

MA2011

Mechatronics System Interfacing (Part-II)

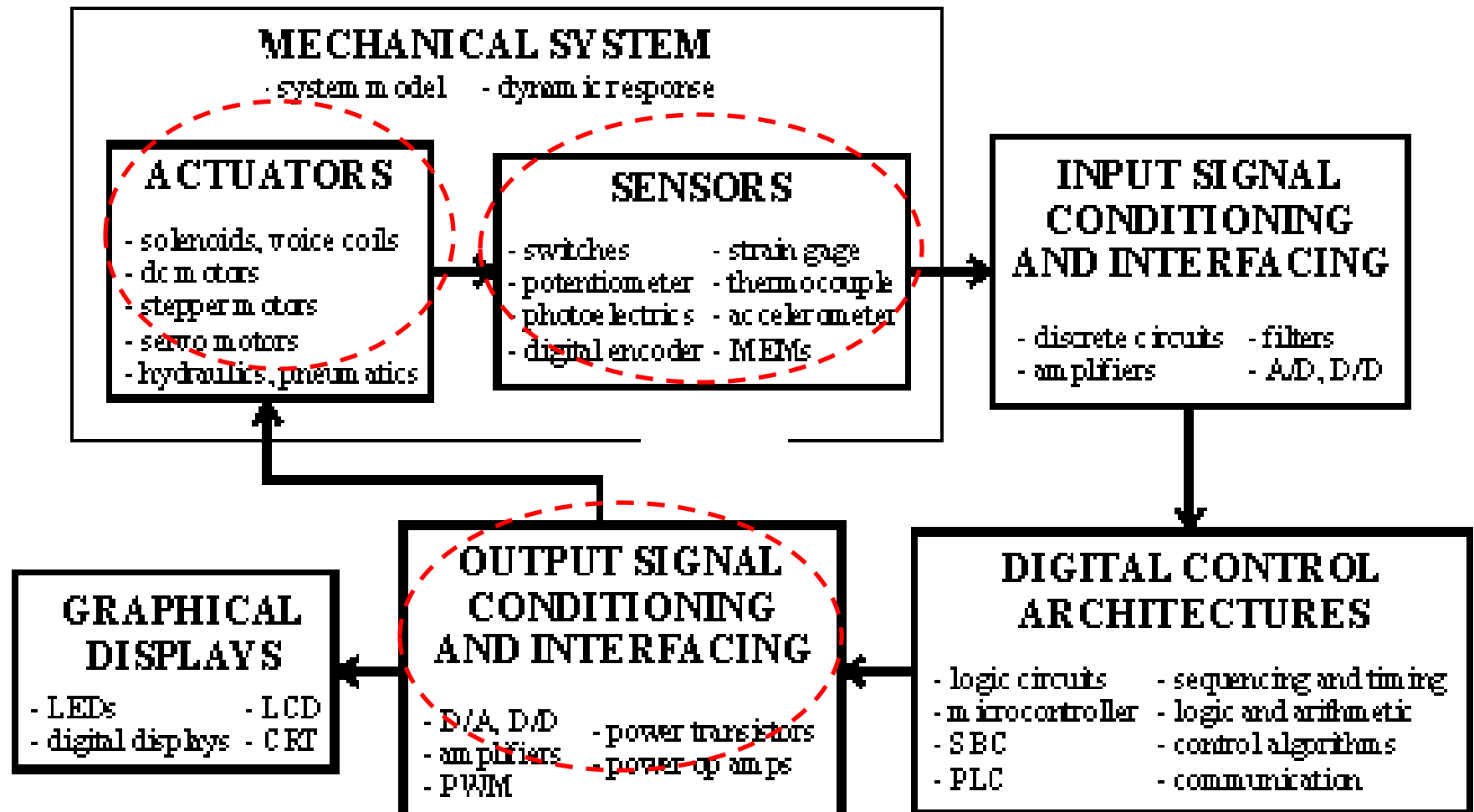
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topics in the text-book



Sensors

Basic Definitions

sensors: basic definitions

- transducers:
 - convert one form of energy into another
 - non necessarily to perform a measurement
- sensors:
 - produce an output signal (typically electrical) for the purpose of sensing a physical phenomenon

sensor classification

- analog vs digital
 - light on/off switch vs light dimmer
- passive vs active
 - passive sensors: do not require external power supplies, they draw the energy from the input signal itself
- null vs deflection type
 - null type: any deflection due to the measured quantity is balanced by an opposing calibrated force
- subject of measurement
 - mechanical, optical, thermal etc...



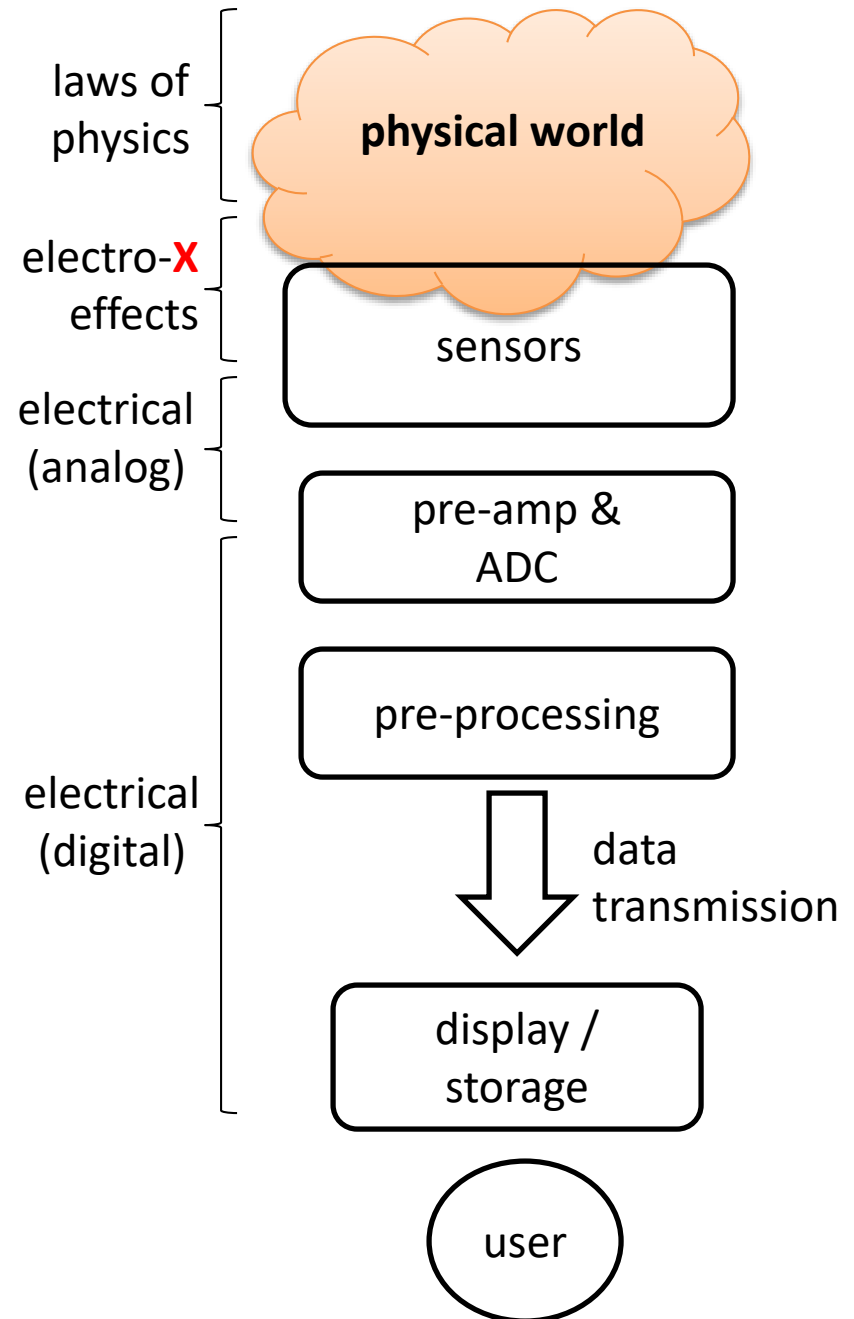
deflection-type



null-type

instrumentation systems

- sensing module
 - **X**=mechanical, thermal, optical, pyro, piezo,...
- conversion module
 - from analog to digital
- pre-processing
 - variable manipulation module
- data transmission
 - wired/wireless, over the web...
- presentation/storage
 - to the final user



basic concepts

- I/O characteristic function or response:

- **input:** stimulus or measurand
 - (temperature, pressure, strain...)
- **output:** electrical signal
 - (voltage, current, frequency, phase...)

- sensitivity S

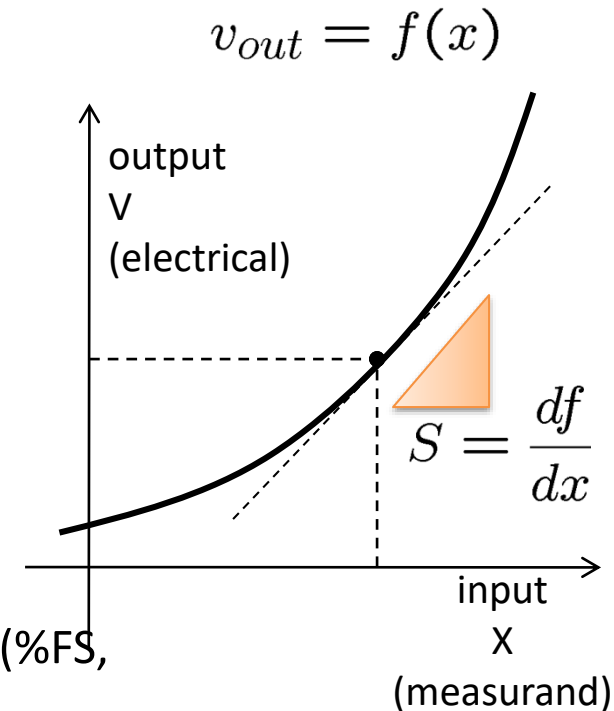
- output variation / input variation
 - slope $S = df/dx$

- resolution

- minimum change of the measurand that can be reliably detected
 - limited by noise, bit-conversion,...

- accuracy

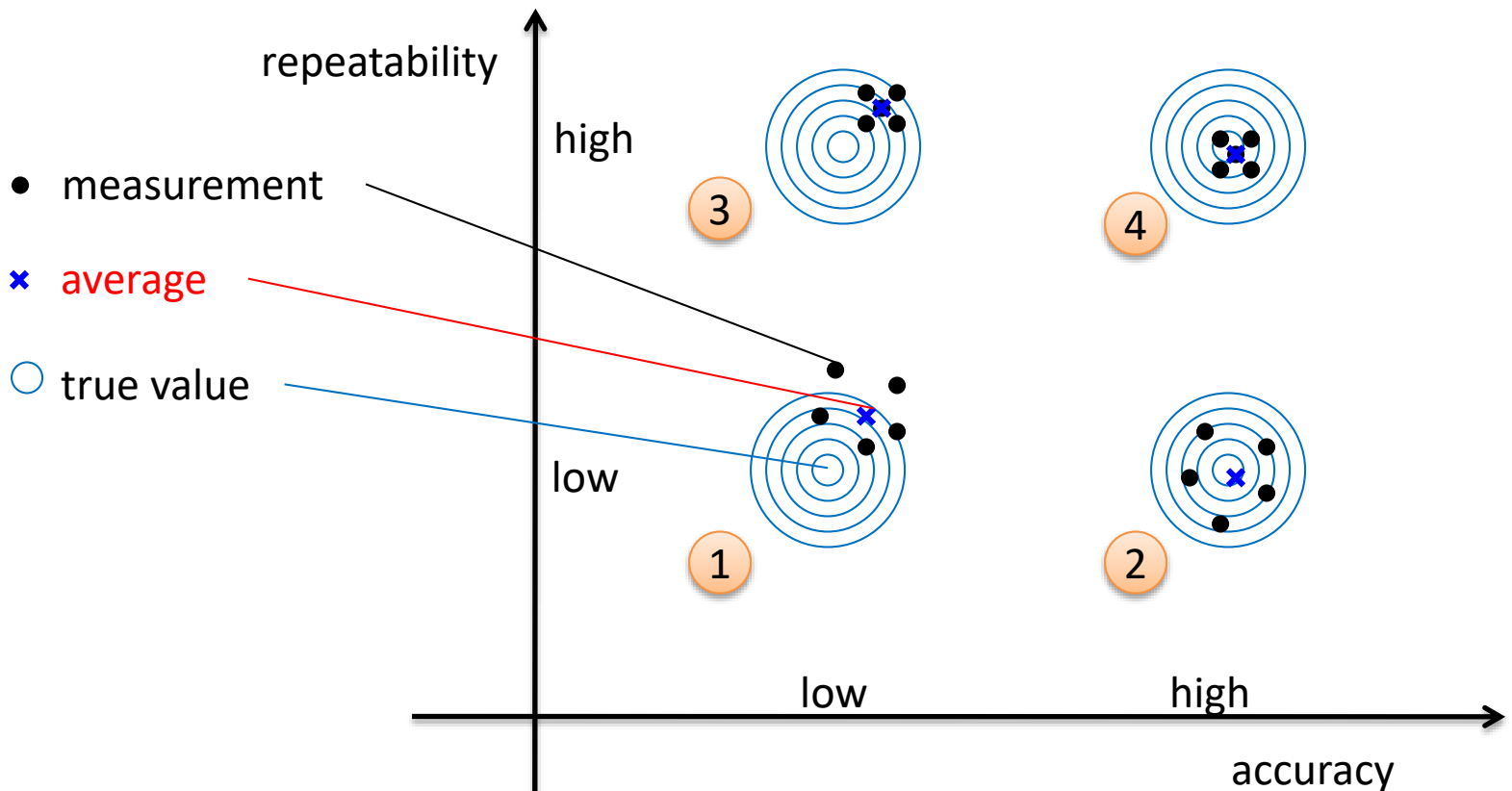
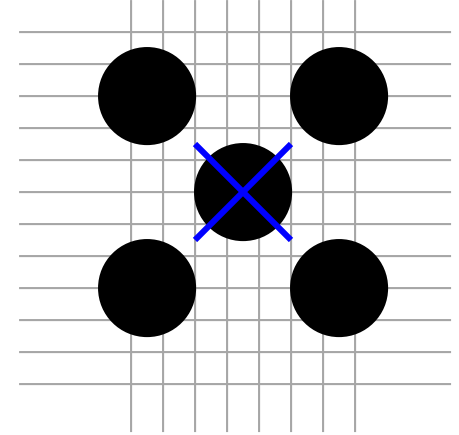
- difference of the measurement from the true value (%FS, Full Scale)



