

# Kalman filtering for GNSS positioning and sensor integration: part I

**AAE4203 – Guidance and Navigation**

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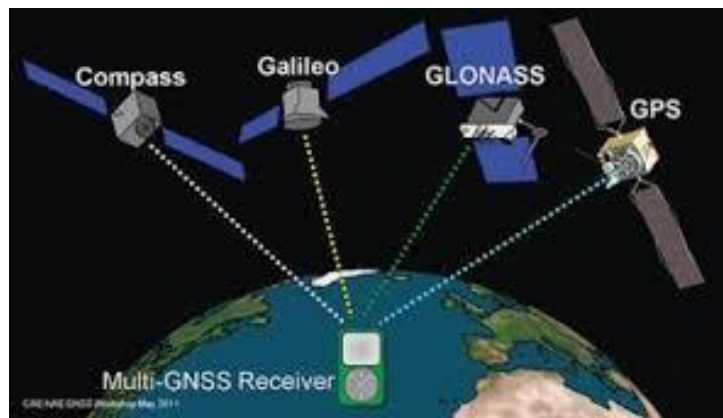
*Week 10, 7 Nov 2022*

# Outline

- > Overview About State Estimation
- > Kalman Filtering
- > GNSS Doppler based Velocity Estimation
- > Principle of Kalman Filtering
- > GNSS Positioning with Kalman Filtering
  - Loosely Coupled GNSS Positioning
  - Tightly Coupled GNSS Positioning

# GNSS Benefits

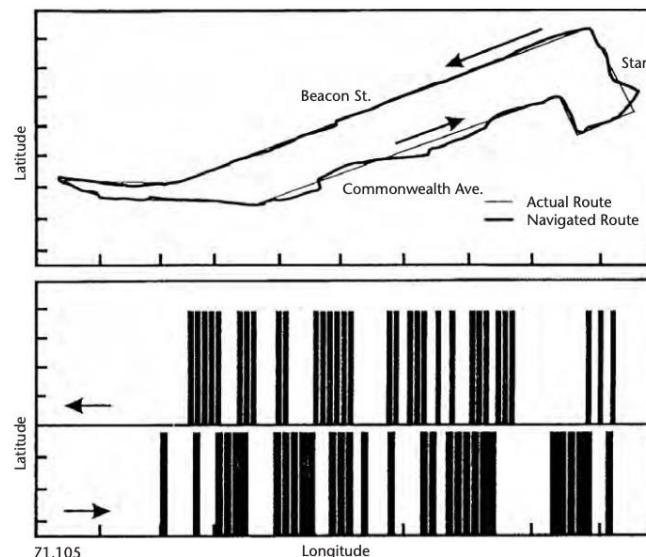
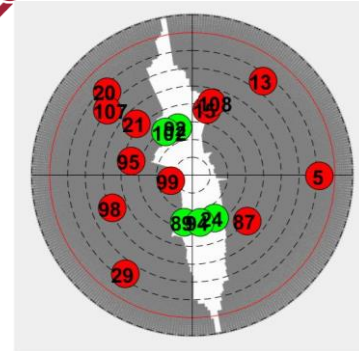
- > GNSS provides a high long-term position accuracy with errors limited to a few meters (stand-alone), while user equipment is available for less than \$100 (€80).



# GPS Drawbacks

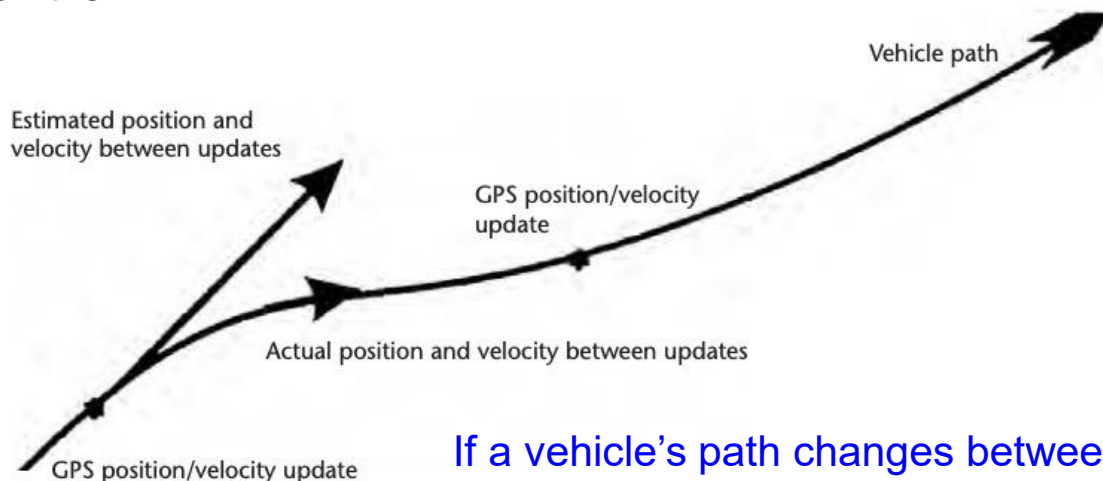
- > One primary concern with using GPS as a stand-alone source for navigation is **signal interruption**.

If the number of usable satellites is less than three, some receivers have the option of not producing a solution or extrapolating the last position and velocity solution forward in what is called *dead-reckoning* (DR) navigation. Inertial navigation systems (INSs) can be used as a flywheel to provide navigation during shading outages.



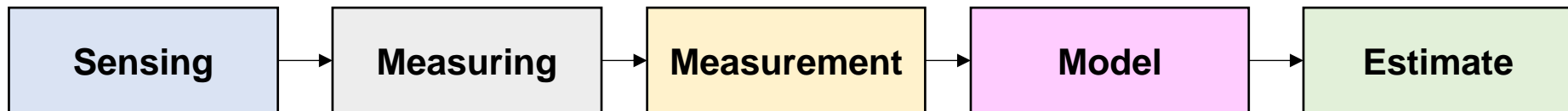
# GPS Drawbacks

- > The **low update rate** of the GPS observations in some equipment is also of concern in real-time applications, especially those related to vehicle control.

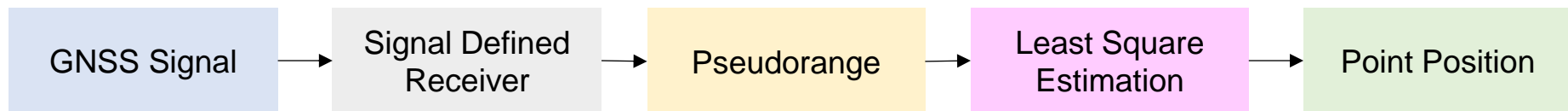


If a vehicle's path changes between updates, the extrapolation of the last GPS measurement produces an error in the estimated and true position.

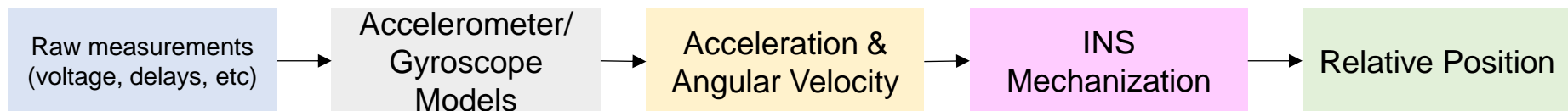
# Framework of Sensors to Navigation



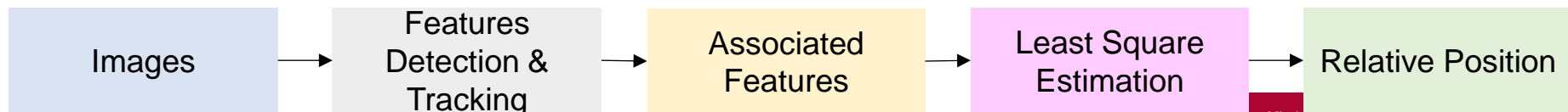
## *GNSS*



## *INS*



## *Visual Navigation*



# Sensor Integration Based on Kalman Filter

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# How the Sensor Fusion Problem Looks like...

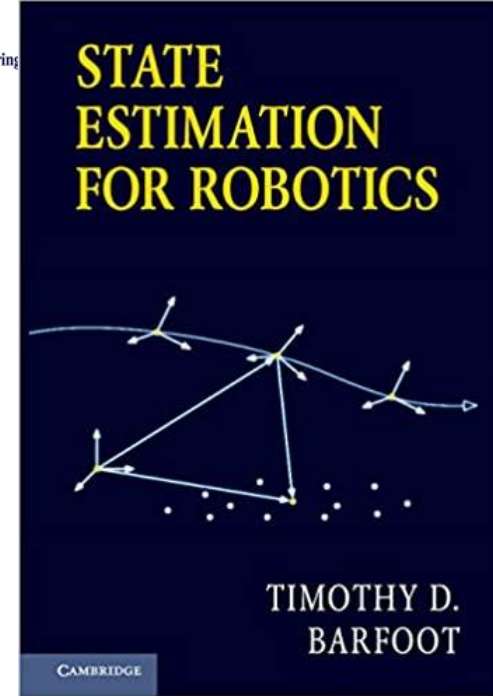
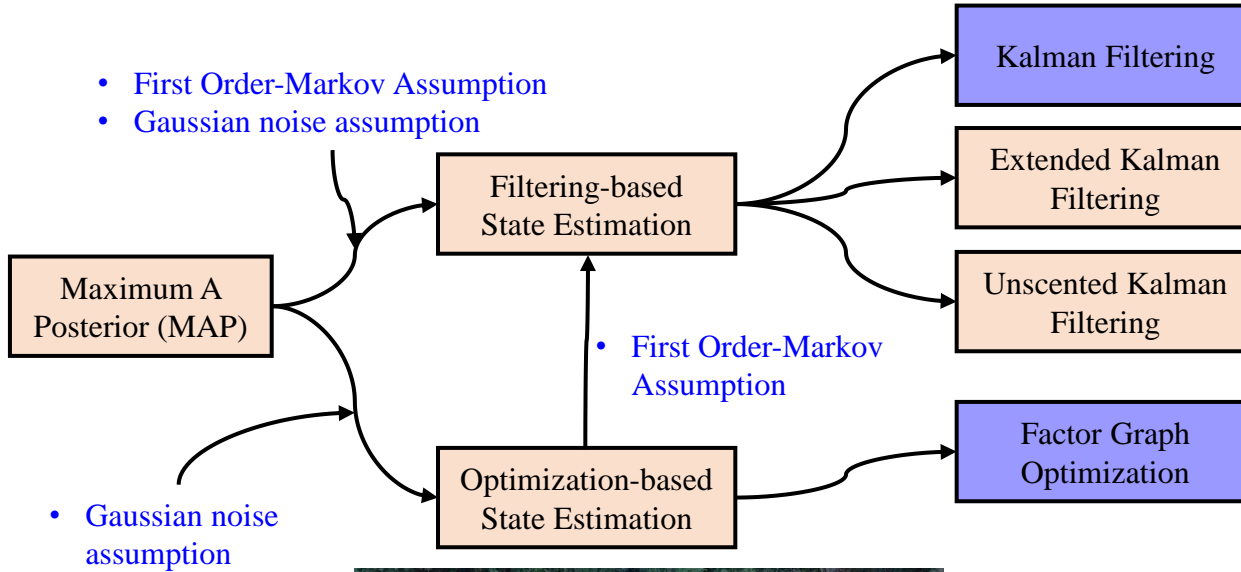
- > GPS provide the position in  $x, y, z$
- > IMU provide the linear angular velocity in  $x, y, z$
- > GPS Doppler provide velocity in  $x, y, z$
- > Visual positioning provide relative motion  $\Delta x, \Delta y, \Delta z$



How to achieve this sensor fusion by  
combining the positioning from different  
sources?



# State Estimation Methods



# Basics for Probabilistic

Event A and B.  $P(A)$  denotes the probability that the event A happens.

$$P(B|A) = \frac{P(AB)}{P(A)} \longrightarrow P(AB) = P(A|B)P(B) \longrightarrow P(B|A) = \frac{P(A|B)P(B)}{P(A)}$$

Event  $z$  denotes the measurements.  $P(z)$  denotes the probability that the event A happens.

$$P(\mathbf{x}_k | \mathbf{z}_1, \dots, \mathbf{z}_k) = \frac{P(\mathbf{z}_0, \dots, \mathbf{z}_k | \mathbf{x}_k) P(\mathbf{x}_k)}{P(\mathbf{z}_0, \dots, \mathbf{z}_k)}$$

The probabilistic view of the state estimation is: given a set of measurements  $(\mathbf{z}_0, \dots, \mathbf{z}_k)$ , can we find a best state  $\mathbf{x}_k$  to maximize the conditional probabilistic  $P(\mathbf{x}_k | \mathbf{z}_1, \dots, \mathbf{z}_k)$ ?

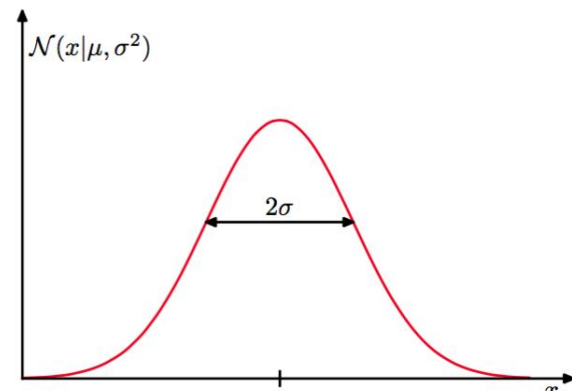
# How do we describe the uncertainty with Gaussian Noise?

> Gaussian distribution

$$p(x) \sim N(\mu, \sigma^2)$$

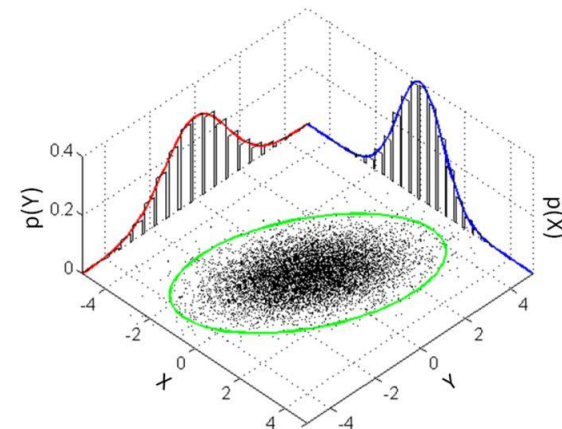
1D (univariate)

$$p(x) = \frac{1}{\sqrt{2\pi}\sigma} e^{-\frac{1}{2} \frac{(x-\mu)^2}{\sigma^2}}$$



2D+ (multi variate)

$$p(\mathbf{x}) = \frac{1}{(2\pi)^{d/2} |\Sigma|^{1/2}} e^{-\frac{1}{2} (\mathbf{x}-\mu)' \Sigma^{-1} (\mathbf{x}-\mu)}$$



# How to understand the $P(\mathbf{z}_0, \dots, \mathbf{z}_k | \mathbf{x}_k)$

Observation function for pseudorange (code) measurement

$$\underbrace{\rho_{r,t}^S}_{\text{Pseudorange}} = \underbrace{r_{r,t}^S}_{\substack{\text{Range} \\ \text{distance}}} + c(\underbrace{\delta_{r,t}}_{\substack{\text{Receiver clock} \\ \text{Bias (1~2m)}}} - \underbrace{\delta_{r,t}^S}_{\substack{\text{Satellite clock} \\ \text{bias}}}) + \underbrace{I_{r,t}^S}_{\substack{\text{ionospheric delay} \\ \text{Distance (1~2m)}}} + \underbrace{T_{r,t}^S}_{\substack{\text{tropospheric delay} \\ \text{Distance (1~2m)}}} + \underbrace{\varepsilon_{r,t}^S}_{\substack{\text{multipath effects, NLOS} \\ \text{receptions, receiver noise,} \\ \text{antenna phase-related noise} \\ \text{(0~100m)}}}$$

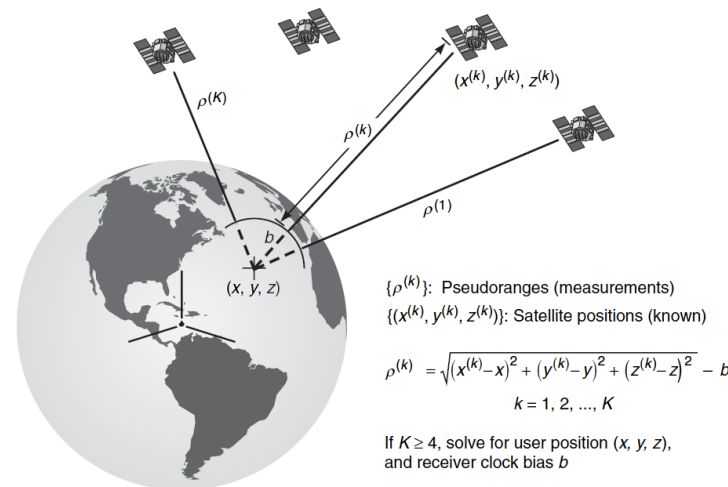
$$\downarrow$$

$$\|\mathbf{p}_t^{G,S} - \mathbf{p}_{r,t}^G\|$$

$$\rho_{r,t}^S \longrightarrow \mathbf{z}_{r,t}^S \quad \mathbf{p}_{r,t}^G \longrightarrow \mathbf{x}_{r,t}^G$$

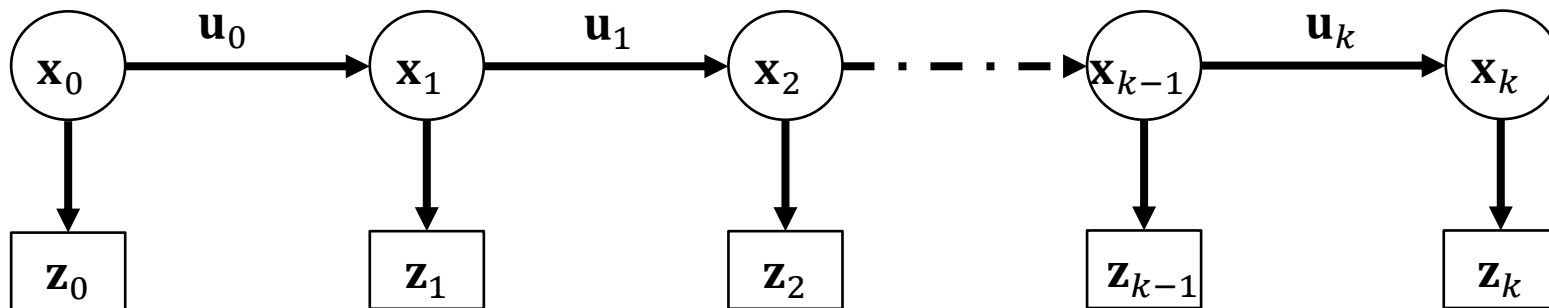
$$P(\mathbf{x}_k | \mathbf{z}_1, \dots, \mathbf{z}_k) = \frac{P(\mathbf{z}_0, \dots, \mathbf{z}_k | \mathbf{x}_k) P(\mathbf{x}_k)}{P(\mathbf{z}_0, \dots, \mathbf{z}_k)}$$

**Formulate the  $P(\mathbf{x}_k | \mathbf{z}_1, \dots, \mathbf{z}_k)$  for the GNSS pseudorange measurements!**



# Estimation Formulation

$\mathbf{u}_i$ : IMU Measurement  
 $\mathbf{z}_i$ : GNSS measurement



**States set**

$$\chi = \{\mathbf{x}_0, \mathbf{x}_1, \mathbf{x}_2, \dots, \mathbf{x}_k\}$$

The **maximum a posteriori (MAP)** estimate is given by

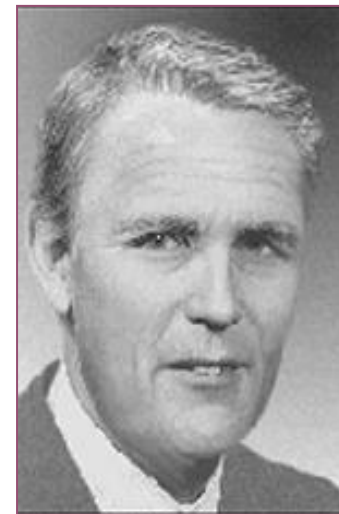
**Optimal State Set**

$$\hat{\chi} = \arg \max_{\chi} (P(\chi | \mathbf{Z}, \mathbf{U})) \quad P(\chi | \mathbf{Z}, \mathbf{U}) = \prod_k P(\mathbf{z}_k | \mathbf{x}_k) P(\mathbf{x}_0) \prod_k P(\mathbf{x}_k | \mathbf{x}_{k-1}, \mathbf{u}_{k-1})$$

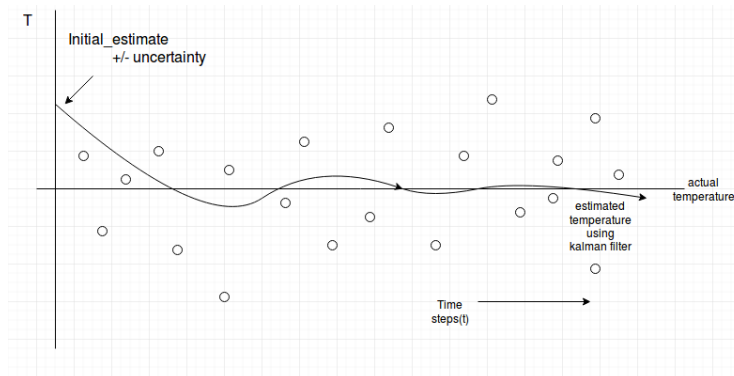
Bayesian theory

# Kalman Filter

- > In early 1960, American engineer R.E. Kalman discovered a linear minimum variance ESTIMATION method——**Kalman Filter**. Soon in space technology (such as flying Navigation system, missile guidance, and determination of satellite orbit and attitude) has been applied.
- > Optimal combination of MEASUREMENT and PROPAGATION



Rudolf Emil Kalman







<http://www.cs.cuhk.edu/~weich/kalman/media/pdf/Kalman960.pdf>

## A New Approach to Linear Filtering and Prediction Problems<sup>1</sup>

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The classical filtering and prediction problem is re-examined using the *Bode-Shannon* representation of random processes and the "state transition" method of analysis of dynamic systems. New results are:

(1) The formulation and methods of solution of the problem without modification to stationary and nonstationary statistics and to growing-memory and infinite-memory filters.

