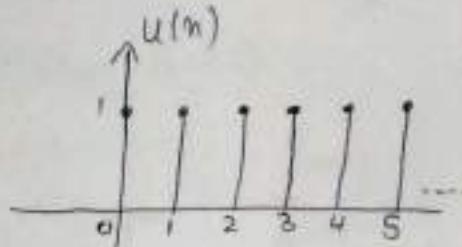
$$= Lt \underset{N \rightarrow \infty}{\frac{A^2}{N+1}} \sum_{n=0}^{N-1} |u(n)|^2 = Lt \underset{N \rightarrow \infty}{\frac{A^2}{N+1}} \left[1^2 + 1^2 + 1^2 + \dots + \underbrace{1^2}_{\text{term}} \right]$$

$$= Lt \underset{N \rightarrow \infty}{\frac{A^2 \times N+1}{N+1}} = A^2$$

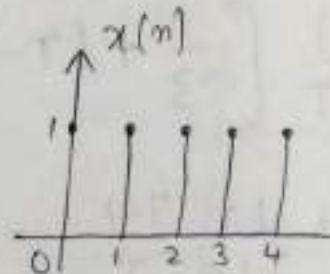
$$\boxed{F = \omega}$$

$$\text{Pb) } x(n) = u(n) - u(n-5)$$

Sol:-



$x(n)$ = Energy signal



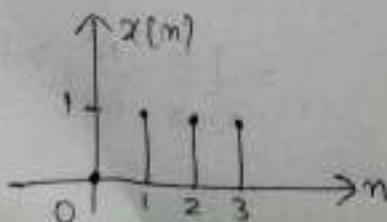
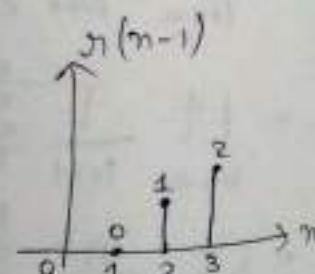
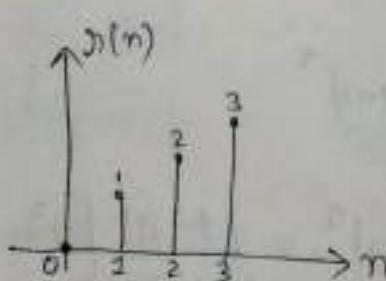
$$E = \sum_{n=0}^4 |x(n)|^2$$

$$= |x(0)|^2 + |x(1)|^2 + |x(2)|^2 + \dots + |x(4)|^2$$

$$\underset{\text{IMP}}{\star} = 1 + 1 + 1 + 1 + 1 = 5$$

$$\text{Pb) } x(n) = \gamma(n) - \gamma(n-1)$$

Sol:-



$$\begin{aligned}
 P &= \lim_{N \rightarrow \infty} \frac{1}{N+1} \sum_{n=0}^N |x(n)|^2 \\
 &= \lim_{N \rightarrow \infty} \frac{1}{N+1} \left[|x(0)|^2 + |x(1)|^2 + \dots + |x(N)|^2 \right] \\
 &= \lim_{N \rightarrow \infty} \frac{1}{N+1} [0 + 1 + 1 + 1 + \dots] \\
 &= \lim_{N \rightarrow \infty} \frac{N}{N+1} = 1 \\
 \therefore P &= 1
 \end{aligned}$$

Pb) $x(n) = \left(\frac{1}{2}\right)^n u(n)$

Sol:- s1) as $n \rightarrow \infty$, $|x(n)| = \left(\frac{1}{2}\right)^n u(\infty) = 0 \times 1 = 0$

Energy signal

s2) $E = \sum_{-\infty}^{\infty} |x(n)|^2 =$ $x(n) = \left(\frac{1}{2}\right)^n, n \geq 0$
 $= 0, \text{else}$

~~$E = \sum_{-\infty}^0 |0|^2 + \sum_0^{\infty} \left(\frac{1}{2}\right)^{2n}$~~

$$= \sum_0^{\infty} \left(\frac{1}{4}\right)^n$$

$$= 1 + \frac{1}{4} + \left(\frac{1}{4}\right)^2 + \left(\frac{1}{4}\right)^3 + \dots \infty$$

$$= \frac{1}{1 - 1/4} = \frac{4}{3}$$

$$\therefore E = 4/3$$

$$\text{Pb) } x(n) = 3^n \cdot u(-n-1)$$

$$\text{Sol: si) as } n \rightarrow \infty \quad |x(n)| = 3^\infty \cdot u(-\infty) \\ = 0$$

\therefore Energy signal

$$x(n) = 3^n \cdot 1, \quad -n-1 \geq 0 \Rightarrow n \leq -1 \\ = 3^n \cdot 0, \quad n > -1.$$

$$\boxed{\begin{aligned} x(n) &= 3^n, & n \leq -1 \\ &= 0, & n > -1 \end{aligned}}$$

$$E = \sum_{-\infty}^{\infty} |x(n)|^2 = \sum_{-\infty}^{-1} |3^n|^2 + \sum_{-1}^{\infty} |0|^2 \\ = \sum_{-\infty}^{-1} 3^{2n} = \sum_{1}^{\infty} 3^{-2n} \\ = \left\{ \left(\frac{1}{9} \right)^n \right\} = \frac{1}{9} + \frac{1}{9^2} + \frac{1}{9^3} + \dots$$

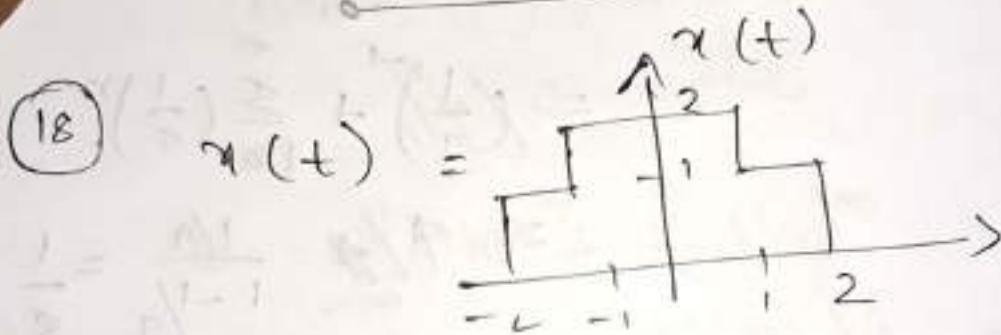
$$\left[\because S_\infty = \frac{a}{1-r} \right] \quad = \quad \frac{1/9}{1 - 1/9} = \frac{1}{8},$$

$$(17) \quad x(n) = n, \quad 0 \leq n \leq 4 \\ = 0, \text{ else}$$

Sol finite duration signal
 \Rightarrow energy signal

$$E = \sum_{n=0}^4 |x(n)|^2 \\ = \sum_{n=0}^4 n^2 = 0^2 + 1^2 + 2^2 + 3^2 + 4^2 \\ = 1 + 4 + 9 + 16 = 30$$

$$\boxed{E = 30}$$



$$E = \int_{-\infty}^{\infty} |x(t)|^2 dt$$

$$E = \int_{-2}^2 |x(t)|^2 dt \\ = \int_{-2}^{-1} 1^2 dt + \int_{-1}^1 2^2 dt + \int_1^2 1^2 dt \\ = [t]_{-2}^{-1} + [4t]_1^2 + [t]_1^2 \\ = 1 + 8 + 1 \\ = 10$$

06/02/2020

$$(9) \quad x(n) = A \sin(n\omega_0 + \theta)$$

Given:- $x(n) = A \sin(n\omega_0 + \theta)$
Sol \therefore It's Power signal.

$$\left\{ P = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^N |x(n)|^2 \right\}$$

$$\left\{ P = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=0}^{N-1} |x(n)|^2 \right\}$$

$$\left\{ P = \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{n=1}^N |x(n)|^2 \right\}$$

$$\left\{ P = \lim_{N \rightarrow \infty} \frac{1}{N+1} \sum_{n=-N/2}^{N/2} |x(n)|^2 \right\}$$

$$P = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^N |A^2 \sin^2(n\omega_0 + \theta)|$$

$$= \lim_{N \rightarrow \infty} \frac{A^2}{2N+1} \sum_{n=-N}^N \left(1 - \frac{\cos 2(n\omega_0 + \theta)}{2} \right)$$

$$= \lim_{N \rightarrow \infty} \frac{A^2}{2(2N+1)} \left[\sum_{n=-N}^N 1 - \sum_{n=-N}^N \cos 2(n\omega_0 + \theta) \right]$$

$$= \lim_{N \rightarrow \infty} \frac{A^2}{2(2N+1)} \sum_{n=-N}^N 1 - \frac{1}{2(2N+1)} \sum_{n=-N}^N \cos 2(n\omega_0 + \theta)$$

$$P = \lim_{N \rightarrow \infty} \frac{A^2}{2(2N+1)} (2N+1) - 0$$

$$\left\{ P = A^2/2 \right\}$$

$$20) \quad x(n) = \{ \underset{-2}{\overset{1}{1}}, \underset{1}{-2}, \underset{0}{1}, \underset{1}{2}, \underset{0}{3} \}.$$

This is discrete-time energy signal

$$\begin{aligned} E &= \sum_{-\infty}^{\infty} |x(n)|^2 \\ &= \sum_{-2}^{2} |x(n)|^2 \\ &= |x(-2)|^2 + |x(-1)|^2 + |x(0)|^2 \\ &\quad + |x(\cancel{1})|^2 + |x(2)|^2 \\ &= 1^2 + 4^2 + 1^2 + 2^2 + 3^2 \\ &= 1 + 4 + 1 + 4 + 9 \\ E &= 19 \text{ Units} \end{aligned}$$

$$21) \quad n(t) = 8 \sin t \cos 3t$$

Given:- $n(t) = 8 \sin t \cos 3t$

$$\begin{aligned} x(t) &= \frac{1}{2} (8 \sin t - 8 \sin 2t) \\ &= \frac{1}{2} \sin t - \frac{1}{2} \sin 2t \end{aligned}$$

We have solved

$$\begin{aligned} \text{if } A \sin(\omega t + \phi) &= A^2 / 2 \\ &= \frac{(1/2)^2}{2} + \frac{(-1/2)^2}{2} \\ &= 1/8 + 1/8 = 1/4 \end{aligned}$$

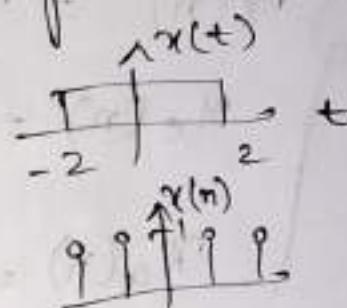
* Even & Odd Signals:-

- These class of signals is based upon the symmetry property of signals
- If $x(t)$ or $x(n)$ is said an even signal if the following condition is satisfied.

$$x(t) | x(n)$$

$$x(-t) = x(t)$$

$$x(-n) = x(n)$$

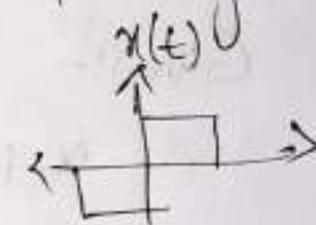


- A signal is said to be an odd signal if it satisfied the following condition.

$$x(t) | x(n)$$

$$x(-t) = -x(t)$$

$$x(-n) = -x(n)$$



Note:- A signal which is neither even nor odd can be expressed as a sum of even function & odd function as follows.

$$\left\{ \begin{array}{l} x(t) = x_e(t) + x_o(t) \\ x_e(t) = \frac{x(t) + x(-t)}{2} \\ x_o(t) = \frac{x(t) - x(-t)}{2} \end{array} \right.$$

$$\left\{ \begin{array}{l} x(n) = x_e(n) + x_o(n) \\ x_e(n) = \frac{x(n) + x(-n)}{2} \\ x_o(n) = \frac{x(n) - x(-n)}{2} \end{array} \right.$$

Find the even & odd parts of the following

① $x(t) = v(t)$ function.

Sol Given:- $x(t) = v(t)$

$$x_e(t) = \frac{v(t) + v(-t)}{2}$$

$$x_o(t) = \frac{v(t) - v(-t)}{2}$$

\therefore neither even nor odd

② $x(t) = t^2 + t + 1$

$$x(-t) = t^2 - t + 1 \neq x(t) \text{ not even}$$

$$\neq -x(t) \text{ not odd.}$$

\therefore neither even nor odd

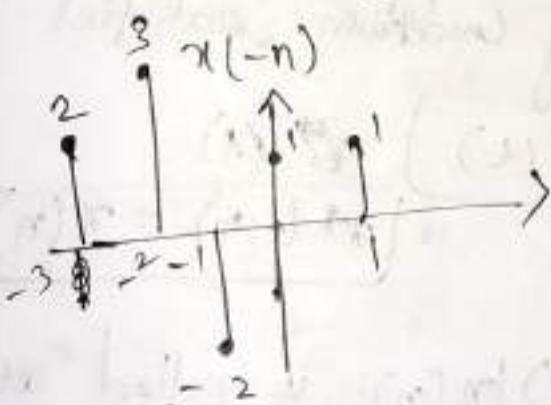
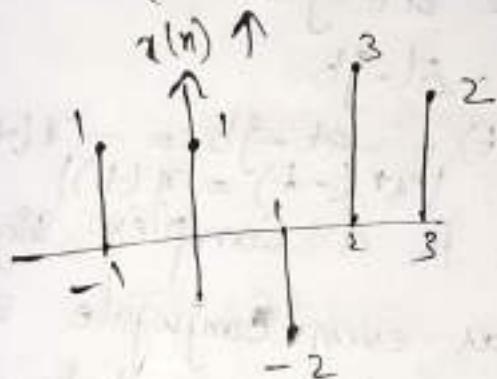
$$\mathbf{x}_e(t) = \frac{t^2 + t + 1 - (t^2 - t + 1)}{2}$$

$$\mathbf{x}_e(t) = t^2 + 1$$

$$\mathbf{x}_o(t) = \frac{t^2 + t + 1 - (t^2 - t + 1)}{2}$$

$$\mathbf{x}_o(t) = t$$

$$(3) \quad \mathbf{x}(n) = \{1, 1, -2, 3, 2\}$$



$$\mathbf{x}_e(n) = \left\{ 1, \frac{3}{2}, -\frac{1}{2}, 1, -\frac{1}{2}, \frac{3}{2}, 1 \right\}$$

$$\mathbf{x}_o(n) = \left\{ -1, -\frac{3}{2}, \frac{3}{2}, 0, -\frac{3}{2}, \frac{3}{2}, 1 \right\}$$

* If $x(t) | x(n)$ is a complex signal.

$$x(t) = a + jb t$$

$$x(-t) = a - jb t$$

$$x^*(-t) = a + jb t$$

$$x^*(-t) = x(t)$$

even conjugate symmetric

$$x(t) = \bar{a}t + \bar{jb}$$

$$x^*(t) = \bar{a}t - \bar{jb}$$

$$x^*(-t) = -\bar{a}t - \bar{jb} = -\bar{x}(t)$$

$$(x^*(-t)) = \bar{x}(t)$$

> If $x(t) | x(n)$ is a complex signal, then

it is called as even conjugate signal

If the following condition satisfies

$$\boxed{\begin{aligned} x^*(-t) &= x(t) \\ x^*(-n) &= x(n) \end{aligned}}$$

> A signal $x(t) | x(n)$ is called as odd conjugate signal if the following condition satisfies.

$$\left\{ \begin{array}{l} x^*(-t) = -x(t) \\ x^*(-n) = -x(n) \end{array} \right.$$

$$1) x(t) = e^{jt}$$

Given:- $x(t) = e^{jt}$

$$\begin{aligned}x(-t) &= e^{-jt} \\&= \cos(-t) + j\sin(-t) \\&= \cos t - j\sin t\end{aligned}$$

$$x^*(-t) = \cos t + j\sin t$$

$$x^*(-t) = x(t)$$

\therefore even conjugate signal

(001) $x^*(t) = e^{-jt}$

$$x^*(-t) = e^{jt}$$

$$x^*(-t) = x(t)$$

\therefore even conjugate signal

$$2) x(t) = j e^{jt}$$

Given $x^*(t) = -j e^{-jt}$ $(a+jb)^* = a-jb$

$$x^*(-t) = -j e^{jt} \quad (jb)^* = -jb$$

$$x^*(-t) = -x(t)$$

\therefore odd conjugate signal

$$\textcircled{3} \quad x(t) = x_{ec}(t) + x_{oc}(t)$$

If $x(+)|x(n)$ is neither even nor odd then it can be expressed as a sum of even conjugate function and odd conjugate function

C.T.S

$$x(t) = x_{ec}(t) + x_{oc}(t)$$

$$x_{ec}(t) = \frac{x(t) + x^*(-t)}{2}$$

$$x_{oc}(t) = \frac{x(t) - x^*(-t)}{2}$$

D.T.S

$$x(n) = x_{ec}(n) + x_{oc}(n)$$

$$x_{ec}(n) = \frac{x(n) + x^*(-n)}{2}$$

$$x_{oc}(n) = \frac{x(n) - x^*(-n)}{2}$$

$$3) \quad x(t) = a + jbt^2$$

$$\underline{x^*(t)} = a - jbt^2$$

$$x^*(-t) = a - jbt^2 \neq x(t)$$

$$\neq -x(t)$$

$$x_{ec} = \frac{a + jbt^2 + a - jbt^2}{2} = a$$

$$x_{oc} = \frac{a + jbt^2 - (a - jbt^2)}{2} = jbt^2$$

$$*) \begin{array}{ll} 0 + 0 = 0 & 0 \rightarrow \text{odd} \\ e + e = e & e \rightarrow \text{even} \\ e \cdot 0 = 0 & \end{array}$$

$$\frac{d}{dt}(e) = \text{odd}$$

$$\frac{d}{dt}(0) = \text{even}$$

→ Prove that sum of two odd function signals over time t is odd and sum of two even signals is even

Proof:- a) In General Consider two odd functions

$$\begin{aligned} n_1(t) &\text{ & } n_2(t) \text{ odd functions} \\ n_1(-t) &= -n_1(t) \\ n_2(-t) &= -n_2(t) \quad n(-t) \\ &= -x(t) \end{aligned}$$

$$\text{Let } x(t) = n_1(t) + n_2(t)$$

Apply time reversal on both side

$$\begin{aligned} x(-t) &= n_1(-t) + n_2(-t) \\ &= -n_1(t) - n_2(t) \\ &= - (n_1(t) + n_2(t)) \end{aligned}$$

$$x(-t) = -x(t)$$

∴ $x(t)$ is odd
∴ sum of two odd function is odd

b) two even

Consider

$n_1(t)$ & $n_2(t)$ be two even functions

$$n_1(-t) = n_1(t)$$

$$n_2(-t) = n_2(t)$$

Id $x(t) = n_1(t) + n_2(t)$

Time reversal on both sides.

$$x(-t) = n_1(-t) + n_2(-t)$$

$$x(-t) = n_1(t) + n_2(t)$$

$$x(-t) = x(t)$$

\therefore Sum of two even functions is even

c) prove that the product of even odd is odd function.

Id consider $n_1(t)$ as odd function
 $n_2(t)$ as even function

$$n_1(-t) = -n_1(t) \text{ by def}$$

$$n_2(-t) = n_2(t)$$

Id $x(t) = n_1(t) \cdot n_2(t)$

$$x(t) = n_1(t) \cdot n_2(t)$$

Time reversal on both side

$$x(-t) = n_1(-t) \cdot n_2(-t)$$

$$n_1(-t) = -n_1(t) \quad n_2(-t) = n_2(t)$$

$$x(-t) = -x(t)$$

\therefore Product of even & odd function
is odd

(d) Proof product of two even function is even

Ques Consider $n_1(t)$ & $n_2(t)$ be
even functions

$$n_1(-t) = n_1(t) \quad \text{by def.}$$

$$n_2(-t) = n_2(t)$$

$$\text{Let } x(t) = n_1(t) \cdot n_2(t)$$

Time reversal.

$$x(-t) = n_1(-t) \cdot n_2(-t)$$

$$n_1(-t) = n_1(t) \quad n_2(-t) = n_2(t)$$

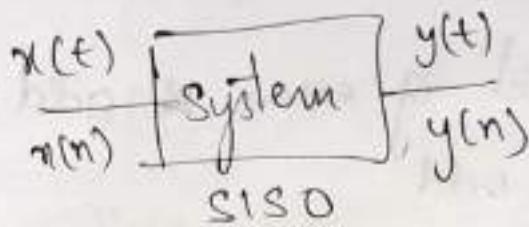
$$x(-t) = x(t)$$

\therefore even function.

* Systems:-

→ In General, A system can be mathematically modelled as follows

$$\left\{ \begin{array}{l} y(t) = T[x(t)] \\ \text{continuous time system} \end{array} \right.$$



$$y(t) = \int x(t) dt$$

$$y(n) = T[x(n)] \quad \text{discrete time syst.}$$

$$y(n) = x(n-1)$$

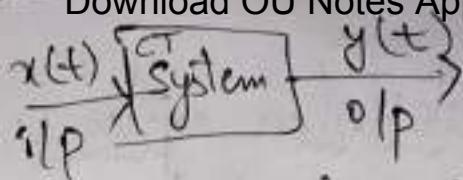
→ System can be broadly classified into two types

i) Continuous time System

ii) Discrete time System

i) Continuous time System:-

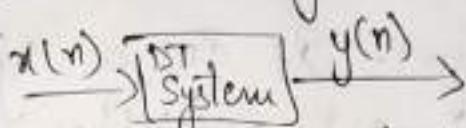
It is a system which transforms one continuous time signal into another of continuous time signal



$$y(t) = T[x(t)] + t$$

Eg:- Integrator :- $y(t) = \int_{-\infty}^t x(\tau) d\tau$
Amplifier, differentiator, filters

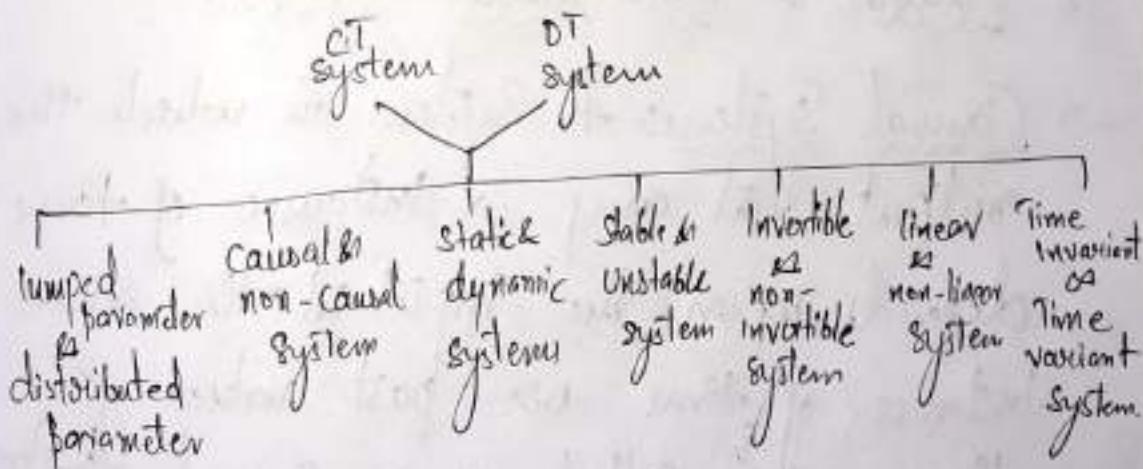
b) Discrete time signal :-



It is a system which transforms an S/I/P discrete time signal into O/I/P discrete time signal.

$$y(n) = T[x(n)]$$

Eg:- Digital computer, processor, register



1) Lumped & Distributed Systems:-

→ Lumped System:- A system in which the physical variables are just functions of time.

Eg:- Resistor, Inductor and Capacitor
(R.L.C)

→ Distributed System:- It is a system in which physical variables are functions of space and time.

Eg:- Optical fibre, Transmission line,
co-axial cable, wave guide
These systems are used at high frequencies

2) Causal & non-Causal Systems:-

→ Causal System:- A system in which the output at any instance of time depends upon the input at the same instance of time or past values of time and is called as a causal system.

Such systems are physically realizable.

eg:-

$$1) y(t) = x(t-2) \quad 2) y(t) = x(t)$$

$$t=0$$

$$y(0) = x(-2)$$

$$y(-1) = x(-3)$$

$$y(0) = x(0)$$

$$y(1) = x(1)$$

1) & 2) are causal signals.

o/p depends on
past i/p

o/p depends on
present i/p

→ Non-Causal Systems: - A System in which output at any instance of time depends upon the future values of the input or output is called as a non-causal system.

$$\text{Eq: } 1) y(t) = 2x(t+2)$$

$$y(0) = 2x(2)$$

non-causal system

$$y(-1) = 2x(1)$$

o/p depends on future i/p

$$2) y(t) = x(t) + x(t+2)$$

$$y(0) = x(0) + x(2)$$

* Investigate the Causality in the following Systems:-

$$1) y(t) = u(\sqrt{t})$$

Sol Test 1:- $y(2) = u(\sqrt{2}) = u(1.414)$

Test 2:- $y(1/2) = u(0.707)$

$$y(0.5) = u(0.707)$$

non-causal

$$2) y(t) = u(t^2)$$

Sol Test 1: $y(2) = u(u)$ non-causal

Test 2: $y(0.5) = u(0.25)$

causal

∴ Non-causal

$$3) y(t) = u(-t)$$

Sol $y(-1) = u(1)$

$$y(1) = u(-1)$$

non-causal

$$4) y(n) = u(n) u(n-2)$$

Sol Test 1: $y(1) = u(1) u(-1)$ causal

$$y(-1) = u(-1) u(-3)$$
 causal

∴ Its causal

5) $y(n) = n(n/2)$
sol $y(2) = n(2 \cdot 2)$
 $y(-2) = n(-2 \cdot 2)$
 $y(-2) = n(-1)$
 \therefore non-causal

6) $y(n) = n(\sqrt{n})$
sol test:- $y(4) = n(2)$
 $y(2) \neq$ causal

Note:- Don't take fractional values (< 1)
for discrete time system Test

3) Static & Dynamic System:-

Static System:- A system in which the output at any instance of time depends upon the input at the same instance of time is called as a static system.

Eg:- $y(t) = x(t)$
 $y(n) = n(n)$

A purely resistance network is static system.

static system are also called as memoryless system.

Dynamic System:- A system in which the output at any instance of time depends upon i/p and any other instance of time (future or past or both) is called as a Dynamic System

$$\text{eg } y(t) = x(t-2)$$

$$y(t) = x(t+2)$$

$$y(n) = x(n-2) + x(n+2)$$

Dynamic systems are memory system

Note:- 1) All static systems are causal, but all causal systems are not static

$$\text{Eg:- } y(t) = x(t) \quad y(t) = x(t-2)$$

2) All non-causal systems are dynamic

$y(n) = x(n-2) + x(n+2)$ but all dynamic systems are non-causal.

$$\text{Eg:- } y(t) = x(t-2)$$

Investigate the memory in following systems

$$(1) y(t) = x(\sqrt{t})$$

Sol Test 1:- $y(2) = x(1.414)$
Dynamic system

$$2) y(t) = x(t^2)$$

Sol Test 1:- $y(2) = x(4)$
Dynamic system

$$3) y(t) = x(-t)$$

Sol $y(1) = x(-1)$
Dynamic system

$$4) y(n) = x(n/2)$$

Sol $y(2) = x(1)$
Dynamic system.

$$5) y(t) = \int_{-\infty}^t z(\tau) d\tau$$

Sol Let $\int z(\tau) d\tau = x(\tau)$

$$y(t) = |x(\tau)| \Big|_{-\infty}^t$$

$$y(t) = x(t) - x(-\infty)$$

(Present) (Past)

Dynamic system.

Causal system

* Causality & memory:-

$$1) \quad y(n) = \sum_{k=0}^n x(k)$$

Sol Test 1:- $y(1) = \sum_0^1 x(k)$
 $y(1) = x(0) + x(1)$

Test 2:- $y(-1) = \sum_0^{-1} x(k)$

$$y(-1) = x(-1) + x(0)$$

non-causal & dynamic

$$2) \quad y(n) = \sum_{-\infty}^n x(k)$$

Sol Test 1:- $y(1) = \sum_{-\infty}^1 x(k)$
 $= x(1) + x(0) + x(-1) + \dots + x(-\infty)$
 (causal)

Test 2:-

$$y(-1) = \sum_{-\infty}^{-1} x(k)$$

$$y(-1) = x(-1) + x(-2) + \dots + x(-\infty)$$

causal and dynamic

$$3) y(n) = \sum_{-\infty}^n x(k)$$

$$y(1) = \sum_{-\infty}^2 x(k)$$

$$y(1) = x(2) + x(1) + x(0) + x(-1) \dots$$

non-causal $x(-\infty)$

$$y(-1) = \sum_{-\infty}^0 x(k)$$

$$y(-1) = x(0) + x(-1) + x(-2) + \dots$$

$x(-\infty)$

\therefore non-causal & dynamic

$$4) y(t) = \frac{d}{dt} x(t)$$

A differentiator is a system which needs two different values present and past to be evaluated so it should be a causal dynamic system.

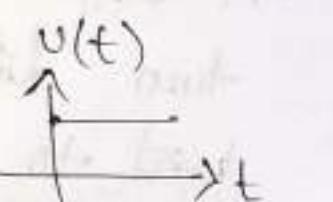
4) Invertible & Non-Invertible Systems:-

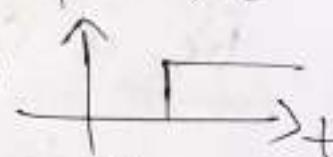
Invertible :- A system whose input can be recovered back by some kind of operation is called as an invertible system, otherwise the system is non-invertible system.

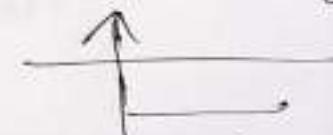
Note:- For a system to be invertible there must be a one-to-one relationship b/w $s(p)$ & $s(p)$

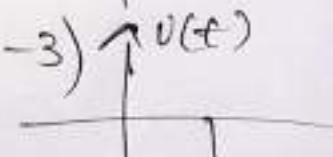
Eg:- Invertible System

$$y(t) = x(t-3)$$

Test 1:- $x_1(t) = u(t)$ 

$$y_1(t) = u(t-3)$$
 

Test 2:- $x_2(t) = -u(t)$ 

$$y_2(t) = -u(t-3)$$
 

for $x_1 \neq x_2$
 $y_1 \neq y_2$

The inverse system is invertible
 System for above system is
 $y'(t) = x'(t+3) \therefore [y(t) = x(t+3)]$

eg:- Non-Invertible System

$$y(t) = |\alpha(t)|, y(t) = \alpha^2(t)$$

\downarrow are non-invertible systems

$$\text{Test 1: } n_1(t) = u(t)$$

$$y_1(t) = v(t)$$

$$\text{Test 2: } n_2(t) = -v(t)$$

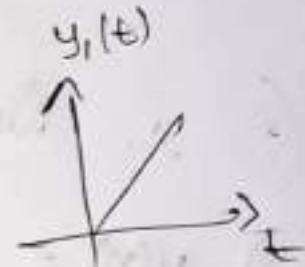
$$y_2(t) = v(t)$$

$$1) y(t) = \int_{-\infty}^t \alpha(\tau) d\tau$$

sol

$$\alpha(\tau) = v(\tau)$$

$$y(t) = \int_{-\infty}^t v(\tau) d\tau = g(t)$$



$$n_2(t) = -v(t)$$

$$y_2(t) = \int_{-\infty}^t -v(\tau) dt = -g(t)$$

for $n_1 \neq n_2$

$$y_1 \neq y_2$$



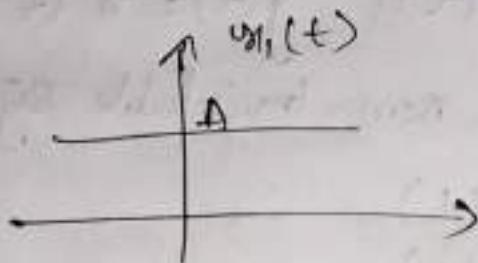
Invertible system is

$$y(t) = \frac{d}{dt} \alpha(t)$$

Invertible

$$2) \quad y(t) = \frac{d}{dt} (x(t))$$

Sol



$$y_1(t) = 0$$



$$y_2(t) = 0$$

∴ Non-invertible

$$3) \quad y(n) = u(n)u(n-1)$$

Sol

$$\begin{cases} u_1(n) = v(n) \\ \end{cases}$$

$$u_1(n) = v(n)v(n-1)$$

$$\begin{cases} u_2(n) = -v(n) \\ \end{cases}$$

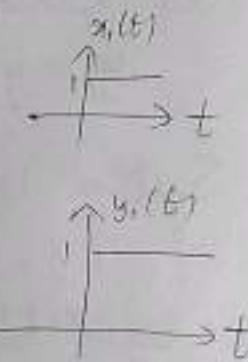
$$y_2(n) = -v(n)(-v(n-1))$$

$$y_2(n) = v(n)v(n-1)$$

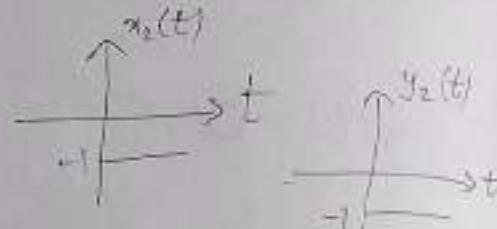
∴ Non-invertible system

$$4) \quad y(t) = x^3(t)$$

Sol. Test 1: $x_1(t) = u(t)$
 $y_1(t) = u^3(t)$



Test 2: $x_2(t) = -u(t)$
 $y_2(t) = -u^3(t)$



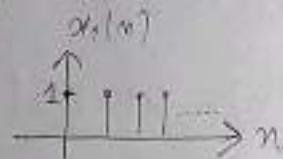
Invertible system

Inverse system is $y = \sqrt[3]{x(t)}$ *not*



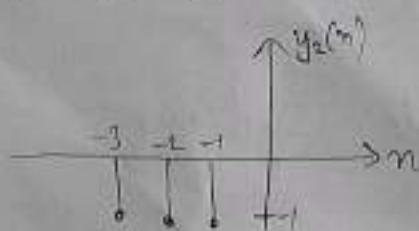
$$5) \quad y(n) = x(1-n) \Rightarrow \text{similar to } y(t) = x(1-t)$$

Sol. Test 1: $x_1(n) = u(n)$
 $y_1(n) = u(1-n)$



Test 2: $x_2(n) = u(n)$

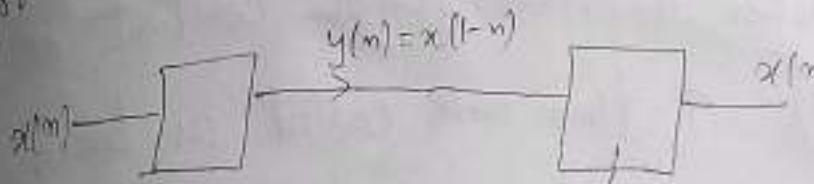
$$y_2(n) = -u(1-n)$$



Invertible

inverse
system

$$y(n) = x(1-n)$$



$$\begin{aligned} y(n) &= x(1-n) \\ &= x(1-(1-n)) = x(n) \end{aligned}$$

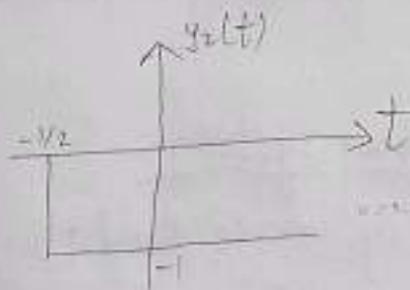
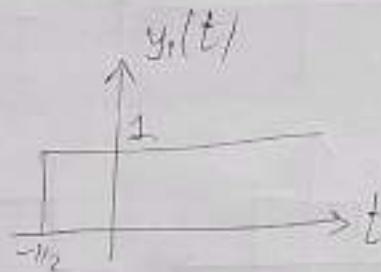
6) $y(t) = x(2t)$

Sl. $x(t) = u(t+1)$

$$y(t) = u(2t+1)$$

$$x_1(t) = -u(t+1)$$

$$y_2(t) = -u(2t+1)$$



Invertible system

Inverse system is $y(t) = x(t/2)$

7) $y(t) = \cos(x(t))$

Sl. $x_1(t) = 2\pi$

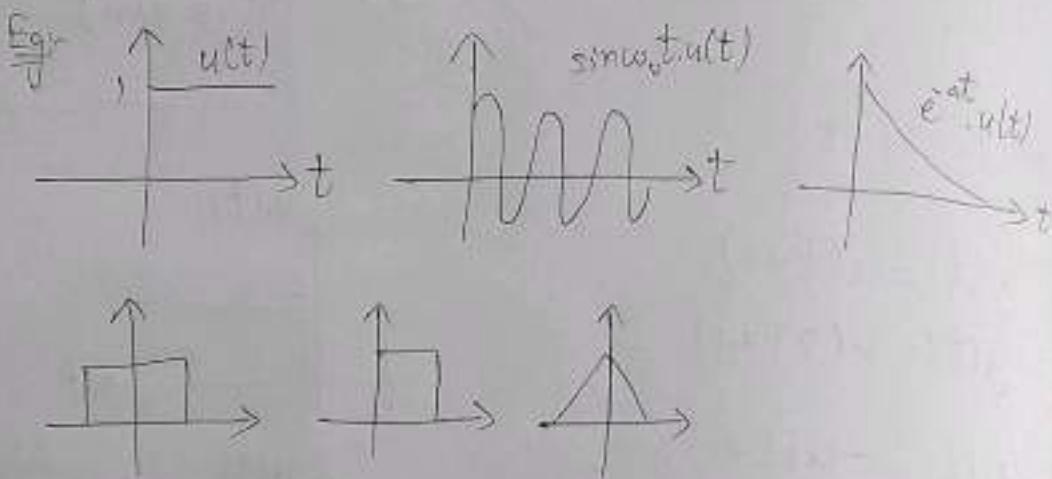
$$y_1(t) = \cos 2\pi = 1$$

Sl. $x_2(t) = 4\pi \quad y_2(t) = \cos 4\pi = 1$

\therefore Non-invertible

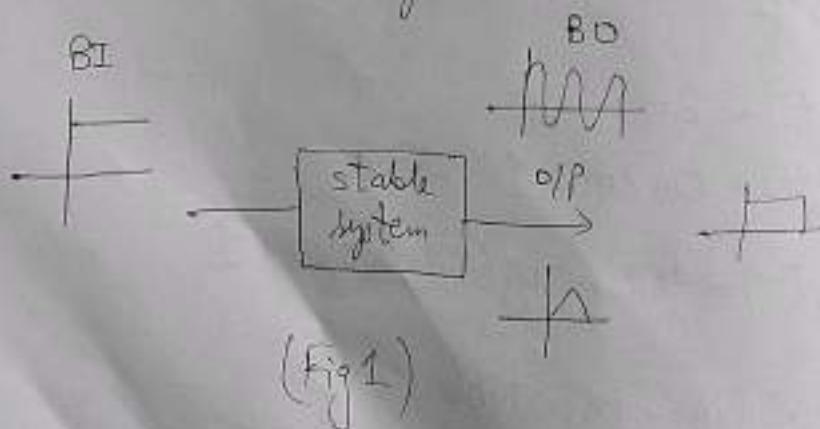
5) Bounded signals:-

Signals whose amplitude remain constant or die down with \uparrow time are called as bounded signals.
 $\text{As } t \rightarrow \infty, A \rightarrow 0, \text{ constant}$



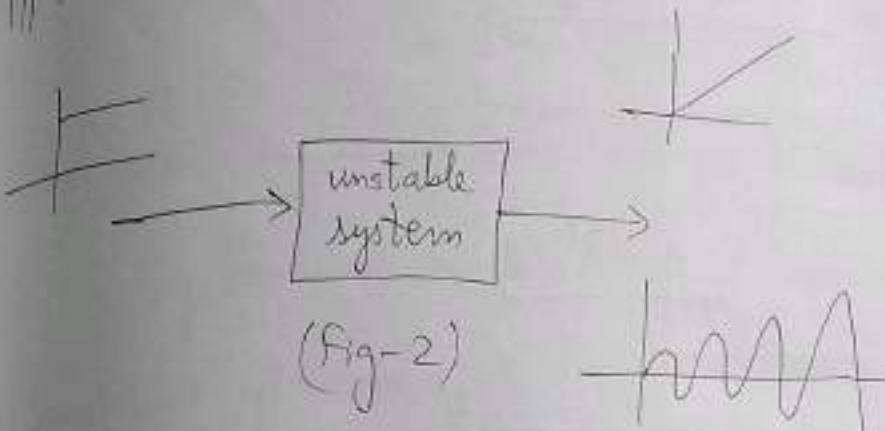
Stable system:-

A system is said to be BIBO (bounded IIP bounded OIP) stable system if it generates a bounded output for every bounded input, as illustrated in the figure below:



A system is said to be BIBO unstable if it generates an unbounded o/p for an bounded I/P.

