

4.3

$$\epsilon_s = 11.7 \times 8.854 \times 10^{-12} \text{ F/m}$$

$$N_B = 10^{15} \text{ cm}^{-3} = 10^{22} \text{ m}^{-3}$$

$$X_d \approx \sqrt{\frac{2\epsilon_s}{qN_B} (0.75\text{V})} = 311 \text{ nm}$$

$$C_s = \frac{\epsilon_s}{X_d} = 3.33 \times 10^{-4} \text{ F/m}^2$$

4.9

- For $V_{gs} = 0\text{V}$, $V_{gs} < V_{TN}$, so the MOSFET is in cutoff. Therefore $i_D \approx 0\text{A}$.

For $V_{gs} = 1\text{V}$, device in linear (triode) operation.

- for $V_{gs} = 1\text{V}$, $V_{gs} - V_{TN} = 0.2\text{V} < V_{DS} = 0.25\text{V}$

Transistor is in saturation.

$$i_D = \frac{K_n}{2} (V_{gs} - V_{TN})^2$$

$$K_n = K'_n \left(\frac{W}{L}\right) = 200 \times 10^{-6} \left(\frac{5\mu}{0.5\mu}\right) = 2 \times 10^{-3} \text{ A/V}^2$$

$$i_D = \frac{2 \times 10^{-3} \text{ A/V}^2}{2} (0.2)^2 = 40 \mu\text{A}$$

- for $V_{gs} = 2\text{V}, 3\text{V}$, $V_{DS} < V_{gs} - V_{TN}$, therefore MOSFET is in linear (triode)

$$i_D = K_n (V_{gs} - V_{TN} - V_{DS}/2) V_{DS}$$

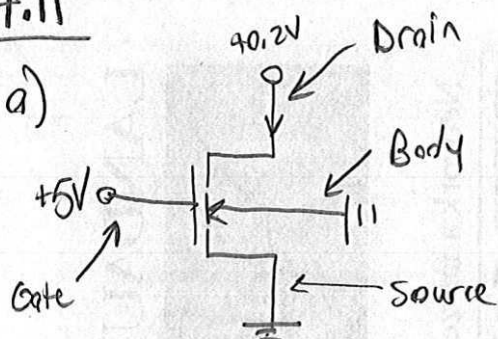
$$i_D = 2 \times 10^{-3} \text{ A/V}^2 (V_{gs} - 0.8\text{V} - \frac{0.25\text{V}}{2}) (0.25\text{V})$$

$$i_D = 537 \mu\text{A} \text{ for } V_{gs} = 2\text{V}$$

$$i_D = 1.04 \text{ mA} \text{ for } V_{gs} = 3\text{V}$$

4.11

a)



The source terminal is the one shorted to the body.

$$V_{ds} = 0.2V < V_{gs} - V_t = 4.3V, \text{ so linear operation}$$

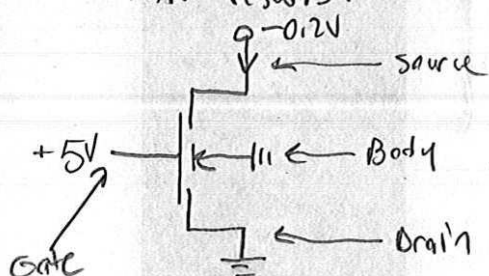
$$i_D = K_n \left(V_{gs} - V_t - \frac{V_{ds}}{2} \right) V_{ds}$$

$$= K_n \left(4.3V - \frac{0.2V}{2} \right) (0.2V)$$

$$i_D = (0.84) K_n = (0.84) (100 \times 10^{-6} \text{ A/V}^2) \left(\frac{10}{1} \right) = \boxed{840 \mu\text{A}}$$

K_n from table $\frac{W}{L}$

b) Because there is a negative voltage on the top terminal, the source and drain are reversed. there is also a body bias that results.



$$V_{gs} = 5V - (-0.2V) = 5.2V$$

$$V_{sb} = -0.2V - 0.2V = -0.4V$$

$$V_{ds} = 0V - (-0.2V) = 0.2V$$

$$V_{TN} = V_{T0} + \gamma \left(\sqrt{V_{sb} + 2\phi_F} - \sqrt{2\phi_F} \right)$$

$$V_{TN} = (0.7V) + (0.75) (\sqrt{-0.2 + 0.6} - \sqrt{0.6})$$

$$= 0.593V$$

Transistor is in linear region

$$i_D = K_n \left(V_{gs} - V_{TN} - \frac{V_{ds}}{2} \right) V_{ds}$$

$$= (100 \times 10^{-6} \text{ A/V}^2) \left(\frac{10}{1} \right) \left(5.2V - 0.593V - \frac{0.2V}{2} \right) (0.2V)$$