instrument static parameters

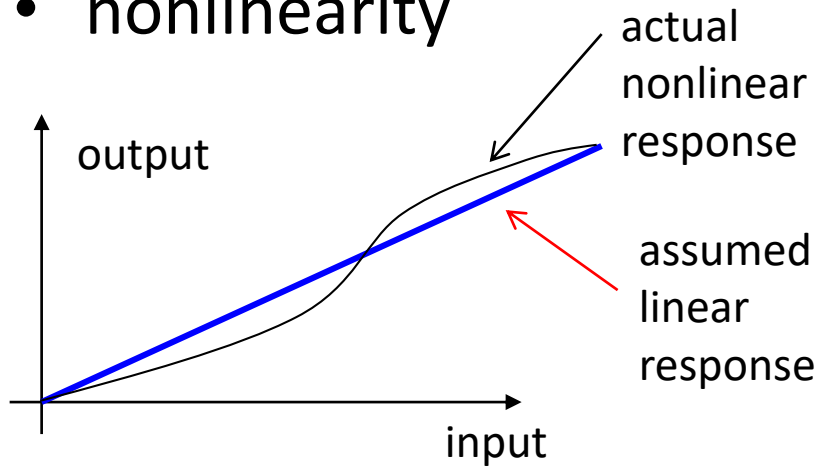
- repeatability

- how well a system or device can reproduce an outcome in unchanged conditions

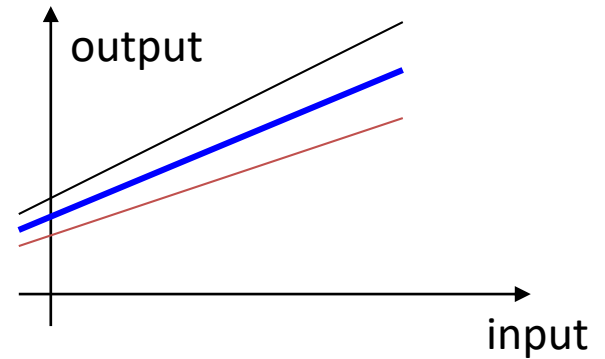


main instrument errors

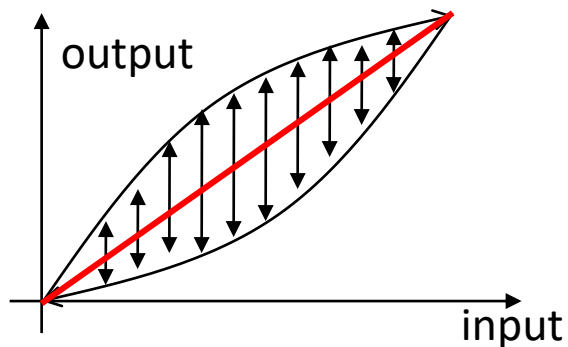
- nonlinearity



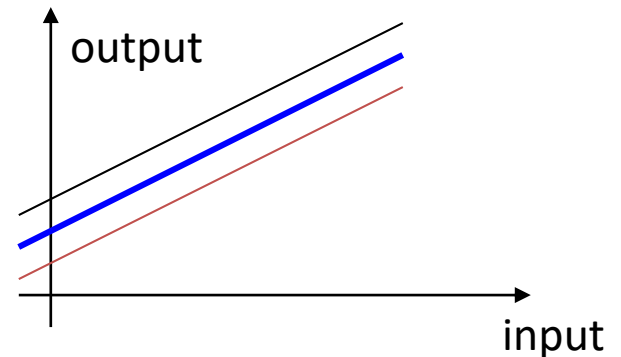
- sensitivity error



- hysteresis



- zero-shift error

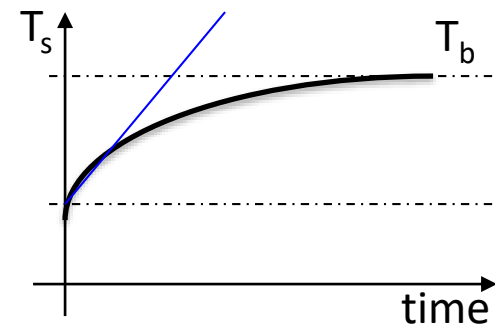
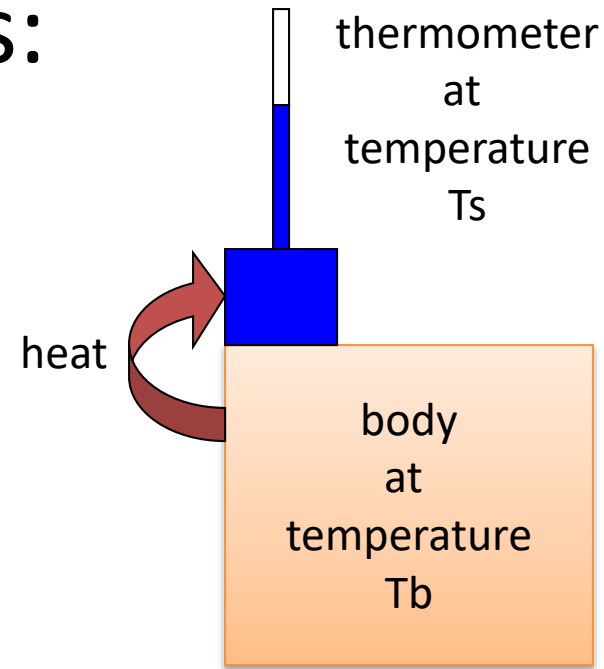


response dynamics

first and second order systems

1st order systems: an example

- thermal measurements
 - heat $q = (T_b - T_s) / R$
 - R : thermal resistance
 - materials, geometries ...
 - heat-up: $dT_s/dt = q / C$
 - C : thermal capacitance
 - if $dT_b/dt = 0 \dots$
 - $T_s(t) = T_{s0} + (T_b - T_{s0}) (1 - e^{-t/RC})$



In general, the time response of a 1st order system is
 $x(t) = a + b e^{-t/\tau}$ where τ is the time constant.

1st order systems: forced and natural response

- **general** (forced) equation

- time constant τ

- forcing input $f(t)$

- initial condition x_0

$$\begin{cases} \frac{dx(t)}{dt} + \frac{x(t)}{\tau} = f(t) \\ x(0) = x_0 \end{cases}$$

uniqueness
of solutions

- **natural** (unforced) equation

- solution

- satisfying the
initial condition $x_N(0) = K = x_0$

$$x_N(t) = Ke^{-t/\tau}$$

$$\begin{cases} \frac{dx_N(t)}{dt} + \frac{x_N(t)}{\tau} = 0 \\ x(0) = x_0 \end{cases}$$

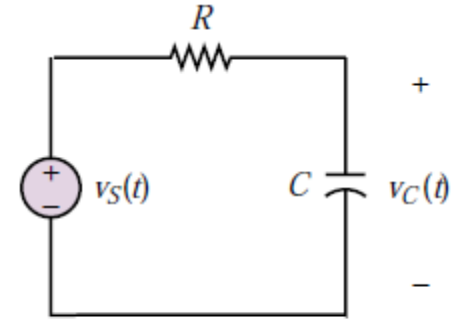
$$x_N(t) = x_0 e^{-t/\tau}$$

1st order systems: time constant

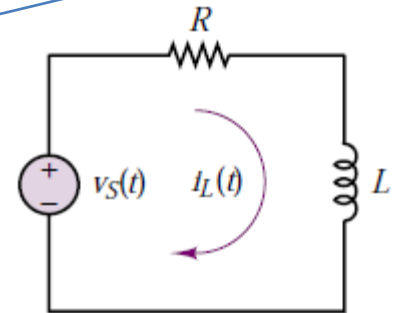
- electrical time constants
 - what are the units of RC or L/R?

$$RC = \frac{\text{voltage}}{\text{current}} \cdot \frac{\text{charge}}{\text{voltage}} = \frac{\text{charge}}{\text{charge/time}} = \text{time}$$

$$\frac{L}{R} = \frac{\text{voltage}}{\text{current/time}} \cdot \frac{\text{current}}{\text{voltage}} = \text{time}$$



$$RC \text{ circuit: } \frac{dv_C}{dt} - \frac{1}{RC} v_C - \frac{1}{RC} v_S = 0$$



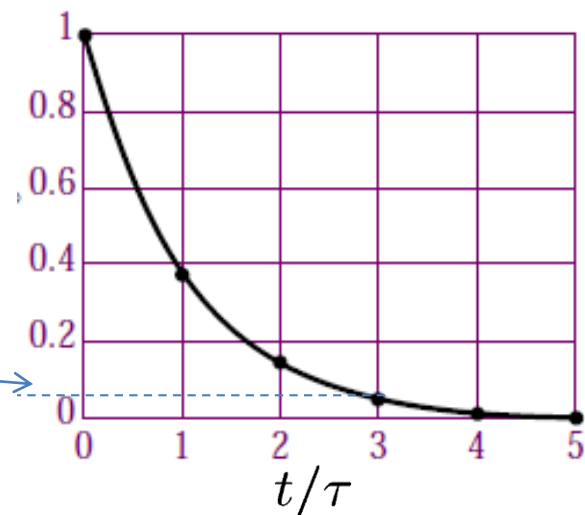
$$RL \text{ circuit: } \frac{di_L}{dt} - \frac{R}{L} i_L - \frac{1}{L} v_S = 0$$

$$x(t) = x_0 e^{-t/\tau}$$

| $\frac{x(t)}{x_0}$ | n |
|--------------------|---|
| 1 | 0 |
| 0.3679 | 1 |
| 0.1353 | 2 |
| 0.0498 | 3 |
| 0.0183 | 4 |
| 0.0067 | 5 |

after 3 time-constants
the signal decays to
5%
of its initial value

$$\frac{x(t)}{x_0}$$



1st order systems

response to **DC** forcing inputs

- we have the following problem
 - where $f(t)=F_0$ is a constant
- $$\begin{cases} \frac{dx(t)}{dt} + \frac{x(t)}{\tau} = F_0 \\ x(0) = x_0 \end{cases}$$
- let's look for a particular (forced) solution $x_F(t)$
 - consider a **DC** steady-state solution $x_F(t) = x_{SS}$
- $$\cancel{\frac{dx_{SS}}{dt}} + \frac{x_{SS}}{\tau} = F_0 \Rightarrow \frac{x_{SS}}{\tau} = F_0 \Rightarrow x_{SS} = F_0\tau = x_\infty$$
- now, let's determine a general solution
 - including the natural solution
$$x(t) = x_N(t) + x_{SS}(t) = Ke^{-t/\tau} + x_\infty$$
 - satisfying the initial condition
- $$x(0) = K + x_\infty = x_0 \Rightarrow \boxed{x(t) = \underbrace{(x_0 - x_\infty)e^{-t/\tau}}_{\text{transient}} + \underbrace{x_\infty}_{\text{steady state}}}$$

second order dynamics

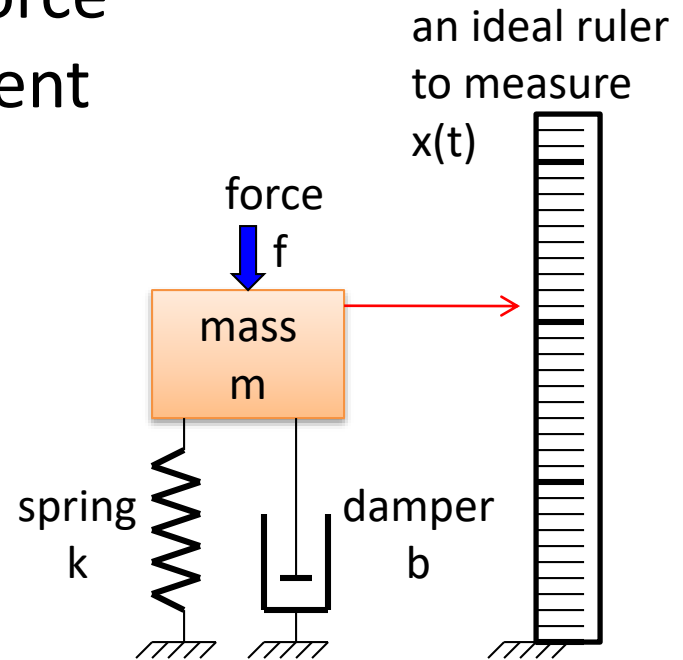
- we can indirectly estimate the force from a displacement measurement

- dynamic equations

- $f = m\ddot{x} + b\dot{x} + kx$

- frequency response

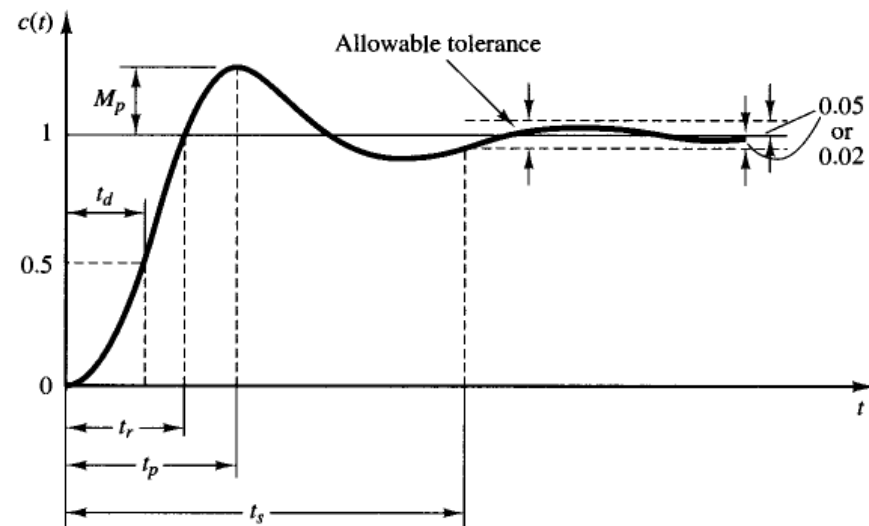
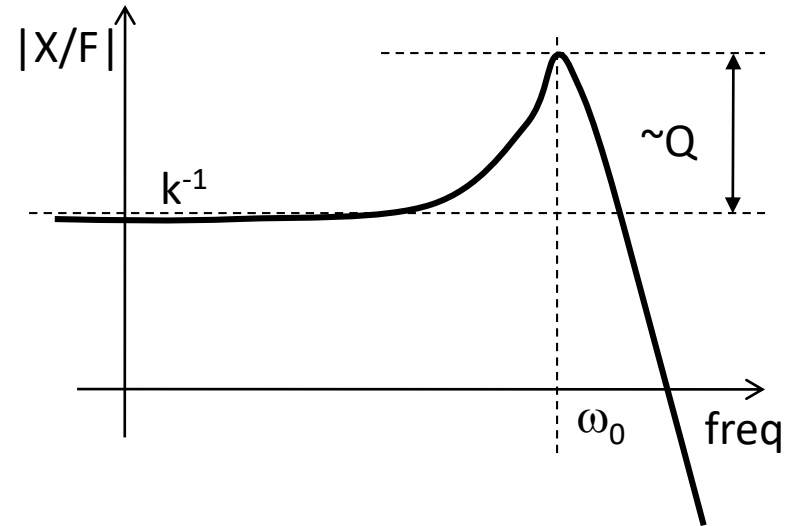
- $$\frac{X}{F} = \frac{k^{-1}}{-\frac{\omega^2}{\omega_0^2} + \frac{j\omega}{Q\omega_0} + 1}$$



second order dynamics

$$\frac{X}{F} = \frac{k^{-1}}{-\frac{\omega^2}{\omega_0^2} + \frac{j\omega}{Q\omega_0} + 1}$$

- dynamic response characterized by
 - frequency response
 - resonance $\omega_0^2 = k/m$
 - bandwidth
 - mechanical Q: $Q^2 = km / b^2$
 - $Q=0.5 \rightarrow$ critical damping
 - time response
 - rise time
 - settling time
 - overshoot
 - ...

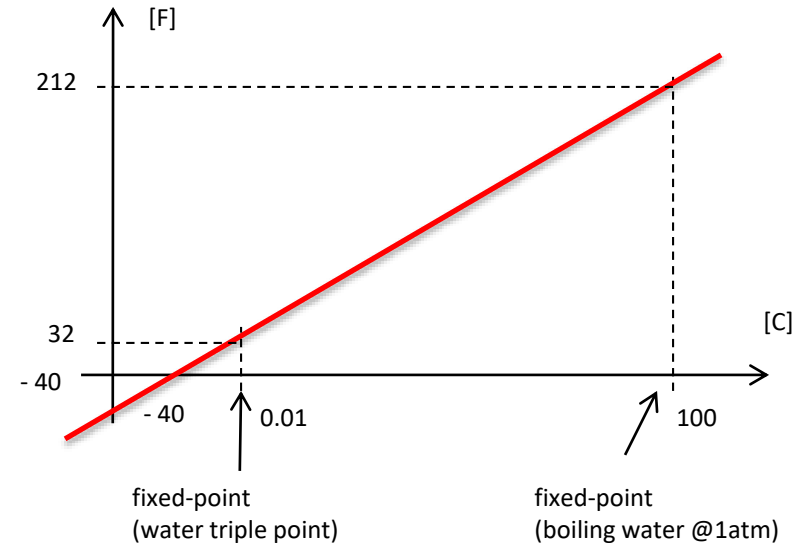


measuring temperature

RTD and Thermistors

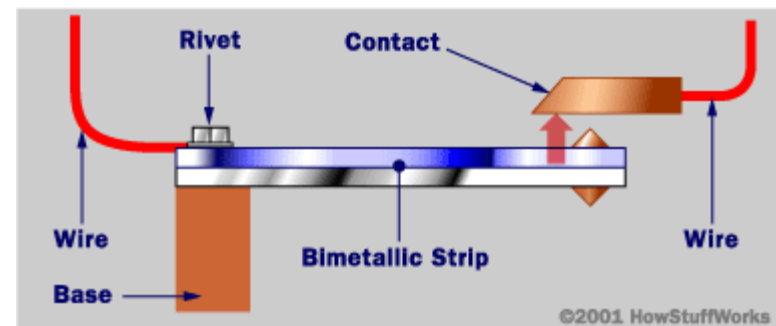
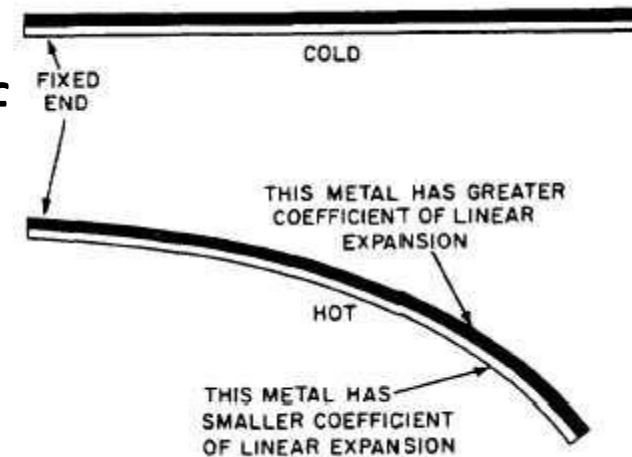
temperature scales