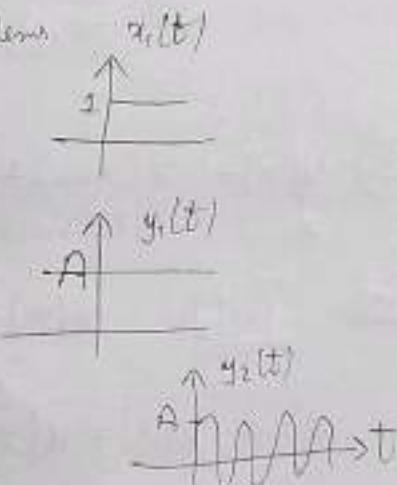
I/P:



Q) $y(t) = Ax(t) \rightarrow$ only 1 test can be done
in these problems

Test 1:- $x_1(t) = u(t)$

$$y_1(t) = A u(t)$$



Test 2:- $x_2(t) = \sin \omega_0 t, u(t)$

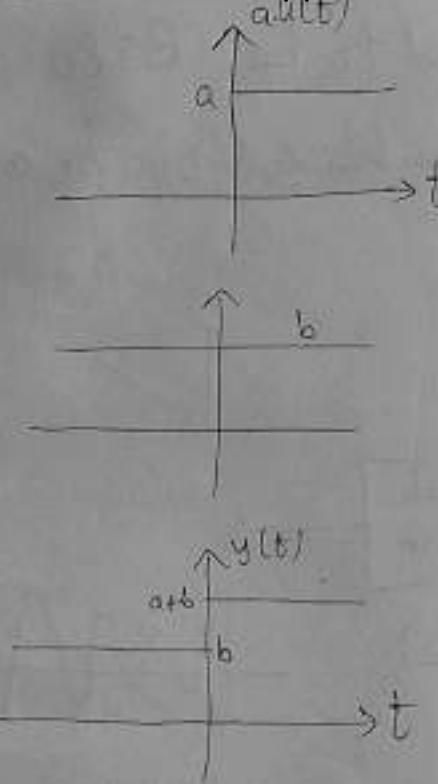
$$y_2(t) = A \sin \omega_0 t, u(t)$$

Since the o/p is bounded for every bounded I/P, it is a stable system.

$$y(t) = ax(t) + b$$

Test 1:- $x_1(t) = u(t)$

$$y_1(t) = a \cdot u(t) + b$$



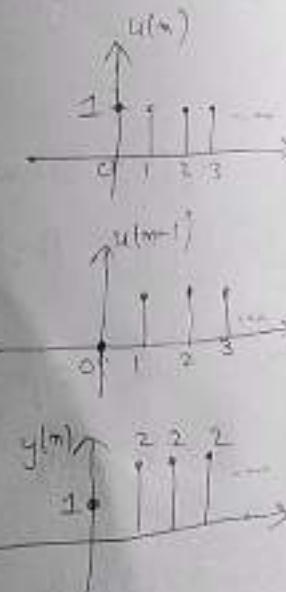
Stable system,

$$3.) y(n) = x(n-1) + x(n)$$

Sol:- Test 1:- $x(n) = u(n)$

$$y(n) = u(n) + u(n-1)$$

∴ The system is stable

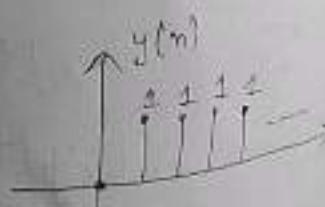


$$4.) y(n) = x(n), x(n-1)$$

Sol:- let $x(n) = u(n)$

$$y(n) = u(n)u(n-1)$$

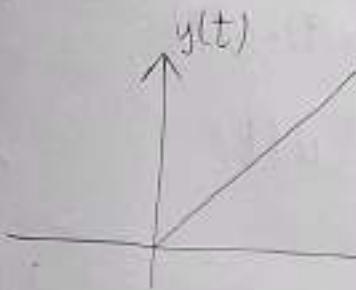
∴ Stable system



$$\text{Q3. } y(t) = t \cdot x(t)$$

$$\text{Let } x(t) = u(t)$$

$$\therefore y(t) = t \cdot u(t)$$



\therefore output is unbounded (ramp signal)

\therefore System is unstable.

Ans for $y(n) = n, x(n)$

$$\text{Q4. } y(t) = (t+5)u(t)$$

$$\text{Sol. Let } x(t) = u(t)$$

$$y(t) = (t+5)u(t)$$

$$y(t) = t \cdot u(t) + 5u(t)$$



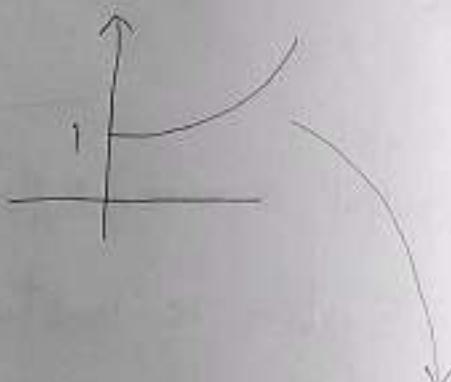
System is unstable

$$7) y(t) = e^{2t} \cdot x(t)$$

Sol. let $x(t) = u(t)$

$$y(t) = e^{2t} \cdot u(t)$$

: unstable



$$8) y(t) = e^{2t} x(-t)$$

Sol. $x(t) = u(t)$

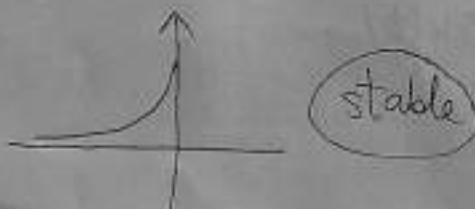
$$y(t) = e^{2t} \cdot u(-t)$$

$$= e^{2t}, t < 0$$

$$= 0, t \geq 0$$

\Rightarrow exponentially
growing again
with causal
given by
unstable

$$\begin{bmatrix} -t > 0 \\ t \leq 0 \end{bmatrix}$$



$$9) y(n) = e^{3n} \cdot x(4-n)$$

Sol. $x(n) = u(n)$

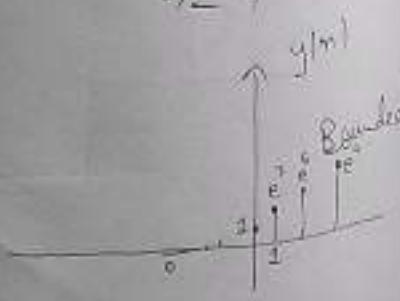
$$y(n) = e^{3n} \cdot u(4-n)$$

$$= e^{3n} \cdot 1, n \leq 4$$

$$= 0, n > 4$$

Stable

$$\begin{aligned} 4-n &\geq 0 \\ n &\leq 4 \\ n &\in \mathbb{N} \end{aligned}$$



$$\begin{aligned} m &> 0 \\ B &= 0 \end{aligned}$$

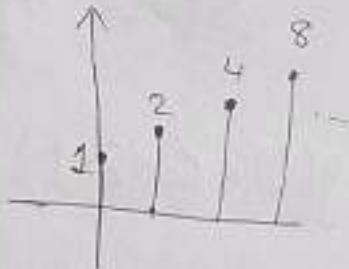
$$\text{i) } y(n) = 2^n \cdot x(n)$$

Sol:- $x(n) = u(n)$

$$y(n) = 2^n \cdot u(n)$$

$$= 2^n, \quad n \geq 0$$

$$= 0, \quad n < 0$$



$n \rightarrow \infty$

$A \rightarrow 0$

Unstable

$$\text{ii) } y(n) = 2^n x(-n)$$

Sol:- $x(n) = u(n)$

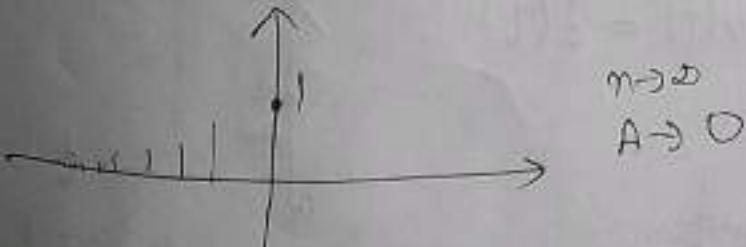
$$y(n) = 2^n u(-n)$$

$n \geq 0$

$n \leq 0$

$$= 2^n \quad n \leq 0$$

$$= 0 \quad n > 0$$

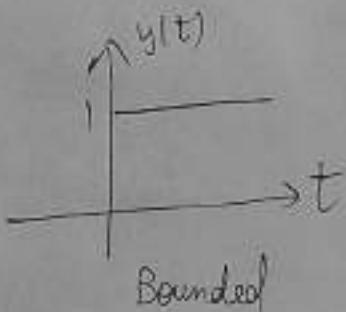


$\therefore \text{Stable} //$

$$1) y(t) = x^2(t)$$

Sol:- $x(t) = u(t)$

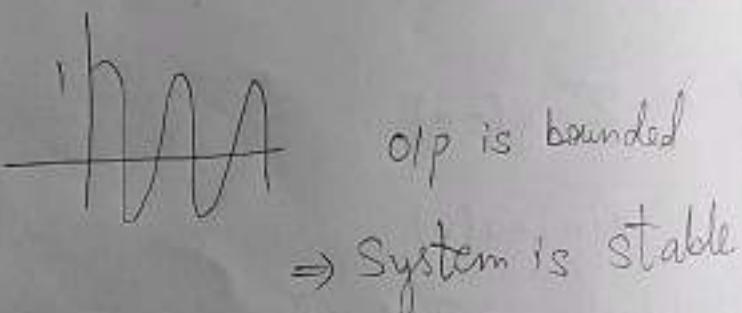
$$y(t) = u^2(t) = u(t) \times u(t)$$



Stable system

$$2) y(t) = \cos(2\pi t) \cdot x(t)$$

Sol:- $y(t) = \cos(2\pi t) u(t)$

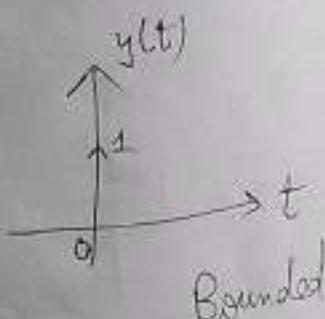


$$3) y(t) = \frac{d}{dt} x(t)$$

Sol:- $x(t) = u(t)$

$$y(t) = \frac{d}{dt} u(t) = \delta(t)$$

Stable system



$$y(t) = \int_{-\infty}^t x(\tau) d\tau$$

~~Eqn~~

$$x(t) = u(t)$$

$$y(t) = \int_{-\infty}^t u(\tau) d\tau$$

$$= \sigma(t)$$



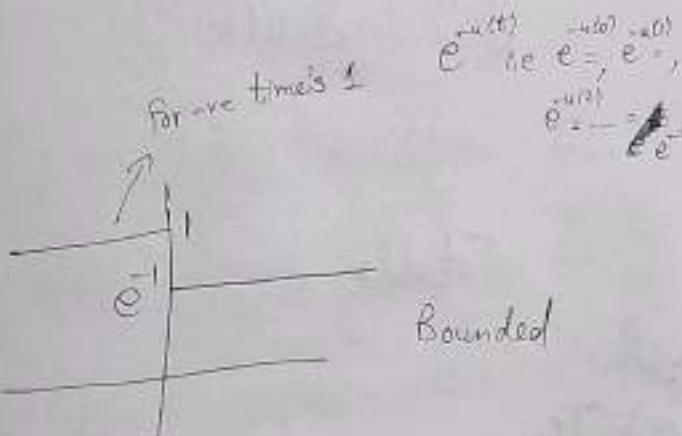
\therefore unstable system.

5) $y(t) = e^{x(t)}$

~~Eqn~~

$$x(t) = u(t)$$

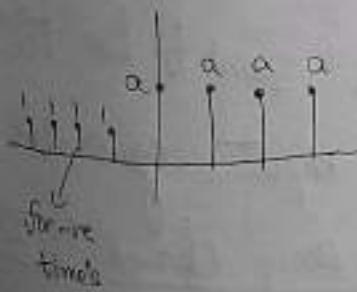
$$y(t) = e^{-u(t)}$$



Stable.

6) $y(n) = a^{x(n)}$

7) $y(n) = a^{u(n)}$



Stable

$$7.) y(n) = \sum_{-\infty}^{\infty} x(k)$$

Sol:- $x(k) = u(k)$

$$\begin{aligned} y(n) &= \sum_{-\infty}^{\infty} u(k) \\ &= \sum_0^1 u(k) = \sum_0^1 (1) = 5 \end{aligned}$$

$$8.) y(n) = \sum_{n=n_0}^{n=n_1} x(k)$$

Sol:- $x(k) = u(k)$

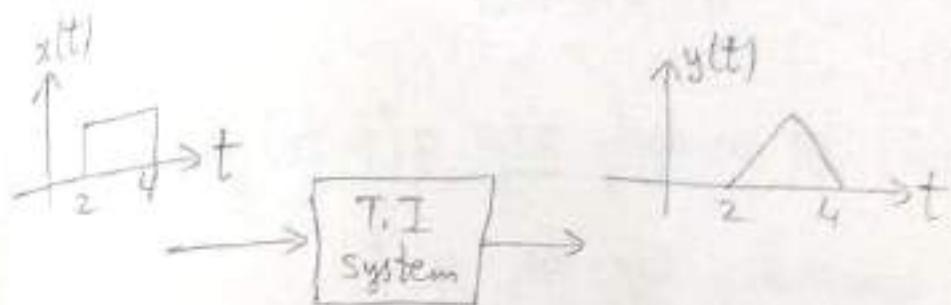
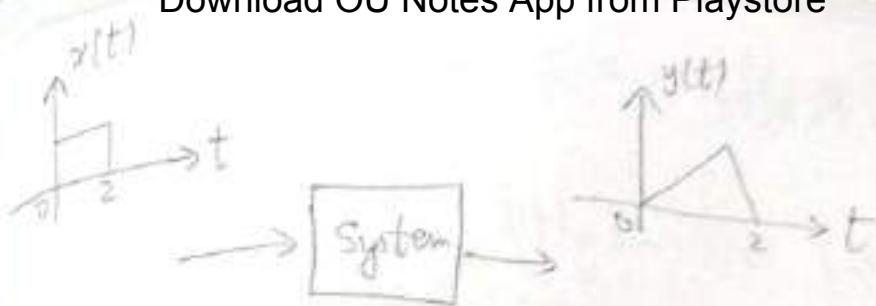
$$y(n) = \sum_{n=n_0}^{n=n_1} u(k)$$

Stable

(2M)
6.) Time invariant & Time variant systems.

A system is said to be time invariant if its behaviour is independent of time.

Mathematically, a system is said to be time invariant, if a delayed input generates a delayed output without any change in shape as illustrated in the figure below



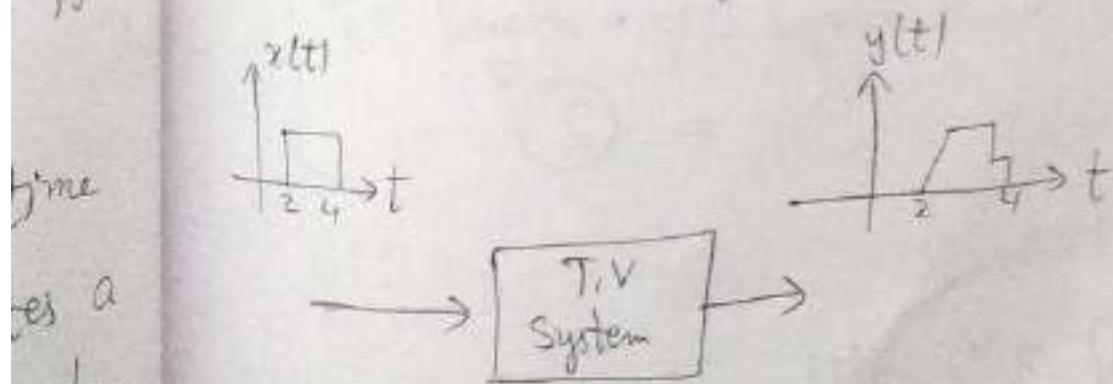
i.e if $x(t) \rightarrow y(t)$

then if $x(t-t_0) \rightarrow y'(t) = y(t-t_0)$

then the system is said to be time invariant.

Time variant system:-

If the system doesn't follow the above condition, then it is time dependent & is said to be time variant system which can be illustrated as shown in the figure below.



Time
yes a
shape

(a) $y(t) = A \cdot x(t)$

Sol:- Step 1) $x(t) \rightarrow y(t)$
 $y(t) = A \cdot x(t)$

Step 2) Let us apply I/p $x(t-t_0)$ & let
 the corresponding o/p be $y'(t)$
 $x(t-t_0) \rightarrow y'(t)$
 $y'(t) = A \cdot x(t-t_0) \leftarrow \textcircled{1}$

Step 3) Let us find $y(t-t_0)$

$y(t-t_0) = \cancel{A} \cdot x(t-t_0) \leftarrow \textcircled{2}$

Comparing step $\textcircled{2}$ & step $\textcircled{3}$

They are equal.

\therefore Time invariant system.

[i.e o/p due to delayed I/p = delayed o/p]
 $\textcircled{1} = \textcircled{2}$

Pb) $y(t) = t \cdot x(t)$

Sol:- S1) $x(t) \rightarrow y(t)$
 $y(t) = t \cdot x(t)$

52) Let us apply Z.P. $x(t-t_0)$ & let
 the corr. o/p be $y'(t)$
 $x(t-t_0) \rightarrow y'(t)$
 $y'(t) = t \cdot x(t-t_0) - ①$

53) Let us find $y(t-t_0)$

$$y(t-t_0) = (t-t_0) \cdot x(t-t_0) - ②$$

$$\therefore y'(t) \neq y(t-t_0)$$

It is a time variant system.

Q) Check whether T.I or T.V if $y(t) = x(t) + x(t-2)$

Sol. 51) $x(t) \rightarrow y(t)$ $y(n) = x(n) + x(n-2)$
 $y(t) \rightarrow x(t) + x(t-2)$

52) Let us apply Z.P. $x(t-t_0)$ & let the corr.
 o/p be $y'(t)$

$$x(t-t_0) \rightarrow y'(t)$$

$$y'(t) = x(t-t_0) + x(t-t_0-2) - ①$$

$$53) y(t-t_0) = x(t-t_0) + x(t-t_0-2) - ②$$

$$① = ② \text{ i.e. } y'(t) = y(t-t_0)$$

\therefore Time invariant

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check whether the following are time
variant & time invariant.

$$3) y(t) = x(t) + x(t-2)$$

$$\underline{\text{Sol}} \text{ Step 1:- } x(t) \rightarrow y(t)$$

$$y(t) = x(t) + x(t-2)$$

s₂):- let us apply ZP $x(t-t_0)$

& let corresponding ZP be

$$y'(t)$$

$$x(t-t_0) \rightarrow y'(t)$$

$$y'(t) = x(t-t_0) + x(t-t_0-2)$$

s₃) If we find $y(t-t_0)$

$$y(t-t_0) = x(t-t_0) + x(t-2-t_0)$$

$$\text{s₄) } y'(t) = y(t-t_0)$$

\therefore The above is time
invariant

$$4) y(n) = x(n)x(n-5)$$

s1) $x(n) \rightarrow y(n)$

$$\Rightarrow y(n) = x(n)x(n-5)$$

s2) Let us Apply SLP $x(n-n_0)$

& let corresponding SLP be $y'(n)$

$$x(n-n_0) \rightarrow y'(n)$$

$$y'(n) = x(n-n_0)x(n-5-n_0)$$

$$y'(n) = x(n-n_0)x(n-n_0-5)$$

s3) Let us find $y(n-n_0)$

$$y(n-n_0) = x(n-n_0)x(n_0-5-n_0)$$

$$y(n-n_0) = y'(n)$$

∴ Time Invariant

$$5) y(t) = x(-t)$$

s1) $x(t) \rightarrow y(t)$

$$y(t) = x(-t)$$

s2) $x(t-t_0) \rightarrow y'(t)$

$$y'(t) = x(-t-t_0)$$

$$S3) \quad y(t-t_0) = n[-(t-t_0)]$$

$$y(t-t_0) = n(t_0-t)$$

$$y(t-t_0) \neq y'(t)$$

\therefore Time Variant

$$6) \quad \frac{dy(t)}{dt} + y(t) = n(t)$$

$$\underline{\underline{S1}} \quad n(t) \rightarrow y(t)$$

$$\frac{dy(t)}{dt} + y(t) = n(t) \rightarrow ①$$

$$S2) \quad n(t-t_0) \rightarrow y'(t)$$

$$\frac{dy(t)}{dt} + y'(t) = n(t-t_0) \rightarrow ②$$

S3) Substitute eq ① & ②

$$\frac{dy'(t)}{dt} + y'(t) = \frac{dy(t-t_0)}{dt} + y(t-t_0)$$

Comparing similar terms

$$y'(t) = y(t-t_0)$$

$$y(t) = y(t-t_0)$$

\therefore Time Invariant

$$1) \frac{d^2y}{dt^2} + \frac{dy(t)}{dt} + 2y(t) + 2 = x(t)$$

s1) $x(t) \rightarrow y(t)$

$$+ \frac{d^2y(t)}{dt^2} + \frac{dy(t)}{dt} + 2y(t) + 2 = x(t) \quad \text{--- } ①$$

s2) $x(t-t_0) \rightarrow y'(t)$

$$+ \frac{d^2y'(t)}{dt^2} + \frac{dy'(t)}{dt} + 2y'(t) + 2 = x(t-t_0) \quad \text{--- } ②$$

s3) Substitute ① in ②

$$+ \frac{d^2y(t)}{dt^2} + \frac{dy(t)}{dt} + 2y(t) + 2$$

$$\stackrel{(t-t_0)^2}{=} + \frac{d^2y(t-t_0)}{dt^2} + \frac{dy(t-t_0)}{dt}$$

$$+ 2y(t-t_0) + 2$$

$$ty'(t) \neq y(t-t_0)(t-t_0)$$

\therefore Time variant

$$8) y(n-1) + 2y(n) = x(n)$$

s1) Given:- $x(n) \rightarrow y(n)$

$$y(n-1) + 2y(n) = x(n)$$

$$x(n-n_0) \rightarrow y'(n)$$

$$y'(n-1) + 2y'(n) = x(n-n_0)$$

$y(n-n_0)$

33) sub ① in ②

$$y'(n-1) + 2y'(n) = y(n-1-n_0) \\ + 2y(n-n_0)$$

$$y(n-1) = y(n-n_0 - 1)$$

$$y(n-1+1) = y(n-n_0 - 1 + 1)$$

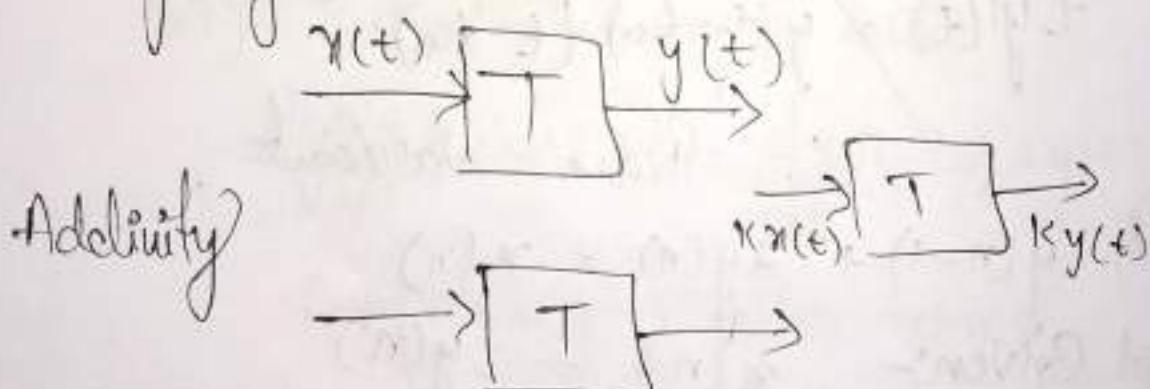
$$y(n) = y(n-n_0)$$

$$y'(n) = y(n-n_0)$$

∴ Time Invariant

* 7) Linear and Non-linear System :-

Homogeneity



Additivity

$$(a)x + (b)y = ax + by$$

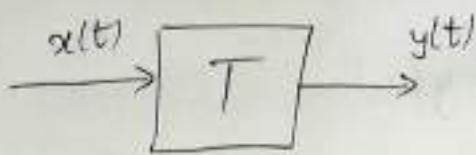
$$(a+b)x = ax + bx$$

$$(ab)x = a(bx)$$

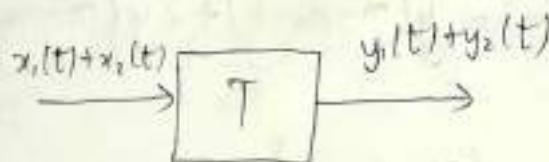
$$(a+b)(c+d)x = acx + adx + bcx + bdx$$

7.) Linear and non-linear Systems

Homogeneity



Additivity



Superposition



→ A system is said to be linear if it satisfies the property of superposition, i.e

if $x_1(t) \rightarrow y_1(t)$ and
 $x_2(t) \rightarrow y_2(t)$ then

$$\text{if } \alpha x_1 + \beta x_2 \rightarrow y'(t) = \alpha y_1 + \beta y_2$$

Otherwise it is a non-linear system.

$$y(t) = Ax(t)$$

Let $x_1(t)$ generates an o/p $y_1(t)$

$$\Rightarrow x_1(t) \rightarrow y_1(t)$$

$$y_1(t) = Ax_1(t)$$

Let $x_2(t)$ generates an o/p $y_2(t)$

$$x_2(t) \rightarrow y_2(t)$$

$$\Rightarrow y_2(t) = Ax_2(t)$$

S2) Id 8lp

 $\alpha x_1(t) + \beta x_2(t)$ generates an o/p $y_3(t)$

$$\alpha x_1(t) + \beta x_2(t) \rightarrow y_3(t)$$

$$y_3(t) = \alpha x_1(t) + \beta x_2(t)$$

S3) Id w/ find $\alpha y_1(t) + \beta y_2(t)$

$$\alpha y_1(t) + \beta y_2(t) = \alpha Ax_1(t) + \beta Ax_2(t)$$

$$\therefore y_3(t) = \alpha y_1(t) + \beta y_2(t)$$

 \therefore the system is linear

$$2) y(n) = Ax(n) + b$$

Sol) Let $x_1(n) \rightarrow y_1(n)$

$$y_1(n) = Ax_1(n) + b$$

Id $x_2(n) \rightarrow y_2(n)$

$$y_2(n) = Ax_2(n) + b$$

S2) Id 8lp

 $\alpha x_1(n) + \beta x_2(n)$ generates $y_3(n)$

$$y_3(n) = A(\alpha x_1(n) + \beta x_2(n)) + b$$

ss) Let us find $\alpha y_1(n) + \beta y_2(n)$

$$\alpha y_1(n) + \beta y_2(n) = \alpha(Ax_1(n) + b) \\ + \beta(-Ax_2(n) + b)$$

$$y_3(n) \neq \alpha y_1(n) + \beta y_2(n)$$

∴ It's non-linear

3) $y(t) = x^2(t)$

s1) Let $x_1(t) \rightarrow y_1(t)$
s2) $y_1(t) \rightarrow x_1^2(t)$

Let $x_2(t) \rightarrow y_2(t)$
 $y_2(t) \rightarrow x_2^2(t)$

s2) Let S(p).

