Lecture 6
Frequency Domain Filtering (2)

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Outline

- Discrete Convolution Theorem (wraparound problem)
- Frequency domain filtering fundamentals
- Frequency domain filtering procedure
- Typical Lowpass filtering
- Typical Highpass filtering
- Other filtering (Homomorphic filtering, Bandreject/Bandpass filtering)

Discrete Convolution Theorem

Convolution theorem

$$f(x,y) \star h(x,y) \Leftrightarrow F(u,v)H(u,v)$$

$$f(x,y)\,h(x,y) \Leftrightarrow \frac{1}{MN}F(u,v) \star H(u,v)$$

Basis of Frequency Domain Filtering!

So, What we can do with frequency domain filtering?

1D case:

线性卷积

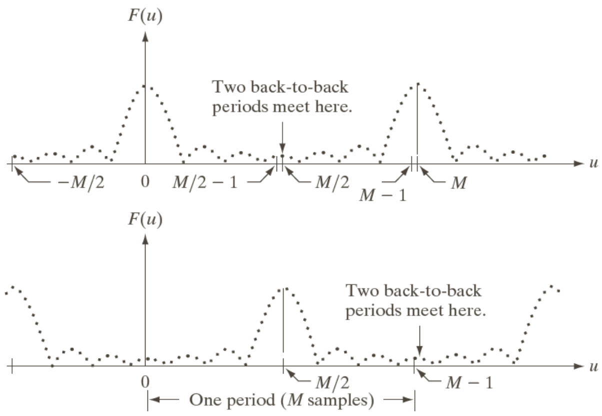
$$h[n] \otimes x[n]$$
$$= \mathcal{IDFT}\{\mathcal{DFT}\{h[n]\} \cdot \mathcal{DFT}\{x[n]\}\}$$

Is that correct?

Periodicity

- $f(x,y) = f(x+k_1M,y) = f(x,y+k_2N) = f(x+k_1M,y+k_2N)$
- $F(u,v) = F(u+k_1M,v) = F(u,v+k_2N) = F(u+k_1M,v+k_2N)$

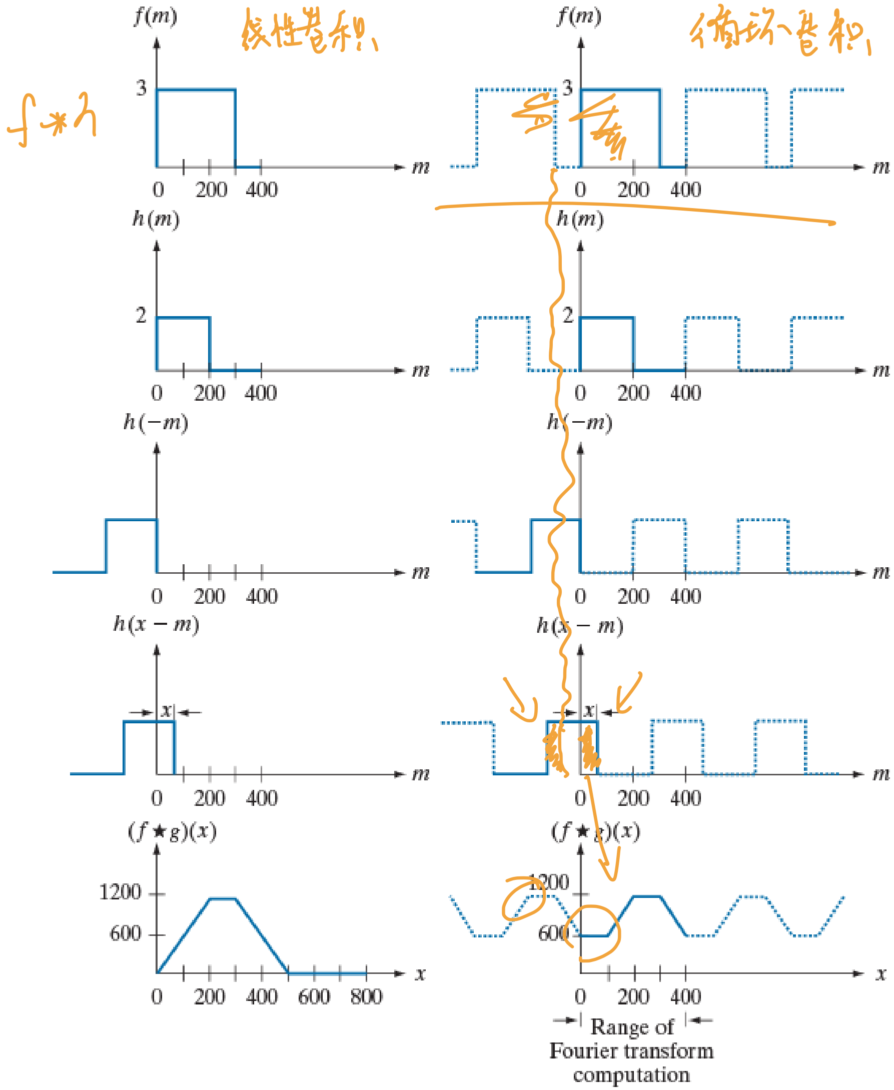
Where k_1 and k_2 are integers



$$f(x,y) \star h(x,y) \Leftrightarrow F(u,v)H(u,v)$$

What is the truly inverse Fourier Transform of *F(u,v)*?

Wraparound problem



a f
b g
c h
d i
e j

FIGURE 4.27 Left column: Spatial convolution computed with Eq. (3-44), using the approach discussed in Section 3.4. Right column: Circular convolution. The solid line in (j) is the result we would obtain using the DFT, or, equivalently, Eq. (4-48). This erroneous result can be remedied by using zero padding.

$f[n] \otimes h[n]$ $\quad$ $f[n]\,Ⓝ\,h[n]$

One important DFT property

- Circular Convolution: Let $x_1[n]$ and $x_2[n]$ be length N with DFT $X_1[k]$ and $X_2[k]$

$$x_1[n] \circledN x_2[n] \leftrightarrow X_1[k] \cdot X_2[k]$$

  - Very useful!!! (for linear convolutions with DFT)

- Multiplication (Modulation): Let $x_1[n]$ and $x_2[n]$ be length N with DFT $X_1[k]$ and $X_2[k]$

卷线错误
⇒ 周期使不该
出现能量出现了

$$x_1[n] \cdot x_2[n] \leftrightarrow \frac{1}{N} X_1[k] \circledN X_2[k]$$

Zero padding

- Zero padding is necessary for applying the convolution theorem

  - **Zero padding**

$$f_p(x,y)=\begin{cases} f(x,y), & 0\le x\le A-1, 0\le y\le B-1 \\ 0, & A\le x\le P, B\le y\le Q \end{cases}$$

$$h_p(x,y)=\begin{cases} h(x,y), & 0\le x\le C-1, 0\le y\le D-1 \\ 0, & C\le x\le P, D\le y\le Q \end{cases}$$

Where $f(x,y)$**:** $A\times B$ image**;** $h(x,y)$**:** $C\times D$ image; $P\ge A+C-1$; $Q\ge B+D-1$

Linear Convolution using DFT

- In practice we can implement a circular convolution using the DFT property:

$$h[n] \circledast x[n] = x_{zp}[n] \, Ⓝ \, h_{zp}[n]$$

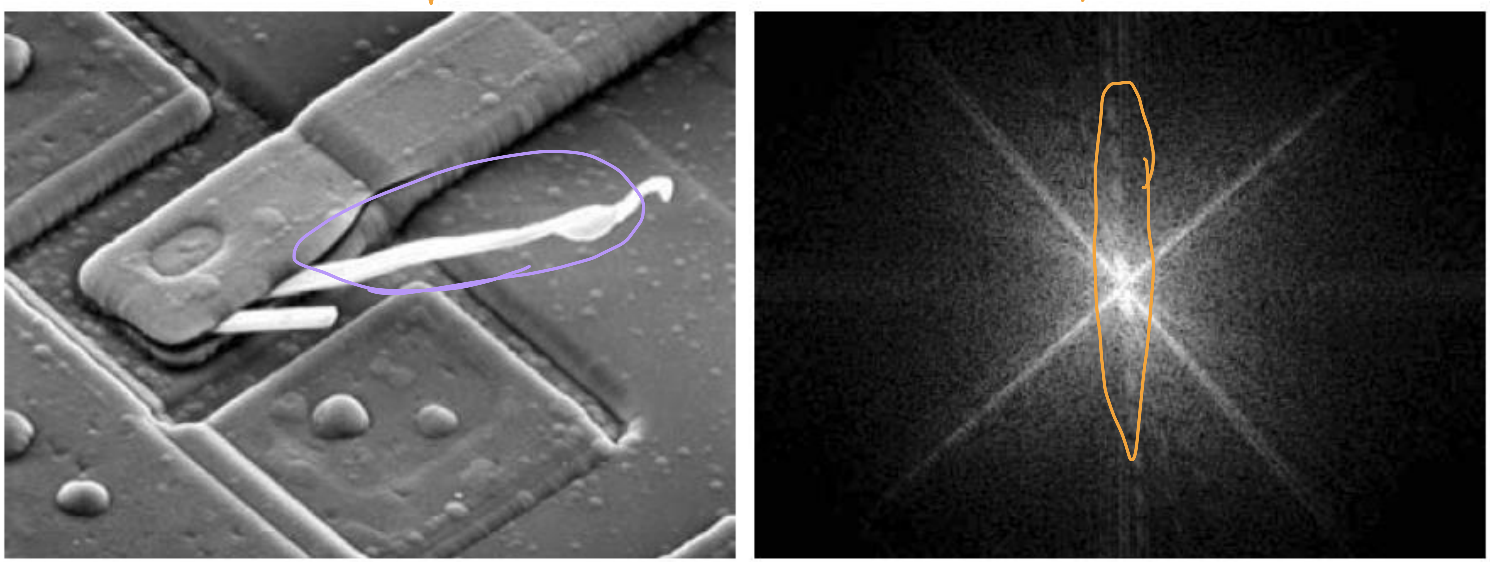
$$= \mathcal{IDFT}\{\mathcal{DFT}\{x_{zp}[n]\} \cdot \mathcal{DFT}\{h_{zp}[n]\}\}$$

补零

- Advantage: DFT can be computed with $N\log_2 N$ complexity (FFT algorithm!)
- Drawback: Must wait for all the samples -- huge delay -- incompatible with real-time implementation (but **not a problem for an image!**)

Frequency domain filtering fundamentals

- How shall we read the Fourier Spectrum?
  - 45 degree? 45°方向上边缘高频信息
  - White Components?



a b

FIGURE 4.29 (a) SEM image of a damaged integrated circuit. (b) Fourier spectrum of (a). (Original image courtesy of Dr. J. M. Hudak, Brockhouse Institute for Materials Research, McMaster University, Hamilton, Ontario, Canada.)