- Kelvin [K], Celsius [C], Fahrenheit [F]
 - $[C] = [K] - 273.15$
 - $[F] = 1.8 * [C] + 32$
- based on
 - fixed-points
 - i.e. temperatures at phase transition, triple points...
 - size of the degree
 - e.g. 1/100 of the difference between icy and boiling water
 - interpolation method in-between fixed points
 - does 50[C] correspond to the level mercury which is half-way between the 0[C] and 100[C] levels?
- ITS-90 standard



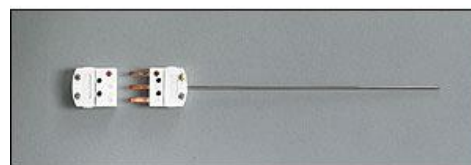
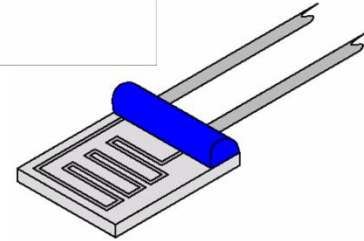
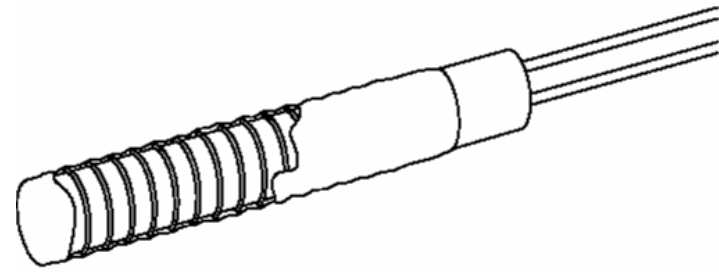
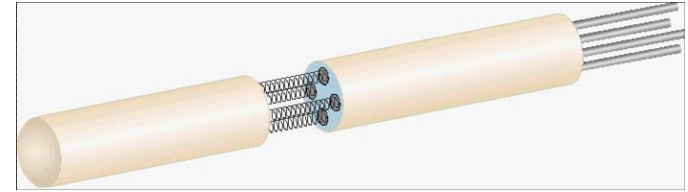
bimetallic thermometers

- differential thermal expansion of different metals
 - metal A and B bonded at temperature T_1
 - bending occurs at different temperatures
- furnace thermostat
 - switch-control



resistance temperature detectors (RTD)

- based on changes of resistance with temperature
 - **metal wire** on insulating support
 - eliminate mechanical strain
 - encasing
 - minimize environment influence (e.g. corrosion)



Typical RTD Probes



Thick Film Omega Film Element



Glass sealed Bifilar Winding



Thin Film Omega TFD Element

RTD: linearity range

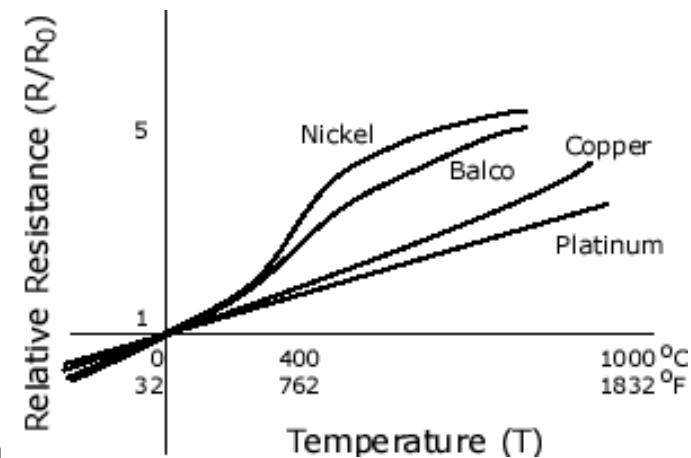
- for a given material, a linear relationship can be assumed for a **limited range**

- $$R/R_0 = 1 + \alpha(T - T_0)$$

- R resistance at temp. T ([C] or [K])
- R₀: resistance at temp. T₀
- α: temp. coefficient

- platinum

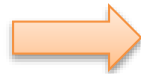
- ±0.3% over the range 0-200 [C]
- ±1.2% over the range 200-800 [C]



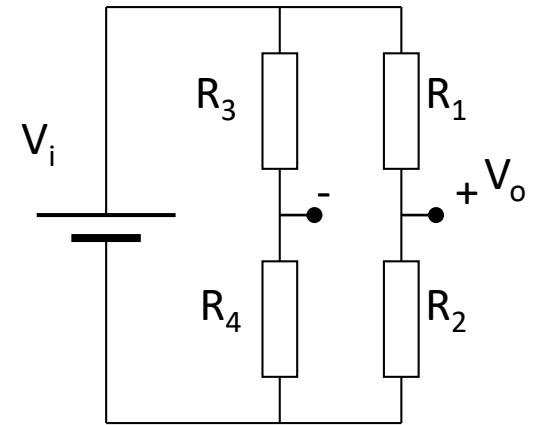
Wheatstone bridge

- bridge equations

$$\begin{cases} \frac{V^+}{V_i} = \frac{R_2}{R_1 + R_2} \\ \frac{V^-}{V_i} = \frac{R_4}{R_3 + R_4} \end{cases}$$



$$\frac{V_o}{V_i} = \frac{R_2}{R_1 + R_2} - \frac{R_4}{R_3 + R_4}$$



- bridge balance condition:

$$V_o = 0 \quad \Leftrightarrow \quad R_1 R_4 = R_2 R_3$$

(product of opposite sides)

RTD: numerical example

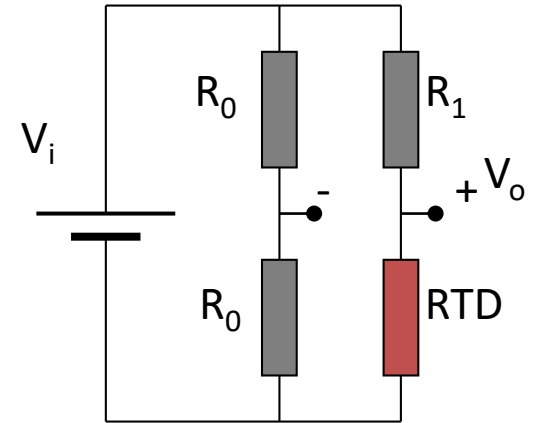
- An RTD forms one arm of an equal-arm Wheatstone bridge

$$R_0 = R_0 = 25\Omega$$

$$\text{At } 0^\circ\text{C}, \text{RTD} = 25\Omega$$

$$\text{and } \alpha = 0.003925\text{ }^\circ\text{C}^{-1}$$

- If R_3 required to balance the bridge is 37.36Ω , find the temperature of the RTD



RTD: numerical example (solution)

- Bridge-balance condition

$$\cancel{R_0} * RTD = R_1 * \cancel{R_0}$$

$$\text{i.e. } RTD = R_1 = 37.36\Omega$$

From $R_{RTD}/R_0 = 1 + \alpha(T - T_0)$, numerically

$$\frac{37.36\Omega}{25\Omega} = 1 + 0.003925(T - 0)$$

i.e.

$$T = 126^{\circ}\text{C}$$

RTD: measurements

- Wheatstone bridge
 - **low resistance** (conductors)
 - subject to **self-heating**
- lead-wire effects
 - 2-wires
 - long wires are also subject to temp-resistance changes
 - $RTD + 2 \cdot r_0 = R_1$
 - 3-wires
 - $RTD + r_0 = R_1 + r_0 \rightarrow RTD = R_1$
 - 4-wires

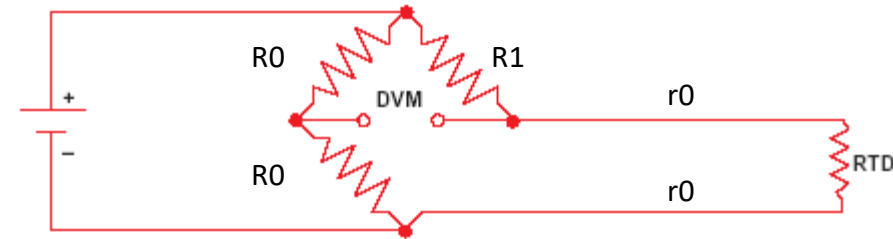
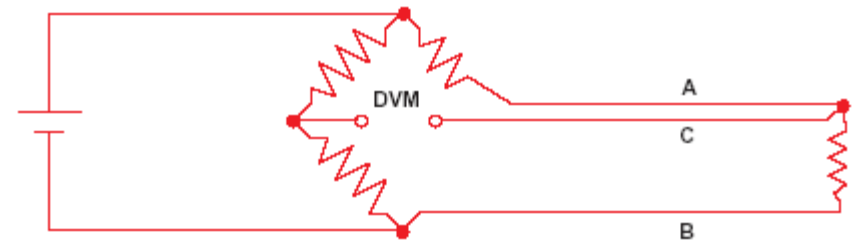
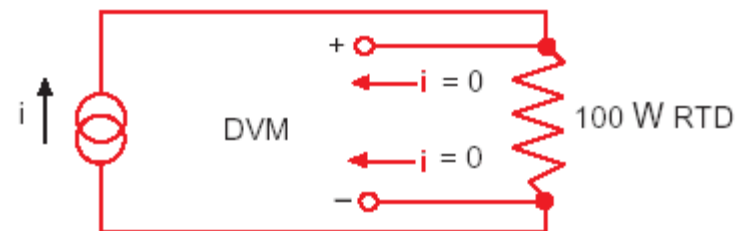


Figure 40



3-WIRE BRIDGE

Figure 41

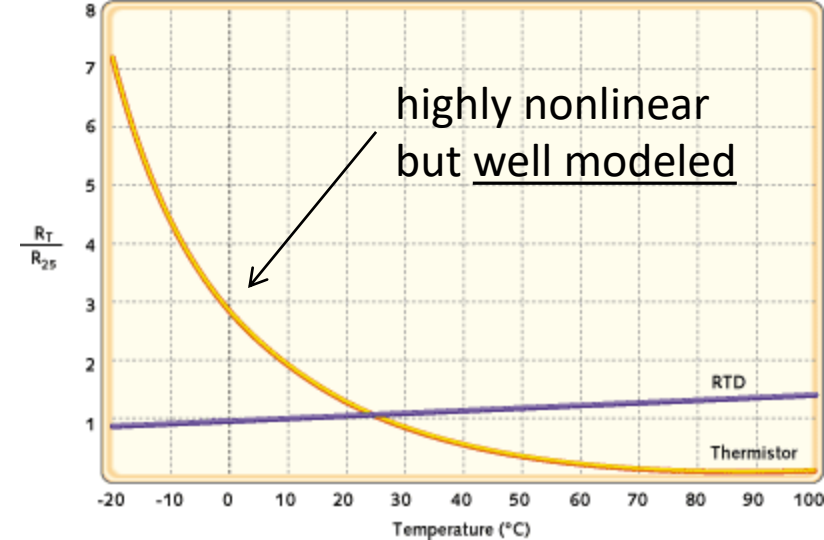


4-WIRE OHMS MEASUREMENT

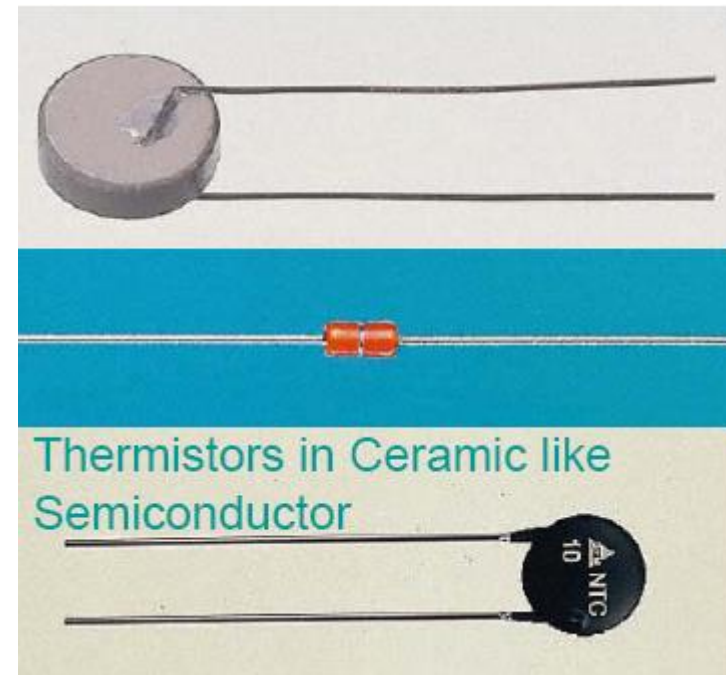
Figure 42

thermistors

- thermally sensitive resistors
 - ceramic-like **semiconductors**
 - R_0 much larger than RTD
 - resistance decreases rapidly with temperature
 - high-sensitivity
 - ruggedness
 - fast time-response



$$R = R_0 e^{\beta \left(\frac{1}{T} - \frac{1}{T_0} \right)}$$

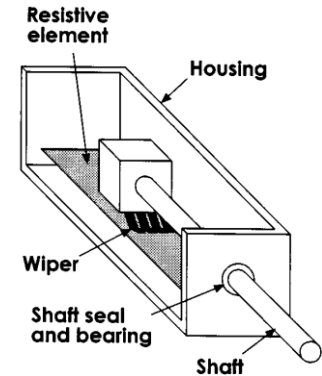


measuring displacement (linear/angular)

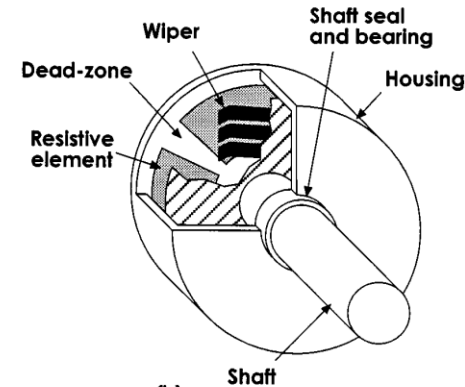
resistive sensors

resistive sensors

- potentiometer (aka 'pot')
 - **3-terminal** electromechanical device based on a conductive **wiper** sliding against a fixed, resistive element
 - many varieties (quality/function)
 - rheostats, trimmers, volume control,



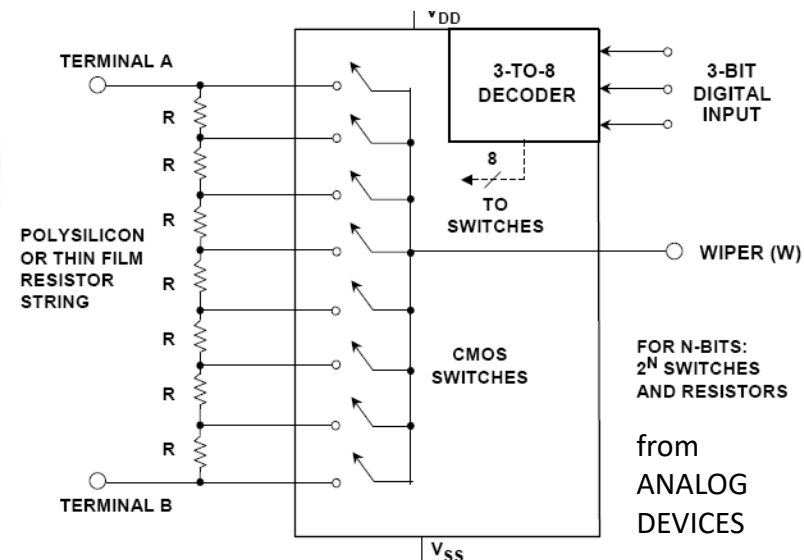
(a)



(b)

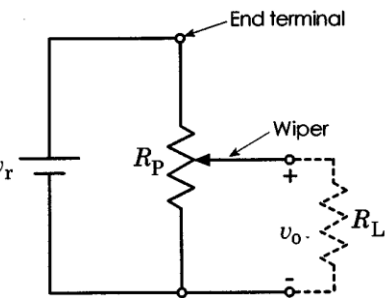


- precision potentiometers
 - manually or **digitally** tunable

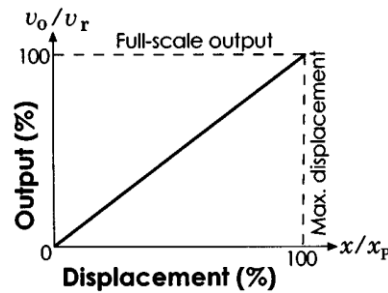


using potentiometers in electrical circuits

- voltage divider

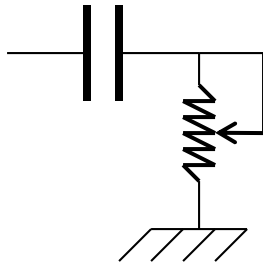


(a)