$\alpha x_1(t) + \beta x_2(t)$ generates $y_3(t)$

$$y_3(t) = (\alpha x_1(t) + \beta x_2(t))^2 \\ = \alpha^2 x_1^2(t) + \beta^2 x_2^2(t) + 2\alpha\beta x_1 x_2$$

ss) $\alpha y_1(t) + \beta y_2(t) =$

$$\alpha x_1^2(t) + \beta x_2^2(t)$$

$$y_3(t) \neq \alpha y_1(t) + \beta y_2(t)$$

∴ non-linear

$$|a+b| < |a| + |b|$$

4) $y(t) = |\pi(t)|$

S1) $y_1(t) = |\pi_1(t)|$

$$y_2(t) = |\pi_2(t)|$$

S2) $y_3(t) = |\alpha\pi_1(t) + \beta\pi_2(t)|$

S3) $\alpha y_1(t) + \beta y_2(t)$

$$= \alpha |\pi_1(t)| + \beta |\pi_2(t)|$$

\therefore non-linear

5) $y(t) = \int_{-\infty}^t \pi(\tau) d\tau$

S1) $y_1(t) = \int_{-\infty}^t \pi_1(\tau) d\tau$

$$y_2(t) = - \int_{-\infty}^t \pi_2(\tau) d\tau$$

S2) $y_3(t) = \int_{-\infty}^t \pi_1(\tau) \alpha + \pi_2(\tau) \beta d\tau$

$$\alpha y_1(t) + \beta y_2(t)$$

$$= \alpha \int_{-\infty}^t \pi_1(\tau) d\tau + \beta \int_{-\infty}^t \pi_2(\tau) d\tau$$

$$= y_3(t)$$

\therefore It is linear

$$6) \frac{dy(t)}{dt} + 2y(t) = x(t)$$

s1) $\frac{dy_1(t)}{dt} + 2y_1(t) = n_1(t)$

$$\frac{dy_2(t)}{dt} + 2y_2(t) = n_2(t)$$

s2) $y_3(t) \Rightarrow$

$$\frac{dy_3(t)}{dt} + 2y_3(t) = \alpha n_1(t) + \beta n_2(t)$$

s3) $\alpha y_1(t) + \beta y_2(t)$

Substitute s1 in s2

at $\frac{dy_3(t)}{dt} + 2y_3(t) = \alpha \left(\frac{dy_1(t)}{dt} + 2y_1(t) \right)$

comparing $+ \beta \left(\frac{dy_2(t)}{dt} + 2y_2(t) \right)$

$$\frac{dy_3(t)}{dt} = \alpha \frac{dy_1(t)}{dt} + \beta \frac{dy_2(t)}{dt}$$

$$y_3(t) = \alpha y_1(t) + \beta y_2(t)$$

$$y_3(t) = \alpha y_1(t) + \beta y_2(t)$$

linear system

$$7) \frac{d^2y(t)}{dt^2} + 2\frac{dy(t)}{dt} + 2y(t) + 2 = r(t)$$

$$\stackrel{\text{sol}}{=} 3) \frac{d^2y_1(t)}{dt^2} + 2\frac{dy_1(t)}{dt} + 2y_1(t) + 2 = n_1(t)$$

$$\frac{d^2y_2(t)}{dt^2} + 2\frac{dy_2(t)}{dt} + 2y_2(t) + 2 = n_2(t)$$

$$S_2) \alpha y_1(t) + \beta y_2(t) \rightarrow y_3(t)$$

$$\frac{d^2y_3(t)}{dt^2} + 2\frac{dy_3(t)}{dt} + 2y_3(t) + 2 = \alpha n_1(t) + \beta n_2(t)$$

S3) Sub S1 in S2

$$\begin{aligned} \Rightarrow \frac{d^2y_3(t)}{dt^2} + 2\frac{dy_3(t)}{dt} + 2y_3(t) + 2 \\ = \alpha \left[\frac{d^2y_1(t)}{dt^2} + 2\frac{dy_1(t)}{dt} \right. \\ \left. + 2y_1(t) + 2 \right] \\ + \beta \left[\frac{d^2y_2(t)}{dt^2} + 2\frac{dy_2(t)}{dt} \right. \\ \left. + 2y_2(t) + 2 \right] \end{aligned}$$

Comparing

$$\frac{d^2y_3(t)}{dt^2} = \frac{d^2}{dt^2} (\alpha y_1(t) + \beta y_2(t))$$

$$y_3(t) = \alpha y_1(t) + \beta y_2(t)$$

$$\frac{dy_3(t)}{dt} = \frac{d(\alpha y_1(t) + \beta y_2(t))}{dt}$$

$$y_3(t) = \alpha y_1(t) + \beta y_2(t)$$

$$y_3(t) = \alpha y_1(t) + \beta y_2(t)$$

$$2 = 2\alpha + 2\beta \\ \alpha + \beta = 1 \text{ (non-linear)}$$

\therefore System is not linear

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$$g) y(n-1) + 2y(n) = x(n)$$

$$g) y_1(n-1) + 2y_1(n) = x_1(n)$$

$$g) y_2(n-1) + 2y_2(n) = x_2(n)$$

$$S2) y_3(n-1) + 2y_3(n) = \alpha x_1(n) + \beta x_2(n)$$

$$S3) y_3(n-1) + 2y_3(n) = \alpha(y_1(n-1) + 2y_1(n)) \\ + \beta(y_2(n-1) + 2y_2(n)).$$

Comparing

$$y_3(n-1) = \alpha y_1(n-1) + \beta y_2(n-1)$$

$$y_3(n-1+1) = \alpha y_1(n-1+1) + \beta y_2(n-1+1)$$

$$y_3(n) = \alpha y_1(n) + \beta y_2(n)$$

$$y_3(n) = \alpha y_1(n) + \beta y_2(n)$$

\therefore Linear,,

$$(9) \quad y(n-2) + 2y(n-1) + ny(n) + 2 = x(n)$$

$$\underline{\text{S1}} \quad y_1(n-2) + 2y_1(n-1) + ny_1(n) + 2 = x_1(n)$$

$$y_2(n-2) + 2y_2(n-1) + ny_2(n) + 2 = x_2(n)$$

$$\text{S2} \quad y_3(n-2) + 2y_3(n-1) + ny_3(n) + 2 \\ = \alpha x_1(n) + \beta x_2(n)$$

Sub ① in ②

$$y_3(n-2) + 2y_3(n-1) + ny_3(n) + 2 = \\ \alpha(y_1(n-2) + 2y_1(n-1) + ny_1(n) + 2) \\ + \beta(y_2(n-2) + 2y_2(n-1) + ny_2(n) + 2)$$

$$y_3(n-2) = \alpha y_1(n-2) + \beta y_2(n-2)$$

$$y_3(n) = \alpha y_1(n) + \beta y_2(n)$$

$$y_3(n-1) = \alpha y_1(n-1) + \beta y_2(n-1)$$

$$y_3(n) = \alpha y_1(n) + \beta y_2(n)$$

$$2 = 2\alpha + 2\beta$$

$$\alpha + \beta = 1.$$

Non-linear

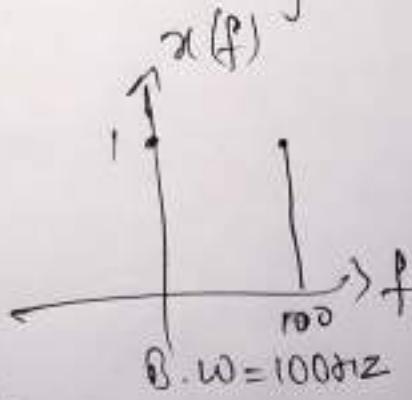
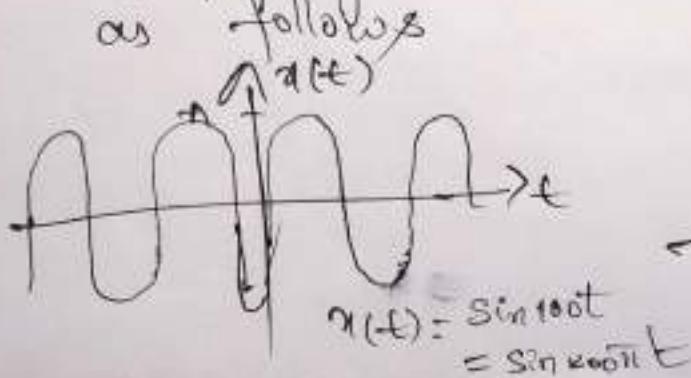
15/02/2020

UNIT - IV

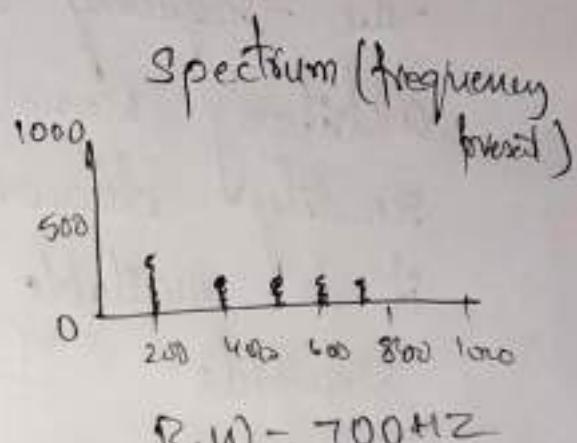
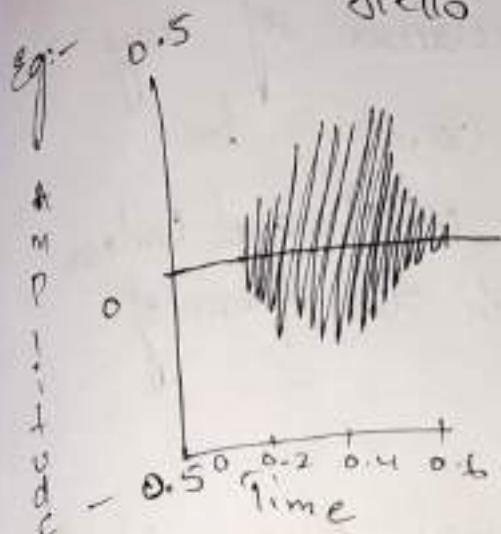
Fourier Series:-

- So far we have seen the time domain representation of a signal i.e. $x(t)$. But very often there are some issues that cannot be tackled using this of time domain like signal bandwidth and noise analysis. Both these issues can be only tackled using one more important version of signal called the frequency domain representation of a signal.
- It is also called as frequency spectrum of a signal.

- It is a plot of Amplitude vs frequency content in a signal as follows

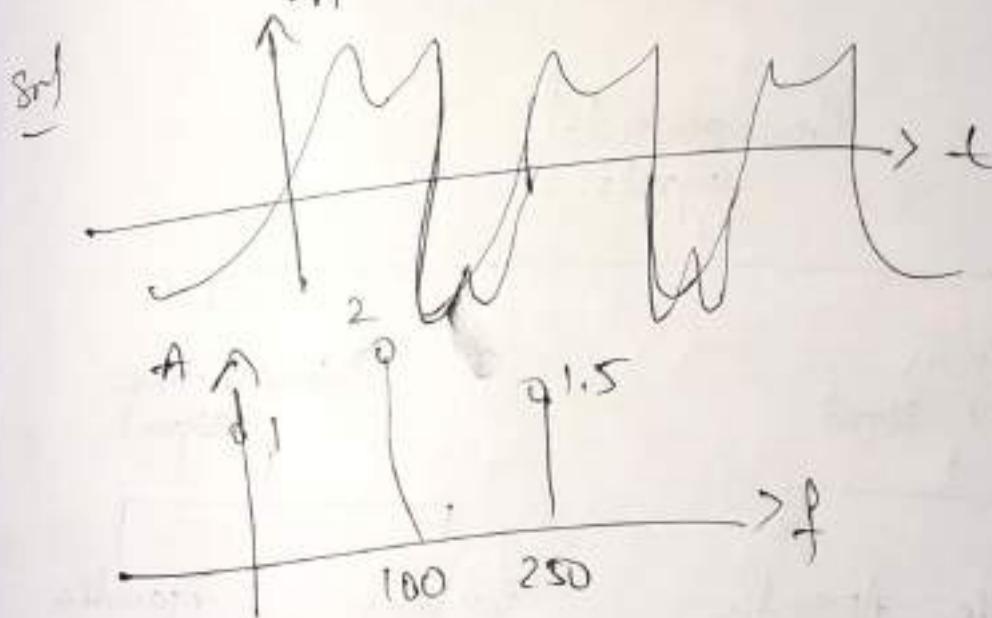


waveform of
Hello



) Draw the frequency spectrum of the following signal

$$x(t) = 1 + 2\sin 200\pi t + \frac{3}{2}\sin 500\pi t$$

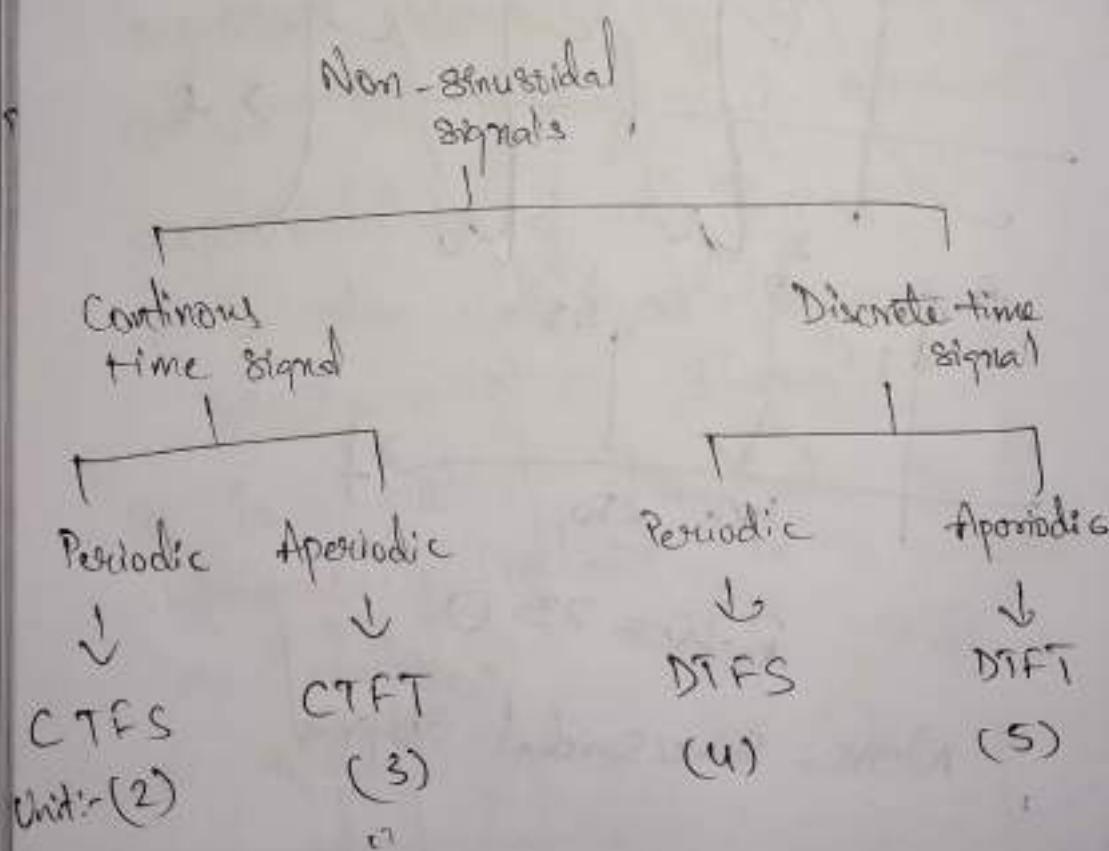


$$B.W = 250$$

Non-sinusoidal signal

Note:- The above example shows that the frequency spectrum of any arbitrary signal cannot be readily drawn until and unless it is available in the form of sinusoids.

> The above theory was observed by Fourier, and he proposed a method of converting any arbitrary non-sinusoidal signal into sinusoids, which came to be known as Fourier series.



* Discrete Time Fourier Series:-

(DTFS)

- It is a method developed by Fourier to convert a time domain discrete signal into a frequency domain representation.
- The following expression quantifies this theory:

$$x(n) \rightarrow D_k$$

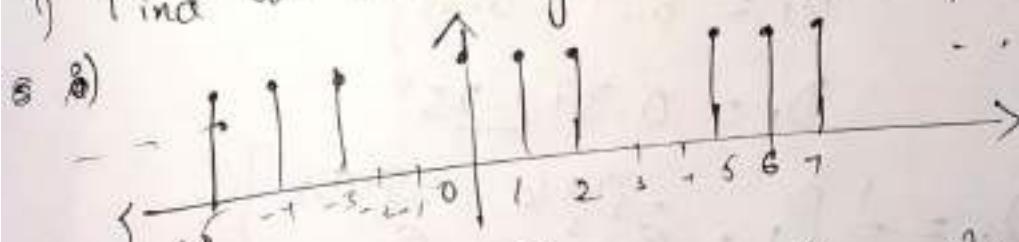
$$\{ x(n) = \sum_{k=0}^{N-1} D_k e^{j\omega_0 k n} \}$$

$N \rightarrow$ Time period of $x(n)$

where

$$\{ D_k = \frac{1}{N} \sum_{n=0}^{N-1} x(n) e^{-j\omega_0 k n}, k = 0 \rightarrow N-1 \}$$

Q) Find the DTFS of the following D.T signal
draw spectrum



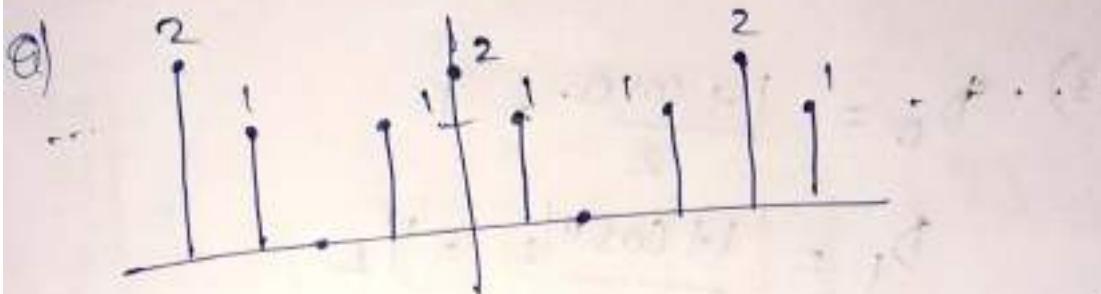
(or) $x(n) = \{ 1, 1, 0, 0 \}, N = 5$

Sol: 1) $N = 5$ (time period)

Fourier Series:-DFTS:-

$$x(n) = \sum_{k=0}^{N_0-1} D_k e^{j k \omega_0 n}$$

$$D_k = \frac{1}{N} \sum_{n=0}^{N_0-1} x(n) e^{-j k \omega_0 n}, k=0 \rightarrow N_0-1$$



S1) $N_0 = 4, \omega_0 = \frac{2\pi}{N_0} = \frac{2\pi}{4} = \pi/2$

S2) $D_k = \frac{1}{4} \sum_{n=0}^3 x(n) e^{-j k \omega_0 n}$
 $= \frac{1}{4} \sum_{n=0}^3 x(n) e^{-j k \pi/2 n}, k=0 \rightarrow 3$

$$\Rightarrow D_k = \frac{1}{4} \left[x(0) e^0 + x(1) e^{-j k \pi/2} + x(2) e^{-j k \pi} + x(3) e^{-j k 3\pi/2} \right]$$

$$D_k = \frac{1}{4} \left[2 + e^{-j k \pi/2} + 0 + e^{-j k 3\pi/2} \right]$$

$$= \frac{1}{4} \left[2 + e^{-j k \pi/2} + e^{-j k (2\pi - \pi/2)} \right], k \rightarrow 0 \rightarrow 3$$

$$= \frac{1}{4} \left[2 + e^{-j k \pi/2} + e^{j k \pi/2} \right]$$

$k=0 \rightarrow 3$

$$\boxed{\begin{aligned} e^{-j(2\pi-\theta)} \\ \text{formula} \end{aligned}}$$

$$\Rightarrow D_K = \frac{1}{4} \left[2 + \cos k\pi/2 - j \sin k\pi/2 + \cos k\pi/2 + j \sin k\pi/2 \right]$$

$$D_K = \frac{1}{4} \left[2 + 2 \cos k\pi/2 \right]$$

$$\Rightarrow D_K = \frac{1}{2} \left[1 + \cos k\pi/2 \right]$$

33) $D_0 = \frac{1 + \cos 0}{2} = 1$

$$D_1 = \frac{1 + \cos \pi/2}{2} = 1/2$$

$$D_2 = \frac{1 + \cos \pi}{2} = 0$$

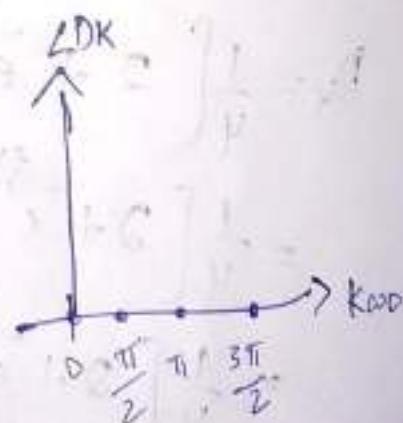
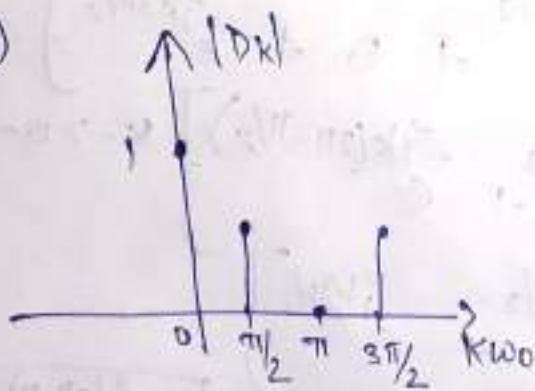
$$D_3 = \frac{1 + \cos 3\pi/2}{2} = 1/2$$

Su) $x(n) = \sum_{K=0}^3 D_K e^{j k \pi/2 n}$

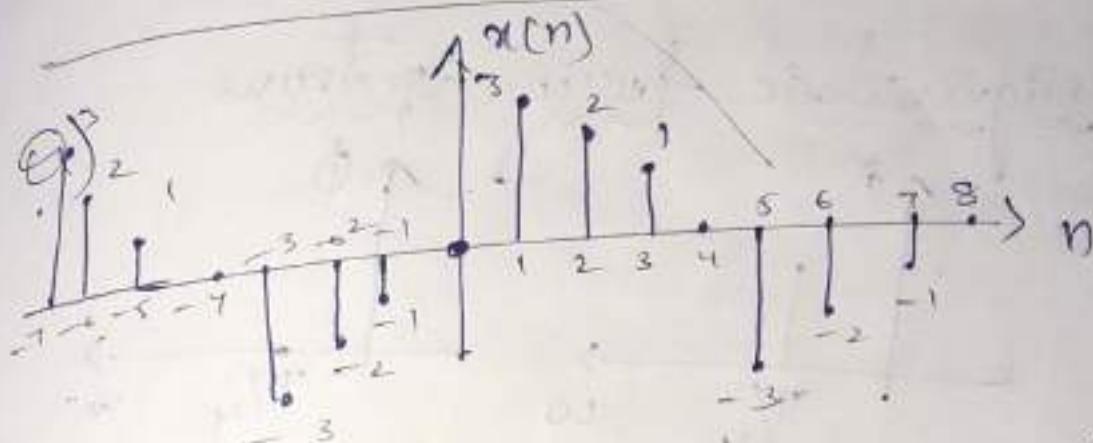
$$x(n) = D_0 e^0 + D_1 e^{j \pi/2 n} + D_2 e^{j \pi n} + D_3 e^{j 3\pi/2 n}$$

$$x(n) = 1 + 1/2 e^{j \pi/2 n} + 0 e^{j \pi n} + 1/2 e^{j 3\pi/2 n}$$

33)



a)



$$S1) \quad x(n) = \{0, -3, -2, -1, 0, 3, 2, 1\}$$

$$N_0 = 8 \quad \omega_0 = 2\pi/8 = \pi/4$$

$$S2) \quad D_k = \frac{1}{8} \sum_{n=0}^7 x(n) e^{-jkn\pi/4} \quad k = 0 \rightarrow 7$$

$$D_k = \frac{1}{8} \left[0 + 3e^{-j\pi/4} + 2e^{-j3\pi/2} + e^{-j5\pi/2} \right. \\ \left. + 0 + (-3)e^{-j5\pi/4} + (-2)e^{-j\pi/4} + (-1)e^{-j7\pi/4} \right] \Big|_{k=0 \rightarrow 7}$$

$$D_k = \frac{1}{8} \left[3e^{-j\pi/4} + 2e^{-j3\pi/2} + e^{-j5\pi/4} \right. \\ \left. - 3e^{-j\pi(2\pi - 3\pi/4)} - jk \right]$$

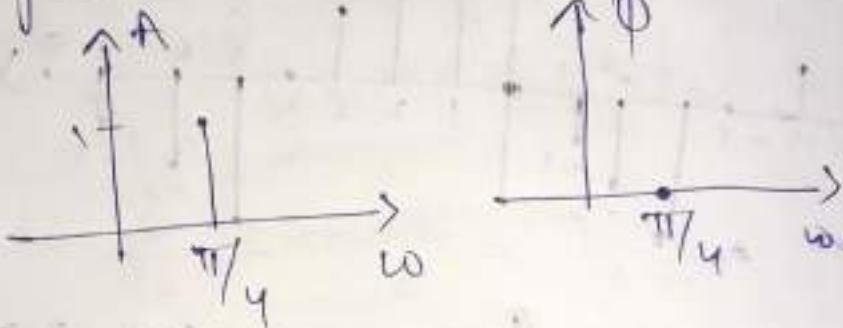
Fourier Series

→ To draw a Fourier Spectrum of the following signal.

$$1) x(n) = \cos \frac{\pi}{4} n$$

Sol Given:- $x(n) = \cos \frac{\pi}{4} n$

Trigonometric Fourier spectrum



EFS:-

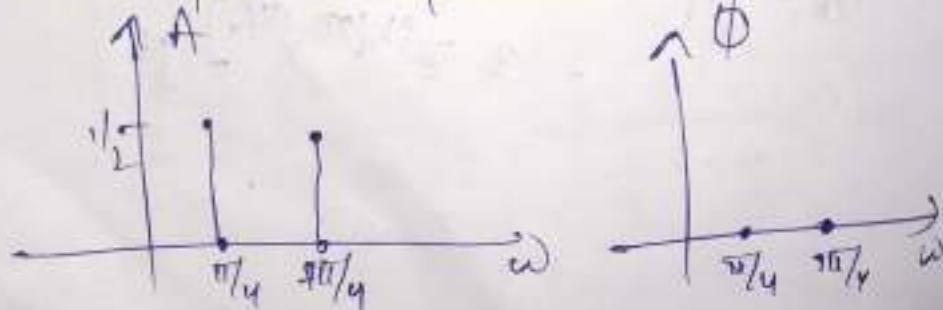
$$x(n) = \sum_{k=0}^{N-1} D_k e^{j k \omega n}$$

$$x(n) = \frac{e^{j \pi/4 n}}{2} + \frac{e^{-j \pi/4 n}}{2}$$

$$\boxed{\frac{e^{-j(2\pi - \theta)}}{2} = e^{j\theta}}$$

$$x(n) = \frac{e^{j \pi/4 n}}{2} + \frac{e^{j 3\pi/4 n}}{2}$$

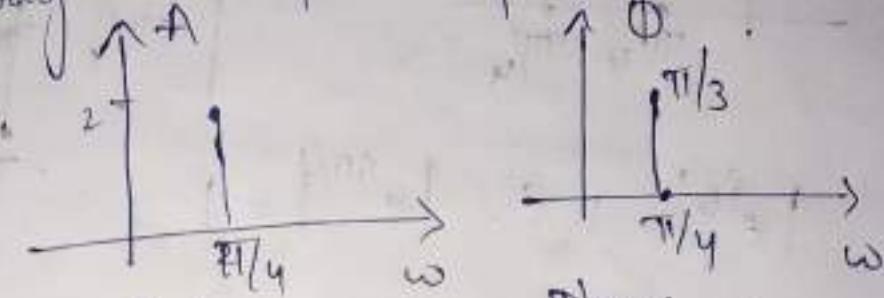
Exponential Fourier spectrum



$$x(n) = 2 \cos(\pi/4n + \pi/3)$$

Given:- $x(n) = 2 \cos(\pi/4n + \pi/3)$

(a) Trigonometric Fourier Spectrum



magnitude spectrum Phase spectrum

frequency Spectrum

Fourier Spectrum

(b) Exponential Fourier Spectrum

$$x(n) = \sum_{k=0}^{N-1} D_k e^{j k \omega n}$$

$$x(n) = \frac{2}{2} e^{j(\pi/4n + \pi/3)} + \frac{2}{2} e^{-j(\pi/4n + \pi/3)}$$

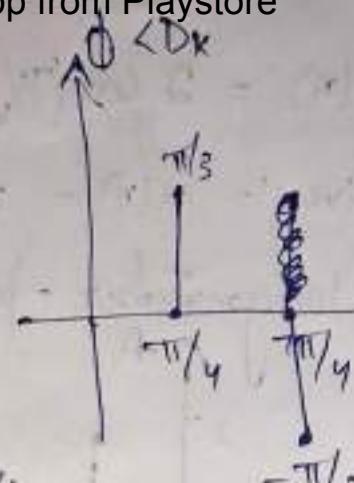
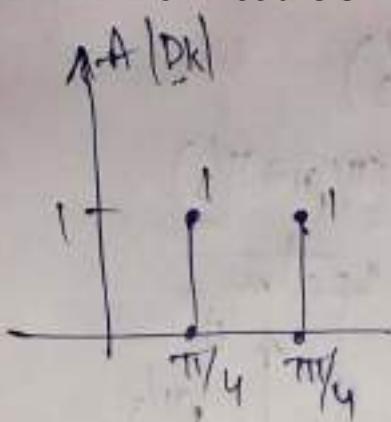
$$= e^{j\pi/3} e^{j\pi/4n} + e^{-j\pi/3} e^{-j\pi/4n}$$

$$= e^{j\pi/3} e^{j\pi/4n} + e^{-j\pi/3} e^{j\pi/4n}$$

$$\therefore \begin{cases} 2\pi - \pi/3 = \pi/4 \\ e^{j(2\pi - 0)} = e^{j0} \end{cases}$$

$$x(n) = e^{j\pi/3} e^{-j\pi/4n} + e^{-j\pi/3} e^{j\pi/4n}$$

Exponential Fourier Series



$$|e^{j0}| = 1 \therefore |e^{j3\pi/3}| = 1$$

$$3) x(n) = 1 + \frac{3}{2} \cos \frac{2\pi}{3} n + 2 \sin \frac{2\pi}{3} n$$

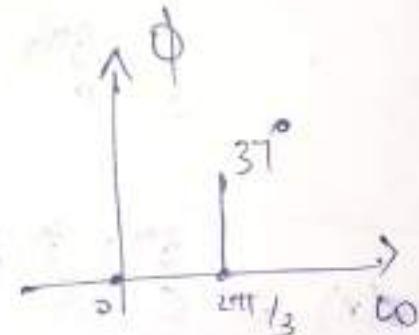
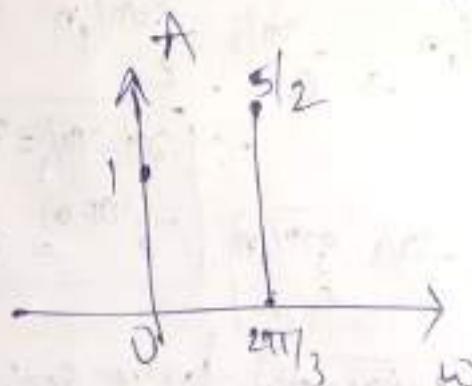
Sol

$$\begin{aligned} A \sin \theta &\rightarrow B \cos \theta \\ &= \sqrt{A^2 + B^2} \sin(\theta + \tan^{-1} B/A) \end{aligned}$$

Trigonometric Polar representation
Fourier spectrum

$$x(n) = 1 + \frac{5}{2} \sin\left(\frac{2\pi}{3} + \tan^{-1} \frac{3}{4}\right)$$

$$x(n) = 1 + \frac{5}{2} \sin(2\pi/3 + 51^\circ)$$



Exponential power series.

11B

$$a(n) = 1 + \frac{5}{2} \sin\left(\frac{2\pi}{3}n + 37^\circ\right)$$

$$\left[\sin \theta = \frac{e^{j\theta} - e^{-j\theta}}{2j} \right]$$

$$a(n) = 1 + \frac{5}{4j} e^{j(2\pi/3)n + 37^\circ} + \frac{5}{4j} e^{-j(2\pi/3)n - 37^\circ}$$

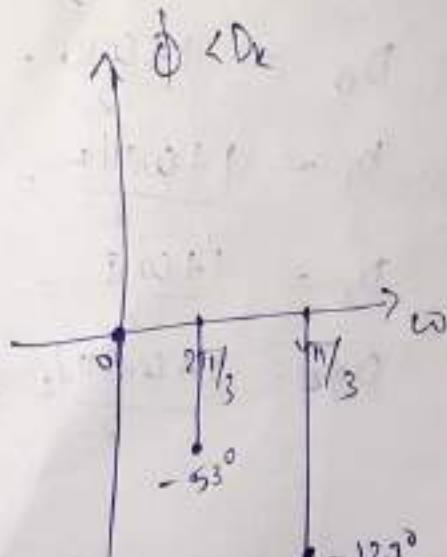
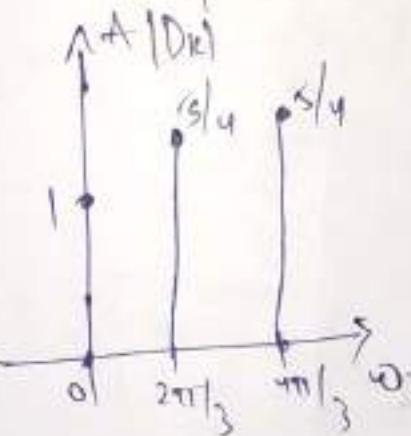
$$a(n) = 1 + \frac{5}{4j} e^{j37^\circ} e^{j2\pi/3n} + \frac{5}{4j} e^{-j37^\circ} e^{-j2\pi/3n}$$

$$a(n) = 1 + \frac{5}{4j} e^{j37^\circ} e^{j2\pi/3n} - \frac{5}{4j} e^{-j37^\circ} e^{j4\pi/3n}$$

$$a(n) = 1 + \frac{5}{4} (-j) e^{j37^\circ} e^{j2\pi/3n} - \frac{5}{4} (-j) e^{-j37^\circ} e^{j4\pi/3n}$$

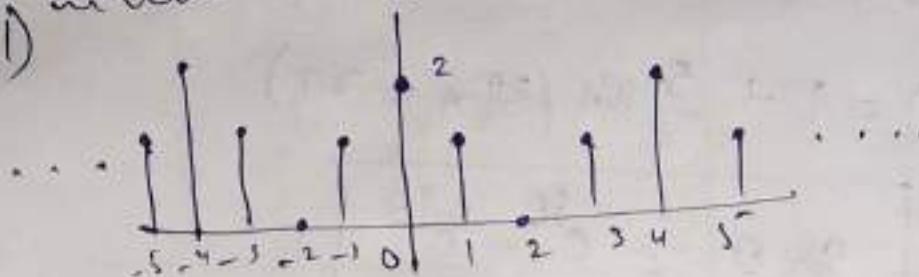
$$= 1 + \frac{5}{4} e^{-j90^\circ} e^{j37^\circ} e^{j2\pi/3n} - \frac{5}{4} e^{-j90^\circ} e^{-j37^\circ} e^{j4\pi/3n}$$

$$\left\{ a(n) = 1 + \frac{5}{4} e^{-j53^\circ} e^{j2\pi/3n} - \frac{5}{4} e^{-j127^\circ} e^{j4\pi/3n} \right\}$$



Problems On DFTS:-

1)



$$\text{Sol} \quad N_0 = 4, \quad \omega_0 = 2\pi/4 = \pi/2$$

$$D_K = \frac{1}{N_0} \sum_{n=0}^{N_0-1} x(n) e^{-jkn\omega_0} \quad k = 0 \rightarrow (4-1),$$

$$D_K = \frac{1}{4} \sum_{n=0}^3 x(n) e^{-jkn\omega_0} \quad k = 0 \rightarrow 3$$

$$D_K = \frac{1}{4} \left[x(0) e^0 + x(1) e^{-j\pi/2} + x(2) e^{-j\pi} + x(3) e^{-j3\pi/2} \right]$$

$$D_K = \frac{1}{4} \left[2 + 1 e^{-jK\pi/2} + \frac{1}{e^{-j3K\pi/2}} \right]$$

$$D_K = \frac{1}{4} \left[2 + e^{-jK\pi/2} + e^{-jK(3\pi - \pi/2)} \right] \quad k = 0 \rightarrow 3$$

$$D_K = \frac{1}{4} \left[2 + e^{-jK\pi/2} + e^{jK\pi/2} \right] \quad \text{cutoff} \\ \text{cyclic}$$

$$D_K = \frac{1}{4} \left[2 + 2 \cos(k\pi/4) \right], \quad k = 0 \rightarrow 3$$

$$D_K = \frac{1 + \cos K\pi/2}{2}, \quad k = 0 \rightarrow 3$$

$$D_0 = \frac{1 + \cos 0}{2} = 1$$

$$D_1 = \frac{1 + \cos \pi/2}{2} = 1/2$$

$$D_2 = \frac{1 + \cos \pi}{2} = 0$$

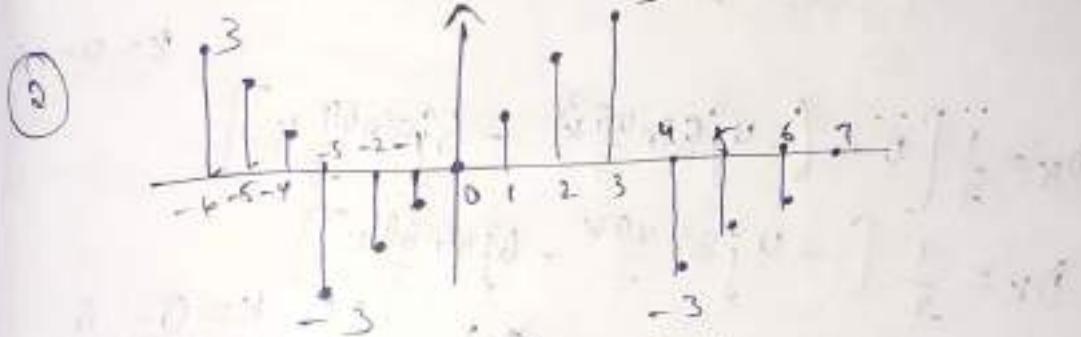
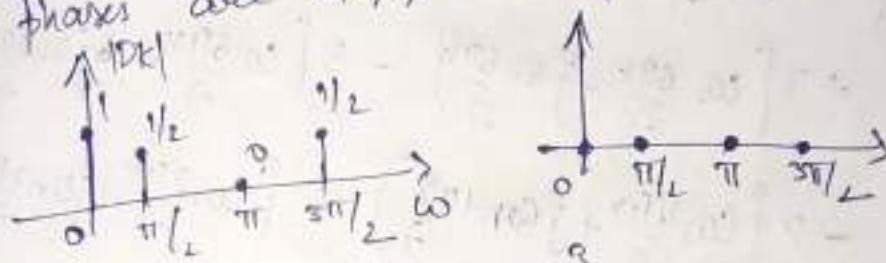
$$D_3 = \frac{1 + \cos 3\pi/2}{2} = 1/2$$

$$x(n) = \sum_{k=0}^{N_0} D_k e^{j k \pi / N} = \sum_{k=0}^3 D_k e^{j k \pi / 7}$$

$$x(n) = D_0 e^0 + D_1 e^{j \pi / 7 n} + D_2 e^{j 2\pi / 7 n} + D_3 e^{j 3\pi / 7 n}$$

$$x(n) = 1 + \frac{1}{2} e^{j \pi / 2 n} + 0 e^{j \pi n} + \frac{1}{2} e^{j 3\pi / 2 n}$$

