

# Frequency Domain

## (Part-1)

The terms ***Function*** and ***Signal*** are used interchangeably in the context of Frequency domain processing

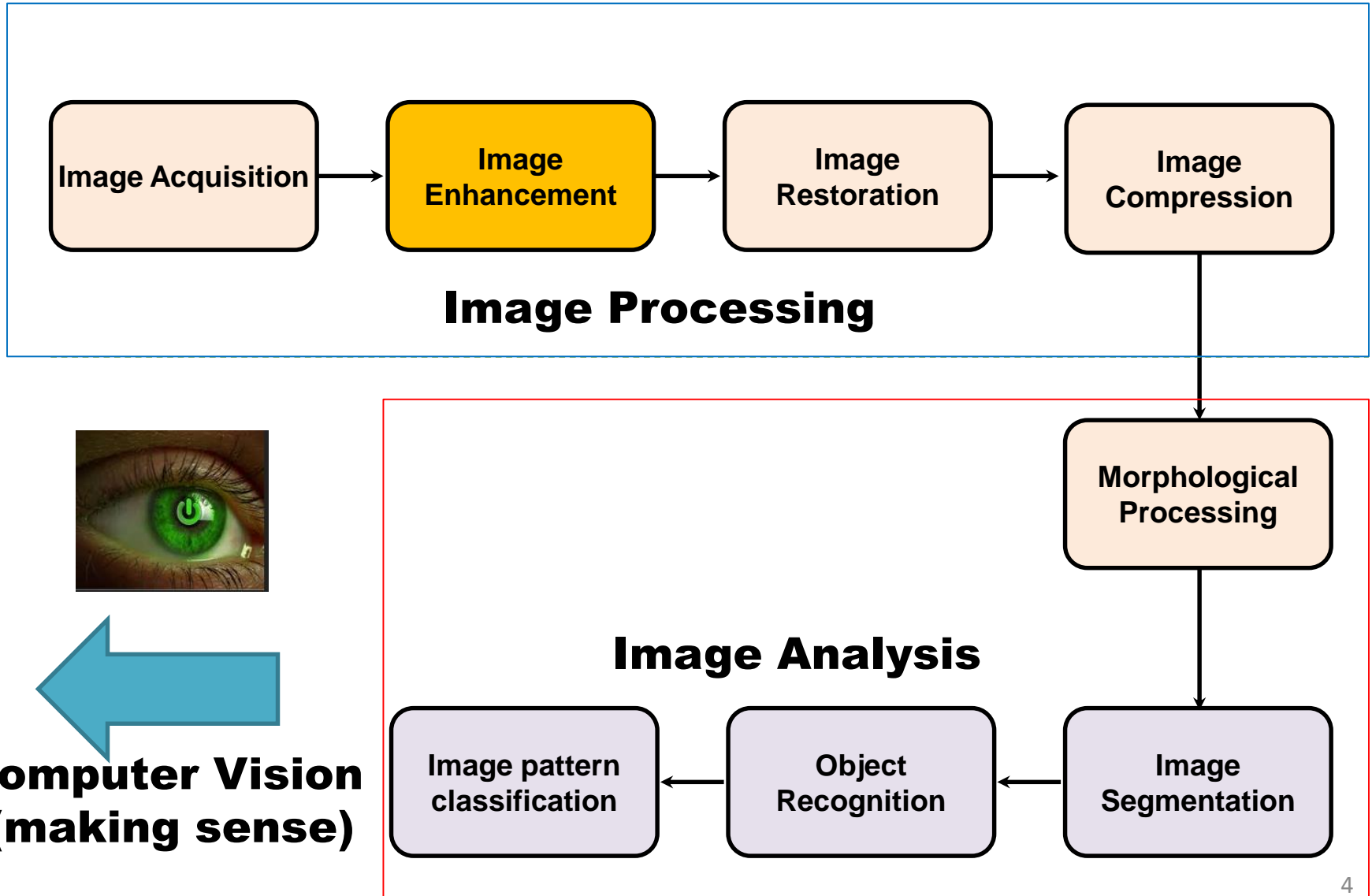
# Recap

- Sharpening spatial filters
  - Foundation of image sharpening
  - Using second-order derivative for image sharpening (Laplacian)
  - Unsharp masking and high boost filtering
  - Using first-order derivative for image sharpening (Gradient)
  - Highpass, Bandreject, and Bandpass filters from lowpass filters
  - Combining spatial enhancement methods

# Lecture Objectives

- Introduction to Frequency Domain
  - Background
  - Sinusoidal Waves
  - Complex Numbers
- Fourier Series
- Impulse
- Fourier Transform
- Convolution of Continuous Functions

# Key Stages in DIP

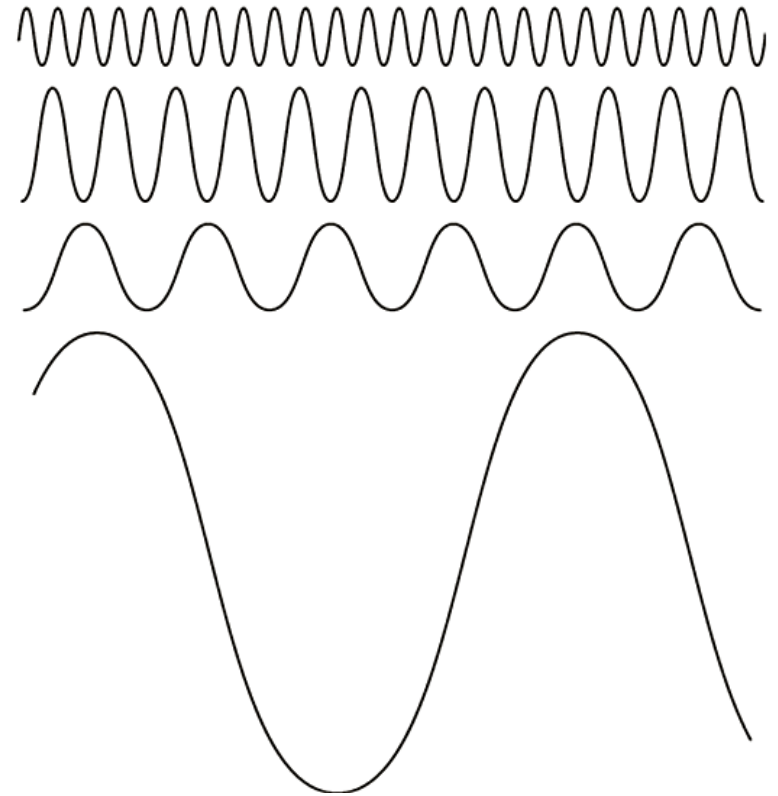


# Background

# Representing Periodic Function as Sines & Cosines



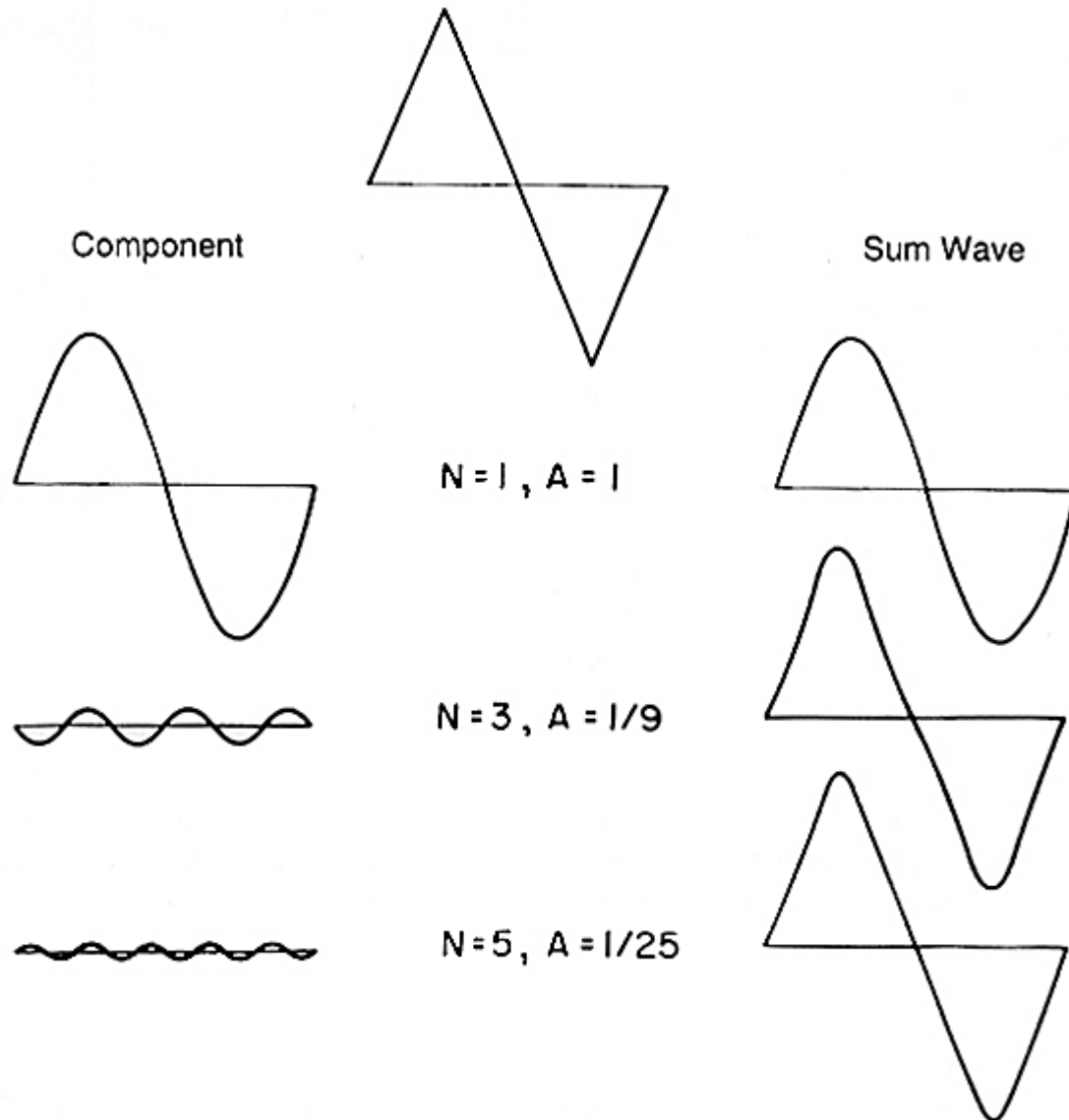
Periodic function



Sines and Cosines

**Fourier series** - Any *periodic function* can be expressed as the **sum of sines and/or cosines** of *different frequencies*, each multiplied by a *different coefficient*.