Simplest frequency domain filtering

- Set the DC component to 0

仅反应平均 intensity
整体亮度情况

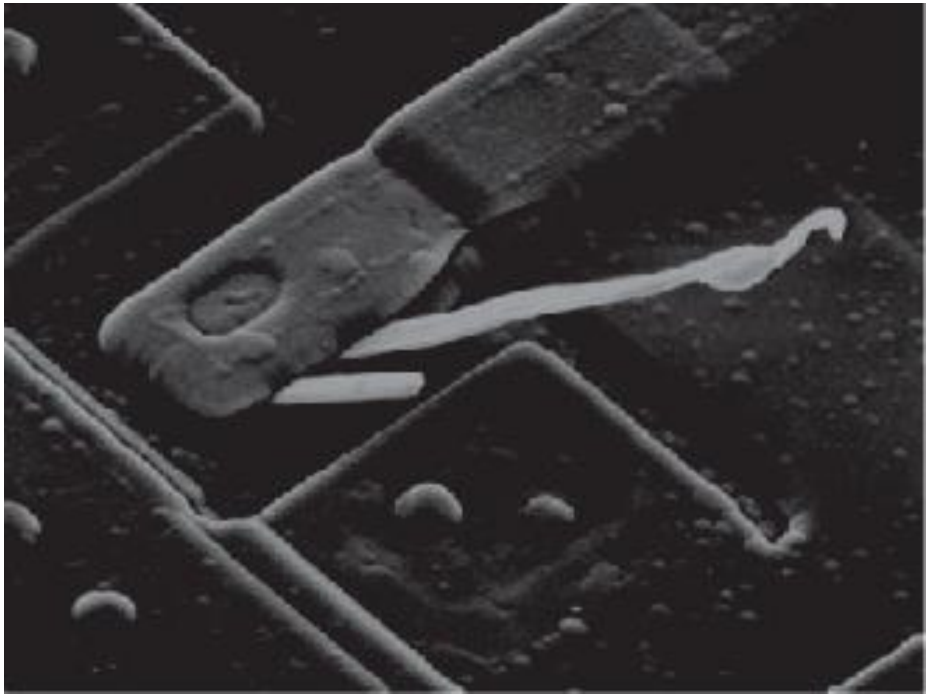
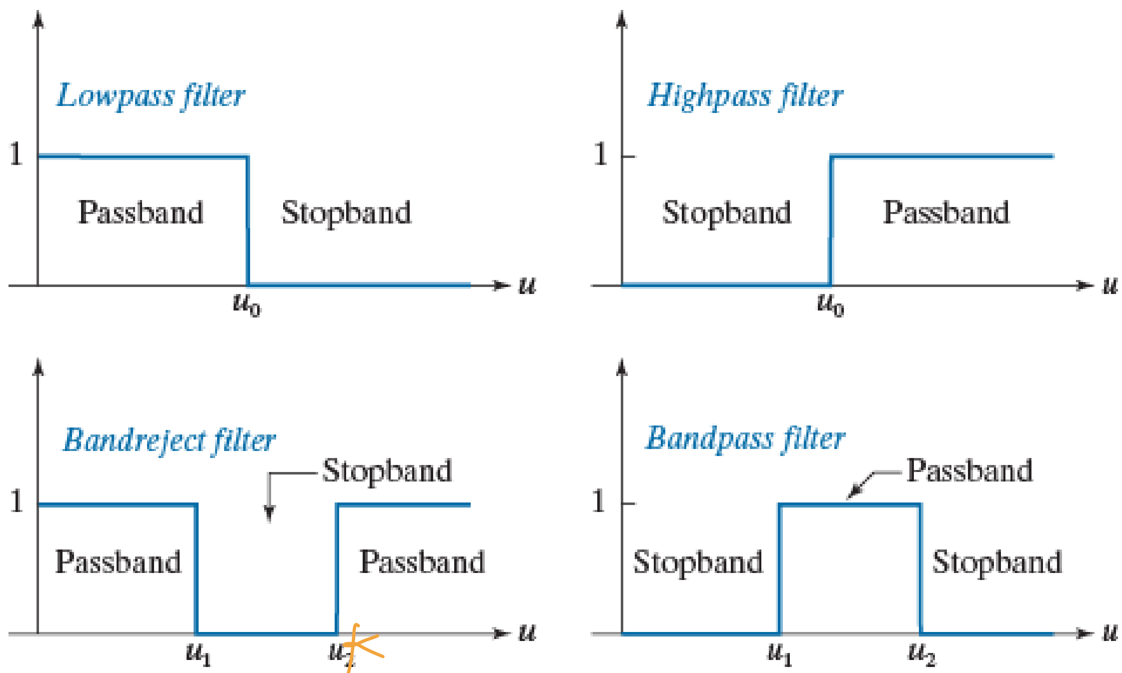


FIGURE 4.29
Result of filtering the image in Fig. 4.28(a) with a filter transfer function that sets to 0 the dc term, $F(P/2, Q/2)$, in the centered Fourier transform, while leaving all other transform terms unchanged.

Filtering in frequency domain

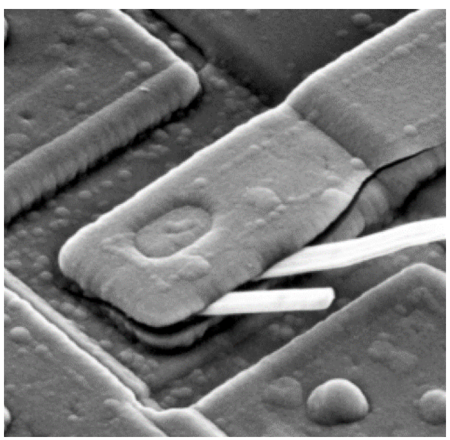
- 4 types of filters



Frequency Domain Filtering

Basic Filtering form: $g(x,y) = \mathcal{F}^{-1}[H(u,v)F(u,v)]$

Original image



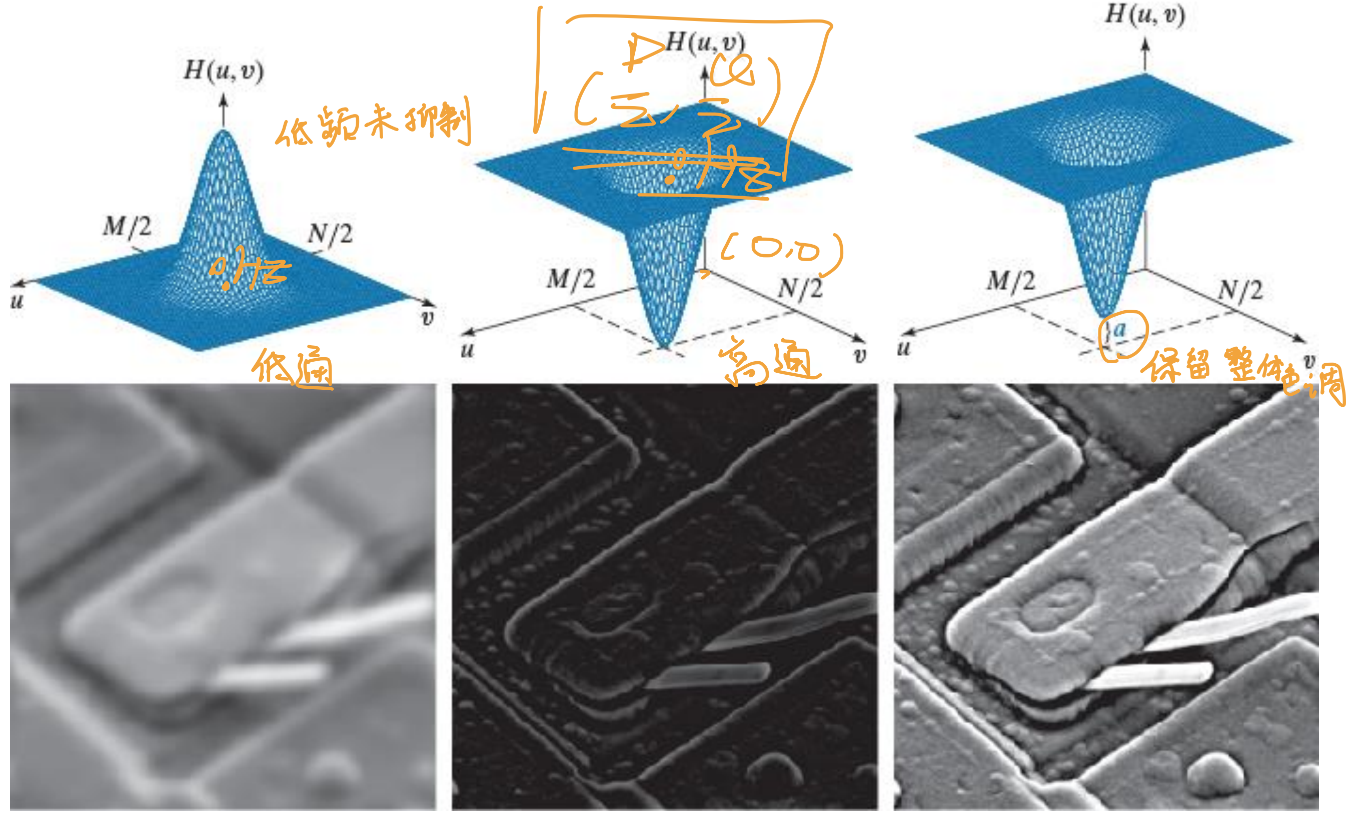


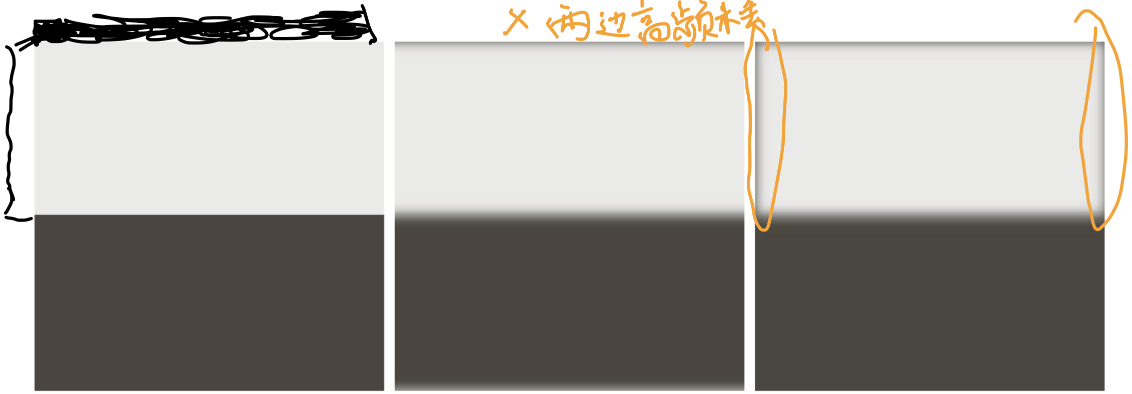
FIGURE 4.30 Top row: Frequency domain filter transfer functions of (a) a lowpass filter, (b) a highpass filter, and (c) an offset highpass filter. Bottom row: Corresponding filtered images obtained using Eq. (4-104). The offset in (c) is $a = 0.85$, and the height of $H(u,v)$ is 1. Compare (f) with Fig. 4.28(a).

Padding or not?

mirror

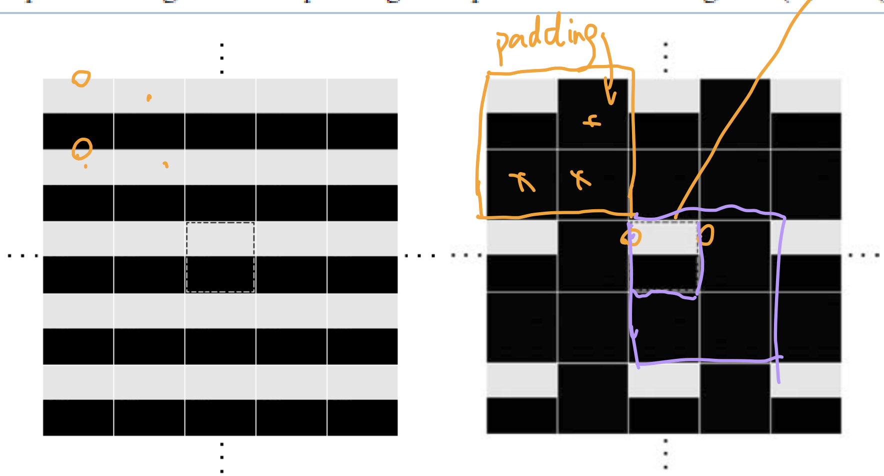
× 两边高频[illegible]

低通去高频



a b c

FIGURE 4.31 (a) A simple image. (b) Result of blurring with a Gaussian lowpass filter without padding. (c) Result of lowpass filtering with zero padding. Compare the vertical edges in (b) and (c).



a b

FIGURE 4.32 (a) Image periodicity without image padding. (b) Periodicity after padding with 0's (black). The dashed areas in the center correspond to the image in Fig. 4.31(a). Periodicity is inherent when using the DFT. (The thin white lines in both images are superimposed for clarity; they are not part of the data.)

Steps for frequency domain filtering

一般步骤：频率域滤波

1. Given an input image f(x,y) of size MxN, obtain the padding parameters P and Q. Typically, P = 2M and Q = 2N.

2. Form a padded image, $f_p(x,y)$ of size PxQ by appending the necessary number of zeros to f(x,y)

3. Multiply $f_p(x,y)$ by $(-1)^{x+y}$ to center its transform 即 fftshift

   不改变F的相位信息. ✓ 要=0

4. Compute the DFT, F(u,v) of the image from step 3

   zero phase shift filter $H(u,v) = |H(u,v)|e^{i0}$

5. Generate a real, symmetric filter function*, H(u,v), of size PxQ with center at coordinates (P/2, Q/2)

*generate from a given spatial filter, we pad the spatial filter, multiply the expadded array by $(-1)^{x+y}$, and compute the DFT of the result to obtain a centered H(u,v).