The above signal dictates that $x(n)$ contains frequency components. $\omega = 0, \pi / 2, \pi \& 3\pi / 2$ & has corresponding Fourier coefficients as 1, 1/2, 0, 1/2. Phases are 0, 0, 0, & 0 respectively.



$$N_0 = 7, \omega_0 = 2\pi / N_0 = 2\pi / 7$$

$$D_K = \frac{1}{7} \sum_{n=0}^6 x(n) e^{-j \frac{2\pi}{7} n k}, \quad k = 0 \rightarrow 6$$

$$D_K = \frac{1}{7} \left[x(0) e^0 + x(1) e^{-j \frac{2\pi}{7} k} + x(2) e^{-j \frac{4\pi}{7} k} + x(3) e^{-j \frac{6\pi}{7} k} + x(4) e^{-j \frac{8\pi}{7} k} + x(5) e^{-j \frac{10\pi}{7} k} + x(6) e^{-j \frac{12\pi}{7} k} \right]$$

$$D_K = \frac{1}{7} \left[1 + e^{-j \frac{2\pi}{7} k} + 2 e^{-j \frac{4\pi}{7} k} + 3 e^{-j \frac{6\pi}{7} k} - 3 e^{-j \frac{8\pi}{7} k} - 2 e^{-j \frac{10\pi}{7} k} - e^{-j \frac{12\pi}{7} k} \right]$$

$$D_K = \frac{1}{7} \left[0 + e^{j\frac{3k\pi}{7}} + 2e^{j\frac{5k\pi}{7}} + 3e^{j\frac{6k\pi}{7}} - 3e^{j\frac{5k(2\pi-\frac{\pi}{7})}{7}} - 2e^{j\frac{5k(2\pi-\frac{4\pi}{7})}{7}} - 1 \cdot e^{j\frac{5k(2\pi-\frac{2\pi}{7})}{7}} \right]$$

$$D_K = \frac{1}{7} \left[0 + e^{j\frac{3k\pi}{7}} + 2e^{j\frac{5k\pi}{7}} + 3e^{j\frac{6k\pi}{7}} - 3e^{j\frac{5k(6\pi)}{7}} - 2e^{j\frac{5k(4\pi)}{7}} - e^{j\frac{5k(2\pi)}{7}} \right]$$

$$D_K = \frac{1}{7} \left[0 + \cos \frac{2\pi k}{7} \cdot j \sin \frac{2\pi k}{7} + 2 \left[\cos \frac{4\pi k}{7} - j \sin \frac{4\pi k}{7} \right] + 3 \left[\cos \frac{6\pi k}{7} - j \sin \frac{6\pi k}{7} \right] - 3 \left[\cos \frac{10\pi k}{7} + j \sin \frac{10\pi k}{7} \right] - 2 \left[\cos \frac{12\pi k}{7} + j \sin \frac{12\pi k}{7} \right] - \left[\cos \frac{14\pi k}{7} + j \sin \frac{14\pi k}{7} \right] \right]$$

$$K = 0 - 6$$

$$D_K = \frac{1}{7} \left[0 + (-4j \sin \frac{4\pi k}{7}) - 6j \sin \frac{6\pi k}{7} \right]$$

$$D_K = \frac{1}{7} \left[-4j \sin \frac{4\pi k}{7} - 6j \sin \frac{6\pi k}{7} \right]$$

$$D_0 = 0$$

$$D_1 = -0.0581j$$

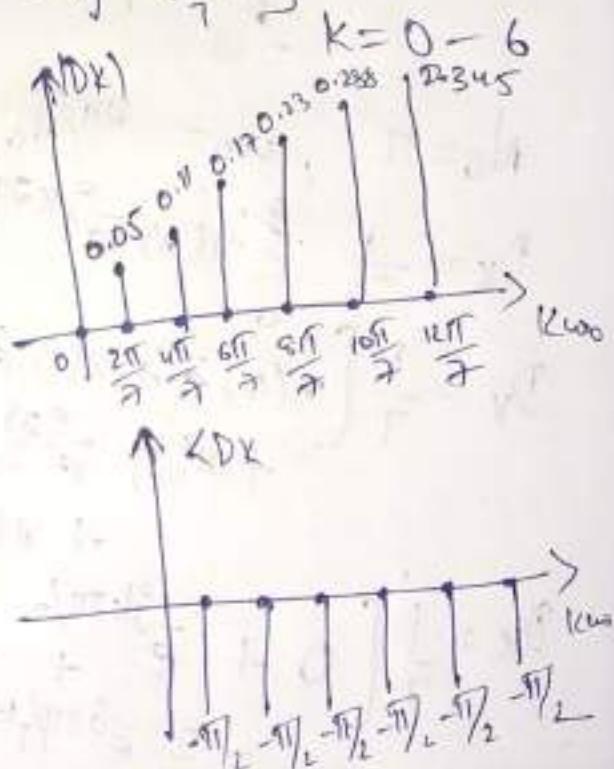
$$D_2 = -0.116j$$

$$D_3 = -0.174j$$

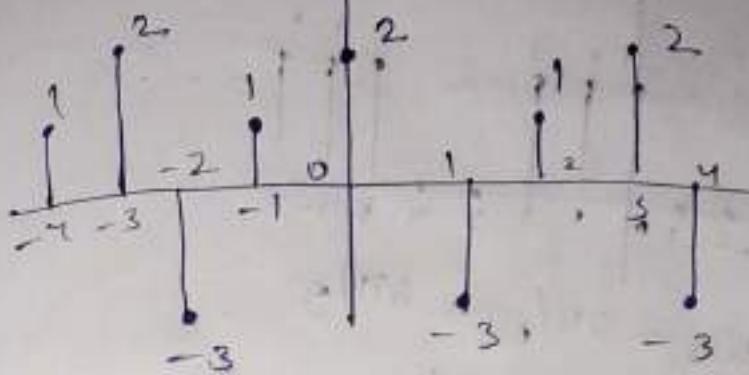
$$D_4 = -0.231j$$

$$D_5 = -0.288j$$

$$D_6 = -0.345j$$



(3)



$$\text{sol } N_0 = 3, \omega_0 = 2\pi/3 = 2\pi/3$$

$$D_K = \frac{1}{N_0} \sum_{n=0}^{N_0-1} x(n) e^{-jkn\omega_0} \quad k = 0 \rightarrow 2.$$

$$D_K = \frac{1}{3} \sum_{n=0}^2 x(n) e^{-jn2\pi/3}$$

$$D_K = \frac{1}{3} [x(0) \bar{e}^0 + x(1) \bar{e}^{-j2\pi/3} + x(2) \bar{e}^{-j4\pi/3}]$$

$$D_K = \frac{1}{3} [2 - 3 \bar{e}^{-j2\pi/3} + 1 \bar{e}^{-j4\pi/3}]$$

$$D_K = \frac{1}{3} [2 - 3 \bar{e}^{-j2\pi/3} - 1 \bar{e}^{j2\pi/3}]$$

$$D_K = \frac{1}{3} [2 - 3 e^{-j2\pi/3} + e^{j2\pi/3}] \quad k=0 \rightarrow 2$$

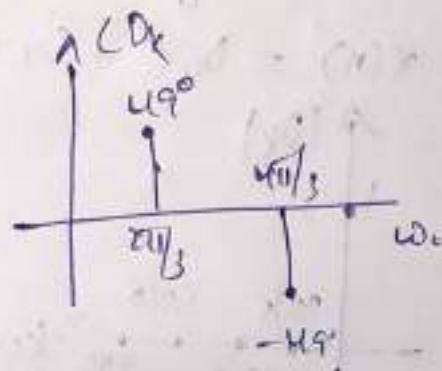
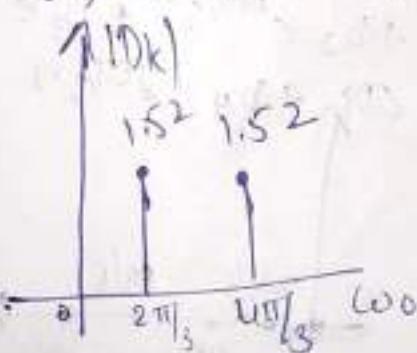
$$D_0 = \frac{1}{3} [2 - 3 + 1] = 0$$

$$D_1 = 1 + 1.5j = 1.52 \angle 49^\circ$$

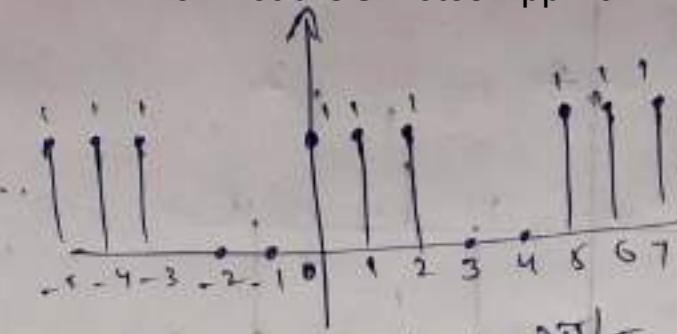
$$D_2 = \frac{1}{3} [2 - 3 (\bar{e}^{j2\pi/3}) + e^{j2\pi/3}]$$

$$= 1 - 1.5j = 1.52 \angle -49^\circ$$

$$x(n) = D_0 e^{j0^\circ} + D_1 e^{j2\pi/3 n} + D_2 e^{j4\pi/3 n}$$



(4)



$$N_0 = 5, \omega_0 = 2\pi/N_0 \approx 2\pi/5$$

$$D_K = \frac{1}{N_0} \sum_{n=0}^{N_0-1} x(n) e^{-jkn2\pi/5}, \quad k = 0 \rightarrow 4$$

$$D_K = \frac{1}{5} [x(0)e^0 + x(1)e^{-j2\pi/5} + x(2)e^{-j4\pi/5} + x(3)e^{-j6\pi/5} + x(4)e^{-j8\pi/5}]$$

$$D_K = \frac{1}{5} \left[1 + e^{-j2\pi k/5} + e^{-j4\pi k/5} + e^{-j6\pi k/5} + e^{-j8\pi k/5} \right], \quad k = 0 \rightarrow 4$$

$$D_K = \frac{1}{5} \left[1 + 2\cos\frac{2\pi k}{5} + 2\cos\frac{4\pi k}{5} \right], \quad k = 0 \rightarrow 4$$

$$D_0 = \frac{1}{5} [1 + 2 + 2] = 1$$

$$D_1 = \frac{1}{5} \left[1 + 2\cos\frac{2\pi}{5} + 2\cos\frac{4\pi}{5} \right] = 0$$

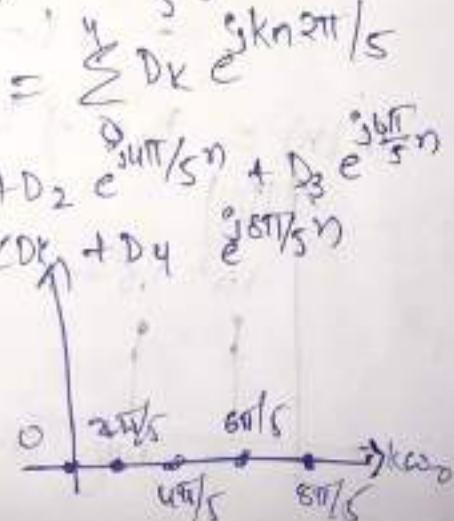
$$D_2 = \frac{1}{5} \left[1 + 2\cos\frac{4\pi}{5} + 2\cos\frac{8\pi}{5} \right] = 0$$

$$D_3 = \frac{1}{5} \left[1 + 2\cos\frac{6\pi}{5} + 2\cos\frac{12\pi}{5} \right] = 0$$

$$D_4 = \frac{1}{5} \left[1 + 2\cos\frac{8\pi}{5} + 2\cos\frac{16\pi}{5} \right] = 0$$

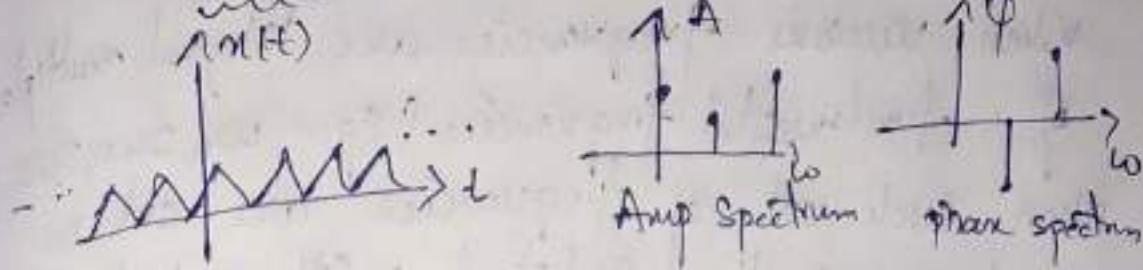
$$x(n) = \sum_{k=0}^{N_0-1} D_k e^{jkn\omega_0} = \sum_{k=0}^{N_0-1} D_k e^{jkn2\pi/5}$$

$$x(n) = D_0 e^0 + D_1 e^{j2\pi/5} n + D_2 e^{j4\pi/5} n + D_3 e^{j6\pi/5} n + D_4 e^{j8\pi/5} n$$



Unit:- 02

Continuous time Fourier Series (CTFS)



Goussier stated that any arbitrary non-sinusoidal signal can be expressed as a weighted sum of sinusoids, each of which has different amplitudes and frequencies which are harmonically related with each other as shown below

$$n(t) = \sum_{k=0}^{\infty} a_k \cos(k\omega_0 t) + b_k \sin(k\omega_0 t)$$

Non-Sinusoid

Trigonometric Fourier Series

$$n(t) = a_0 + \sum_{k=1}^{\infty} a_k \cos(k\omega_0 t) + b_k \sin(k\omega_0 t)$$

D.C component of $n(t)$

$$n(t) = a_0 + (a_1 \cos\omega_0 t + b_1 \sin\omega_0 t) + \\ a_2 \cos 2\omega_0 t + b_2 \sin 2\omega_0 t + \\ (a_3 \cos 3\omega_0 t + b_3 \sin 3\omega_0 t) + \dots$$

Note:-

Harmonic frequencies:-

When various frequencies are integral multiple of fundamental frequency (Eg:- $\omega_0, 2\omega_0, 3\omega_0$, then such set of frequencies are said to be harmonically related with each other.

ω_0 = fundamental frequency

ω_0 = 1st harmonic frequency.

$2\omega_0$ = 2nd harmonic freq

$a_1, b_1 \rightarrow$ Fourier Coefficients of 1st harmonic freq

$a_2, b_2 \rightarrow$ Fourier coeff of 2nd harmonic freq.

$a_1 \cos \omega_0 t + b_1 \sin \omega_0 t \rightarrow$ 1st harmonic

$a_2 \cos 2\omega_0 t + b_2 \sin 2\omega_0 t \rightarrow$ 2nd component

Trigonometric Identities:-

$$1) \int_0^T \cos k\omega_0 t dt = \int_0^T \sin k\omega_0 t dt = 0$$

$$\begin{aligned} \text{Sol} \quad & \int_0^T \cos k\omega_0 t dt = [\sin k\omega_0 t]_0^T \\ &= \sin k\omega_0 T - \sin 0 \\ &= \sin k\omega_0 T - 0 = \frac{\sin k\pi}{T} T \\ &= \sin k\pi = 0 \end{aligned}$$

$$\textcircled{2} \quad \int_{T_0}^T \sin \omega_0 k t \cos \omega_0 m t = 0, \quad \forall k \neq m$$

$$\textcircled{3} \quad \int_{T_0}^T \sin k \omega_0 t \sin m \omega_0 t dt = \int_{T_0}^T \cos k \omega_0 t + \cos m \omega_0 t dt \\ = 0, \quad k \neq m \\ = \frac{\pi}{2}, \quad k = m$$

$$\textcircled{4} \quad \int_{T_0}^T \sin^2 k \omega_0 t dt = \int_{T_0}^T \cos^2 k \omega_0 t dt = \frac{T_0}{2}$$

* Expressions for a_0, a_k, b_k - Trigonometric Fourier coefficients.

(2) Consider the Fourier series

$$x(t) = a_0 + \sum_{k=1}^{\infty} (a_k \cos \omega_0 k t + b_k \sin \omega_0 k t)$$

Integrate both sides w.r.t t
for duration T_0

$$\int_{T_0}^T x(t) dt = a_0 \int_{T_0}^T 1 dt + \sum_{k=1}^{10} \left[a_k \int_{T_0}^T \cos k \omega_0 t dt + b_k \int_{T_0}^T \sin k \omega_0 t dt \right]$$

$$\int_{T_0}^T x(t) dt = a_0 T_0 \quad T_0 = \frac{2\pi}{\omega_0}$$

$$\int_{T_0}^T x(t) dt = a_0 T_0 \\ \Rightarrow a_0 = \frac{1}{T_0} \int_{T_0}^T x(t) dt$$

Note:- The expression $a_0' = \frac{1}{T_0} \int_{t_0}^{t_0} x(t) dt$ dictates that the D.C content in the signal $x(t)$ is simply the average value of the signal.

a_K :-

$$x(t) = a_0 + \sum_{k=1}^{\infty} a_k \cos k\omega_0 t + b_k \sin k\omega_0 t$$

Integrate w.r.t $\cos k\omega_0 t$ / $\sin k\omega_0 t$ over duration $\langle T_0 \rangle$

$$\int_{t_0}^{t_0} x(t) \cos k\omega_0 t dt = a_0 \int_{t_0}^{t_0} \cos k\omega_0 t dt$$

$$+ \sum_{k=1}^{\infty} a_k \int_{t_0}^{t_0} \cos k\omega_0 t \cos k\omega_0 t dt$$

$$+ \sum_{k=1}^{\infty} b_k \int_{t_0}^{t_0} \sin k\omega_0 t \cos k\omega_0 t dt$$

$$\int_{t_0}^{t_0} x(t) \cos k\omega_0 t dt = a_0(0) +$$

$$\sum_{k=1}^{\infty} a_k \int_{t_0}^{t_0} \cos k\omega_0 t \cos k\omega_0 t dt + 0$$

$$\int_{t_0}^{t_0} x(t) \cos k\omega_0 t dt = a_k \int_{t_0}^{t_0} \cos k\omega_0 t \cos k\omega_0 t dt$$

$$+ \sum_{k=1, k \neq m}^{\infty} a_k \int_{t_0}^{t_0} \cos k\omega_0 t \cos m\omega_0 t dt$$

$$\int_{T_0} \pi(t) \cos k\omega_0 t dt = a_k T_0 / 2$$

$$\Rightarrow a_k = \frac{2}{T_0} \int_{T_0} \pi(t) \cos k\omega_0 t dt$$

bx:- multiply B.S with $\sin k\omega_0 t$ & \int_{T_0}

$$\int_{T_0} \pi(t) \sin k\omega_0 t dt = a_0 \int_{T_0} \sin k\omega_0 t dt$$

$$\rightarrow \sum_{k=1}^{\infty} a_k \int_{T_0} \sin k\omega_0 t \cos k\omega_0 t dt$$

$$+ \sum_{k=1}^{\infty} b_k \int_{T_0} \sin k\omega_0 t \sin k\omega_0 t dt$$

$$\int_{T_0} \pi(t) \sin k\omega_0 t dt = 0 + 0 + \sum_{k=1}^{\infty} b_k \int_{T_0} \sin k\omega_0 t \sin k\omega_0 t dt$$

$$\int_{T_0} \pi(t) \sin k\omega_0 t = b_k \int_{T_0} \sin k\omega_0 t \sin k\omega_0 t dt$$

$$+ \sum_{k=1}^{\infty} b_k \int_{T_0} \sin k\omega_0 t \sin k\omega_0 t dt$$

$$= b_k (T_0/2) \rightarrow 0$$

$$b_k = \frac{2}{T_0} \int_{T_0} \pi(t) \sin k\omega_0 t$$

Polar form of Fourier Series / cosine form of Fourier series

$$y(t) = a_0 + \sum_{k=1}^{\infty} \sqrt{a_k^2 + b_k^2} \left[\frac{a_k}{\sqrt{a_k^2 + b_k^2}} \cos \omega_0 t + \frac{b_k}{\sqrt{a_k^2 + b_k^2}} \sin \omega_0 t \right]$$

$$\frac{a_k}{\sqrt{a_k^2 + b_k^2}} = \cos \theta_k \quad \frac{b_k}{\sqrt{a_k^2 + b_k^2}} = \sin \theta_k$$

$$\tan \theta_k = \frac{b_k}{a_k}$$

$$\theta_k = \tan^{-1} \left(\frac{b_k}{a_k} \right)$$

$$A_k = \sqrt{a_k^2 + b_k^2}$$

$$x(t) = a_0 + \sum_{k=1}^{\infty} A_k \left[\cos \theta_k \cos \omega_0 t + \sin \theta_k \sin \omega_0 t \right]$$

$$y(t) = a_0 + \sum_{k=1}^{\infty} A_k \cos(\omega_0 t - \theta_k)$$

$$y(t) = A_0 + \sum_{k=1}^{\infty} A_k \cos(\omega_0 t - \theta_k)$$

$$A_0 = a_0$$

$$A_k = \sqrt{a_k^2 + b_k^2}$$

$$\theta_k = \tan^{-1} \left(\frac{b_k}{a_k} \right)$$

Exponential Fourier Series (EFS)

$$x(t) = \sum_{k=-\infty}^{\infty} C_k e^{j k \omega_0 t}$$

According to this series any arbitrary signal can be expressed as a weighted sum of complex exponentials as follows, each of which are harmonically related to each other.

$$\left[\int_{T_0} e^{j(k-m)\omega_0 t} dt = T_0, \quad k=m \right. \\ \left. = 0, \quad k \neq m \right]$$

$$x(t) = \sum_{k=-\infty}^{\infty} C_k e^{j k \omega_0 t}$$

multiply B.S. with $\overline{e^{j m \omega_0 t}} \times \int dt$

$$\int_{T_0} \overline{e^{j m \omega_0 t}} dt x(t) = \sum_{k=-\infty}^{\infty} C_k \int_{T_0} e^{j k \omega_0 t} e^{-j m \omega_0 t} dt$$

$$\int_{T_0} x(t) \overline{e^{j m \omega_0 t}} dt = \sum_{k=-\infty}^{\infty} C_k \int_{T_0} e^{j k \omega_0 t} e^{-j m \omega_0 t} dt$$

$$\int_{T_0} x(t) \overline{e^{j m \omega_0 t}} dt = C_k \int_{T_0} e^{j k \omega_0 t} dt + \sum_{k=-\infty, k \neq m}^{\infty} C_k \int_{T_0} e^{j k \omega_0 t} e^{-j m \omega_0 t} dt$$

$$\int_{T_0} x(t) \overline{e^{j m \omega_0 t}} dt = C_k T_0$$

$$\boxed{C_k = \frac{1}{T_0} \int_{T_0} x(t) \overline{e^{j m \omega_0 t}} dt}$$

* Relationship b/w T.F & E.F

Consider T.F

$$x(t) = a_0 + \sum_{k=1}^{\infty} a_k \cos \omega k t \rightarrow \sum_{k=1}^{\infty} b_k \sin \omega k t$$

$$x(t) = a_0 + \sum_{k=1}^{\infty} a_k \left(\frac{e^{j\omega k t} + e^{-j\omega k t}}{2} \right) + \sum_{k=1}^{\infty} b_k \left(\frac{e^{j\omega k t} - e^{-j\omega k t}}{2j} \right)$$

$$x(t) = a_0 + \sum_{k=1}^{\infty} \left(\frac{a_k + b_k}{2} \right) e^{j\omega k t} + \sum_{k=1}^{\infty} \left(\frac{a_k - b_k}{2j} \right) e^{-j\omega k t}$$

$$x(t) = a_0 + \sum_{k=1}^{\infty} \left(\frac{a_k - j b_k}{2} \right) e^{j\omega k t} + \sum_{k=1}^{\infty} \left(\frac{a_k + b_k}{2j} \right) e^{-j\omega k t}$$

$$x(t) = a_0 + \sum_{k=1}^{\infty} \left(\frac{a_k - j b_k}{2} \right) e^{j\omega k t} + \sum_{k=-1}^{\infty} \left(\frac{a_k + j b_k}{2} \right) e^{j\omega k t}$$

$$x(t) = a_0 + \sum_{k=1}^{\infty} \left(\frac{a_k - j b_k}{2} \right) e^{j\omega k t} + \sum_{k=-1}^{\infty} \left(\frac{a_{-k} + j b_{-k}}{2} \right) e^{j\omega k t}$$

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{j\omega k t} = c_0 + \sum_{k=1}^{\infty} c_k e^{j\omega k t} + \sum_{k=-1}^{\infty} c_{-k} e^{j\omega k t}$$

$$c_0 = a_0$$

$$c_k = \frac{a_k - j b_k}{2}, \quad c_{-k} = \frac{a_{-k} + j b_{-k}}{2}$$

$$c_{-k} = \frac{a_k + j b_k}{2}$$

$$\boxed{C_{-k} = C_k^*}$$

Notes:-

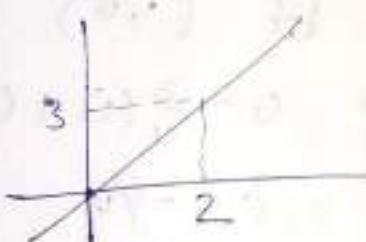
$$\textcircled{1} \cos k\pi = (-1)^k = -1, n \rightarrow \text{odd} - 1, 3, 5, 7, \dots \\ = +1, n \rightarrow \text{even} = 2, 4, 6.$$

$$\textcircled{2} \sin k\pi = 0, \forall k$$

$$\textcircled{3} \cos \frac{k\pi}{2} = 0, \cancel{\text{for } k=1, 3, 5} \\ = 1, \cancel{\text{for } k=2, 6, 10} \\ = -1, \cancel{\text{for } k=4, 8, 12} \\ k=1, 5, 9, 13$$

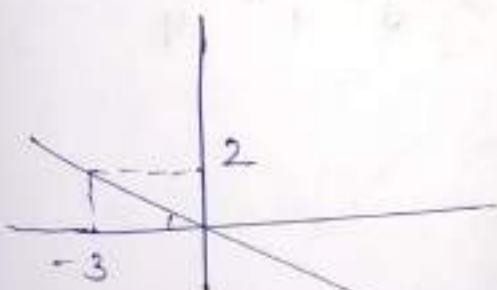
$$\textcircled{4} \sin \frac{k\pi}{2} = 1, \\ = -1, k=3, 7, 11, 15 \\ = 0, k=\text{even}$$

Co-ordinate Geometry :-

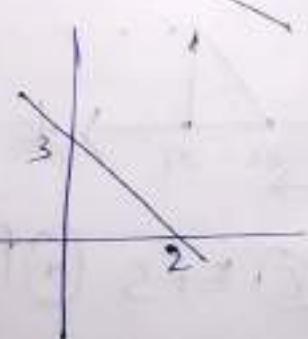


$$y = mx$$

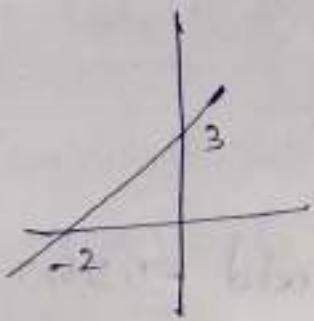
$$y = \frac{3}{2}x$$



$$y = -\frac{2}{3}x$$



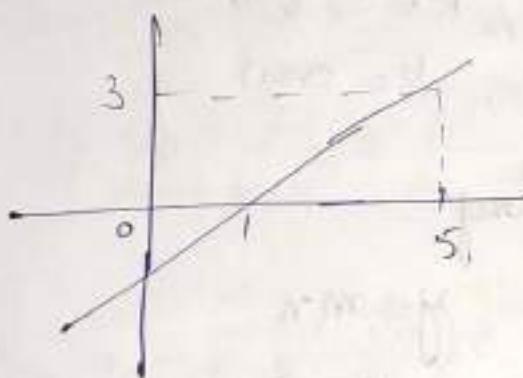
$$y = mx + c \\ y = -\frac{3}{2}x + 3$$



$$y = \frac{3}{2}x + 3$$



$$y = \frac{3}{2}x - 3$$



$$y = mx + c$$

$$m = \frac{3}{4}$$

$$y = \frac{3}{4}x + c$$

$$\text{lit } (1, 0)$$

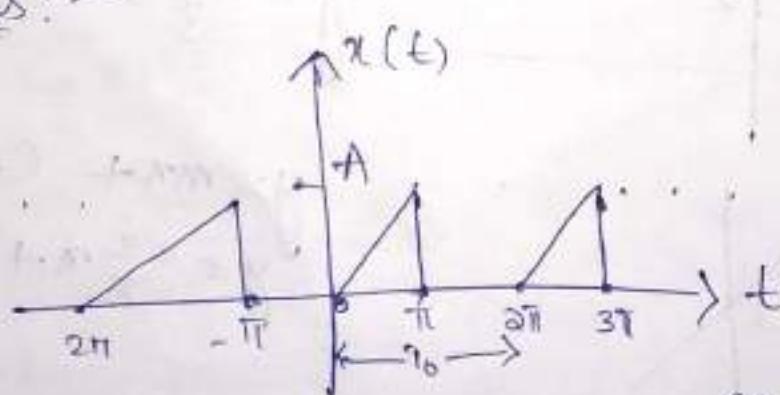
$$0 = \frac{3}{4}(1) + c$$

$$c = -\frac{3}{4}$$

$$\Rightarrow y = \frac{3}{4}x - \frac{3}{4}$$

Problems:

①



- ① TFS ② PFS ③ EFS ④ Fourier spectrum

$$T_0 = 2\pi, \omega_0 = \frac{2\pi}{T_0} = \frac{2\pi}{2\pi} = 1$$

$$\boxed{\omega_0 = 1}$$

we have $y = mt$

$$n(t) = mt$$

$$n(t) = \frac{A}{\pi}t, 0 \leq t \leq \pi$$

$$= 0, \pi \leq t \leq 2\pi$$

$$a_0 = \frac{1}{T_0} \int_{T_0}^0 n(t) dt$$

$$a_0 = \frac{1}{2\pi} \int_{2\pi}^0 n(t) dt$$

$$\Rightarrow \frac{1}{2\pi} \int_0^0 n(t) = \frac{1}{2\pi} \left[\int_0^\pi n(t) dt + \int_\pi^{2\pi} n(t) dt \right]$$

$$= \frac{1}{2\pi} \left[\int_0^\pi \frac{A}{\pi} t dt + 0 \right]$$

$$= \frac{A}{2\pi^2} \int_0^\pi t dt$$

$$= \frac{A}{2\pi^2} \left[\frac{t^2}{2} \right]_0^\pi$$

$$= \frac{A}{2\pi^2} \left[\frac{\pi^2}{2} \right] = A/y$$

$$\boxed{a_0 = A/y}$$

$$a_k = \frac{2}{\pi} \int_{-\pi}^{\pi} x(t) \cos k \omega_0 t \, dt, \quad \omega_0 = 1$$

$$a_k = \frac{2}{\pi} \int_0^{2\pi} x(t) \cos k \omega_0 t \, dt$$

$$a_k = \frac{1}{\pi} \int_0^{\pi} A + \cos kt \, dt + \frac{1}{\pi} \int_{\pi}^{2\pi} 0 \cos kt \, dt$$

$$a_k = \frac{A}{\pi^2} \int_0^{\pi} \cos kt \, dt \quad \text{STATE}$$

$$a_k = \frac{A}{\pi^2} \left[t \left(\frac{\sin kt}{k} \right) \Big|_0^{\pi} - \int_0^{\pi} \frac{\sin kt}{k} \, dt \right]$$

$$a_k = \frac{A}{\pi^2} \left[\frac{t}{k} (\sin k\pi - \sin 0) - \frac{1}{k} \left[\frac{\sin kt}{k} \Big|_0^{\pi} \right] \right]$$

$$a_k = \frac{A}{\pi^2} \left[t \frac{\sin kt}{k} + \frac{\cos kt}{k^2} \Big|_0^{\pi} \right]$$

$$a_k = \frac{A}{\pi^2 k^2} \left[\frac{\cos \pi k}{k} - \frac{\cos 0}{k} \right]$$

$$a_k = \frac{A}{\pi^2 k^2} \left[-1 - \frac{1}{k} \right], \quad k = \text{odd}$$

$$a_k = \frac{A}{\pi^2 k^2} \cdot [+1 - 1], \quad k = \text{even}$$

$$\left. \begin{aligned} a_k &= \frac{-2A}{\pi^2 k^2}, & k &= \text{odd} \\ a_k &= 0, & k &= \text{even} \end{aligned} \right\}$$

$$b_K = \frac{A}{\pi^2} \int_0^\pi t \sin kt dt \Big|_{\pi}$$

$$b_K = \frac{A}{\pi^2} \left[t \left[-\frac{\cos kt}{k} \right] - \int_0^\pi -\frac{\cos kt}{k} dt \right] \Big|_{\pi}$$

$$b_K = \frac{A}{\pi^2} \left[-t \frac{\cos kt}{k} + \frac{\sin kt}{k^2} \right] \Big|_{\pi}$$

$$b_K = \frac{A}{\pi^2} \left[-\frac{\pi \cos \pi k}{k} + 0 \right]$$

$$b_K = \frac{-A \cos \pi k}{\pi k}$$

$$\begin{cases} b_K = \frac{A}{\pi k}, k = \text{odd } 1, 3, 5, 7, \\ b_K = \frac{-A}{\pi k}, k = \text{even } 2, 4, 6, 8 \end{cases}$$

$$x(t) = a_0 + \sum_{K=1}^{\infty} a_K \cos \omega_0 k t + b_K \sin \omega_0 t$$

$$x(t) = \frac{A}{4} + \sum_{\substack{K=1, 3, 5 \\ K=\text{odd}}}^{\infty} -\frac{2A}{\pi^2 k^2} \cos kt + \sum_{\substack{K=1 \\ K=\text{odd}}}^{\infty} \frac{A}{\pi k} \sin kt$$

$$+ \sum_{\substack{K=2 \\ K=\text{even}}}^{\infty} -\frac{A}{\pi k} \sin kt$$

$$x(t) = \frac{A}{4} + \sum_{\substack{K=1 \\ K=\text{odd}}}^{\infty} \sqrt{\frac{4A^2}{\pi^4 k^4} + \frac{A^2}{\pi^2 k^2}} \cos \left(Kt - \tan^{-1} \left(\frac{\pi k}{-2A/\pi k^2} \right) \right)$$

$$+ \sum_{K=2}^{\infty} \frac{A}{\pi k} \cos \left(Kt - \frac{\pi}{2} \right)$$

$$x(t) = \frac{A}{4} + \sum_{\substack{K=1 \\ K=\text{odd}}}^{\infty} \sqrt{\frac{4A^2}{\pi^4 k^4} + \frac{A^2}{\pi^2 k^2}} \cos \left(Kt - \tan^{-1} \left(\frac{\pi k}{2} \right) \right)$$

$$+ \sum_{\substack{K=2 \\ K=\text{even}}}^{\infty} \frac{A}{\pi k} \cos \left(Kt - \frac{\pi}{2} \right)$$