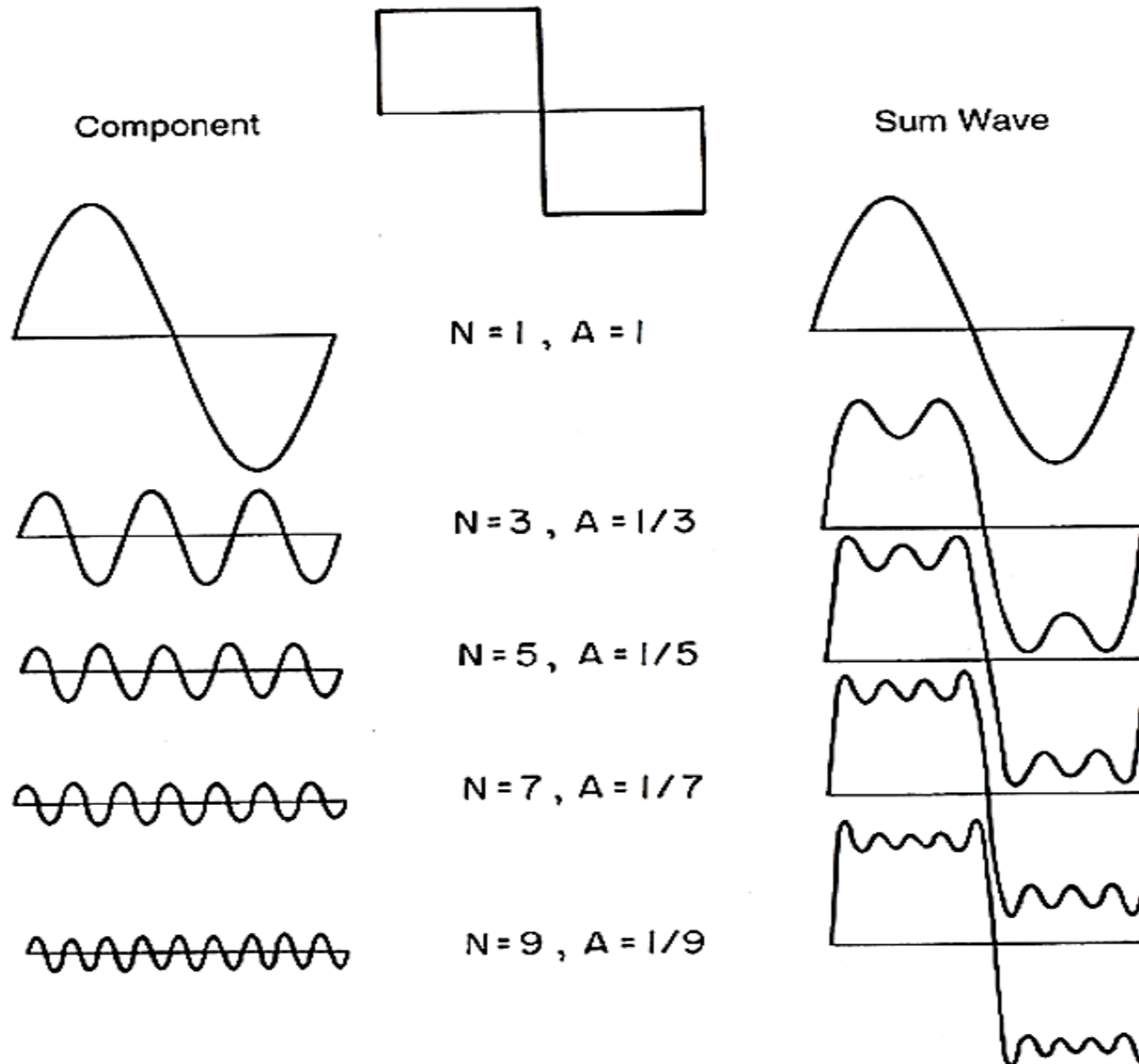
# Representing Periodic Function as Sines & Cosines



Making a triangle wave with a sum of harmonics.

Adding in higher frequencies makes the triangle tips sharper and sharper

# Representing Periodic Function as Sines & Cosines



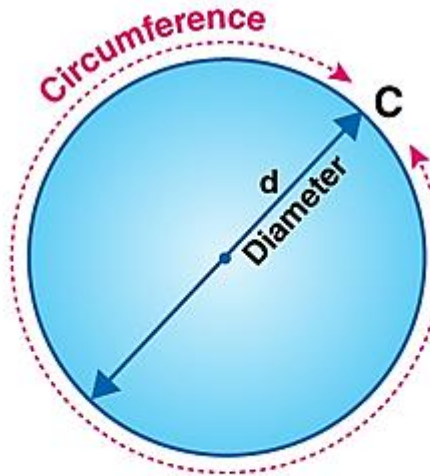


# Sinusoidal Waves

**Remember !!!  $\pi$  radians =  $180^\circ$**

# Why $\pi$ value is 3.14159....

- The value of Pi ( $\pi$ ) is the **ratio** of the *circumference of a circle to its diameter*.

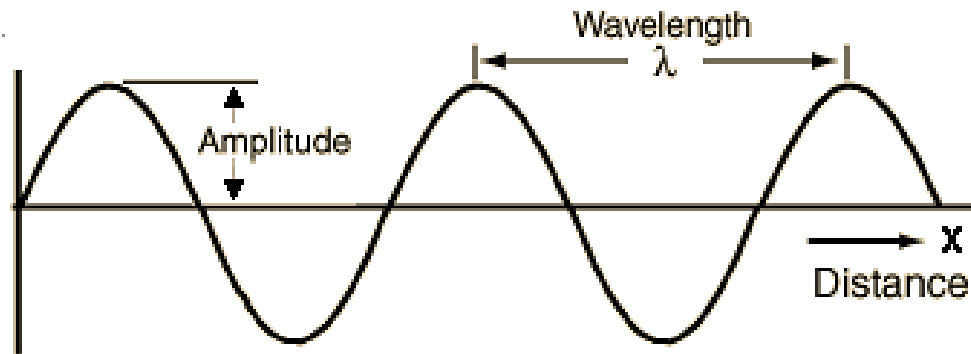


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$$\frac{\text{Circumference}}{\text{Diameter}} = \pi = 3.14159.....$$

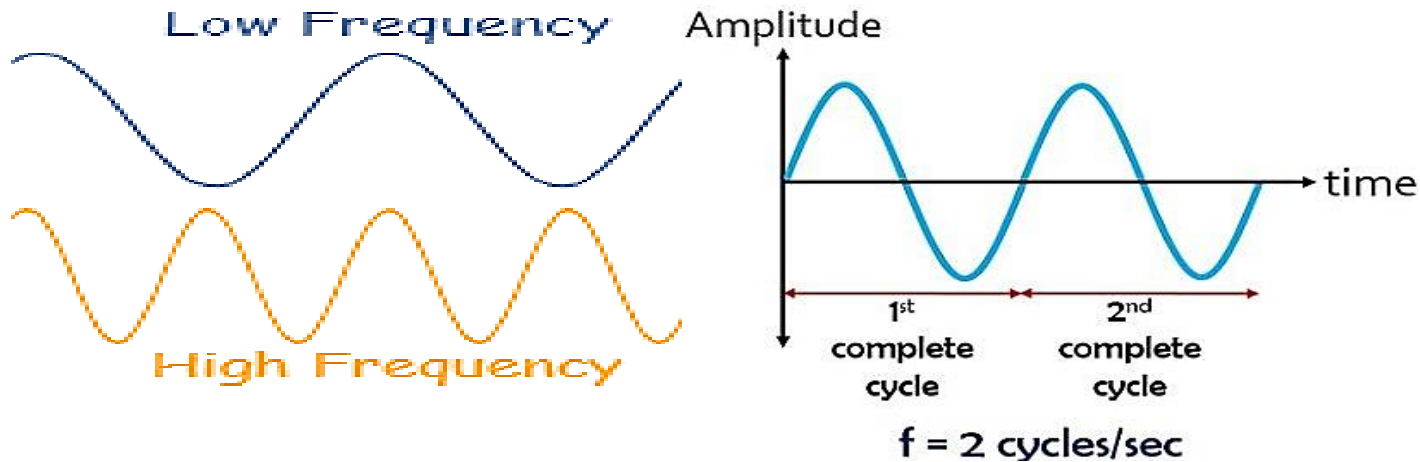
# Properties of a Wave

- Wavelength (  $\lambda$  ) is the length from one peak to the next (or from any point to the next matching point).
- Amplitude (  $A$  ) is the height from the center line to the peak (or to the trough). **OR**, we can measure the height from highest to lowest points and divide that by 2.



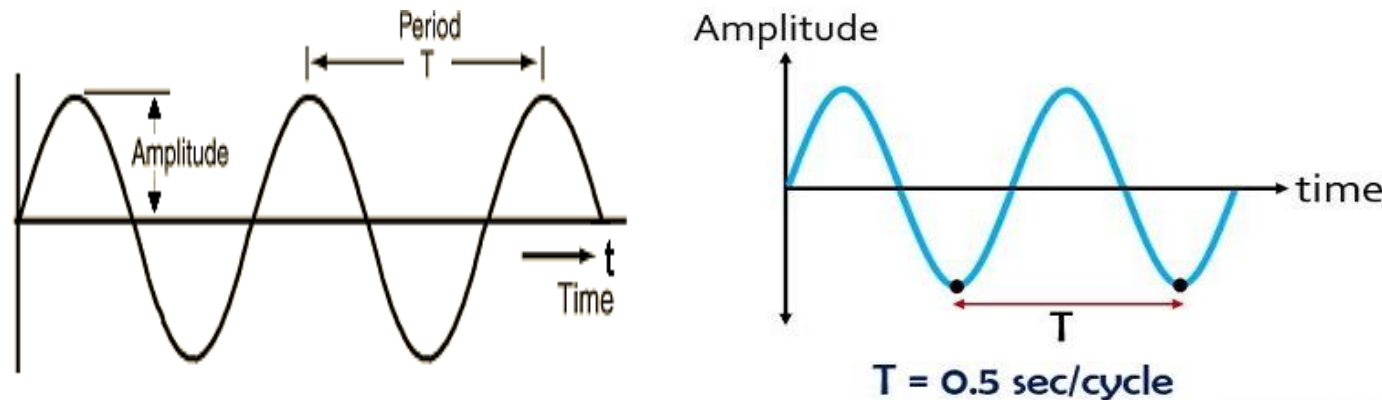
# Properties of a Wave

- Frequency ( $f$ )** describes the number of waves that pass a fixed place in a given amount of time.



