

ELEC 3004/7312: Digital Linear Systems: Signals & Control

## Tutorial 5 (Week 10): System Poles Zeros and PID Control

By: Ye Tian

---

### Review: System Poles and Zeros

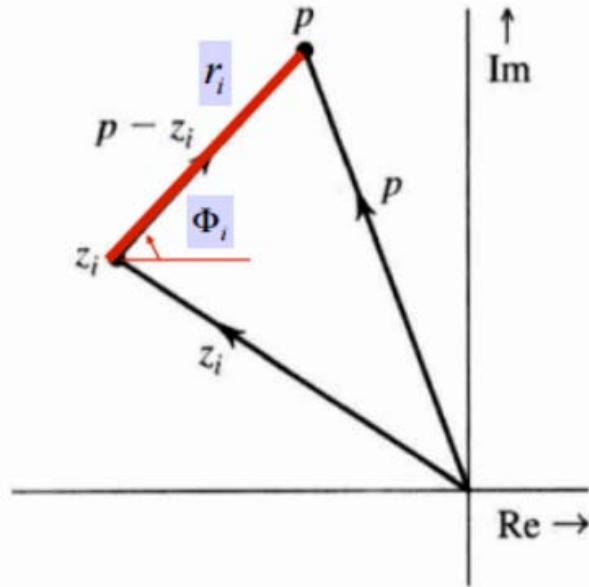
A general system transfer function can be expressed in a poles-zeros form:

$$G(s) = \frac{B(s)}{A(s)} = k \frac{(s - z_1)(s - z_2) \cdots (s - z_m)}{(s - \lambda_1)(s - \lambda_2) \cdots (s - \lambda_n)}$$

The value of the transfer function at a complex frequency  $s = p$  is:

$$G(s)|_{s=p} = k \frac{(p - z_1)(p - z_2) \cdots (p - z_m)}{(p - \lambda_1)(p - \lambda_2) \cdots (p - \lambda_n)}$$

Take the  $i$  th term of the numerator  $p - z_i$  for example, it is the vector subtraction between vector  $p$  and vector  $z_i$ . This vector  $p - z_i$  can be illustrated graphically as:



Given the vector length  $r_i$  and the angle  $\Phi_i$ , the expression in polar coordinates is:

$$p - z_i = r_i e^{j\Phi_i}.$$

The transfer function thus can be written as:

$$G(s)|_{s=p} = k \frac{(r_1 e^{j\Phi_1})(r_2 e^{j\Phi_2}) \cdots (r_m e^{j\Phi_m})}{(d_1 e^{j\theta_1})(d_2 e^{j\theta_2}) \cdots (d_n e^{j\theta_n})}$$

Therefore, when the complex frequency  $s$  is close to one of the zeros  $z_i$ , the amplitude of  $G$  is a small value due to the small  $r_i$  value; when the complex frequency  $s$  is close to one of the poles, the amplitude of  $G$  is a large value (tends to infinity) due to a  $d_j$  value which tends to zero.

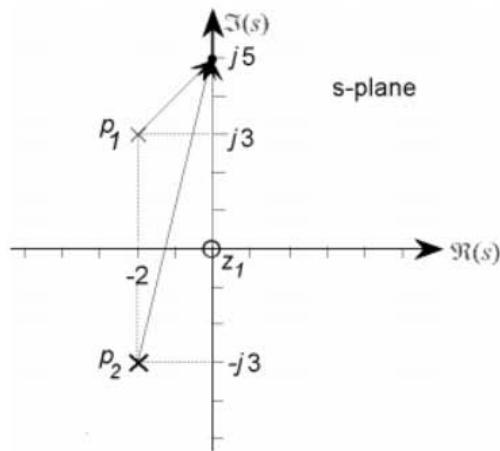
Exercise 1: A second-order system has a pair of complex conjugate poles at  $s = -2 \pm j3$  and a single zero at the origin of the  $s$ -plane. Find the transfer function and use the pole-zero plot to evaluate the transfer function at  $s = 0 + j5$ . (The phase response can be ignored in this question)