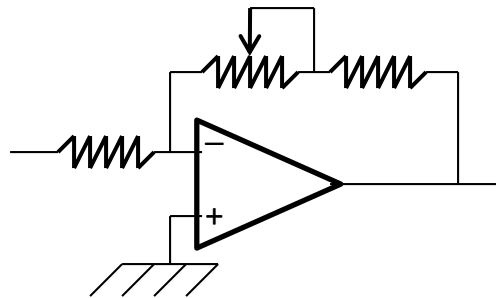


(b)

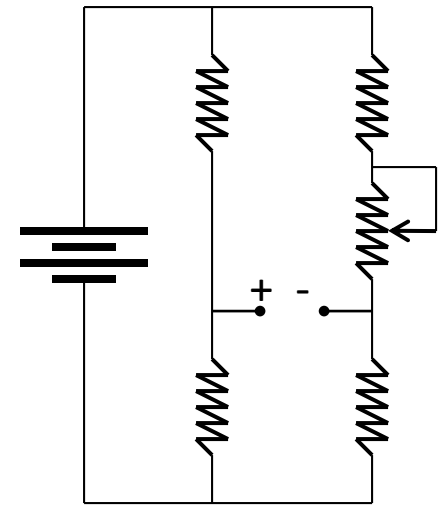
- variable resistance



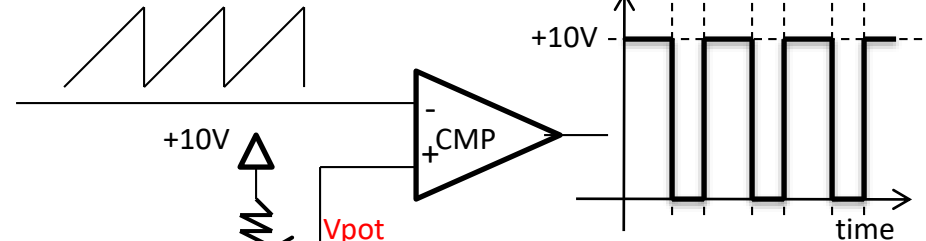
tunable filters



variable gain amplifier



Wheatstone bridge
with adjustable offset

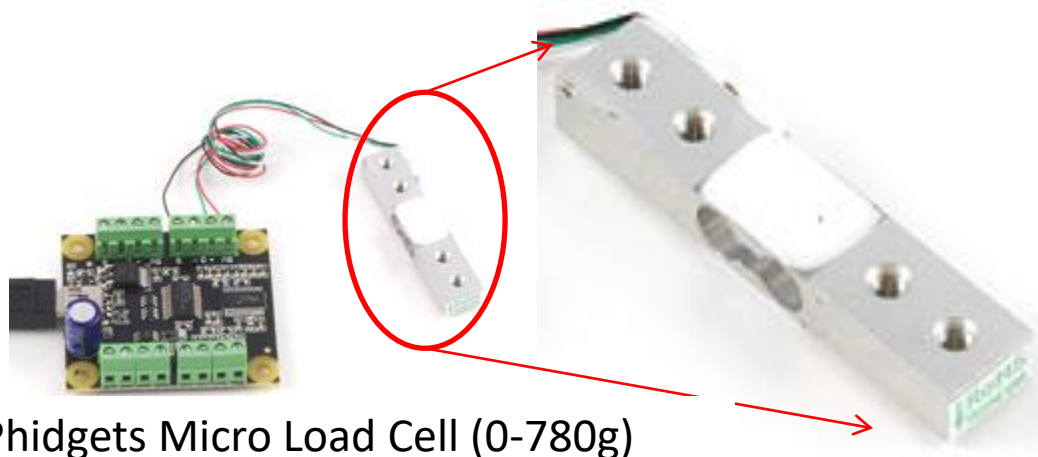
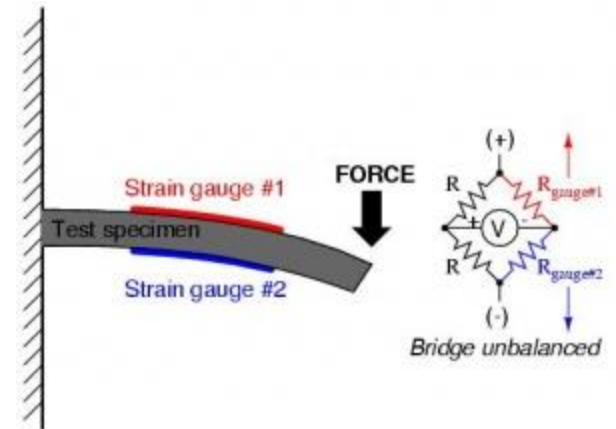
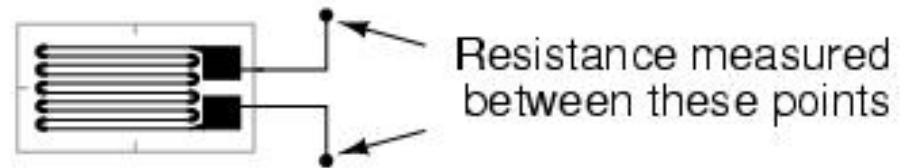


measuring forces through displacement

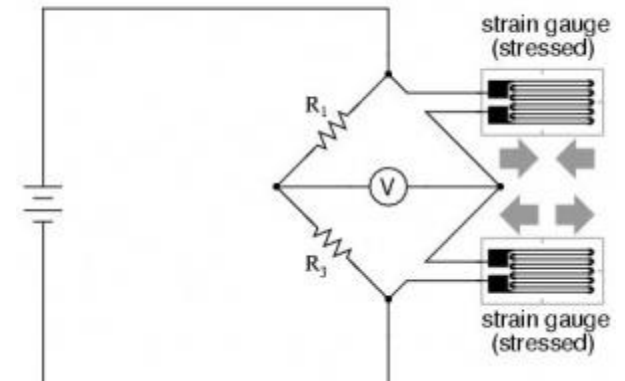
force sensors

strain gauges

Force measurements
always indirectly via
deformations

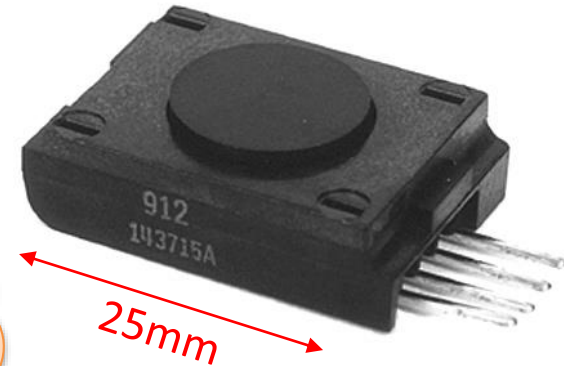


Half-bridge strain gauge circuit

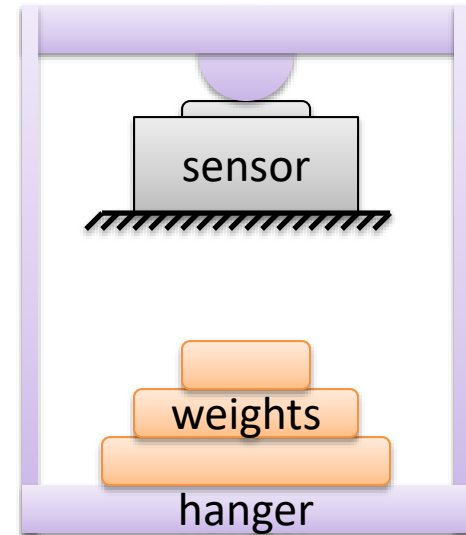
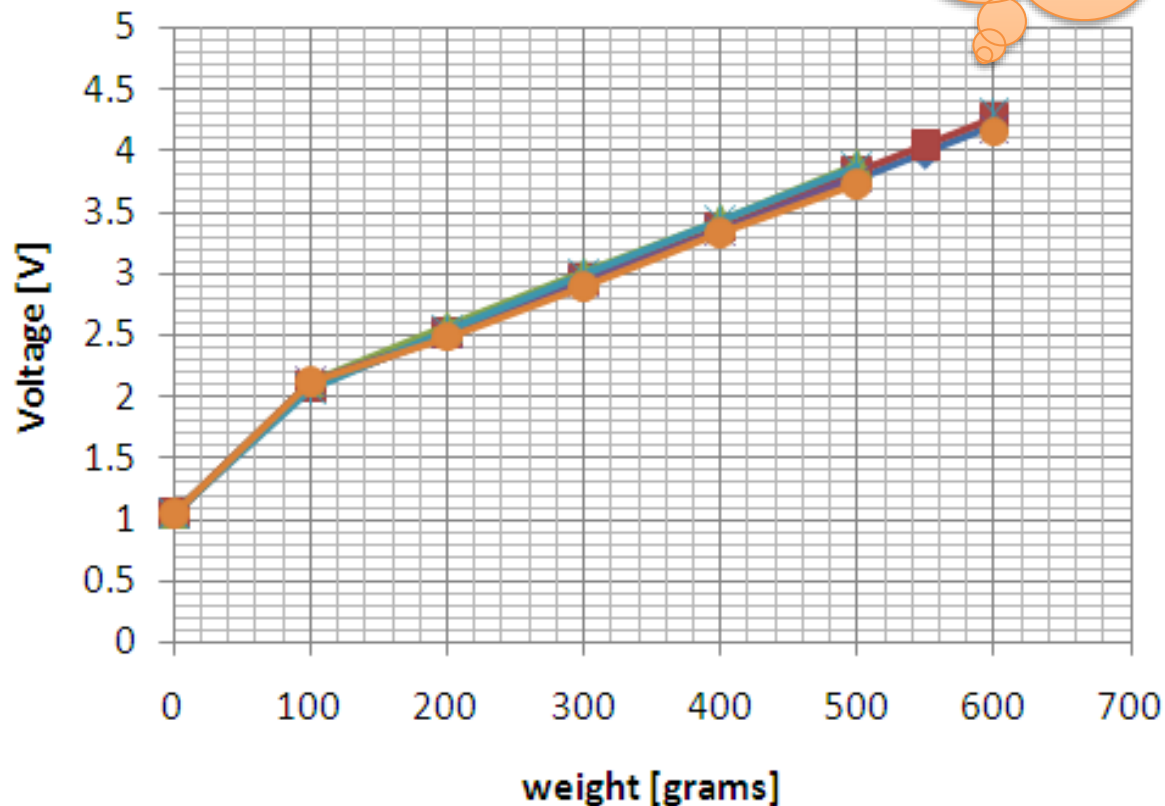


Honeywell FS01 piezoresistive force sensor

- calibration curve

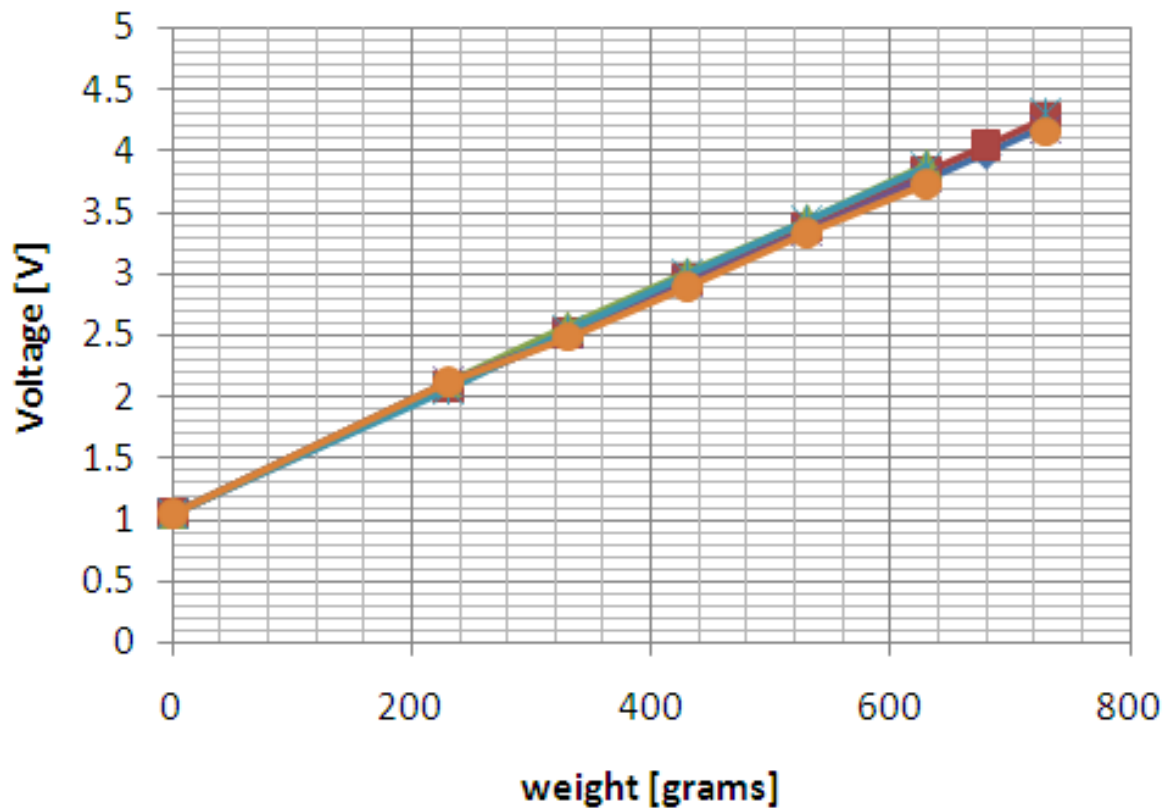
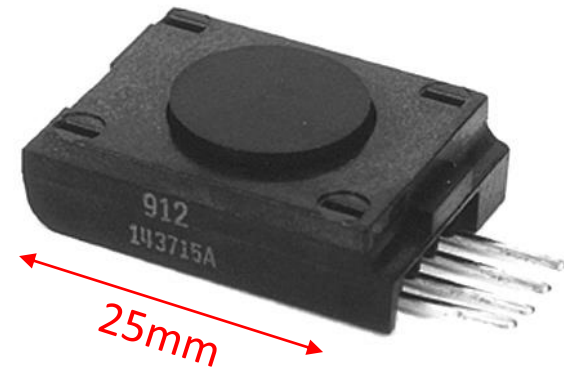


What is
wrong with
this graph?