(2) A nonlinear difference (or differential) equation is derived for the covariance matrix of the optimal estimation error. From the solution of this equation the coefficients of the difference (or differential) equation of the optimal linear filter are obtained without further calculations.

(3) The filtering problem is shown to be the dual of the noise-free regulator problem. The new method developed here is applied to two well-known problems, confirming and extending earlier results.

The discussion is largely self-contained and proceeds from first principles; basic concepts of the theory of random processes are reviewed in the Appendix.

## Introduction

AN IMPORTANT class of theoretical and practical problems in communication and control is of a statistical nature. Such problems are: (i) Prediction of random signals; (ii) separation of random signals from random noise; (iii) detection of signals of known form (pulses, sinusoids) in the presence of random noise.

In his pioneering work, Wiener [1] showed that problems (i) and (ii) lead to the so-called Wiener-Hopf integral equation; he also gave a method (spectral factorization) for the solution of this integral equation in the practically important special case of stationary statistics and rational spectral densities.

Many extensions and generalizations followed Wiener's basic work. Zadeh and Ragazzini solved the finite-memory case [2]. Concurrently and independently of Bode and Shannon [3], they also gave a simplified method [2] of solution. Bode discussed the nonstationary Wiener-Hopf equation [4]. These results are now in standard texts [5-6]. A somewhat different approach along these main lines has been given recently by Darlington [7]. For extensions to sampled signals, see, e.g., Franklin [8], Leech [9]. Another approach based on the eigenfunctions of the Wiener-Hopf equation (which applies also to nonstationary problems whereas the preceding methods in general do not) has been pioneered by Davis [10] and applied by many others, e.g., Shinozaki [11], Blum [12], Pugsachev [13], Solodovnikov [14].

In all these works, the objective is to obtain the specification of a linear dynamic system (Wiener filter) which accomplishes the prediction, separation, or detection of a random signal.<sup>1</sup>

Present methods for solving the Wiener problem are subject to a number of limitations which seriously curtail their practical usefulness:

- (1) The optimal filter is specified by its impulse response. It is not a simple task to synthesize the filter from such data.
- (2) Numerical determination of the optimal impulse response is often quite involved and poorly suited to machine computation. The solution gets rapidly worse with increasing complexity of the problem.
- (3) Important generalizations (e.g., growing-memory filters, nonstationary prediction) require new derivations, frequently of considerable difficulty to the nonspecialist.
- (4) The mathematics of the derivations are not transparent. Fundamental assumptions and their consequences tend to be obscured.

This paper introduces a new look at this whole assemblage of problems, sidestepping the difficulties just mentioned. The following are the highlights of the paper:

- (5) *Optimal Estimator and Optimal Projections.* The Wiener problem is approached from the point of view of conditional distributions and expectations. In this way, basic facts of the Wiener theory are quickly obtained; the scope of the results and the fundamental assumptions appear clearly. It is seen that all statistical calculations and results are based on first and second order averages; no other statistical data are needed. Thus difficulty (4) is eliminated. This method is well known in probability theory (see pp. 75-78 and 148-155 of Doob [15]) and pp. 455-464 of Loève [16]) but has not yet been used extensively in engineering.

(6) *Models for Random Processes.* Following, in particular, Bode and Shannon [3], arbitrary random signals are represented (up to second order average statistical properties) as the output of a linear dynamic system excited by independent or uncorrelated random signals ("white noise"). This is a standard trick in the engineering applications of the Wiener theory [2-7]. The approach taken here differs from the conventional one only in the way in which linear dynamic systems are described. We shall emphasize the concepts of *state* and *state transition*; in other words, linear systems will be specified by systems of first-order difference (or differential) equations. This point of view is

$$\Phi(t+1) = \Phi(t) \bar{X}(t-1) + u(t) \quad (23)$$

has  $\Phi^*$  is also the transition matrix of the linear dynamic system over the error.

From (23) we obtain at once a recursion relation for the covariance matrix  $P^*(t)$  of the optimal error  $X(t-1)$ . Noting that  $(t)$  is independent of  $X(t)$  and therefore of  $X(t-1)$  we get

$$P^*(t+1) = E\bar{X}(t+1)\bar{X}^T(t-1) \quad (24)$$

$$\begin{aligned} &= E\bar{X}(t+1) + E\bar{X}(t-1)\bar{X}^T(t-1)\Phi^T(t+1; t) + Q(t) \\ &= E\bar{X}(t+1) + E\bar{X}(t-1)\bar{X}^T(t-1)\Phi^T(t+1; t) + Q(t) \\ &= E\bar{X}(t+1) + E\bar{X}(t-1)\bar{X}^T(t-1) + Q(t) \end{aligned} \quad (24)$$

here  $Q(t) = E u(t)u^T(t)$ . There remains the problem of obtaining an explicit formula for  $\Phi^*$  (and thus also for  $\Phi^*$ ). Since,

$$\bar{X}(t+1) = \bar{X}(t) + \bar{X}(t) - \bar{X}(t) + \bar{X}(t+1) - \bar{X}(t) \quad (25)$$

orthogonal to  $\bar{Y}(t-1)$ , it follows that by (19) that

$$0 = E\bar{X}(t+1) + E\bar{X}(t-1)\bar{Y}^T(t-1) \quad (26)$$

$$= E\bar{X}(t+1) + E\bar{X}(t-1) + E\bar{X}(t-1)\bar{Y}^T(t-1) - E\bar{X}(t-1) \quad (27)$$

$$= E\bar{X}(t+1) + E\bar{X}(t-1) - E\bar{X}(t-1) + E\bar{X}(t-1) \quad (28)$$

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$$X^*(t) = \bar{X}(t) - E\bar{X}(t) | Y(t) \quad (23)$$

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$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (48)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (49)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (50)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (51)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (52)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (53)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (54)$$

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$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (57)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (58)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (59)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (60)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (61)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (62)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (63)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (64)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (65)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (66)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (67)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (68)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (69)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (70)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (71)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (72)$$

$$= \bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) + E\bar{X}(t) - E\bar{X}(t) \quad (73)$$

the matrices occurring in equation (31) and the covariance matrix of  $X(t)$  are found after simple calculations. We have, for all  $t \geq 0$ ,

$$\Phi(t+1)X^*(t) = \frac{1}{C_1(t)} \begin{bmatrix} C_2(t) \\ C_1(t) \\ C_1(t) - C_2(t) \end{bmatrix} \quad (23)$$

$$\Phi(t+1)X^*(t) = \frac{1}{C_1(t)} \begin{bmatrix} C_2(t) \\ C_1(t) \\ C_1(t) - C_2(t) \end{bmatrix} \quad (24)$$

$$\Phi(t+1)X^*(t) = \frac{1}{C_1(t)} \begin{bmatrix} C_2(t) \\ C_1(t) \\ C_1(t) - C_2(t) \end{bmatrix} \quad (25)$$

$$\Phi(t+1)X^*(t) = \frac{1}{C_1(t)} \begin{bmatrix} C_2(t) \\ C_1(t) \\ C_1(t) - C_2(t) \end{bmatrix} \quad (26)$$

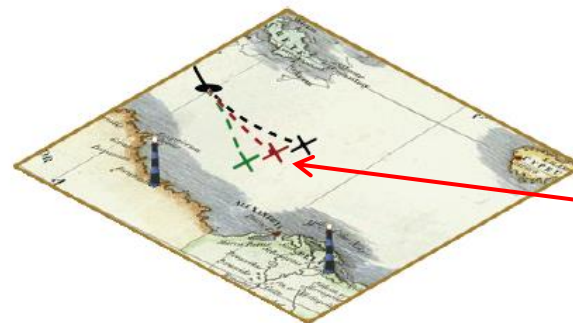
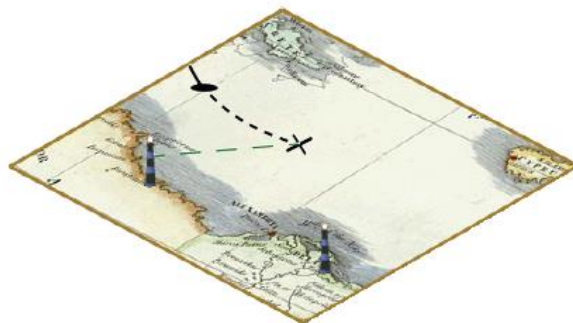
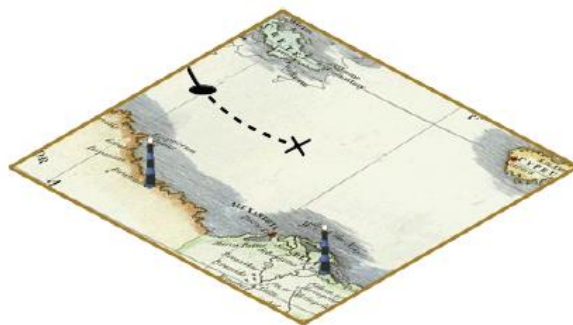
$$\Phi(t+1)X^*(t) = \frac{1}{C_1(t)} \begin{bmatrix} C_2(t) \\ C_1(t) \\ C_1(t) - C_2(t) \end{bmatrix} \quad (27)$$

$$\Phi(t+1)X^*(t) = \frac{1}{C_1(t)} \begin{bmatrix} C_2(t) \\ C_1(t) \\ C_1(t) - C_2(t) \end{bmatrix} \quad (28)$$

$$\Phi(t+1)X^*(t) = \frac{1}{C_1(t)} \begin{bmatrix} C_2(t) \\ C_1(t) \\ C_1(t) - C_2(t) \end{bmatrix} \quad (29)$$

$$\Phi(t+1)X^*(t) = \frac{1}{C_1(t)} \begin{bmatrix} C_2(t) \\ C_1(t) \\ C_1(t) - C_2(t) \end{bmatrix} \quad (30)$$

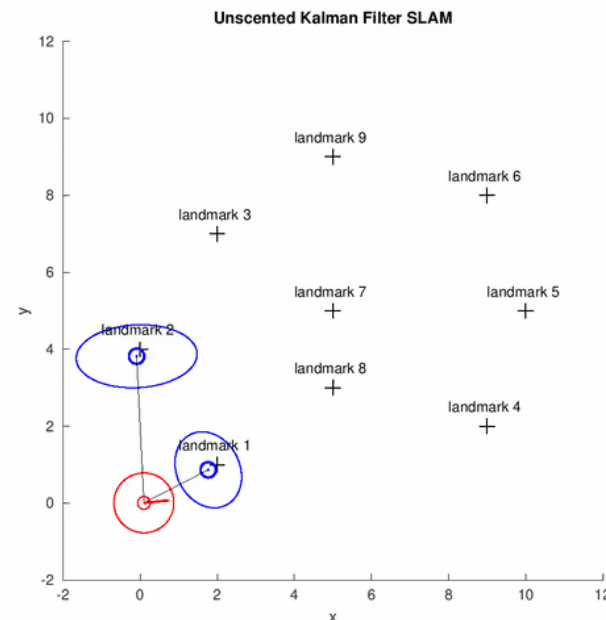
# Kalman Filter——Navigation application using dead reckoning and visual measurement to landmark



**Predicted and  
corrected position  
of the ship**



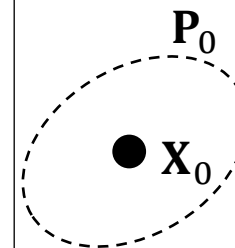
# Examples of Kalman Filter in Navigation



# Definitions

- > **X**, The *state vector* is the set of parameters describing a system, known as *states*, which the Kalman filter estimates.
- > **P**, Associated with the state vector is an *error covariance matrix*. This represents the uncertainties in the Kalman filter's state estimates and the degree of correlation between the errors in those estimates.

North

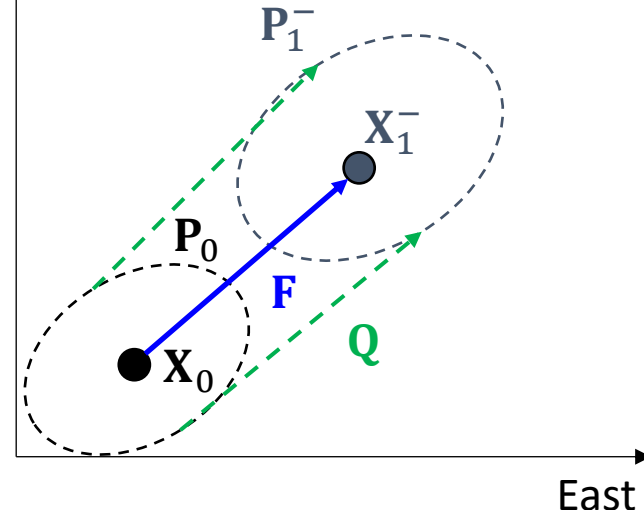


East

# Definitions

- > **F**, The *system model*, also known as the process model or time-propagation model, describes how the Kalman filter states and error covariance matrix vary with time. (The system model is deterministic for the states as it is based on known properties of the system.)
- > **Q**, A state uncertainty should also be increased with time to account for unknown changes in the system that cause the state estimate to go out of date in the absence of new measurement information. This variation in the true values of the states is known as *system noise* or *process noise*.

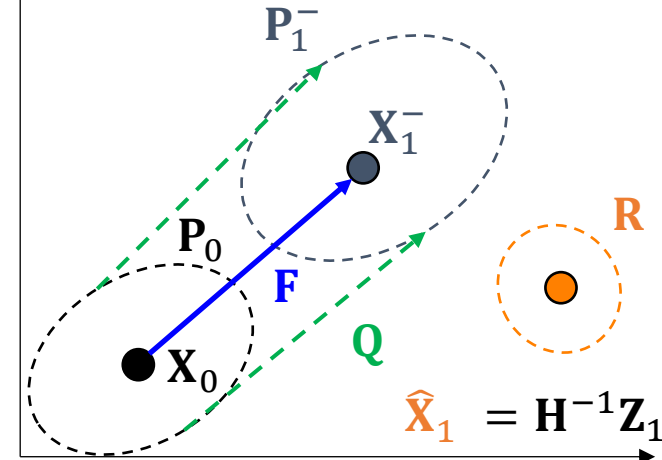
North



# Definitions

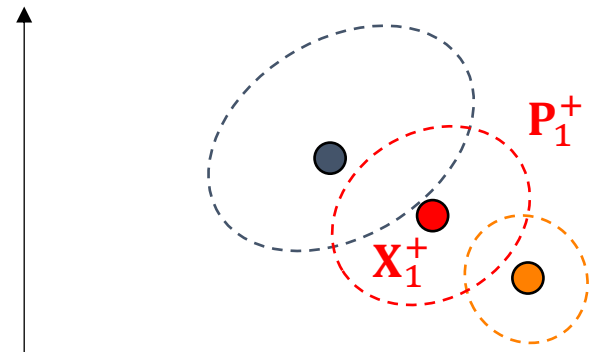
- > **Z**, The *measurement vector* is a set of simultaneous measurements of properties of the system which are functions of the state vector.
- > **R**, Associated with the measurement vector is a *measurement noise covariance matrix* which describes the statistics of the noise on the measurements.
- > **H**, **Z=H(X)=HX**, The *measurement model* describes how the measurement vector varies as a function of the true state vector (as opposed to the state vector estimate) in the absence of measurement noise.

North



North

East



# Key Equations of Kalman Filter

$$> \mathbf{X}_t^- = \mathbf{F}\mathbf{X}_{t-1}^+$$

$$> \mathbf{P}_t^- = \mathbf{F}\mathbf{P}_{t-1}^+ \mathbf{F}^T + \mathbf{Q}$$

State  
Propagation

$$> \Delta \mathbf{Z}_t = \mathbf{Z}_t - \mathbf{H}\mathbf{X}_t^-$$

$$> \mathbf{S}_t = \mathbf{H}\mathbf{P}_t^- \mathbf{H}^T + \mathbf{R}$$

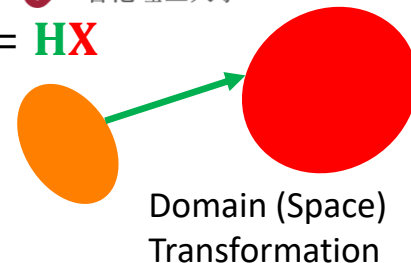
$$> \mathbf{K}_t = \mathbf{P}_t^- \mathbf{H}^T \mathbf{S}_t^-$$

$$> \mathbf{X}_t^+ = \mathbf{X}_t^- + \mathbf{K}_t \Delta \mathbf{Z}_t$$

$$> \mathbf{P}_t^+ = (\mathbf{I} - \mathbf{K}_t \mathbf{H}) \mathbf{P}_t^-$$

Measurement  
Update

$$\mathbf{Z} = \mathbf{H}\mathbf{X}$$



# Measurement Innovation, $\Delta \mathbf{Z}_t$

>  $\Delta \mathbf{Z}_t = \mathbf{Z}_t - \mathbf{H}\mathbf{X}_t^-$

> Meaning: The difference between propagation and measurement in the domain of  $\mathbf{Z}$  (measurement)

>  $\Delta \mathbf{Z}_t = 0$ , What does it mean?

