

## EE5801 Tutorial 4 Solutions

Q1

$$\epsilon_r = 2.9$$

$$Z_0 = 75 \Omega$$

$$Z_L = 300 \Omega$$

$$l = 38 \text{ mm} = 0.038 \text{ m}$$

$$\text{since } u_p = \frac{\omega}{k} = \frac{c}{\sqrt{\epsilon_r}}, \text{ we have}$$

$$k = \frac{\omega}{c} \sqrt{\epsilon_r} = 71.33 \text{ rad/m}$$

$$Z_{in} = Z_0 \frac{Z_L + jZ_0 \tan(kl)}{Z_0 + jZ_L \tan(kl)} = 82.90 + j118.01 \Omega.$$

Q2

From page 5 of lecture notes: characteristic impedance  $= \sqrt{L/C} = (h/w)120\pi$  for air-spaced parallel plate. Thus,  $Z_0 = (1\text{cm}/2\text{cm})120\pi = 60\pi$

From page 8 of lecture note, assuming lossless (although there is a resistance due to skin depth  $R = L/(\sigma A) = 20\text{cm}/(\sigma w \delta)$ , where skin depth  $\delta = 1/\sqrt{\pi f \mu \sigma} = 2.3\mu\text{m}$  and thus  $R = 20\text{cm}/(\sigma x 2\text{cm} x 2.3\mu\text{m}) = 0.073\Omega$  or  $0.30 \Omega/\text{m}$

When the ground is a perfect short ( $0\Omega$ ), the input impedance is  $Z_{in} = -jZ_0 \tan(kz)$

Thus,  $Z_{in} = -j60\pi \tan(16.755 \times 0.2) = -j60\pi \times 0.212 = -j40\Omega$

If the ground impedance is a typical  $5\Omega$ , then the input impedance is

$$Z_{in} = Z_0 \frac{Z_L + jZ_0 \tan(kz)}{Z_0 + jZ_L \tan(kz)} = 60\pi \frac{5 + j60\pi \times 0.212}{60\pi + j5 \times 0.212} = 60\pi \frac{5 + j40}{60\pi + j1.06}$$

$$Z_{in} = 60\pi \frac{40.31e^{82.87^\circ}}{188.5e^{0.32^\circ}} = 40.3e^{82.55^\circ} = 5.23 + j40$$

Q3

$$Z_{in2} = Z_{02} \frac{Z_{L2} + jZ_{02} \tan(2\pi 3/8)}{Z_{02} + jZ_{L2} \tan(2\pi 3/8)} = 118.042 - j92.553 \Omega$$

$$Z_{in3} = Z_{03} \frac{Z_{L3} + jZ_{03} \tan(2\pi 3/8)}{Z_{03} + jZ_{L3} \tan(2\pi 3/8)} = 95.244 - j30.472 \Omega$$

Total impedance as seen from  $BB'$ :

$$Z_{BB'} = Z_{in2} // Z_{in3} = 54.818 - j26.575 \Omega$$

Total impedance as seen from  $AA'$ :

$$Z_{AA'} = Z_{01} \frac{Z_{BB'} + jZ_{01} \tan(2\pi 3/8)}{Z_{01} + jZ_{BB'} \tan(2\pi 3/8)} = 126.884 - j332.878 \Omega$$

$$\text{Total current at } AA': I_{AA'} = \frac{100 \angle 0^\circ}{300 + Z_{AA'}} = 0.1456 + j0.1136 = 0.1847 \angle 0.6623 \text{ A}$$

Average power supplied to transmission line at  $AA'$ :

$$P_{AA'} = \frac{1}{2} |I_{AA'}|^2 \text{Re}[Z_{AA'}] = \frac{1}{2} (0.1847)^2 (126.884) = 2.1643 \text{ W}.$$

Since transmission line 1 is lossless, the average power supplied to the parallel transmission lines at  $BB'$  is  $P_{BB'} = P_{AA'} = 2.1643 \text{ W}$ . The total average power supplied to the two antennas must therefore be equal to  $P_{BB'}$ , since transmission lines 2 and 3 are lossless.

But

$$P_{BB'} = \frac{1}{2} \text{Re}[V_{BB'} I_{BB'}^*] = \frac{1}{2} \text{Re}[V_{BB'} (V_{BB'}^* / Z_{BB'}^*)] = \frac{1}{2} |V_{BB'}|^2 \text{Re}[1 / Z_{BB'}^*] = \frac{1}{2} |V_{BB'}|^2 (0.01477) \\ \rightarrow |V_{BB'}|^2 = 293.067$$

$$P_{L2} = \frac{1}{2} |V_{BB'}|^2 \text{Re}[1 / Z_{in2}^*] = 0.7688 \text{ W}$$

$$P_{L3} = \frac{1}{2} |V_{BB'}|^2 \text{Re}[1 / Z_{in3}^*] = 1.3956 \text{ W}$$

Note that  $P_{L2} + P_{L3} = P_{BB'} = 2.1643 \text{ W}$

Q4

Input impedance of the antenna  $Z_{in} = 35 + j10 \Omega$

Normalized input impedance is  $z_{in} = \frac{Z_{in}}{Z_0} = \frac{35 + j10}{50} = 0.7 + j0.2$

Normalized input admittance is  $y_{in} = \frac{1}{z_{in}} = 1.32 - j0.38$

Steps:

1. Enter  $y_{in}$  into the Smith chart as shown. Note that  $y_{in}$  is diagonally opposite  $z_{in}$ .
2. Turn clockwise (towards generator) to point  $A$  which gives the first solution. Point  $A$ 's normalized admittance is  $1.0 - j0.44$ . The distance traveled from  $y_{in}$  to point  $A$  measuring on the Smith chart is  $d_1$  which is

$$d_1 = (0.3585 - 0.3045)\lambda = 0.054\lambda$$

This is the position of the stub from the load.

3. To cancel the reactive part of  $-j0.44$ , an open-circuit parallel stub, whose normalized admittance at the open end is at point  $S$  on the Smith chart, must have

a length  $\ell_1$  measuring (towards generator) from point  $S$  to point  $S_A$  on the Smith chart in order to give a normalized admittance of  $j0.44$ . That is,

$$\ell_1 = 0.0660\lambda$$

This is the length of the stub.

4. There is another possible solution. This is achieved by turning (towards generator) from  $y_{in}$  to point  $B$  on the Smith chart. Point  $B$ 's normalized admittance is  $1.0 + j0.44$ . The distance traveled from  $y_{in}$  to point  $B$  measuring on the Smith chart is  $d_2$  which is

$$d_2 = (0.500 - 0.3045)\lambda + 0.1415\lambda = 0.337\lambda$$

This is the position of the stub from the load for the second solution.

5. To cancel the reactive part of  $j0.44$ , an open-circuit parallel stub, whose normalized admittance at the open end is at point  $S$  on the Smith chart, must have a length  $\ell_2$  measuring (towards generator) from point  $S$  to point  $S_B$  on the Smith chart in order to give a normalized admittance of  $-j0.44$ . That is,

$$\ell_2 = 0.434\lambda$$

This is the length of the stub for the second solution.