零频率在中心

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Steps for frequency domain filtering

6. Form the product G(u,v) = H(u,v)F(u,v) using array multiplication 对应元素 ×.

7. Obtain the processed image

$$g_p(x,y) = \left\{ real\left[\Im^{-1}\left[G(u,v) \right] \right] \right\} (-1)^{x+y}$$

误差

还原③

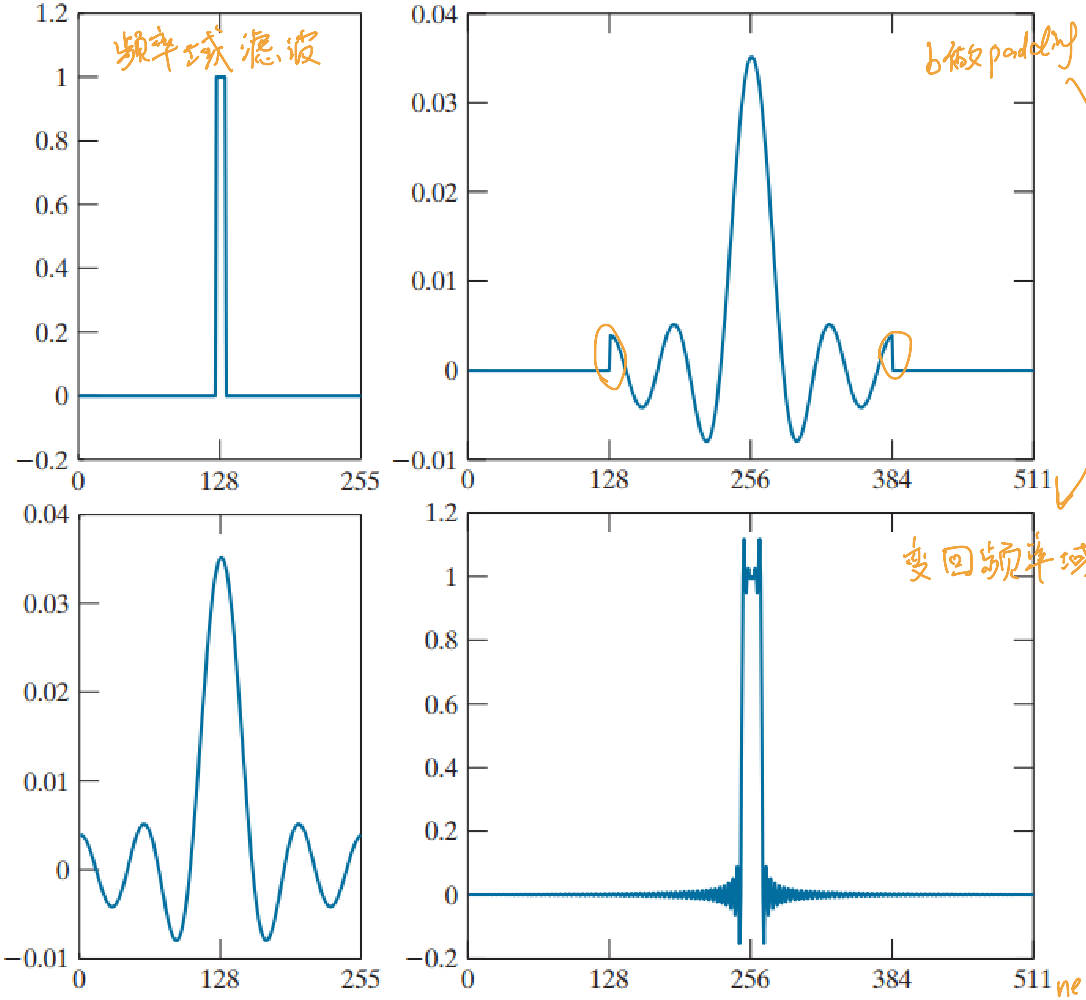
8. Obtain the final processed result, g(x,y), by extracting the MxN region from the top, left quadrant 象限 of $g_p(x,y)$

裁剪 Padding MxN

Note

a	c
b	d

FIGURE 4.33
(a) Filter transfer function specified in the (centered) frequency domain. (b) Spatial representation (filter kernel) obtained by computing the IDFT of (a). (c) Result of padding (b) to twice its length (note the discontinuities). (d) Corresponding filter in the frequency domain obtained by computing the DFT of (c). Note the ringing caused by the discontinuities in (c). Part (b) of the figure is below (a), and (d) is below (c).



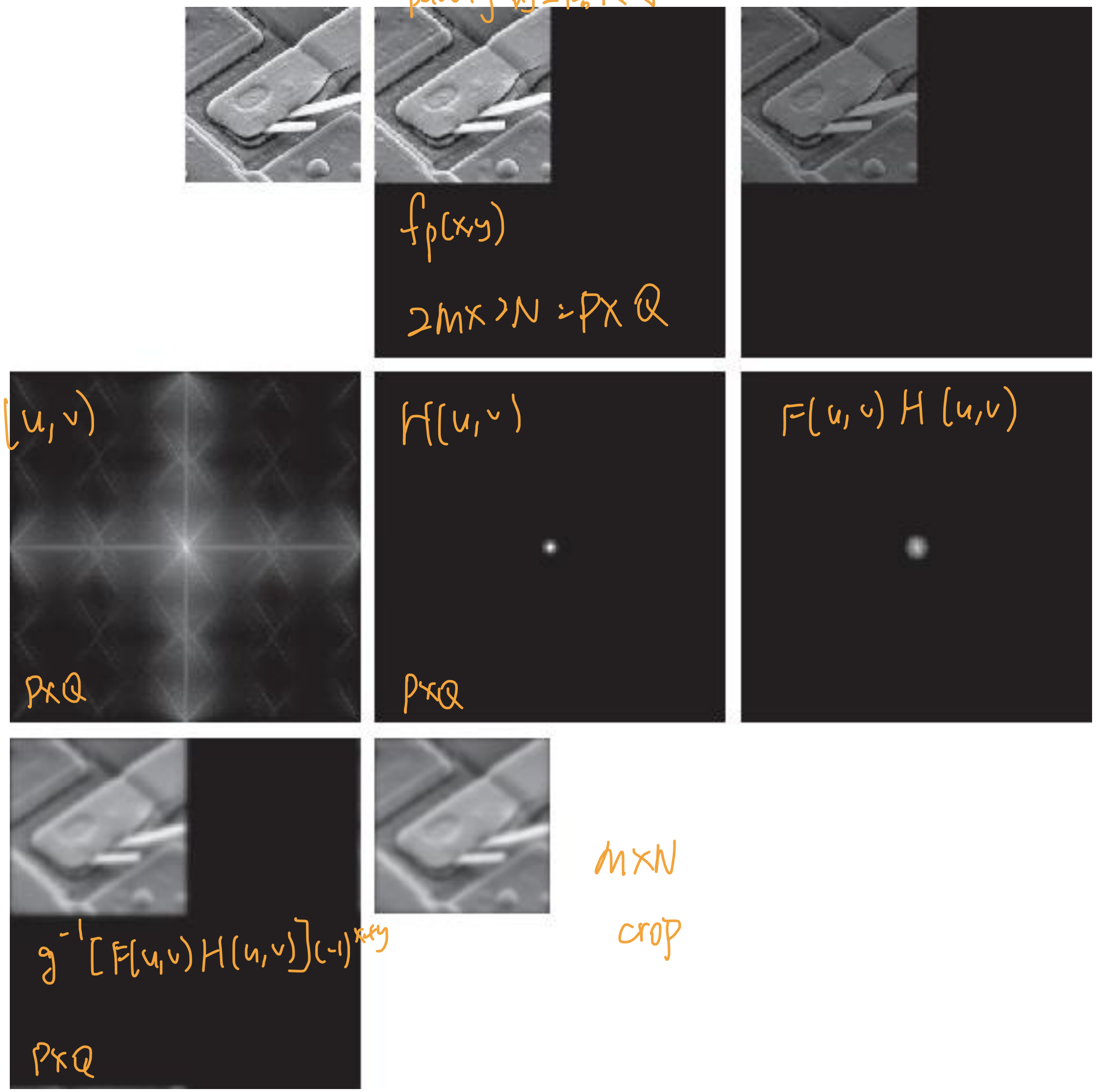
need 缓慢过渡到0

由于截断误差

Steps for frequency domain filtering

HW2

$f_p(x,y)(-1)^{x+y}$



Fourier 逆

$M \times N$

crop

a b c
d e f
g h

FIGURE 4.35 (a) An $M \times N$ image, f. (b) Padded image, f_p of size $P \times Q$. (c) Result of multiplying f_p by $(-1)^{x+y}$. 频谱 (d) Spectrum of F. (e) Centered Gaussian lowpass filter transfer function, H, of size $P \times Q$. (f) Spectrum of the product HF. (g) Image g_p, the real part of the IDFT of HF, multiplied by $(-1)^{x+y}$. (h) Final result, g, obtained by extracting the first M rows and N columns of g_p.

Filtering in Spatial and Frequency Domains

减少运算量 (时域不如频域)

kernel size 要大

- **Frequency filters ⇒ Spatial filter** $H(u,v) \Rightarrow h(x,y)$

逆变换到空间域上

Design filter in frequency domain → Compute IDFT → Full size spatial filter → Smaller spatial masks

- **Gaussian Filter**

变到空间域上也是 Gaussian

σ ⇒ 胖瘦 频域 σ↑ 胖 spatial 瘦

$$H(u) = Ae^{-\frac{u^2}{2\sigma^2}} \Leftrightarrow h(x) = \sqrt{2\pi}\sigma A e^{-2\pi^2\sigma^2x^2}$$ 低通

高通

$$H(u) = Ae^{-\frac{u^2}{2\sigma_1^2}} - Be^{-\frac{u^2}{2\sigma_2^2}} \Leftrightarrow h(x) = \sqrt{2\pi}\sigma_1 A e^{-2\pi^2\sigma_1^2x^2} - \sqrt{2\pi}\sigma_2 B e^{-2\pi^2\sigma_2^2x^2}$$

二维

$$H(u,v) = Ae^{-\frac{u^2+v^2}{2\sigma^2}} \Leftrightarrow h(x,y) = A2\pi\sigma^2 e^{-2\pi^2\sigma^2(x^2+y^2)}$$

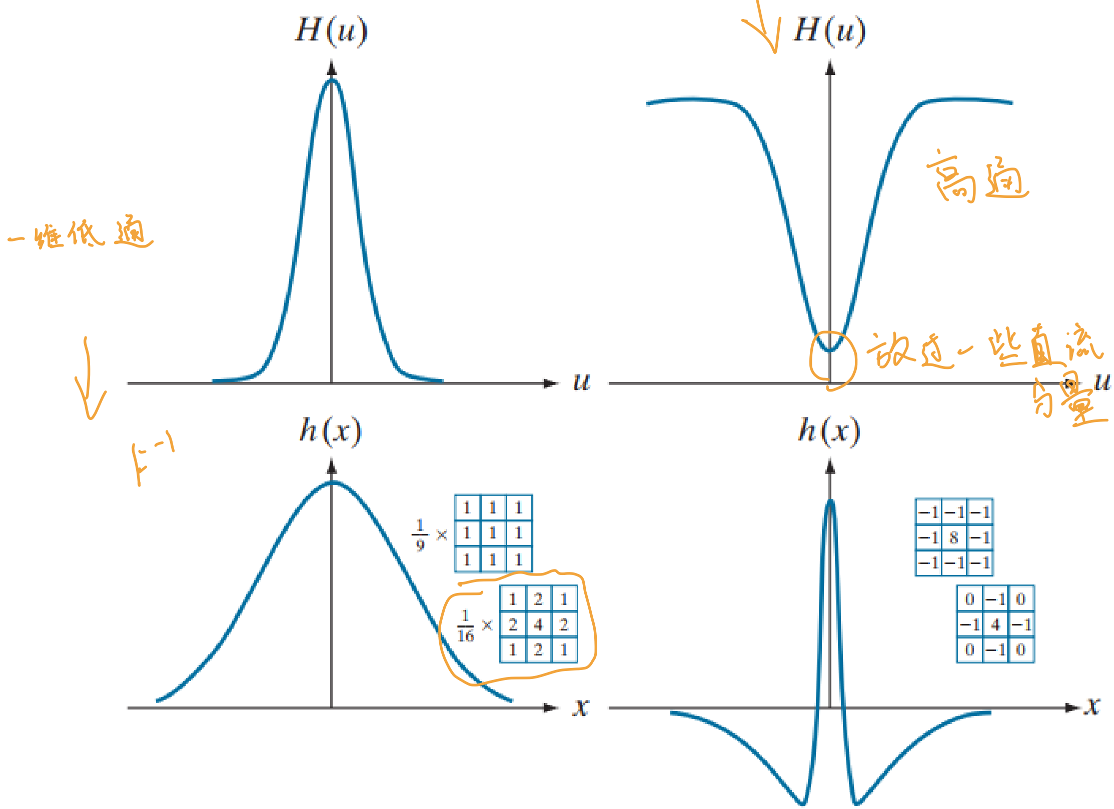
Filtering in Spatial and Frequency Domains