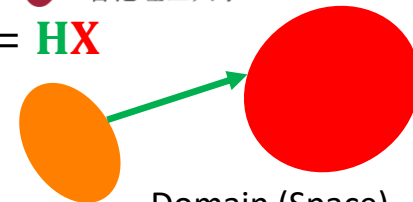
>  $\Delta \mathbf{Z}_t \neq 0$ , What shall we do?

# Kalman Filter Gain, $K_t$

$$> K_t = P_{t-1}^- H^T S_t^{-1}$$

$$> S_t = H P_{t-1}^- H^T + R$$

$$Z = HX$$



Domain (Space)  
Transformation

$$Z^T = X^T H^T$$

If we forget there is no denominator in Matrix...

$$> K_t = \frac{P_{t-1}^- H^T}{H P_{t-1}^- H^T + R}$$

The propagated state covariance  
(uncertainty) in the domain of the  
transpose of  $Z$  (measurement)

$$> K_t = \frac{(F P_{t-1}^+ F^T + Q) H^T}{H P_{t-1}^- H^T + R}$$

The propagated state covariance (uncertainty)  
in the domain of  $Z$  (measurement)

+ measurement covariance (uncertainty)  $R$

The updated state covariance (uncertainty)  
considering (+) propagation covariance (uncertainty)

# Kalman Filter Gain, $K_t$

$$> \mathbf{K}_t = \frac{(\mathbf{F}\mathbf{P}_{t-1}^+ \mathbf{F}^T + \mathbf{Q})\mathbf{H}^T}{\mathbf{H}\mathbf{P}_{t-1}^- \mathbf{H}^T + \mathbf{R}}$$

>  $\mathbf{R} \gg \mathbf{Q}$ , what does it mean?

>  $\mathbf{R} \ll \mathbf{Q}$ , what does it mean?

>  $\mathbf{R} = \mathbf{Q}$ , what does it mean?



# Simplified the Models to Identity Matrix, $\mathbf{I}$

$$\mathbf{F} = \mathbf{I}$$

$$\mathbf{H} = \mathbf{I}$$

$$\mathbf{B} = \mathbf{I}$$

$$> \mathbf{X}_t^- = \mathbf{F}\mathbf{X}_{t-1}^+ + \mathbf{B}\mathbf{U}_t$$

$$> \mathbf{P}_t^- = \mathbf{F}\mathbf{P}_{t-1}^+ \mathbf{F}^T + \mathbf{Q}$$

$$> \Delta \mathbf{Z}_t = \mathbf{Z}_t - \mathbf{H}\mathbf{X}_t^-$$

$$> \mathbf{S}_t = \mathbf{H}\mathbf{P}_{t-1}^- \mathbf{H}^T + \mathbf{R}$$

$$> \mathbf{K}_t = \mathbf{P}_{t-1}^- \mathbf{H}^T \mathbf{S}_t^{-1}$$

$$> \mathbf{X}_t^+ = \mathbf{X}_t^- + \mathbf{K}_t \Delta \mathbf{Z}_t$$

$$> \mathbf{P}_t^+ = (\mathbf{I} - \mathbf{K}_t \mathbf{H}) \mathbf{P}_{t-1}^-$$

$$> \mathbf{X}_t^- = \mathbf{X}_{t-1}^+ + \mathbf{U}_t$$

$$> \mathbf{P}_t^- = \mathbf{P}_{t-1}^+ + \mathbf{Q}$$

$$> \Delta \mathbf{Z}_t = \mathbf{Z}_t - \mathbf{X}_t^-$$

$$> \mathbf{S}_t = \mathbf{P}_{t-1}^- + \mathbf{R}$$

$$> \mathbf{K}_t = \mathbf{P}_{t-1}^- \mathbf{S}_t^{-1}$$

$$> \mathbf{X}_t^+ = \mathbf{X}_t^- + \mathbf{K}_t \Delta \mathbf{Z}_t$$

$$> \mathbf{P}_t^+ = (\mathbf{I} - \mathbf{K}_t) \mathbf{P}_{t-1}^-$$

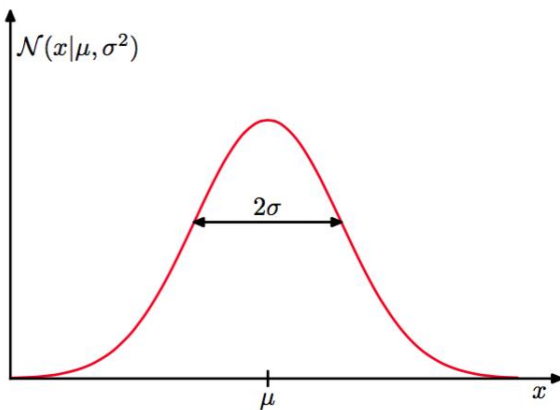
# How do we describe the uncertainty?

> Gaussian distribution

1D (univariate)

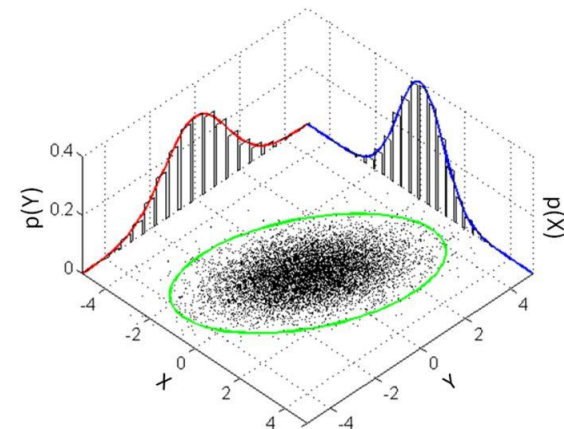
$$p(x) \sim N(\mu, \sigma^2)$$

$$p(x) = \frac{1}{\sqrt{2\pi}\sigma} e^{-\frac{1}{2} \frac{(x-\mu)^2}{\sigma^2}}$$

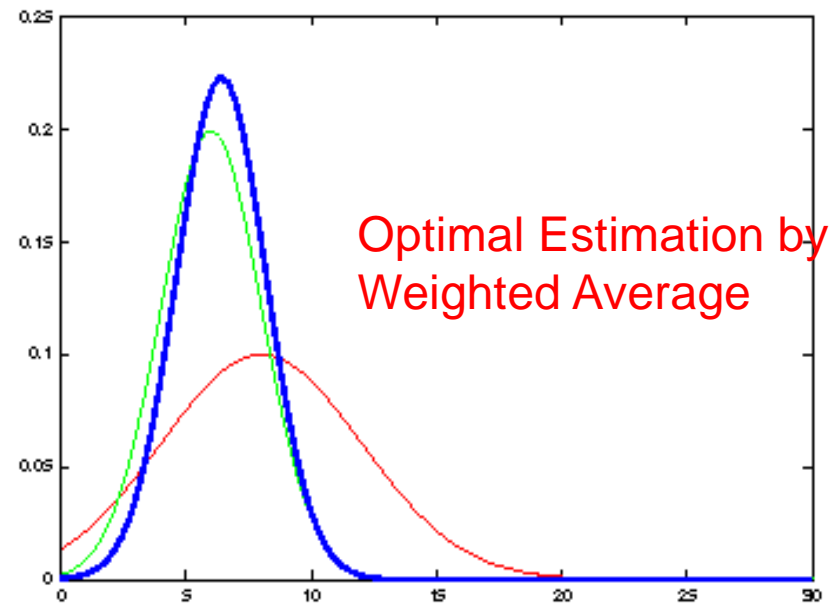
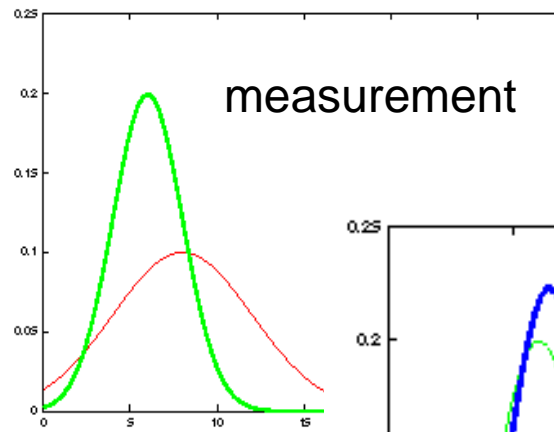
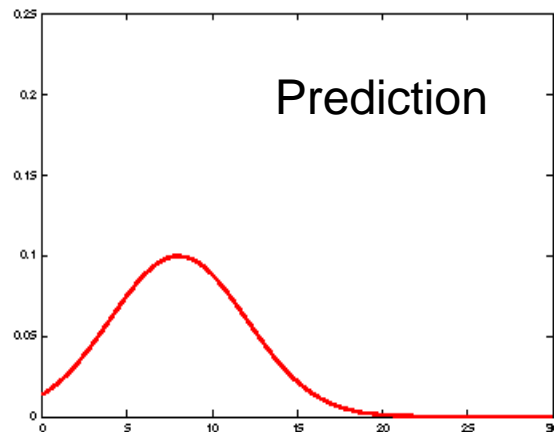


2D+ (multi variate)

$$p(\mathbf{x}) = \frac{1}{(2\pi)^{d/2} |\Sigma|^{1/2}} e^{-\frac{1}{2} (\mathbf{x}-\mu)' \Sigma^{-1} (\mathbf{x}-\mu)}$$



# What Kalman Filter tries to achieve?



# Fusion of Gaussian distribution in 1D

After some calculation and rearrangements

> Distribution of two measurements

$$\hat{q}_1 = q_1 \text{ with variance } \sigma_1^2$$

$$\hat{q}_2 = q_2 \text{ with variance } \sigma_2^2$$

Kalman Gain

$$\hat{q} = q_1 + \frac{\sigma_1^2}{\sigma_1^2 + \sigma_2^2} (q_2 - q_1)$$

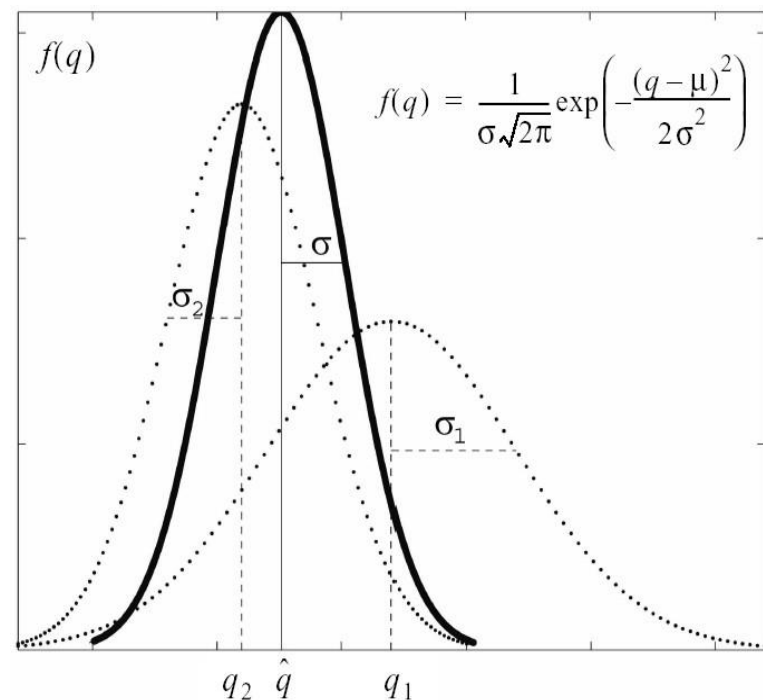
> Weighted least-square

$$S = \sum_{i=1}^n w_i (\hat{q} - q_i)^2$$

> Finding minimum error

$$\frac{\partial S}{\partial \hat{q}} = \frac{\partial}{\partial \hat{q}} \sum_{i=1}^n w_i (\hat{q} - q_i)^2 = 2 \sum_{i=1}^n w_i (\hat{q} - q_i) = 0$$

If  $w_i = 1/\sigma_i^2$



# Question: Why Kalman Filter is **Optimal**?

- > What does it mean to Optimal?
  - Maximum the Probability of the Estimation!
  
- > What assumptions are made in Kalman Filter so that it can achieve Optimal?
  - A. Gaussian Random Variable (Gaussian Noise)
  - B. 1<sup>st</sup> Order of Markov Chain

# Kalman Filter in GNSS

- > Example of the GNSS loosely-coupled pseudorange/Doppler integration using Kalman filter (this is done through the handwriting in visualizer)
- > Example of the GNSS tightly-coupled pseudorange/Doppler integration using Kalman filter

# Doppler Effect



- > Receiver is set in the car.
- > Received frequency is
- > “cs” is speed of light.
- > Doppler frequency “ $f_D$ ” is equal to “ $f_{obs} - f_{source}$ ”
- > FLL (frequency lock loop) tries to estimate “ $f_D$ ”.
- > Once we can estimate “ $f_D$ ”, “ $v_o$ ” can be resolved.

$$f_{obs} = f_s \frac{cs - v_o}{cs - v_s}$$

# Velocity Estimation

- > Velocity estimation in GPS is just same as shown in the previous slide.
- > The differences are as follows.
- \* **3 dimension velocity ( $\mathbf{v}_x, \mathbf{v}_y, \mathbf{v}_z$ ) have to be estimated.**
- \* **Frequency in the receiver is based on on-board clock.**
- \* **Measurement is pseudorange rate, which calculated from Doppler frequency AND satellite velocity.**
- > 4 unknown variables ( $\mathbf{v}_x, \mathbf{v}_y, \mathbf{v}_z, f_{\text{clk}}$ ) have to be estimated using at least 4 visible satellites. DOP is also important.
- > Velocity estimation is same as position estimation.



# Receiver Velocity Estimation from the Doppler Measurements

## Measurements from Doppler

$$\mathbf{y} = (-\lambda_i D_{r,i}^1, -\lambda_i D_{r,i}^2, -\lambda_i D_{r,i}^3, \dots, -\lambda_i D_{r,i}^m)^T$$

## Observation function

$$\mathbf{h}(\mathbf{x}) = \begin{pmatrix} r_r^1 + cd\dot{t}_r - cd\dot{T}^1 \\ r_r^2 + cd\dot{t}_r - cd\dot{T}^2 \\ r_r^3 + cd\dot{t}_r - cd\dot{T}^3 \\ \vdots \\ r_r^m + cd\dot{t}_r - cd\dot{T}^m \end{pmatrix} \quad \mathbf{H} = \begin{pmatrix} -\mathbf{e}_r^{1T} & 1 \\ -\mathbf{e}_r^{2T} & 1 \\ -\mathbf{e}_r^{3T} & 1 \\ \vdots & \vdots \\ -\mathbf{e}_r^{mT} & 1 \end{pmatrix} \quad (\text{F.6.28})$$

Can you try to formulate the steps for GNSS velocity estimation similar to the position estimation? 😊

The range-rate  $\dot{r}_r^s$  between the receiver and the satellite in these equations is derived from:

$$\dot{r}_r^s = \mathbf{e}_r^{sT} \left( \mathbf{v}^s(t^s) - \mathbf{v}_r \right) + \frac{\omega_e}{c} \left( v_y^s x_r + y^s v_{x,r} - v_x^s y_r - x^s v_{y,r} \right) \quad (\text{F.6.29})$$

# Q&A

# Thank you for your attention 😊

## Q&A

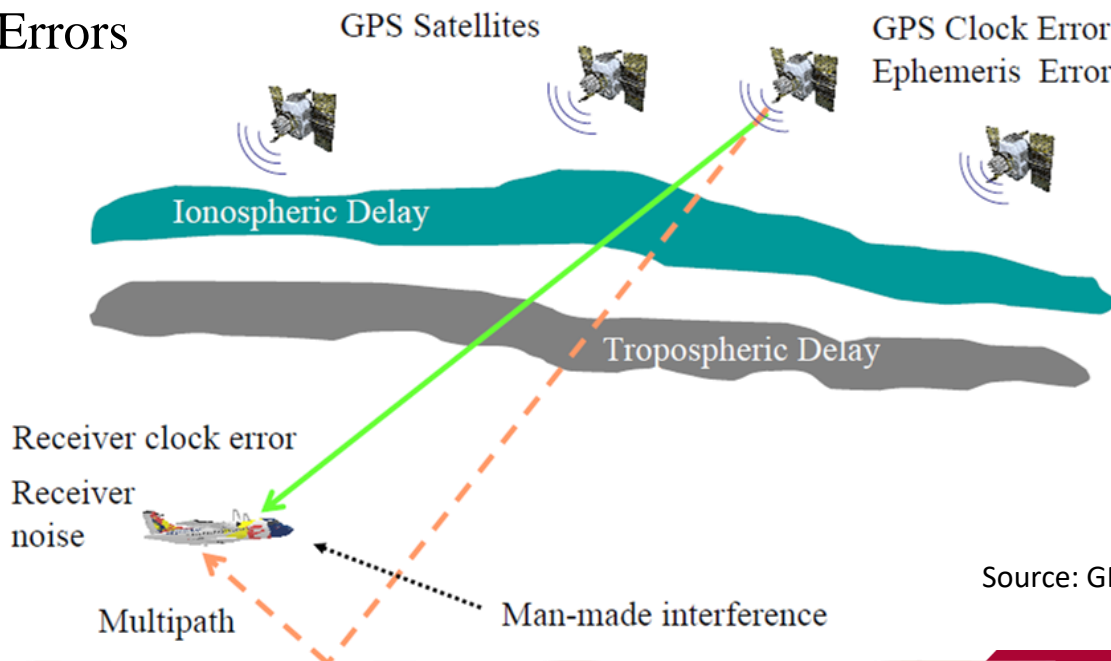
Dr. Weisong Wen

If you have any questions or inquiries,  
please feel free to contact me.

Email: [welson.wen@polyu.edu.hk](mailto:welson.wen@polyu.edu.hk)

# Source of Measurements Errors

- > Control Segment Errors
- > Signal Propagation Modeling Errors
- > Measurement Errors



Source: GPS Lab. Stanford Univ.

# Why we discuss about measurement errors ?

- > Back to bias and noise errors discussion, noise errors of pseudo-range can be mitigated to some degree using **carrier phase smoothing technique**.
- > On the other hand, you have to estimate **bias errors** as accurate as possible **by yourself** to improve positioning performance.
- > All kinds of improved techniques are essentially same in terms of estimating or eliminating bias or noise errors.