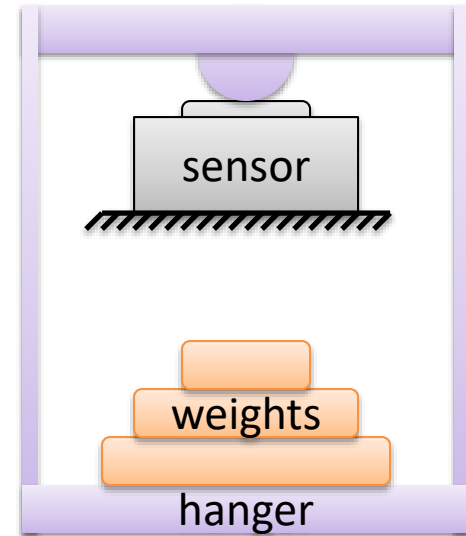


Honeywell FS01 piezoresistive force sensor

- calibration curve



- ◆— 1st
- 2nd
- ▲— 3rd
- ×— 4th
- *— 5th
- 6th



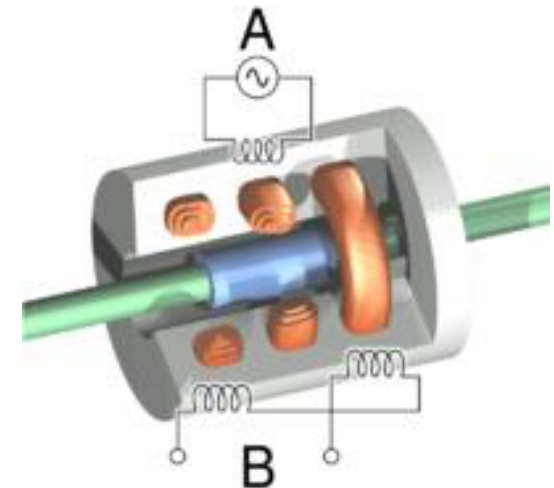
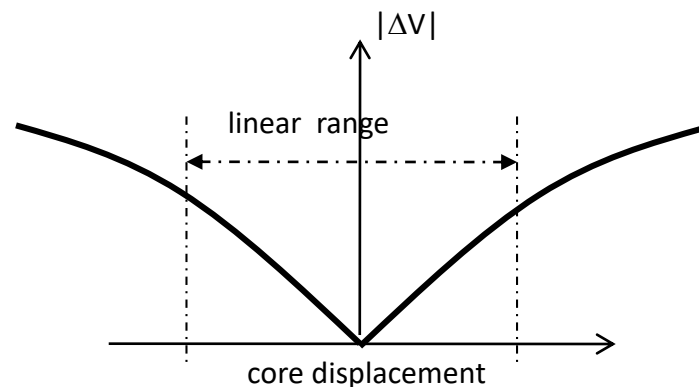
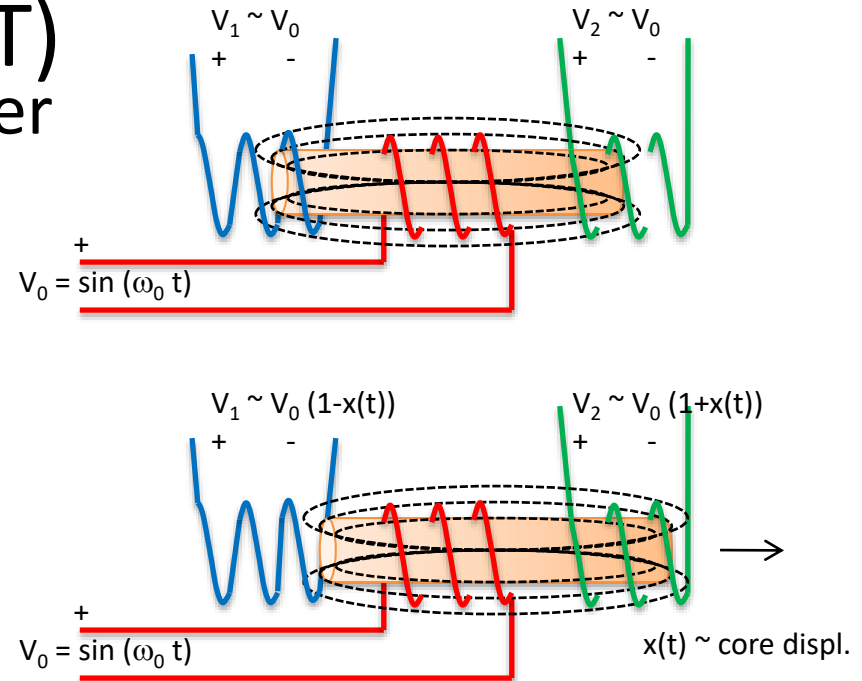
measuring displacement (linear/angular)

inductive sensors and
amplitude modulation

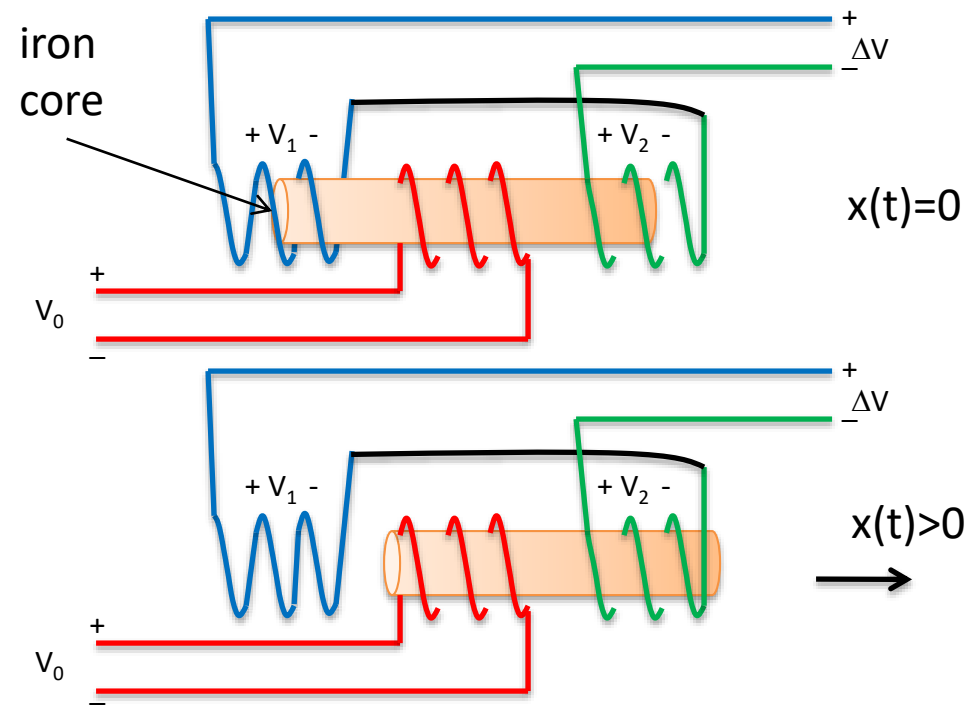
Linear Variable Differential Transformer (LVDT)

- a type of electrical transformer
 - measuring **linear** displacement
 - **variable** coupling via sliding ferromagnetic core
 - primary coil (AC driven, kHz)
 - two secondary coils
 - **differential** voltage

$$V_{OUT} = \Delta V = V_2 - V_1 \simeq x(t) V_0$$

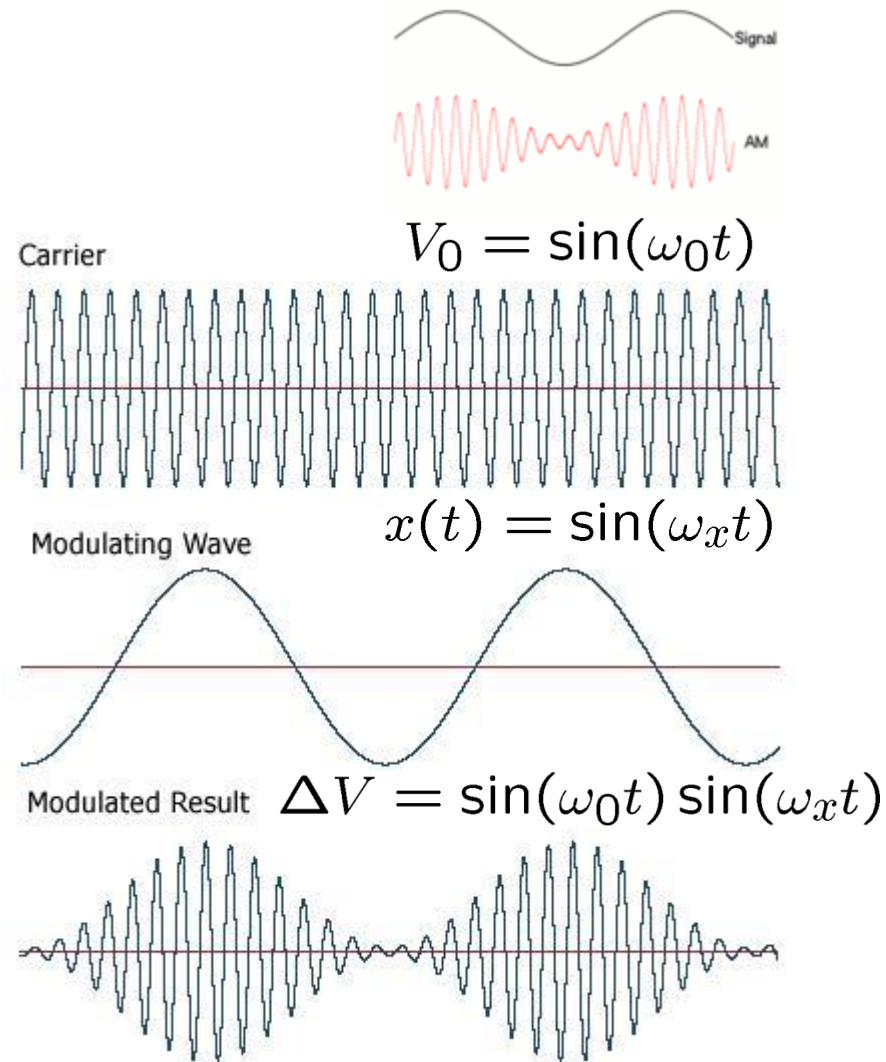


LVDT: amplitude modulation (AM)

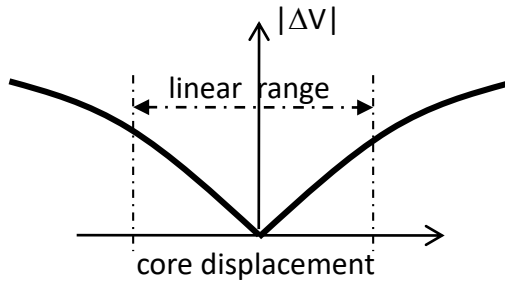
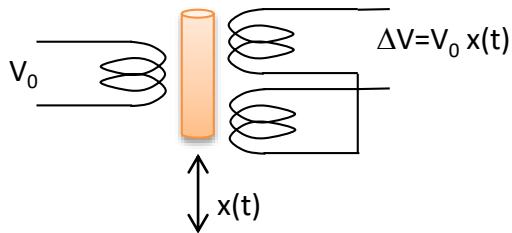


the **amplitude** of the output voltage is **modulated** by the physical displacement:

$$\Delta V \approx V_0 x(t) = \sin(\omega_0 t) \sin(\omega_x t)$$



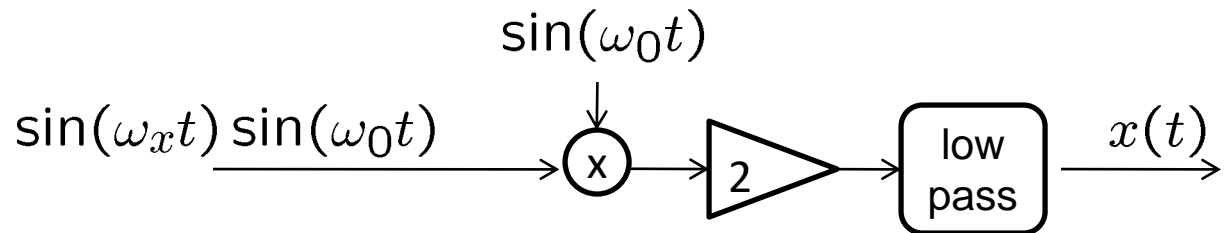
LVDT: amplitude demodulation



$$V_0 = \sin(\omega_0 t)$$

$$x(t) = \sin(\omega_x t)$$

$$\omega_x \ll \omega_0$$



$$\sin(\omega_x t) \sin^2(\omega_0 t)$$

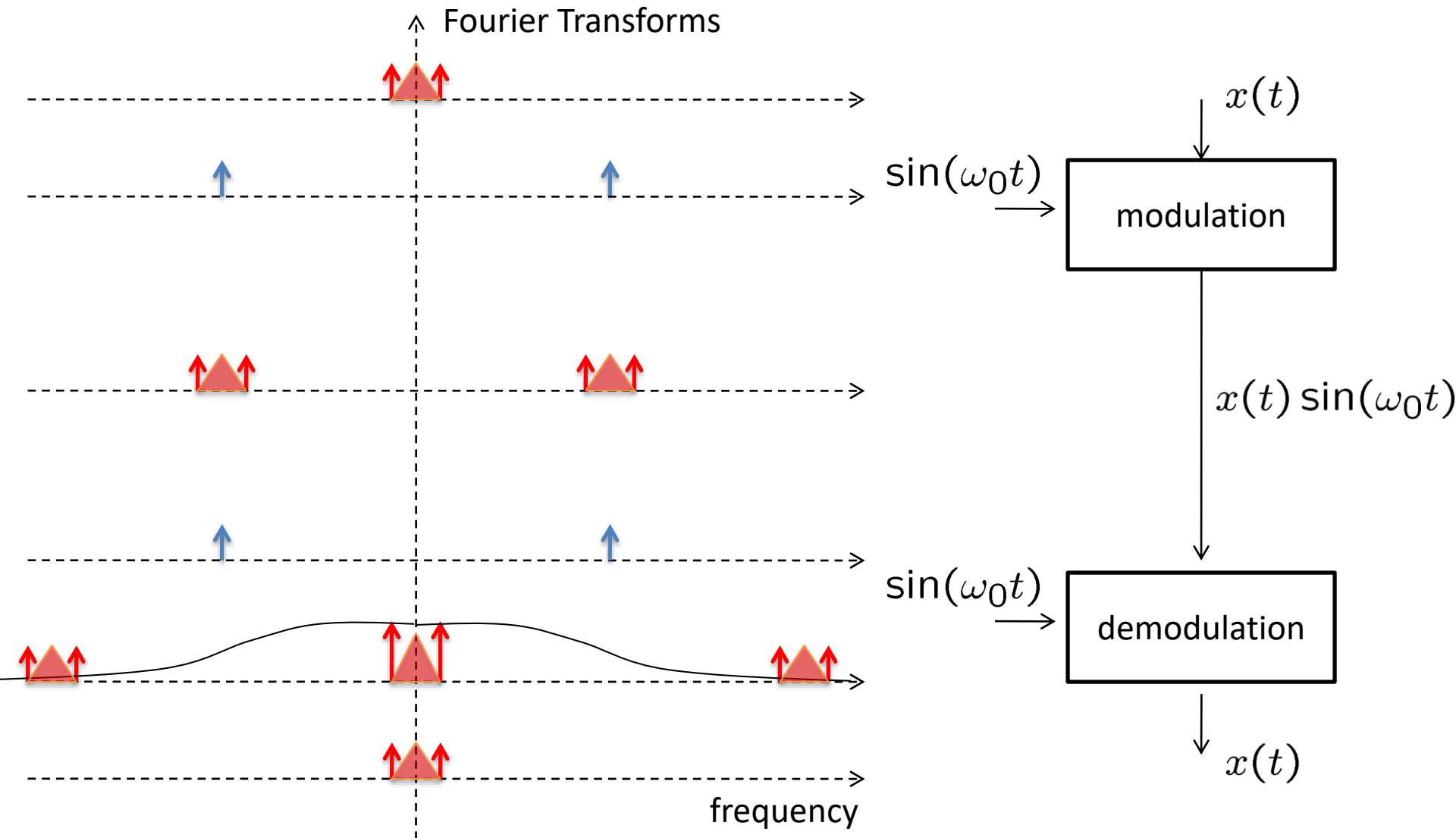
$$\sin(\omega_x t) \frac{1 - \cos(2\omega_0 t)}{2}$$

$$\frac{1}{2} \sin(\omega_x t) + \frac{\sin((2\omega_0 + \omega_x)t) - \sin((2\omega_0 - \omega_x)t)}{4}$$

low frequency

high frequency
unwanted - to be filtered-out

Amplitude Modulation/Demodulation

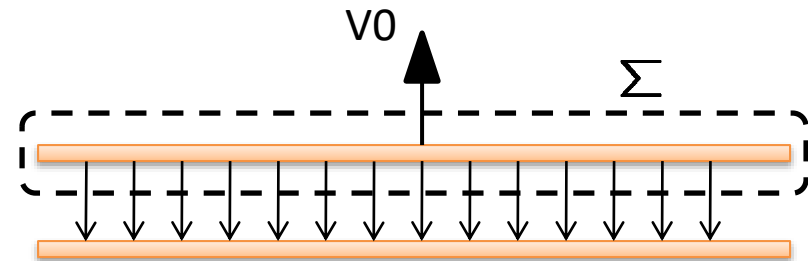
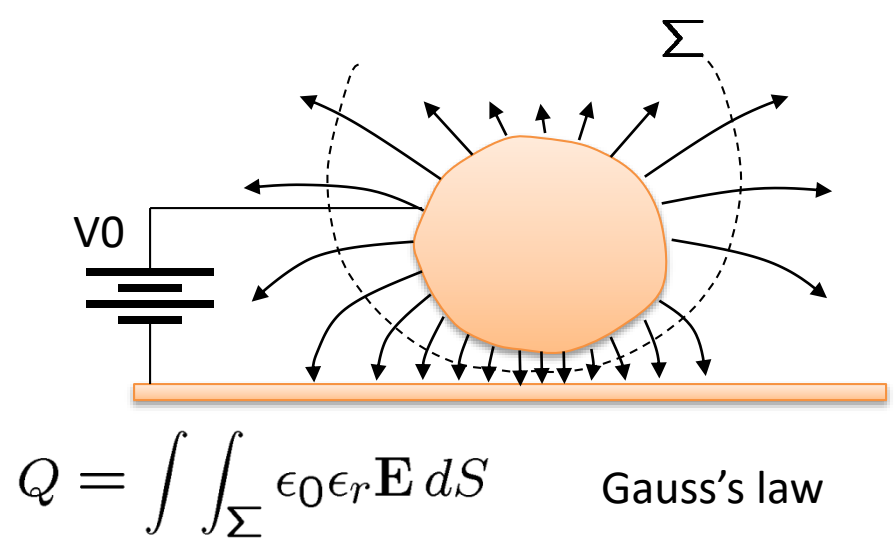


measuring displacement (linear/angular)