➢Frequency filters ⇒ Spatial filter $H(u,v) \Rightarrow h(x,y)$

$$H(u) = Ae^{-\frac{u^2}{2\sigma^2}} \Leftrightarrow h(x) = \sqrt{2\pi}\sigma A e^{-2\pi^2\sigma^2x^2}$$

A>B, σ1>σ2

$$H(u) = Ae^{-\frac{u^2}{2\sigma_1^2}} - Be^{-\frac{u^2}{2\sigma_2^2}} \Leftrightarrow h(x) = \sqrt{2\pi}\sigma_1 A e^{-2\pi^2\sigma_1^2x^2} - \sqrt{2\pi}\sigma_2 B e^{-2\pi^2\sigma_2^2x^2}$$



a	c
b	d

FIGURE 4.36 (a) A 1-D Gaussian lowpass transfer function in the frequency domain. (b) Corresponding kernel in the spatial domain. (c) Gaussian highpass transfer function in the frequency domain. (d) Corresponding kernel. The small 2-D kernels shown are kernels we used in Chapter 3.

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Even and Odd Functions

- **Even function**

$f(0) = f(4-0) = f(4)$

$$w_e(x,y) = w_e(M-x, N-y)$$

$$f = \{f(0), f(1), f(2), f(3)\} = \{2,1,1,1\}$$

$$f(0) = f(4), f(1) = f(3), f(2) = f(2), f(3) = f(1)$$

$\frac{M}{2}$

M is even, then $\{\boldsymbol{a}, \boldsymbol{b}, \boldsymbol{c}, \boldsymbol{b}\}$

$M=5$ $\frac{M}{2}=2.5$

M is odd, then $\{\boldsymbol{a}, \boldsymbol{b}, \boldsymbol{c}, \boldsymbol{c}, \boldsymbol{b}\}$

0 1 2 3

实偶与实(偶)，相位0

- **Odd function**

$$w_o(x,y) = -w_o(M-x, N-y)$$

$$g = \{g(0), g(1), g(2), g(3)\} = \{0, -1, 0, 1\}$$

周期且互为相反数

$$\boxed{g(0) = -g(4) = 0}, g(1) = -g(3), g(2) = -g(2), g(3) = -g(1)$$

轴上必须为0

$\frac{M}{2}$

M is even, then $\{\boldsymbol{0}, \boldsymbol{b}, \boldsymbol{0}, -\boldsymbol{b}\}$

$\frac{M}{2}=2.5$

M is odd, then $\{\boldsymbol{0}, \boldsymbol{b}, \boldsymbol{c}, -\boldsymbol{c}, -\boldsymbol{b}\}$

实虚与纯虚(奇)，相角90°

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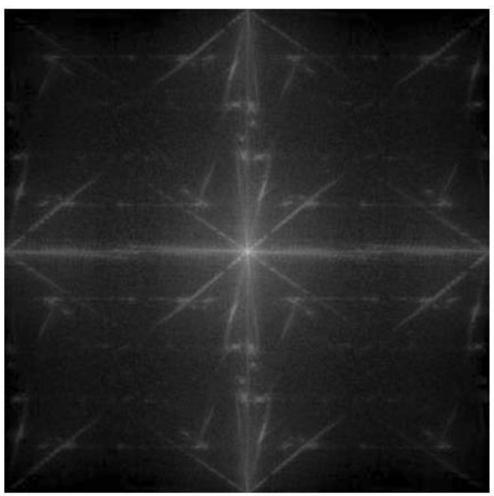
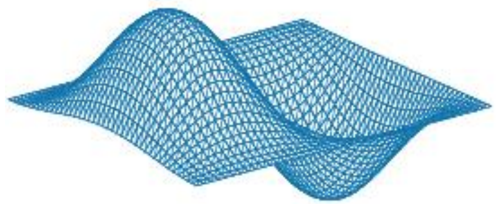
Spatial and Frequency Filtering



−1	0	1
−2	0	2
−1	0	1

提取水平方向信息



- **Zero Padding**

变成P×Q
kernel移到
可以奇/偶函数
⇒中心 $\frac{M}{2}$

$$\cdots\cdots\quad \begin{array}{c|ccccc|}
0 & 0 & 0 & 0 & 0 & 0 \\ \hline
0 & 0 & 0 & 0 & 0 & 0 \\
0 & 0 & -1 & 0 & 1 & 0 \\
0 & 0 & -2 & 0 & 2 & 0 \\
0 & 0 & -1 & 0 & 1 & 0 \\
0 & 0 & 0 & 0 & 0 & 0 \\ \hline
\end{array}\quad \cdots\cdots$$

- **Odd Function (奇函数)**

$$w_o(x,y) = -w_o(-x,-y)$$

$$w_o(x,y) = -w_o(M-x, N-y)$$

$f(x,y)$ real and odd $\Leftrightarrow$ $F(u,v)$ imaginary and odd

Procedure

(3)人为把谱实部置为0

The procedure used to generate $H(u,v)$ is: (1) multiply $h_p(x,y)$ by $(-1)^{x+y}$ to center the frequency domain filter; (2) compute the forward DFT of the result in (1) to generate $H(u,v)$; (3) set the real part of $H(u,v)$ to 0 to account for parasitic real parts (we know that H has to be purely imaginary because h_p is real and odd); and (4) multiply the result by $(-1)^{u+v}$. This last step reverses the multiplication of $H(u,v)$ by $(-1)^{u+v}$, which is implicit when $h(x,y)$ was manually placed in the center of $h_p(x,y)$. Figure 4.38(a) shows a perspective plot of $H(u,v)$, and Fig. 4.38(b) shows $H(u,v)$ as an image. Note the antisymmetry in this image about its center, a result of $H(u,v)$ being odd. Function $H(u,v)$ is used as any other frequency domain filter transfer function. Figure 4.38(c) is the result of using the filter transfer function just obtained to filter the image in Fig. 4.37(a) in the frequency domain, using the step-by-step filtering procedure outlined earlier. As expected from a derivative filter, edges were enhanced and all the constant intensity areas were reduced to zero (the grayish tone is due to scaling for display). Figure 4.38(d) shows the result of filtering the same image in the spatial domain with the Sobel kernel $h(x,y)$, using the procedure discussed in Section 3.6. The results are identical.

4) Translation to center of the frequency rectangle, $(M/2, N/2)$

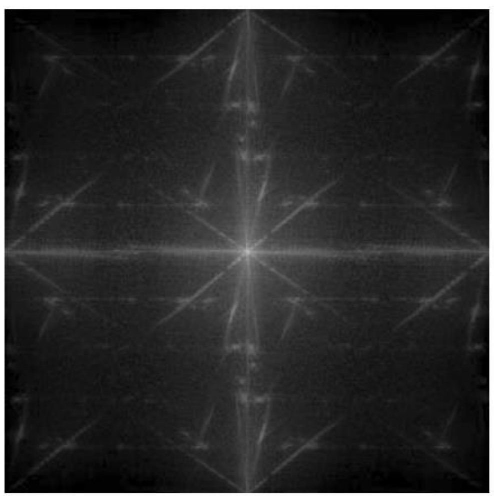
手动挪

$$f(x,y)(-1)^{x+y} \Leftrightarrow F(u-M/2, v-N/2)$$

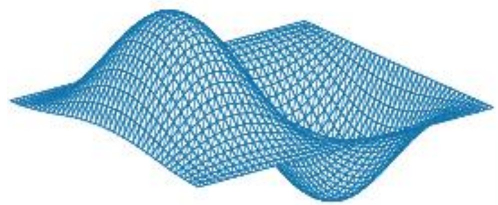
$$\boxed{f(x-M/2, y-N/2) \Leftrightarrow F(u,v)(-1)^{u+v}}$$

乘回去

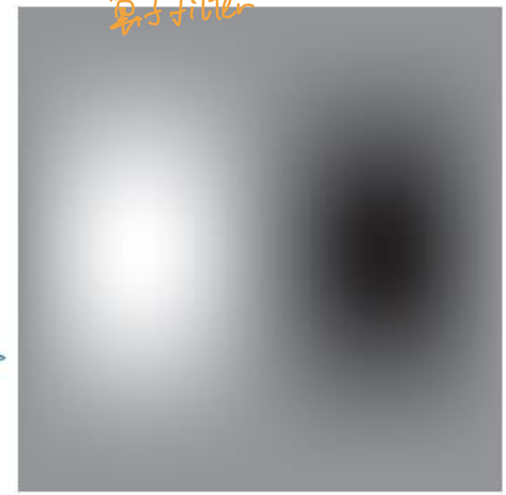
Spatial and Frequency Filtering



−1	0	1
−2	0	2
−1	0	1



算子filter

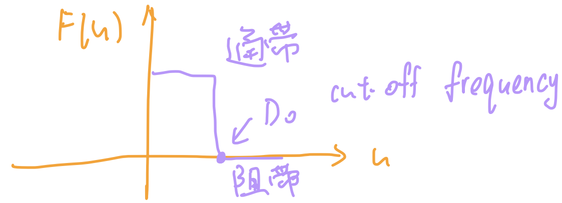


a b
c d

FIGURE 4.38
(a) A spatial kernel and perspective plot of its corresponding frequency domain filter transfer function.
(b) Transfer function shown as an image.
(c) Result of filtering Fig. 4.37(a) in the frequency domain with the transfer function in (b).
(d) Result of filtering the same image in the spatial domain with the kernel in (a). The results are identical.

Lowpass filtering

- Ideal Lowpass filter
- Butterworth Lowpass filter
- Gaussian Lowpass filter



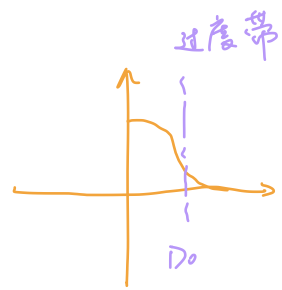


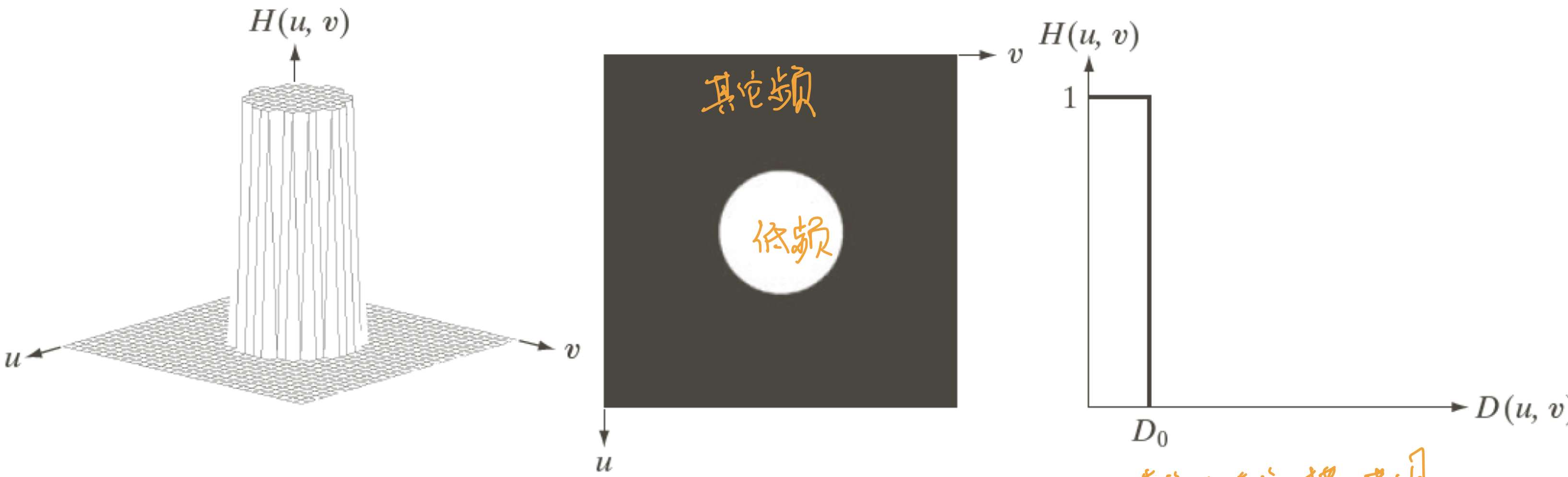
TABLE 4.5
Lowpass filter transfer functions. D_0 is the cutoff frequency, and n is the order of the Butterworth filter.