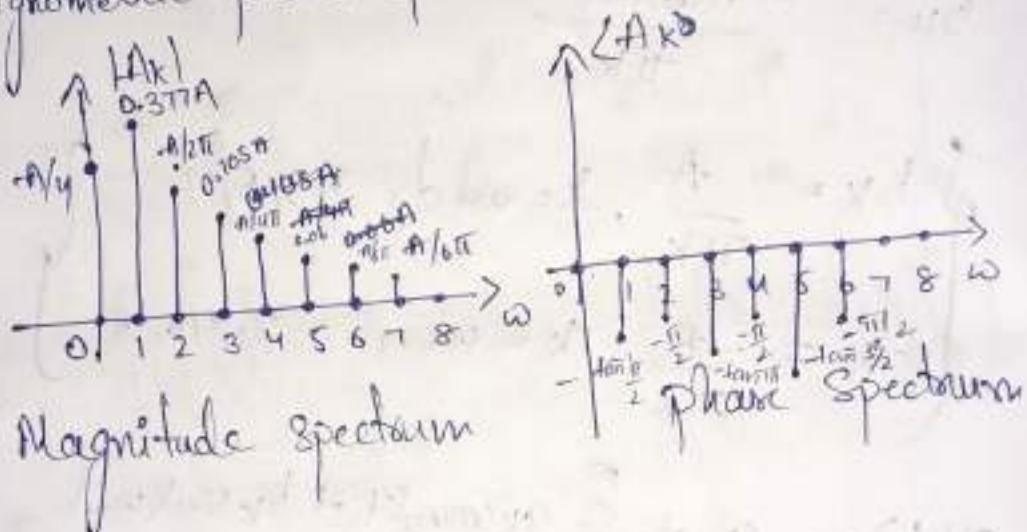
$$\sqrt{\frac{4A^2}{\pi^4 k^4} + \frac{A^2}{\pi^2 k^2}} = \frac{A}{\pi k} \sqrt{\frac{4}{\pi^2 k^2} + 1}$$

$$k=1, \frac{A}{\pi} \sqrt{\frac{4}{\pi^2} + 1} = 0.377A$$

$$k=3, \frac{A}{3\pi} \sqrt{\frac{4}{9\pi^2} + 1} = 0.108A$$

$$k=5, \frac{A}{5\pi} \sqrt{\frac{4}{25\pi^2} + 1} = 0.06A$$

Trigonometric Fourier Spectrum



(c) EFS

$$x(t) = \sum_{k=-\infty}^{\infty} C_k e^{jkw_0 t}$$

$$x(t) = C_0 + \sum_{k=-\infty, k \neq 0}^{\infty} C_k e^{jkw_0 t}$$

$$x(t) = \frac{A}{4} + \sum_{k=-\infty, k \neq 0}^{\infty} C_k e^{jkw_0 t}$$

$$C_k = \frac{a_k - j b_k}{2}$$

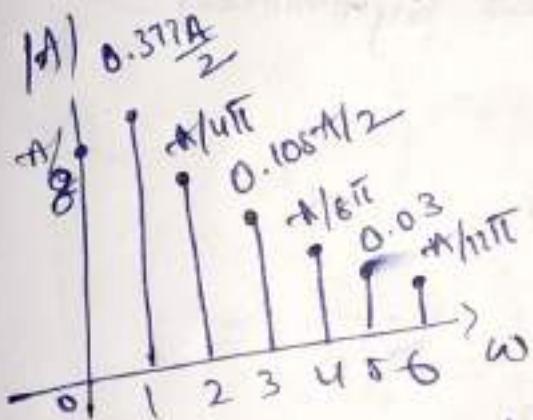
$$C_k = \frac{-2A}{\pi^2 k^2 + \pi^2} \cdot \frac{jA}{\pi k}, \quad k = \text{odd}$$

$$C_K = \frac{-A}{\pi^2 k^2} - \frac{3A}{2\pi k}, \quad k=\text{odd}$$

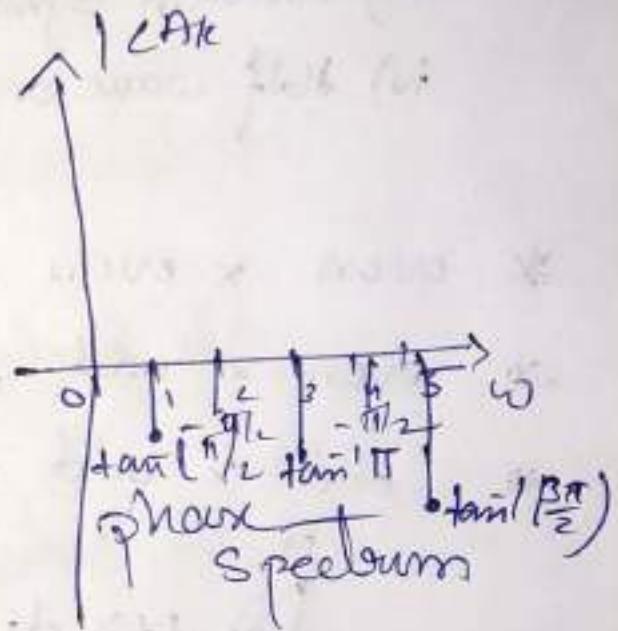
$$C_K = \begin{cases} 0 & k-\text{even} \\ -\frac{3(-A/\pi k)}{2} & k-\text{odd} \end{cases}$$

$$C_K = \frac{3A}{2\pi k}, \quad k=\text{even}$$

$$x(t) = \frac{A}{4} + \sum_{k=1,3,5}^{\infty} \left(\frac{-A}{\pi^2 k^2} - \frac{3A}{2\pi k} \right) e^{jkt} + \sum_{k=2,4,6}^{\infty} \frac{3A}{2\pi k} e^{jkt} \\ + \sum_{k=-1,-3,-5}^{-\infty} \left(\frac{-A}{\pi^2 k^2} + \frac{3A}{2\pi k} \right) e^{-jkt} + \sum_{k=2,4,6}^{\infty} \left(\frac{-3A}{2\pi k} \right) e^{-jkt}$$



magnitude spectrum



Effect of Symmetry on Fourier Series.

→ It is observed that signals with some kind of symmetry will have a relatively easier Fourier series analysis.

→ There are 4 kinds of symmetry

- * i) Even Symmetry

- * ii) Odd Symmetry

- * iii) Hidden Symmetry

- * iv) Half wave or Rotational Symmetry

* even × even = even

* even × odd = odd

* odd × odd = even

$$\int_{-\pi/2}^{\pi/2} x_e(t) dt = 2 \int_0^{\pi/2} x_e(t) dt$$

$$\int_{-\pi/2}^{\pi/2} x_o(t) dt = 2 \int_0^{\pi/2} x_o(t) dt$$

$$\int_{-\pi/2}^{\pi/2} x_h(t) dt = 0$$

$$\int_{-\pi/2}^{\pi/2} x_r(t) dt = 0$$

① Even Symmetry

Let $x(t)$ be even function

$$a_0 = \frac{1}{T_0} \int_0^{T_0} x(t) dt$$

$$= \frac{1}{T_0} 2 \int_0^{\pi_0/2} x(t) dt$$

$$\left[a_0 = \frac{2}{T_0} \int_0^{\pi_0/2} x(t) dt \right] \text{ even function}$$

$$a_k = \frac{2}{T_0} \int_0^{\pi_0} x(t) \cos kt dt$$

$$= \frac{2}{T_0} 2 \int_0^{\pi_0/2} x(t) \cos kt dt$$

$$\left\{ a_k = \frac{2}{T_0} \int_0^{\pi_0/2} x(t) \cos kt dt \right\}$$

$$b_k = \frac{2}{T_0} \int_0^{\pi_0} x(t) \sin kt dt$$

$$\boxed{b_k = 0} \quad \text{even} \times \text{odd} = \text{odd}$$

$$a_0 = \frac{2}{T_0} \int_0^{\pi_0/2} x(t) dt \quad \boxed{x(t) = a_0 + \sum_{k=1}^{\infty} a_k \cos kt}$$

$$a_k = \frac{2}{T_0} \int_0^{\pi_0/2} x(t) \cos kt dt$$

$$b_k = 0 \quad \therefore \text{For even periodic signal will have only DC & cosine terms in its F.S expression}$$

② Odd Symmetry

If $x(t)$ be odd function.

$$a_0 = \frac{1}{T_0} \int_0^{T_0} x(t) dt$$

Odd

$$a_0 = \frac{1}{T_0} \{ 0 \} \Rightarrow \boxed{a_0 = 0}$$

$$a_k = \frac{2}{T_0} \int_0^{T_0} x(t) \cos k\omega_0 t dt$$

$$a_k = \frac{2}{T_0} \int_0^{T_0} \text{odd} dt = 0$$

$$\boxed{a_k = 0}$$

$$b_k = \frac{2}{T_0} \int_0^{T_0} x(t) \sin k\omega_0 t dt$$

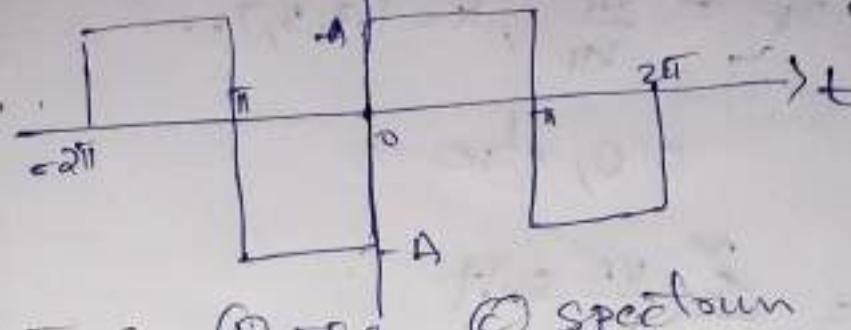
$$= \frac{2}{T_0} \cdot 2 \int_0^{\pi/2} x(t) \sin k\omega_0 t dt$$

$$b_k = \frac{4}{T_0} \int_0^{\pi/2} x(t) \sin k\omega_0 t dt$$

$$\boxed{x(t) = \sum_{k=1}^{\infty} b_k \sin k\omega_0 t}$$

∴ odd periodic signal will have
only b_k term

(B)



- (a) TFS (b) EFS (c) spectrum

If the above signal is odd.

then $a_0 = a_{k=0} = 0$

$$b_k = \frac{4}{T_0} \int_0^{\pi/2} x(t) \sin kt dt$$

because $T_0 = 2\pi$, $\omega_0 = \frac{2\pi}{T_0} = 1$

$$\boxed{x(t) = A, 0 \leq t \leq \pi}$$

$$b_k = \frac{4}{2\pi} \int_0^\pi A \sin kt dt$$

$$= \frac{2A}{\pi} \left[-\frac{\cos kt}{k} \right]_0^\pi$$

$$= \frac{2A}{k\pi} [-\cos k\pi + \cos 0]$$

$$b_k = -\frac{2A}{k\pi} [\cos k\pi - 1]$$

when $k = \text{odd}$

$$b_k = -\frac{2A}{k\pi} [-1 - 1] = \frac{4A}{k\pi}$$

when $k = \text{even}$

$$b_k = -\frac{2A}{k\pi} [1 - 1] = 0$$

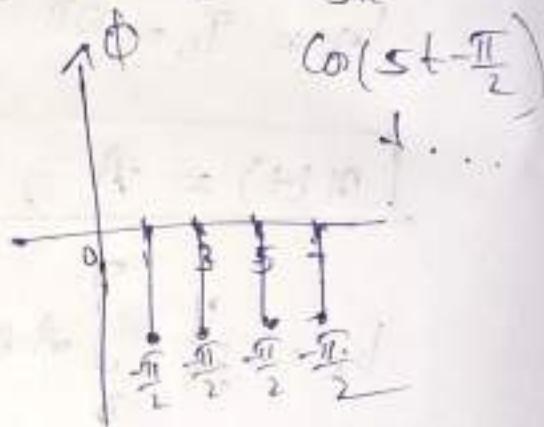
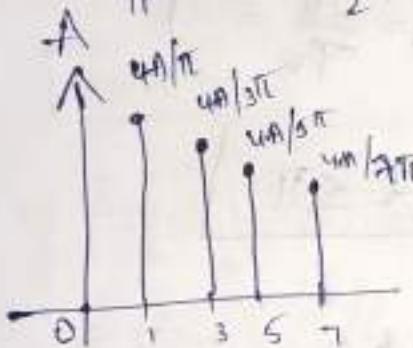
$$\Rightarrow b_k = \frac{4A}{k\pi}, k=1, 3, 5, 7, \dots \\ = 0, \text{ else}$$

$$n(t) = \sum_{k=1,3,5}^{\infty} \frac{4A}{k\pi} \sin kt$$

$$n(t) = \frac{4A}{\pi} \sin t + \frac{4A}{3\pi} \sin 3t + \frac{4A}{5\pi} \sin 5t + \dots$$

$$\boxed{\sin kt = \cos(kt - 90^\circ)}$$

$$n(t) = \frac{4A}{\pi} \cos\left(t - \frac{\pi}{2}\right) + \frac{4A}{3\pi} \cos\left(t - \frac{3\pi}{2}\right) + \frac{4A}{5\pi} \cos\left(t - \frac{5\pi}{2}\right)$$

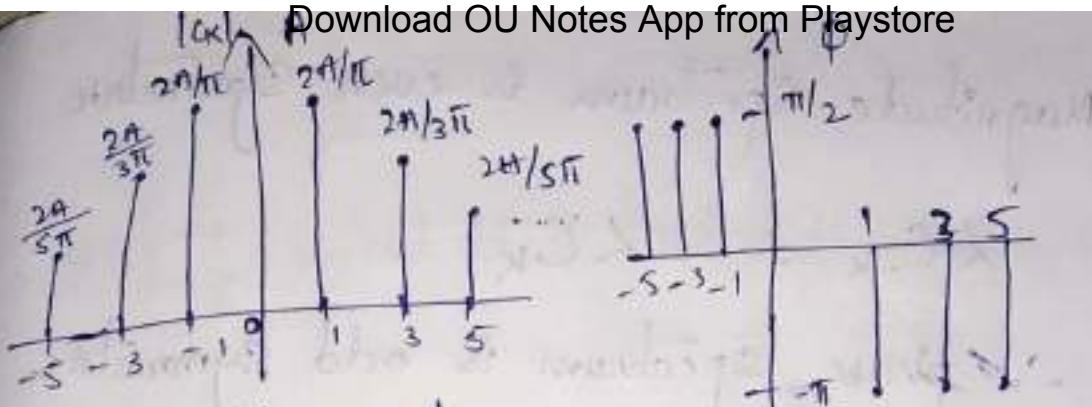


Spectrum

$$\text{E.F.S: } n(t) = \sum_{k=-\infty}^{\infty} C_k e^{jkt}$$

$$C_k = \frac{a_k - j b_k}{2} = -j \frac{b_k}{2} = -j \frac{4A}{2k\pi}, k=\text{odd} \\ = 0, k=\text{even}$$

$$C_{-k} = C_k^* = \frac{j2A}{k\pi} \quad n(t) = \sum_{k=-1}^{\infty} \frac{2A}{k\pi} e^{jkt} + \sum_{k=1}^{\infty} \frac{-2A}{k\pi} e^{-jkt}$$



Amplitude spectrum

phase spectrum

→ Amplitude Spectrum is always even.

Symmetric

→ phase spectrum is always odd symmetric

* Prove that Amplitude Spectrum is even
symmetric & phase Spectrum is odd
symmetric for a real signal

sol As it is known that C_k is in general
is complex, can be represented as

$$C_k = a + jb \quad |C_k| = \sqrt{a^2 + b^2}$$

$$\arg C_k = \tan^{-1}(b/a)$$

then

$$C_{-k} = a - jb = C_k^*$$

$$|C_{-k}| = \sqrt{a^2 + b^2}$$

$$\arg C_{-k} = -\tan^{-1}(b/a)$$

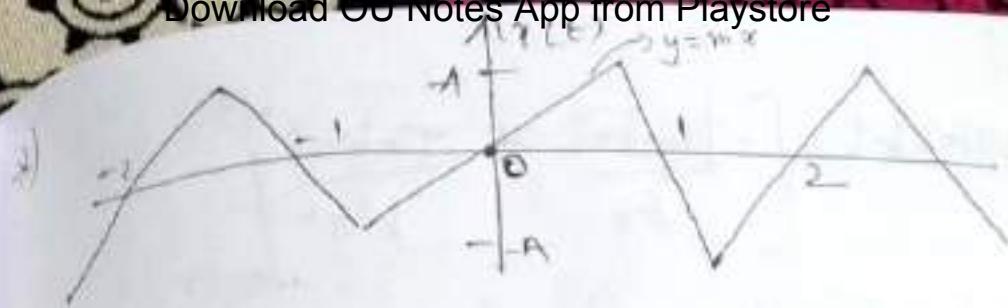
$$\therefore |C_k| = |C_{-k}|$$

Magnitude spectrum is even symmetric

$$\angle C_{-k} = -\angle C_k$$

∴ phase spectrum is odd symmetric

odd
val



repeated

odd symmetric

$$a_0 = a_k = 0$$

$$b_k = \frac{4}{T_0} \int_{0}^{T_0/2} x(t) \sin k\pi t dt$$

$$s) T_0 = 2 \quad \omega_0 = \frac{2\pi}{2} = \pi$$

$$x(t) = 2At, \quad 0 \leq t \leq \frac{1}{2}$$

$$x(t) = -2At + 2A$$

a)

$$= 2A(1-t), \quad \frac{1}{2} \leq t \leq 1$$

$$y = mx \\ y = \frac{A}{V_2} x$$

$$y = mx + c$$

$$y = -2At + C$$

$$0 = -2A(1) + C$$

$$C = 2A$$

x(t)

$$= 2 \left[\int_0^{\pi/2} 2At \sin kt dt + \int_{\pi/2}^1 2A(1-t) \sin kt dt \right]$$

sc

$$= 4A \left[\int_0^{\pi/2} t \sin kt dt + \int_{\pi/2}^1 \sin kt - \int_{\pi/2}^1 t \sin kt dt \right]$$

$$\int_{\pi/2}^1 \sin kt dt = \left[\frac{-\cos kt}{k\pi} \right]_{\pi/2}^1 = \frac{1}{k\pi} \left[\cos \frac{k\pi}{2} - \cos k\pi \right]$$

$$\int_0^{\pi/2} t \sin kt dt = -\frac{\cos kt}{k\pi} - \int_0^{\pi/2} -\frac{\cos kt}{k\pi} dt$$

$$\int_0^{t_2} t \sin kt = \left[-\frac{t \cos kt}{k\pi} + \frac{\sin kt}{k^2\pi^2} \right]_0^{t_2}$$

$$= \pi \left\{ \cancel{\left[\frac{-t \cos kt}{2k\pi} \right]} - \left(\frac{1}{2} \cos \frac{k\pi}{2} + 0 \right) + \frac{1}{k\pi} \left[\frac{\sin kt}{2} \right]_{0^+}^{t_2} \right\}$$

$$= -\frac{\cos k\pi/2}{2k\pi} + \frac{\sin k\pi/2}{k^2\pi^2}$$

$$\int_{t_2}^1 t \sin kt = \left. -\frac{t \cos kt}{k\pi} + \frac{\sin kt}{k^2\pi^2} \right|_{t_2}^1$$

$$= -\left(\frac{1}{k\pi} \left(\cos k\pi - \frac{1}{2} \cos \frac{k\pi}{2} \right) + \frac{1}{k^2\pi^2} \left(\sin k\pi - \sin \frac{k\pi}{2} \right) \right)$$

$$= \frac{\cos k\pi}{2k\pi} - \frac{\cos k\pi}{k\pi} + \frac{\sin k\pi}{k^2\pi^2} - \frac{\sin \frac{k\pi}{2}}{k^2\pi^2}$$

$$b_k = 4A \left[-\frac{\cos k\pi/2}{2k\pi} + \frac{\sin k\pi/2}{k^2\pi^2} + \frac{\cos k\pi}{2k\pi} - \frac{\cos k\pi}{k\pi} + \right. \\ \left. -\frac{\cos k\pi/2}{2k\pi} + \frac{\cos k\pi}{k\pi} - \frac{\sin k\pi}{k^2\pi^2} + \frac{\sin \frac{k\pi}{2}}{k^2\pi^2} \right]$$

$$b_k = 4A \left[\frac{2 \sin k\pi/2}{k^2\pi^2} - \frac{\sin k\pi}{k^2\pi^2} \right]$$

$$b_k = 4A \left(\frac{2 \sin k\pi/2}{k^2 \pi^2} - 0 \right)$$

$$= \frac{8A \sin \frac{k\pi}{2}}{k^2 \pi^2}, \quad k = 1, 5, 9, 13, \dots$$

$$= -\frac{8A}{k^2 \pi^2}, \quad k = 3, 7, 11, 15, \dots$$

$$= 0, \quad k = \text{even}$$

$$v(t) = \sum_{k=1}^{\infty} b_k c_k \sin kt$$

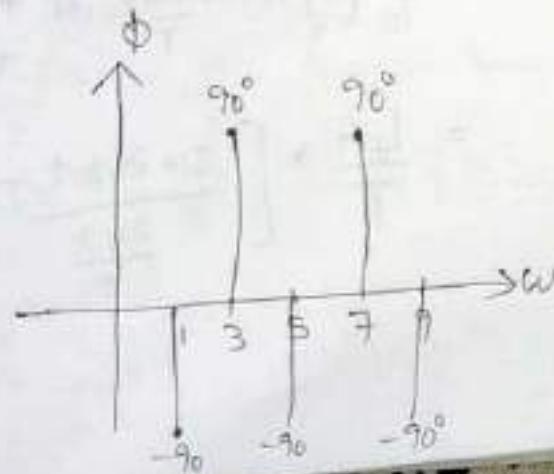
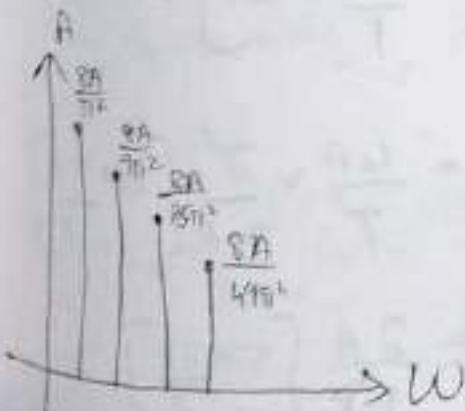
$$= \frac{8A}{\pi^2} \sin \pi t + \frac{8A}{9\pi^2} \sin 3\pi t + \frac{8A}{25\pi^2} \sin 5\pi t + \dots$$

$$\frac{8A}{49\pi^2} \sin 7\pi t + \dots$$

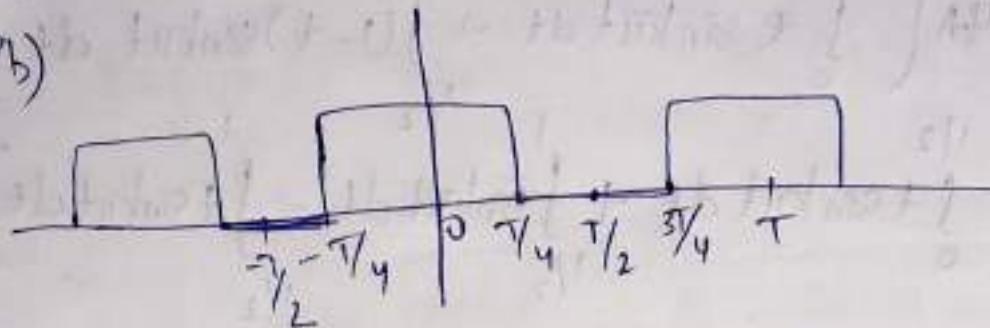
$$v(t) = \frac{8A}{\pi^2} \cos(\pi t - \varphi_0) + \frac{8A}{9\pi^2} \cos(3\pi t + \varphi_0)$$

$$\begin{aligned} \sin kt &= \cos(kt + \pi/2) \\ \sin kt &= \cos(kt - \pi/2) \end{aligned}$$

$$+ \frac{8A}{25\pi^2} \cos(5\pi t - \varphi_0) + \frac{8A}{49\pi^2} \cos(7\pi t + \varphi_0) + \dots$$



Pb)



Sol $\because n(t)$ is even symmetric
 $\therefore b_k = 0$

$$T_0 = T \quad \omega_0 = \frac{2\pi}{T_0} = \frac{2\pi}{T}$$

$$\begin{aligned} n(t) &= A & 0 \leq t \leq T/4 \\ &= 0 & T/4 \leq t \leq T/2 \end{aligned}$$

$$\begin{aligned} a_0 &= \frac{2}{T_0} \int_0^{T_0/2} n(t) dt \\ &= \frac{2}{T_0} \int_0^{T/2} n(t) dt \\ &= \frac{2}{T_0} \left[\int_0^{T/4} A dt + \int_{T/4}^{T/2} 0 \cdot dt \right] \end{aligned}$$

$$a_0 = \frac{2A}{4} = \frac{A}{2}$$

$a_0 = A/2$

$$a_k = \frac{4}{T_0} \int_0^{T_0/2} x(t) \cos k\omega_0 t dt.$$

$$= \frac{4}{T_0} \left[\int_0^{T_0/4} A \cos \frac{2\pi}{T} t dt + \int_{T_0/4}^{T_0/2} 0 \cdot \cos \frac{2\pi}{T} t dt \right]$$

$$= \frac{4A}{T_0} \left[\frac{\sin 2\pi k t}{T} \Big|_0^{T_0/4} \right]$$

$$= \frac{2A}{\pi k} \times \frac{T}{2\pi k} \left[\sin \frac{\pi k}{2} \right]$$

$$= \frac{2A}{\pi k} \left[\sin \frac{\pi k}{2} \right]$$

$$\sin \frac{\pi k}{2} = +1, \quad k = 1, 5, 9, 13$$

$$= -1, \quad k = 3, 7, 11, 15$$

$$a_k = \frac{2A}{\pi k}, \quad k = 1, 5, 9, 13$$

$$= -\frac{2A}{\pi k}, \quad k = 3, 7, 11, 15$$

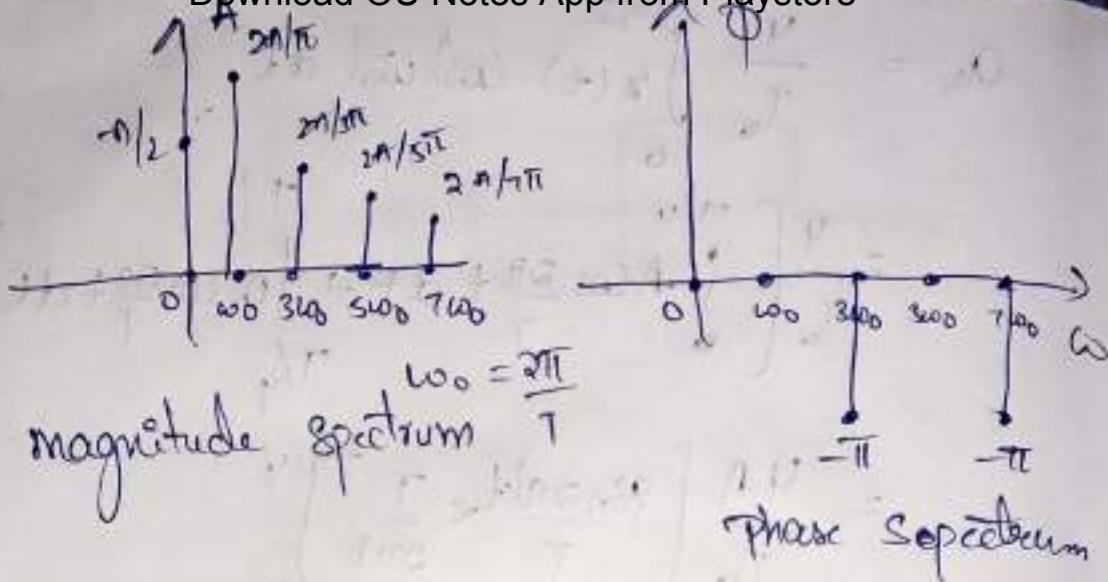
$$= 0, \quad k = 2, 4, 6, \dots$$

$$x(t) = a_0 + \sum_{k=1}^{\infty} a_k \cos \omega_0 t$$

$$= \frac{A}{2} + \frac{2A}{\pi} \cos \frac{2\pi}{T} t - \frac{2A}{3\pi} \cos \frac{3\pi}{T} t$$

$$+ \frac{2A}{5\pi} \cos \frac{5\pi}{T} t - \frac{2A}{7\pi} \cos \frac{7\pi}{T} t$$

$$x(t) = a_0 + \frac{2A}{\pi} \cos \omega_0 t + \frac{2A}{3\pi} \cos (3\omega_0 t - \pi) + \frac{2A}{5\pi} \cos (5\omega_0 t) + \frac{2A}{7\pi} \cos (7\omega_0 t - \pi)$$

E.F.s

$$v(t) = \sum_{-\infty}^{\infty} C_k e^{j k \omega_0 t}, \quad \omega_0 = \frac{2\pi}{T}$$

$$C_0 = a_0 = A/2$$

$$C_k = \frac{a_k - j b_k}{2} = \frac{A}{\pi k}, \quad k=1, 3, 5, \dots$$

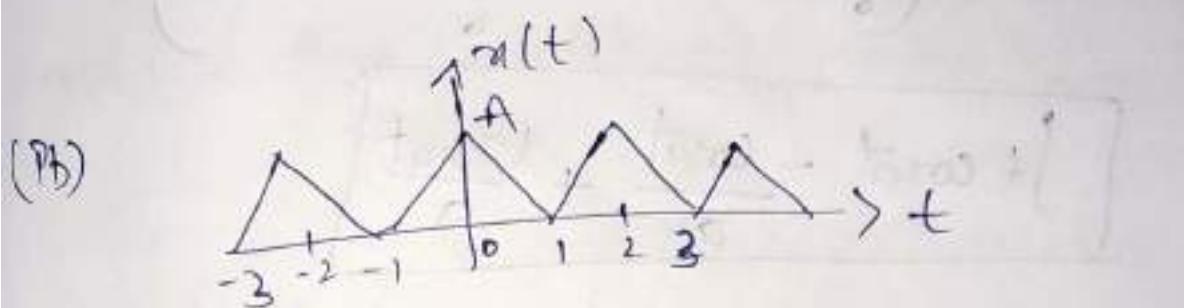
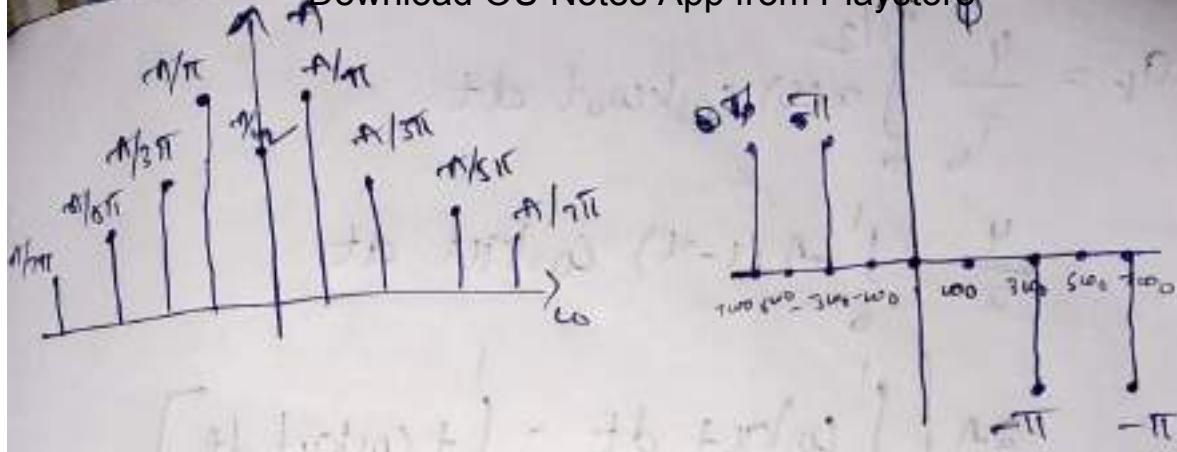
$$= -\frac{A}{\pi k}, \quad k=3, 7, 11, \dots$$

$$v(t) = C_0 + \sum_{k=-\infty}^{-1} C_k e^{jk\omega_0 t} + \sum_{k=1}^{\infty} C_k e^{jk\omega_0 t}$$

$$v(t) = \frac{A}{2} + \sum_{k=-\infty} C_k e^{jk\omega_0 t} + \sum_{k=1, 3, 5}^{\infty} \left(\frac{A}{\pi k}\right) e^{jk\omega_0 t}, \quad \omega_0 = \frac{2\pi}{T}$$

$$= \frac{A}{2} + \left(\sum_{k=-\infty}^{-1} \left(\frac{A}{\pi k}\right)\right) e^{jk\omega_0 t} + \left(\sum_{k=1, 3, 5}^{\infty} \left(\frac{A}{\pi k}\right)\right) e^{jk\omega_0 t}$$

$$= \frac{A}{2} + \frac{A}{\pi} e^{j\omega_0 t} + \frac{A}{3\pi} e^{j3\omega_0 t} + \frac{A}{5\pi} e^{j5\omega_0 t} + \dots + \left(-\frac{A}{\pi} e^{-j\omega_0 t} - \frac{A}{3\pi} e^{-j3\omega_0 t} - \frac{A}{5\pi} e^{-j5\omega_0 t}\right)$$



sol Even Symmetric
 $b_k = 0$, $a_0 = ?$, $a_k = ?$

$$T_0 = 2, \omega_0 = \pi$$

$$x(t) = At + A, 0 \leq t \leq \pi/2$$

$$= A(1-t), 0 < t < \pi/2$$

$$a_0 = \frac{2}{T_0} \int_0^{T_0/2} x(t) dt = \frac{2}{2} \int_0^1 x(t) dt$$

$$= A \int_0^1 1-t dt$$

$$= A \left[t - \frac{t^2}{2} \right]_0^1$$

$$\boxed{a_0 = -A/2}$$

$$a_k = \frac{4}{T_0} \int_{0}^{T_0/2} u(t) \cos k\omega_0 t \, dt$$

$$\frac{4}{2} \cdot \int_0^{\frac{T_0}{2}} A(1-t) \cos k\pi t \, dt$$

$$2A \left(\int_0^{\frac{T_0}{2}} \cos k\pi t \, dt - \int_0^{\frac{T_0}{2}} t \cos k\pi t \, dt \right)$$

$$\boxed{\int t \cos at = \frac{\cos at}{a^2} + \frac{t \sin at}{a}}$$

$$a_k = 2A \left[\frac{\sin k\pi t}{k\pi} \Big|_0^{\frac{T_0}{2}} - \left[\frac{\cos k\pi t}{k\pi} + \frac{t \sin k\pi t}{k\pi} \Big|_0^{\frac{T_0}{2}} \right] \right]$$

$$= 2A \left[\frac{2 \sin k\pi}{k\pi} - \left[\frac{\cos k\pi - \cos 0}{k^2\pi^2} + \left[\frac{\sin k\pi - 0 \sin 0}{k\pi} \right] \right] \right]$$

$$a_k = -\frac{2A}{k^2\pi^2} (\cos k\pi - 1)$$

$k = \text{odd}$

$$a_k = -\frac{2A}{k^2\pi^2} (-1 - 1) = \frac{4A}{k^2\pi^2}$$

$k = \text{even}$

$$a_k = -\frac{2A}{k^2\pi^2} (1 - 1) = 0$$

$$a_k = \frac{4A}{k^2\pi^2}, k=1, 3, 5$$

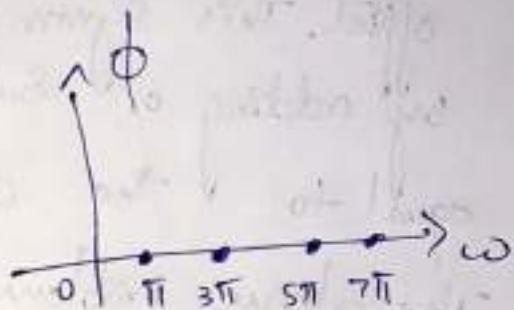
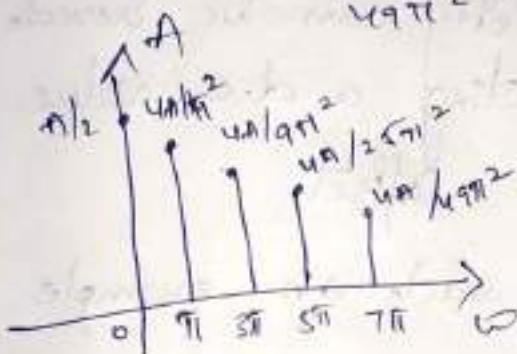
$= 0, k = \text{even}$

$$a_0 = A/2$$

$$b_k = 0$$

$$v(t) = \frac{A}{2} + \frac{4A}{\pi^2} \cos \pi t + \frac{4A}{9\pi^2} \cos 3\pi t + \frac{4A}{25\pi^2} \cos 5\pi t$$

$$+ \frac{4A}{49\pi^2} \cos 7\pi t$$



E.F.S

$$v(t) = \sum_{k=-\infty}^{\infty} c_k e^{j\omega_0 kt}, \omega_0 = 2\pi/f_1$$

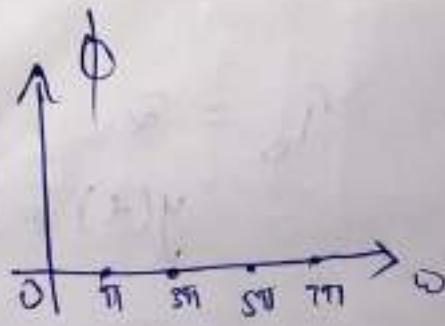
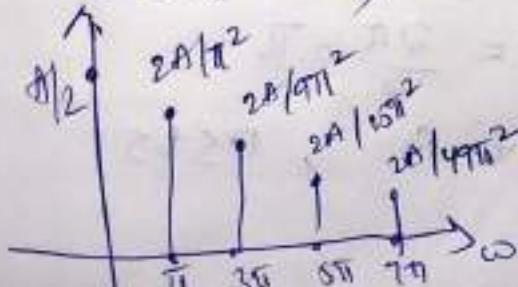
$$c_0 = a_0 = A/2, \omega_0 = \frac{2\pi}{2} = \pi$$

$$c_k = \frac{a_k - j b_k}{2} = \frac{4A}{\pi^2 k^2}, k = \text{odd}$$

$$c_{-k} = \frac{4A}{\pi^2 k^2} = 0, k = \text{even}$$

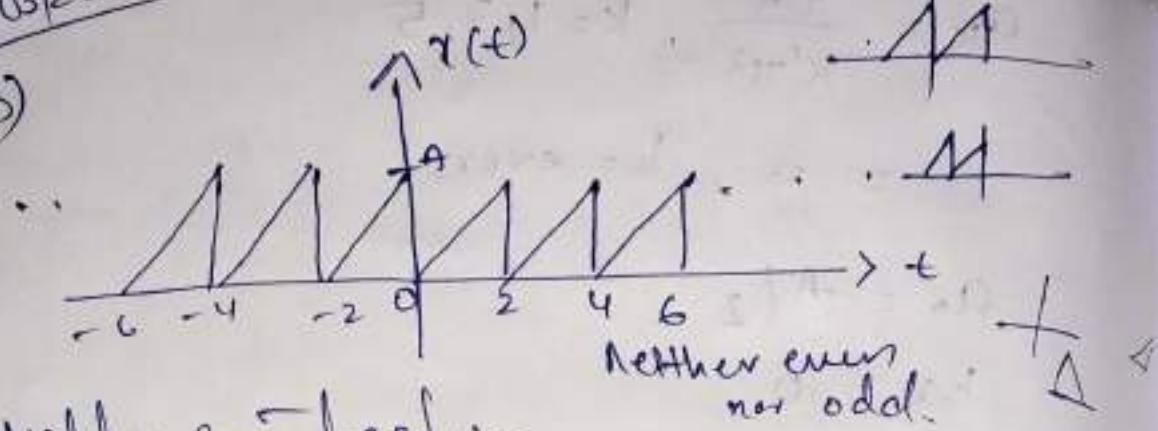
$$v(t) = \sum_{k=-\infty}^{\infty} \left(\frac{4A}{\pi^2 k^2} \right) e^{j\omega_0 kt}$$

$$v(t) = \sum_{k=-\infty}^{\infty} \left(\frac{4A}{\pi^2 k^2} \right) e^{j\omega_0 kt}$$