$$\text{Frequency} = \frac{1}{\text{Period}}$$

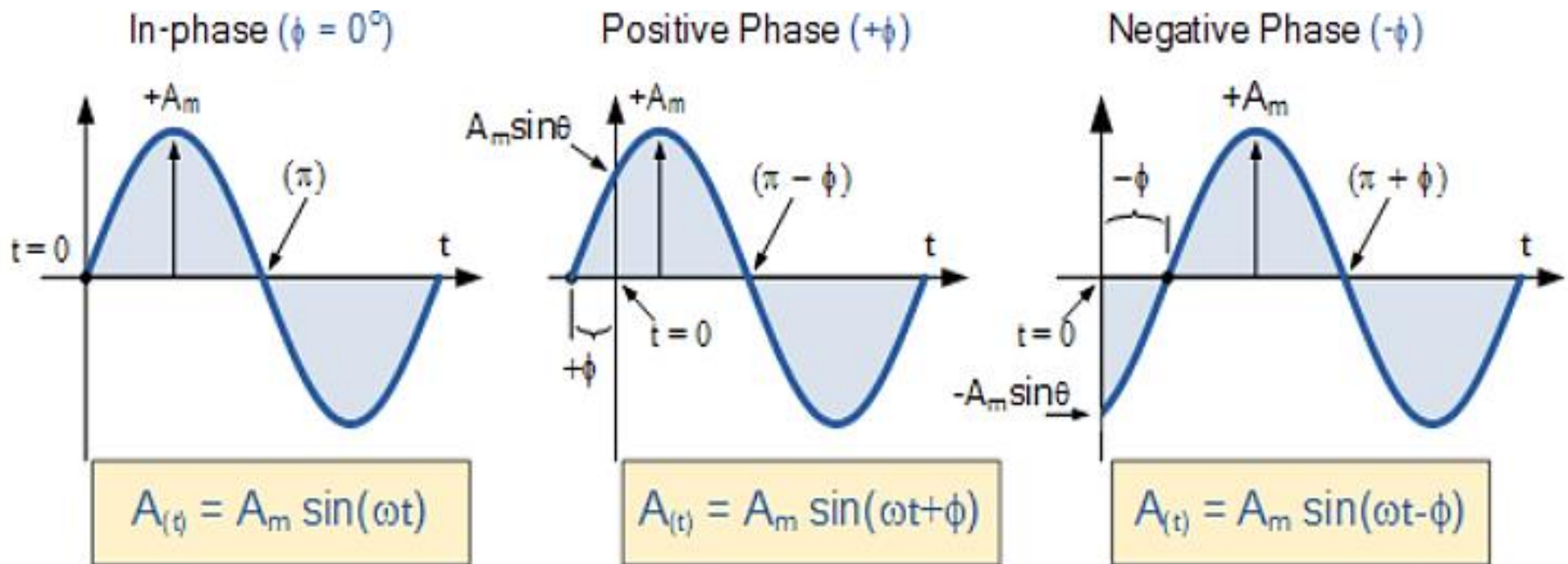
- Period ( $T$ )** is the time it takes to complete one cycle of the wave.



$$\text{Period} = \frac{1}{\text{Frequency}}$$

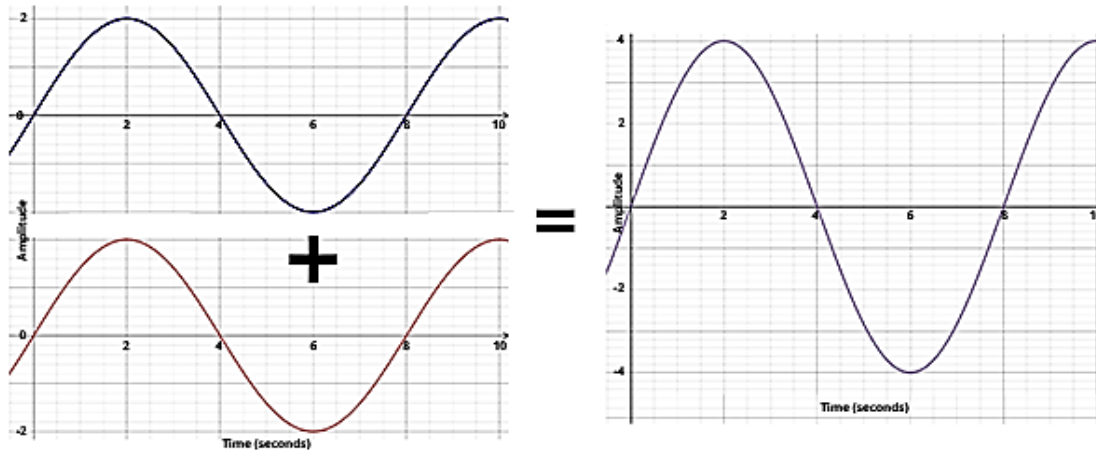
# Properties of a Wave

- Phase ( $\phi$ ) represents an angular shift of a wave(s) and is measured in radians (or degrees).



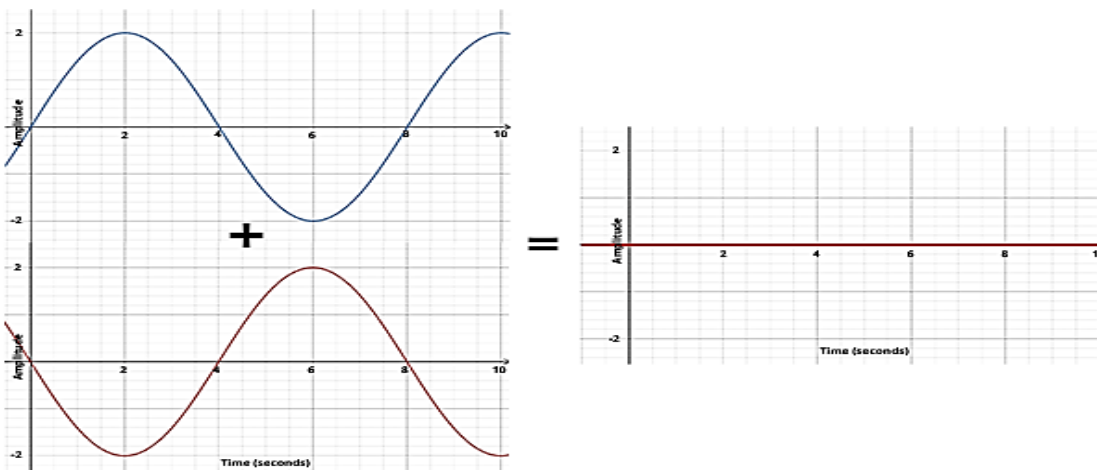
# Properties of a Wave

- Phase Difference - “in-phase”



Waves that are **in-phase** add to produce a wave with an amplitude equal to the sum of the amplitudes of the two waves.

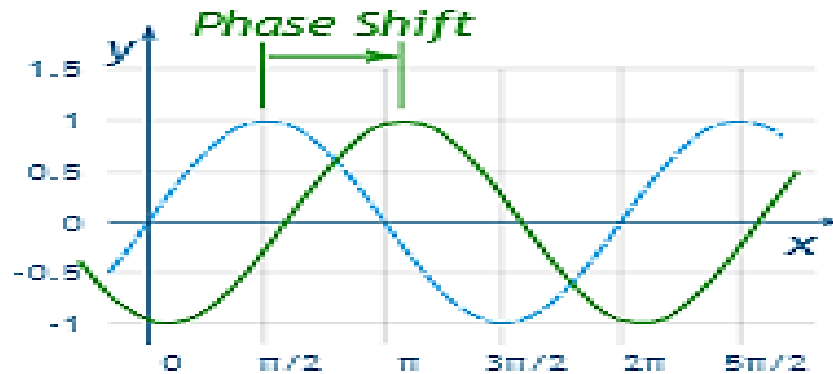
- Phase Difference - “out-of-phase”



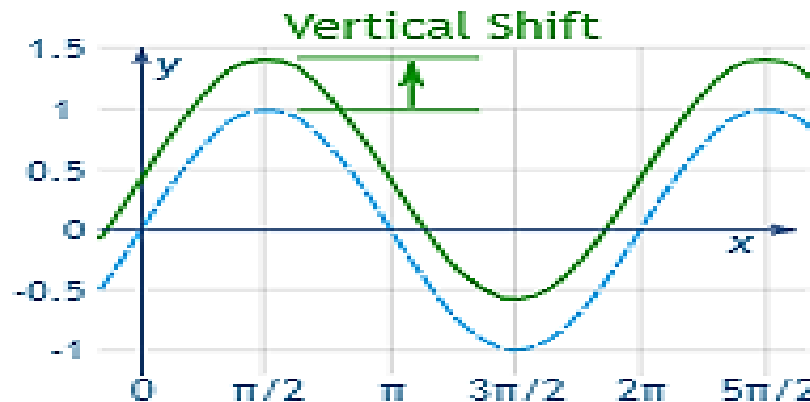
Waves that are **out-of-phase** exactly cancel each other when added together.

# Properties of a Wave

- Phase Shift is how far one wave is shifted **horizontally** from the other wave.



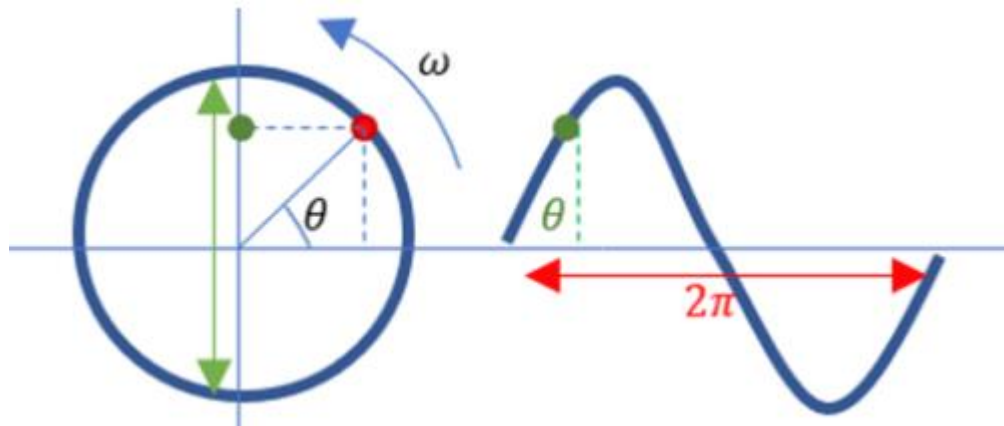
- Vertical Shift is how far one wave is shifted **vertically** from the other wave.



# Properties of a Wave

- Angular velocity ( $\omega$ ) refers to the angular displacement per unit time (for example, in rotation) **OR** the rate of change of the phase of a sinusoidal waveform measured in degrees (or radians) per second.