Ideal	Gaussian	Butterworth
$H(u,v) = \begin{cases} 1 & \text{if } D(u,v) \le D_0 \\ 0 & \text{if } D(u,v) > D_0 \end{cases}$	$H(u,v) = e^{-D^2(u,v)/2D_0^2}$	$H(u,v) = \dfrac{1}{1 + \left[D(u,v)/D_0\right]^{2n}}$

Ideal Lowpass filter

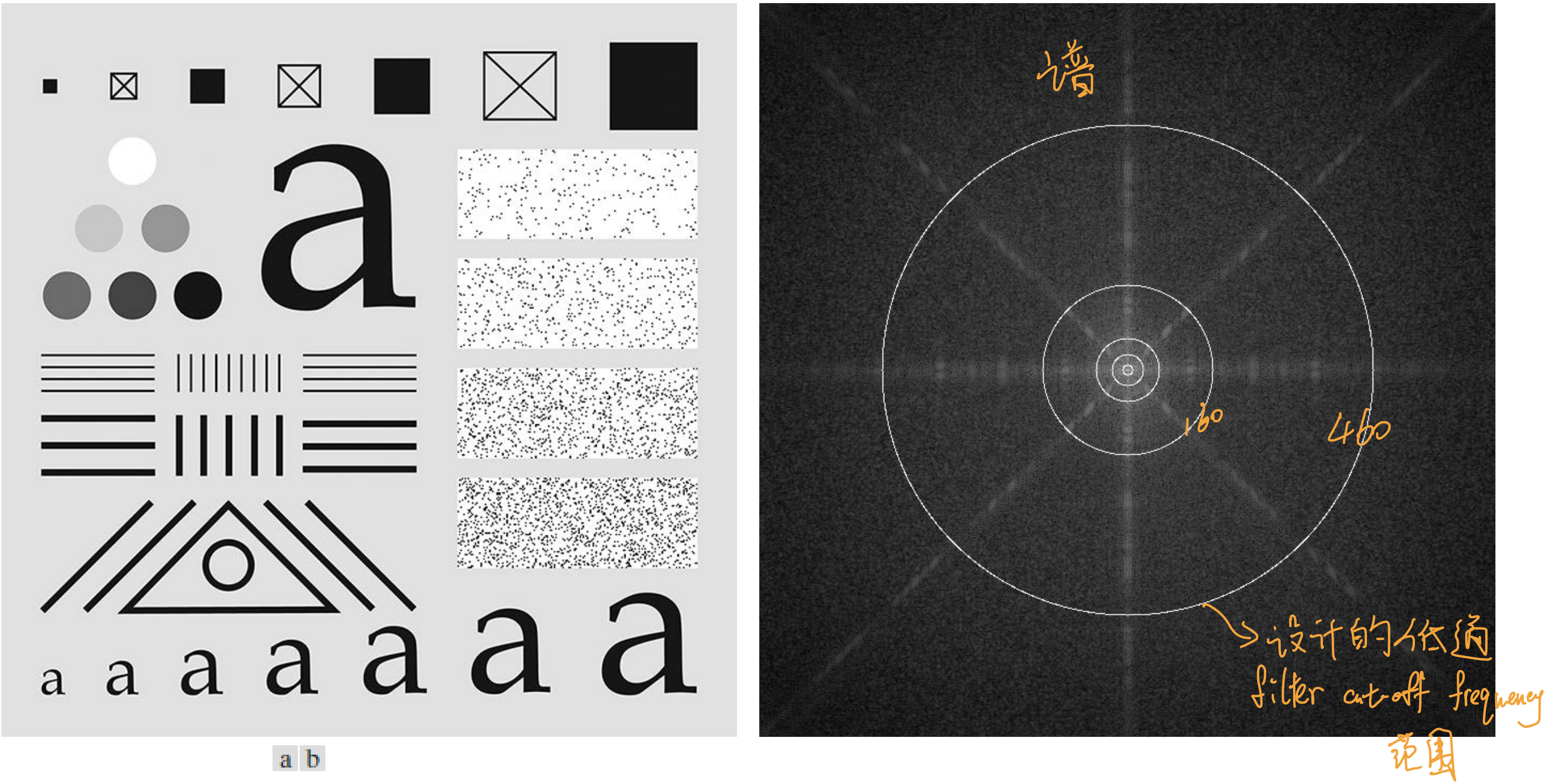
Ideal Lowpass Filter (ILPF):

$$H(u,v) = \begin{cases} 1, & D(u,v) \leq D_0 \\ 0, & D(u,v) > D_0 \end{cases} \qquad D(u,v) = \left[(u - \frac{P}{2})^2 + (v - \frac{Q}{2})^2 \right]^{1/2}$$



D_0越小越模糊

Ideal Lowpass filter (cutoff frequency)



a b

FIGURE 4.40 (a) Test pattern of size 688 × 688 pixels, and (b) its spectrum. The spectrum is double the image size as a result of padding, but is shown half size to fit. The circles have radii of 10, 30, 60, 160, and 460 pixels with respect to the full-size spectrum. The radii enclose 86.9, 92.8, 95.1, 97.6, and 99.4% of the padded image power, respectively.

Ideal Lowpass filter (cutoff frequency)

60

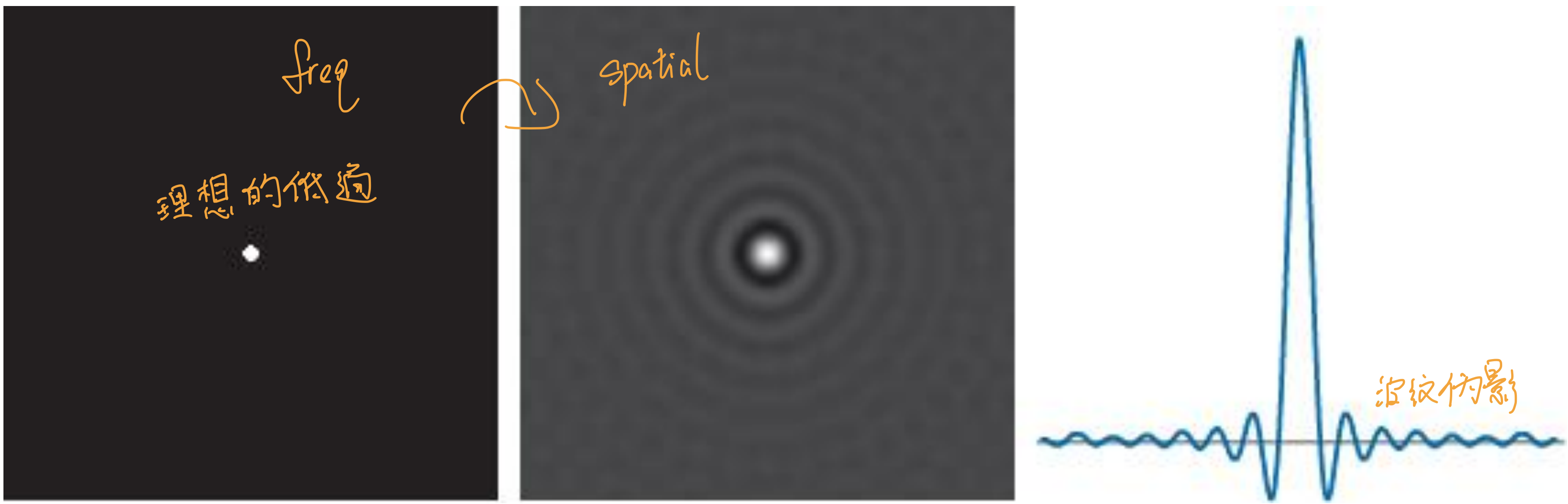
波纹样的伪影

why无伪影? freq大 => spatial窄 => 波纹很快衰减

a b c
d e f

FIGURE 4.41 (a) Original image of size 688 × 688 pixels. (b)–(f) Results of filtering using ILPFs with cutoff frequencies set at radii values 10, 30, 60, 160, and 460, as shown in Fig. 4.40(b). The power removed by these filters was 13.1, 7.2, 4.9, 2.4, and 0.6% of the total, respectively. We used mirror padding to avoid the black borders characteristic of zero padding, as illustrated in Fig. 4.31(c).

Ideal Lowpass filter (cutoff frequency)



a b c

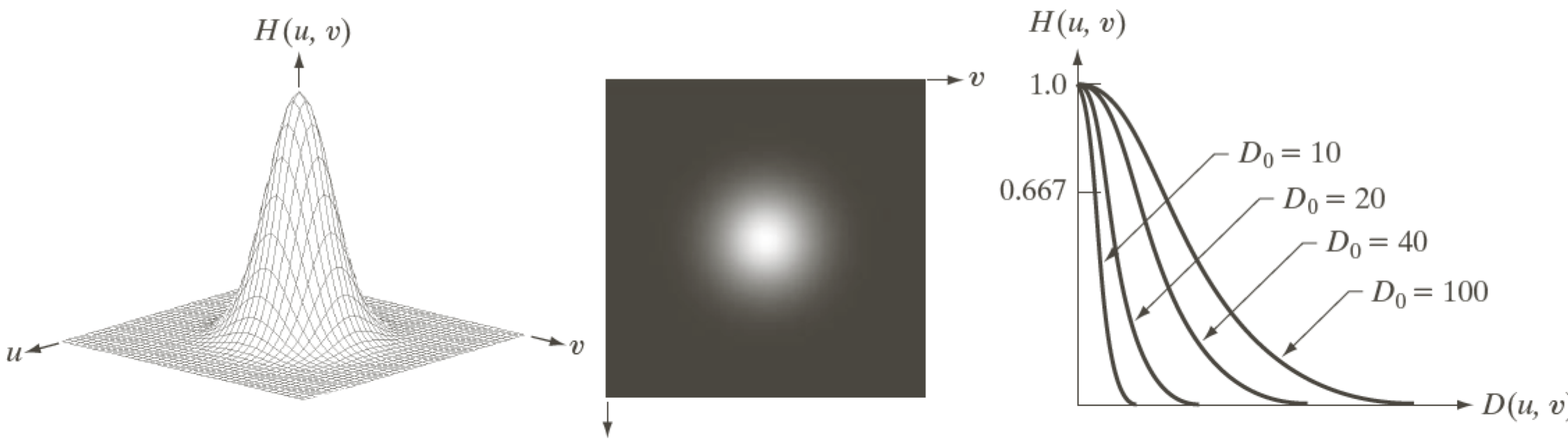
FIGURE 4.42 (a) Frequency domain ILPF transfer function. (b) Corresponding spatial domain kernel function. (c) Intensity profile of a horizontal line through the center of (b).