8/03/2020

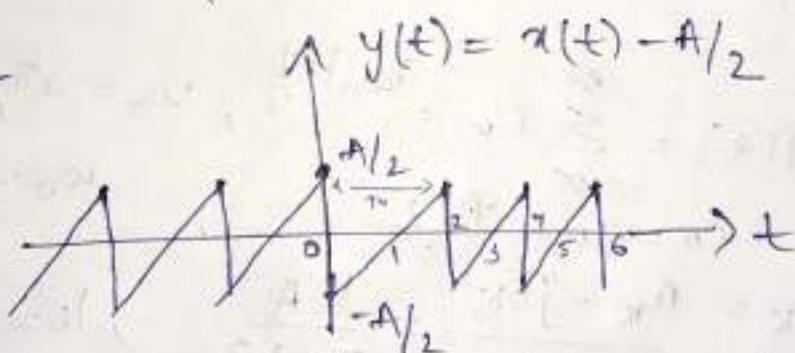
(PB)

Hidden Symmetry:-

In some signals some kind of symmetry exists but is hidden due to a d.c offset. This symmetry can be revealed by adding or subtracting a d.c. value equal to the d.c. offset.

The above problem is such an example.

Step1:-



\therefore This is odd symmetry

$$a_0 = a_k = 0$$

but $b_k = \frac{4}{T_0} \int_0^{T_0/2} x(t) \sin k\omega_0 t dt$

$$\omega_0 = 2, \omega_0 = \frac{2\pi}{2} = \pi$$

$$y(t) = \frac{A}{2}(-t-1), 0 \leq t \leq 1$$

$$b_{1K} = \frac{A}{2} \int_0^1 \frac{A}{2} (t-1) \sin k\pi t dt$$

$$= 2 \times \frac{A}{2} \int_0^1 (t-1) \sin k\pi t dt$$

$$b_K = A \int_0^1 t \sin k\pi t dt - A \int_0^1 \sin k\pi t dt$$

$$\boxed{\begin{aligned} \int t \sin at dt &= \frac{\sin at}{a^2} - \frac{t \cos at}{a} \\ \int t \cos at dt &= \frac{\cos at}{a^2} + \frac{t \sin at}{a} \end{aligned}}$$

$$b_K = A \left[\frac{\sin k\pi t}{k^2\pi^2} - \frac{t \cos k\pi t}{k\pi} \right]_0^1 + A \left[\frac{\cos k\pi t}{k\pi} \right]_0^1$$

$$b_K = A \left[\frac{\sin k\pi}{k^2\pi^2} - \frac{\cos k\pi}{k\pi} - [0] \right] + A \left[\frac{\cos k\pi}{k\pi} - \frac{1}{k\pi} \right]$$

$$b_{1K} = A \left[\frac{\sin k\pi}{k^2\pi^2} - \frac{\cos k\pi}{k\pi} \right] + \frac{A \cos k\pi}{k\pi} - \frac{A}{k\pi}$$

$$= \frac{A \sin k\pi}{k^2\pi^2} - \frac{A}{k\pi} \quad \because \sin k\pi = 0$$

$$= 0 - \frac{A}{k\pi} \quad \forall K$$

$$\boxed{b_{1K} = -\frac{A}{k\pi} \quad \forall K}$$

f. s

$$y(t) = \sum_{k=1}^{\infty} b_k \sin k\omega_0 t, \quad \omega_0 = \pi$$

$$y(t) = \sum_{k=1}^{\infty} \left(-\frac{A}{k\pi}\right) \sin kt$$

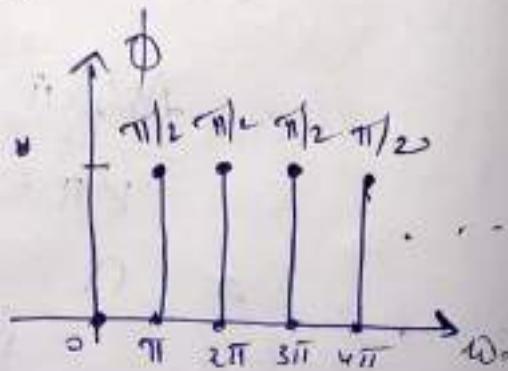
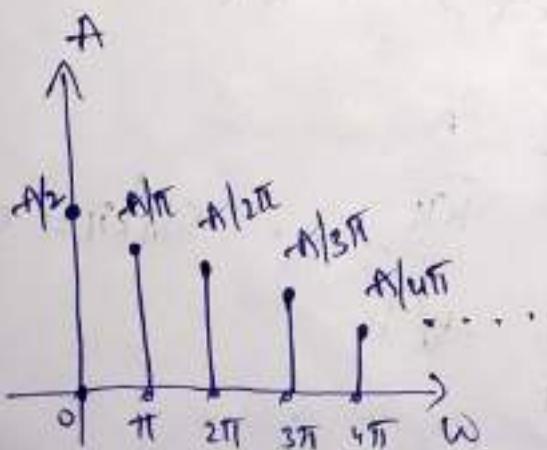
$$x(t) - \frac{A}{2} = \sum_{k=1}^{\infty} \left(-\frac{A}{k\pi}\right) \sin kt + \frac{A}{2}$$

$$\boxed{x(t) = \frac{A}{2} + \sum_{k=1}^{\infty} \left(-\frac{A}{k\pi}\right) \sin kt}$$

$$x(t) = \frac{A}{2} + \left(-\frac{A}{\pi} \sin t - \frac{A}{2\pi} \sin 2\pi t - \frac{A}{3\pi} \sin 3\pi t - \frac{A}{4\pi} \sin 4\pi t + \dots \right)$$

$$\boxed{-\sin kt = \cos(kt + \pi/2)}$$

$$x(t) = \frac{A}{2} + \frac{A}{\pi} \cos(\pi t + \pi/2) + \frac{A}{2\pi} \cos(2\pi t + \pi/2) + \frac{A}{3\pi} \cos(3\pi t + \pi/2) + \dots$$



Exponential Function

$$n(t) = \sum_{k=-\infty}^{\infty} C_k e^{j\omega_0 k t}, \quad \omega_0 = \pi$$

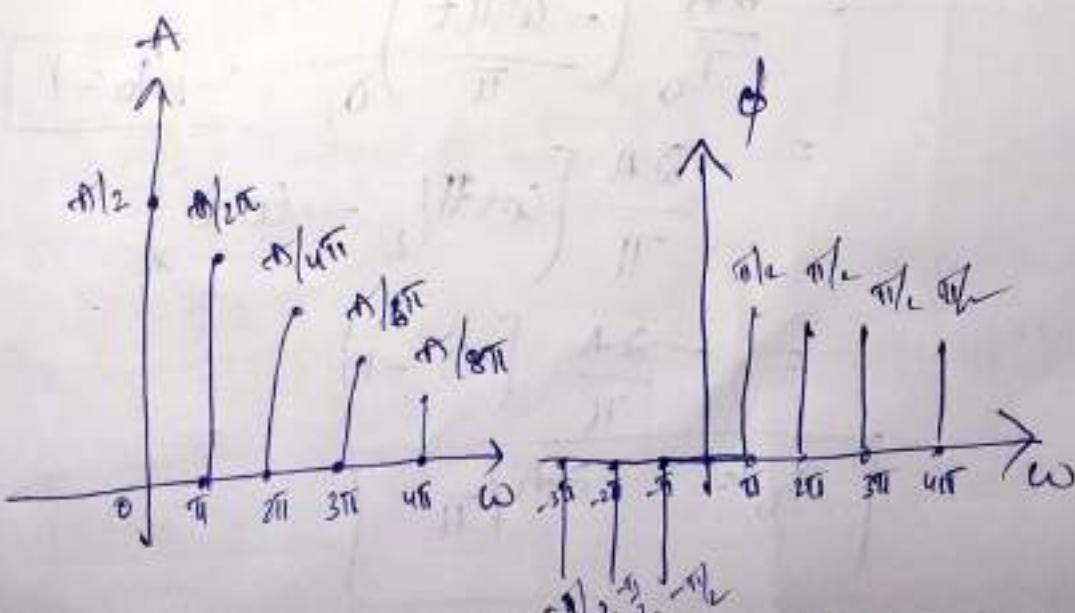
$$C_0 = a_0 = 0$$

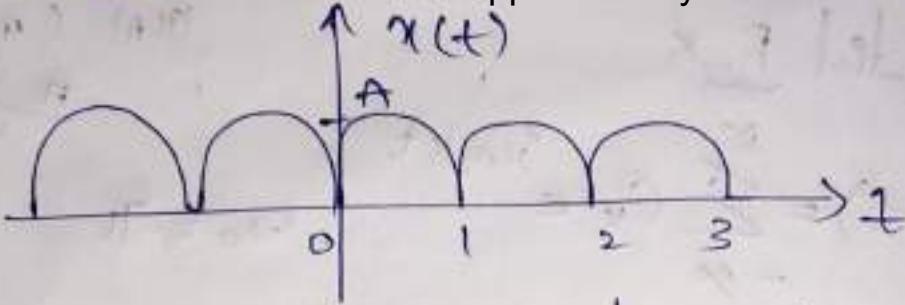
$$C_k = \frac{a_k - j b_k}{2} = -j \left(-\frac{A}{k\pi} \right) = \frac{Aj}{k\pi} \quad \forall k$$

$$C_k = C_k^* = -\frac{Aj}{k\pi} \quad \forall k.$$

$$n(t) = \sum_{\substack{k=-\infty \\ k \neq 0}}^{\infty} C_k e^{jk\omega_0 t} \quad [\omega_0 = \pi]$$

$$n(t) = \sum_{k=-\infty}^{\infty} \left(\frac{Aj}{k\pi} \right) e^{jk\pi t}$$





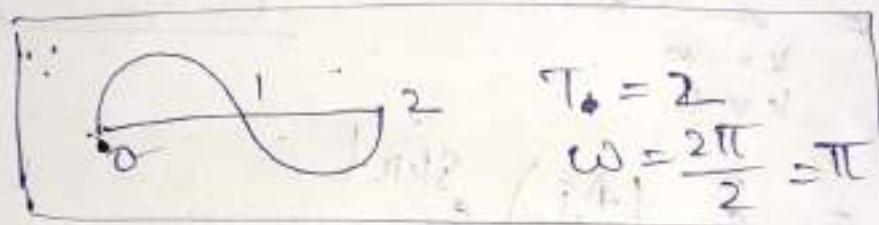
s1) $x(t)$ is even symmetric $b_k = 0$

s2) $a_0 = a_k \neq ?$

$$T_0 = 1$$

$$\omega_0 = \frac{2\pi}{T_0} = 2\pi$$

$$x(t) = A \sin \pi t, \quad 0 \leq t \leq 1/2$$



$$a_0 = \frac{2}{T_0} \int_0^{1/2} x(t) dt$$

$$= \frac{2A}{T_0} \int_0^{1/2} \sin \pi t dt$$

$$= \frac{2A}{T_0} \left[-\frac{\cos \pi t}{\pi} \right]_0^{1/2}$$

$$T_0 = 1$$

$$= -\frac{2A}{\pi} \left[\cos \frac{\pi}{2} - \cos 0 \right]$$

$$= -\frac{2A}{\pi} (0 - 1)$$

$$a_0 = -2A/\pi$$

$$a_K = \frac{4}{\pi} \int_0^{\pi/2} x(t) \cos k \omega t dt, \quad \omega_0 = 2\pi$$

$$a_K = \frac{4A}{\pi} \int_0^{\pi/2} \sin \pi t \cos 2\pi k t dt$$

$$a_K = \frac{4A}{\pi} \int_0^{\pi/2} \left(\sin((1+2k)\pi t) + \sin((1-2k)\pi t) \right) dt$$

$$a_K = 2A \int_0^{\pi/2} \sin((1+2k)\pi t) + \sin((1-2k)\pi t) dt$$

$$a_K = 2A \left[\left[\frac{-\cos((1+2k)\pi t)}{\pi(1+2k)} \right]_0^{\pi/2} + \left[\frac{-\cos((1-2k)\pi t)}{\pi(1-2k)} \right]_0^{\pi/2} \right]$$

$$a_K = 2A \left[\frac{-\cos((1+2k)\pi)}{\pi(1+2k)} + \frac{\cos 0}{\pi(1+2k)} + \left[\frac{-\cos((1-2k)\pi)}{\pi(1-2k)} + \frac{\cos 0}{\pi(1-2k)} \right] \right]$$

$$a_K = \frac{2A}{\pi} \left[\frac{-\cos((1+2k)\pi)}{1+2k} + \frac{1}{1+2k} - \frac{\cos((1-2k)\pi)}{1-2k} + \frac{1}{1-2k} \right]$$

$$\therefore \left[\cos(2k+1) \frac{\pi}{2} = \cos(2k-1) \frac{\pi}{2} = 0 \right]$$

$$a_K = \frac{2A}{\pi} \left[\frac{1}{1+2k} + \frac{1}{1-2k} \right]$$

$$a_K = \frac{2A}{\pi} \left[\frac{1-2k+1+2k}{1-4k^2} \right] = \frac{4A}{\pi(1-4k^2)}$$

$$\boxed{a_K = \frac{4A}{\pi(1-4k^2)}} \quad \forall k$$

$$x(t) = a_0 + \sum_{k=1}^{\infty} a_k \cos k\omega_0 t$$

$$x(t) = \frac{2A}{\pi} + \sum_{k=1}^{\infty} \frac{4A}{\pi(1-4k^2)} \cos 2\pi kt$$

$$x(t) = \frac{2A}{\pi} - \frac{4A}{3\pi} \cos 2\pi t - \frac{4A}{15\pi} \cos 4\pi t - \frac{4A}{35\pi} \cos 6\pi t$$

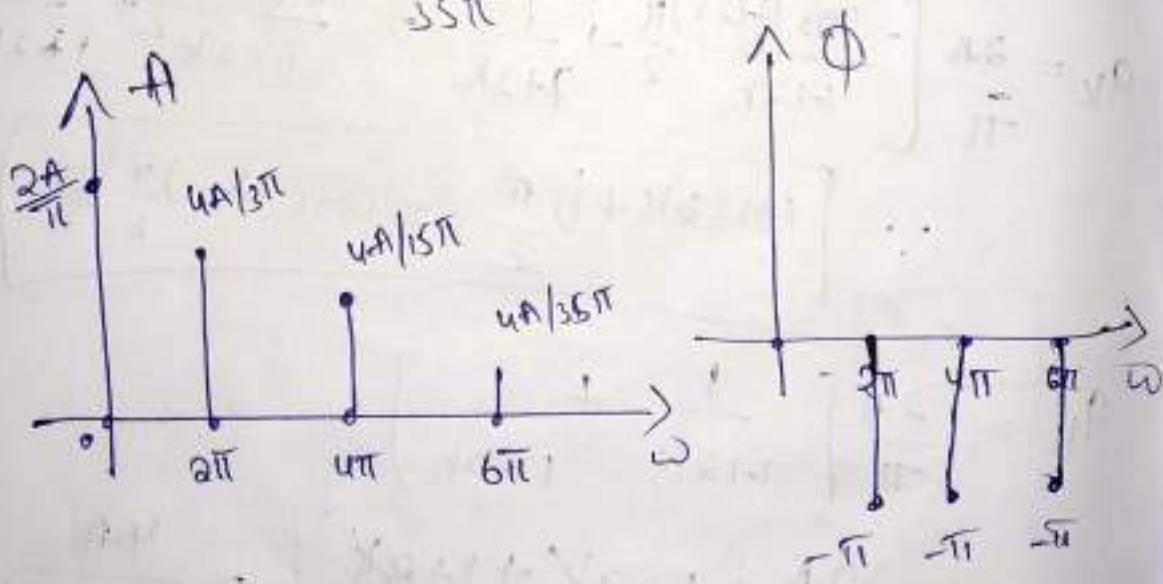
$$x(t) = \frac{2A}{\pi} + \frac{4A}{3\pi} \cos(2\pi t - \pi) + \frac{4A}{15\pi} \cos(4\pi t - \pi)$$

$$+ \frac{4A}{35\pi} \cos(6\pi t - \pi) + \dots$$

$$-\cos kt = \cos(kt - \pi)$$

$$x(t) = \frac{2A}{\pi} + \frac{4A}{3\pi} \cos(2\pi t - \pi) + \frac{4A}{15\pi} \cos(4\pi t - \pi)$$

$$+ \frac{4A}{35\pi} \cos(6\pi t - \pi) + \dots$$



E.P.8

$$x(t) = \sum_{-\infty}^{\infty} c_k e^{j\omega_0 k t}, \quad \omega_0 = 2\pi$$

$$c_0 = a_0 = 2A/\pi$$

$$c_k = \frac{a_k - j b_k}{2} = \frac{2A}{\pi(1-4k^2)}$$

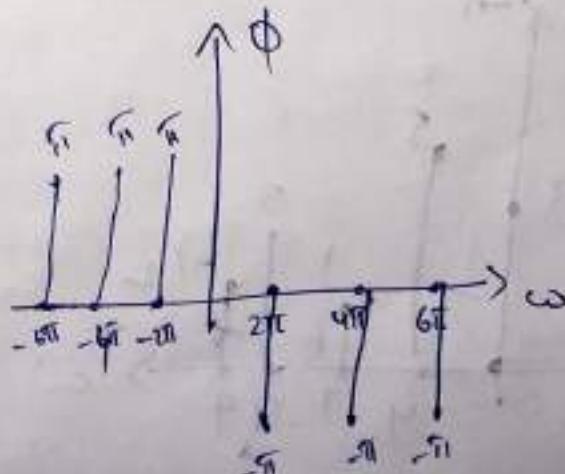
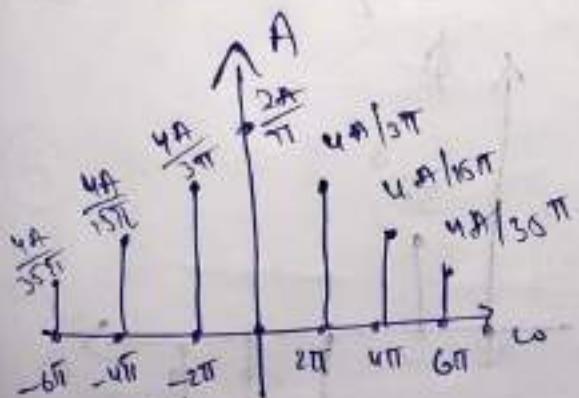
$$c_{-k} = c_k^* = \frac{2A}{\pi(1-4k^2)}$$

$$x(t) = a_0 + \sum_{k=-\infty}^{\infty} c_k e^{j\omega_0 k t}$$

$$x(t) = \frac{2A}{\pi} + \sum_{k=-\infty}^{\infty} \frac{2A}{\pi(1-4k^2)} e^{j2\pi k t}$$

$$x(t) = \frac{2A}{\pi} + \sum_{k=-\infty}^{\infty} \frac{2A}{\pi(1-4k^2)} e^{j4\pi k t}$$

$$\Rightarrow x(t) = \frac{2A}{\pi} + \sum_{k=-\infty}^{\infty} \frac{2A}{\pi(1-4k^2)} e^{j8\pi k t}$$



3) Draw the Frequency Spectrum of

$$\begin{aligned} \eta(t) = & 3 + 4 \sin 4t + 3 \cos 5t - \cos(7t - 75^\circ) \\ & + 3 \sin(8t - 100^\circ) + \frac{3}{2} \sin(9t - 150^\circ) \end{aligned}$$

SQ $\eta(t) = 3 + 4 \sin 4t + 3 \cos 4t - \cos(7t - 75^\circ)$

$$+ 3 \sin(8t - 100^\circ) - \frac{3}{2} \sin(9t - 150^\circ)$$

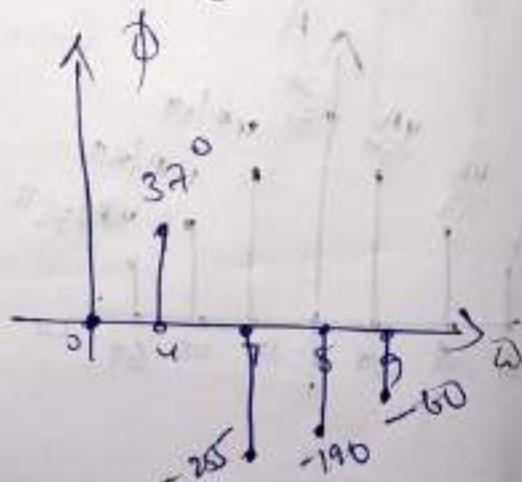
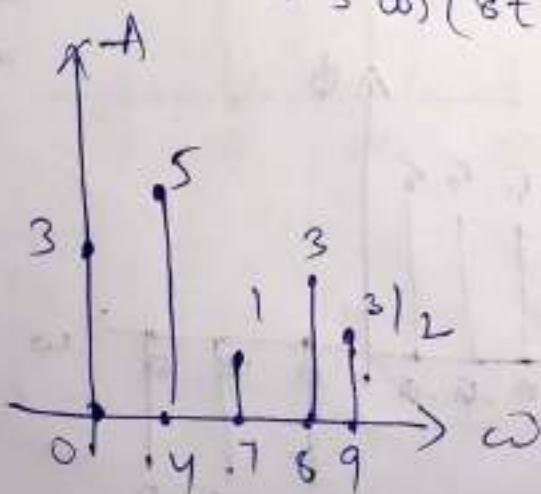
✓ $B \sin \theta + A \cos \theta = \sqrt{A^2 + B^2} \cos(\theta - \tan^{-1}\left(\frac{B}{A}\right))$ ✓

$$\begin{aligned} \eta(t) = & 3 + 5 \cos\left(4t + \tan^{-1}\left(\frac{3}{4}\right)\right) + \cos(7t - 15^\circ) \\ & + 3 \cos(8t - 100^\circ - \pi/2) + \frac{3}{2} \cos(9t - 150^\circ + \frac{\pi}{2}) \end{aligned}$$

$\begin{cases} -\cos \theta = \cos(\theta - \pi) \\ \sin \theta = \cos(\theta - \pi/2) \\ -\sin \theta = \cos(\theta + \pi/2) \end{cases}$

$$\eta(t) = 3 + 5 \cos(4t + 37^\circ) + \cos(7t - 25^\circ)$$

$$+ 3 \cos(8t - 190^\circ) + \frac{3}{2} \cos(9t - 60^\circ)$$

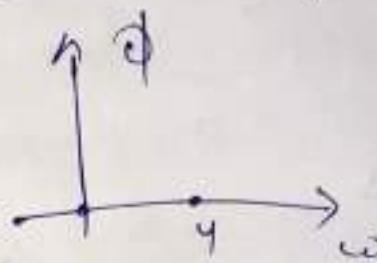
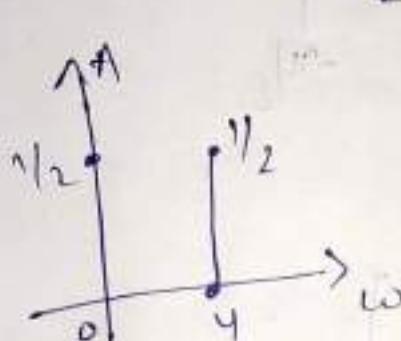


$$(4) x(t) = \cos^2 \omega t$$

- sol = (a) Draw the spectrum
 (b) find the E.F.s & draw the spectrum

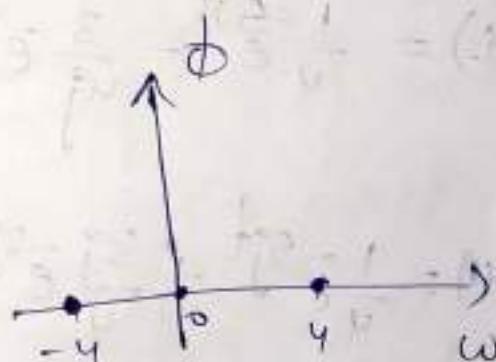
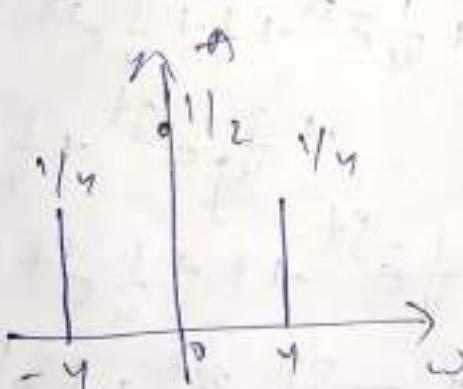
$$\text{sol} \quad x(t) \Rightarrow \frac{1 + \cos \omega t}{2}$$

$$x(t) \Rightarrow \frac{1}{2} + \frac{1}{2} \cos \omega t$$



$$(b) x(t) = \frac{1}{2} + \frac{1}{2} \left(e^{j\omega t} + e^{-j\omega t} \right)$$

$$= \frac{1}{4} e^{j\omega t} + \frac{1}{2} + \frac{1}{4} e^{-j\omega t}$$

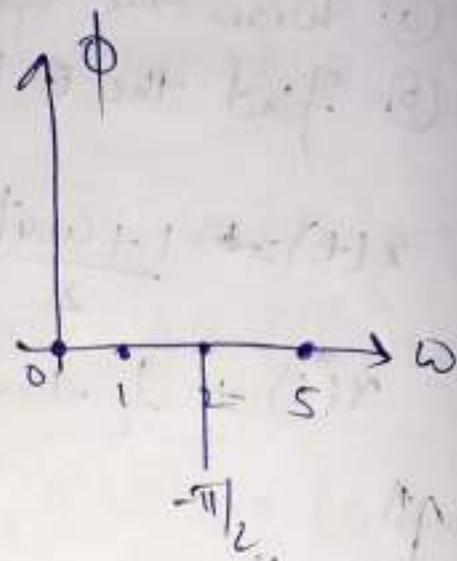
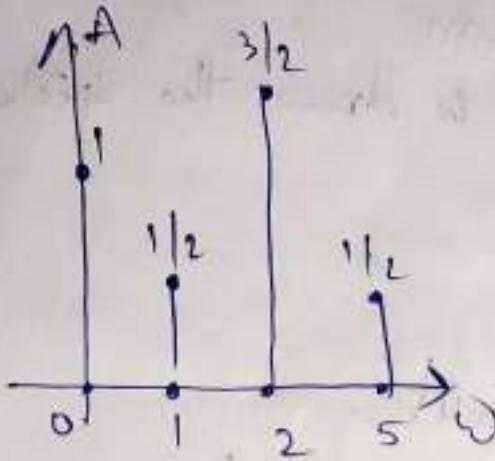


$$(5) x(t) = 1 + \sin \omega t \cos 3t \Rightarrow \frac{3}{2} \sin \omega t$$

$$\text{sol} \quad x(t) = 1 + \frac{1}{2} [\cos(\omega t) + \cos 3t] + \frac{3}{2} \cos \left(\omega t - \frac{\pi}{2} \right)$$

$$= 1 + \frac{1}{2} \cos \omega t + \frac{3}{2} \cos \left(\omega t - \frac{\pi}{2} \right) + \frac{1}{2} \cos 3t$$

$$\gamma(t) = -1 + \frac{1}{2} \cos t + \frac{3}{2} \cos(2t - 90^\circ) + \frac{1}{2} \cos 5t$$

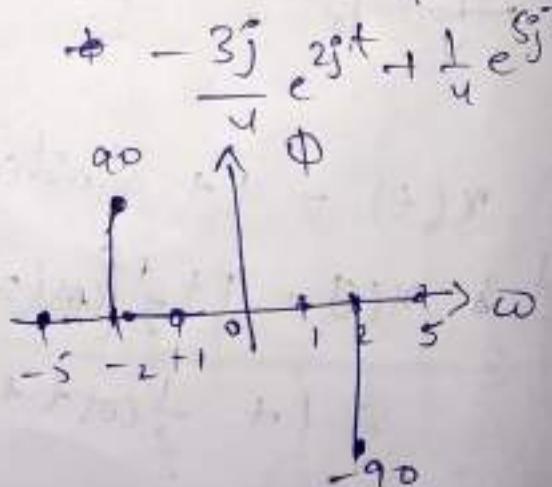
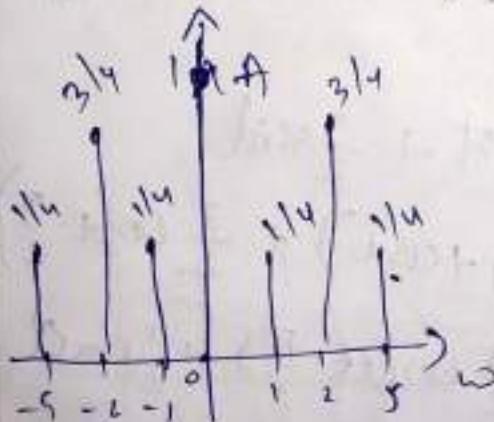


E.F.S

$$\alpha(t) = 1 + \frac{1}{2} \left[e^{\frac{jt}{2}} + e^{-\frac{jt}{2}} \right] + \frac{3}{2} \left[\frac{e^{2jt} - e^{-2jt}}{2j} \right] + \frac{1}{2} \left[e^{\frac{5jt}{2}} + e^{-\frac{5jt}{2}} \right]$$

$$\alpha(t) = \frac{1}{4} e^{-5jt} - \frac{3}{4} e^{2jt} + \frac{1}{4} e^{jt} + 1 + \frac{1}{4} e^{jt} + \frac{3}{4} e^{2jt} + \frac{1}{4} e^{5jt}$$

$$\alpha(t) = \frac{1}{4} e^{-5jt} + \frac{3}{4} e^{2jt} + \frac{1}{4} e^{jt} + 1 + \frac{1}{4} e^{jt} \rightarrow -\frac{3}{4} e^{2jt} + \frac{1}{4} e^{5jt}$$

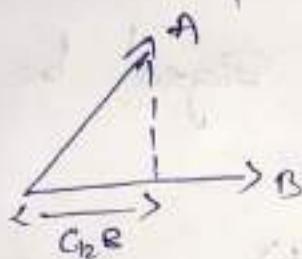


* Analogy b/w Signals & Vectors :-

> Vectors

> Signals are similar to vectors as they also have are dictated by some magnitude and some directions

Component of a vector



$$\underline{A} \approx C_{12} B$$

$$C_{12} = \frac{\underline{A} \cdot B}{|B|^2} = \cos \theta$$

Let \underline{A} & B be the vectors as shown in the figure above then the component of A on B can be mathematically expressed as follows

$$\underline{A} \approx C_{12} B$$

the best optimum value of C_{12}
so that A closely approximates B

$$C_{12} = \frac{\underline{A} \cdot B}{|B|^2} \quad (\text{perpendicular projection})$$

Component of a Signal:-

Let $x(t)$ & $y(t)$ be two different real signals.

