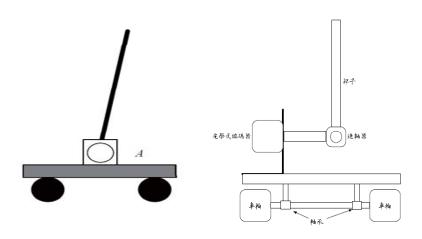
Inverted Pendulum Car

電機工程學系 鄭智湧

正面側面示意圖





車與桿倒單擺系統機構

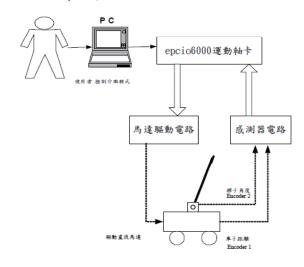
系統的機械結構由直流馬達、齒輪、車輪、 編碼器組成車子的部分; 角度編碼器、桿子 組合成倒單擺部份







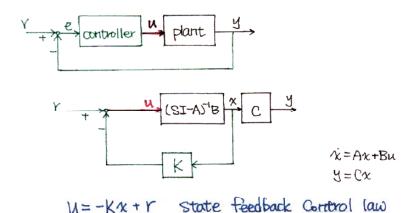
系統結構圖



State Feedback Controller Design



State Feedback Design





Eigenvalue (pole) Placement

```
Thm:

consider the system \dot{x} = Ax + bu with the state feedback U = -Kx

If the system is controllable then eigenvalues of A-bk can be arbitrarily moved around the complex plane (with complex conjugate pairs)
```

```
M
```

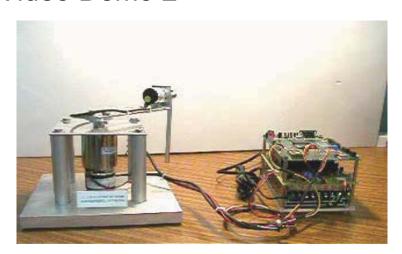
```
interrupt void eva_timer1_isr(void)
   QEP1 = EvaRegs.T2CNT;
                                   // counting value of pendulum encoder
   QEP2 = EvbReqs.T3CNT;
                                   // counting value of car encoder
   arm = QEP1*0.0031416;
                           // theta=QEP2*(2*pi/2000) rad
       arm_dot = ((arm-arm_old)/0.001);
                                              // sampling time=0.001s
       arm old = arm;
   pend dot = ((pend-pend old)/0.001);
                                             // sampling time=0.001s
      pend old = pend;
       U = balance LQR();
 // state feedback control law for balancing control
 float balance LQR(void)
  law = -((k1*arm)+(k2*pend)+(k3*arm_dot)+(k4*pend_dot));
  return law;
```

Video Demo 1





Video Demo 2



Simulation Study



IP Modeling

For pendulum,

where

and

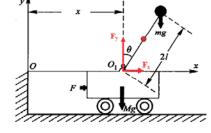
For cart,

$$F_x = m\frac{d^2}{dt^2}(x + l\sin\theta)$$

 $J\ddot{\theta} = F_y l \sin \theta - F_x l \cos \theta$

$$F_y - mg = m\frac{d^2}{dt^2}(l\cos\theta)$$

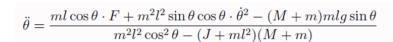
$$F - F_x = M \frac{d^2x}{dt^2}$$



$$(J+ml^2)\ddot{\theta} + ml\cos\theta \cdot \ddot{x} = mlg\sin\theta$$

$$(M+m)\ddot{x} + ml(\cos\theta \cdot \ddot{\theta} - \sin\theta \cdot \dot{\theta}^2) = F$$

Note: Lagrange dynamics is another approach for modeling.



$$\ddot{x} = \frac{(J+ml^2)F + lm(J+ml^2)\sin\theta \cdot \dot{\theta}^2 - m^2l^2g\sin\theta\cos\theta}{(J+ml^2)(M+m) - m^2l^2\cos^2\theta}$$

Linearization $\dot{\theta}^2 \approx 0, \sin \theta \approx \theta, \cos \theta \approx 1$

$$\ddot{\theta} = \frac{(M+m)mlg\theta - mlF}{J(M+m) + mMl^2}$$

$$\ddot{x} = \frac{(J+ml^2)F - m^2l^2g\theta}{J(M+m) + mMl^2}$$



Assume
$$M=1, m=1, 2l=0.6, g=10, J=\frac{ml^2}{3}=0.03$$

$$\begin{split} \ddot{\theta} &= \frac{0.3\cos\theta\cdot F + 0.09\sin\theta\cdot\cos\theta\cdot\dot{\theta}^2 - 6\sin\theta}{0.09\cos^2\theta - 0.24} \end{split}$$
 Model (NL)
$$\ddot{x} &= \frac{0.12F + 0.036\sin\theta\cdot\dot{\theta}^2 - 0.9\sin\theta\cdot\cos\theta}{0.24 - 0.09\cos^2\theta} \end{split}$$

$$\ddot{\theta} = 40\theta - 2F$$
 Model (L)
$$\ddot{x} = -6\theta + 0.8F$$



State space model

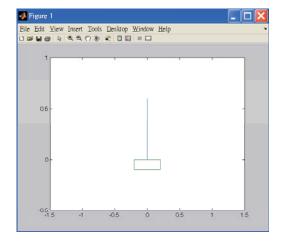
$$x_1 = \theta, x_2 = \dot{\theta}, x_3 = x, x_4 = \dot{x}$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \\ \dot{x}_4 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 40 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 \\ -6 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} + \begin{bmatrix} 0 \\ -2 \\ 0 \\ 0.8 \end{bmatrix} F$$