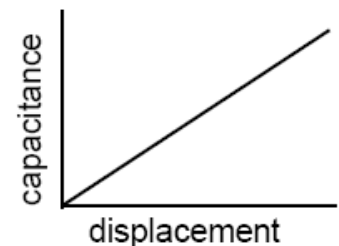
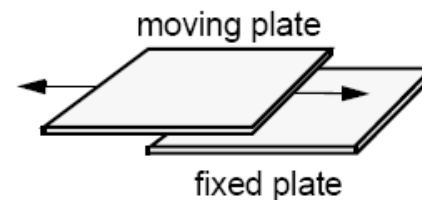
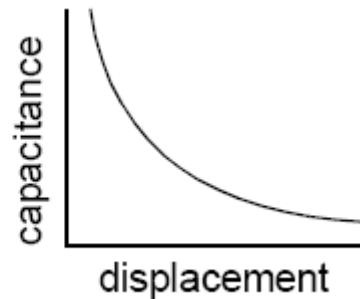
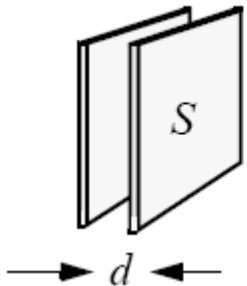
capacitive sensors
(proximity sensor)

capacitive sensing: principle

- capacitance definition
 - $C := Q / V$
- ideal case
 - infinite parallel plates
- applications
 - proximity sensing

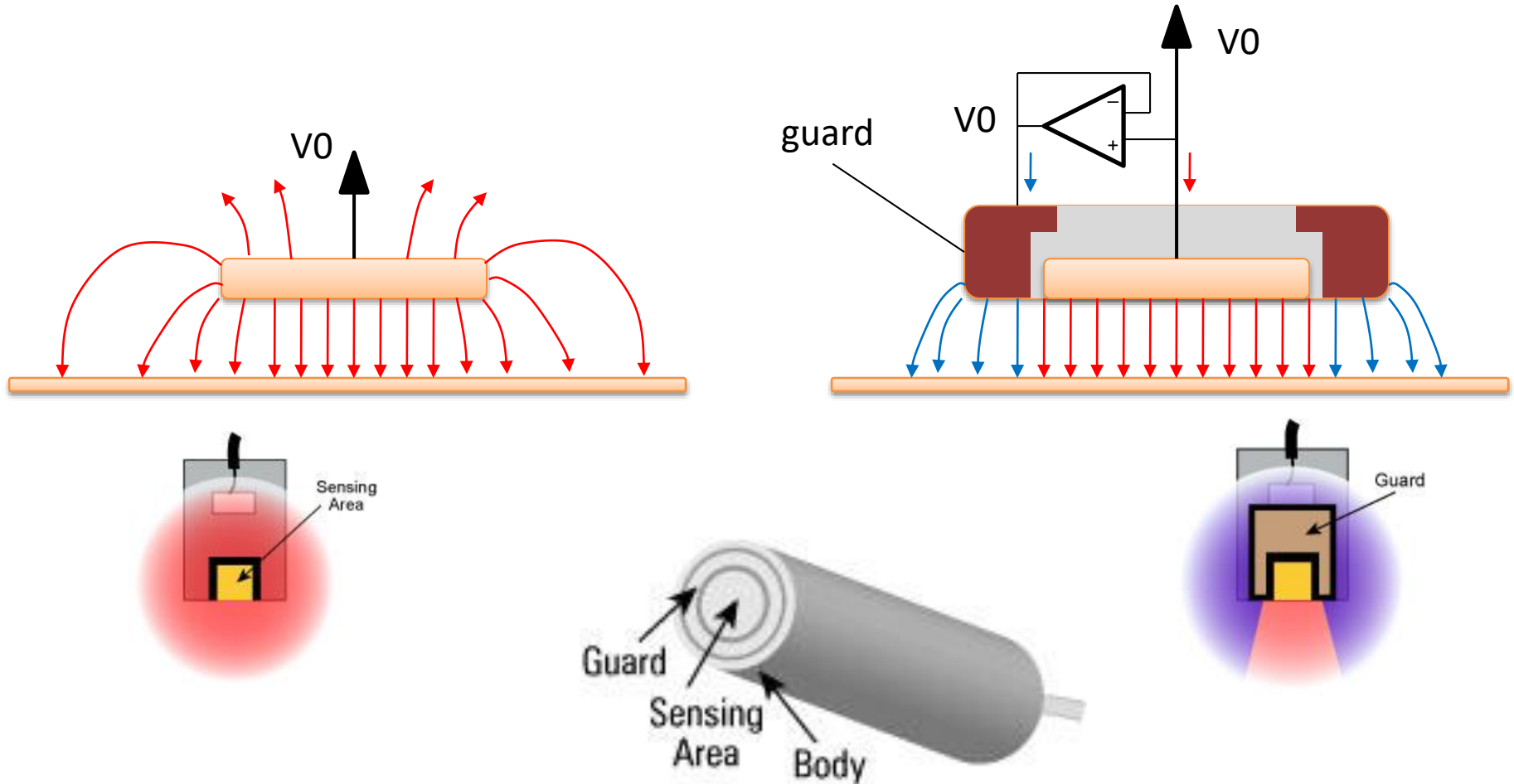


$$C = \frac{Q}{V} = \frac{\epsilon_0 \epsilon_r E S}{E d} = \frac{\epsilon_0 \epsilon_r S}{d}$$



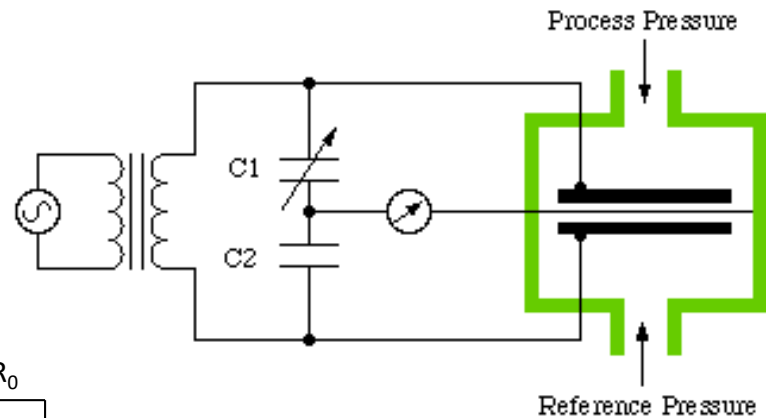
capacitive sensing: guard electrode

- the guard electrode limits field-fringing effects

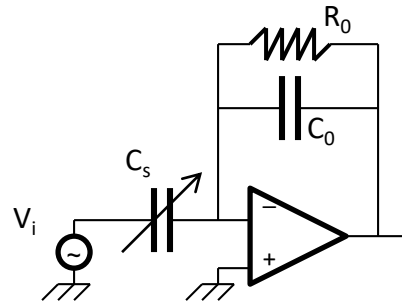


interfacing with capacitive sensors: AC

- AC bridge



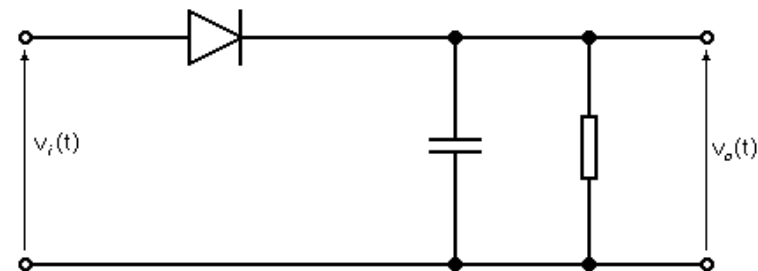
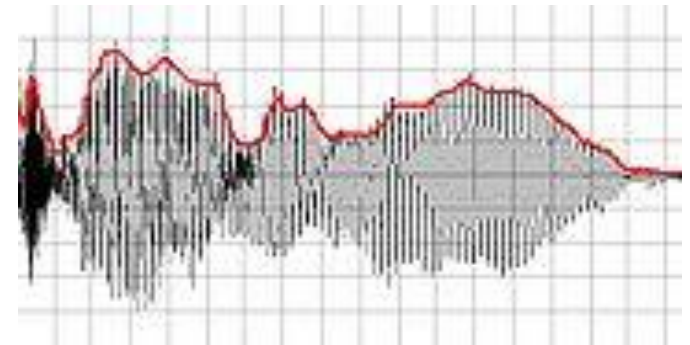
- AC driving



- Modulation

- envelope demodulator

- simplest demodulation
- for non-negative signals



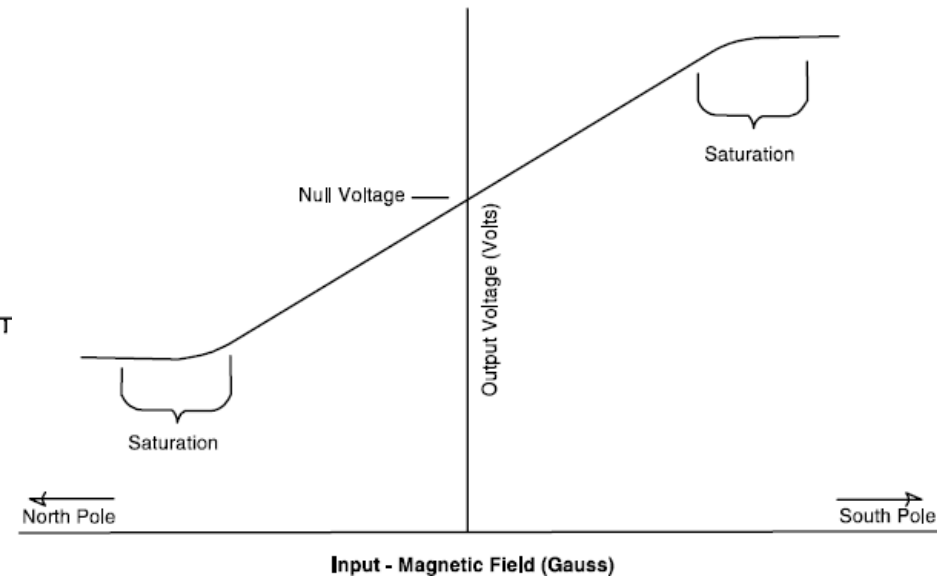
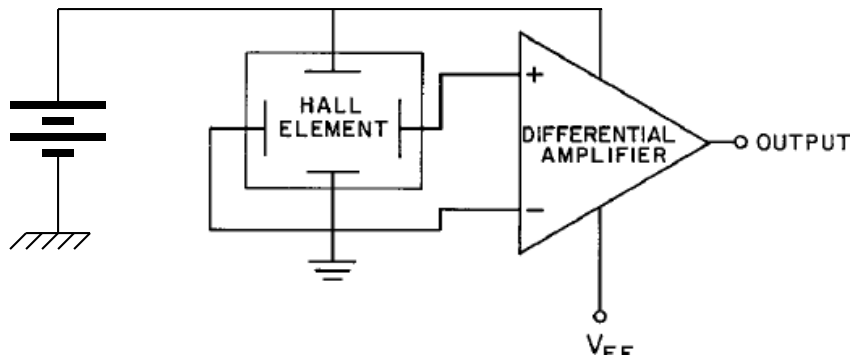
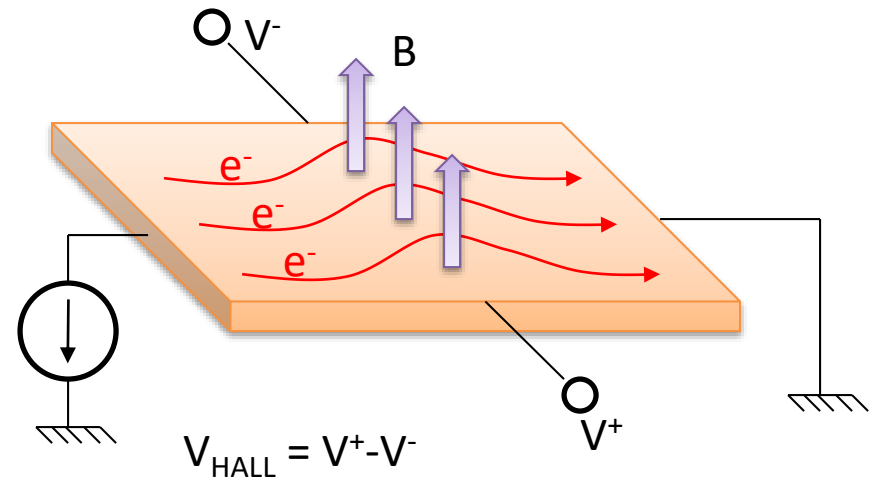
proximity sensors