Extending the theory developed for vectors to signals we can write

$$x(t) \approx c_{12}y(t) \quad \{t_1 \leq t \leq t_2\}$$

Let the error b/w two signal be denoted like as shown

$$e(t) = x(t) - c_{12}y(t)$$

To minimize the error between two signals either the energy or Power (Mean square value) of the ~~error~~ signal should be minimized

$$\begin{aligned} P_e(t) &= \frac{1}{t_2-t_1} \int_{t_1}^{t_2} e^2(t) dt \\ &= \frac{1}{t_2-t_1} \int_{t_1}^{t_2} (x(t) - c_{12}y(t))^2 dt \end{aligned}$$

$$MSE = P_{ct}(t) = \frac{1}{t_2 - t_1} \left[\int_{t_1}^{t_2} x^2(t) dt - 2C_{12} \int_{t_1}^{t_2} x(t)y(t) dt + \int_{t_1}^{t_2} y^2(t) dt \right]$$

In order to minimize the mean square error we need to choose a suitable value of C_{12} such that the error is minimum.

That is $\frac{d(MSE)}{dC_{12}} = 0$

$$\frac{1}{t_2 - t_1} \left[0 - 2 \int_{t_1}^{t_2} x(t)y(t) dt + 2C_{12} \int_{t_1}^{t_2} y^2(t) dt \right] = 0$$

$$C_{12} \int_{t_1}^{t_2} y^2(t) dt = \int_{t_1}^{t_2} x(t)y(t) dt$$

$$C_{12} = \frac{\int_{t_1}^{t_2} x(t)y(t) dt}{\int_{t_1}^{t_2} y^2(t) dt}$$

The above expression looks very much similar to the expression derived for vectors

thus bringing out a perfect analogy between signals & vectors

Special Case :-

If $c_{12} = 0$, then $x(t)$ is independent of $y(t)$ and they are entirely dissimilar to each other such

Signals are said to be orthogonal to each other

$$c_{12} = 0$$

i.e. $\int_{t_1}^{t_2} x(t) y(t) dt = 0$

$\therefore x(t) \perp y(t)$ are orthogonal

If the signals are complex then

$$c_{12} = \frac{\int_{t_1}^{t_2} x(t) y^*(t) dt}{\int_{t_1}^{t_2} y^2(t) dt}$$

and the condition of orthogonality

$$\boxed{\int_{t_1}^{t_2} x(t) y^*(t) dt = 0}$$

1) Prove that the functions $\sin n\omega t$ & $\cos m\omega t$ are orthogonal for every $m \neq n$ within the range $t_0, t_0 + \frac{2\pi}{\omega_0}$

$$\begin{aligned}
 & \int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} \sin n\omega t \cos m\omega t dt \\
 & \Rightarrow \frac{1}{2} \int_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} [\sin(n+m)\omega t + \sin(n-m)\omega t] dt \\
 & \Rightarrow \frac{1}{2} \left[-\frac{\cos(n+m)\omega t}{(n+m)\omega_0} - \frac{\cos(n-m)\omega t}{(n-m)\omega_0} \right]_{t_0}^{t_0 + \frac{2\pi}{\omega_0}} \\
 & = -\frac{1}{2\omega_0} \left[\frac{\cos(n+m)2\pi}{(n+m)} + \frac{\cos(n-m)2\pi}{(n-m)} - \left[\frac{\cos 0}{m+n} + \frac{\cos 0}{n-m} \right] \right] \\
 & = -\frac{1}{2} [0] = 0 \quad \text{Cos } k\pi = 1
 \end{aligned}$$

\therefore orthogonal

2) Prove that the complex exponential functions are orthogonal functions

~~shift, subtract, (0, $\frac{2\pi}{\omega_0}$)~~

Let the Complex Exponential be
 $x(t) = e^{j\omega_0 t}$ & $y(t) = e^{j\omega_0 t}, (0, \frac{2\pi}{\omega_0})$

$$\int_{t_1}^{t_2} x(t) y^*(t) dt = 0$$

$$\int_0^{2\pi/\omega_0} e^{jn\omega_0 t} \bar{e}^{jm\omega_0 t} dt$$

$$\Rightarrow \int_0^{2\pi/\omega_0} e^{j(n-m)\omega_0 t} dt$$

$$\Rightarrow \left[\frac{e^{j(n-m)\omega_0 t}}{j(n-m)\omega_0} \right]_0^{2\pi/\omega_0}$$

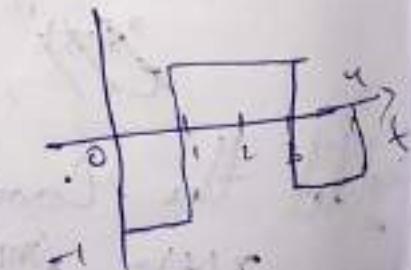
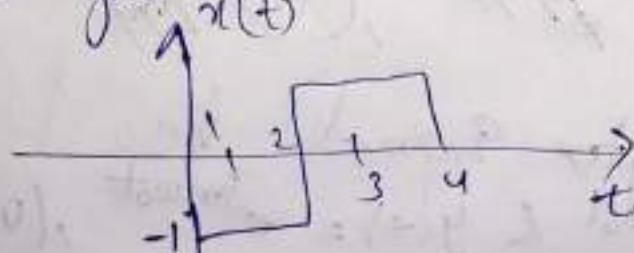
$$= \frac{e^{j(n-m)2\pi} - 1}{j(n-m)\omega_0}$$

$$= \frac{1 - 1}{j(n-m)\omega_0}$$

$$= 0$$

\therefore orthogonal.

- ③ prove that the above signals are orthogonal.



Sol

$$\begin{aligned}
 & \int_{t_1}^{t_2} m(t) y(t) dt \\
 &= \int_0^1 +1 dt + \int_1^2 (-1) dt + \int_2^3 +1 dt + \int_3^7 -1 dt \\
 &= 0 \\
 \therefore & \text{Orthogonal}
 \end{aligned}$$

Q. A rectangular function $x(t)$ is defined by the following function.

$$\begin{aligned}
 x(t) = 1, & 0 \leq t \leq \pi \\
 & \leq -1, \quad \pi \leq t \leq 2\pi
 \end{aligned}$$

approximate the above following by a single sinusoid $\sin t$ over the interval $(0, 2\pi)$

such that the mean square error is minimum. Evaluate the mean square error in this approximation

Sol Let $x(t)$ be approximated in terms of $\sin t$ as follows.

$$m(t) \approx c_{12} y(t)$$

$$x(t) = c_{12} \sin t$$

Q.

$$\begin{aligned}
 C_{12} &= \frac{\int_{t_1}^{t_2} n(t) y(t) dt}{\int_{t_1}^{t_2} y^2(t) dt} \\
 &= \frac{\int_0^{\pi} 1 \cdot \sin t dt + \int_{-\pi}^0 (-1) \sin t dt}{\int_0^{2\pi} \sin^2 t dt} \\
 &= \frac{-\cos t \Big|_0^{\pi} + \cos t \Big|_{-\pi}^{2\pi}}{\int_0^{2\pi} \frac{1 - \cos 2t}{2} dt} \\
 C_{12} &= \frac{-(\cos \pi - \cos 0) + (\cos 2\pi - \cos 0)}{\frac{1}{2}(2\pi) - \frac{1}{2}(\sin 2\pi - \sin 0)} \\
 &\boxed{C_{12} = \frac{4}{\pi}}
 \end{aligned}$$

$$n(t) \approx \frac{4}{\pi} \sin t$$

$$MSE = \frac{1}{t_2 - t_1} \left[\int_{t_1}^{t_2} n^2(t) dt - 2c_{12} \int_{t_1}^{t_2} n(t)y(t) dt + c_{12}^2 \int_{t_1}^{t_2} y^2(t) dt \right]$$

$$\begin{aligned}
 \text{MSE} &= \frac{1}{2\pi} \left[\int_0^{2\pi} n^2(t) dt - 2c_{12} \int_0^{2\pi} n(t)y(t) dt \right. \\
 &\quad \left. + c_{12}^2 \int_0^{2\pi} y^2(t) dt \right] \\
 &= \frac{1}{2\pi} \left[\int_0^{\pi} 1 \cdot dt + \int_{\pi}^{2\pi} (\sin t)^2 dt - 2 \times \frac{4}{\pi} \left[\int_0^{\pi} 1 \cdot \sin t dt + \int_{\pi}^{2\pi} \sin t dt \right] \right. \\
 &\quad \left. + \frac{16}{\pi^2} \left[\int_0^{2\pi} \sin^2 t dt \right] \right] \\
 &= \frac{1}{2\pi} \left[\pi + \pi - \frac{8}{\pi} \left[-\cos t \Big|_0^{2\pi} \right] + \cos t \Big|_{\pi}^{2\pi} \right. \\
 &\quad \left. + \frac{16}{\pi^2} \left[\int_0^{2\pi} \frac{1 - \cos 2t}{2} dt \right] \right] \\
 &= \frac{1}{2\pi} \left[2\pi - \frac{8}{\pi} [(1 - \cos \pi) + (\cos 2\pi - \cos \pi)] + \frac{16}{\pi^2} \left[\frac{t}{2} - \frac{1}{2} \sin t \Big|_0^{2\pi} \right] \right] \\
 &= \frac{1}{2\pi} \left[2\pi - \frac{8}{\pi} [2 + 2] + \frac{16}{\pi^2} \left[\frac{2\pi - 0}{2} - \frac{1}{4} (\sin 4\pi - \sin 0) \right] \right] \\
 &= \frac{1}{2\pi} \left[2\pi - \frac{32}{\pi} + \frac{16}{\pi^2} [2\pi] \right] \\
 &= \frac{1}{2\pi} \left[2\pi + \frac{16}{\pi} - \frac{32}{\pi} \right] = \\
 &= \frac{1}{2\pi} \left[2\pi - \frac{16}{\pi} \right] = 1 - \frac{8}{\pi^2} = 0.189
 \end{aligned}$$

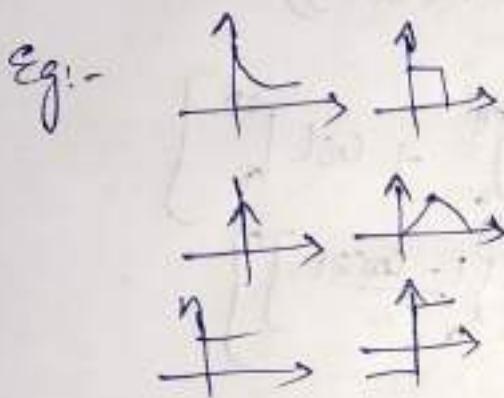
$$\begin{aligned}
 \text{MSE} &= 0.189 \\
 &< 18.9\%
 \end{aligned}$$

11/03/2020

UNIT:- 03:-

Fouier Transform :-

Aperiodic signal $\xrightarrow{\text{F.T.}}$ Frequency domain
 $x(t)$ $\xrightarrow{\text{F.T.}}$ $X(\omega)$.



FT is a tool or concept developed by Fourier & is available to obtain frequency domain representation of aperiodic signal.

Fouier transform:-

$$x(t) \xrightarrow{\text{F.T.}} X(\omega)$$

$$\mathcal{F}[x(t)] = X(\omega)$$

$$\Rightarrow X(\omega) = \mathcal{F}[x(t)]$$

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

$$x(\omega) = \int_{-\infty}^{\infty} x(t) e^{j\omega t} dt$$

fourier
operator

Inverse Fourier transform (IFT): -

$$X(\omega) \xrightarrow{F^{-1}} x(t)$$

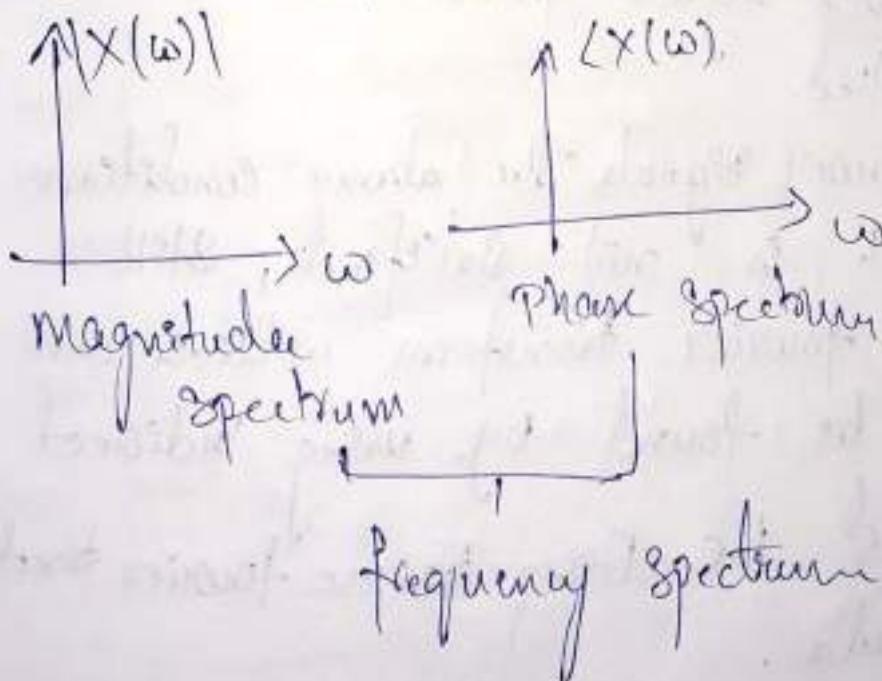
$$x(t) = F^{-1}[X(\omega)]$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

Note:- $\xrightarrow{\text{in}} X(j\omega)$

① $X(\omega)$ is in General Complex

$$\therefore X(\omega) = |X(\omega)| \angle X(\omega)$$



② Dirichlet's Conditions:-

For power a signal $x(t)$ to have Fourier transform it should be absolutely integrable.

All Energy signals satisfy this condition.

For FT to exists

$$|x(\omega)| < \infty$$

$$\therefore \int_{-\infty}^{\infty} |x(t)| |\tilde{e}^{j\omega t}| dt < \infty$$

$$\boxed{\int_{-\infty}^{\infty} |x(t)| dt < \infty}$$

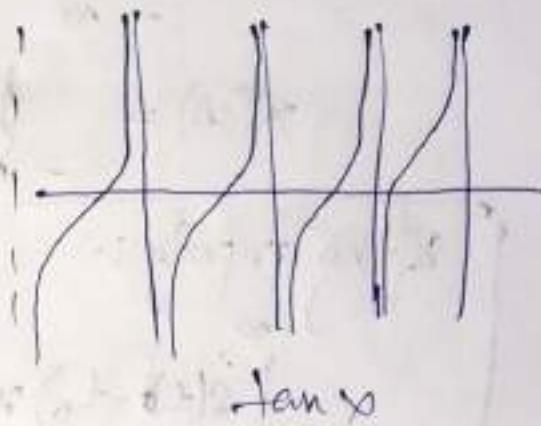
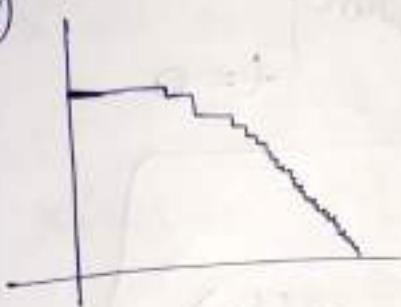
For such signals formula can be directly applied.

For power signals the above condition though is not satisfied, still have Fourier transform which can only be found by using Indirect Method i.e. Using inverse Fourier transform formula.

Signals which are neither energy nor power will not have Fourier transform

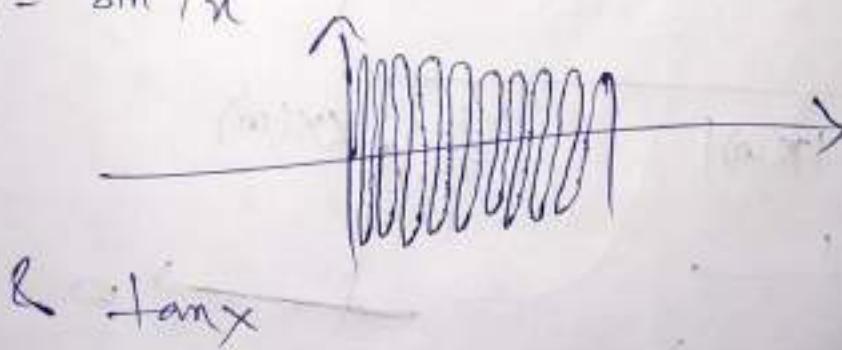
- ② For Fourier transform to exist, signals should have finite number of discontinuities within a finite duration and the number of discontinuity must be finite

Eg:-



- ③ For a Fourier transform to exist the signal should have finite number of maxima within a finite duration and the number of maxima must also be finite

Eg:- \sin^2/n



Problem:-

① $F[\delta(t)]$ Find fourier transfer of impulse signal

\therefore It is a finite duration signal

$$x(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

$$x(\omega) = \int_{-\infty}^{\infty} \delta(t) \cdot e^{-j\omega t} dt$$

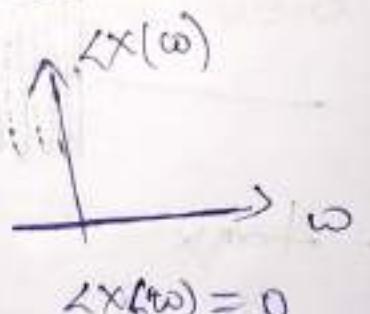
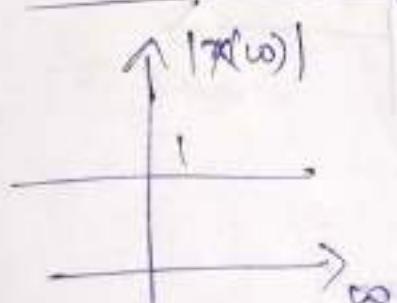
$$x(\omega) = \left[e^{-j\omega t} \right]_{t=0}$$

Sifting property :-

$$\int_{-\infty}^{\infty} \delta(t-t_0) x(t) = x(t_0)$$

$$|x(\omega)| = 1$$

$$|\delta(t)| \xrightarrow{F} 1$$



$$\angle X(\omega) = 0$$

④ Note:-

All Aperiodic signals have continuous spectrum, whereas all periodic signals have discrete spectrum.

⑤ $F[\bar{e}^{at}v(t)]$:

$$\stackrel{\text{def}}{=} x(t) = \bar{e}^{at}v(t)$$

$$x(\omega) = \int_{-\infty}^{\infty} \bar{e}^{at}v(t) dt \stackrel{?}{=} \int_{-\infty}^{\infty} \bar{e}^{at}v(t) e^{-j\omega t} dt$$

$$x(\omega) = \int_0^{\infty} \bar{e}^{at}v(t) e^{-j\omega t} dt$$

$$x(\omega) = \frac{\bar{e}^{(a+j\omega)t}}{-a-j\omega} \Big|_0^{\infty}$$

$$= -\frac{1}{a+j\omega} [0-1]$$

$$\boxed{x(\omega) = 1/a+j\omega}$$

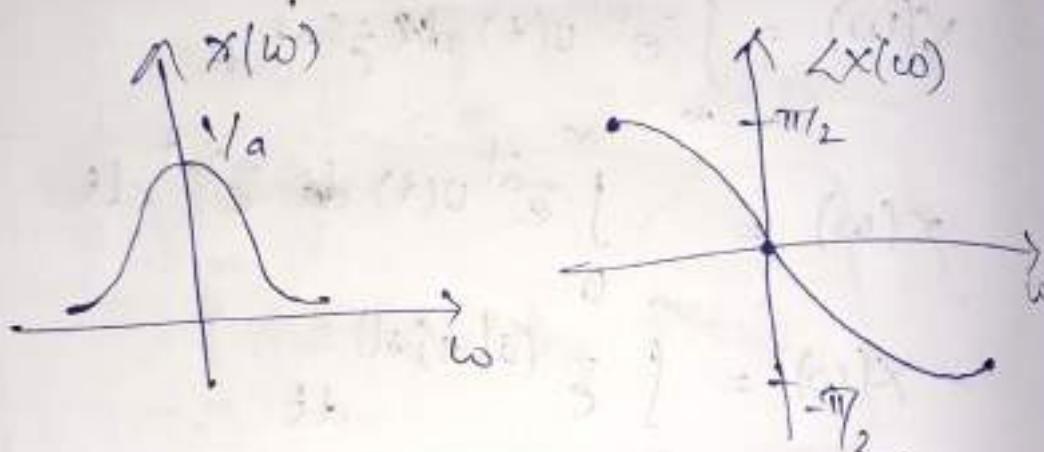
$$\boxed{\bar{e}^{at}v(t) \xrightarrow{F} \frac{1}{a+j\omega}}$$

$$X(\omega) = \frac{a - j\omega}{a^2 + \omega^2}$$

$$|X(\omega)| = \frac{1}{\sqrt{a^2 + \omega^2}}$$

$$\angle X(\omega) = \tan^{-1} \left(\frac{-\omega / \sqrt{a^2 + \omega^2}}{a / \sqrt{a^2 + \omega^2}} \right)$$

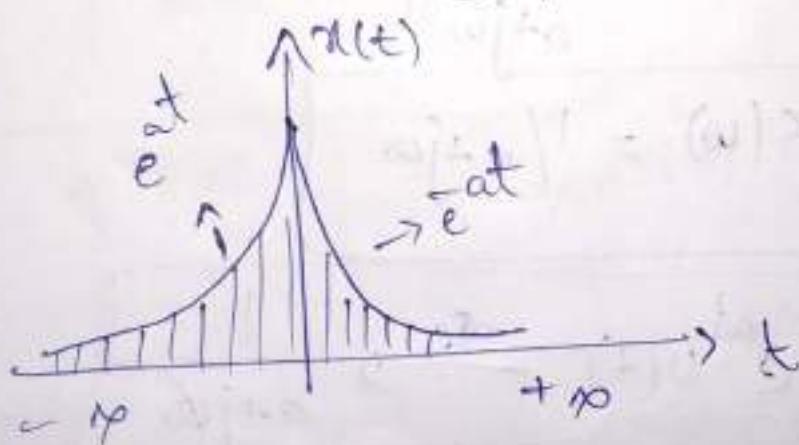
$$\angle X(\omega) = -\tan^{-1}(\omega/a)$$



③ $x(t) = e^{-at}$

Sol $x(t) = e^{at}, t < 0$

$= e^{at}, t > 0$



$$x(\omega) = \int_{-\infty}^{\infty} x(t) e^{j\omega t} dt = e^{\infty} = \frac{1}{e^{\infty}} = \frac{1}{\infty} = 0$$

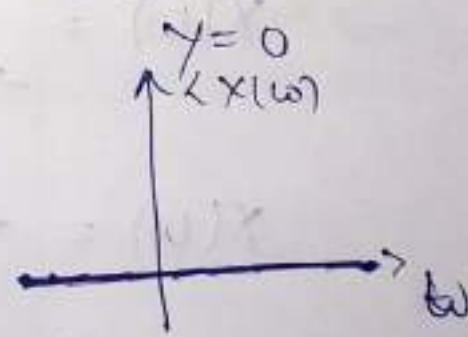
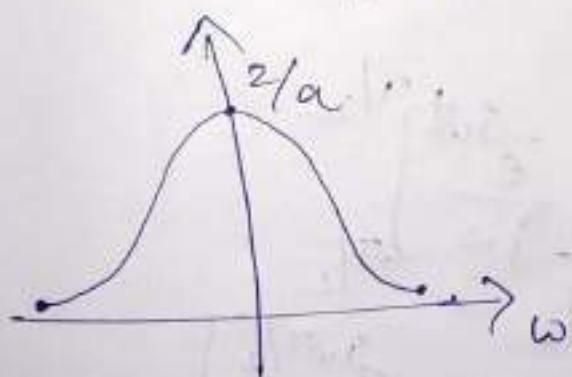
$$x(\omega) = \int_{-\infty}^{\infty} e^{-at} e^{-j\omega t} dt$$

$$\begin{aligned} x(\omega) &= \int_{-\infty}^0 e^{(a-j\omega)t} dt + \int_0^{\infty} e^{-t(a+j\omega)} dt \\ &= \left[\frac{e^{(a-j\omega)t}}{a-j\omega} \right]_{-\infty}^0 + \left[\frac{e^{-(a+j\omega)t}}{-a-j\omega} \right]_0^{\infty} \\ &= \frac{1-0}{a-j\omega} - \frac{(0-1)}{a+j\omega} \end{aligned}$$

$$x(\omega) = \frac{1}{a+j\omega} + \frac{1}{a-j\omega} = \frac{2a}{a^2+\omega^2}$$

$$\boxed{e^{-at} \xrightarrow{F} \frac{2a}{a^2+\omega^2}}$$

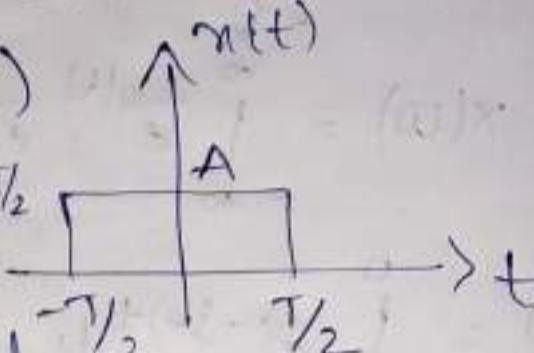
$$|X(\omega)| = \frac{2a}{a^2+\omega^2} \quad \angle X(\omega) = 0^\circ$$



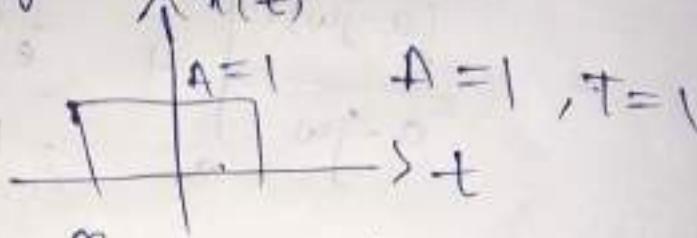
$$\textcircled{4} \quad x(t) = A \operatorname{rect}(t/T)$$

$$\Leftrightarrow A \operatorname{rect}(t/T)$$

$$\begin{aligned} A \operatorname{rect}(t/T) &= A, -T/2 \leq t \leq T/2 \\ &= 0, \text{ else} \end{aligned}$$



Unit rectangular function



$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

$$X(\omega) = \int_{-\infty}^{\infty} A \operatorname{rect}(t/T) e^{-j\omega t} dt$$

$$X(\omega) = \int_{-T/2}^{T/2} A \operatorname{rect}(t/T) e^{-j\omega t} dt$$

$$X(\omega) = \int_{-T/2}^{T/2} A e^{-j\omega t} dt$$

$$X(\omega) = A \left[\frac{e^{-j\omega T/2}}{-j\omega} \right]_{-T/2}^{T/2}$$

$$X(\omega) = \frac{A}{-j\omega} \left[e^{-j\omega T/2} - e^{j\omega T/2} \right]$$

$$x(\omega) = \frac{2A}{j\omega} \left[\frac{e^{j\frac{\omega T}{2}} - e^{-j\frac{\omega T}{2}}}{2j} \right]$$

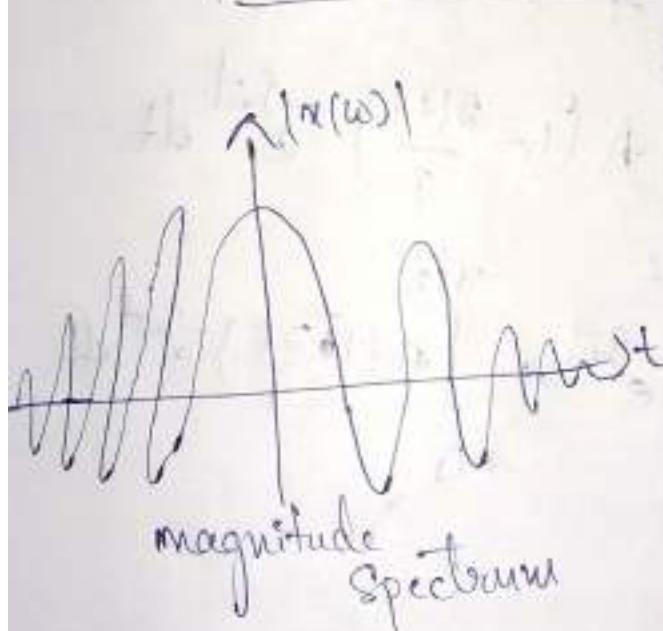
$$x(\omega) = \frac{2A}{\omega} \left[\sin\left(\frac{\omega T}{2}\right) \right]$$

$$x(\omega) = \frac{2A}{\omega} \times \frac{\omega T}{2} \frac{\sin(\omega T/2)}{\omega T/2}$$

$$\boxed{x(\omega) = AT \operatorname{Sa}\left(\frac{\omega T}{2}\right)}$$

$$\boxed{\operatorname{Sa}(x) = \frac{\sin x}{x}}$$

$$\therefore \boxed{-A \operatorname{rect}\left(\frac{t}{T}\right) \xrightarrow{f} AT \operatorname{Sa}\left(\frac{\omega T}{2}\right)}$$



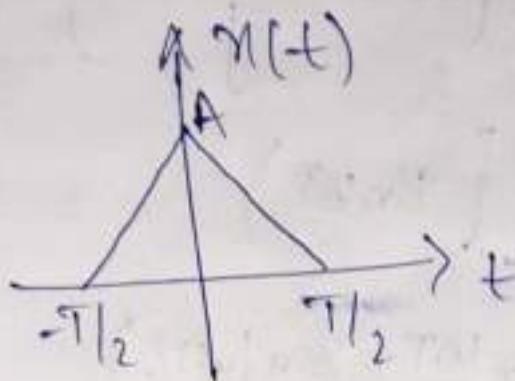
$$(05) \quad x(\omega) = \frac{2A}{\omega} \sin\left(\frac{\omega T}{2}\right)$$

$$\operatorname{sinc}(x) = \frac{\sin x}{\pi x}$$

$$= \frac{2A}{\omega} \frac{\sin\left(\frac{\pi \omega T}{2\pi}\right)}{\frac{\pi \omega T}{2\pi}} \times \frac{\pi \omega T}{2\pi}$$

$$\boxed{x(\omega) = AT \operatorname{sinc}\left(\frac{\omega T}{2\pi}\right)}$$

$$\textcircled{5} \quad n(t) = A \Delta(t/T)$$



$$n(t) = A \left(1 - \frac{2t}{T}\right), \quad 0 \leq t \leq T/2$$

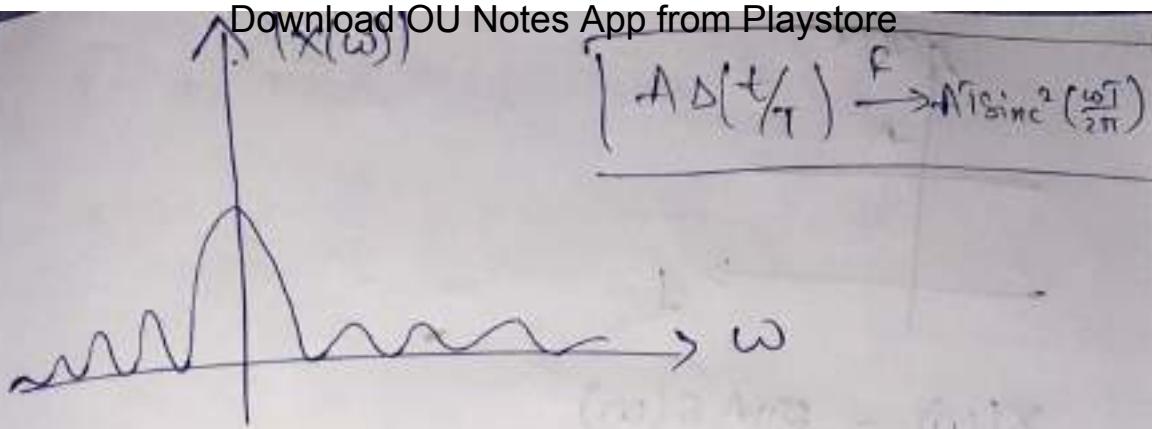
$$= A \left(1 + \frac{2t}{T}\right), \quad -T/2 \leq t \leq 0$$

$$n(t) = A \left(1 - \frac{2|t|}{T}\right), \quad |t| \leq T/2$$

$$X(\omega) = \int_{-T/2}^{T/2} A \left(1 - \frac{2|t|}{T}\right) e^{-j\omega t} dt$$

$$X(\omega) = \int_{-T/2}^0 A \left(1 + \frac{2t}{T}\right) e^{j\omega t} dt + \int_0^{T/2} A \left(1 - \frac{2t}{T}\right) e^{-j\omega t} dt$$

$$X(\omega) = AT \sin^2 \left(\frac{\omega T}{2} \right)$$



⑥ $F^{-1}[\delta(\omega)]$ find the inverse fourier transform

$$\text{sol } x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} x(\omega) e^{j\omega t} d\omega$$

$$= \frac{1}{2\pi} \int_{-\infty}^{\infty} \delta(\omega) e^{j\omega t} d\omega$$

∴ softening property

$$x(t) = \frac{1}{2\pi} (1) = \frac{1}{2\pi}$$

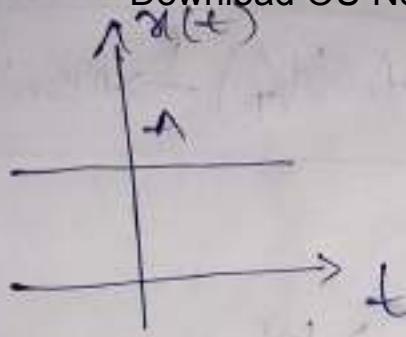
$$F^{-1}[\delta(\omega)] = \frac{1}{2\pi}$$

$$\Rightarrow F\left[\frac{1}{2\pi}\right] = \delta(\omega)$$

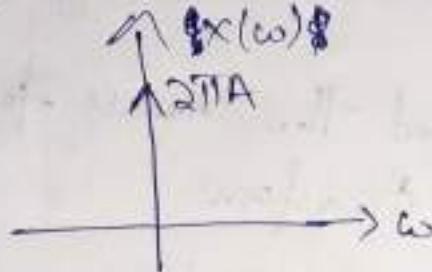
$$F[1] = 2\pi \delta(\omega)$$

$$1 \xrightarrow{F} 2\pi \delta(\omega)$$

$$A \xrightarrow{F} 2\pi A \delta(\omega)$$



$$x(\omega) = 2\pi A \delta(\omega)$$



(7) $F^{-1}[\delta(\omega - \omega_0)]$

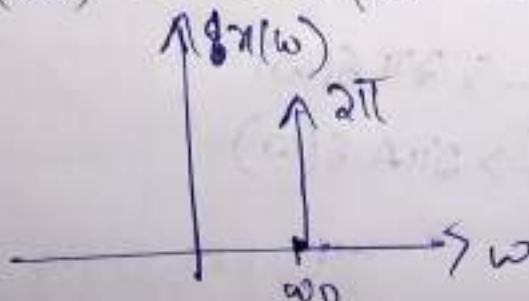
$$\stackrel{def}{=} n(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \delta(\omega - \omega_0) \cdot e^{j\omega t} d\omega$$

$$n(t) = \frac{1}{2\pi} [e^{j\omega_0 t}]$$

$$F\left[\frac{e^{j\omega_0 t}}{2\pi}\right] = \delta(\omega - \omega_0)$$

$$\boxed{F[e^{j\omega_0 t}] = 2\pi \delta(\omega - \omega_0)}$$

$$x(\omega) = 2\pi \delta(\omega - \omega_0)$$



$$\textcircled{8} \quad F^{-1} [\delta(\omega + \omega_0)]$$

$$\text{def} \quad x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \delta(\omega + \omega_0) e^{j\omega t} dt.$$