# Observation Model for Pseudorange/Carrier Measurements

## Observation function for pseudorange (code) measurement

$$\underbrace{\rho_{r,t}^S}_{\text{Pseudorange}} = \underbrace{r_{r,t}^S}_{\substack{\text{Range} \\ \text{distance}}} + c(\underbrace{\delta_{r,t}}_{\substack{\text{Receiver clock} \\ \text{Bias (1~2m)}}} - \underbrace{\delta_{r,t}^S}_{\substack{\text{Satellite clock} \\ \text{bias}}}) + \underbrace{I_{r,t}^S}_{\substack{\text{ionospheric delay} \\ \text{Distance (1~2m)}}} + \underbrace{T_{r,t}^S}_{\substack{\text{tropospheric delay} \\ \text{Distance (1~2m)}}} + \underbrace{\epsilon_{r,t}^S}_{\substack{\text{multipath effects, NLOS} \\ \text{receptions, receiver noise,} \\ \text{antenna phase-related noise} \\ \text{(0~100m)}}}$$

$\|\mathbf{p}_t^{G,S} - \mathbf{p}_{r,t}^G\|$

## Observation function for carrier-phase measurement

$$\underbrace{\varphi_{r,t}^S}_{\text{Carrier-phase}} = \underbrace{r_{r,t}^S}_{\substack{\text{Range} \\ \text{distance}}} + c(\underbrace{\delta_{r,t}}_{\substack{\text{Receiver clock} \\ \text{Bias (1~2m)}}} - \underbrace{\delta_{r,t}^S}_{\substack{\text{Satellite clock} \\ \text{bias}}}) + \underbrace{I_{r,t}^S}_{\substack{\text{ionospheric delay} \\ \text{Distance (1~2m)}}} + \underbrace{T_{r,t}^S}_{\substack{\text{tropospheric delay} \\ \text{Distance (1~2m)}}} + \underbrace{\epsilon_{r,t}^S}_{\substack{\text{multipath effects, NLOS} \\ \text{receptions, receiver noise,} \\ \text{antenna phase-related noise} \\ \text{(0~100m)}}} + \underbrace{N_{r,t}^S}_{\text{Ambiguity}}$$

$\|\mathbf{p}_t^{G,S} - \mathbf{p}_{r,t}^G\|$

To use the carrier-phase measurements, the ambiguity need to be resolved.

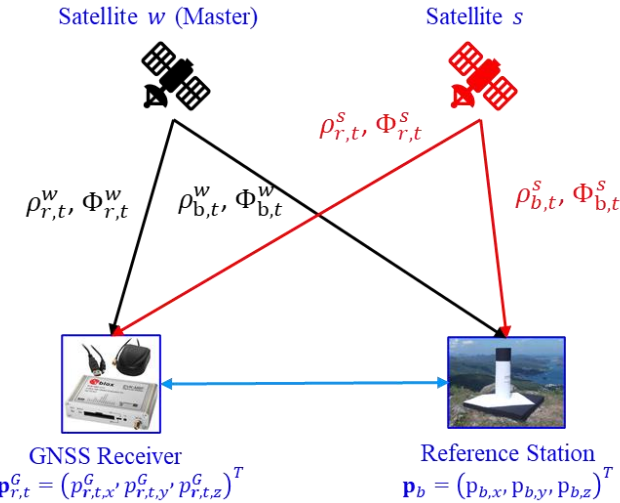
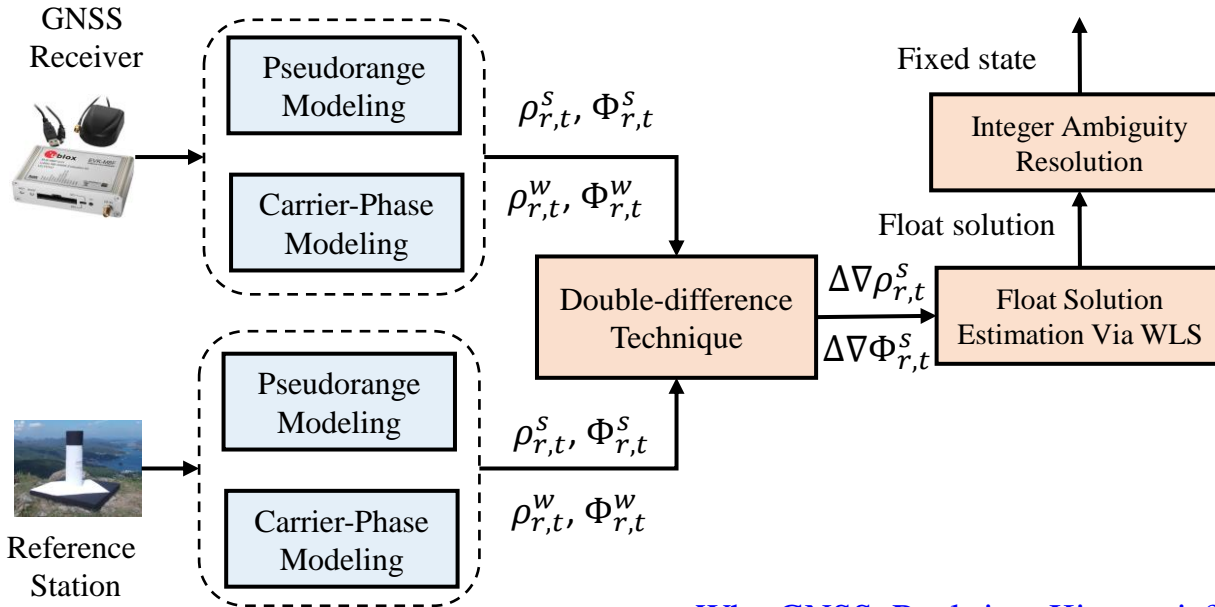
For each carrier-phase measurement, you got an unknown variable  $N_{r,t}^S$ !

# RTK (Real Time Kinematic)

- The concept of **RTK** is same as **DGPS**.
- RTK uses **carrier phase measurements**. DGPS uses pseudo-range measurements.
- GPS receiver is able to measure 1/100 of wavelength of L1 frequency (19 cm).
- If you have high-end receiver, you know your position **within 1-2cm accuracy** as long as you have 5 or more LOS satellites.

# Overview of GNSS Real-time Kinematic

WLS\*: Weighted Least Squares  
DD\*: Double-difference



## Why GNSS Real-time Kinematic?

- Remove the error from receiver/satellite clock bias, atmosphere error using double-difference technique.
- Use the high-accuracy carrier-phase measurements.

$\rho_{r,t}^s$ : Pseudorange measurement  
 $\Phi_{r,t}^s$ : Carrier-phase measurement  
 $\Delta\nabla\rho_{r,t}^s$ : DD Pseudorange measurement  
 $\Delta\nabla\Phi_{r,t}^s$ : DD Carrier-phase measurement

# Observation Model for Pseudorange/Carrier Measurements

## Observation function for pseudorange (code) measurement

$$\underbrace{\rho_{r,t}^S}_{\text{Pseudorange}} = \underbrace{r_{r,t}^S}_{\substack{\text{Range} \\ \text{distance}}} + c(\underbrace{\delta_{r,t}}_{\substack{\text{Receiver clock} \\ \text{Bias (1~2m)}}} - \underbrace{\delta_{r,t}^S}_{\substack{\text{Satellite clock} \\ \text{bias}}}) + \underbrace{I_{r,t}^S}_{\substack{\text{ionospheric delay} \\ \text{Distance (1~2m)}}} + \underbrace{T_{r,t}^S}_{\substack{\text{tropospheric delay} \\ \text{Distance (1~2m)}}} + \underbrace{\epsilon_{r,t}^S}_{\substack{\text{multipath effects, NLOS} \\ \text{receptions, receiver noise,} \\ \text{antenna phase-related noise} \\ \text{(0~100m)}}}$$

$\|\mathbf{p}_t^{G,S} - \mathbf{p}_{r,t}^G\|$

## Observation function for carrier-phase measurement

$$\underbrace{\psi_{r,t}^S}_{\substack{\text{Carrier-phase} \\ \text{range}}} = \underbrace{r_{r,t}^S}_{\substack{\text{Range} \\ \text{distance}}} + c(\underbrace{\delta_{r,t}}_{\substack{\text{Receiver clock} \\ \text{Bias (1~2m)}}} - \underbrace{\delta_{r,t}^S}_{\substack{\text{Satellite clock} \\ \text{bias}}}) + \underbrace{I_{r,t}^S}_{\substack{\text{ionospheric delay} \\ \text{Distance (1~2m)}}} + \underbrace{T_{r,t}^S}_{\substack{\text{tropospheric delay} \\ \text{Distance (1~2m)}}} + \underbrace{\epsilon_{r,t}^S}_{\substack{\text{multipath effects, NLOS} \\ \text{receptions, receiver noise,} \\ \text{antenna phase-related noise} \\ \text{(0~100m)}}} + \underbrace{N_{r,t}^S}_{\substack{\text{Ambiguity}}}$$

$\|\mathbf{p}_t^{G,S} - \mathbf{p}_{r,t}^G\|$

To use the carrier-phase measurements, the ambiguity need to be resolved.



# Single Difference Pseudorange Measurements

Single difference between the GNSS receiver and the reference station to remove the atmosphere errors:

$$\rho_{r,t}^s = r_{r,t}^s + c(\delta_{r,t} - \delta_{r,t}^s) + I_{r,t}^s + T_{r,t}^s + \varepsilon_{r,t}^s$$

$$\rho_{b,t}^s = r_{b,t}^s + c(\delta_{b,t} - \delta_{b,t}^s) + I_{b,t}^s + T_{b,t}^s + \varepsilon_{b,t}^s$$

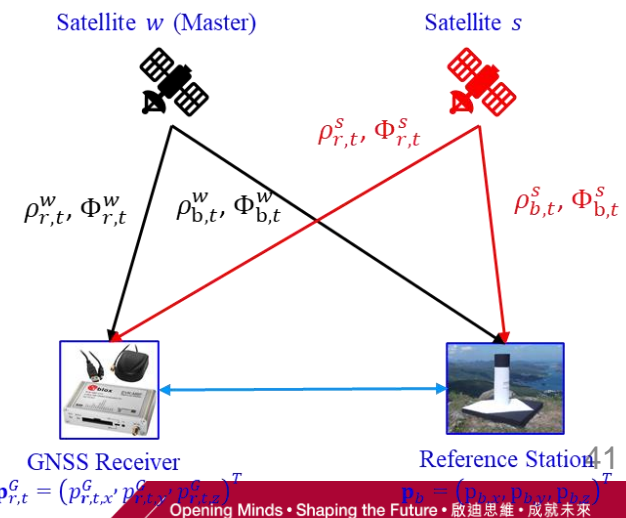
$$\Delta\rho_{r,t}^s = \rho_{r,t}^s - \rho_{b,t}^s = r_{r,t}^s - r_{b,t}^s + c(\delta_{r,t} - \delta_{r,t}^s) - c(\delta_{b,t} - \delta_{b,t}^s) \quad \text{Satellite } s$$

$$\rho_{r,t}^w = r_{r,t}^w + c(\delta_{r,t} - \delta_{r,t}^w) + I_{r,t}^w + T_{r,t}^w + \varepsilon_{r,t}^w$$

$$\rho_{b,t}^w = r_{b,t}^w + c(\delta_{b,t} - \delta_{b,t}^w) + I_{b,t}^w + T_{b,t}^w + \varepsilon_{b,t}^w$$

$$\Delta\rho_{r,t}^w = \rho_{r,t}^w - \rho_{b,t}^w = r_{r,t}^w - r_{b,t}^w + c(\delta_{r,t} - \delta_{r,t}^w) - c(\delta_{b,t} - \delta_{b,t}^w) \quad \text{Satellite } w \text{ (Master)}$$

**Assumption:** GNSS receiver and the reference station are close with the same atmosphere errors



# Double Difference Pseudorange Measurements

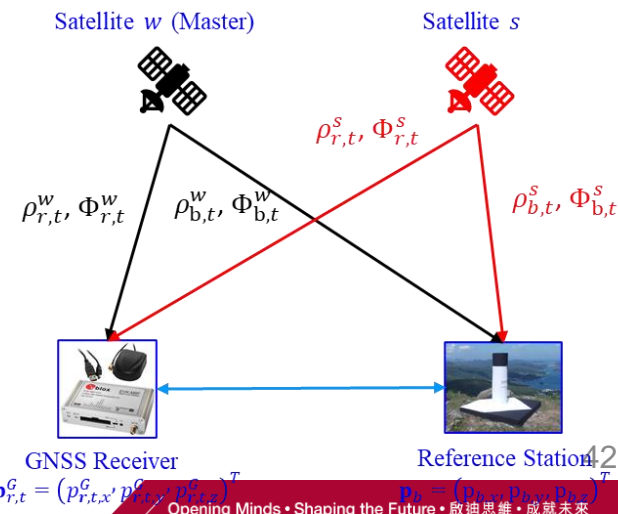
Second difference between the master satellite and the satellite  $s$  to remove the atmosphere errors:

$$\Delta\rho_{r,t}^s = \rho_{r,t}^s - \rho_{b,t}^s = r_{r,t}^s - r_{b,t}^s + c(\delta_{r,t} - \delta_{r,t}^s) - c(\delta_{b,t} - \delta_{b,t}^s) \quad \text{Satellite } s$$

$$\Delta\rho_{r,t}^w = \rho_{r,t}^w - \rho_{b,t}^w = r_{r,t}^w - r_{b,t}^w + c(\delta_{r,t} - \delta_{r,t}^w) - c(\delta_{b,t} - \delta_{b,t}^w) \quad \text{Satellite } w \text{ (Master)}$$

$$\Delta\nabla\rho_{r,t}^s = \Delta\rho_{r,t}^s - \Delta\rho_{r,t}^w = \rho_{r,t}^s - \rho_{b,t}^s - \rho_{r,t}^w - \rho_{b,t}^w \quad \text{DD measurements}$$

**Assumption:** GNSS receiver and the reference station are close with the same atmosphere errors



# Single Difference Carrier-phase Measurements

Single difference between the GNSS receiver and the reference station to remove the atmosphere errors:

$$\psi_{r,t}^s = r_{r,t}^s + c(\delta_{r,t} - \delta_{r,t}^s) + I_{r,t}^s + T_{r,t}^s + \varepsilon_{r,t}^s + N_{r,t}^s$$

$$\psi_{b,t}^s = r_{b,t}^s + c(\delta_{b,t} - \delta_{b,t}^s) + I_{b,t}^s + T_{b,t}^s + \varepsilon_{b,t}^s + N_{b,t}^s \quad \text{Satellite } s$$

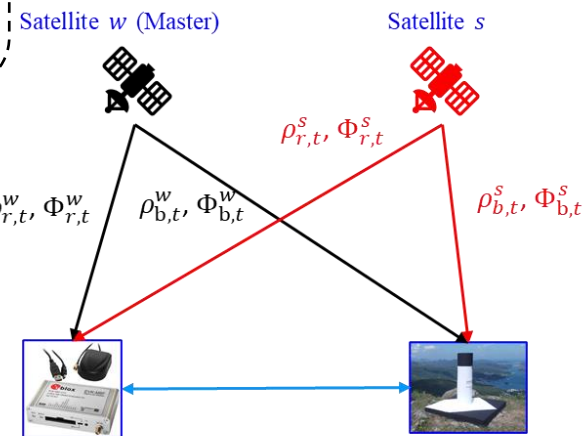
$$\Delta\psi_{r,t}^s = \psi_{r,t}^s - \psi_{b,t}^s = r_{r,t}^s - r_{b,t}^s + c(\delta_{r,t} - \delta_{r,t}^s) - c(\delta_{b,t} - \delta_{b,t}^s) + N_{r,t}^s - N_{b,t}^s$$

$$\psi_{r,t}^w = r_{r,t}^w + c(\delta_{r,t} - \delta_{r,t}^w) + I_{r,t}^w + T_{r,t}^w + \varepsilon_{r,t}^w + N_{r,t}^w$$

$$\psi_{b,t}^w = r_{b,t}^w + c(\delta_{b,t} - \delta_{b,t}^w) + I_{b,t}^w + T_{b,t}^w + \varepsilon_{b,t}^w + N_{b,t}^w \quad \text{Satellite } w$$

$$\Delta\psi_{r,t}^w = \psi_{r,t}^w - \psi_{b,t}^w = r_{r,t}^w - r_{b,t}^w + c(\delta_{r,t} - \delta_{r,t}^w) - c(\delta_{b,t} - \delta_{b,t}^w) + N_{r,t}^w - N_{b,t}^w$$

**Assumption:** GNSS receiver and the reference station are close with the same atmosphere errors



GNSS Receiver

Reference Station

$$\mathbf{p}_{r,t}^G = (p_{r,t,x}^G, p_{r,t,y}^G, p_{r,t,z}^G)^T$$

$$\mathbf{p}_b = (p_{b,x}, p_{b,y}, p_{b,z})^T$$

# Double Difference Pseudorange Measurements

Second difference between the master satellite and the satellite  $s$  to remove the atmosphere errors:

Satellite  $s$

$$\Delta\psi_{r,t}^s = \psi_{r,t}^s - \psi_{b,t}^s = r_{r,t}^s - r_{b,t}^s + c(\delta_{r,t} - \delta_{r,t}^s) - c(\delta_{b,t} - \delta_{b,t}^s) + N_{r,t}^s - N_{b,t}^s$$

Satellite  $w$  (Master)

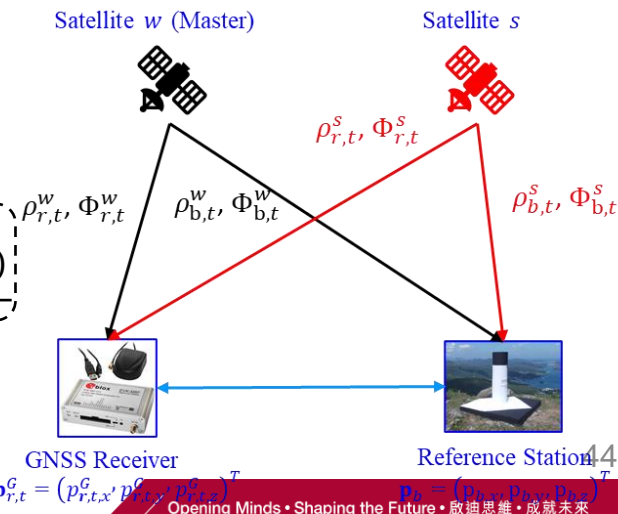
$$\Delta\psi_{r,t}^w = \psi_{r,t}^w - \psi_{b,t}^w = r_{r,t}^w - r_{b,t}^w + c(\delta_{r,t} - \delta_{r,t}^w) - c(\delta_{b,t} - \delta_{b,t}^w) + N_{r,t}^w - N_{b,t}^w$$

DD measurements

$$\Delta\Delta\psi_{r,t}^s = \Delta\psi_{r,t}^s - \Delta\psi_{r,t}^w = \rho_{r,t}^s - \rho_{b,t}^s - \rho_{r,t}^w - \rho_{b,t}^w + \underbrace{(N_{r,t}^s - N_{b,t}^s) - (N_{r,t}^w - N_{b,t}^w)}$$

DD Ambiguity:  $\Delta\Delta N_{r,t}^s$

**Assumption:** GNSS receiver and the reference station are close with the same atmosphere errors



# GNSS Real-time Kinematic Positioning: Float Solution Estimation

Estimate the float solution via weighted least square positioning

$$\begin{bmatrix} \mathbf{p}_{r,t}^G \\ \Delta \nabla N_{r,t}^1 \\ \Delta \nabla N_{r,t}^2 \\ \vdots \\ \Delta \nabla N_{r,t}^{m-1} \end{bmatrix} = \left( \mathbf{G}_t^G{}^T \mathbf{W}_t \mathbf{G}_t^G \right)^{-1} \mathbf{G}_t^G{}^T \mathbf{W}_t \begin{bmatrix} \Delta \nabla \rho_{r,t}^1 \\ \Delta \nabla \rho_{r,t}^2 \\ \vdots \\ \Delta \nabla \rho_{r,t}^{m-1} \\ \Delta \nabla \psi_{r,t}^1 \\ \Delta \nabla \psi_{r,t}^2 \\ \vdots \\ \Delta \nabla \psi_{r,t}^{m-1} \end{bmatrix}$$

Output

$\mathbf{p}_{r,t}^G$ : Float solution of position of GNSS receiver

$\Delta \nabla N_{r,t}^1, \Delta \nabla N_{r,t}^2, \dots$ : Float ambiguity

$\left( \mathbf{G}_t^G{}^T \mathbf{W}_t \mathbf{G}_t^G \right)^{-1}$ : Covariance matrix

$\mathbf{p}_{r,t}^G$ : Position of GNSS receiver

$\mathbf{W}_t$ : Weighting matrix

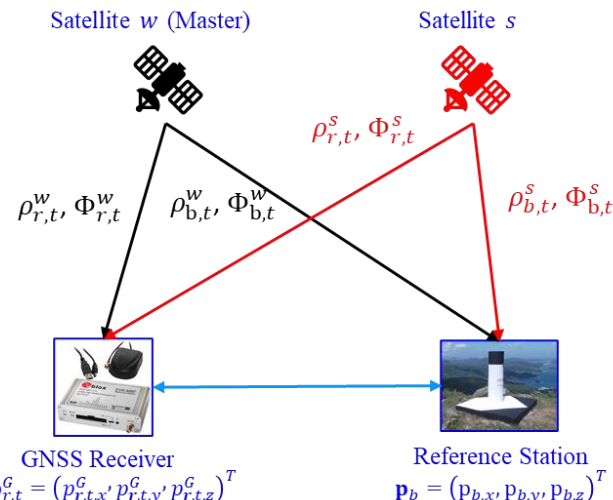
$m$ : number of satellite

$\mathbf{G}_t^G$ : Observation matrix

$\Delta \nabla \rho_{r,t}^{m-1}$ : DD pseudorange measurements

$\Delta \nabla \psi_{r,t}^{m-1}$ : DD carrier-phase measurements

How to formulate the  
linear least square problem  
of the GNSS-RTK  
Formulate it!