Gaussian Lowpass filter

$$H(u,v) = e^{-\frac{D(u,v)^2}{2D_0^2}}$$

Where $H(u,v) = 0.607$ when $D(u,v) = D_0$

D_0↑ 放过频频越多

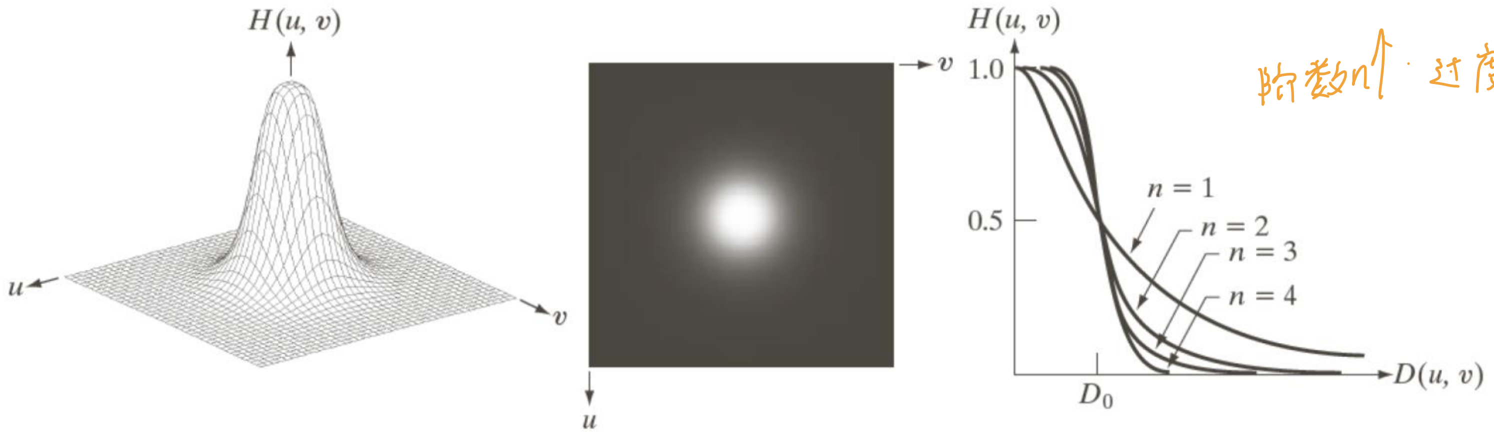


Gaussian Lowpass filter

Butterworth Lowpass filter

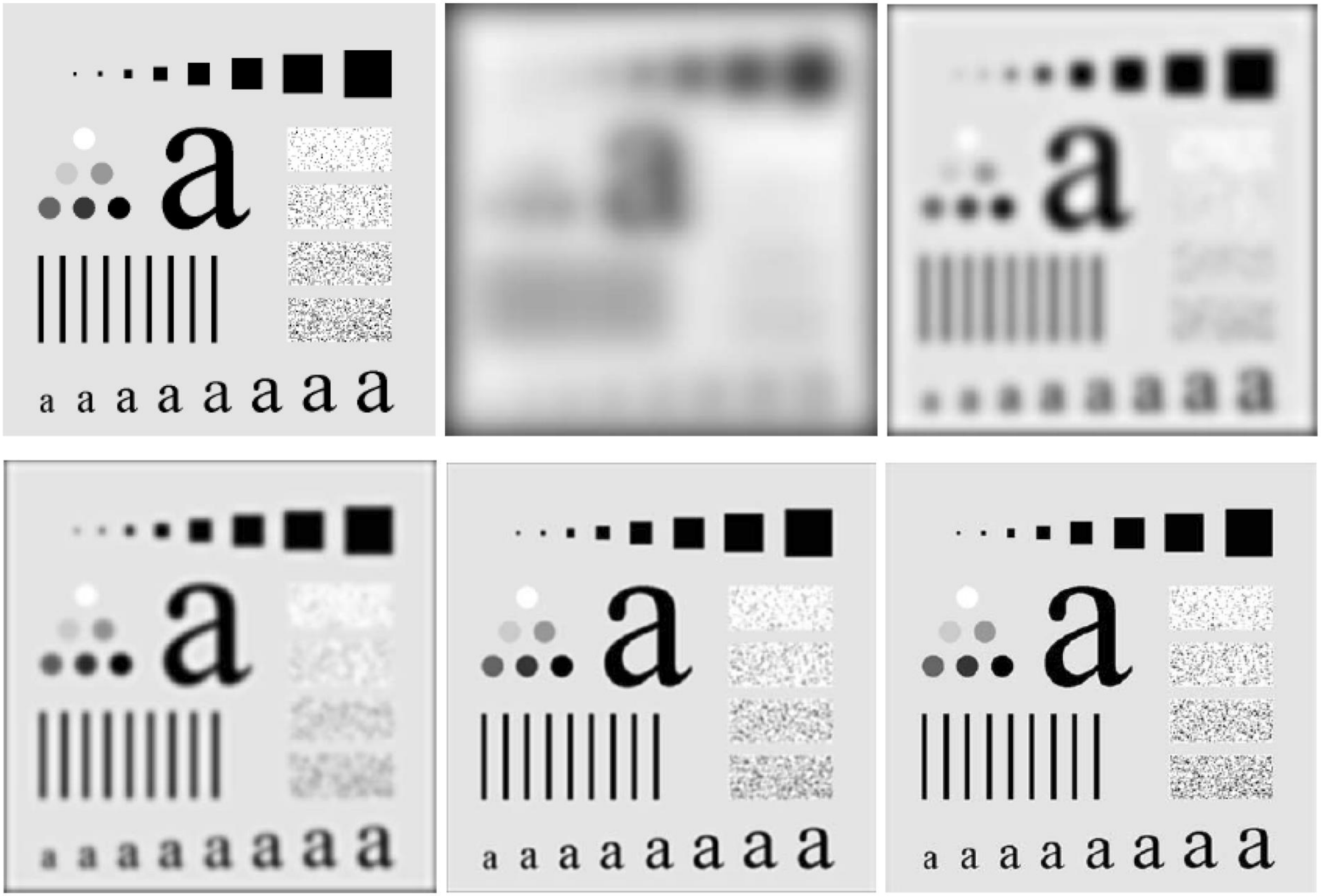
$$H(u,v) = \frac{1}{1 + \left[\frac{D(u,v)}{D_0}\right]^{2n}}$$

Where $D(u,v) = \left[(u - \frac{P}{2})^2 + (v - \frac{Q}{2})^2\right]^{1/2}$, and $H(u,v) = 0.5$ when $D(u,v) = D_0$



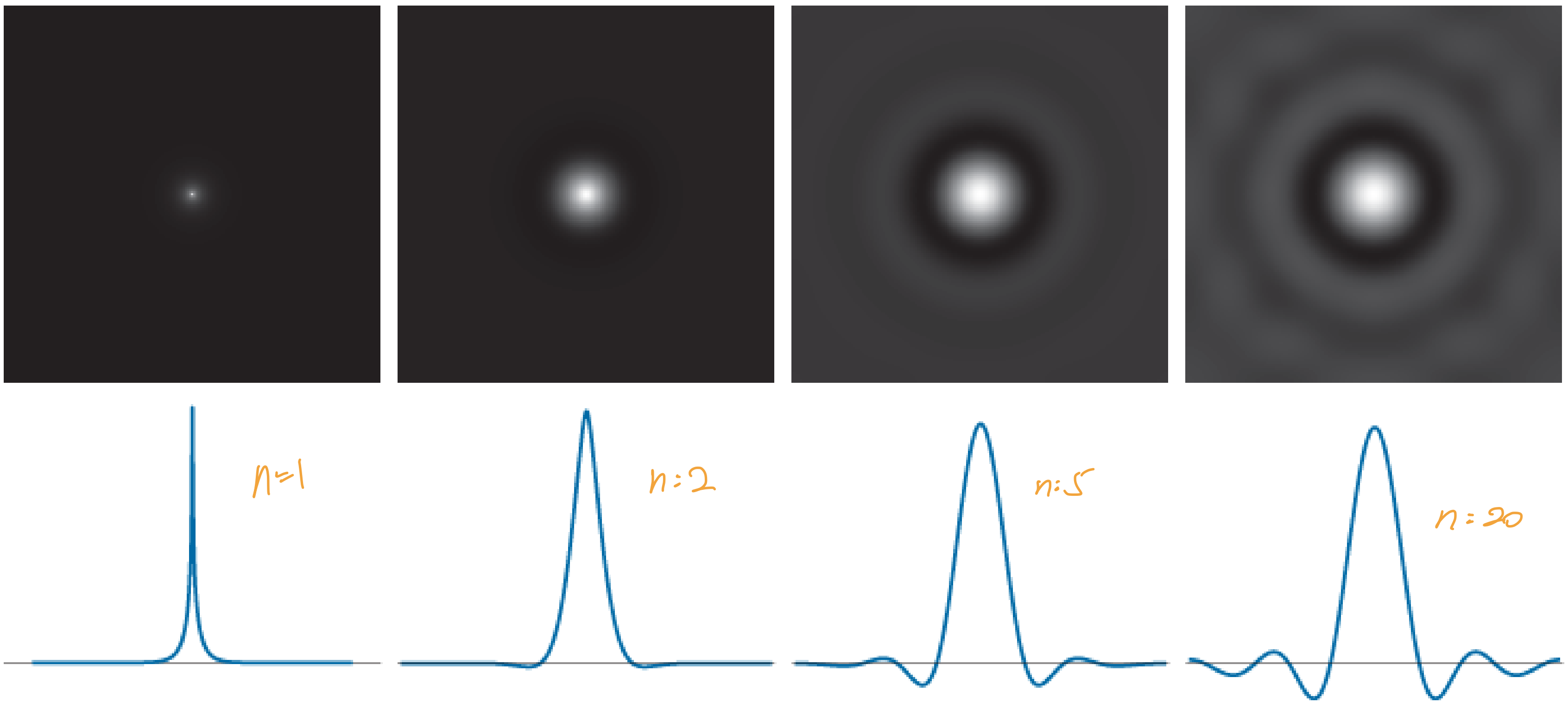
阶数n↑· 过度带越陡

Butterworth Lowpass filter



Butterworth Lowpass filter

- Order of BLPF



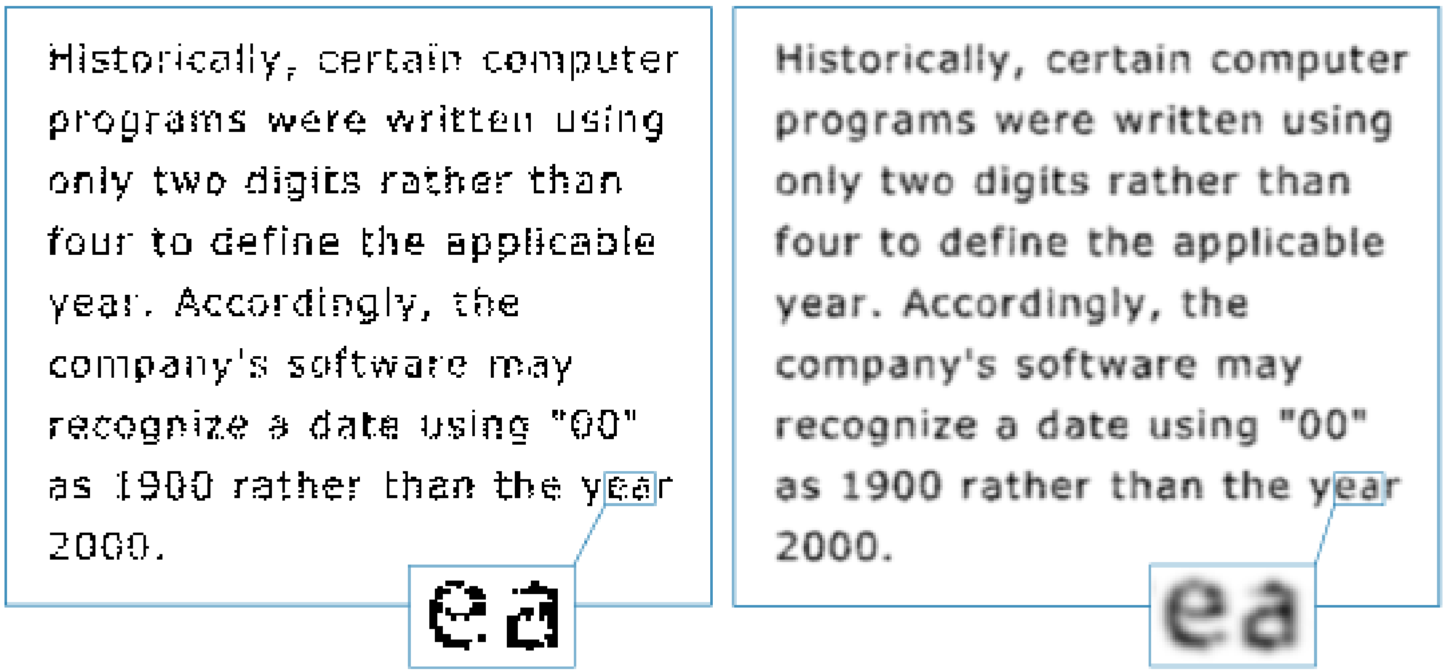
a	b	c	d
e	f	g	h

FIGURE 4.47 (a)–(d) Spatial representations (i.e., spatial kernels) corresponding to BLPF transfer functions of 1000 × 1000 pixels, cut-off frequency of 5, and order 1, 2, 5, and 20, respectively. (e)–(h) Corresponding inten profiles through the center of the filter functions.