$$= \boxed{901 \mu\text{A}}$$

4.10

$$a) i_D = K_n (V_{GS} - V_{TN} - \frac{V_{DS}}{2}) V_{DS}$$

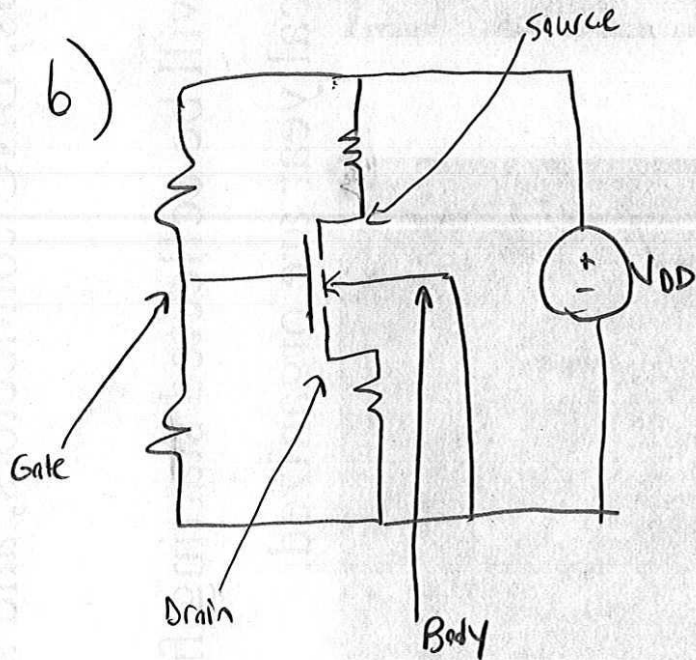
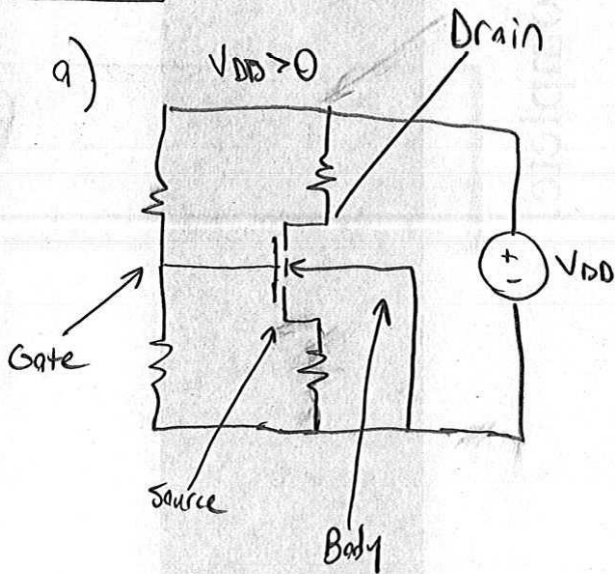
$$\frac{1}{R_{on}} = \frac{i_D}{V_D} = K_n (V_{GS} - V_{TN} - \frac{V_{DS}}{2}) \approx K_n (V_{GS} - V_{TN}) \text{ for } V_{DS} \rightarrow 0V$$

$$\frac{1}{R_{on}} = K_n' \left(\frac{W}{L}\right) (V_{GS} - V_{TN})$$

$$\left(\frac{W}{L}\right) = \frac{R_{on}}{K_n' (V_{GS} - V_{TN})} = \frac{500 \Omega}{100 \mu A/V^2 (5V - 0.75V)} = 4.7$$

$$b) \left(\frac{W}{L}\right) = \frac{1}{(500 \Omega) (100 \times 10^{-6} A/V^2) (3.3V - 0.75V)} = 7.8$$

4.25



c) If R_1, R_2, R_3 , and R_4 are chosen such that the transistor is on in part a, then most likely it will be off in part b because the gate will no longer be biased so that the transistor is on.

4.34

- a) • For both transistors, $V_{GS} = V_{DS}$, therefore the transistors are in saturation for any V_{GS} , V_{DS} .
- Also, because the transistors are identical and current through them the same, the V_{DS} must be the same. Therefore the $V_{DS} = \frac{1}{2} V_{DD}$ for both transistors.

$$a) \quad i_D = \frac{K_n}{2} (V_{GS} - V_{TN})^2 (1 + \lambda V_{DS})$$

$$i_D = \frac{1}{2} (100 \times 10^{-6} \text{ A/V}^2) (10) \left(\frac{1}{2} (10) - 0.75 \right)^2$$

$$\boxed{i_D = 2.13 \text{ mA}}$$

- b) The current is simply twice of part A

$$i_D = \frac{1}{2} (100 \times 10^{-6} \text{ A/V}^2) (20) \left(\frac{1}{2} (10) - 0.75 \right)^2$$

$$\boxed{i_D = 4.25 \text{ mA}}$$

$$c) \quad i_D = \frac{1}{2} (100 \times 10^{-6} \text{ A/V}^2) (10) \left(\frac{1}{2} (10) - 0.75 \right)^2 \left(1 + (0.04) \left(\frac{1}{2} (10) \right) \right)$$

$$\boxed{i_D = 2.55 \text{ mA}}$$

4.49

Saturation begins when $V_{ds} = V_{gs} - V_T$. Inspecting the figure, we see that saturation begins at the following V_{ds} , V_{gs} values

V_{gs}	$V_{ds}(\text{sat})$
-2	-1.5
-3	-2.5
-4	-3.5
-5	-4.5

Therefore $V_{Tp} = -0.5V$.

Because threshold voltage less than zero, the device is enhancement mode

To estimate K_p , we see $i_D \approx 4000 \mu A$ for $V_{gs} = -5V$ in saturation.

$$i_D \approx \frac{1}{2} K_p (V_{gs} - V_{Tp})^2$$

$$K_p \approx \frac{2i_D}{(V_{gs} - V_{Tp})^2} = \frac{2(4000 \mu A)}{(-5V - (-0.5V))^2} = 395 \mu A/V^2$$

using K_p' from table 4.6

$$\frac{W}{L} = \frac{K_p}{K_p'} = \frac{395 \mu A/V^2}{40 \mu A/V^2} = 9.88$$