Check controllability (under-actuated or non-holonomic)



IP graphics test





test_IP_gra.m

- length=0.6; len=0.2; width=0.1;
- x_pos=0; angle=pi/3;
- N=100;
- for i=1:N
- x pos=x pos+1/N;
- angle=angle+1/N;
- stick_x=[x_pos, x_pos+length*cos(angle)];
- stick_y=[0 , length*sin(angle)];
- car_x=[x_pos-len x_pos+len x_pos-len x_pos-len];
- car_y=[0 0 -width -width 0];
- plot(stick_x,stick_y,car_x,car_y); axis([-1 1 -0.5 1]);
- MA(:,i)=getframe;
- end



ODE simulation

$$\dot{x} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 40 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 \\ -6 & 0 & 0 & 0 \end{bmatrix} x + \begin{bmatrix} 0 \\ -2 \\ 0 \\ 0.8 \end{bmatrix} u$$

function dx=IP(t,x)

% state feedback design

$$u = -[-95.4861 - 22.3475 - 7.0711 - 12.5201] x;$$

$$dx = [0 1 0 0]$$

40 0 0 0

0001

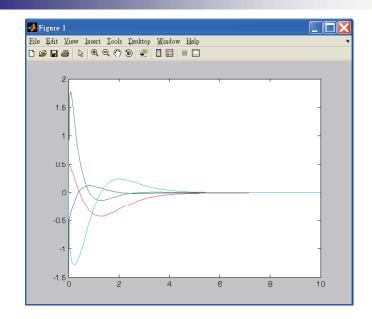
-6 0 0 0]*x+[0 -2 0 0.8]'*u;



IPcar.m

- length=0.6; len=0.2; width=0.1;
- [t,x]=ode23(@IP,[0 10],[-0.5 0 0.5 0]);
- N=max(size(x));
- for i=1:N
- angle=pi/2-x(i,1);
- x pos=x(i,3);
- stick_x=[x_pos, x_pos+length*cos(angle)];
- stick_y=[0 , length*sin(angle)];
- car_x=[x_pos-len x_pos+len x_pos+len x_pos-len x_pos-len];
- car_y=[0 0 -width -width 0];
- plot(stick_x,stick_y,car_x,car_y); axis([-1.5 1.5 -0.5 1]);
- MA(:,i)=getframe;
- end







Acker in Matlab

% Use the example discussed in class.

% The desired eigenvalues are at -2, -1, -1 A=[1 0 0; -1 0 2; 0 -1 1]; B=[1;0;0]; J=[-2 -1 -1]; K=acker(A,B,J)

% The result is K = [6 -8 0]

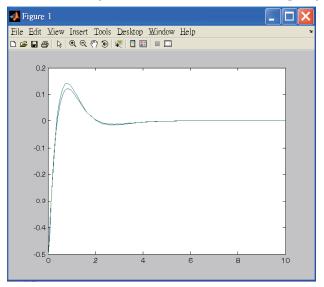


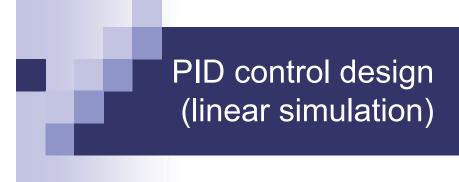
IPcar_NL and IP_NL

- function dx=IP_NL(t,x)
- % LQR design
- u=-[-95.4861 -22.3475 -7.0711 -12.5201]*x;
- dx=zeros(4,1);
- = dx(1)=x(2);
- $dx(2) = (0.3*\cos(x(1))*u+0.09*\sin(x(1))*\cos(x(1))*x(1)^2-6*\sin(x(1))) / (0.09*\cos(x(1))^2-0.24);$
- = dx(3)=x(4);
- $dx(4) = (0.12*u + 0.036*sin(x(1))*x(1)^2 0.9*sin(x(1))*cos(x(1)))/(0.24-0.09*cos(x(1))^2);$



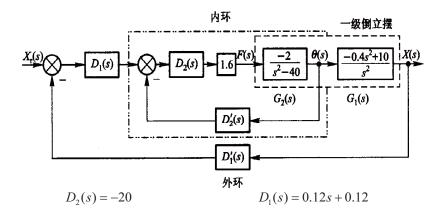
Comparion (NL could diverge)







PID control of angle and position



 $D_1'(s) = 1$



IPpid.m

- G2= tf([-2],[1 0 -40]);
- D2= tf([-32],[1]);
- D2p= tf([0.175 1.625],[1]);
- sys_inner= feedback(series(D2,G2),D2p);
- G1= tf([-0.4 0 10],[1 0 0]);
- D1= tf([0.12 0.12],[1]);
- D1p= tf([1],[1]);
- sys_outer= feedback(series(series(D1,sys_inner),G1),D1p);
- sys_angle= feedback(series(D1,sys_inner),series(G1,D1p));



- t=0:0.1:10;
- pos=0.1*impulse(sys_outer,t);

 $D_2'(s) = 0.175 \,\mathrm{s} + 1.625$

- the=0.1*impulse(sys angle,t);
- % plot(t,the,t,pos)
- length=0.6; len=0.2; width=0.1;
- N=max(size(the));
- for i=1:N
- angle=pi/2-the(i);
- x_pos=pos(i);
- stick_x=[x_pos, x_pos+length*cos(angle)];
- stick_y=[0 , length*sin(angle)];
- car_x=[x_pos-len x_pos+len x_pos+len x_pos-len x_pos-len];
- car_y=[0 0 -width -width 0]
- plot(stick_x,stick_y,car_x,car_y); axis([-1.5 1.5 -0.5 1]);
- MA(:,i)=getframe;
- end



Pendulum angle and cart position

