Transmission of Correlated Sources over a MAC: A Gaussian Approximation-Based Analysis

Vincent Y. F. Tan

Institute for Infocomm Research (I2R) National University of Singapore (NUS)

October 2, 2012

■ We have a 2-DMS S^n , T^n with distribution

$$p_{S^n,T^n}(s^n,t^n) = \prod_{i=1}^n p_{S,T}(s_i,t_i)$$

■ They are to be separately encoded so $x_1^n = \phi_n^1(s^n) \in \mathcal{X}_1^n$ and $x_2^n = \phi_n^2(t^n) \in \mathcal{X}_2^n$

■ We have a 2-DMS S^n , T^n with distribution

$$p_{S^n,T^n}(s^n,t^n) = \prod_{i=1}^n p_{S,T}(s_i,t_i)$$

- They are to be separately encoded so $x_1^n = \phi_n^1(s^n) \in \mathcal{X}_1^n$ and $x_2^n = \phi_n^2(t^n) \in \mathcal{X}_2^n$
- Then the codewords (x_1^n, x_2^n) are sent across a DM-MAC

$$W^{n}(y^{n}|x_{1}^{n},x_{2}^{n}) = \prod_{i=1}^{n} W(y_{i}|x_{1i},x_{2i})$$

■ We have a 2-DMS S^n , T^n with distribution

$$p_{S^n,T^n}(s^n,t^n) = \prod_{i=1}^n p_{S,T}(s_i,t_i)$$

- They are to be separately encoded so $x_1^n = \phi_n^1(s^n) \in \mathcal{X}_1^n$ and $x_2^n = \phi_n^2(t^n) \in \mathcal{X}_2^n$
- Then the codewords (x_1^n, x_2^n) are sent across a DM-MAC

$$W^{n}(y^{n}|x_{1}^{n},x_{2}^{n}) = \prod_{i=1}^{n} W(y_{i}|x_{1i},x_{2i})$$

■ Decoder receives y^n and estimates \hat{s}^n and \hat{t}^n , i.e., $(\hat{s}^n, \hat{t}^n) = \psi_n(y^n)$

■ We have a 2-DMS S^n , T^n with distribution

$$p_{S^n,T^n}(s^n,t^n) = \prod_{i=1}^n p_{S,T}(s_i,t_i)$$

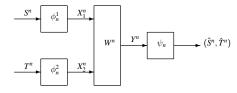
- They are to be separately encoded so $x_1^n = \phi_n^1(s^n) \in \mathcal{X}_1^n$ and $x_2^n = \phi_n^2(t^n) \in \mathcal{X}_2^n$
- Then the codewords (x_1^n, x_2^n) are sent across a DM-MAC

$$W^{n}(y^{n}|x_{1}^{n},x_{2}^{n}) = \prod_{i=1}^{n} W(y_{i}|x_{1i},x_{2i})$$

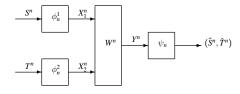
- Decoder receives y^n and estimates \hat{s}^n and \hat{t}^n , i.e., $(\hat{s}^n, \hat{t}^n) = \psi_n(y^n)$
- What is the condition on $p_{S,T}$ and W so that

$$\lim_{n\to\infty}\mathbb{P}[(\hat{S}^n,\hat{T}^n)\neq(S^n,T^n)]=0$$

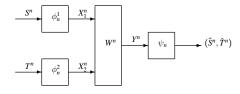




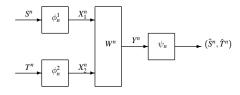
Multi-terminal joint source-channel coding problem



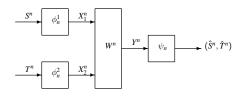
- Multi-terminal joint source-channel coding problem
- First studied by Cover-El Gamal-Salehi (1980)



- Multi-terminal joint source-channel coding problem
- First studied by Cover-El Gamal-Salehi (1980)
- Unlike the MAC for channel coding, S^n and T^n can be correlated



- Multi-terminal joint source-channel coding problem
- First studied by Cover-El Gamal-Salehi (1980)
- Unlike the MAC for channel coding, S^n and T^n can be correlated
- The ratio of channel uses to source symbols is assumed to be 1



- Multi-terminal joint source-channel coding problem
- First studied by Cover-El Gamal-Salehi (1980)
- Unlike the MAC for channel coding, S^n and T^n can be correlated
- The ratio of channel uses to source symbols is assumed to be 1
- Second-order coding analysis has been done for
 - Slepian-Wolf (Tan-Kosut 2012) Will review
 - DM-MAC (Tan-Kosut 2012, Huang-Moulin 2012 and MolavianJazi-Laneman 2012) – Will review
 - JSCC (Wang-Ingber-Kochman 2011, Kostina-Verdú 2012)