**Solution:** From the problem description

$$\begin{aligned} H(s) &= K \frac{s}{(s - (-2 + j3))(s - (-2 - j3))} \\ &= K \frac{s}{s^2 + 4s + 13} \end{aligned} \quad (27)$$

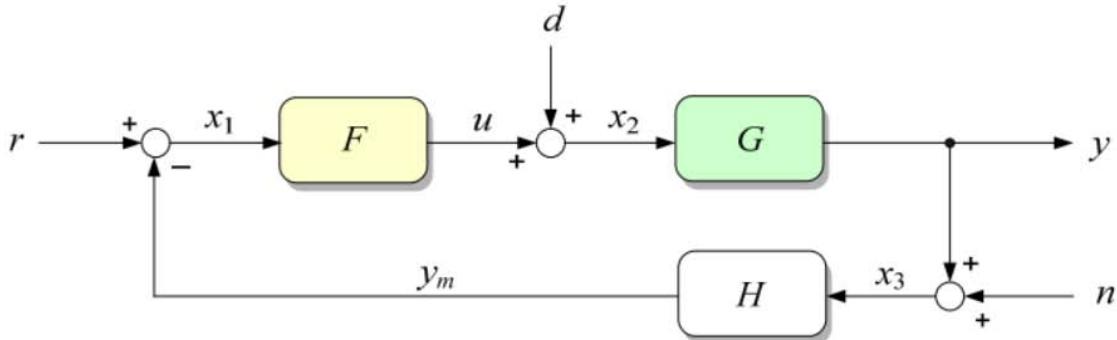
The pole-zero plot is shown in Fig. 6. From the figure the transfer function is

$$\begin{aligned} |H(s)| &= K \frac{\sqrt{(0-5)^2}}{\sqrt{(0-(-2))^2 + (5-3)^2} \sqrt{(0-(-2))^2 + (5-(-3))^2}} \\ &= K \frac{5}{4\sqrt{34}} \end{aligned} \quad (28)$$



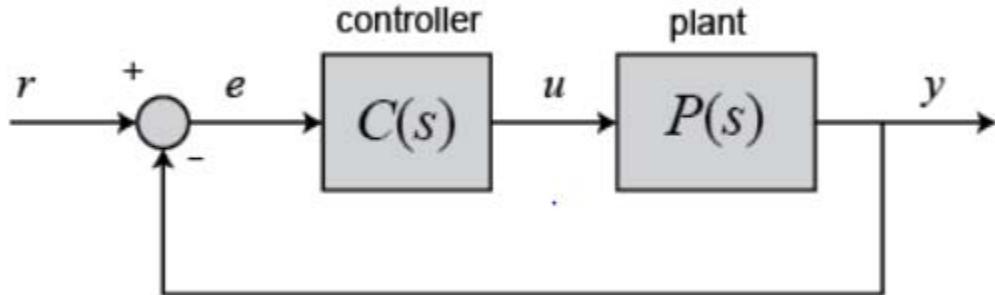
## Basic Feedback Equations

Exercise 2: Write the expression of  $y$  using  $r$ ,  $d$  and  $n$ . (Hint: It may be helpful to define intermediate variables,  $x_{1,2,3}$  in this case, for the output of the nodes with signal addition/subtraction)



## A short PID control review and a simple exercise

For a unity feedback system:



The output of a PID controller in the time domain is as follows:

$$u(t) = K_p e(t) + K_i \int e(t) dt + K_d \frac{de}{dt}$$

The variable ( $e$ ) represents the tracking error, the difference between the desired input value ( $r$ ) and the actual output ( $y$ ). This error signal ( $e$ ) will be sent to the PID controller, and the controller computes both the derivative and the integral of this error signal. The control signal ( $u$ ) to the plant is equal to the proportional gain ( $K_p$ ) times the magnitude of the error plus the integral gain ( $K_i$ ) times the integral of the error plus the derivative gain ( $K_d$ ) times the derivative of the error.