# GNSS Real-time Kinematic Positioning: Ambiguity Resolution

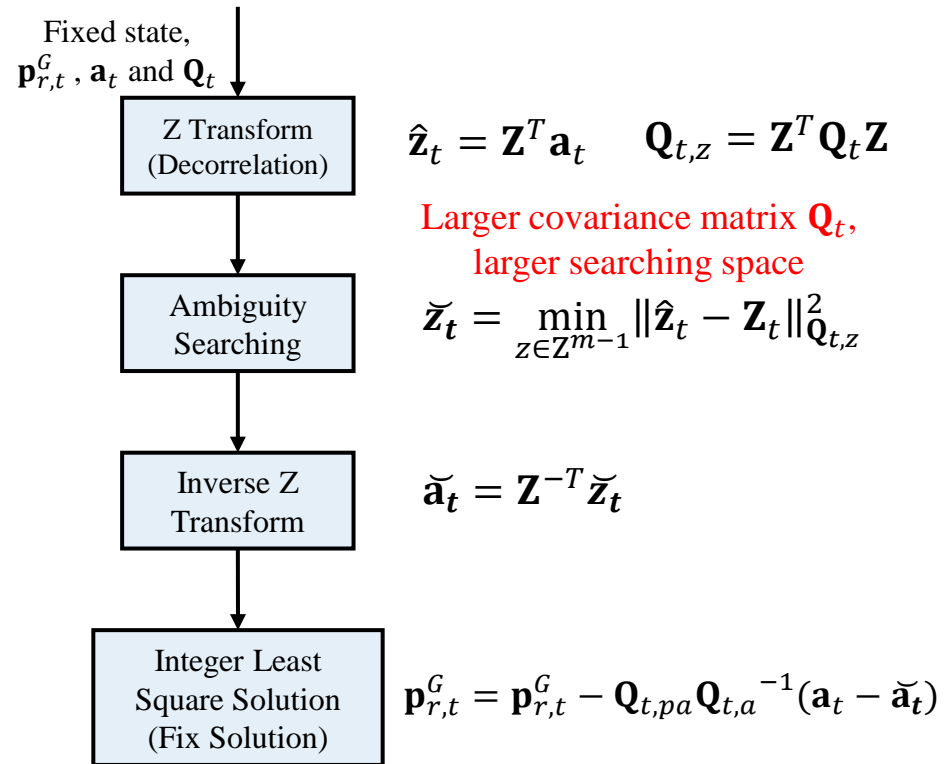
$\mathbf{p}_{r,t}^G$ : Float solution of position of GNSS receiver

$\mathbf{a}_t = \Delta \nabla N_{r,t}^1, \Delta \nabla N_{r,t}^2, \dots$ : Float ambiguity

$\mathbf{Q}_t = \left( \mathbf{G}_t^G \mathbf{W}_t \mathbf{G}_t^G \right)^{-1}$ : Covariance matrix

$$\mathbf{G}_t^G = \begin{bmatrix} \frac{p_{t,x}^{G,1} - p_{r,t,x}^G}{\|\mathbf{p}_t^{G,1} - \mathbf{p}_{r,t}^G\|} & \frac{p_{t,y}^{G,1} - p_{r,t,y}^G}{\|\mathbf{p}_t^{G,1} - \mathbf{p}_{r,t}^G\|} & \frac{p_{t,z}^{G,1} - p_{r,t,z}^G}{\|\mathbf{p}_t^{G,1} - \mathbf{p}_{r,t}^G\|} & 0 & \dots & 0 \\ \vdots & \vdots & \vdots & \vdots & \ddots & \vdots \\ \frac{p_{t,x}^{G,m-1} - p_{r,t,x}^G}{\|\mathbf{p}_t^{G,m-1} - \mathbf{p}_{r,t}^G\|} & \frac{p_{t,y}^{G,m-1} - p_{r,t,y}^G}{\|\mathbf{p}_t^{G,m-1} - \mathbf{p}_{r,t}^G\|} & \frac{p_{t,z}^{G,m-1} - p_{r,t,z}^G}{\|\mathbf{p}_t^{G,m-1} - \mathbf{p}_{r,t}^G\|} & 0 & \dots & 0 \\ \frac{p_{t,x}^{G,1} - p_{r,t,x}^G}{\|\mathbf{p}_t^{G,1} - \mathbf{p}_{r,t}^G\|} & \frac{p_{t,y}^{G,1} - p_{r,t,y}^G}{\|\mathbf{p}_t^{G,1} - \mathbf{p}_{r,t}^G\|} & \frac{p_{t,z}^{G,1} - p_{r,t,z}^G}{\|\mathbf{p}_t^{G,1} - \mathbf{p}_{r,t}^G\|} & 1 & \dots & 0 \\ \vdots & \vdots & \vdots & \vdots & \ddots & \vdots \\ \frac{p_{t,x}^{G,m-1} - p_{r,t,x}^G}{\|\mathbf{p}_t^{G,m-1} - \mathbf{p}_{r,t}^G\|} & \frac{p_{t,y}^{G,m-1} - p_{r,t,y}^G}{\|\mathbf{p}_t^{G,m-1} - \mathbf{p}_{r,t}^G\|} & \frac{p_{t,z}^{G,m-1} - p_{r,t,z}^G}{\|\mathbf{p}_t^{G,m-1} - \mathbf{p}_{r,t}^G\|} & 0 & \dots & 1 \end{bmatrix}$$

The correct fixed solution relies on the accuracy of float solution  $\mathbf{p}_{r,t}^G$  and the covariance matrix  $\mathbf{Q}_t$ .



# Statistical Methods for Estimation and Optimization

**Statistical method**

**States set**

$$\mathbf{X} = \{\mathbf{x}_0, \mathbf{x}_1, \mathbf{x}_2, \dots, \mathbf{x}_k\}$$

$$\mathbf{x} = \begin{bmatrix} x_0 \\ \vdots \\ x_m \end{bmatrix}$$

**Measurement set**

$$\mathbf{Z} = \{\mathbf{z}_0, \mathbf{z}_1, \mathbf{z}_2, \dots, \mathbf{z}_k\}$$

$$\mathbf{z} = \begin{bmatrix} z_0 \\ \vdots \\ z_n \end{bmatrix}$$

**Frequentist**

$$P(\mathbf{z}|\mathbf{x})$$

event-centric estimation, fully based on data. Should be used in Big Data application.

**Maximum Likelihood Estimation (MLE)**

$$\hat{\mathbf{x}} = \arg \max_{\mathbf{x} \in \mathbf{X}} P(\mathbf{Z}|\mathbf{X})(\mathbf{z}|\mathbf{x})$$

**Bayesians**

$$P(\mathbf{x}|\mathbf{z}) = \frac{P(\mathbf{z}|\mathbf{x})P(\mathbf{x})}{P(\mathbf{z})}$$

observer-centric estimation, required a knowledgeable observer (good prior-information)

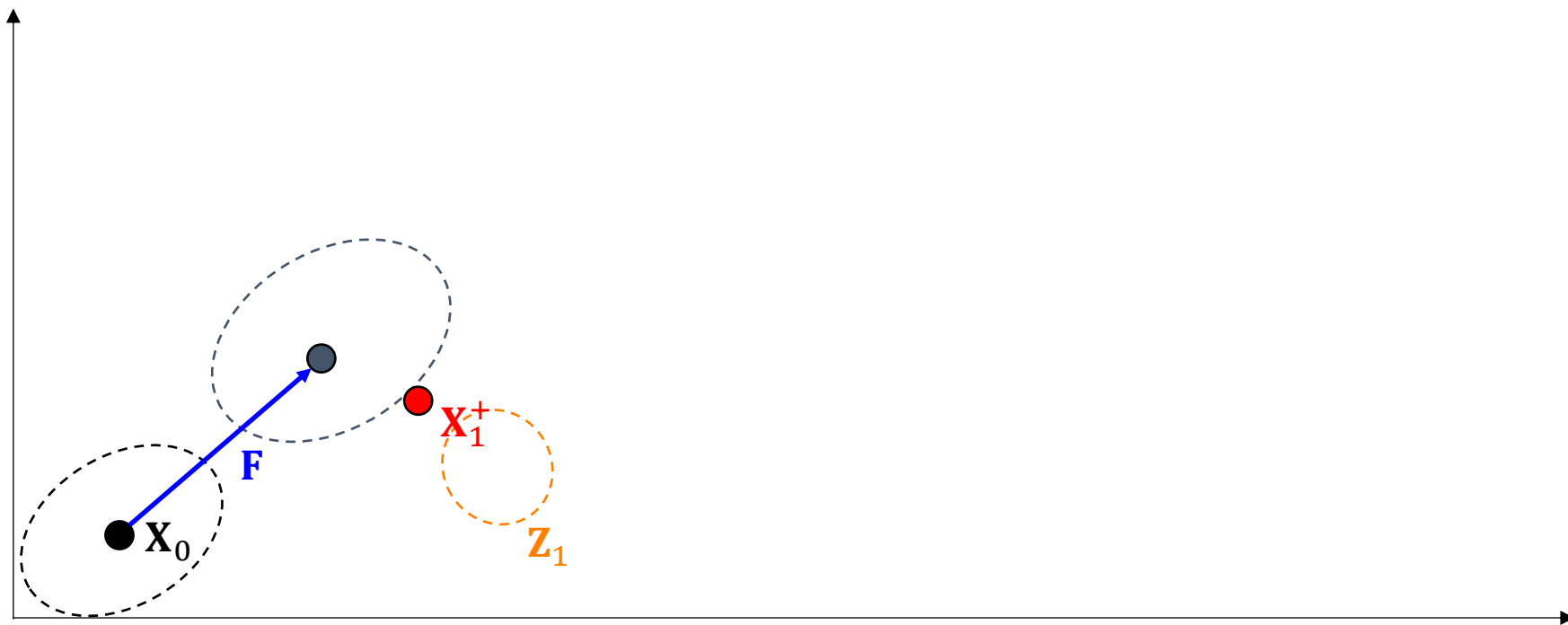
**Maximum a Posterior Estimation (MAP)**

$$\hat{\mathbf{x}} = \arg \max_{\mathbf{x} \in \mathbf{X}} P(\mathbf{Z}|\mathbf{X})(\mathbf{x}|\mathbf{z})$$

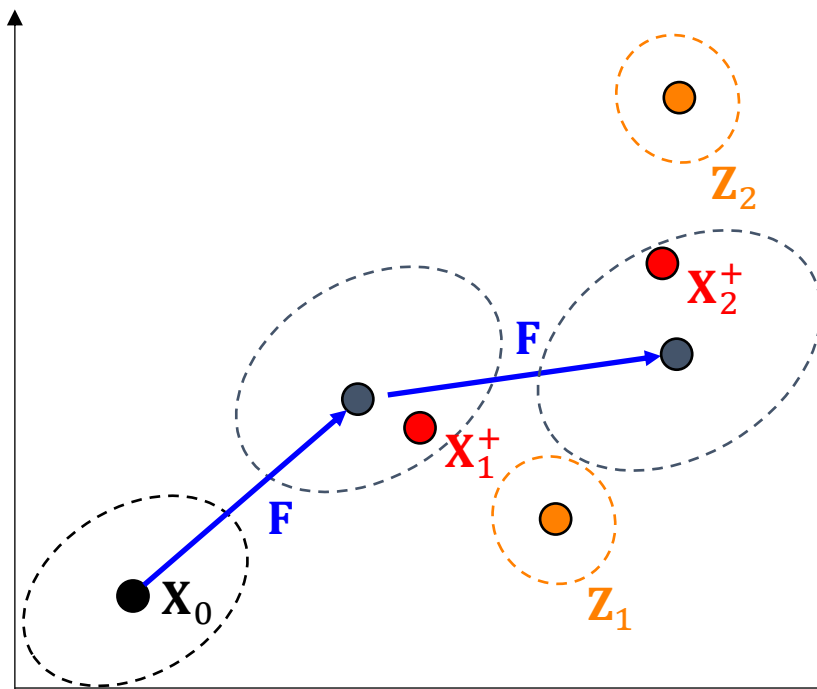
**Optimization, i.e., Kalman filter, factor graph optimization, etc.**



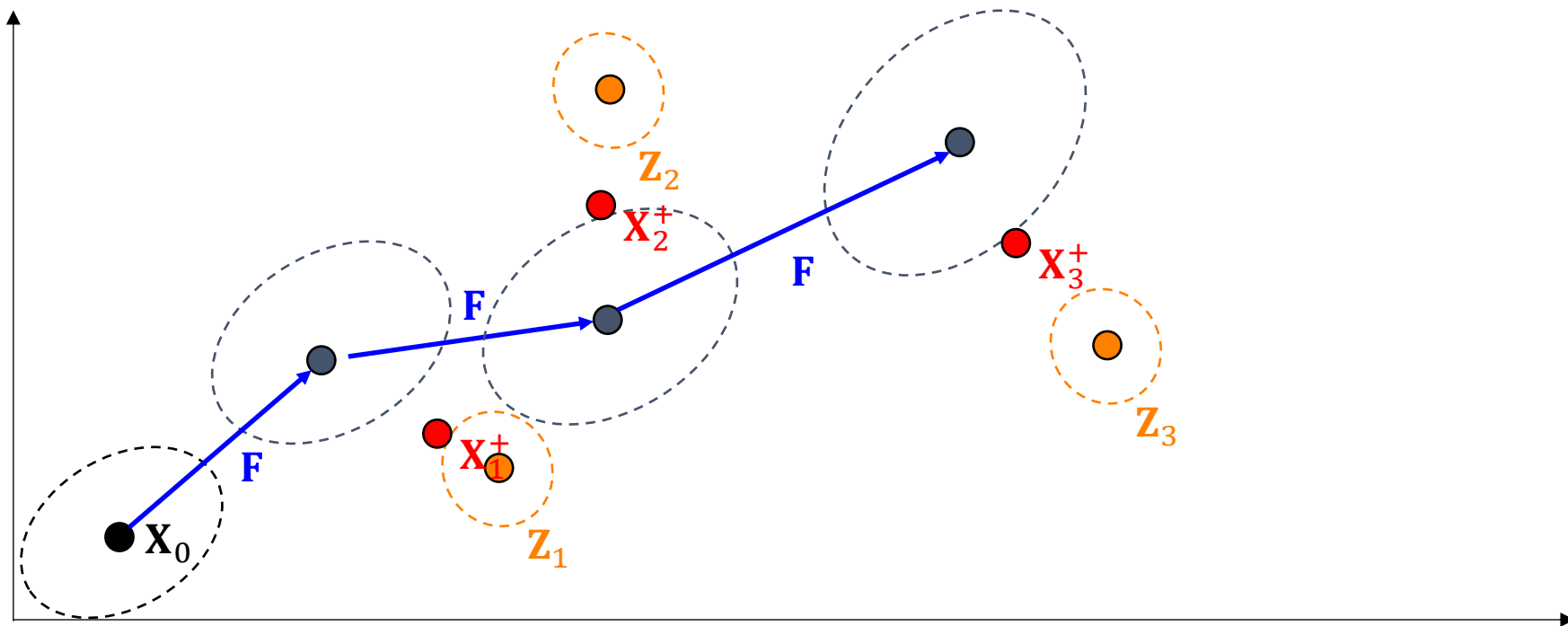
# Batch (including all the data in the past) optimization



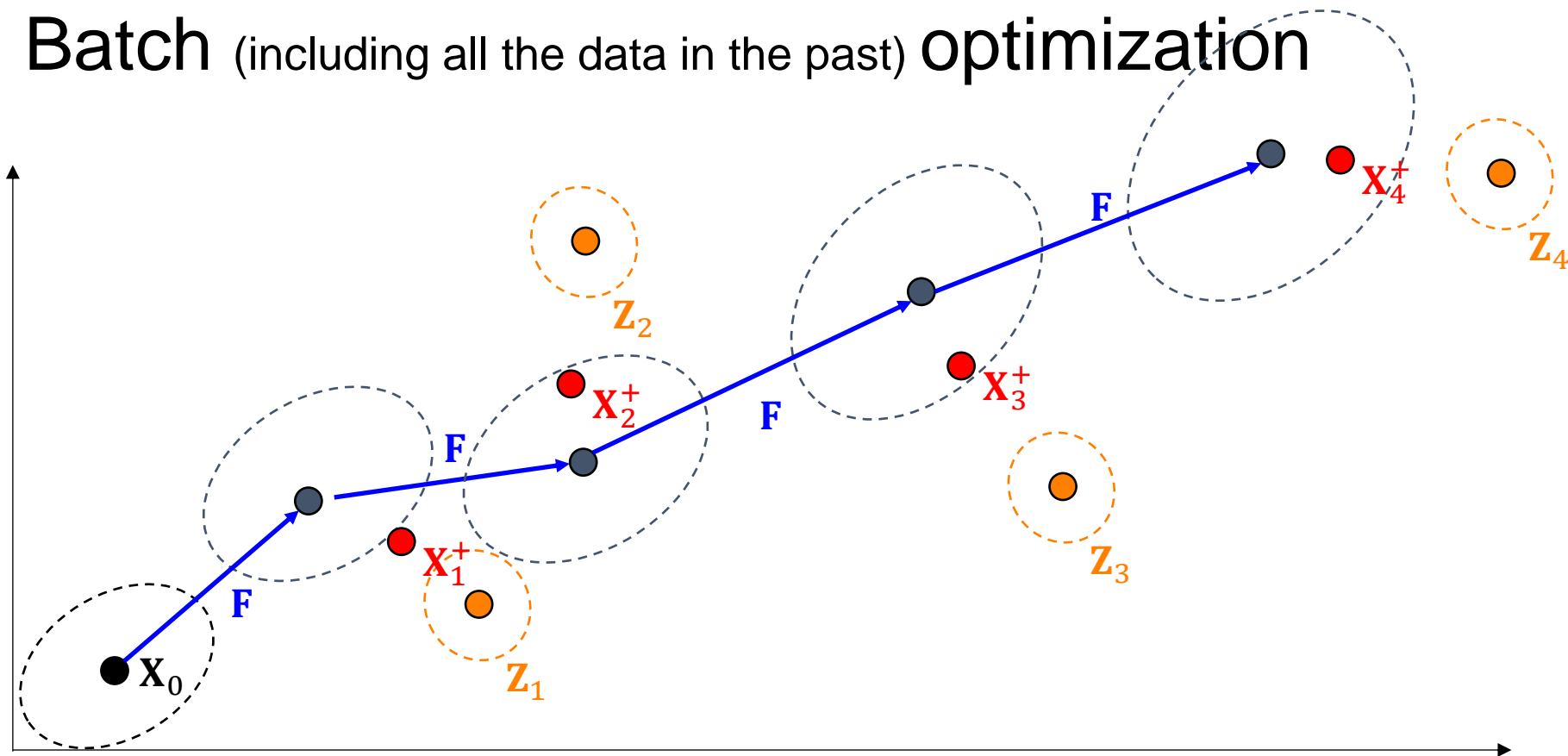
# Batch (including all the data in the past) optimization