■ We propose a joint source-channel coding scheme to determine the set of $(p_{S,T}, W)$ that are (n, ϵ) -transmissible

- We propose a joint source-channel coding scheme to determine the set of $(p_{S,T}, W)$ that are (n, ϵ) -transmissible
- (n, ϵ) -transmissibility means that we consider a finite blocklength n and allow a decoding error probability ϵ

- We propose a joint source-channel coding scheme to determine the set of $(p_{S,T}, W)$ that are (n, ϵ) -transmissible
- (n, ϵ) -transmissibility means that we consider a finite blocklength n and allow a decoding error probability ϵ
- Unlike the Cover-El Gamal-Salehi (CES) condition, all constraints on entropies and mutual informations are coupled through an information dispersion matrix

- We propose a joint source-channel coding scheme to determine the set of $(p_{S,T}, W)$ that are (n, ϵ) -transmissible
- lacksquare (n,ϵ) -transmissibility means that we consider a finite blocklength n and allow a decoding error probability ϵ
- Unlike the Cover-El Gamal-Salehi (CES) condition, all constraints on entropies and mutual informations are coupled through an information dispersion matrix
- Our result states that (S,T) is (n,ϵ) -transmissible over W if

$$\mathbf{I} - \mathbf{H} \in \frac{S_{\mathbf{V}}(\epsilon)}{\sqrt{n}} + O\left(\frac{\log n}{n}\right) \mathbf{1}_6$$

- We propose a joint source-channel coding scheme to determine the set of $(p_{S,T}, W)$ that are (n, ϵ) -transmissible
- (n, ϵ) -transmissibility means that we consider a finite blocklength n and allow a decoding error probability ϵ
- Unlike the Cover-El Gamal-Salehi (CES) condition, all constraints on entropies and mutual informations are coupled through an information dispersion matrix
- Our result states that (S,T) is (n,ϵ) -transmissible over W if

$$\mathbf{I} - \mathbf{H} \in \frac{\mathcal{S}_{\mathbf{V}}(\epsilon)}{\sqrt{n}} + O\left(\frac{\log n}{n}\right) \mathbf{1}_6$$

■ Surprisingly, at blocklength *n*, two inequalities that can be removed for CES cannot be eliminated in our second-order characterization

■ In the asymptotic setting, by combining Slepian-Wolf coding and by encoding the compressed bits using a multiple-access channel code, if there exists $p_Q(q)$, $p_{X_1|Q}(x_1|q)$, $p_{X_2|Q}(x_2|q)$ such that

$$H(S|T) < I(X_1; Y|X_2, Q)$$

 $H(T|S) < I(X_2; Y|X_1, Q)$
 $H(S,T) < I(X_1, X_2; Y|Q)$

then the probability of decoding error tends to zero

■ In the asymptotic setting, by combining Slepian-Wolf coding and by encoding the compressed bits using a multiple-access channel code, if there exists $p_Q(q)$, $p_{X_1|Q}(x_1|q)$, $p_{X_2|Q}(x_2|q)$ such that

$$H(S|T) < I(X_1; Y|X_2, Q)$$

 $H(T|S) < I(X_2; Y|X_1, Q)$
 $H(S, T) < I(X_1, X_2; Y|Q)$

then the probability of decoding error tends to zero

Corresponds to a separation strategy

■ In the asymptotic setting, by combining Slepian-Wolf coding and by encoding the compressed bits using a multiple-access channel code, if there exists $p_Q(q)$, $p_{X_1|Q}(x_1|q)$, $p_{X_2|Q}(x_2|q)$ such that

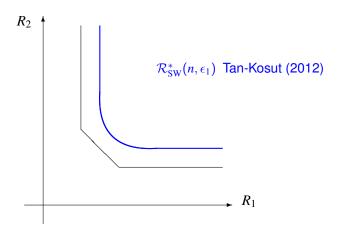
$$H(S|T) < I(X_1; Y|X_2, Q)$$

 $H(T|S) < I(X_2; Y|X_1, Q)$
 $H(S, T) < I(X_1, X_2; Y|Q)$

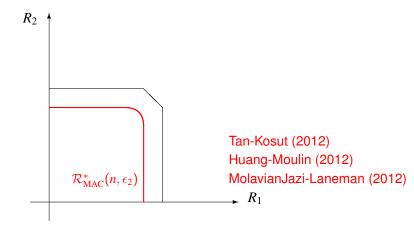
then the probability of decoding error tends to zero

- Corresponds to a separation strategy
- CES showed that this is strictly suboptimal

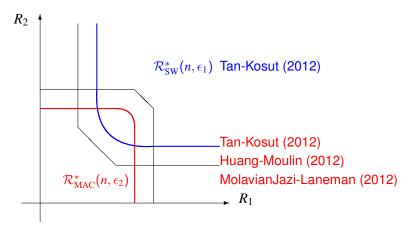
Slepian-Wolf and MAC finite blocklength regions