magnetic and optical

Hall effect

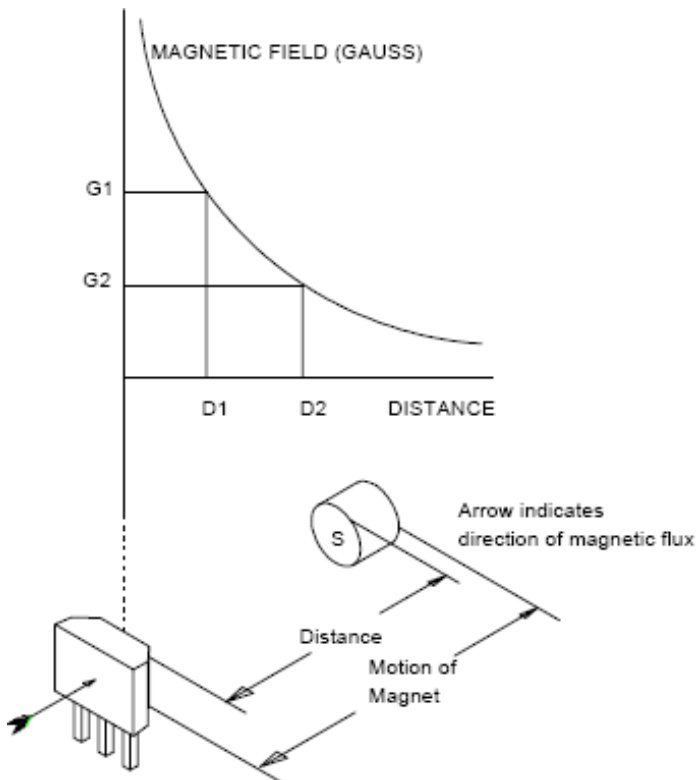
- Lorentz force

$$\vec{F} = q\vec{v} \times \vec{B}$$

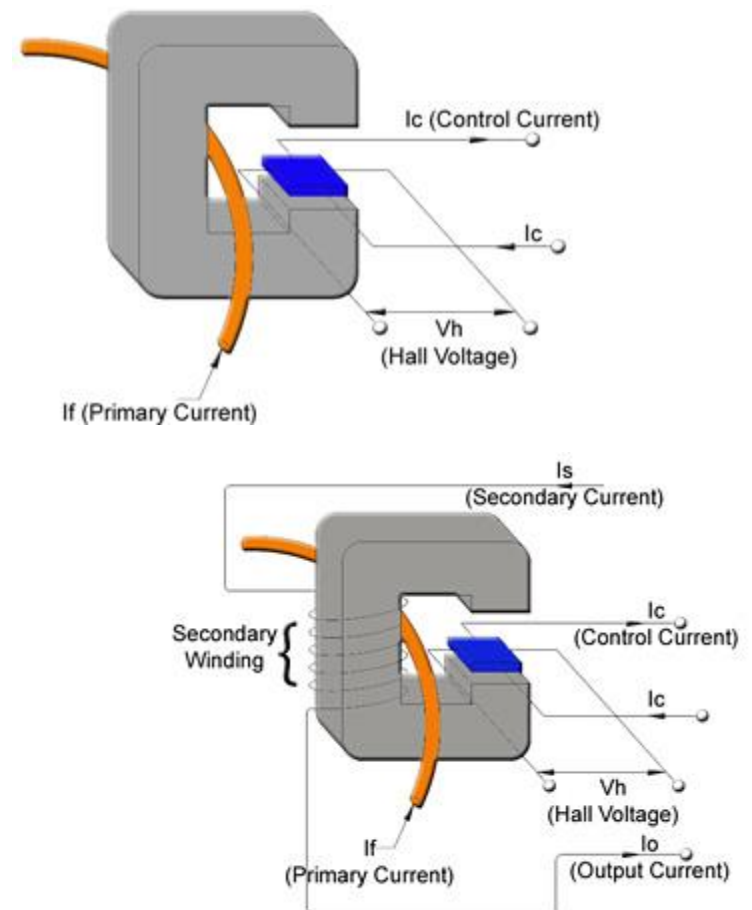


Hall effect sensors applications

- proximity sensor
 - contactless switch

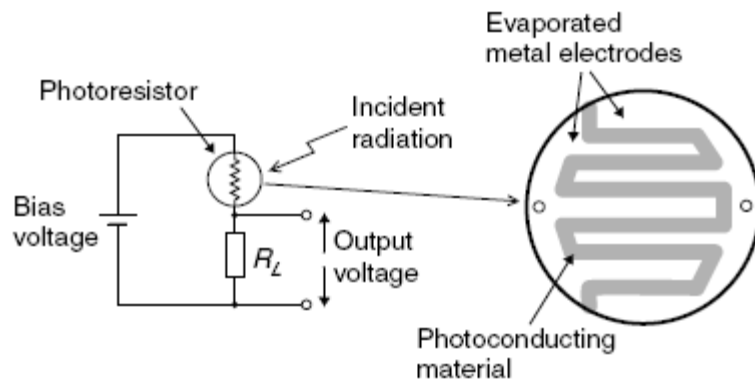


- current sensor

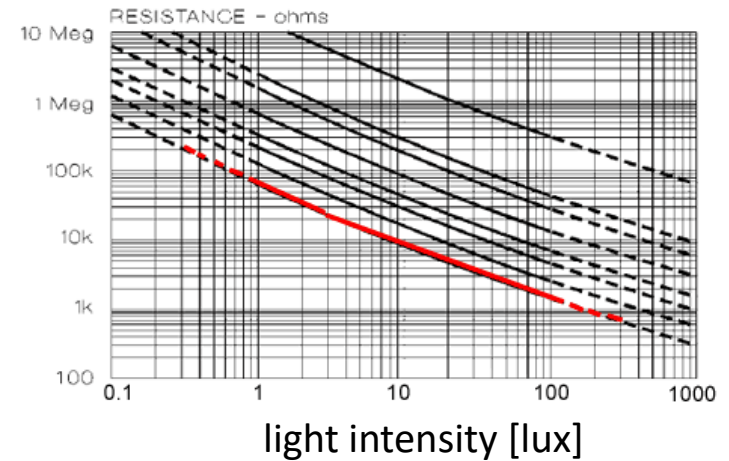


light detectors

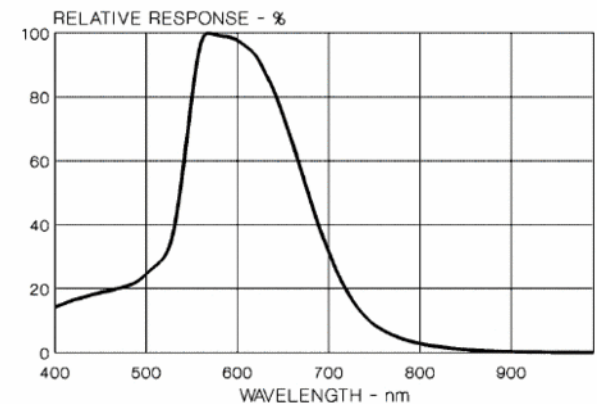
- photo-resistors



Resistance vs. Illumination

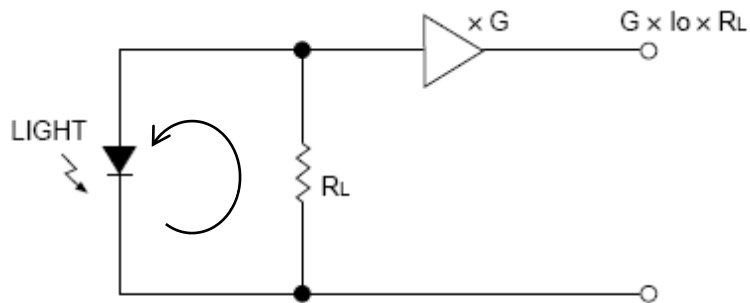


Relative Spectral Response



light detectors

- photo-diodes
 - load resistance



– op-amp circuit

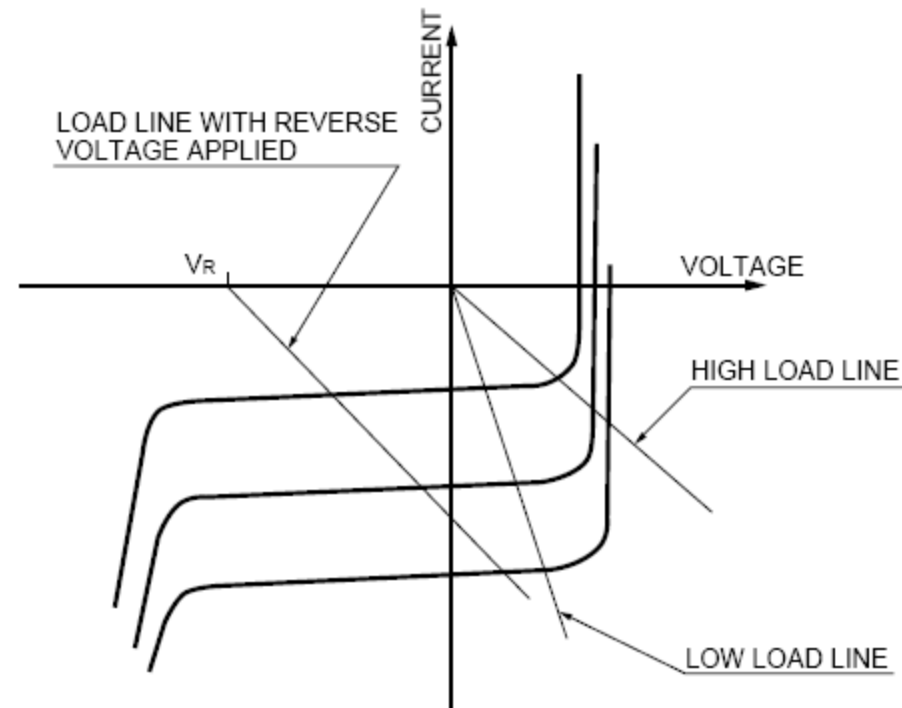
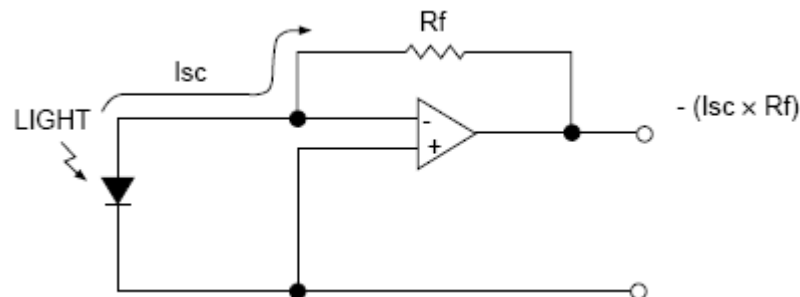
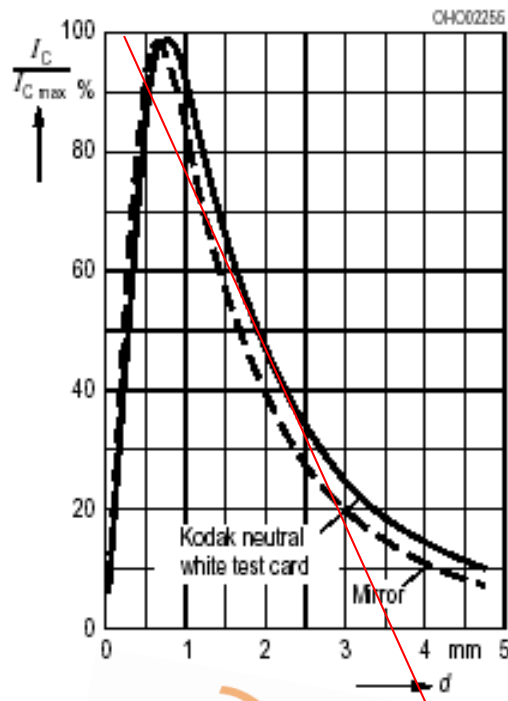
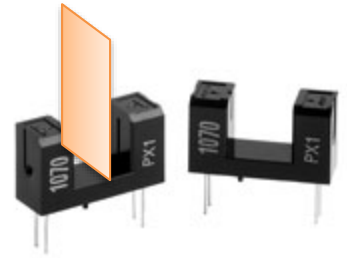
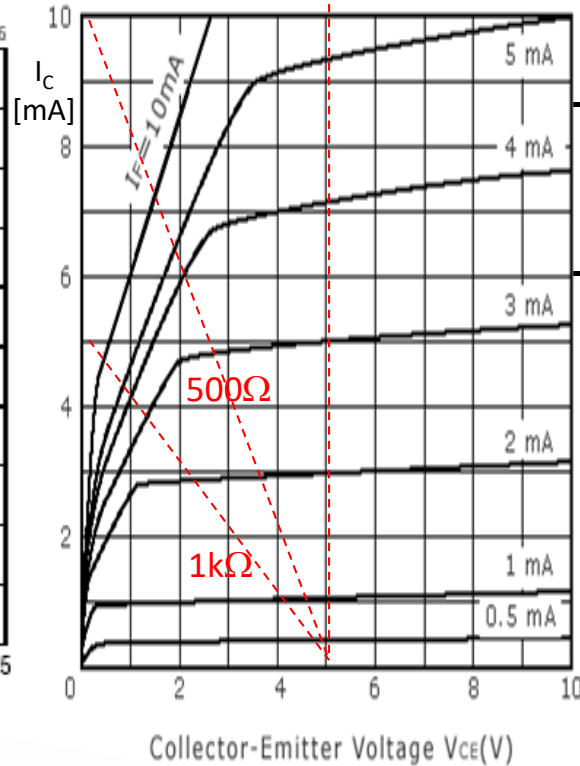


photo-transistors

Collector Current $\frac{I_C}{I_{Cmax}} = f(d)$



Example of Collector Current V.S. Collect-Emittter Voltage

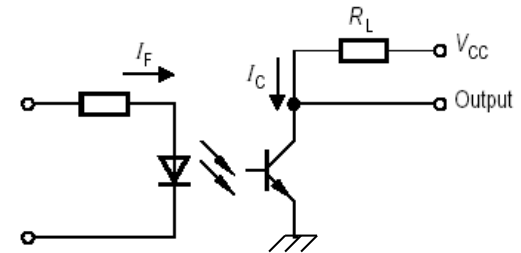


transmissive

- photo-interrupter

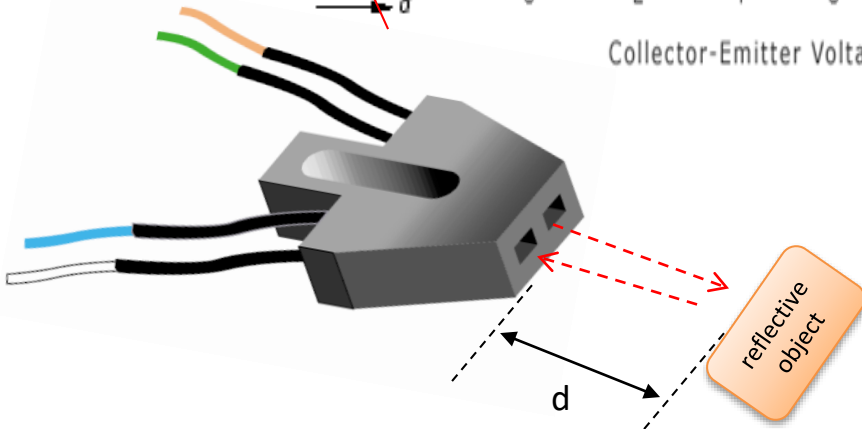
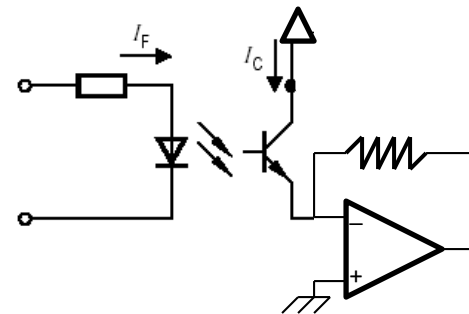
reflective