# Batch (including all the data in the past) optimization

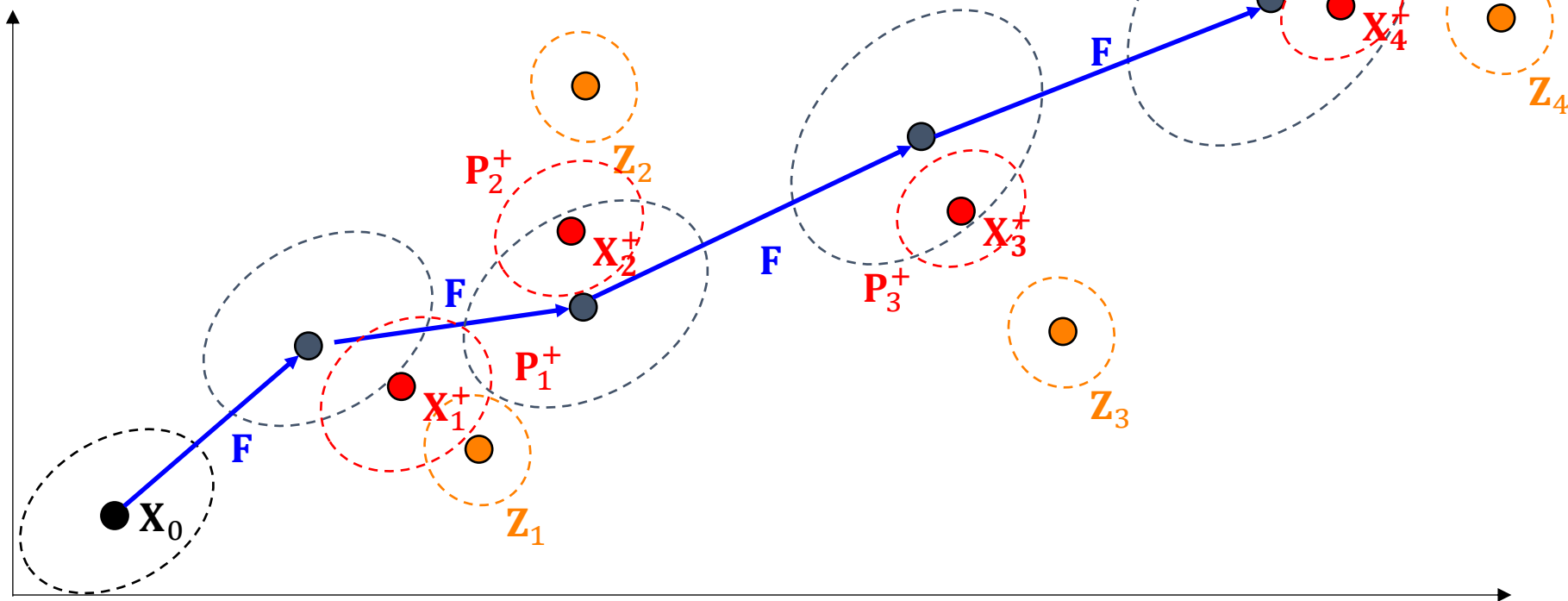


# Batch (including all the data in the past) optimization

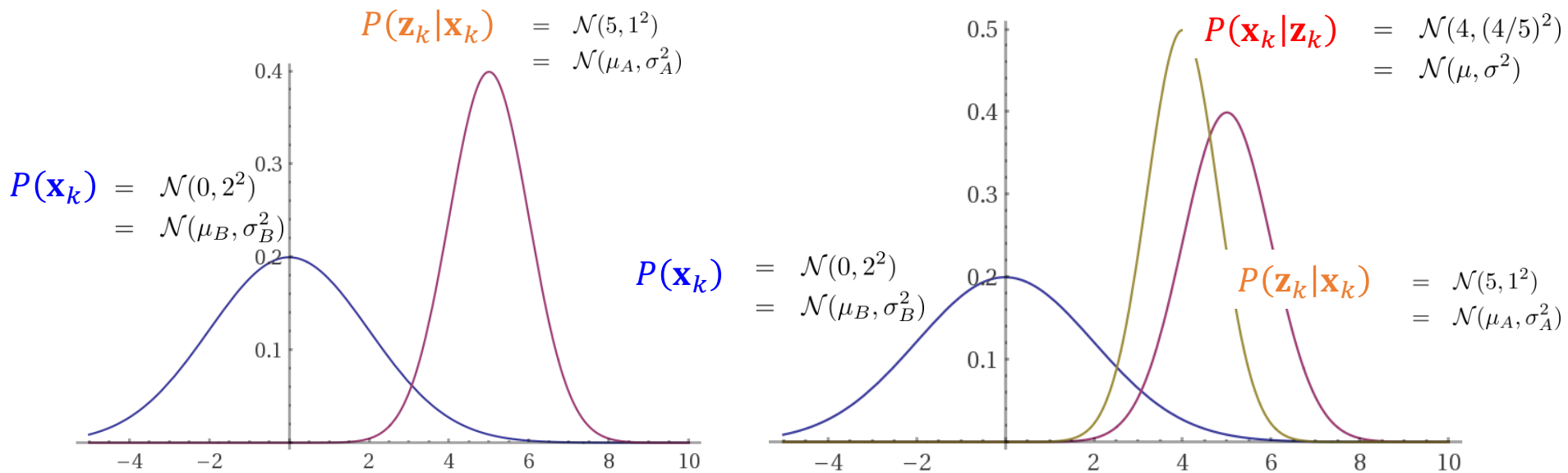


# Kalman Filter Optimization

1<sup>st</sup> order Markov Chain



# Kalman Filter with 1D state——Update step



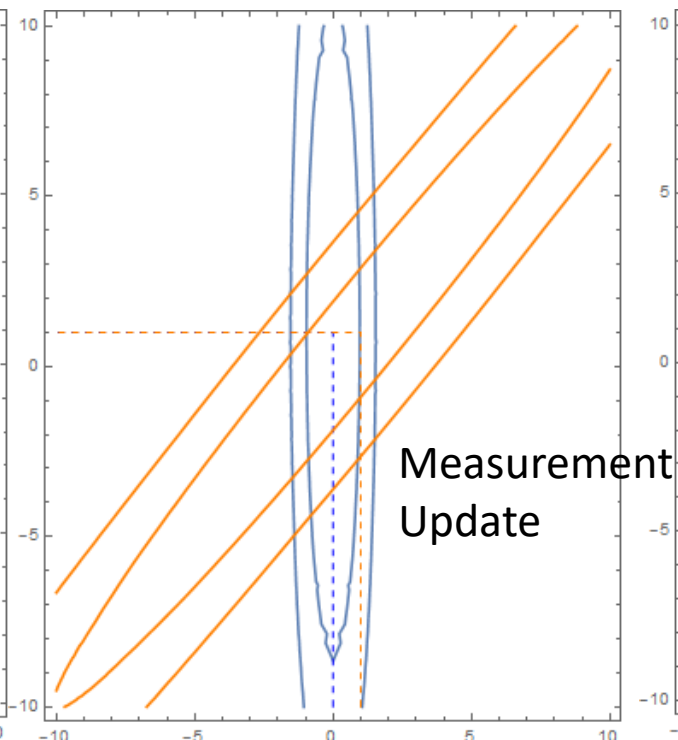
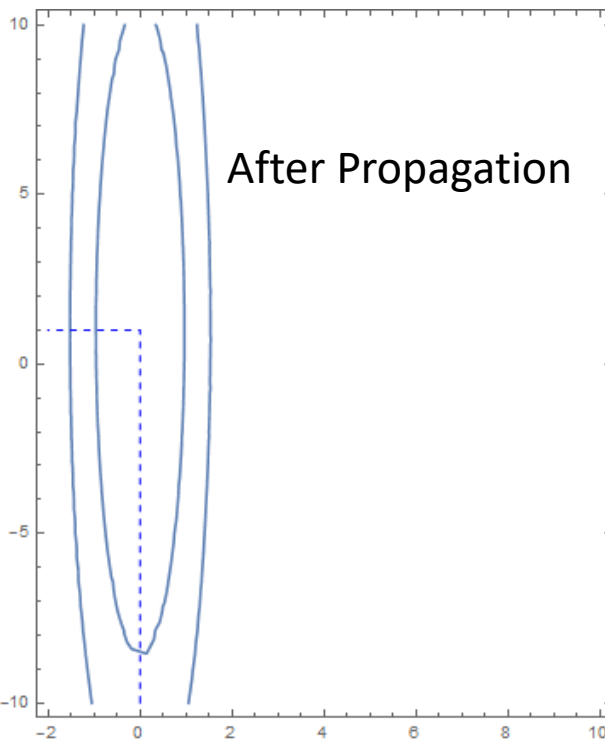
Kalman Gain: specifies how much effect will the measurement have in the posterior, compared to the prediction prior.

Which one do you trust more, your prior or your measurement ?

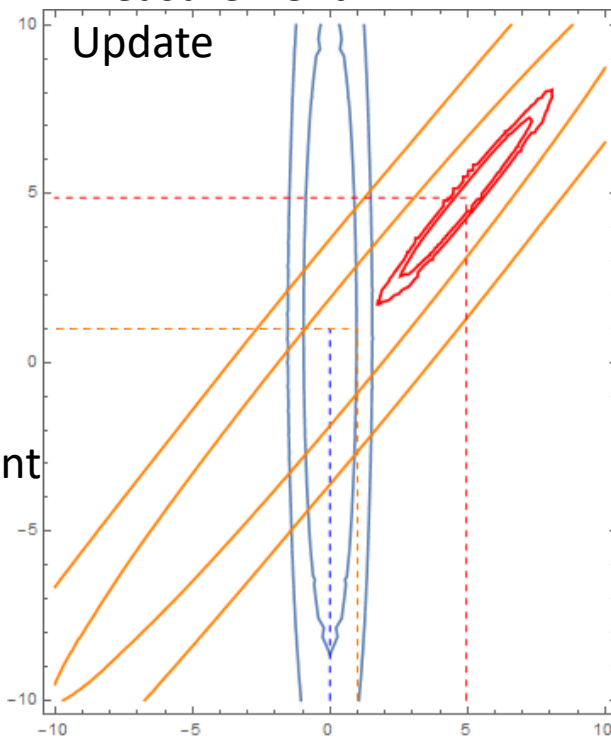
$$\mu = \mu_B + \frac{\sigma_B^2}{\sigma_A^2 + \sigma_B^2} (\mu_A - \mu_B)$$

$$\sigma^2 = \sigma_B^2 - \frac{\sigma_B^2}{\sigma_A^2 + \sigma_B^2} \sigma_B^2$$

# 2D Example



After  
Measurement  
Update





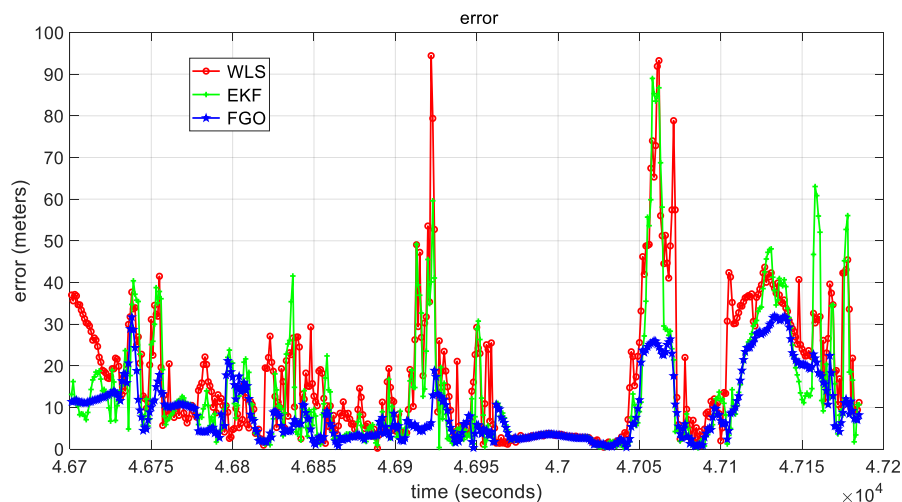
# Kalman filter in GNSS

- > Example of the GNSS loosely-coupled pseudorange/Doppler integration using Kalman filter
- > **Example of the GNSS tightly-coupled pseudorange/Doppler integration using Kalman filter**

# Evaluation of GNSS Positioning

GNSS positioning performance using the three listed methods

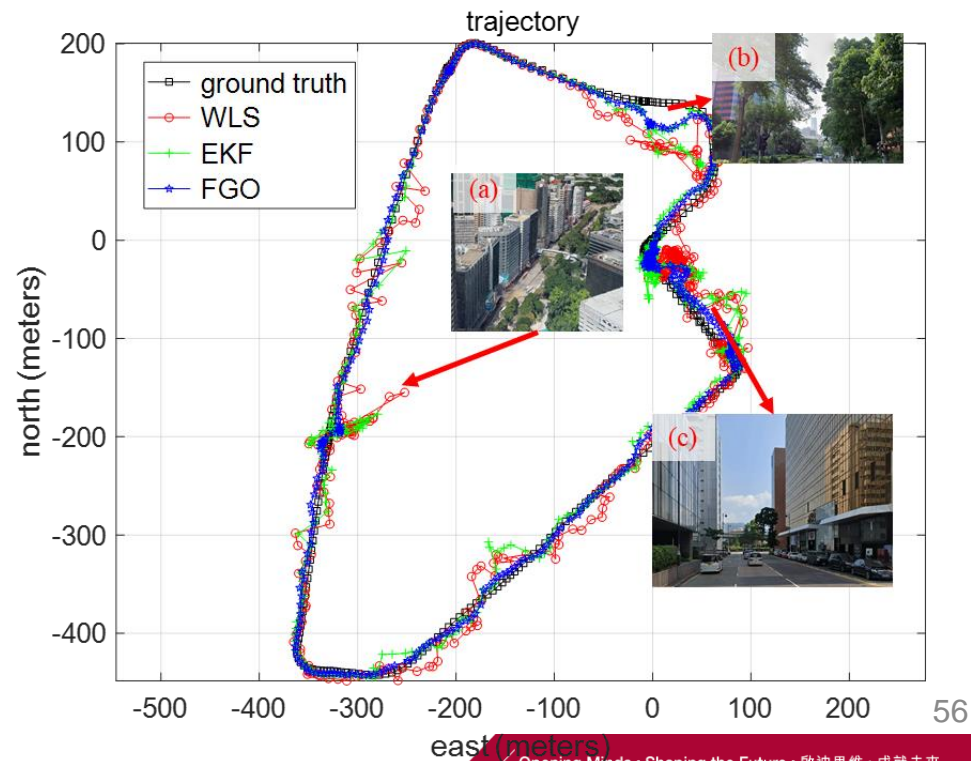
All data	WLS	EKF	FGO
MEAN (m)	17.39	13.61	9.45
STD (m)	16.01	15.19	8.06
MAX (m)	94.43	88.97	31.94
Availability	100%	100%	100%



WLS\*: weighted least square with pseudorange

EKF\*: Pseudorange/Doppler fusion with extended Kalman filter

FGO\*: Pseudorange/Doppler fusion with factor graph optimization



# Evaluation with Huawei P40 Pro



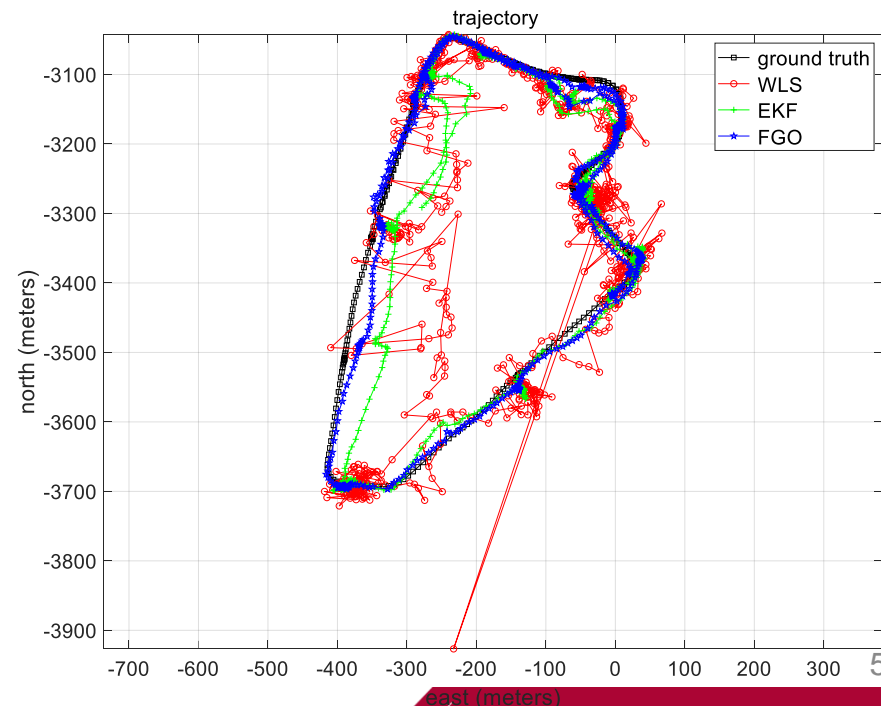
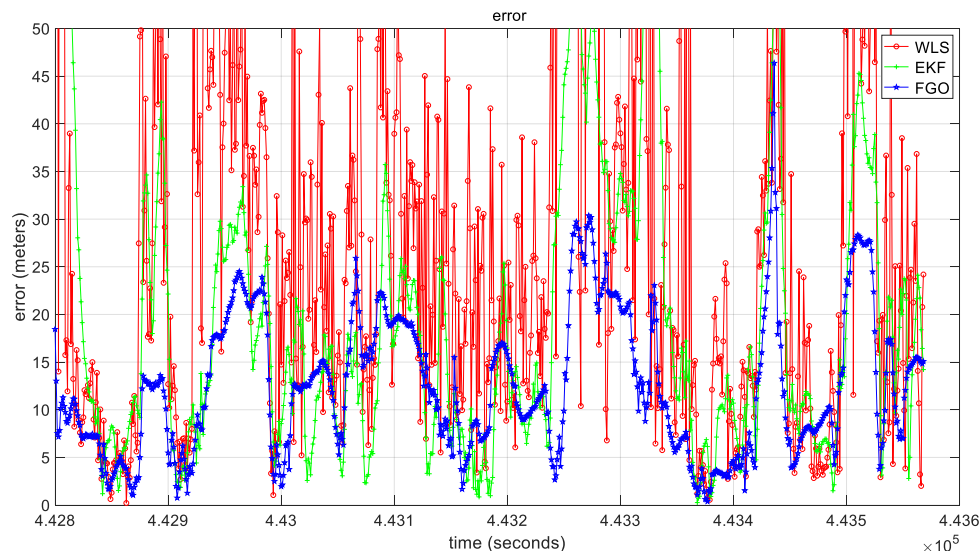
Huawei P40 Pro  
Phone