Slepian-Wolf and MAC finite blocklength regions



Slepian-Wolf and MAC finite blocklength regions



If $\mathcal{R}^*_{SW}(n, \epsilon_1) \cap \mathcal{R}^*_{MAC}(n, \epsilon_2) \neq \emptyset$ for some $p(q), p(x_1|q), p(x_2|q)$, then source is $(n, \tilde{\epsilon})$ -transmissible using a separation strategy where

$$\tilde{\epsilon} = \epsilon_1 + \epsilon_2 - \epsilon_1 \epsilon_2$$

■ Define K to be the common part of S and T (Gács-Körner)

- Define K to be the common part of S and T (Gács-Körner)
- CES: If there exists $p_C(c)$, $p_{X_1|S,C}(x_1|s,c)$, $p_{X_2|T,C}(x_2|t,c)$ such that

$$H(S|T) < I(X_1; Y|X_2, T, C)$$

 $H(T|S) < I(X_2; Y|X_1, S, C)$
 $H(T|S, K) < I(X_1, X_2; Y|K, C)$
 $H(S, T) < I(X_1, X_2; Y)$

then the probability of decoding error tends to zero

- Define K to be the common part of S and T (Gács-Körner)
- CES: If there exists $p_C(c)$, $p_{X_1|S,C}(x_1|s,c)$, $p_{X_2|T,C}(x_2|t,c)$ such that

$$H(S|T) < I(X_1; Y|X_2, T, C)$$

 $H(T|S) < I(X_2; Y|X_1, S, C)$
 $H(T|S, K) < I(X_1, X_2; Y|K, C)$
 $H(S, T) < I(X_1, X_2; Y)$

then the probability of decoding error tends to zero

■ Common part *K* is represented by the independent auxiliary rv *C*

- Define K to be the common part of S and T (Gács-Körner)
- CES: If there exists $p_C(c)$, $p_{X_1|S,C}(x_1|s,c)$, $p_{X_2|T,C}(x_2|t,c)$ such that

$$H(S|T) < I(X_1; Y|X_2, T, C)$$

 $H(T|S) < I(X_2; Y|X_1, S, C)$
 $H(T|S, K) < I(X_1, X_2; Y|K, C)$
 $H(S, T) < I(X_1, X_2; Y)$

then the probability of decoding error tends to zero

- Common part *K* is represented by the independent auxiliary rv *C*
- Chosen to maximize cooperation among encoders

- Define K to be the common part of S and T (Gács-Körner)
- CES: If there exists $p_C(c)$, $p_{X_1|S,C}(x_1|s,c)$, $p_{X_2|T,C}(x_2|t,c)$ such that

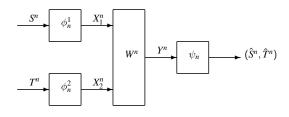
$$H(S|T) < I(X_1; Y|X_2, T, C)$$

 $H(T|S) < I(X_2; Y|X_1, S, C)$
 $H(T|S, K) < I(X_1, X_2; Y|K, C)$
 $H(S, T) < I(X_1, X_2; Y)$

then the probability of decoding error tends to zero

- Common part *K* is represented by the independent auxiliary rv *C*
- Chosen to maximize cooperation among encoders
- Generalization of the CES region when we only allow a blocklength n?

Formal Definition of a Code



Definition

An $(|\mathcal{S}|^n, |\mathcal{T}|^n, n, \epsilon)$ joint source-channel code for transmitting the correlated source (S, T) over the MAC W consists of two encoders $\phi_n^1: \mathcal{S}^n \to \mathcal{X}_1^n, \, \phi_n^2: \mathcal{T}^n \to \mathcal{X}_2^n$ and a decoder $\psi_n: \mathcal{Y}^n \to \mathcal{S}^n \times \mathcal{T}^n$ s.t.

$$\mathbb{P}[(\hat{S}^n, \hat{T}^n) \neq (S^n, T^n)] \leq \epsilon$$

Definition

We say that the source (S,T) can be (n,ϵ) -transmissible over the DM-MAC W if there exists an $(|S|^n, |\mathcal{T}|^n, n, \epsilon)$ joint source-channel code.

Definition

We say that the source (S,T) can be (n,ϵ) -transmissible over the DM-MAC W if there exists an $(|S|^n, |\mathcal{T}|^n, n, \epsilon)$ joint source-channel code.

Definition

Define the set

$$S_{\mathbf{V}}(\epsilon) = \left\{ \mathbf{z} \in \mathbb{R}^d : \mathbb{P}(\mathbf{Z} \le \mathbf{z}) \ge 1 - \epsilon \right\}$$

where $Z \sim \mathcal{N}(0, V)$.