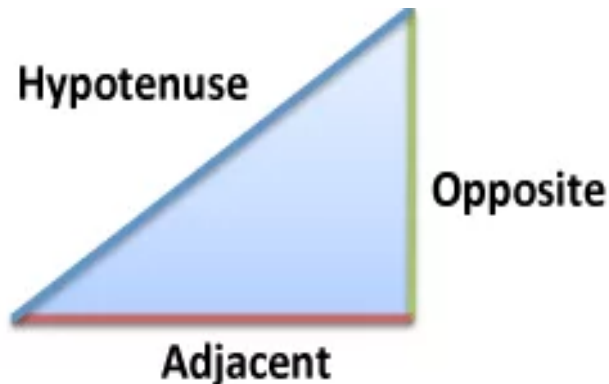
$$\omega = \frac{2\pi}{T} = 2\pi f$$





# Why we need **sine**, **cosine**, **tangent** ?

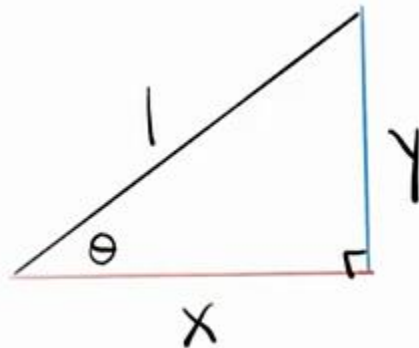
Angles (In Degrees)	0°	30°	45°	60°	90°	180°	270°	360°
Angles (In Radians)	0	$\frac{\pi}{6}$	$\frac{\pi}{4}$	$\frac{\pi}{3}$	$\frac{\pi}{2}$	$\pi$	$\frac{3\pi}{2}$	$2\pi$
sin	0	$\frac{1}{2}$	$\frac{1}{\sqrt{2}}$	$\frac{\sqrt{3}}{2}$	1	0	-1	0
cos	1	$\frac{\sqrt{3}}{2}$	$\frac{1}{\sqrt{2}}$	$\frac{1}{2}$	0	-1	0	1
tan	0	$\frac{1}{\sqrt{3}}$	1	$\sqrt{3}$	Not Defined	0	Not Defined	1



- **sine** is Opposite / Hypotenuse
- **cosine** is Adjacent / Hypotenuse
- **tangent** is Opposite / Adjacent

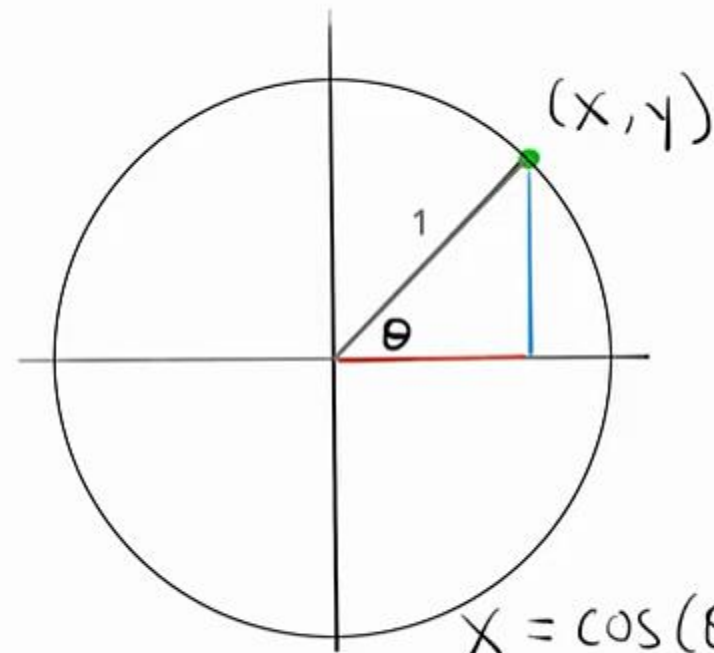
# Why we need Sine, Cosine, Tangent ?

- **sine** and **cosine** give the **value of a point (x, y)** on the circumference of a circle.



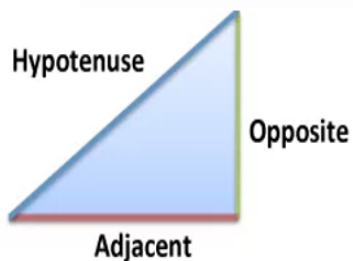
$$\cos(\theta) = \frac{x}{1} = x$$

$$\sin(\theta) = \frac{y}{1} = y$$



$$x = \cos(\theta)$$

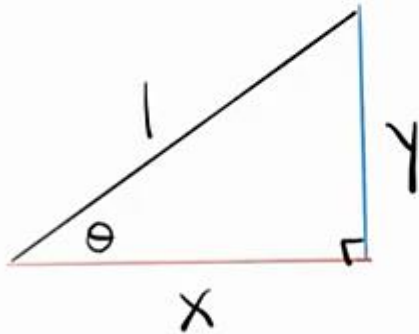
$$y = \sin(\theta)$$



- sine is Opposite / Hypotenuse
- cosine is Adjacent / Hypotenuse
- tangent is Opposite / Adjacent

# Why we need Sine, Cosine, Tangent ?

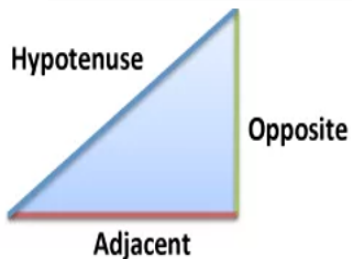
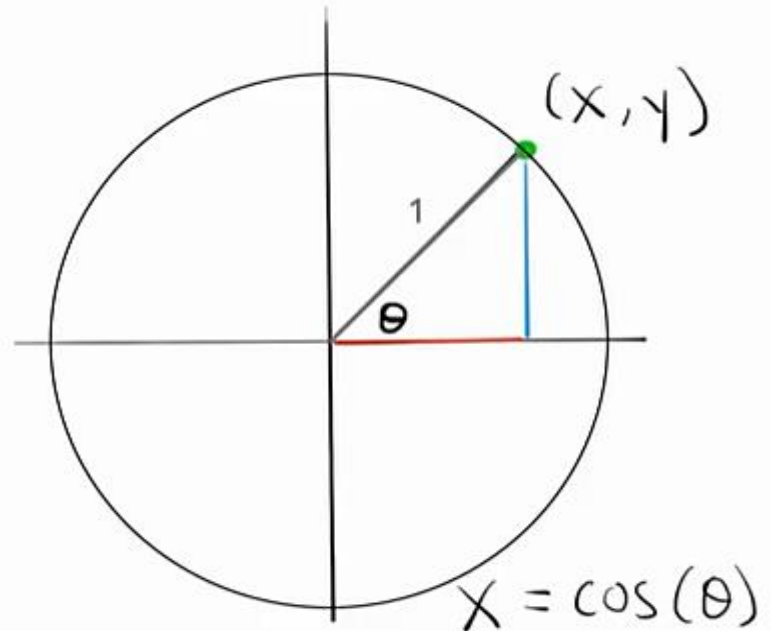
- **tangent** gives the **slope**.



$$\cos(\theta) = \frac{x}{l} = x$$

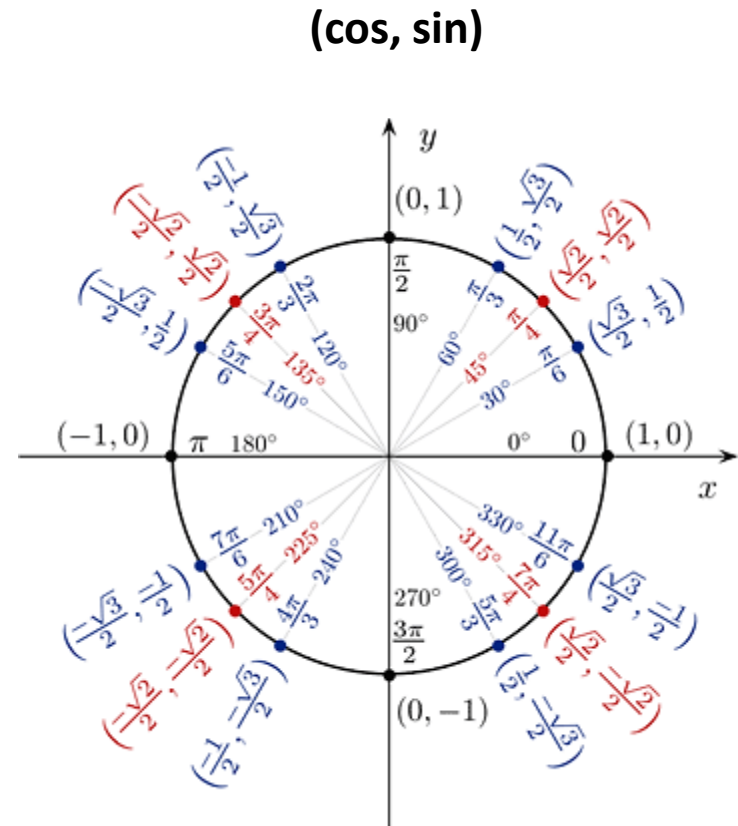
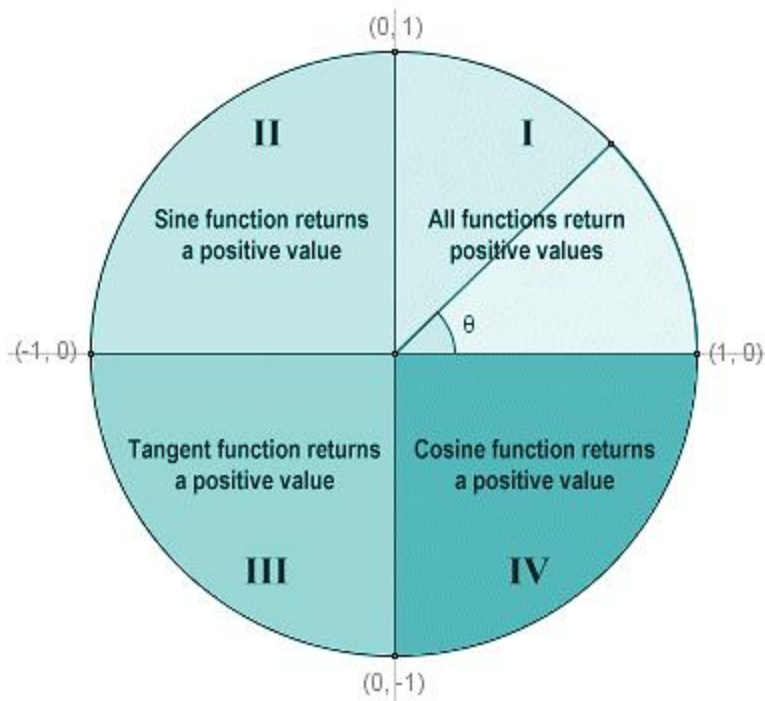
$$\sin(\theta) = \frac{y}{l} = y$$

$$\tan(\theta) = \frac{y}{x} = \frac{\text{up}}{\text{over}} = \frac{\text{rise}}{\text{run}} = \text{slope}$$