Application of GLPF

a b

FIGURE 4.48
(a) Sample text of low resolution (note the broken characters in the magnified view).
(b) Result of filtering with a GLPF, showing that gaps in the broken characters were joined.



Historically, certain computer
programs were written using
only two digits rather than
four to define the applicable
year. Accordingly, the
company's software may
recognize a date using "00"
as 1900 rather than the year
2000.
ea

字母不连续之断点，
当成高频信息，intensity 变化

低通滤波

Application of GLPF

低通滤波除皱纹，光滑
→磨皮

睫毛也模糊

FIGURE 4.49 (a) Original 785×732 image. (b) Result of filtering using a GLPF with $D_0 = 150$. (c) Result of filtering using a GLPF with $D_0 = 130$. Note the reduction in fine skin lines in the magnified sections in (b) and (c).

Highpass filtering

- Ideal Highpass filter
- Butterworth Highpass filter
- Gaussian Highpass filter

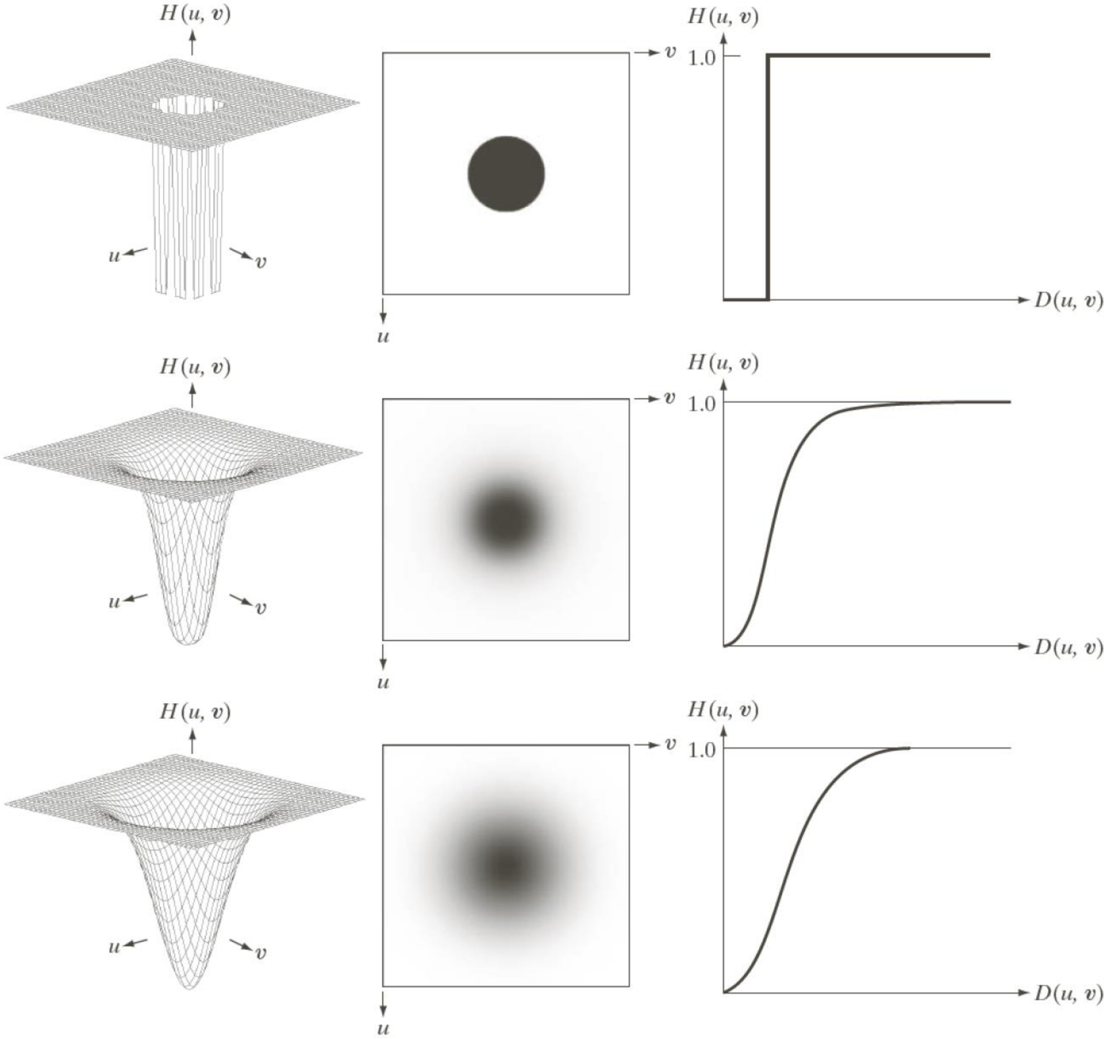
低通取反

$$H_{\text{HP}}(u,v) = 1 - H_{\text{LP}}(u,v)$$

TABLE 4.6
Highpass filter transfer functions. D_0 is the cutoff frequency and n is the order of the Butterworth transfer function.

Ideal	Gaussian	Butterworth
$H(u,v) = \begin{cases} 0 & \text{if } D(u,v) \le D_0 \\ 1 & \text{if } D(u,v) > D_0 \end{cases}$	$H(u,v) = 1 - e^{-D^2(u,v)/2D_0^2}$	$H(u,v) = \dfrac{1}{1 + [D_0/D(u,v)]^{2n}}$

Highpass filtering



Highpass filtering

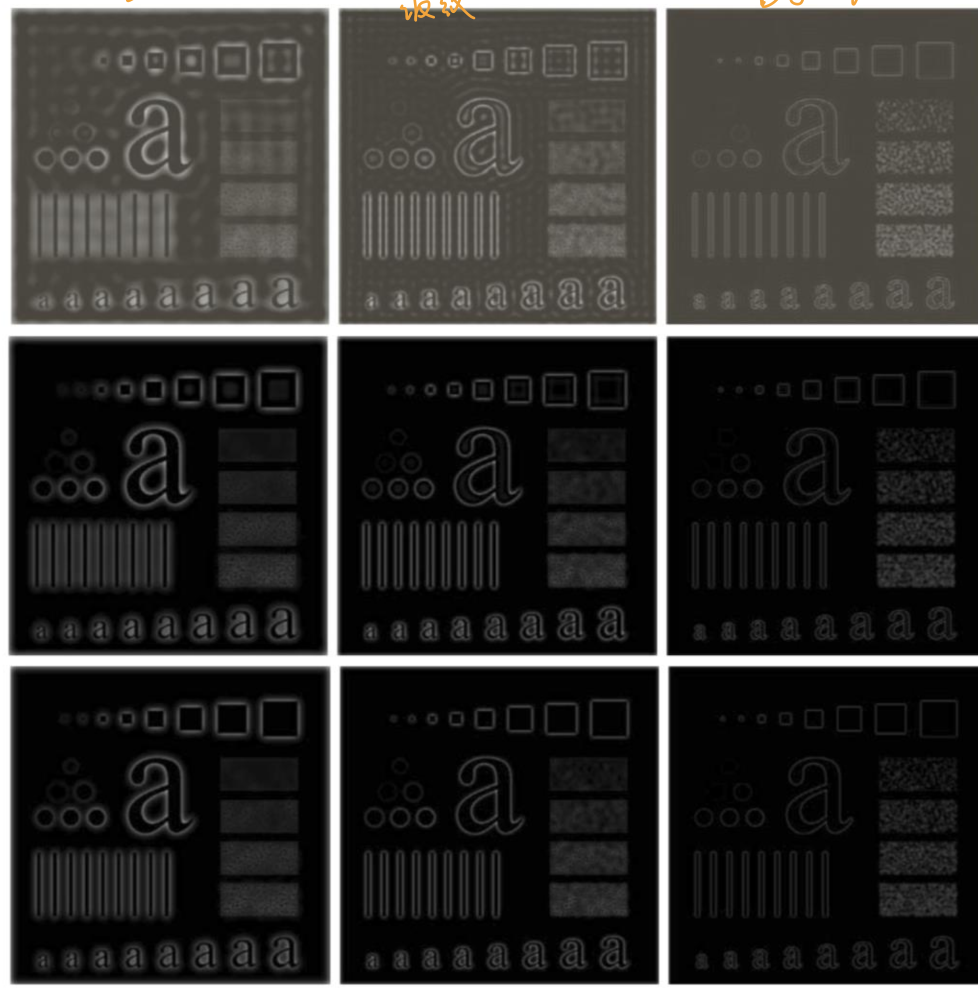
$D_0=60$ 波纹 (振铃现象) $D_0=160$

IHPF 高通

BHPF Butterworth

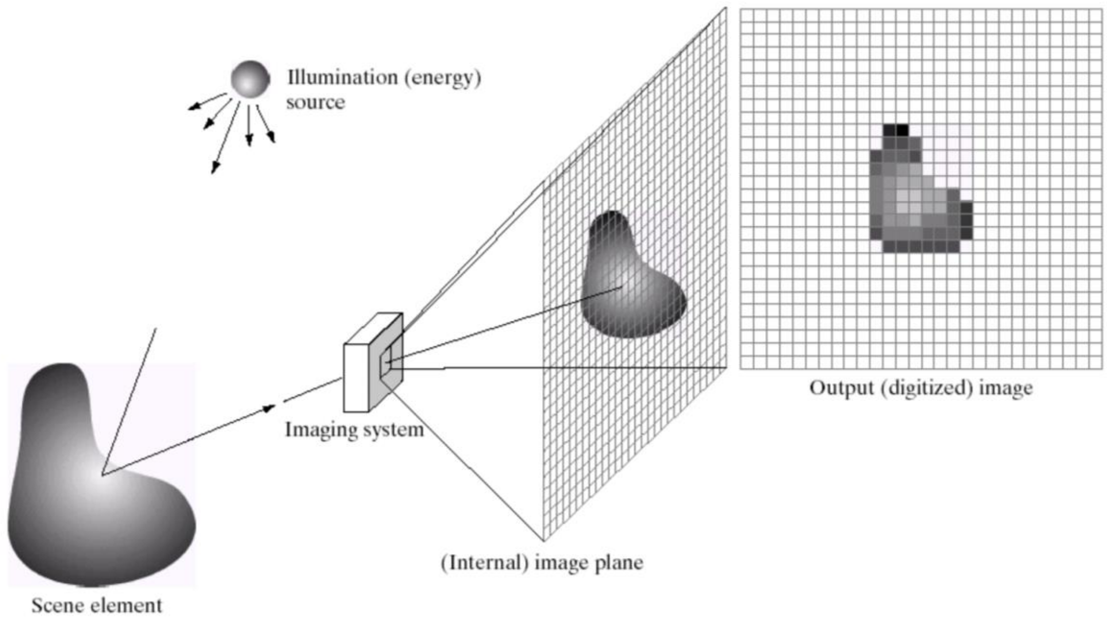
过多能量滤走

GHPF



Highpass Filtering-Homomorphic Filtering

- Homomorphic Filtering （同态滤波）



光照

$$f(x, y) = i(x, y)r(x, y) \qquad 0 < i(x, y) < \infty, 0 \leq r(x, y) < 1$$

该位置对光反射能力

Homomorphic Filtering

- We first transform the multiplicative components to additive components by moving to the log domain.