Definition

We say that the source (S,T) can be (n,ϵ) -transmissible over the DM-MAC W if there exists an $(|S|^n, |\mathcal{T}|^n, n, \epsilon)$ joint source-channel code.

Definition

Define the set

$$S_{\mathbf{V}}(\epsilon) = \left\{ \mathbf{z} \in \mathbb{R}^d : \mathbb{P}(\mathbf{Z} \le \mathbf{z}) \ge 1 - \epsilon \right\}$$

where $Z \sim \mathcal{N}(0, V)$.

■ Analogue of the Q^{-1} function

Definition

We say that the source (S,T) can be (n,ϵ) -transmissible over the DM-MAC W if there exists an $(|S|^n, |\mathcal{T}|^n, n, \epsilon)$ joint source-channel code.

Definition

Define the set

$$S_{\mathbf{V}}(\epsilon) = \left\{ \mathbf{z} \in \mathbb{R}^d : \mathbb{P}(\mathbf{Z} \le \mathbf{z}) \ge 1 - \epsilon \right\}$$

where $Z \sim \mathcal{N}(0, V)$.

- Analogue of the Q^{-1} function
- $\blacksquare \text{ If } d=1,\, \mathcal{S}_{\sigma^2}(\epsilon)=[\sigma Q^{-1}(\epsilon),\infty)$

Definition

We say that the source (S,T) can be (n,ϵ) -transmissible over the DM-MAC W if there exists an $(|S|^n, |\mathcal{T}|^n, n, \epsilon)$ joint source-channel code.

Definition

Define the set

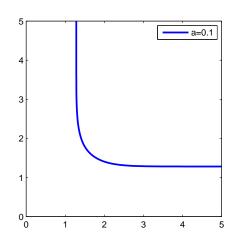
$$S_{\mathbf{V}}(\epsilon) = \left\{ \mathbf{z} \in \mathbb{R}^d : \mathbb{P}(\mathbf{Z} \leq \mathbf{z}) \geq 1 - \epsilon \right\}$$

where $Z \sim \mathcal{N}(0, V)$.

- Analogue of the Q^{-1} function
- $\blacksquare \text{ If } d=1,\, \mathcal{S}_{\sigma^2}(\epsilon)=[\sigma Q^{-1}(\epsilon),\infty)$
- This set was also used to characterize the (n, ϵ) -rate region for Slepian-Wolf coding

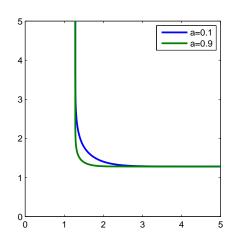
$$\mathcal{S}_{\mathbf{V}}(\epsilon) := \{\mathbf{z} \in \mathbb{R}^d : \mathbb{P}(\mathbf{Z} \leq \mathbf{z}) \geq 1 - \epsilon\} \text{ where } \mathbf{Z} \sim \mathcal{N}(\mathbf{0}, \mathbf{V})$$

$$\mathbf{V} = \begin{bmatrix} 1 & a \\ a & 1 \end{bmatrix}, a = 0.1$$



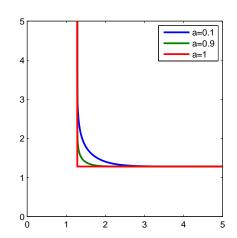
$$\mathcal{S}_{\mathbf{V}}(\epsilon) := \{\mathbf{z} \in \mathbb{R}^d : \mathbb{P}(\mathbf{Z} \leq \mathbf{z}) \geq 1 - \epsilon\} \text{ where } \mathbf{Z} \sim \mathcal{N}(\mathbf{0}, \mathbf{V})$$

$$\mathbf{V} = \begin{bmatrix} 1 & a \\ a & 1 \end{bmatrix}, a = 0.9$$



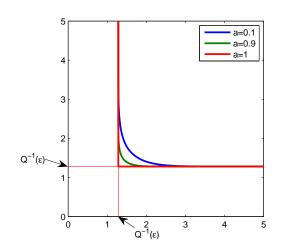
$$\mathcal{S}_{\mathbf{V}}(\epsilon) := \{\mathbf{z} \in \mathbb{R}^d : \mathbb{P}(\mathbf{Z} \leq \mathbf{z}) \geq 1 - \epsilon\} \text{ where } \mathbf{Z} \sim \mathcal{N}(\mathbf{0}, \mathbf{V})$$

$$\mathbf{V} = \begin{bmatrix} 1 & a \\ a & 1 \end{bmatrix}, a = 1.0$$



$$\mathcal{S}_{\mathbf{V}}(\epsilon) := \{\mathbf{z} \in \mathbb{R}^d : \mathbb{P}(\mathbf{Z} \leq \mathbf{z}) \geq 1 - \epsilon\} \text{ where } \mathbf{Z} \sim \mathcal{N}(\mathbf{0}, \mathbf{V})$$