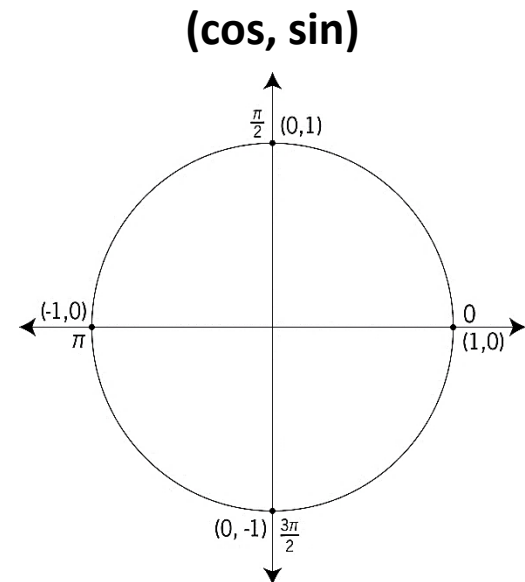
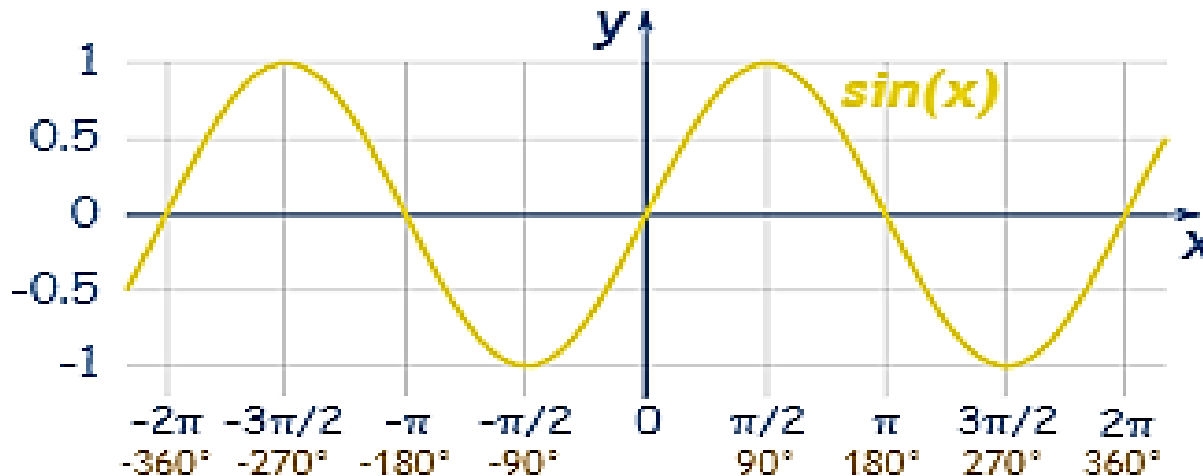
- sine is Opposite / Hypotenuse
- cosine is Adjacent / Hypotenuse
- tangent is Opposite / Adjacent

# Why we need Sine, Cosine, Tangent ?



# What is a Sine Wave ?

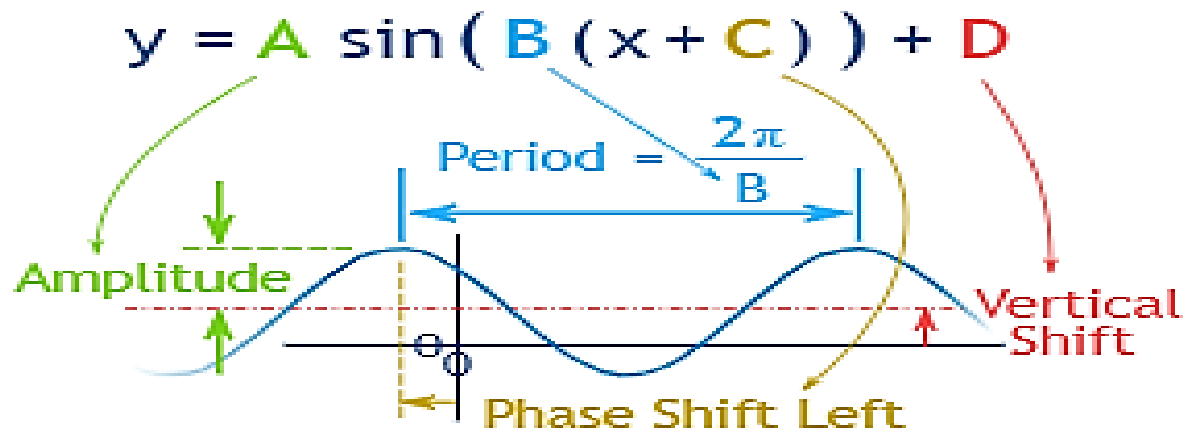
- A basic **sine** wave is an *S-shaped waveform* defined by the mathematical function  **$y = \sin(x)$** .
- The **sine** function has up-down curve (which repeats every  **$2\pi$**  radians, or  **$360^\circ$** ).



- For every cycle, it starts at **0**, heads up to **1** by  $\pi/2$  radians ( **$90^\circ$** ) and then heads down to **-1** and finally heads up to **0**.

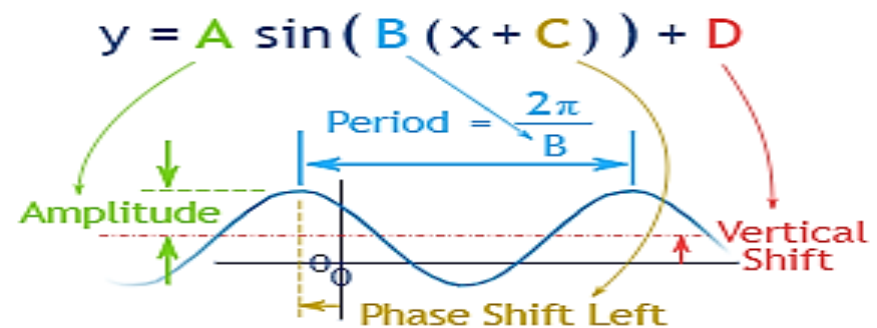
# What is Sine Wave ?

- The general equation of a **sine** wave is given by:



- amplitude is **A**
- period is  **$2\pi/B$**  (defines **periodicity**)
- phase shift is **C** (positive is to the **left**)
- vertical shift is **D**

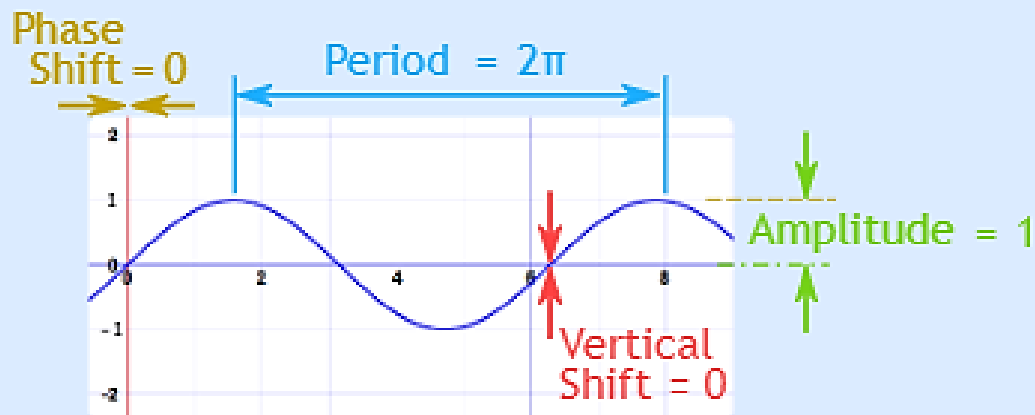
# Sine wave - examples



Example:  $\sin(x)$

This is the basic unchanged sine formula.  $A = 1$ ,  $B = 1$ ,  $C = 0$  and  $D = 0$

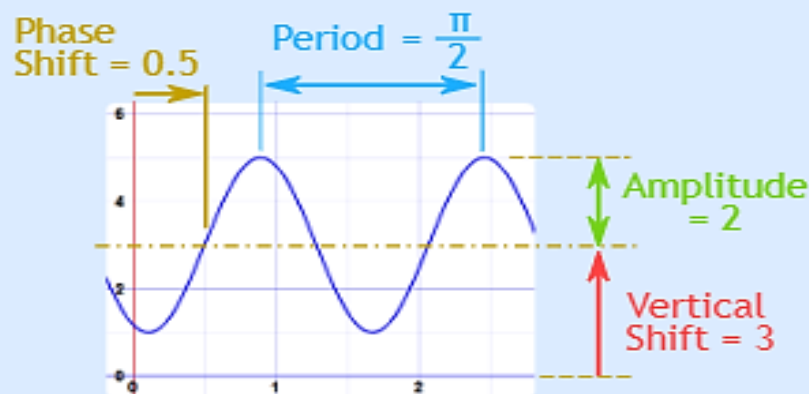
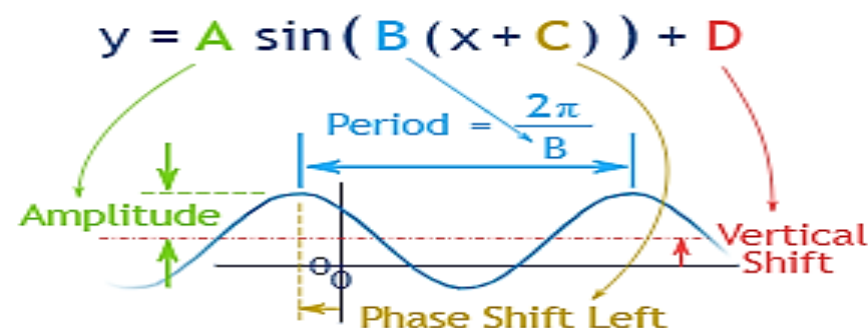
So amplitude is **1**, period is  **$2\pi$** , there is no phase shift or vertical shift:



# Sine wave - examples

Example:  $2 \sin(4(x - 0.5)) + 3$

- amplitude  $A = 2$
- period  $2\pi/B = 2\pi/4 = \pi/2$
- phase shift =  $-0.5$  (or  $0.5$  to the right)
- vertical shift  $D = 3$