The transfer function of a PID controller is

$$K_p + \frac{K_i}{s} + K_d s = \frac{K_d s^2 + K_p s + K_i}{s}$$

$K_p$  = Proportional gain  $K_i$  = Integral gain  $K_d$  = Derivative gain.

## The Characteristics of P I D Controllers

---

A proportional controller ( $K_p$ ) will have the effect of reducing the rise time and will reduce but never eliminate the steady-state error. An integral control ( $K_i$ ) will have the effect of eliminating the steady-state error for a constant or step input, but it may make the transient response slower. A derivative control ( $K_d$ ) will have the effect of increasing the stability of the system, reducing the overshoot, and improving the transient response.

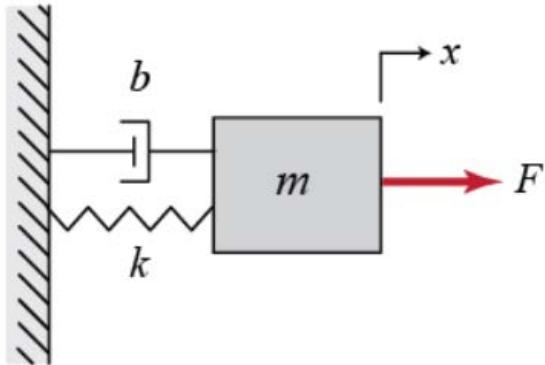
The effects of each of controller parameters,  $K_p$ ,  $K_i$ , and  $K_d$  on a closed-loop system are summarized in the table below.

CLOSE LOOP RESPONSE	RISE TIME	OVERSHOOT	SETTLING TIME	S-S ERROR
$K_p$	Decrease	Increase	Small Change	Decrease
$K_i$	Decrease	Increase	Increase	Eliminate
$K_d$	Small Change	Decrease	Decrease	No Change

Note that these correlations may not be exactly accurate, because  $K_p$ ,  $K_i$ , and  $K_d$  are dependent on each other. In fact, changing one of these variables can change the effect of the other two. For this reason, the table should only be used as a reference when you are determining the values for  $K_i$ ,  $K_p$  and  $K_d$ .

### A simple exercise:

Consider a simple mass-spring-damper system with input F and output x (the displacement of the mass):



1. What is the transfer function between  $X(s)$  and  $F(s)$ ?
2. Let  $m=1\text{kg}$ ,  $b=10\text{Ns/m}$ ,  $k=20\text{N/m}$ , what is the closed-loop transfer function with a proportional controller ( $K_p$ ), with a proportional-integral (PI) controller ( $K_p$ ,  $K_i$ ) and with a proportional-integral-derivative controller ( $K_p$ ,  $K_i$ ,  $K_d$ )?

**Answer to the example above exercise:**

1. The modeling equation of this system is

$$M\ddot{x} + b\dot{x} + kx = F$$

Taking the Laplace transform of the modeling equation, we get

$$Ms^2X(s) + bsX(s) + kX(s) = F(s)$$

The transfer function between the displacement  $X(s)$  and the input  $F(s)$  then becomes

$$\frac{X(s)}{F(s)} = \frac{1}{Ms^2 + bs + k}$$

2. Let

$$M = 1 \text{ kg}$$

$$b = 10 \text{ N s/m}$$

$$k = 20 \text{ N/m}$$

$$F = 1 \text{ N}$$

Plug these values into the above transfer function

$$\frac{X(s)}{F(s)} = \frac{1}{s^2 + 10s + 20}$$

The closed-loop transfer function of the above system with a proportional controller is:

$$\frac{X(s)}{F(s)} = \frac{K_p}{s^2 + 10s + (20 + K_p)}$$

the closed-loop transfer function with a PI control is:

$$\frac{X(s)}{F(s)} = \frac{K_p s + K_i}{s^3 + 10s^2 + (20 + K_p s + K_i)}$$

The closed-loop transfer function of the given system with a PID controller is:

$$\frac{X(s)}{F(s)} = \frac{K_d s^2 + K_p s + K_i}{s^3 + (10 + K_d)s^2 + (20 + K_p)s + K_i}$$