$$\mathbf{V} = \left[\begin{array}{cc} 1 & a \\ a & 1 \end{array} \right]$$



Main Result

Theorem

Let K be the common part of (S,T). If for sufficiently large n, there exists distributions $p_C(c)$, $p_{X_1|S,C}(x_1|s,c)$ and $p_{X_2|T,C}(x_2|t,c)$ such that

$$\begin{bmatrix} I(X_1; Y|X_2, T, C) - H(S|T) \\ I(X_2; Y|X_1, S, C) - H(T|S) \\ I(X_1, C; Y|X_2, T) - H(S|T) \\ I(X_2, C; Y|X_1, S) - H(T|S) \\ I(X_1, X_2; Y|K, C) - H(T, S|K) \\ I(X_1, X_2; Y) - H(S, T) \end{bmatrix} \in \frac{S_{V(C, X_1, X_2)}(\epsilon)}{\sqrt{n}} + \frac{\log n}{n} \mathbf{1}_6,$$

where the information dispersion matrix $V(C, X_1, X_2)$ is the covariance of the random vector of differences between information and entropy densities (see paper).

Main Result

Theorem

Let K be the common part of (S,T). If for sufficiently large n, there exists distributions $p_C(c)$, $p_{X_1|S,C}(x_1|s,c)$ and $p_{X_2|T,C}(x_2|t,c)$ such that

$$\begin{bmatrix} I(X_1; Y|X_2, T, C) - H(S|T) \\ I(X_2; Y|X_1, S, C) - H(T|S) \\ I(X_1, C; Y|X_2, T) - H(S|T) \\ I(X_2, C; Y|X_1, S) - H(T|S) \\ I(X_1, X_2; Y|K, C) - H(T, S|K) \\ I(X_1, X_2; Y) - H(S, T) \end{bmatrix} \in \frac{S_{V(C, X_1, X_2)}(\epsilon)}{\sqrt{n}} + \frac{\log n}{n} \mathbf{1}_6,$$

where the information dispersion matrix $V(C, X_1, X_2)$ is the covariance of the random vector of differences between information and entropy densities (see paper).

Can be improved using a time-sharing rv Q [Huang and Moulin, 2012]

■ Generalization of CES and reduces to CES when $n \to \infty$

- $lue{}$ Generalization of CES and reduces to CES when $n o \infty$
- "Second-order rates" are captured by the information dispersion matrix

$$\frac{\mathcal{S}_{\mathbf{V}(C,X_1,X_2)}(\epsilon)}{\sqrt{n}}$$

- Generalization of CES and reduces to CES when $n \to \infty$
- "Second-order rates" are captured by the information dispersion matrix

$$\frac{S_{\mathbf{V}(C,X_1,X_2)}(\epsilon)}{\sqrt{n}}$$

All constraints are coupled

- Generalization of CES and reduces to CES when $n \to \infty$
- "Second-order rates" are captured by the information dispersion matrix

$$\frac{\mathcal{S}_{\mathbf{V}(C,X_1,X_2)}(\epsilon)}{\sqrt{n}}$$

- All constraints are coupled
- Gaussian approximations and Chernoff bounds in the proof

- Generalization of CES and reduces to CES when $n \to \infty$
- "Second-order rates" are captured by the information dispersion matrix

$$\frac{\mathcal{S}_{\mathbf{V}(C,X_1,X_2)}(\epsilon)}{\sqrt{n}}$$

- All constraints are coupled
- Gaussian approximations and Chernoff bounds in the proof
- \blacksquare But two extra inequalities that cannot be removed when n is finite

$$I(X_1, C; Y|X_2, T) < H(S|T) + \sqrt{\frac{V_3}{n}}$$

 $I(X_2, C; Y|X_3, S) < H(T|S) + \sqrt{\frac{V_4}{n}}$



Recall in the classical case the first constraint is

$$H(S|T) < I(X_1; Y|X_2, T, C)$$

Recall in the classical case the first constraint is

$$H(S|T) < I(X_1; Y|X_2, T, C)$$

which immediately implies "third constraint"

$$H(S|T) < I(X_1, C; Y|X_2, T)$$

because of the chain rule and $I(C; Y|X_2, T) \ge 0$.

Recall in the classical case the first constraint is

$$H(S|T) < I(X_1; Y|X_2, T, C)$$

which immediately implies "third constraint"

$$H(S|T) < I(X_1, C; Y|X_2, T)$$

because of the chain rule and $I(C; Y|X_2, T) \ge 0$.