$$f(x,y) = i(x,y)r(x,y) \qquad 0 < i(x,y) < \infty, 0 \leq r(x,y) < 1$$

Let $z(x,y) = \ln f(x,y) = \ln i(x,y) + \ln r(x,y)$ 取对数

低频 高频

$$\boldsymbol{Z(u,v) = F_i(u,v) + F_r(u,v)}$$

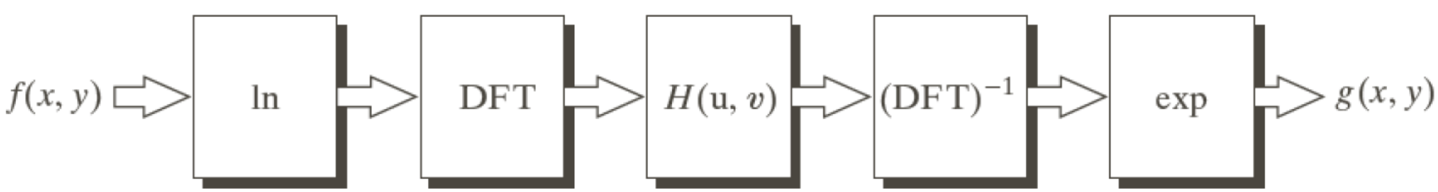
Slow variation Rapid variation

filter

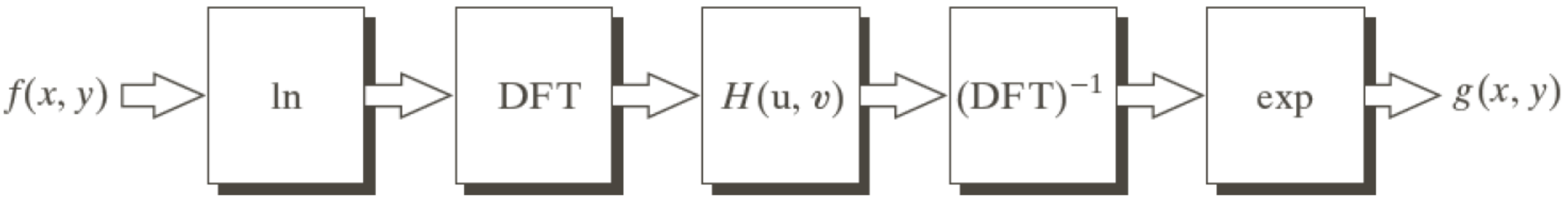
$$s(x,y) = \mathcal{F}^{-1}[H(u,v)Z(u,v)]$$
$$= \mathcal{F}^{-1}[H(u,v)F_i(u,v)] + \mathcal{F}^{-1}[H(u,v)F_r(u,v)]$$

filter 保留反射 => 高通 filter

(也叫同态滤波)



Homomorphic Filtering



- After filtering the image is reconstructed by a inverted DFT and exponential computation.

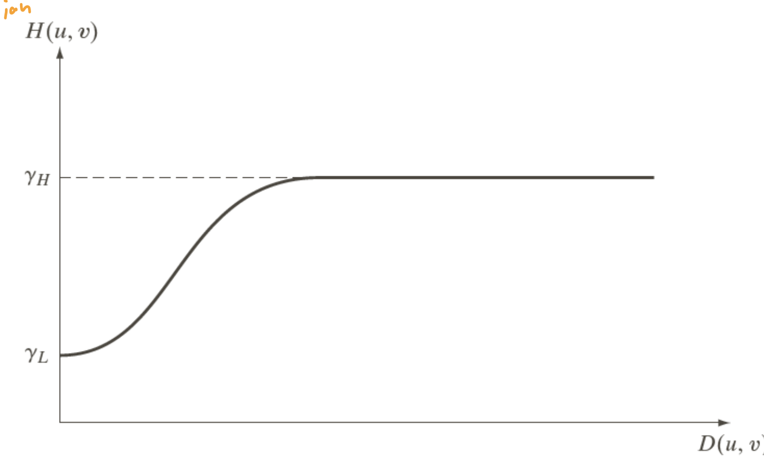
$$\boldsymbol{g(x,y) = e^{s(x,y)} = i_0(x,y)r_0(x,y)}$$

- How to design the H?

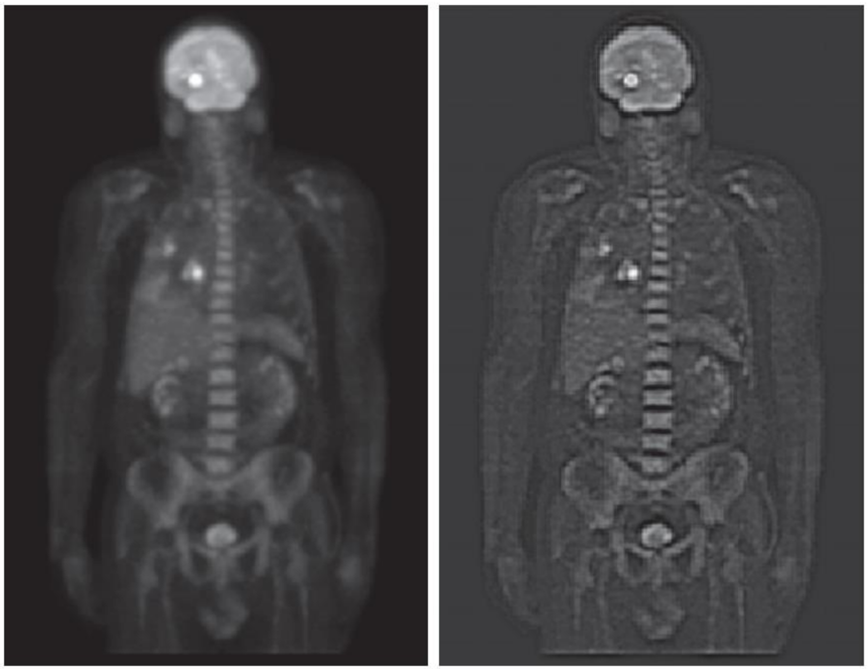
Based on Gaussian

$$H(u,v) = (\gamma_H - \gamma_L)\left[1 - e^{-c\left[\frac{D(u,v)}{D_0}\right]^2} + \gamma_L\right]$$

定义滤波器高频部分放过系数



Homomorphic Filtering



a b

FIGURE 4.60 (a) Full body PET scan. (b) Image enhanced using homomorphic filtering. (Original image courtesy of Dr. Michael E. Casey, CTI Pet Systems.)

丢失contrast

同态滤波，不会丢失信息

Homomorphic Filtering

- Homomorphic filtering is most commonly used for correcting non-uniform illumination in images.
- Illumination typically varies slowly across the image as compared to reflectance which can change quite abruptly at object edges.

光照变化缓慢，把光照与反射系数分离

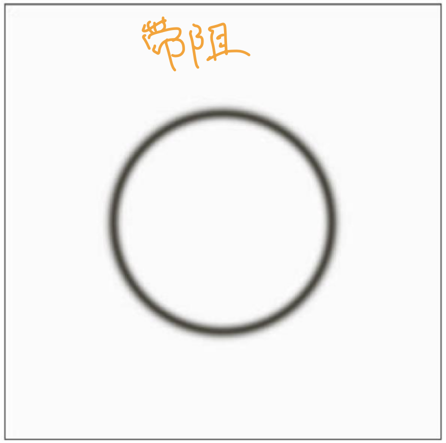
- We use a high-pass filter in the log domain to remove the low-frequency illumination component while preserving the high-frequency reflectance component.

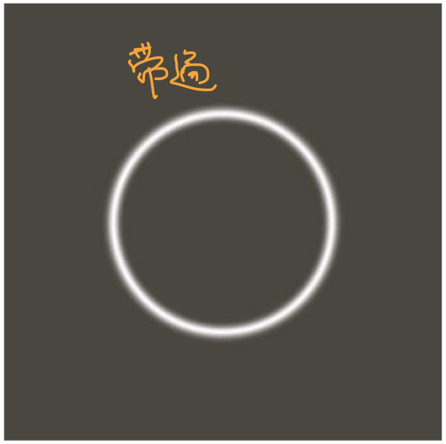
Selective Filtering

- Bandreject and Bandpass filtering

$$H_{\mathrm{BP}}(u,v) = 1 - H_{\mathrm{BR}}(u,v)$$

Ideal	Butterworth	Gaussian
$H(u,v) = \begin{cases} 0 & \text{if } D_0 - \frac{W}{2} \leq D \leq D_0 + \frac{W}{2} \\ 1 & \text{otherwise} \end{cases}$	$H(u,v) = \dfrac{1}{1 + \left[\dfrac{DW}{D^2 - D_0^2}\right]^{2n}}$	$H(u,v) = 1 - e^{-\left[\frac{D^2 - D_0^2}{DW}\right]^2}$





Selective Filtering

圆点（扣一小块）

带通 → 圆环

➢ Notch Filter (陷波滤波器)

- Reject or pass frequencies in predefined neighborhood

关于原点，对称

- Symmetric about the origin for a zero-phase shift filters
- Selectively modify local regions of the DFT

由高通filter设计

两个高（关于原点对称）相乘

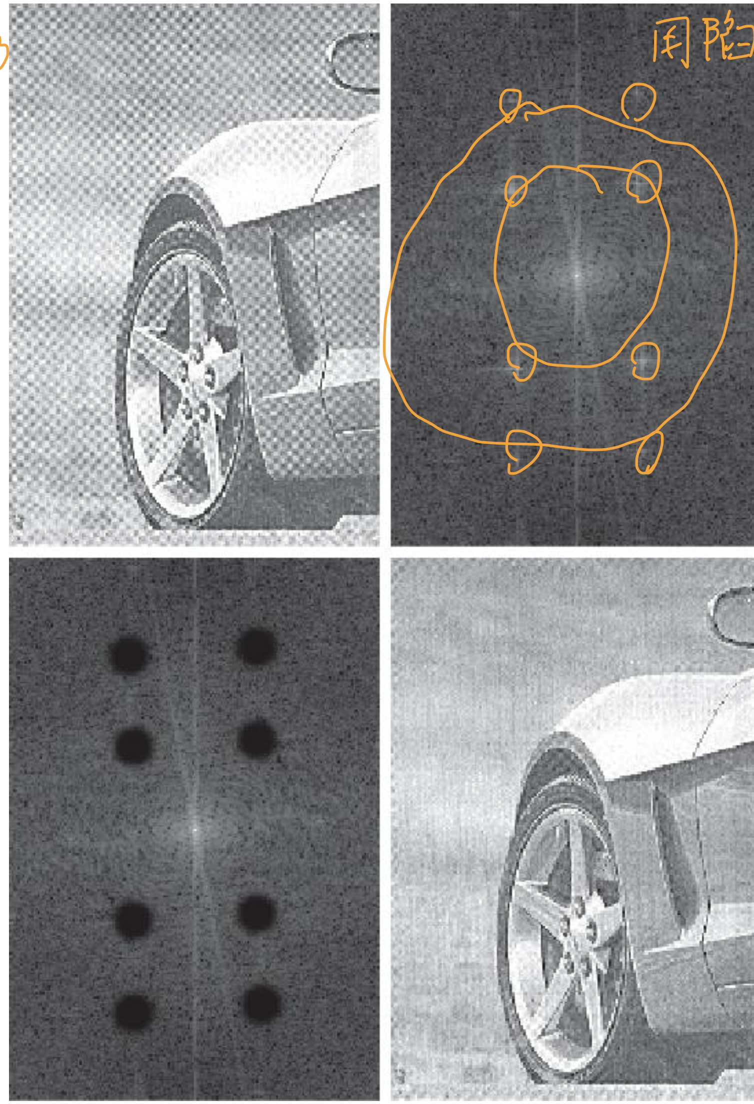
$$H_{\mathrm{NR}}(u,v)=\prod_{k=1}^{Q} H_k(u,v)H_{-k}(u,v)$$

$$H_{\mathrm{NP}}(u,v)=1-H_{\mathrm{NR}}(u,v)$$

Where $H_k(u,v)$, $H_{-k}(u,v)$ are Highpass filters with center at (u_k, v_k) and (u_{-k}, v_{-k})

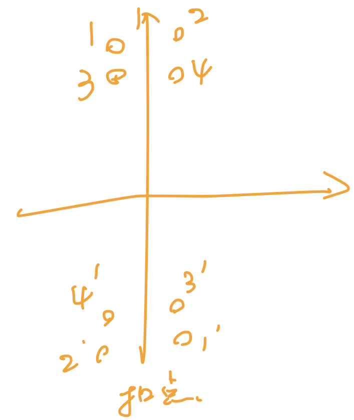
Notch Filter (陷波滤波器)

空间伪影



用陷波(因为分布广)

带阻的环



a	b
c	d

FIGURE 4.64 (a) Sampled newspaper image showing a moiré pattern. (b) Spectrum. (c) Fourier transform multiplied by a Butterworth notch reject filter transfer function. (d) Filtered image.

Take home message

- Zero-padding in spatial domain is necessary for frequency domain filtering
- Know the procedure of frequency domain filtering
- Frequency domain filtering and spatial domain filtering are related
- Know typical LP/HP filters
- 'Ideal filters' are not ideal
- Homomorphic Filtering is able to reduce abnormal illumination effect in image.