All data	WLS	EKF	FGO
MEAN (m)	31.98	19.84	12.541
STD (m)	38.22	15.78	7.48
MAX (m)	701.7	77.28	46.36

WLS\*: weighted least square with  
pseudorange

EKF\*: Pseudorange/Doppler fusion  
with extended Kalman filter

FGO\*: Pseudorange/Doppler fusion  
with factor graph optimization



# Supplementary: GNSS/INS Integration Using Kalman Filtering

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# Inertial navigation system

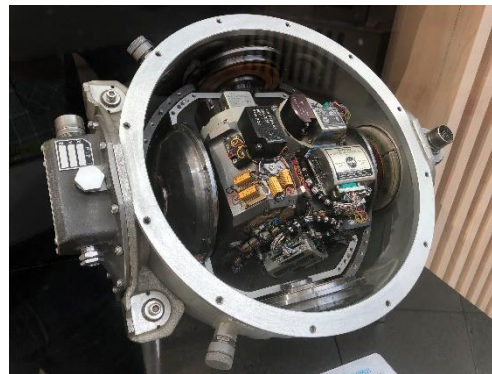
$\hat{\mathbf{a}}_t$ ,  $\hat{\boldsymbol{\omega}}_t$  are the raw accelerometer and gyroscope measurements in the body frame  
 $\mathbf{a}_t$ ,  $\boldsymbol{\omega}_t$  are expected measurements

The cap ^ denotes the noisy measurement or estimation of a certain quantity

$$\hat{\mathbf{a}}_t = \mathbf{a}_t + \mathbf{R}_w^t \mathbf{g}^w + \mathbf{b}_{a_t} + \mathbf{n}_a \quad (1)$$

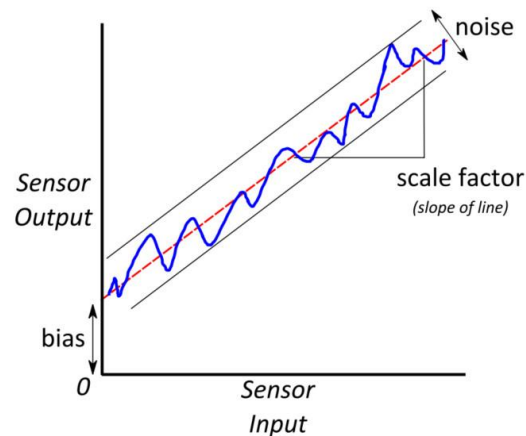
$$\hat{\boldsymbol{\omega}}_t = \boldsymbol{\omega}_t + \mathbf{b}_{\omega_t} + \mathbf{n}_\omega \quad (2)$$

$$\mathbf{n}_a \sim \mathcal{N}(0, \sigma_a^2), \mathbf{n}_\omega \sim \mathcal{N}(0, \sigma_\omega^2)$$



# Error analysis of inertial navigation system

- > The errors of **accelerometer** and **gyroscope** can be divided into:  
**deterministic error & random error.**
- > Deterministic errors can be calibrated in advance including bias, scale...
- > Random error usually assumes that noise obeys Gaussian distribution, including Gaussian white noise, bias random walk...



## Common systematic errors in IMU

bias

noise

scale factor

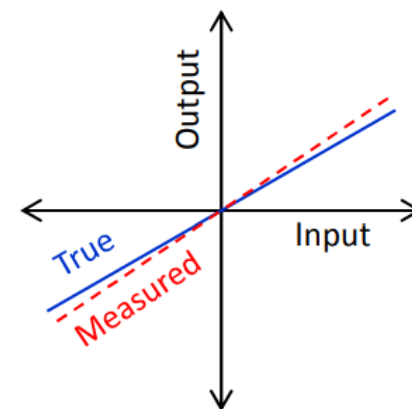
# Deterministic error

(sourcing imperfectness of electrical/mechanical components)

- > **Bias**: In theory, the output of the IMU sensor should be 0 when there is no external action. However, there is a bias **b** to the international data. Influence of accelerometer bias on orientation estimation:

$$\mathbf{V}_{\text{error}} = \mathbf{b}_a t, \mathbf{P}_{\text{error}} = \frac{1}{2} \mathbf{b}_a t^2$$

- > **Scale**: The ratio between the actual value and the sensor output value.





# Deterministic error (sourcing imperfectness of installation)

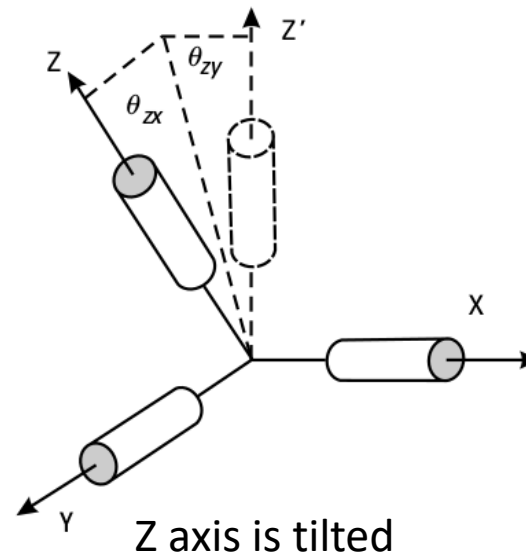
- > **Nonorthogonality/Misalignment** Errors: When manufacturing multi-axis IMU sensors, due to the manufacturing process, the xyz axis may not be vertical.

Scale + Misalignment

$$\begin{bmatrix} l_{ax} \\ l_{ay} \\ l_{az} \end{bmatrix} = \begin{bmatrix} s_{xx} & m_{xy} & m_{xz} \\ m_{yz} & s_{yy} & m_{yz} \\ m_{zx} & m_{zy} & s_{zz} \end{bmatrix} \begin{bmatrix} a_x \\ a_y \\ a_z \end{bmatrix}$$

Measured Acc

True Acc





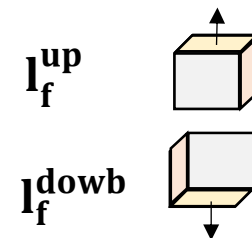
# Deterministic error calibration method—Accelerometer

- > The six-sided method means that the three axes of the accelerometer are placed horizontally up or down for a period of time, and data on the six sides are collected to complete the calibration.

If the axes are orthogonal, it is easy to get bias and scale:

$$\mathbf{b}_a = \frac{\mathbf{l}_f^{\text{up}} + \mathbf{l}_f^{\text{down}}}{2}$$

$$\mathbf{s}_a = \frac{\mathbf{l}_f^{\text{up}} - \mathbf{l}_f^{\text{down}}}{2 \cdot \mathbf{g}} = \begin{bmatrix} s_{a,xx} \\ s_{a,yy} \\ s_{a,zz} \end{bmatrix}$$



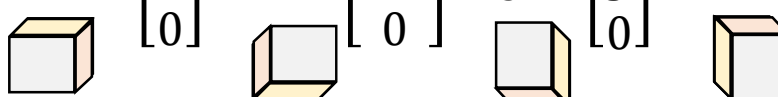
$\mathbf{l}$  is the measured value of a certain axis of the accelerometer,  $\mathbf{g}$  is the local gravity acceleration

# Deterministic error calibration method—Accelerometer

- > When considering the inter-axis error, the relationship between the actual acceleration and the measured value is:

$$\begin{bmatrix} l_{ax} \\ l_{ay} \\ l_{az} \end{bmatrix} = \begin{bmatrix} s_{xx} & m_{xy} & m_{xz} \\ m_{yz} & s_{yy} & m_{yz} \\ m_{zx} & m_{zy} & s_{zz} \end{bmatrix} \begin{bmatrix} a_x \\ a_y \\ a_z \end{bmatrix} + \begin{bmatrix} b_{ax} \\ b_{ay} \\ b_{az} \end{bmatrix}$$

- > When placed horizontally and statically on 6 sides, the theoretical value of acceleration is

$$\mathbf{a}_1 = \begin{bmatrix} g \\ 0 \\ 0 \end{bmatrix}, \mathbf{a}_2 = \begin{bmatrix} -g \\ 0 \\ 0 \end{bmatrix}, \mathbf{a}_3 = \begin{bmatrix} 0 \\ g \\ 0 \end{bmatrix}, \mathbf{a}_4 = \begin{bmatrix} 0 \\ -g \\ 0 \end{bmatrix}, \mathbf{a}_5 = \begin{bmatrix} 0 \\ 0 \\ g \end{bmatrix}, \mathbf{a}_6 = \begin{bmatrix} 0 \\ 0 \\ -g \end{bmatrix}$$


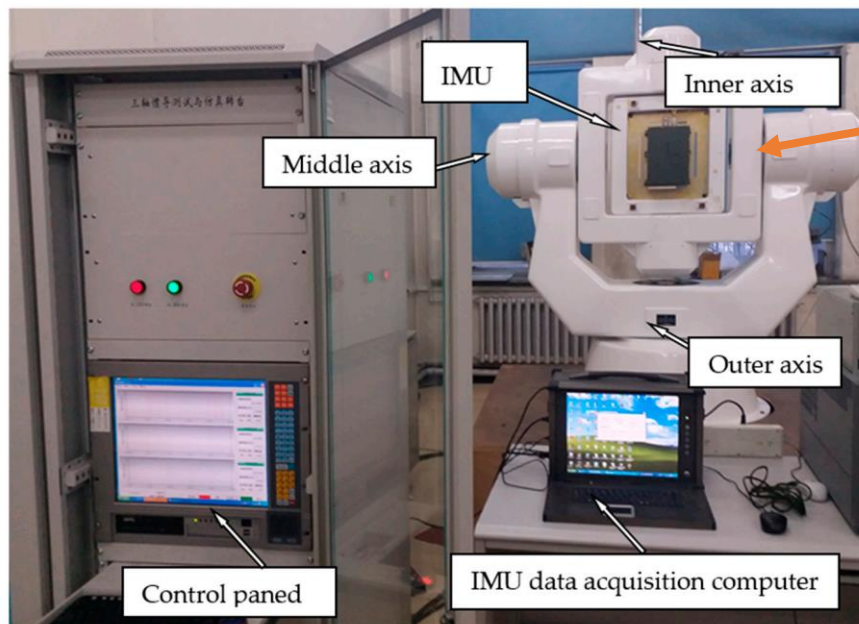
- > Corresponding measurement value matrix  $\mathbf{L}$

$$\mathbf{L} = [l_1 \quad l_2 \quad l_3 \quad l_4 \quad l_5 \quad l_6]$$

- > 12 variables can be obtained by using least squares.

# Deterministic error calibration method——Gyroscope

- > Unlike the six-sided method of accelerometer, the true value of the gyroscope is provided by a high-precision turntable. The 6 faces in this refer to the clockwise and counterclockwise rotation of each axis

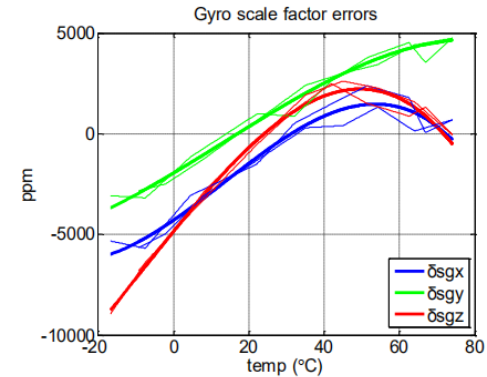
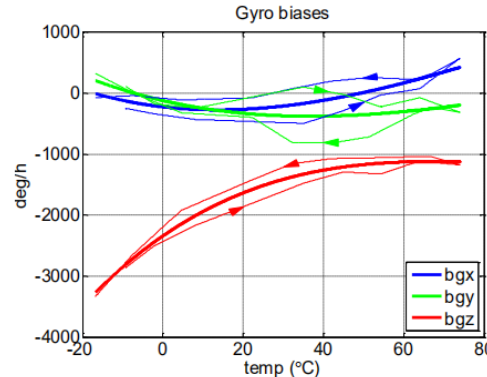


high-precision  
three-axis turntable

# Random error – Unstableness of electrical and mechanical component due to temperature

- > We can calibrate to do temperature compensation on the bias and scale estimated by the sensor, and to obtain the values of bias and scale at different temperatures and draw them into a curve.
- > **Soak method**: control the temperature value of **the constant temperature room**, and then read the sensor value for calibration.

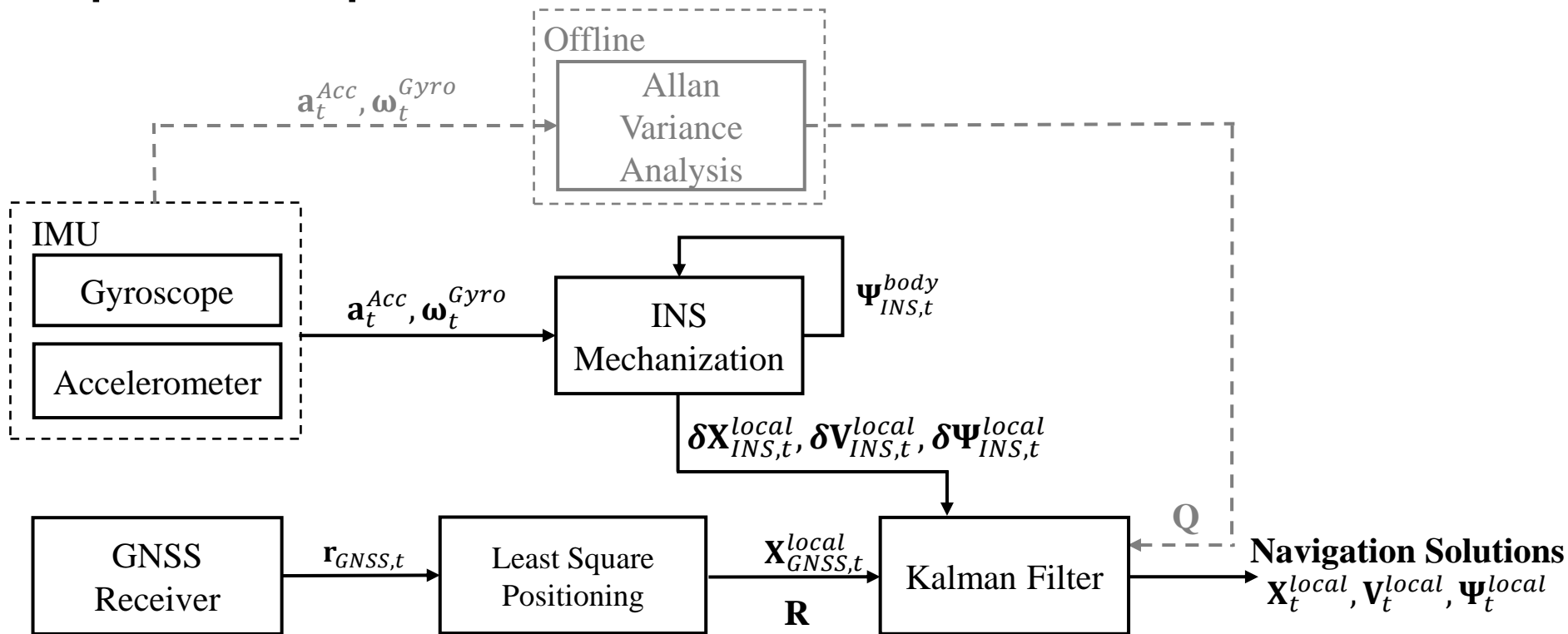
The thin solid lines are the results under separated heating and cooling processes The thick lines are the final curve fitted result .



# Nomenclature

- >  $\mathbf{a}_t^{Acc}$ : measured 3-axis accelerations by the accelerometers at epoch  $t$
- >  $\boldsymbol{\omega}_t^{Gyro}$ : measured 3-axis rotation by the gyroscopes at epoch  $t$
- >  $\mathbf{X}_{INS,t}^{body}$ : estimated 3-axis position in body frame by INS at epoch  $t$
- >  $\mathbf{V}_{INS,t}^{body}$ : estimated 3-axis velocity in body frame by INS at epoch  $t$
- >  $\boldsymbol{\Psi}_{INS,t}^{body}$ : estimated 3-axis orientation in body frame (Euler angles) by INS at epoch  $t$
- >  $\mathbf{B}_{a,t}^{body}$ : estimated 3-axis biases of accelerometers in body frame at epoch  $t$
- >  $\mathbf{B}_{\omega,t}^{body}$ : estimated 3-axis biases of gyroscopes in body frame at epoch  $t$
- >  $\mathbf{W}_{b_a}$ : estimated 3-axis random walk noise of accelerometers in body frame
- >  $\mathbf{W}_{b_\omega}$ : estimated 3-axis random walk noise of gyroscopes in body frame

# Open Loop



# INS Mechanization

Sola, Joan. "Quaternion kinematics for the error-state Kalman filter." *arXiv preprint arXiv:1711.02508* (2017).