But in the finite blocklength setting,

$$H(S|T) < I(X_1; Y|X_2, T, C) + \sqrt{V_1/n}$$

Recall in the classical case the first constraint is

$$H(S|T) < I(X_1; Y|X_2, T, C)$$

which immediately implies "third constraint"

$$H(S|T) < I(X_1, C; Y|X_2, T)$$

because of the chain rule and $I(C; Y|X_2, T) \ge 0$.

But in the finite blocklength setting,

$$H(S|T) < I(X_1; Y|X_2, T, C) + \sqrt{V_1/n}$$

does not necessarily imply

$$H(S|T) < I(X_1, C; Y|X_2, T) + \sqrt{V_3/n}$$

■ Depends on dispersions V_1 and V_3 and their correlations



Recall in the classical case the first constraint is

$$H(S|T) < I(X_1; Y|X_2, T, C)$$

which immediately implies "third constraint"

$$H(S|T) < I(X_1, C; Y|X_2, T)$$

because of the chain rule and $I(C; Y|X_2, T) \ge 0$.

But in the finite blocklength setting,

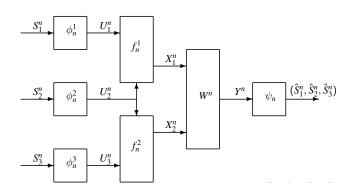
$$H(S|T) < I(X_1; Y|X_2, T, C) + \sqrt{V_1/n}$$

does not necessarily imply

$$H(S|T) < I(X_1, C; Y|X_2, T) + \sqrt{V_3/n}$$

- Depends on dispersions V_1 and V_3 and their correlations
- But region reduces to CES when $n \to \infty$

■ Consider transmitting a 3-DMS (S_1, S_2, S_3) over a MAC $W: \mathcal{X}_1 \times \mathcal{X}_2 \to \mathcal{Y}$



- Input 1 $X_1^n = f_n^1(U_1^n, U_2^n)$ and U_j^n is a stoc. function of S_j^n
- Input 2 $X_2^n = f_n^2(U_2^n, U_3^n)$ and U_j^n is a stoc. function of S_j^n

- Input 1 $X_1^n = f_n^1(U_1^n, U_2^n)$ and U_j^n is a stoc. function of S_j^n
- Input 2 $X_2^n = f_n^2(U_2^n, U_3^n)$ and U_j^n is a stoc. function of S_j^n
- This system was studied by Ahlswede and Han (1983) and specializes to the CES setting. Choose

$$S := S_1, \qquad K := S_2, \qquad T := S_3$$

- Input 1 $X_1^n = f_n^1(U_1^n, U_2^n)$ and U_j^n is a stoc. function of S_j^n
- Input 2 $X_2^n = f_n^2(U_2^n, U_3^n)$ and U_j^n is a stoc. function of S_j^n
- This system was studied by Ahlswede and Han (1983) and specializes to the CES setting. Choose

$$S := S_1, \qquad K := S_2, \qquad T := S_3$$

 A little easier to analyze using information spectrum methods [lwata and Oohama 2005]

- Input 1 $X_1^n = f_n^1(U_1^n, U_2^n)$ and U_j^n is a stoc. function of S_j^n
- Input 2 $X_2^n = f_n^2(U_2^n, U_3^n)$ and U_j^n is a stoc. function of S_j^n
- This system was studied by Ahlswede and Han (1983) and specializes to the CES setting. Choose

$$S := S_1, \qquad K := S_2, \qquad T := S_3$$

- A little easier to analyze using information spectrum methods [lwata and Oohama 2005]
- Results in seven inequalities of which one can be eliminated

- Input 1 $X_1^n = f_n^1(U_1^n, U_2^n)$ and U_j^n is a stoc. function of S_j^n
- Input 2 $X_2^n = f_n^2(U_2^n, U_3^n)$ and U_j^n is a stoc. function of S_j^n
- This system was studied by Ahlswede and Han (1983) and specializes to the CES setting. Choose

$$S := S_1, \qquad K := S_2, \qquad T := S_3$$

- A little easier to analyze using information spectrum methods [lwata and Oohama 2005]
- Results in seven inequalities of which one can be eliminated
- Because we don't have to explicitly recover the common part $S_2 = K$

Proof Idea III: (n, ϵ) -Transmissibility for Ahlswede-Han

Theorem $((n, \epsilon)$ -Transmissibility for the joint source (S_1, S_2, S_3))