In words:

- the **2** tells us it will be 2 times taller than usual, so Amplitude = 2
- the usual period is  $2\pi$ , but in our case that is "sped up" (made shorter) by the **4** in  $4x$ , so Period =  $\pi/2$
- and the  $-0.5$  means it will be shifted to the **right** by **0.5**
- lastly the  $+3$  tells us the center line is  $y = +3$ , so Vertical Shift = 3



# Sine wave - examples

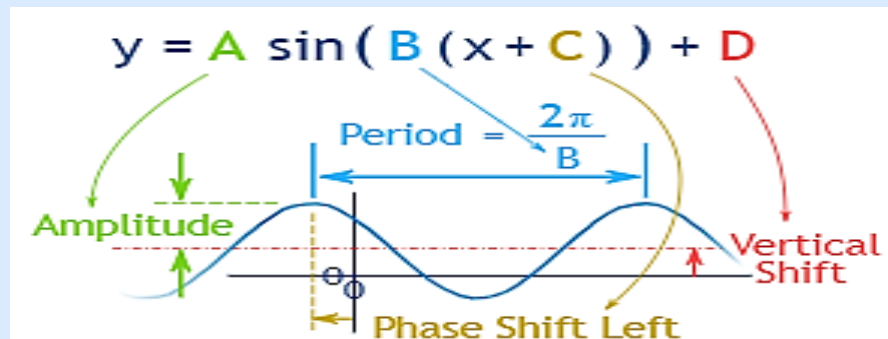
Example:  $3 \sin(100(t + 0.01))$

First we need brackets around the  $(t+1)$ , so we can start by dividing the 1 by 100:

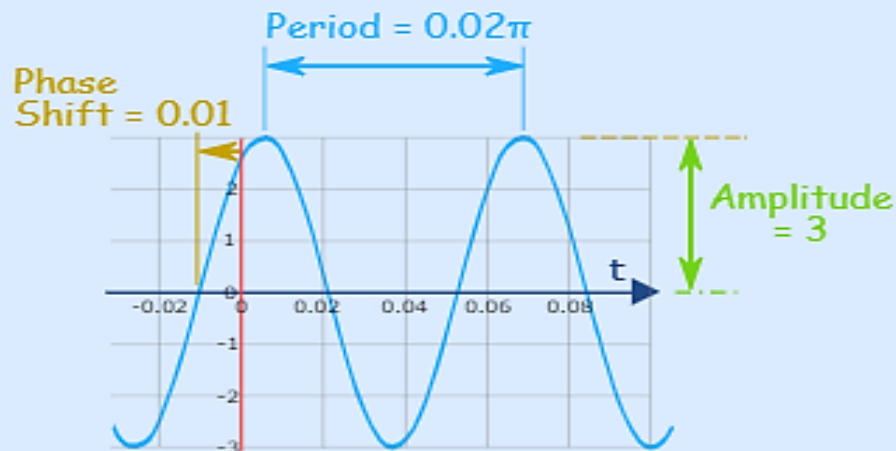
$$3 \sin(100t + 1) = 3 \sin(100(t + 0.01))$$

Now we can see:

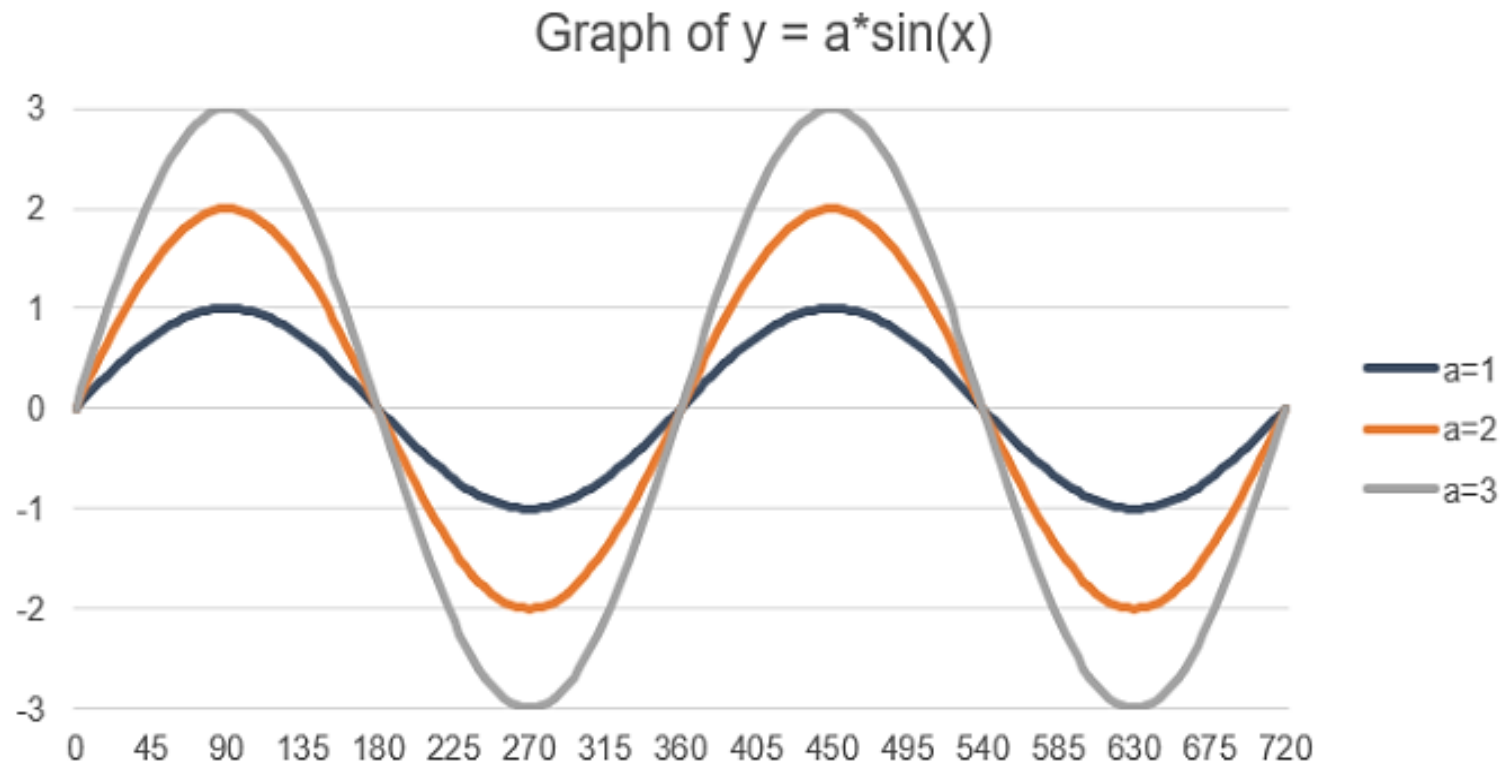
- amplitude is  $A = 3$
- period is  $2\pi/100 = 0.02\pi$
- phase shift is  $C = 0.01$  (to the left)
- vertical shift is  $D = 0$



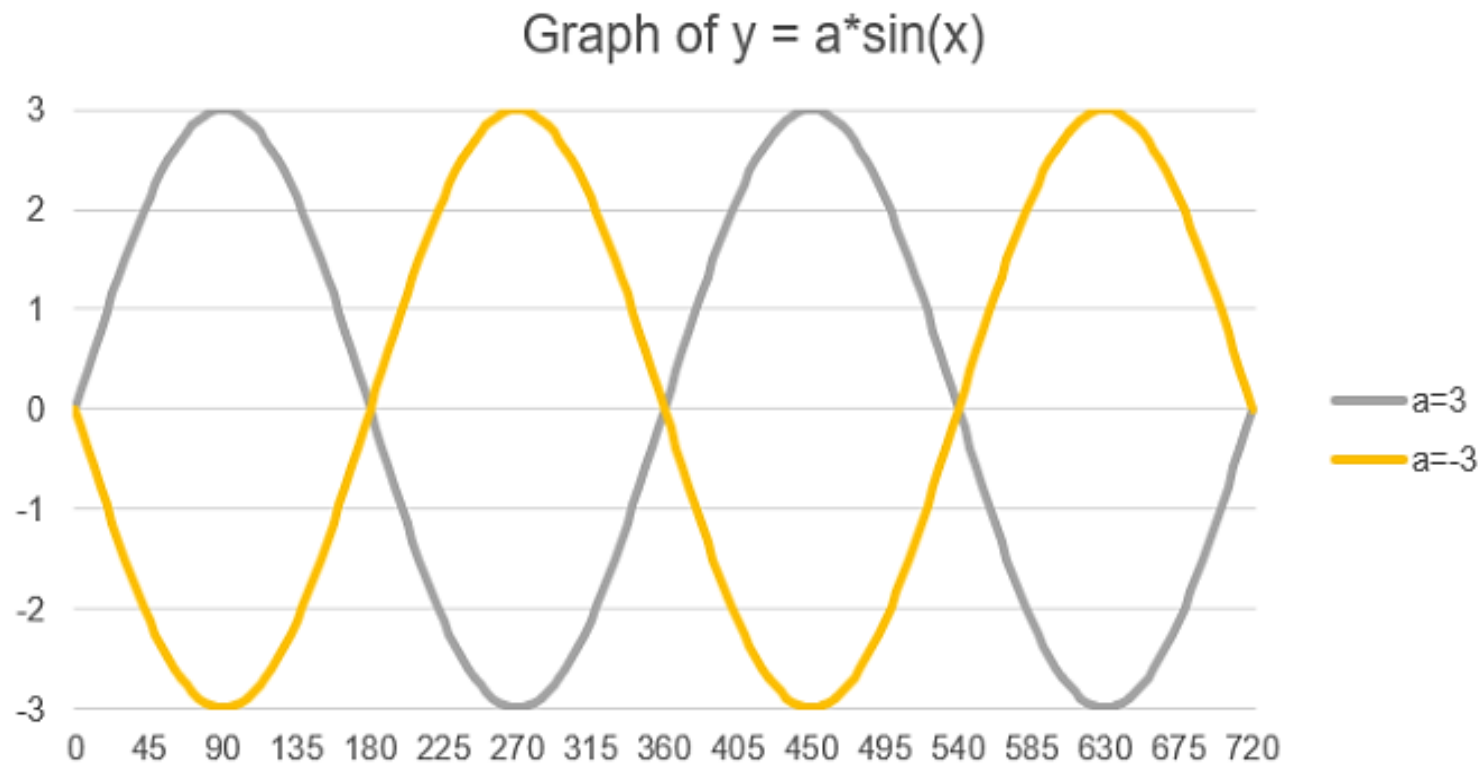
And we get:



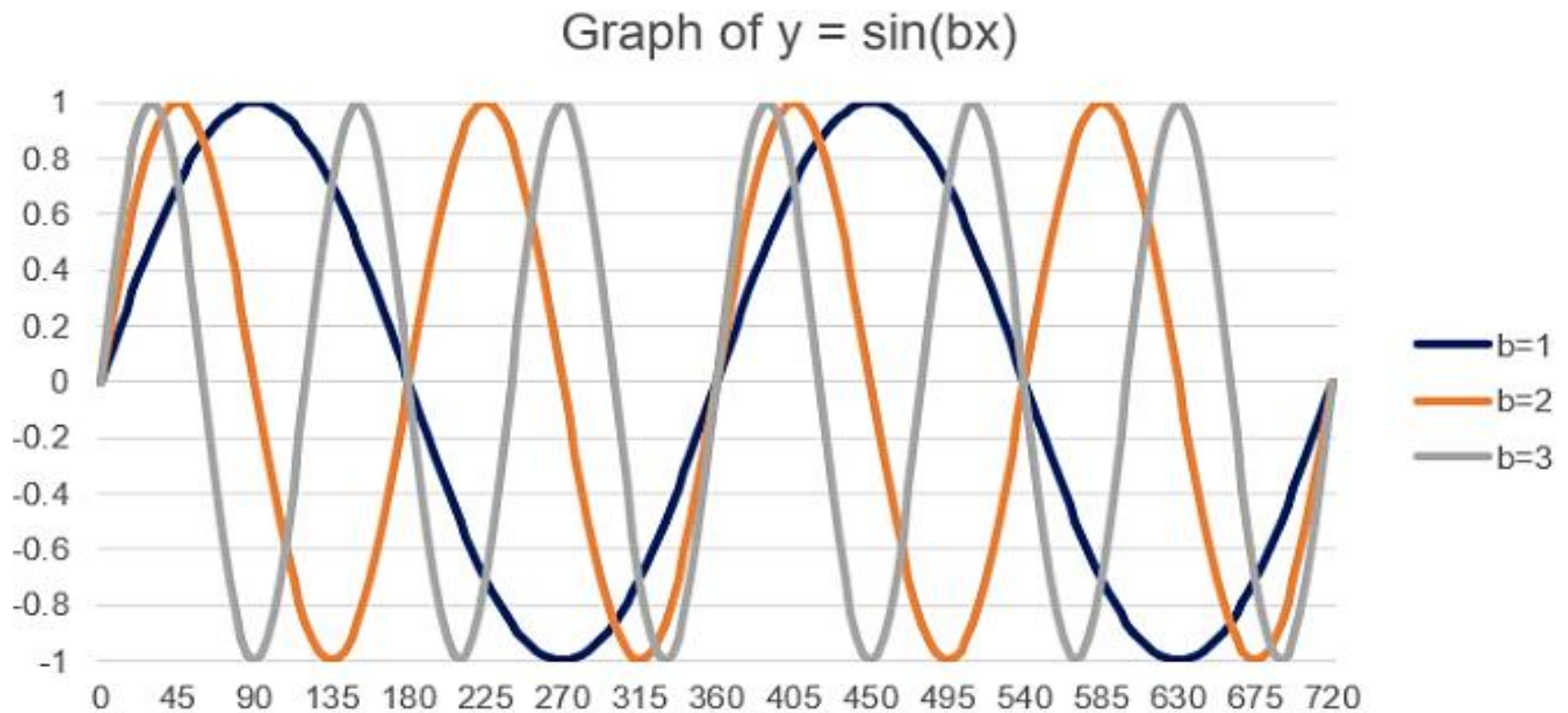
# Sine wave - Variation in Amplitude



# Sine wave - Variation in Amplitude

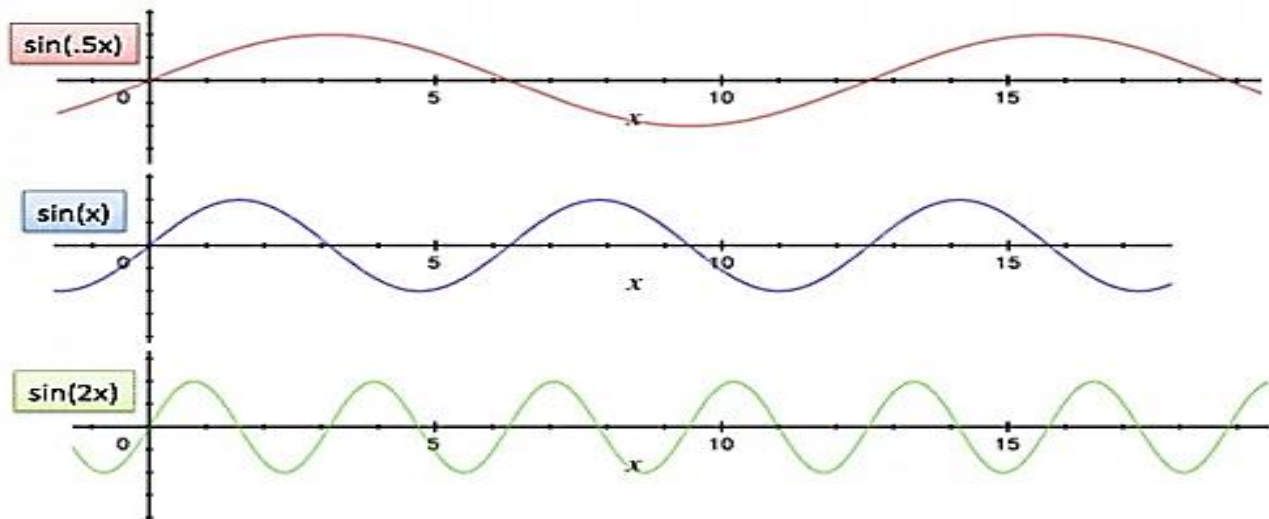


# Sine wave - Variation in Periodicity



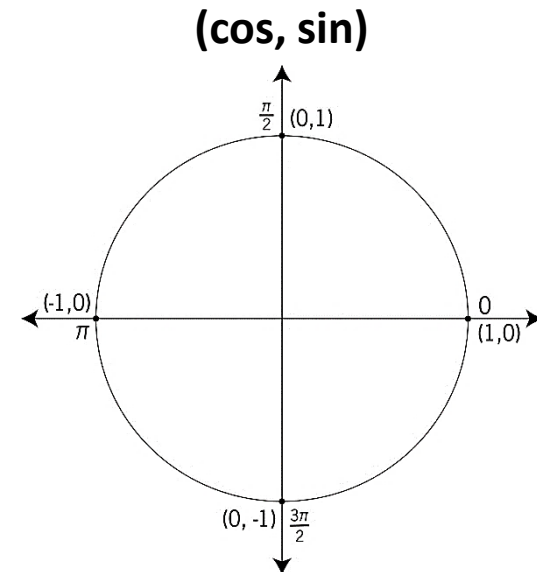
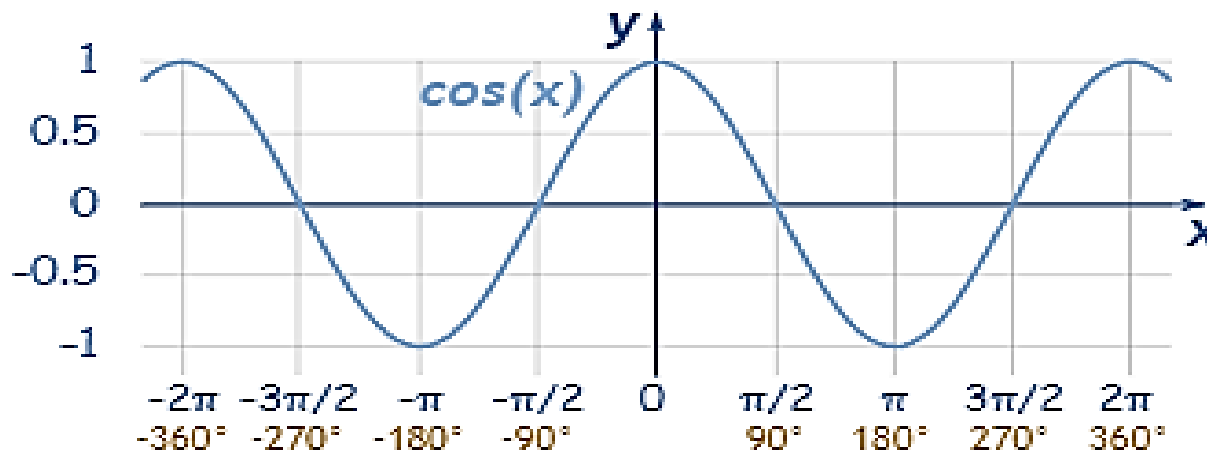
# How **fast** is Sine Wave?

- $\sin(x)$  is the *default* sine wave, that indeed takes  $\pi$  units of time to go from **0** to **1** and back to **0** (or  $2\pi$  for a complete cycle).
- $\sin(2x)$  is a wave that moves **twice as fast**.
- $\sin(0.5x)$  is a wave that moves **twice as slow**.
- So, we use  $\sin(n \cdot x)$  to get a sine wave cycling as fast as we need.



# What is **Cosine** Wave ?

- A basic **cosine** wave is an *S-shaped waveform* defined by the mathematical function  **$y = \cos(x)$** .
- The **cosine** function has up-down curve (which repeats every  **$2\pi$**  radians, or  **$360^\circ$** ).



- For every cycle, it starts at **1**, heads down to **0** by  **$\pi/2$**  radians ( **$90^\circ$** ) and again heads down to **-1** and finally heads up to **0**.