The Euler angle rates obtained by angular velocity:

$$\begin{bmatrix} \dot{\phi} \\ \dot{\theta} \\ \dot{\psi} \end{bmatrix} = \Psi_{INS,t}^{body} = \begin{bmatrix} 1 & \sin(\phi) \tan(\theta) & \cos(\phi) \tan(\theta) \\ 0 & \cos(\phi) & -\sin(\phi) \\ 0 & \sin(\phi) \sec(\theta) & \cos(\phi) \sec(\theta) \end{bmatrix} \omega_t^{Gyro} = \begin{bmatrix} p \\ q \\ r \end{bmatrix}$$

$$\delta \Psi_{INS,t}^{body} = \dot{\Psi}_{INS,t}^{body} \Delta t = \begin{bmatrix} \delta \phi \\ \delta \theta \\ \delta \psi \end{bmatrix}$$

Rotate the Euler angles from Body to Local

$$\Psi_{INS,t}^{local} = \mathbf{R}_{body}^{local} \Psi_{INS,t}^{body}$$

Update the current Euler angles

$$\Psi_{INS,t}^{body} = \Psi_{INS,t-1}^{body} + \delta \Psi_{INS,t}^{body}$$

$$\begin{aligned} \mathbf{R}_{body}^{local} &= \mathbf{R}(X, \phi) \mathbf{R}(Y, \theta) \mathbf{R}(Z, \psi) \\ &= \begin{bmatrix} 1 & 0 & 0 \\ 0 & \cos(\phi) & -\sin(\phi) \\ 0 & \sin(\phi) & \cos(\phi) \end{bmatrix} \begin{bmatrix} \cos(\theta) & 0 & \sin(\theta) \\ 0 & 1 & 0 \\ -\sin(\theta) & 0 & \cos(\theta) \end{bmatrix} \begin{bmatrix} \cos(\psi) & -\sin(\psi) & 0 \\ \sin(\psi) & \cos(\psi) & 0 \\ 0 & 0 & 1 \end{bmatrix} \end{aligned}$$

# INS Mechanization

To remove the gravity from the acceleration

$$\mathbf{a}_{INS,t}^{local} = \mathbf{R}_{body}^{local} \mathbf{a}_{INS,t}^{body} - \mathbf{g}$$

To obtain the change of aircraft in terms of position, velocity and orientation

$$\delta \mathbf{x}_{INS,t}^{local} = \mathbf{x}_{INS,t}^{local} - \mathbf{x}_{INS,t-1}^{local} = \mathbf{v}_{INS,t-1}^{local} \Delta t + \frac{\mathbf{a}_{INS,t}^{local} \Delta t^2}{2}$$

$$\delta \mathbf{v}_{INS,t}^{local} = \mathbf{v}_{INS,t}^{local} - \mathbf{v}_{INS,t-1}^{local} = \mathbf{a}_{INS,t}^{local} \Delta t$$

$$\delta \Psi_{INS,t}^{local} = \Psi_{INS,t}^{local} - \Psi_{INS,t-1}^{local}$$



# Kalman Filter——GNSS/INS(Open Loop)

System States:

$$\begin{aligned}\mathbf{X}_t &= (\mathbf{x}_t^{local}, \mathbf{v}_t^{local}, \boldsymbol{\Psi}_t^{local}) \\ \mathbf{x}_t^{local} &= (x_t^{local}, y_t^{local}, z_t^{local}) \\ \mathbf{v}_t^{local} &= (vx_t^{local}, vy_t^{local}, vz_t^{local}) \\ \boldsymbol{\Psi}_t^{local} &= (\phi_{roll}, \theta_{pitch}, \psi_{yaw})\end{aligned}$$

Propagation model:

$$\mathbf{X}_t^- = \mathbf{F}\mathbf{X}_{t-1}^+ + \mathbf{B}\mathbf{U}_t$$

$$\mathbf{F} = \begin{bmatrix} 1 & \cdots & 0 \\ \vdots & \ddots & \vdots \\ 0 & \cdots & 1 \end{bmatrix} \quad \mathbf{B} = \begin{bmatrix} 1 & \cdots & 0 \\ \vdots & \ddots & \vdots \\ 0 & \cdots & 1 \end{bmatrix}$$

$$\mathbf{U}_t = \begin{bmatrix} \delta \mathbf{x}_{INS,t}^{local} \\ \delta \mathbf{v}_{INS,t}^{local} \\ \delta \boldsymbol{\Psi}_{INS,t}^{local} \end{bmatrix}$$

Measurement model:

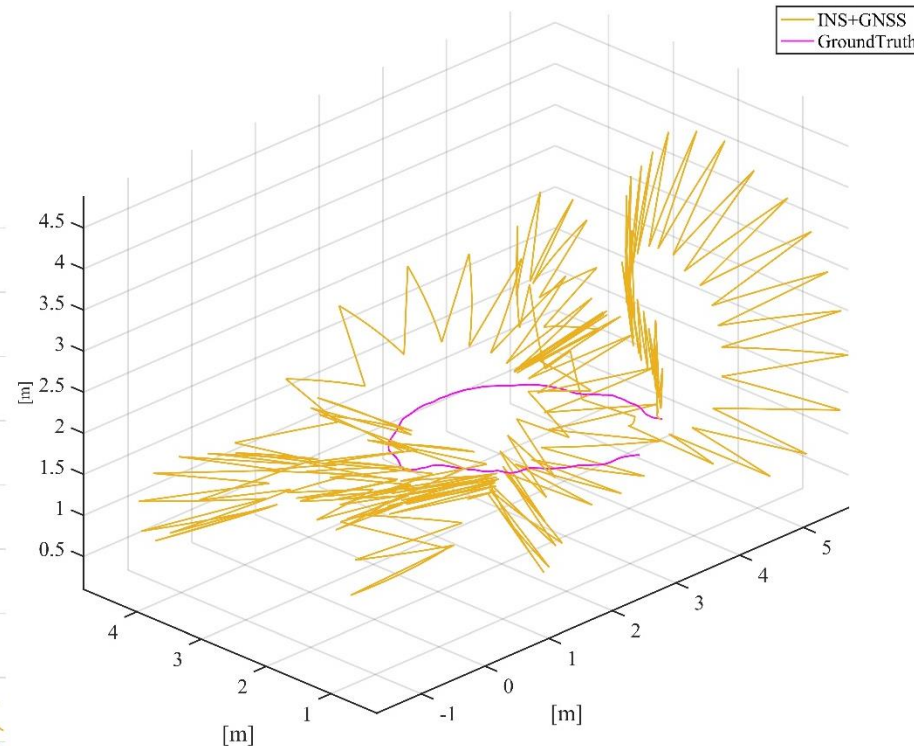
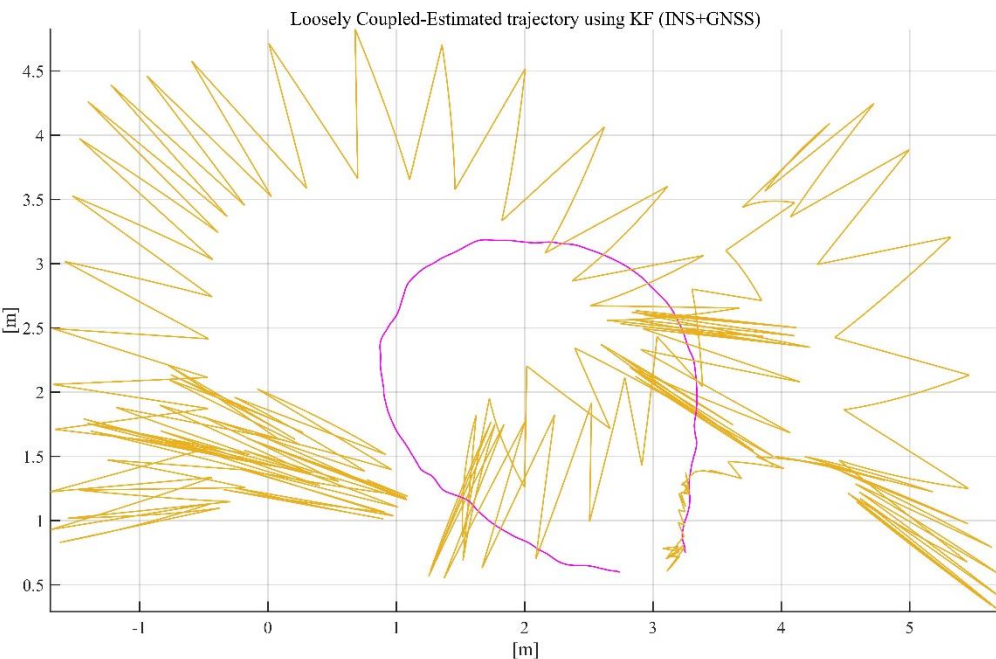
$$\Delta \mathbf{Z}_t = \mathbf{Z}_t - \mathbf{H}\mathbf{X}_t^-$$

$$\mathbf{H} = \begin{bmatrix} \mathbf{I}_{3 \times 3} & \cdots & \mathbf{0} \\ \vdots & \mathbf{0} & \vdots \\ \mathbf{0} & \cdots & \mathbf{0} \end{bmatrix}$$

$$\mathbf{Z}_t = \mathbf{x}_{GNSS,t}^{local} = \begin{bmatrix} x_{GNSS,t}^{local} \\ y_{GNSS,t}^{local} \\ z_{GNSS,t}^{local} \end{bmatrix}$$

# Open Loop

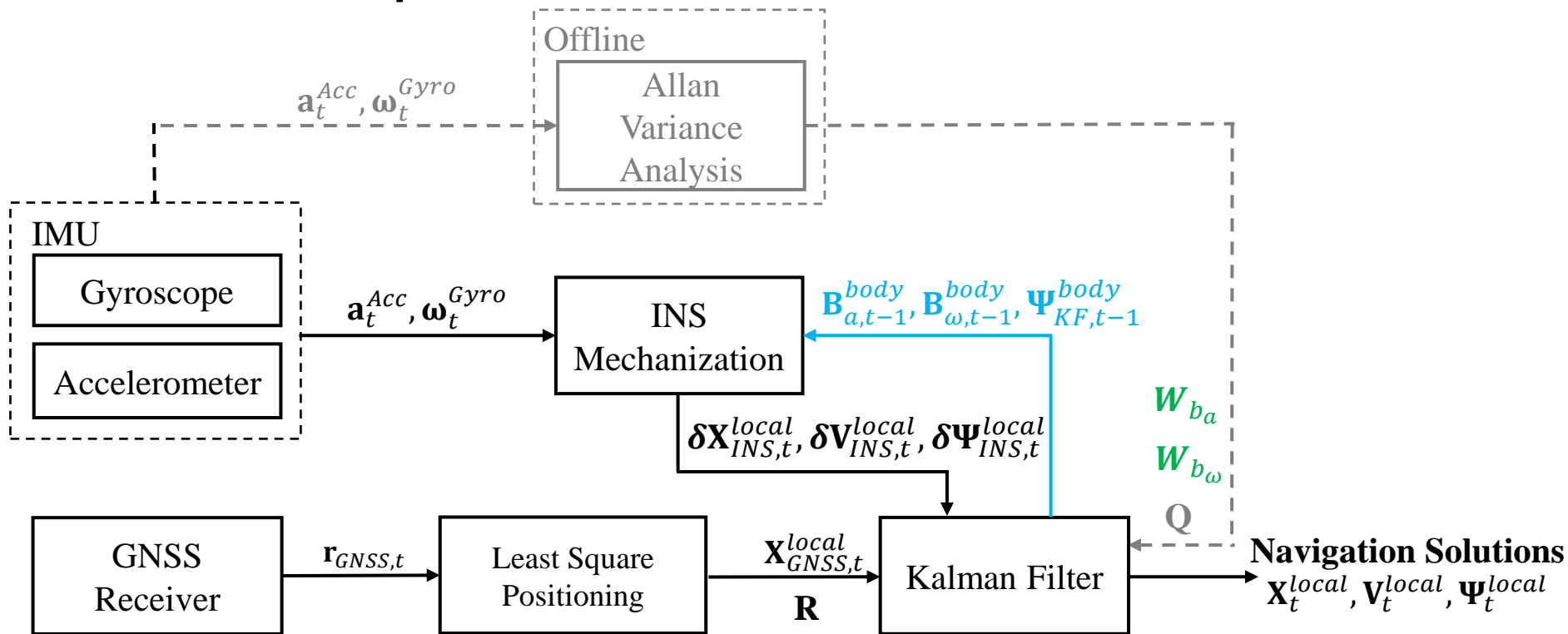
Loosely Coupled-Estimated trajectory using KF (INS+GNSS)



# Closed-loop Correction

- > The estimated position, velocity, and attitude errors are fed back to the inertial navigation processor, where they are used to correct the inertial navigation solution itself.
- > any accelerometer and gyro errors estimated by the Kalman filter are fed back to correct the IMU measurements, as they are input to the inertial navigation equations.
- > Unlike the position, velocity, and attitude corrections, the accelerometer and gyro corrections must be applied on every iteration

# Closed Loop



# INS Mechanization

The Euler angle rates obtained by angular velocity:

$$\begin{bmatrix} \dot{\phi} \\ \dot{\theta} \\ \dot{\psi} \end{bmatrix} = \dot{\Psi}_{INS,t}^{body} = \begin{bmatrix} 1 & \sin(\phi) \tan(\theta) & \cos(\phi) \tan(\theta) \\ 0 & \cos(\phi) & -\sin(\phi) \\ 0 & \sin(\phi) \sec(\theta) & \cos(\phi) \sec(\theta) \end{bmatrix} (\omega_t^{Gyro} - \mathbf{B}_{\omega,t-1}^{body})$$

$$\delta \Psi_{INS,t}^{body} = \dot{\Psi}_{INS,t}^{body} \Delta t = \begin{bmatrix} \delta \phi \\ \delta \theta \\ \delta \psi \end{bmatrix}$$

Rotate the Euler angles from Body to Local

$$\Psi_{INS,t}^{local} = \mathbf{R}_{body}^{local} \Psi_{INS,t}^{body}$$

Update the current Euler angles

$$\Psi_{INS,t}^{body} = \Psi_{KF,t-1}^{body} + \delta \Psi_{INS,t}^{body}$$

$$\Psi_{KF,t-1}^{local} = \mathbf{R}_{body}^{local} \Psi_{KF,t-1}^{body}$$

$$\begin{aligned} \mathbf{R}_{body}^{local} &= \mathbf{R}(X, \phi) \mathbf{R}(Y, \theta) \mathbf{R}(Z, \psi) \\ &= \begin{bmatrix} 1 & 0 & 0 \\ 0 & \cos(\phi) & -\sin(\phi) \\ 0 & \sin(\phi) & \cos(\phi) \end{bmatrix} \begin{bmatrix} \cos(\theta) & 0 & \sin(\theta) \\ 0 & 1 & 0 \\ -\sin(\theta) & 0 & \cos(\theta) \end{bmatrix} \begin{bmatrix} \cos(\psi) & -\sin(\psi) & 0 \\ \sin(\psi) & \cos(\psi) & 0 \\ 0 & 0 & 1 \end{bmatrix} \end{aligned}$$

# INS Mechanization

To remove the gravity from the acceleration

$$\mathbf{a}_{INS,t}^{local} = \mathbf{R}_{body}^{local} (\mathbf{a}_{INS,t}^{body} - \mathbf{B}_{a,t-1}^{body}) - \mathbf{g}$$

To obtain the change of aircraft in terms of position, velocity and orientation

$$\delta \mathbf{x}_{INS,t}^{body} = \mathbf{v}_{INS,t-1}^{local} \Delta t + \frac{\mathbf{a}_{INS,t}^{local} \Delta t^2}{2}$$

$$\delta \mathbf{v}_{INS,t}^{local} = \mathbf{a}_{INS,t}^{local} \Delta t$$

$$\delta \Psi_{INS,t}^{local} = \Psi_{KF,t-1}^{local} - \Psi_{INS,t-1}^{local}$$

# Kalman Filter——GNSS/INS (Closed Loop)

System States:

$$\mathbf{X}_t = (\mathbf{x}_t^{local}, \mathbf{v}_t^{local}, \boldsymbol{\Psi}_t^{local}, \mathbf{B}_{a,t}^{body}, \mathbf{B}_{\omega,t}^{body})$$

$$\mathbf{x}_t^{local} = (x_t^{local}, y_t^{local}, z_t^{local})$$

$$\mathbf{v}_t^{local} = (vx_t^{local}, vy_t^{local}, vz_t^{local})$$

$$\boldsymbol{\Psi}_t^{local} = (\phi_{roll}, \theta_{pitch}, \psi_{yaw})$$

$$\mathbf{B}_{a,t}^{body} = (b_{ax,t}^{body}, b_{ay,t}^{body}, b_{az,t}^{body})$$

$$\mathbf{B}_{\omega,t}^{body} = (b_{\omega x,t}^{body}, b_{\omega y,t}^{body}, b_{\omega z,t}^{body})$$

Propagation model:

$$\mathbf{X}_t^- = \mathbf{F}\mathbf{X}_{t-1}^+ + \mathbf{B}\mathbf{U}_t$$

$$\mathbf{F} = \begin{bmatrix} 1 & \cdots & 0 \\ \vdots & \ddots & \vdots \\ 0 & \cdots & 1 \end{bmatrix} \quad \mathbf{B} = \begin{bmatrix} 1 & \cdots & 0 \\ \vdots & \ddots & \vdots \\ 0 & \cdots & 1 \end{bmatrix}$$

$$\mathbf{U}_t = \begin{bmatrix} \delta \mathbf{x}_{INS,t}^{local} \\ \delta \mathbf{v}_{INS,t}^{local} \\ \delta \boldsymbol{\Psi}_{INS,t}^{local} \\ \mathbf{W}_{b_{\omega}} \\ \mathbf{W}_{b_a} \end{bmatrix}$$

Measurement model:

$$\Delta \mathbf{Z}_t = \mathbf{Z}_t - \mathbf{H}\mathbf{X}_t^-$$

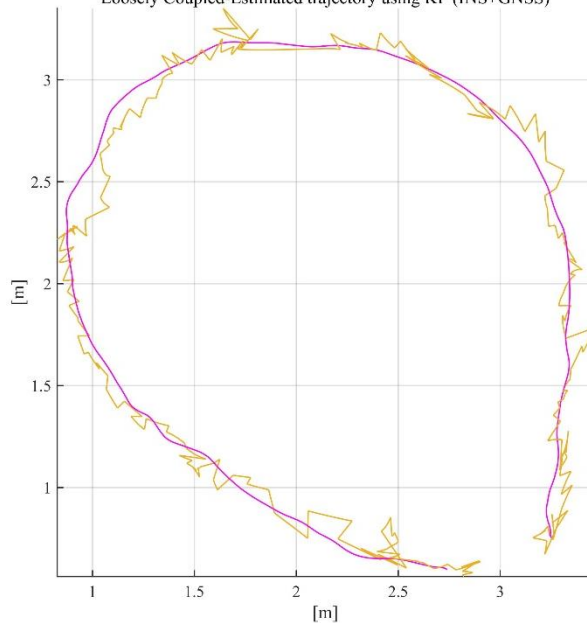
$$\mathbf{H} = \begin{bmatrix} \mathbf{I}_{3 \times 3} & \cdots & \mathbf{0} \\ \vdots & \mathbf{0} & \vdots \\ \mathbf{0} & \cdots & \mathbf{0} \end{bmatrix}$$

$$\mathbf{Z}_t = \mathbf{x}_{GNSS,t}^{local} = \begin{bmatrix} x_{GNSS,t}^{local} \\ y_{GNSS,t}^{local} \\ z_{GNSS,t}^{local} \end{bmatrix}$$

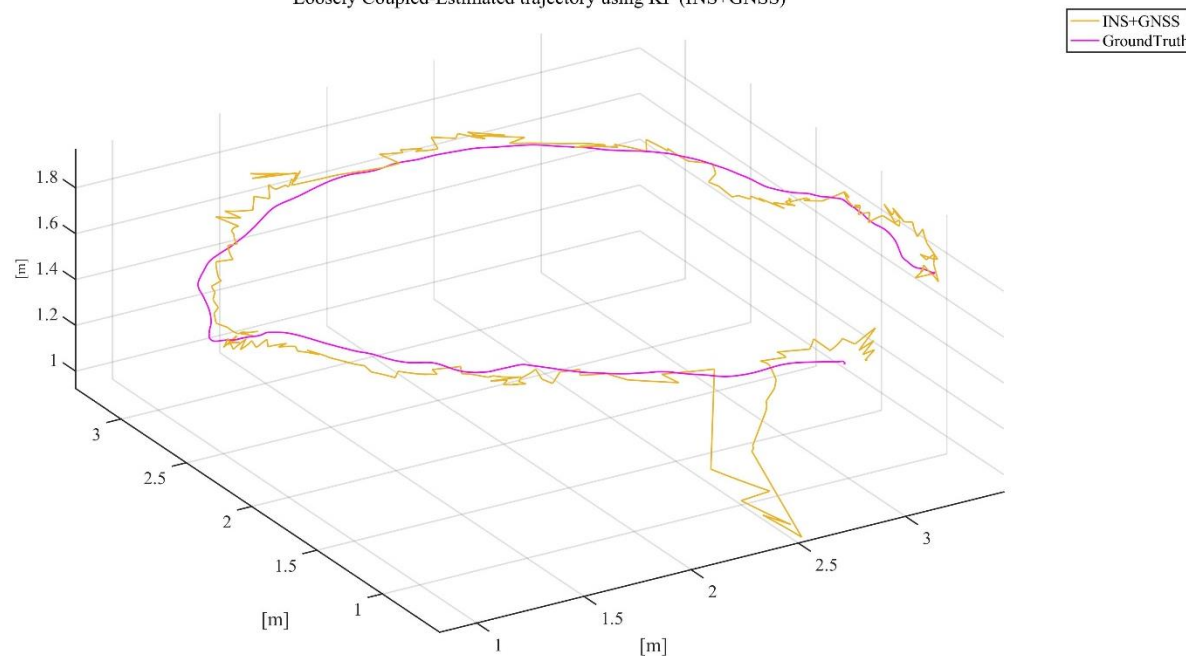


# Closed Loop

Loosely Coupled-Estimated trajectory using KF (INS+GNSS)



Loosely Coupled-Estimated trajectory using KF (INS+GNSS)





# References

- > Chapters 3 and 5, Paul D. Groves, *Principles of GNSS, Inertial, and Multisensor Integrated Navigation Systems, 2nd Edition*, Artech House, 2013.