If for sufficiently large n, there exists distributions $p_{U_1|S_1}, p_{U_2|S_2}, p_{U_3|S_3}$ and functions f_1 and f_2 such that

$$\left[I(U_{\mathcal{A}};Y|U_{\mathcal{A}^c},S_{\mathcal{A}^c})-H(S_{\mathcal{A}}|S_{\mathcal{A}^c}):\emptyset\neq\mathcal{A}\subset[3]\right]\in\frac{\mathcal{S}_{\mathbf{V}(U_1,U_2,U_3)}(\epsilon)}{\sqrt{n}}+\frac{\log n}{n}\mathbf{1}_7,$$

where the information dispersion matrix $V(U_1, U_2, U_3)$ is the covariance of the difference between the information and entropy densities

$$\left[\log \frac{p_{Y|U_{\mathcal{A}},U_{\mathcal{A}^{c}},S_{\mathcal{A}^{c}}}(Y|U_{\mathcal{A}},U_{\mathcal{A}^{c}},S_{\mathcal{A}^{c}})}{p_{Y|U_{\mathcal{A}^{c}},S_{\mathcal{A}^{c}}}(Y|U_{\mathcal{A}^{c}},S_{\mathcal{A}^{c}})} - \log \frac{1}{p_{S_{\mathcal{A}}|S_{\mathcal{A}^{c}}}(S_{\mathcal{A}}|S_{\mathcal{A}^{c}})} : \emptyset \neq \mathcal{A} \subset [3] \right],$$

then the source (S_1, S_2, S_3) is (n, ϵ) -transmissible over W.

Proof Idea IV: (n, ϵ) -Transmissibility for Ahlswede-Han

Construct random i.i.d. codes as per normal

Proof Idea IV: (n, ϵ) -Transmissibility for Ahlswede-Han

- Construct random i.i.d. codes as per normal
- Given y^n , the decoder searches for the unique triple of sequences $(\hat{s}_1^n, \hat{s}_2^n, \hat{s}_3^n)$ such that

$$(\hat{s}_1^n, \hat{s}_2^n, \hat{s}_3^n, u_1^n(\hat{s}_1^n), u_2^n(\hat{s}_2^n), u_3^n(\hat{s}_3^n), y^n) \in \mathcal{T}_{\delta_n}^{(n)},$$

Proof Idea IV: (n, ϵ) -Transmissibility for Ahlswede-Han

- Construct random i.i.d. codes as per normal
- Given y^n , the decoder searches for the unique triple of sequences $(\hat{s}_1^n, \hat{s}_2^n, \hat{s}_3^n)$ such that

$$(\hat{s}_1^n, \hat{s}_2^n, \hat{s}_3^n, u_1^n(\hat{s}_1^n), u_2^n(\hat{s}_2^n), u_3^n(\hat{s}_3^n), y^n) \in \mathcal{T}_{\delta_n}^{(n)},$$

where $\mathcal{T}_{\delta_n}^{(n)}$ is the intersection of all typical sets

$$\begin{split} \mathcal{T}_{\delta_{n}}^{(n)}(\mathcal{A}) &:= \left\{ (s_{1}^{n}, s_{2}^{n}, s_{3}^{n}, u_{1}^{n}, u_{2}^{n}, u_{3}^{n}, y^{n}) : \right. \\ &\left. \frac{1}{n} \log \frac{p_{Y^{n} | U_{\mathcal{A}}^{n}, U_{\mathcal{A}^{c}}^{n}, s_{\mathcal{A}^{c}}^{n}}(y^{n} | u_{\mathcal{A}}^{n}(s_{\mathcal{A}}^{n}), u_{\mathcal{A}^{c}}^{n}(s_{\mathcal{A}^{c}}^{n}), s_{\mathcal{A}^{c}}^{n})}{p_{Y^{n} | U_{\mathcal{A}}^{n}, U_{\mathcal{A}^{c}}^{n}}(y^{n} | u_{\mathcal{A}}^{n}(s_{\mathcal{A}}^{n}), u_{\mathcal{A}^{c}}^{n}(s_{\mathcal{A}^{c}}^{n}))} \\ &\left. - \frac{1}{n} \log \frac{1}{p_{S_{\mathcal{A}}^{n} | S_{\mathcal{A}^{c}}^{n}(s_{\mathcal{A}}^{n} | s_{\mathcal{A}^{c}}^{n})} \ge \delta_{n} \right\} \end{split}$$

Proof Idea V: (n, ϵ) -Transmissibility for Ahlswede-Han

The error events are

$$\mathcal{E}_{0} := \{ (S_{1}^{n}, S_{2}^{n}, S_{3}^{n}, U_{1}^{n}(S_{1}^{n}), U_{2}^{n}(S_{2}^{n}), U_{3}^{n}(S_{3}^{n}), Y^{n}) \notin \mathcal{T}_{\delta_{n}}^{(n)} \}$$

$$\mathcal{E}_{\mathcal{A}} := \{ \exists \, \tilde{s}_{\mathcal{A}}^{n} \neq S_{\mathcal{A}}^{n} : (\tilde{s}_{\mathcal{A}}^{n}, S_{\mathcal{A}^{c}}^{n}, U_{\mathcal{A}}^{n}(\tilde{s}_{\mathcal{A}}^{n}), U_{\mathcal{A}^{c}}^{n}(S_{\mathcal{A}^{c}}^{n}), Y^{n}) \in \mathcal{T}_{\delta_{n}}^{(n)} \}$$