- resistive load



$V_{CC} = 5\text{ V}$
 $R_L = 1k\Omega$

- current-voltage
op-amp



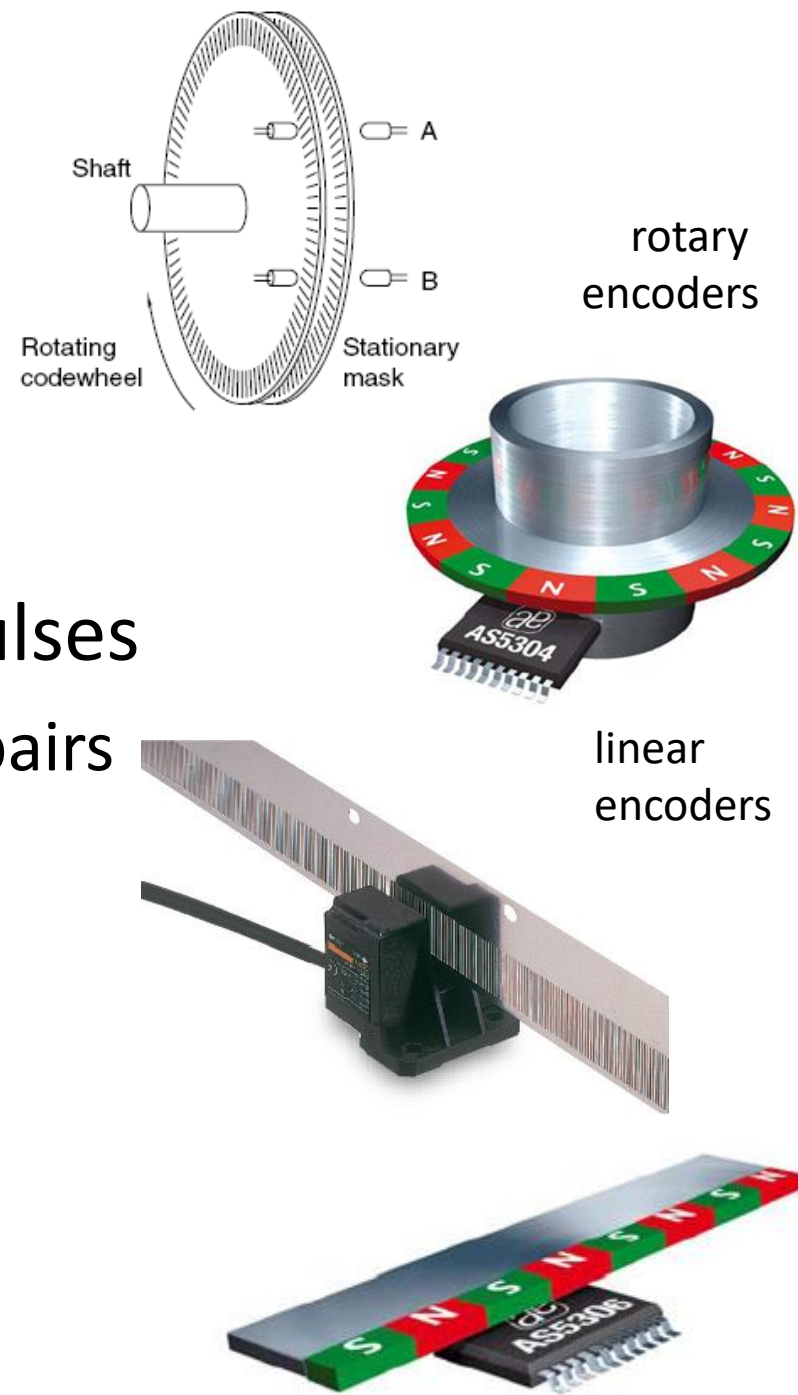
digital encoders

linear / angular

absolute / incremental

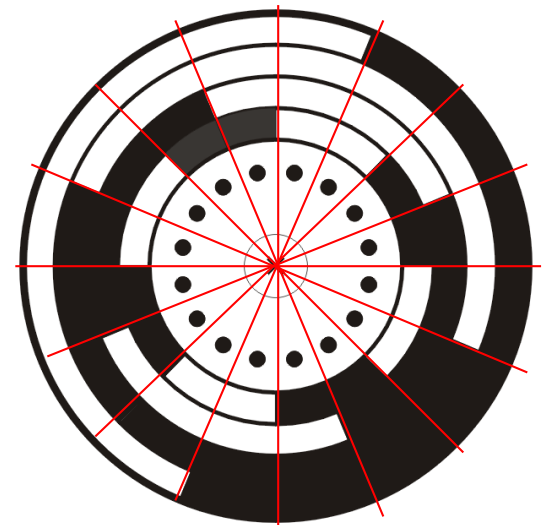
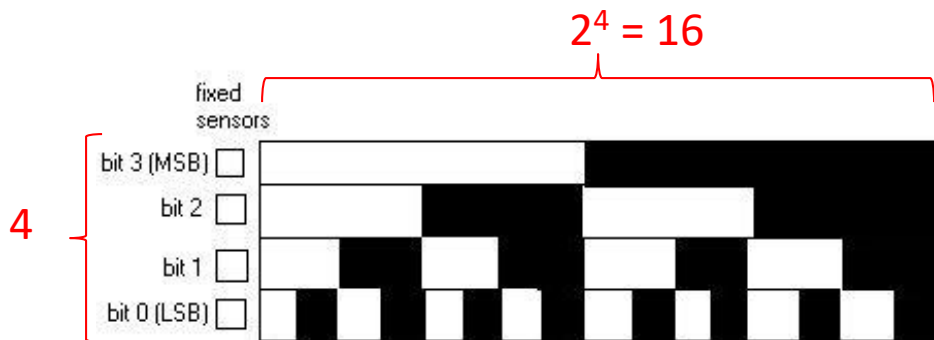
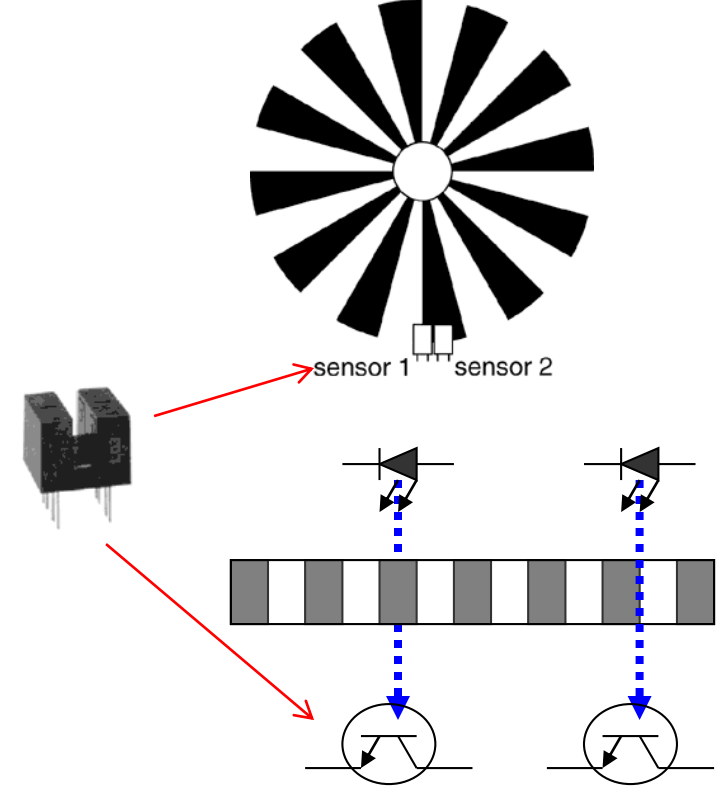
digital encoders

- convert motion
 - either linear or **rotary**
- into a sequence of digital pulses
 - optical transmitter/receiver pairs
 - glass/plastic material
photographically patterned
 - Hall effect sensors
 - coupled with magnetic
rings / bars



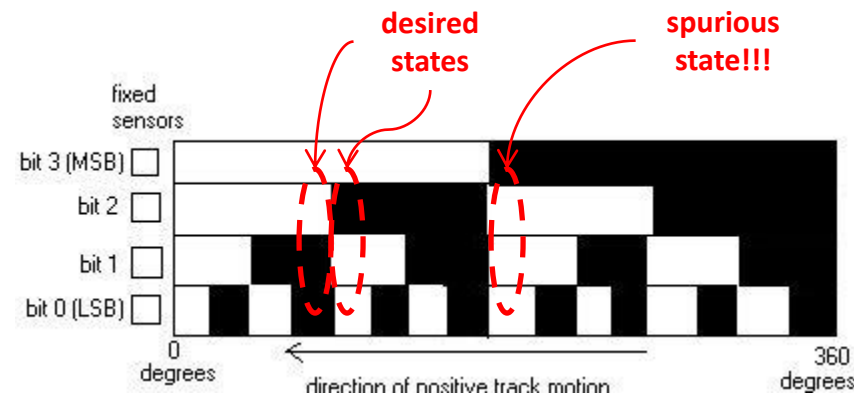
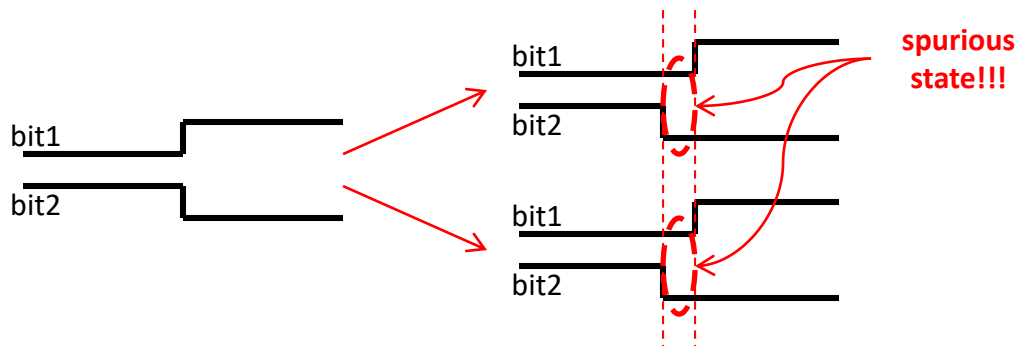
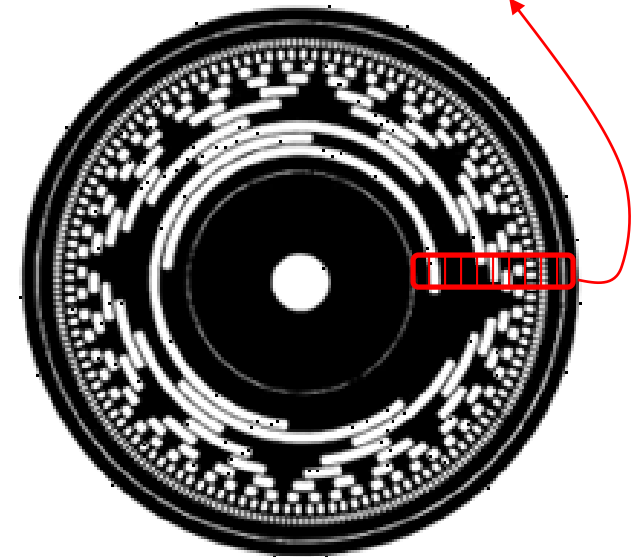
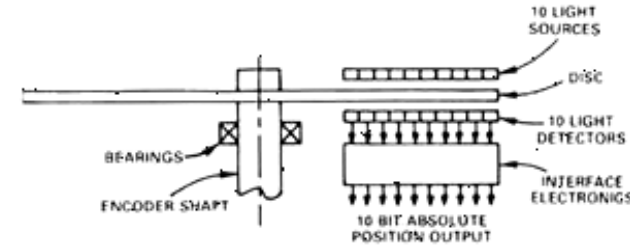
digital encoders

- incremental encoders
 - minimum 2 Tx/Rx pairs
 - encoding steps and direction
- absolute encoders
 - n Tx/Rx pairs for coding 2^n sectors



absolute encoders

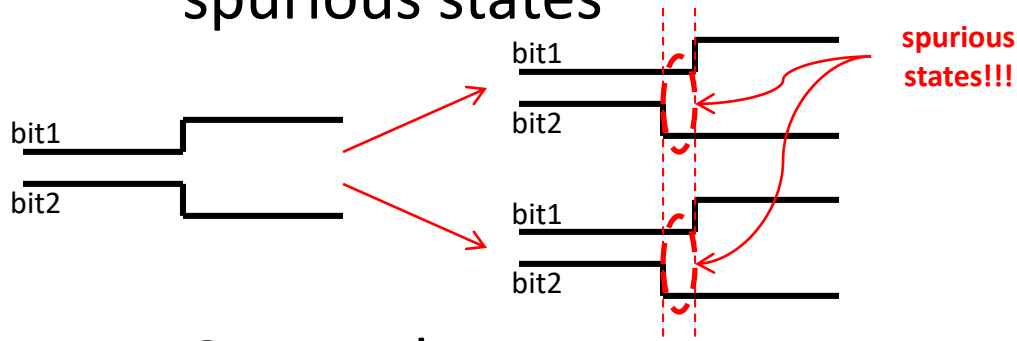
- angular n-bits encoders
 - $360^\circ / 2^n$ resolution
 - $n=10 \rightarrow 360^\circ / 1024 = 0.35^\circ$
 - more expensive
 - require n Tx/Rx pairs
 - **CAVEAT**: spurious states may arise from contemporary transitions



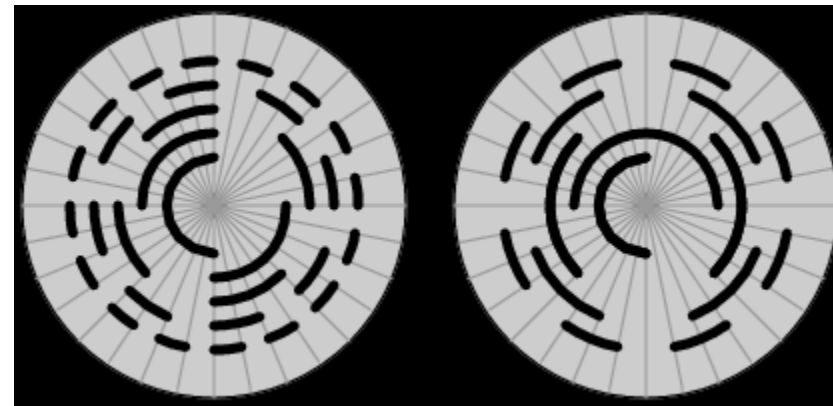
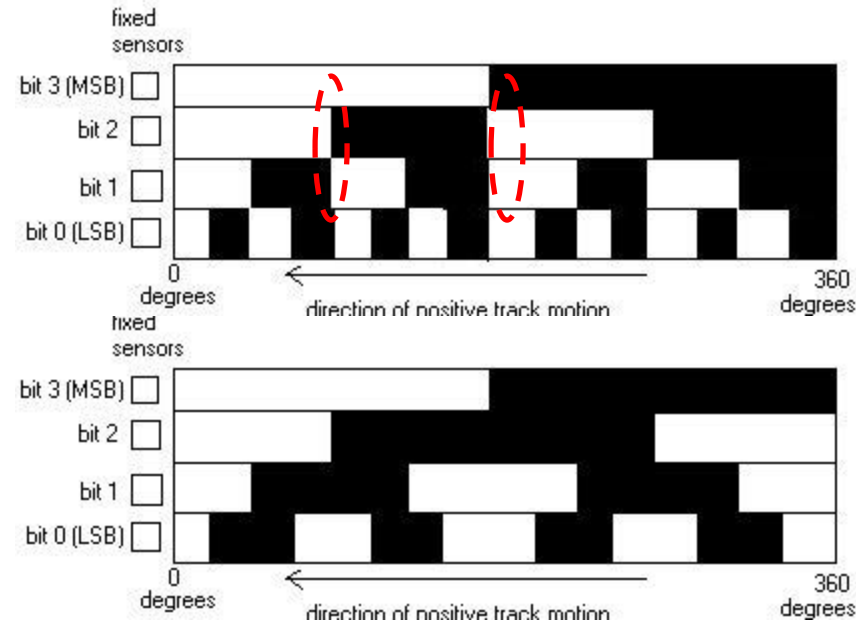
absolute encoders

- natural binary code vs. **Gray code**

- contemporary transitions might lead to temporary spurious states

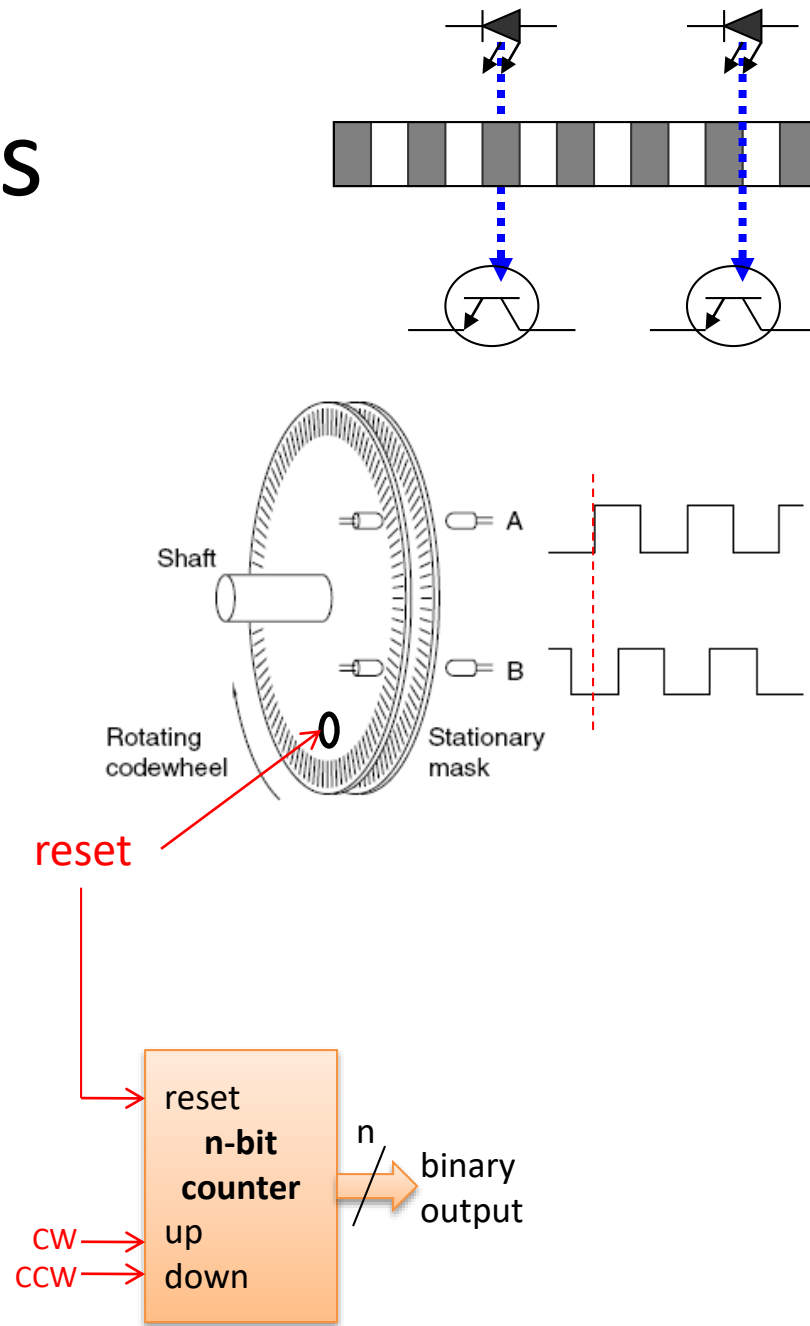
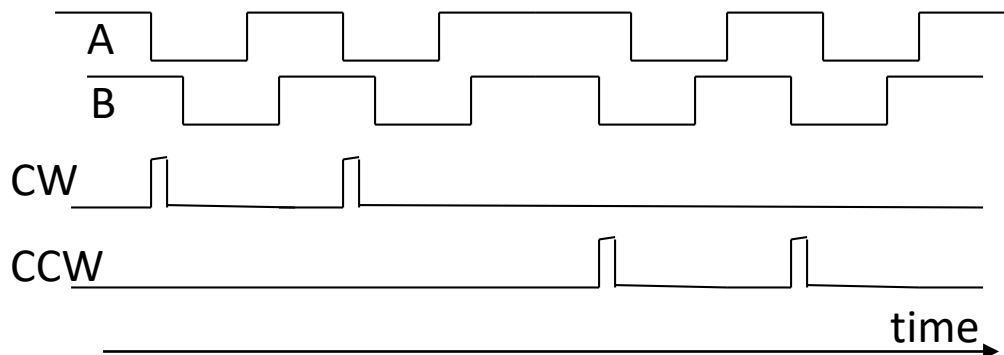


- Gray code ensures no contemporary transitions



incremental encoders

- simpler design
 - a single pair not sufficient to encode the direction
 - 2 Tx/Rx pairs plus a 'reset' position are required
- quadrature signals
 - $\frac{1}{4}$ cycle out-of-phase



measuring higher kinematics

please, refer to this case study

D. Campolo, S. E. Maini, F. Patanè, C. Laschi, P. Dario, F. Keller, E. Guglielmelli, Design of a Sensorized Ball for Ecological Behavioral Analysis of Infants, IEEE Intl. Conf. on Robotics and Automation (ICRA), Rome, Italy, pp. 1529 - 1534, April 10-14, 2007

http://www3.ntu.edu.sg/home/d.campolo/pdf/2007_ICRA_BALL.pdf