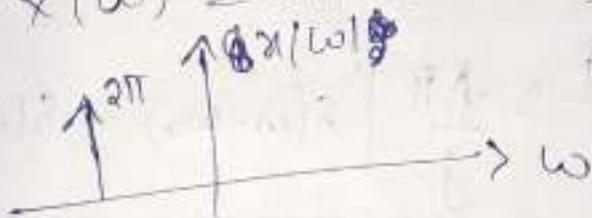
$$x(t) = \frac{1}{2\pi} e^{-j\omega_0 t}$$

$$\boxed{F\left[\frac{e^{-j\omega_0 t}}{2\pi}\right] = \delta(\omega + \omega_0)}$$

$$\text{def} \quad F[e^{-j\omega_0 t}] = 2\pi \delta(\omega + \omega_0)$$

$$\boxed{F[e^{-j\omega_0 t}] = 2\pi \delta(\omega + \omega_0)}$$

$$X(\omega) = 2\pi \delta(\omega + \omega_0)$$



$$\textcircled{9} \quad x(t) = A \cos \omega_0 t$$

$$\text{def} \quad F[A \cos \omega_0 t] = A F[\cos \omega_0 t]$$

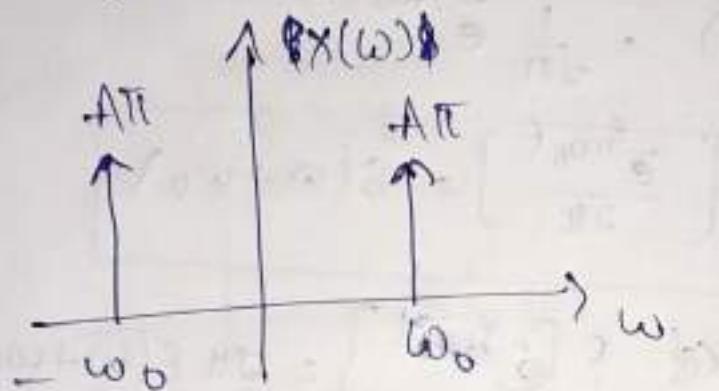
$$= A F\left[\frac{e^{j\omega_0 t}}{2} + \frac{-e^{j\omega_0 t}}{2}\right]$$

$$= \frac{A}{2} F\left[e^{j\omega_0 t} + e^{-j\omega_0 t}\right]$$

$$= \frac{A}{2} \left[f[e^{j\omega_0 t}] + f[e^{-j\omega_0 t}] \right]$$

$$= \frac{A}{2} [2\pi\delta(\omega - \omega_0) + 2\pi\delta(\omega + \omega_0)]$$

$$\mathcal{F}\{A \cos \omega t\} = A\pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$$



$$A \cos \omega t \xrightarrow{\mathcal{F}} A\pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$$

$$A \sin \omega t \xrightarrow{\mathcal{F}} \frac{A\pi}{j} [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$$

$$\textcircled{10} \quad x(t) = A \sin \omega t$$

$$\text{sol } \mathcal{F}\{A \sin \omega t\} = A \mathcal{F}\{\sin \omega t\}$$

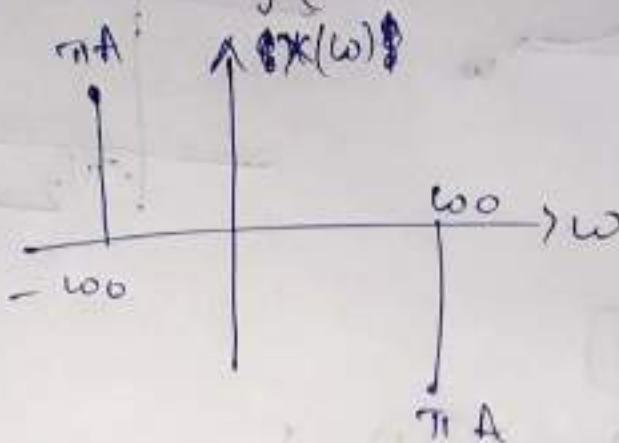
$$= A \left[\frac{e^{j\omega t}}{2j} - \frac{e^{-j\omega t}}{2j} \right]$$

$$= \frac{A}{2j} [f(e^{j\omega t}) - f(e^{-j\omega t})]$$

$$= \frac{A}{2j} [2\pi\delta(\omega - \omega_0) - 2\pi\delta(\omega + \omega_0)]$$

$$= \frac{-A\pi}{j} [\delta(\omega - \omega_0) - \delta(\omega + \omega_0)]$$

$$f[A \sin(\omega_0 t)] = -A\pi j [\delta(\omega + \omega_0) - \delta(\omega - \omega_0)]$$



⑪ $F[\text{sgn}(t)]$

Sol:

$$\text{sgn}(t) = \lim_{a \rightarrow 0} [e^{-at} v(t) - e^{at} v(-t)]$$

$$= \lim_{a \rightarrow 0} \left[\int_{-\infty}^0 e^{-at} v(t) e^{j\omega t} dt - \int_{-\infty}^0 e^{at} v(t) e^{-j\omega t} dt \right]$$

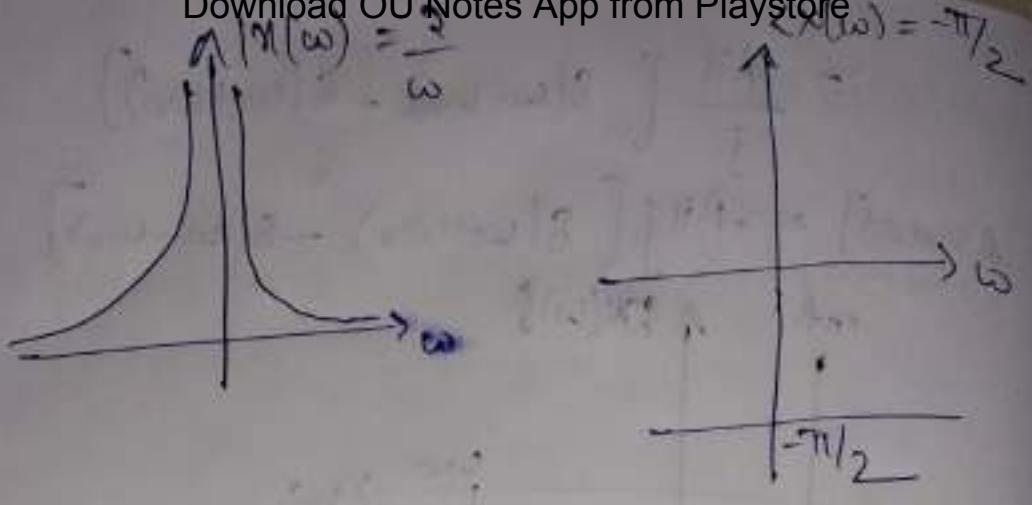
$$= \lim_{a \rightarrow 0} \left[\int_0^\infty e^{-t(a+j\omega)} dt - \int_{-\infty}^0 e^{t(a-j\omega)} dt \right]$$

$$= \lim_{a \rightarrow 0} \left[\frac{1}{a+j\omega} - \left(\frac{1}{a-j\omega} \right) \right]$$

$$= \lim_{a \rightarrow 0} \left[\frac{1}{a+j\omega} - \frac{1}{a-j\omega} \right]$$

$$X(\omega) = \frac{1}{j\omega} - \left(-\frac{1}{j\omega} \right)$$

$$X(\omega) = \frac{2}{j\omega} = \frac{2}{\omega} e^{-90^\circ}$$



$$\textcircled{12} \quad F[v(t)]$$

$$\stackrel{\text{def}}{=} \text{Sgn}(t) = 2v(t) - 1$$

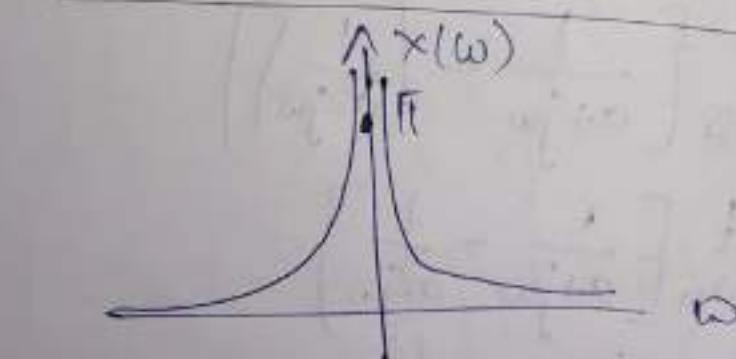
$$v(t) = \frac{1}{2} + \frac{1}{2} \text{Sgn}(t)$$

$$F[v(t)] = F\left[\frac{1}{2} + \frac{1}{2} \text{Sgn}(t)\right]$$

$$= F\left[\frac{1}{2}\right] + \frac{1}{2} F[\text{Sgn}(t)]$$

$$= 2\pi \times \frac{1}{2} S(\omega) + \frac{1}{2} \times \frac{2}{j\omega}$$

$$\boxed{X(\omega) = \pi S(\omega) + \frac{1}{j\omega}}$$



$$\boxed{v(t) \xrightarrow{F} \pi S(\omega) + \frac{1}{j\omega}}$$

Summary:-

$$\text{Stec}(t) \leftrightarrow \text{Sinc}\left(\frac{\omega}{2\pi}\right)$$

Unit rectangular

$x(t)$ Signal	$X(\omega)$ F.T	$\Delta(t) \xrightarrow{F.T} \text{sinc}^2\left(\frac{\omega}{2\pi}\right)$ Unit triangular
$s(t)$		
$e^{at}v(t)$	$1/a+j\omega$	
$e^{at}v(-t)$	$1/a-j\omega$	
$A \text{rect}(t/T)$	$A \text{rect}(\omega T/2), A \text{sinc}(\omega T/2\pi)$	
$A \Delta(t/T)$	$A \Delta(\omega T/2), A \text{sinc}^2(\omega T/2\pi)$	
$e^{j\omega_0 t}$	$2\pi \delta(\omega - \omega_0)$	
$e^{-j\omega_0 t}$	$2\pi \delta(\omega + \omega_0)$	
$\cos \omega_0 t$	$\pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$	
$\sin \omega_0 t$	$\pi j [\delta(\omega + \omega_0) - \delta(\omega - \omega_0)]$	
$\text{sgn}(t)$	$a/j\omega$	
$v(t)$	$\pi \delta(\omega) + 1/j\omega$	
A	$2\pi A \delta(\omega)$	
$e^{at}v(t)$	$2a/\omega + \omega^2$	

Properties of F.T

① Linearity property

$$\text{If } n_1(t) \xrightarrow{F} X_1(\omega)$$

$$n_2(t) \xrightarrow{F} X_2(\omega)$$

$$\text{then } \alpha n_1(t) + \beta n_2(t) \xrightarrow{F} \alpha X_1(\omega) + \beta X_2(\omega)$$

Proof:-

$$F[\alpha n_1(t) + \beta n_2(t)] = \int_{-\infty}^{\infty} (\alpha n_1(t) + \beta n_2(t)) e^{-j\omega t} dt$$

$$= \int_{-\infty}^{\infty} \alpha n_1(t) e^{-j\omega t} dt + \int_{-\infty}^{\infty} \beta n_2(t) e^{-j\omega t} dt$$

$$= \alpha \int_{-\infty}^{\infty} n_1(t) e^{-j\omega t} dt + \beta \int_{-\infty}^{\infty} n_2(t) e^{-j\omega t} dt$$

$$= \alpha [X_1(\omega)] + \beta [X_2(\omega)]$$

$$= \alpha X_1(\omega) + \beta X_2(\omega)$$

Q) $n(t) = 5 - 8\sin^2 t$

Given:- $n(t) = 5 - 8\sin^2 t$

$$n(t) = 5 - \left[\frac{1}{2} - \frac{1}{2} \cos 2t \right]$$

$$n(t) = \frac{9}{2} + \frac{1}{2} \cos 2t$$

Apply Linearity property.

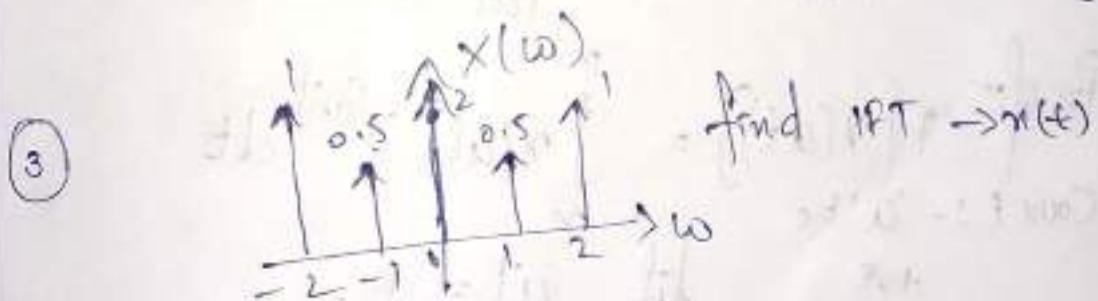
$$\begin{aligned}
 F[x(t)] &= F\left[\frac{9}{2}\right] + \frac{1}{2} F[(\omega_0)^2 t^2] \\
 &= 9\pi \times \frac{9}{2} \delta(\omega) + \frac{1}{2} \pi \left[\delta(\omega - \omega_0) + \delta(\omega + \omega_0) \right] \\
 &= 9\pi \delta(\omega) + \frac{\pi}{2} [\delta(\omega+2) + \delta(\omega-2)]
 \end{aligned}$$

$$F[n(t)] = 9\pi \delta(\omega) + \frac{\pi}{2} [\delta(\omega+2) + \delta(\omega-2)]$$

② $x(t) = 8\sin t + 2\cos 2t$

$$\begin{aligned}
 F[x(t)] &= F[8\sin t] + F[2\cos 2t] \\
 &= F[\sin t] + 2 F[\cos 2t] \\
 &= \frac{1}{j} \pi [\delta(\omega+1) - \delta(\omega-1)] \\
 &\quad + 2\pi [\delta(\omega+2) + \delta(\omega-2)]
 \end{aligned}$$

$$F[x(t)] = \pi \left[2(\delta(\omega+2) + \delta(\omega-2)) + j (\delta(\omega+1) - \delta(\omega-1)) \right]$$



$$\begin{aligned}
 X(\omega) &= 0.5\delta(\omega+2) + \frac{1}{2}\delta(\omega+1) + 2\delta(\omega) \\
 &\quad + \frac{1}{2}\delta(\omega-1) + \delta(\omega-2)
 \end{aligned}$$

$$\begin{aligned}
 f^{-1}[X(\omega)] &= f^{-1} \left\{ \delta(\omega+2) + \frac{1}{2} \delta(\omega+1) + 2\delta(\omega) \right. \\
 &\quad \left. + \frac{1}{2} \delta(\omega-1) + \delta(\omega-2) \right\}
 \end{aligned}$$

$$\hat{f}^{-1}[x(\omega)] = \hat{F}^{-1}[s(\omega+2)] + \frac{1}{2} \hat{F}^{-1}[s(\omega+1)] \\ + 2 \hat{F}^{-1}[s(\omega)] + \frac{1}{2} \hat{F}^{-1}[s(\omega-1)] \\ + \hat{F}^{-1}[s(\omega-2)]$$

$$\hat{f}^{-1}[x(\omega)] = -\frac{e^{2jt}}{2\pi} + \frac{1}{2} \frac{e^{jt}}{2\pi} + \frac{2(1)}{2\pi} \\ + \frac{1}{2} \frac{e^{jt}}{2\pi} + \frac{e^{2jt}}{2\pi}$$

$$\hat{F}^{-1}[x(\omega)] = \frac{1}{\pi} + \frac{2\cos\omega t}{2\pi} + \frac{1}{2} \left(2\cos\omega t \right) \frac{1}{2\pi}$$

* Time Scaling Property:-

If $x(t) \xrightarrow{F} X(\omega)$ then

$$x(at) \xrightarrow{F} \frac{1}{|a|} X\left(\frac{\omega}{a}\right)$$

Proof:- $F[x(at)] = \int_{-\infty}^{\infty} x(at) e^{-j\omega t} dt$

Case I :- a be

+ve

$$dt = at = p \quad dt = \frac{dp}{a}$$

$$= \int_{-\infty}^{\infty} x(p) e^{-j\omega t/a} \frac{dp}{a}$$

$$= \frac{1}{a} \int_{-\infty}^{\infty} x(p) e^{-j\omega p/a} dp$$

$$\boxed{F[n(at)] = \frac{1}{a} \times \left(\frac{\omega}{a}\right)}$$

Case ii) $a < 0$

$$F[n(at)] F[n(-at)] = \int_{-\infty}^{\infty} n(-at) e^{j\omega t} dt$$

$$\text{Let } -at = u \Rightarrow dt = -\frac{1}{a} du$$

$$\begin{aligned} F[n(-at)] &= \frac{1}{a} \int_{-\infty}^{\infty} n(u) e^{\frac{j\omega}{a} u} du \\ &= \frac{1}{a} \int_{-\infty}^{\infty} n(u) e^{\frac{j\omega}{a} u} du \end{aligned}$$

$$F[n(-at)] = \frac{1}{a} \times (-\omega/a)$$

$$\therefore \boxed{F[n(at)] = \frac{1}{|a|} \times \left(\frac{\omega}{a}\right)}$$

$$\textcircled{1} \quad F[n(t+b)]$$

$$\textcircled{2} \quad F[n(\omega at)]$$

$$\text{Sol: } F[n(\omega at)] = \frac{1}{a}$$

$$\cos t = \pi (\delta(\omega+1) + \delta(\omega-1))$$

$$\omega at = \frac{1}{2} \pi [\delta(\frac{\omega+1}{2}) + \delta(\frac{\omega-1}{2})]$$

$$\boxed{\delta(at-b) = \frac{1}{a} \delta(t-\frac{b}{a})}$$

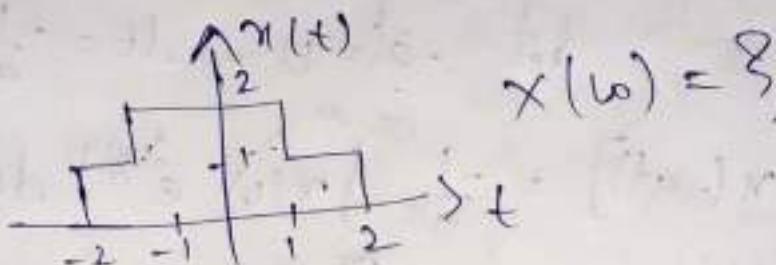
$$\boxed{|\delta(at) = \frac{1}{a} \delta(t)|}$$

$$\cos \omega t = \frac{1}{2} \pi \left[g\left(\frac{\omega+2}{2}\right) + g\left(\frac{\omega-2}{2}\right) \right]$$

$$= \frac{1}{2} \pi [2g(\omega+2) + 2g(\omega-2)]$$

$$\left(\cos \omega t = \pi [g(\omega+2) + g(\omega-2)] \right)$$

(2)



$$x(\omega) = ?$$

Sol method 1:-

$$n(t) = 1, -2 \leq t \leq -1$$

$$= 2, -1 \leq t \leq 1$$

$$= 1, 1 \leq t \leq 2$$

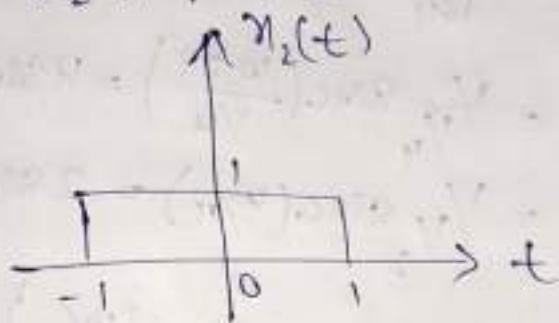
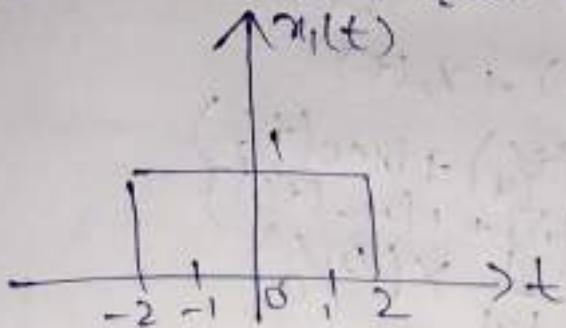
$$x(\omega) = \int_{-\infty}^{\infty} n(t) e^{-j\omega t} dt$$

$$= \int_{-2}^{-1} 1 \cdot e^{-j\omega t} dt + \int_{-1}^{1} 2 \cdot e^{-j\omega t} dt + \int_{1}^{2} 1 \cdot e^{-j\omega t} dt$$

=

method 2:-

$$n(t) = n_1(t) + n_2(t)$$



$$F[n(t)] = F[n_1(t) + n_2(t)]$$

$$F[n(t)] = F[n_1(t)] + F[n_2(t)]$$

$$\rightarrow A \operatorname{rect}\left(\frac{t}{T_1}\right) \longleftrightarrow A T \sin\left(\frac{\omega T}{2\pi}\right)$$

$$F[n_1(t)] = F[n_1(t)] + F[n_2(t)]$$

$$n_1(t) \Rightarrow A=1, T=4, \operatorname{rect}\left(\frac{t}{T}\right) = \operatorname{rect}\left(\frac{t}{4}\right) \\ = 4 \sin\left(\frac{4\omega}{2\pi}\right)$$

$$n_2(t) \Rightarrow A=1, T=2, \operatorname{rect}\left(\frac{t}{T}\right) = \operatorname{rect}\left(\frac{t}{2}\right) \\ = 2 \sin\left(\frac{2\omega}{2\pi}\right)$$

$$\boxed{F[n(t)] = 4 \sin\left(\frac{2\omega}{\pi}\right) + 2 \sin\left(\frac{\omega}{\pi}\right)}$$

Method 3:-

$$\text{Dirac}(t) \longrightarrow \text{sinc}\left(\frac{\omega}{2\pi}\right)$$

$$x(t) = x_1(t) + x_2(t)$$

$$= \text{Dirac}\left(\frac{t}{4}\right) + \text{Dirac}\left(\frac{t}{2}\right)$$

$$f[x(t)] = F[\text{Dirac}\left(\frac{t}{4}\right)] + F[\text{Dirac}\left(\frac{t}{2}\right)]$$

$$x(\omega) \rightarrow \frac{1}{|a|} \times \left(\frac{\omega}{a}\right)$$

$$\text{Dirac}\left(\frac{t}{4}\right) = \frac{1}{1/4} \text{sinc}\left(\frac{\omega/2\pi}{1/4}\right) = 4 \text{sinc}\left(\frac{2\omega}{\pi}\right)$$

$$\text{Dirac}\left(\frac{t}{2}\right) = \frac{1}{1/2} \text{sinc}\left(\frac{\omega}{\pi}\right) = 2 \text{sinc}\left(\frac{\omega}{\pi}\right)$$

$$x(\omega) = 4 \text{sinc}\left(\frac{2\omega}{\pi}\right) + 2 \text{sinc}\left(\frac{\omega}{\pi}\right)$$

3. Time reversal property :-

If $x(t) \xrightarrow{F} X(\omega)$
 then $x(-t) \rightarrow X(-\omega)$

Proof:- From time scaling property

$$x(at) \rightarrow \frac{1}{|a|} X\left(\frac{\omega}{a}\right)$$

$$x(-t) \xrightarrow{F} \frac{1}{|-1|} X\left(\frac{\omega}{-1}\right)$$

$$\boxed{x(-t) \rightarrow X(-\omega)}$$

6) $F[V(-t)]$

$$\text{sol} \quad V(t) \rightarrow \pi g(\omega) + \frac{1}{j\omega} \quad \boxed{\uparrow}$$

From time reversal property

$$\begin{cases} V(-t) \rightarrow \pi g(-\omega) - \frac{1}{j\omega} \\ V(-t) \rightarrow \pi g(\omega) - \frac{1}{j\omega} \end{cases}$$

② $F[\text{sgn}(t)]$

$$\text{sol} \quad \text{sgn}(t) = V(t) - V(-t) \leftarrow 2V(t) - 1$$

$$F[\text{sgn}(t)] = F[V(t)] - F[V(-t)]$$

$$= \pi g(\omega) + \frac{1}{j\omega} - \cancel{\pi g(-\omega)} + \frac{1}{j\omega}$$

$$\boxed{F[\text{sgn}(t)] = \frac{2}{j\omega}}$$

③ $x(t) = e^{at} v(-t)$ [Find $F[x(t)]$] $\times(\omega)$

$$\text{sol} \quad \cancel{e^{at} v(t)} \rightarrow \frac{1}{a+j\omega} \quad (1)$$

$$\cancel{e^{a(-t)} v(-t)} \rightarrow \frac{1}{a-j\omega}$$

$$e^{at} v(-t) \rightarrow \frac{1}{a-j\omega}$$

$$\textcircled{1} \quad X(\omega) = \frac{5j\omega + 12}{(j\omega)^2 + 5j\omega + 6}, \quad x(t) = ?$$

$$\textcircled{2} \quad X(\omega) = \frac{5j\omega + 12}{(j\omega)^2 + 12j\omega + 6}$$

$$X(\omega) = \frac{5j\omega + 12}{(j\omega + 3)(j\omega + 2)}$$

$$X(\omega) = \frac{(12 - 10j - 2j^2)}{j\omega + 2} + \frac{12 - 5j(3) / -3 + 2}{j\omega + 3}$$

$$= \frac{2}{2 + j\omega} + \frac{3}{j\omega + 3}$$

Apply IFT & linearity.

$$\tilde{f}[X(\omega)] = 2 \tilde{f}\left[\frac{1}{2+j\omega}\right] + 3 \tilde{f}\left[\frac{1}{3+j\omega}\right]$$

$$\boxed{x(t) = 2e^{-2t} u(t) + 3e^{-3t} u(t)}$$

$$\textcircled{3} \quad X(\omega) = \frac{1}{2 - \omega^2 + 3j\omega}, \quad x(t) = ?$$

$$\textcircled{4} \quad X(\omega) = \frac{1}{(j\omega)^2 + 3j\omega + 2}$$

$$X(\omega) = \frac{1}{(j\omega + 1)(j\omega + 2)}$$

$$X(\omega) = \frac{1}{j\omega + 1} - \frac{1}{j\omega + 2}$$

$$\boxed{x(\omega) = -e^{-t} v(t) - e^{2t} v(t)}$$

④ Property 4 Time Shifting:-

If $x(t) \xrightarrow{F} X(\omega)$

$$x(t-t_0) \xrightarrow{F} e^{-j\omega t_0} X(\omega)$$

$$n(t+t_0) \xrightarrow{F} e^{j\omega t_0} X(\omega)$$

Proof: $\bar{x}(t) = \bar{f}^{-1}[X(\omega)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$

$$\begin{aligned} \bar{f}^{-1}[e^{-j\omega t_0} X(\omega)] &= \frac{1}{2\pi} \int_{-\infty}^{\infty} e^{-j\omega t_0} X(\omega) e^{j\omega t} d\omega \\ &= \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega(t-t_0)} d\omega \end{aligned}$$

$$\boxed{\bar{f}^{-1}[e^{-j\omega t_0} X(\omega)] = x(t-t_0)}$$

$$F[x(t-t_0)] = e^{-j\omega t_0} X(\omega)$$

$$\boxed{n(t-t_0) \xrightarrow{F} e^{j\omega t_0} X(\omega)}$$

① $F^{-1}[e^{-j\omega t_0}]$

Sol $X(\omega) = e^{-j\omega t_0} \cdot 1$

$x(t) = ?$

Ques? $s(t) \rightarrow 1$

$$s(t-t_0) \rightarrow 1 \cdot e^{-j\omega t_0}$$

$$\textcircled{2} \quad \mathcal{F}^{-1} \left[\frac{e^{j\omega t_0}}{a+j\omega} \right]$$

$$\text{sol} \quad e^{at} v(t) \rightarrow \frac{1}{a+j\omega}$$

$$-e^{a(t-t_0)} v(t-t_0) \rightarrow \frac{e^{j\omega t_0}}{a+j\omega}$$

$$\textcircled{3} \quad \frac{e^{-2j\omega}}{2-j\omega}$$

$$\text{sol} \quad e^{2t} v(-t) \rightarrow \frac{1}{2-j\omega}$$

$$e^{2(t-2)} v(-(t-2)) \rightarrow \frac{e^{-2j\omega}}{2-j\omega}$$

$$e^{2t-4} v(2-t) \rightarrow \frac{e^{-2j\omega}}{2-j\omega}$$

$$\mathcal{F}^{-1} \left[\frac{e^{-2j\omega}}{2-j\omega} \right] = e^{2t-4} v(2-t)$$

$$\textcircled{4} \quad x(t) = v(t+1) + v(t-1)$$

$$\text{sol} \quad X(\omega) = ?$$

$$\mathcal{F}\{x(t)\} = \mathcal{F}\{v(t+1)\} + \mathcal{F}\{v(t-1)\}$$

$$v(t) \rightarrow \pi \delta(\omega) + \frac{1}{j\omega}$$

$$X(\omega) = e^{j\omega} \left[\pi \delta(\omega) + \frac{1}{j\omega} \right] + e^{-j\omega} \left[\pi \delta(\omega) + \frac{1}{j\omega} \right]$$

$$= \frac{e^{j\omega} - e^{-j\omega}}{j\omega} + \pi \left[e^{j\omega} s(\omega) + e^{-j\omega} s(\omega) \right]$$

\therefore soft of property

$$= \frac{2 \cos \omega}{j\omega} + \pi \left[e^j s(\omega) + e^{-j} s(\omega) \right]$$

⑤ $n(t) = s(t-1) - s(t+1)$

sol $x(\omega) = f[n(t)] = f[s(t-1) - s(t+1)]$

$$x(\omega) = f[s(t-1)] - f[s(t+1)]$$

$$\begin{aligned}s(t) &\rightarrow 1 \\s(t-1) &\rightarrow 1 \cdot e^{-j\omega(1)} \\s(t+1) &\rightarrow e^{j\omega}\end{aligned}$$

$$x(\omega) = -e^{-j\omega} + e^{j\omega}$$

$$x(\omega) = 2 \cos \omega$$

⑥ $e^{-2|t-2|} = n(t)$, find $x(\omega)$

sol $\frac{-e^{at}}{e^{at}} \rightarrow \frac{2a}{a^2 + \omega^2}$

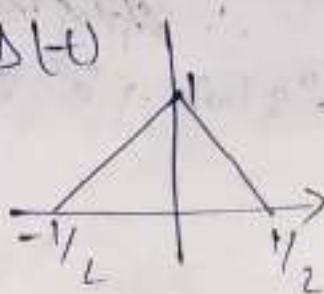
$$e^{-2t+1} \rightarrow \frac{4}{4+\omega^2}$$

$$\frac{-e^{-2|t-2|}}{e^{-2t+1}} \rightarrow \frac{4}{\omega^2 + 4} e^{-j\omega(2)} = \frac{4e^{-2j\omega}}{\omega^2 + 4}$$

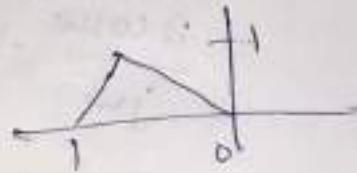
7) $x(t) = A \Delta(t - \frac{1}{2})$

Sol

$\Delta(t)$



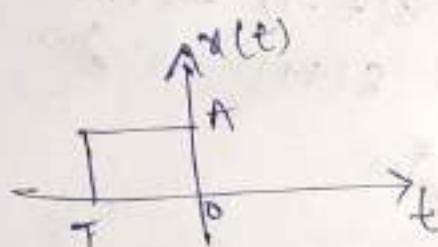
$\Delta(t + \frac{1}{2})$



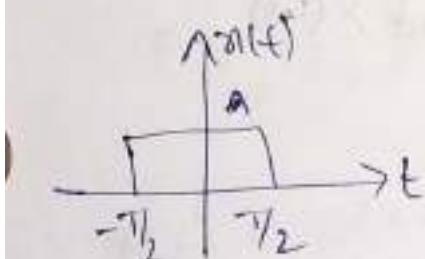
$A \Delta(t/\tau) \rightarrow A \tau \operatorname{sinc}^2\left(\frac{\omega \tau}{2\pi}\right)$

$\Delta(t) \xrightarrow{f} \operatorname{sinc}^2\left(\frac{\omega}{2\pi}\right)$

$\Delta(t + \frac{1}{2}) \rightarrow \operatorname{sinc}^2\left(\frac{\omega}{2\pi}\right) e^{j\frac{\omega}{2}\tau}$

8)find $X(\omega)$ Sol

$A \operatorname{rect}(t/\tau) \rightarrow A \tau \operatorname{sinc}\left(\frac{\omega \tau}{2}\right)$



$x(t) = A \operatorname{rect}\left(\frac{t}{\tau} + \frac{T}{2}\right)$

$A \operatorname{rect}\left(\frac{t}{\tau} + \frac{T}{2}\right) = A \tau \operatorname{sinc}\left(\frac{\omega \tau}{2\pi}\right) e^{j\frac{\omega}{2}\tau}$

9)

$x(t) = -e^{-st} u(t-1)$

Sol

$-e^{-st} u(t) \xrightarrow{f} \frac{1}{3+j\omega}$

$-e^{-s(t-1)} u(t-1) \rightarrow \frac{1}{3-j\omega} - e^{j\omega(1)}$

$$e^3 \cdot e^{-3t} v(t-1) \rightarrow \frac{-j\omega}{3+j\omega}$$

$$-e^{-3t} v(t-1) \rightarrow -e^{-3-j\omega}$$

$$-e^{-3t} v(t-1) \rightarrow \frac{-e^{3-j\omega}}{3+j\omega}$$

⑩ $F[\operatorname{sinc}\left(\frac{t+2}{4}\right)]$

$\stackrel{def}{=} n(t) = \operatorname{sinc}\left(-t/4 + 1/2\right)$

$$\operatorname{sinc}(t) \rightarrow \operatorname{sinc}\left(\frac{\omega}{2\pi}\right)$$

$$\operatorname{sinc}\left(t + \frac{1}{2}\right) \rightarrow e^{j\omega(1/2)} \operatorname{sinc}\left(\frac{\omega}{2\pi}\right)$$

$$\operatorname{sinc}\left(\frac{t}{4} + \frac{1}{2}\right) \rightarrow 4 e^{j(\omega/1/4)/2} \operatorname{sinc}\left(\frac{\omega/1/4}{2\pi}\right)$$

$$\operatorname{sinc}\left(\frac{t}{4} + 1/2\right) \rightarrow 4 e^{\frac{4j\omega}{2} \operatorname{sinc}\left(\frac{4\omega}{2\pi}\right)}$$

$$\operatorname{sinc}\left(\frac{t}{4} + 1/2\right) \rightarrow 4 e^{\frac{2j\omega}{2} \operatorname{sinc}\left(\frac{2\omega}{2\pi}\right)}$$

⑪ If $n(t) \rightarrow X(\omega)$ find F.T of
 $n(-2-t) + n(2-t)$

$\stackrel{def}{=} F[n(-2-t)]$

$$n(t) \rightarrow X(\omega)$$

$$n(t-2) \rightarrow -e^{2j\omega} X(\omega)$$

$$n(-t-2) \rightarrow e^{2j\omega} X(\omega)$$

①

$$F\{x(2-t)\} = F\{x(t+2)\}$$

$$x(t) \rightarrow X(\omega)$$

$$x(t+2) \rightarrow e^{2j\omega} X(\omega)$$

$$x(-t+2) \rightarrow e^{-2j\omega} X(-\omega)$$

$$\begin{aligned} F[x(-2-t) + x(2-t)] &= e^{2j\omega} x(-\omega) + e^{-2j\omega} X(-\omega) \\ &= X(-\omega) [e^{2j\omega} + e^{-2j\omega}] \end{aligned}$$

$$F[x(-2-t) + x(2-t)] = 2\cos(2\omega) X(-\omega),$$