Proof Idea V: (n, ϵ) -Transmissibility for Ahlswede-Han

■ The error events are

$$\mathcal{E}_{0} := \{ (S_{1}^{n}, S_{2}^{n}, S_{3}^{n}, U_{1}^{n}(S_{1}^{n}), U_{2}^{n}(S_{2}^{n}), U_{3}^{n}(S_{3}^{n}), Y^{n}) \notin \mathcal{T}_{\delta_{n}}^{(n)} \}$$

$$\mathcal{E}_{\mathcal{A}} := \{ \exists \, \tilde{s}_{\mathcal{A}}^{n} \neq S_{\mathcal{A}}^{n} : (\tilde{s}_{\mathcal{A}}^{n}, S_{\mathcal{A}^{c}}^{n}, U_{\mathcal{A}}^{n}(\tilde{s}_{\mathcal{A}}^{n}), U_{\mathcal{A}^{c}}^{n}(S_{\mathcal{A}^{c}}^{n}), Y^{n}) \in \mathcal{T}_{\delta_{n}}^{(n)} \}$$

■ By multi-dimensional Berry-Essèen theorem and the choice of source (S_1, S_2, S_3) and channel W

$$\mathbb{P}(\mathcal{E}_0) \approx 1 - \epsilon$$

if I - H satisfies the conditions of the theorem

Proof Idea V: (n, ϵ) -Transmissibility for Ahlswede-Han

The error events are

$$\mathcal{E}_{0} := \{ (S_{1}^{n}, S_{2}^{n}, S_{3}^{n}, U_{1}^{n}(S_{1}^{n}), U_{2}^{n}(S_{2}^{n}), U_{3}^{n}(S_{3}^{n}), Y^{n}) \notin \mathcal{T}_{\delta_{n}}^{(n)} \}$$

$$\mathcal{E}_{\mathcal{A}} := \{ \exists \, \tilde{s}_{\mathcal{A}}^{n} \neq S_{\mathcal{A}}^{n} : (\tilde{s}_{\mathcal{A}}^{n}, S_{\mathcal{A}^{c}}^{n}, U_{\mathcal{A}}^{n}(\tilde{s}_{\mathcal{A}}^{n}), U_{\mathcal{A}^{c}}^{n}(S_{\mathcal{A}^{c}}^{n}), Y^{n}) \in \mathcal{T}_{\delta_{n}}^{(n)} \}$$

■ By multi-dimensional Berry-Essèen theorem and the choice of source (S_1, S_2, S_3) and channel W

$$\mathbb{P}(\mathcal{E}_0) \approx 1 - \epsilon$$

if I - H satisfies the conditions of the theorem

■ By Chernoff bounds, for every $\emptyset \neq \mathcal{A} \subset [3]$,

$$\mathbb{P}(\mathcal{E}_{\mathcal{A}})\approx 0$$



■ Use a Gaussian approximation-based method for analyzing transmission of correlated sources over a MAC

- Use a Gaussian approximation-based method for analyzing transmission of correlated sources over a MAC
- Unlike first-order analysis, second-order characterization does not allow us to eliminate some constraints

- Use a Gaussian approximation-based method for analyzing transmission of correlated sources over a MAC
- Unlike first-order analysis, second-order characterization does not allow us to eliminate some constraints
- In our analysis, the second-order term is the variance of the difference between an information density and an entropy density

- Use a Gaussian approximation-based method for analyzing transmission of correlated sources over a MAC
- Unlike first-order analysis, second-order characterization does not allow us to eliminate some constraints
- In our analysis, the second-order term is the variance of the difference between an information density and an entropy density
- Different from point-to-point JSCC (Wang-Ingber-Kochman 2011, Kostina-Verdú 2012) where the dispersion is the sum of the source and channel dispersions

$$V_{\rm JSCC} = V_{\rm Src}(p_S) + V_{\rm Ch}(W)$$



- Use a Gaussian approximation-based method for analyzing transmission of correlated sources over a MAC
- Unlike first-order analysis, second-order characterization does not allow us to eliminate some constraints
- In our analysis, the second-order term is the variance of the difference between an information density and an entropy density
- Different from point-to-point JSCC (Wang-Ingber-Kochman 2011, Kostina-Verdú 2012) where the dispersion is the sum of the source and channel dispersions

$$V_{\rm JSCC} = V_{\rm Src}(p_S) + V_{\rm Ch}(W)$$

■ By using alternative techniques, we could potentially relax our condition on (n, ϵ) -transmissibility