# What is **Cosine** Wave ?

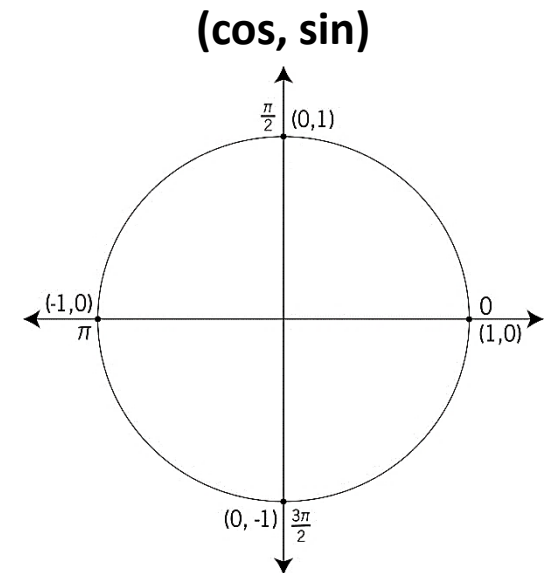
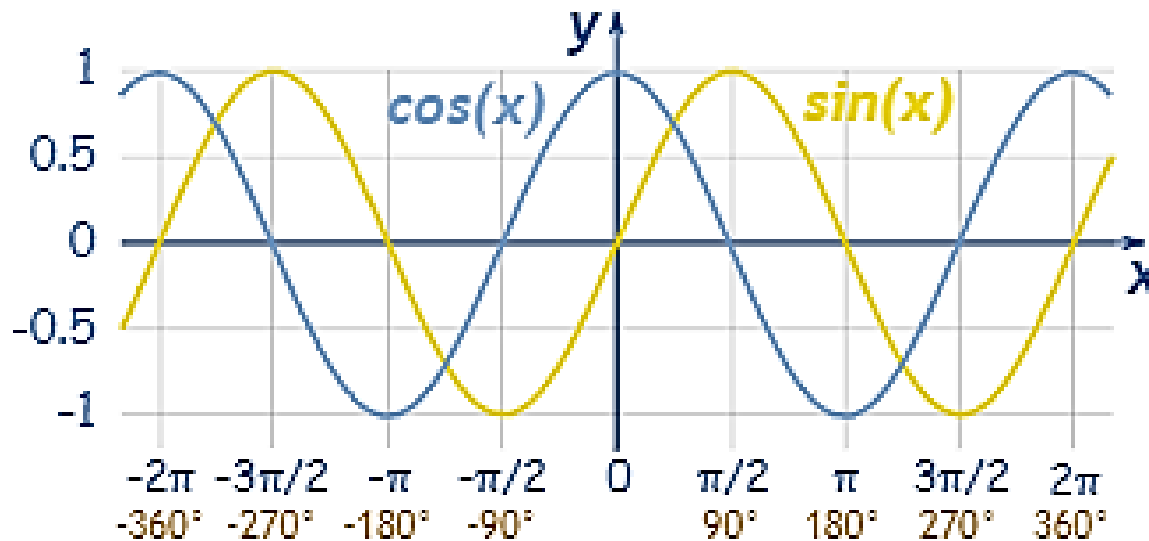
- The general equation of a **cosine** wave is given by:

$$Y = \mathbf{A} \cos(\mathbf{B}(x + \mathbf{C})) + \mathbf{D}$$

- amplitude is **A**
- period is  **$2\pi/B$**
- phase shift is **C** (positive is to the **left**)
- vertical shift is **D**

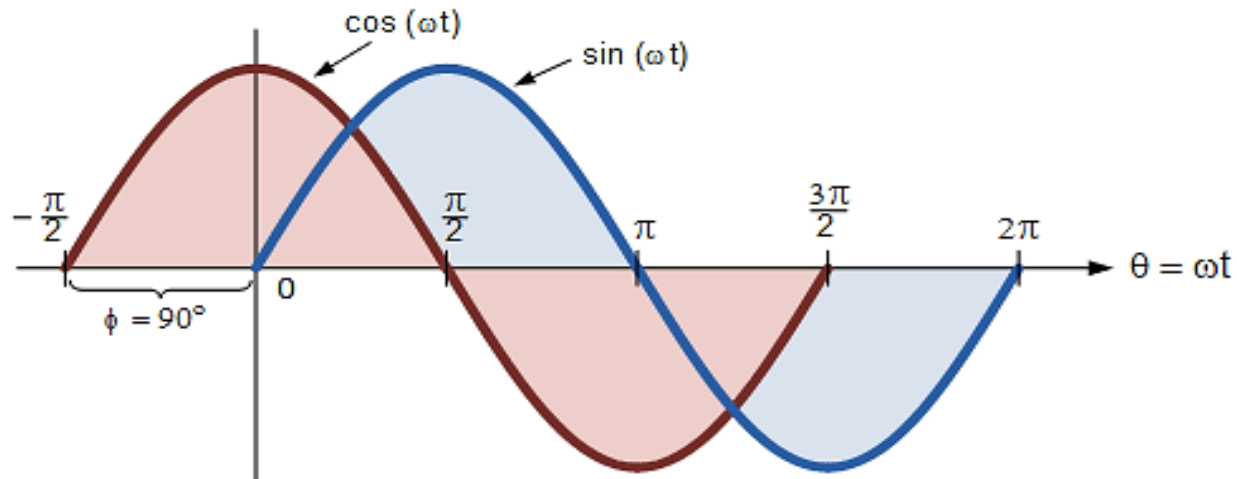
# Plot of Sine and Cosine Waves

- Sine and Cosine are **good friends**: they follow each other, exactly  $\pi/2$  radians (**90°**) apart.





# Plot of Sine and Cosine Waves



- A sine wave and a cosine wave are **90°** ( $\pi/2$  radians) **out of phase** with each other.

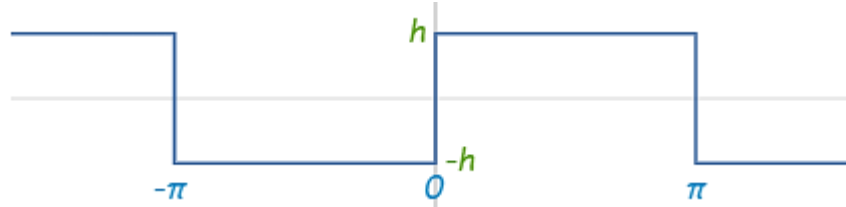
$$\cos(\omega t + \phi) = \sin\left(\omega t + \phi + 90^\circ\right)$$

$$\sin(\omega t + \phi) = \cos\left(\omega t + \phi - 90^\circ\right)$$

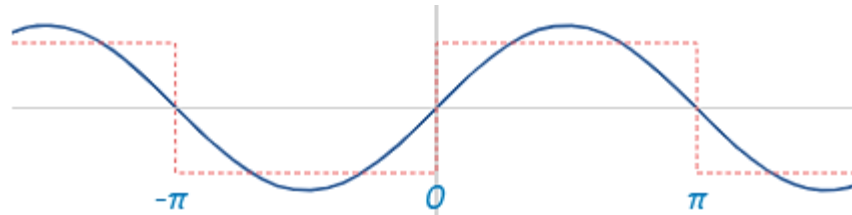
# Forming a Square Wave

<https://www.geogebra.org/graphing>

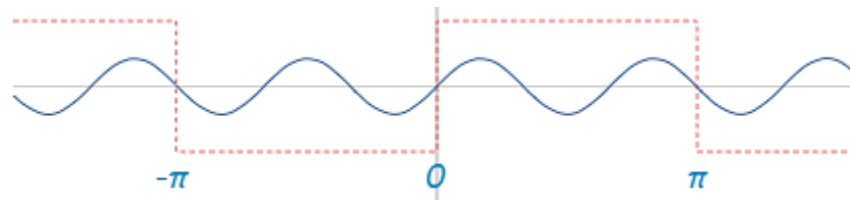
- Can we **add sine** waves to make a **square wave**?



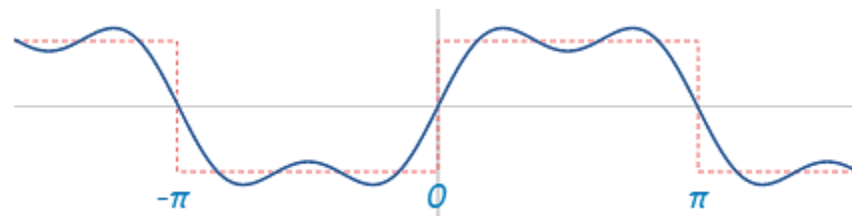
**$\sin(x)$ :**



**$\sin(3x)/3$ :**

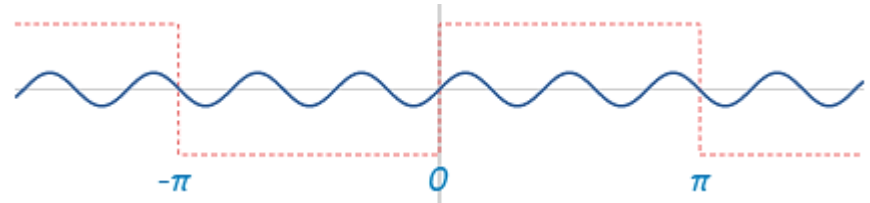


**$\sin(x) + \sin(3x)/3$ :**

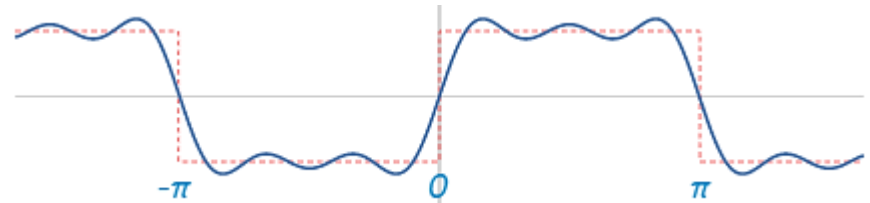


# Forming a Square Wave

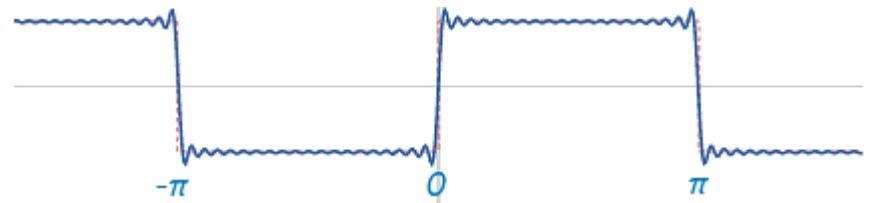
$\sin(5x)/5$ :



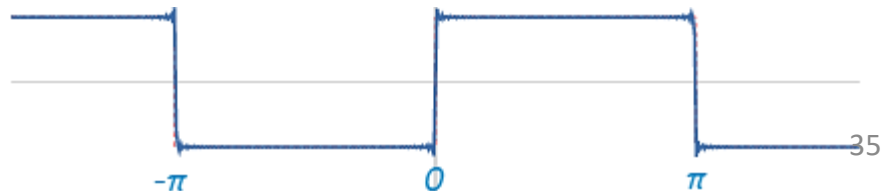
$\sin(x) + \sin(3x)/3 + \sin(5x)/5$ :



$\sin(x) + \sin(3x)/3 + \sin(5x)/5 + \dots + \sin(39x)/39$ :



$\sin(x) + \sin(3x)/3 + \sin(5x)/5 + \dots + \sin(199x)/199$ :



# Complex Number

# Complex Number

- A **Complex Number** is a combination of a **Real Number** and an **Imaginary Number**.



Real Numbers are numbers like:

1	12.38	-0.8625	3/4	$\sqrt{2}$	1998
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Nearly any number you can think of is a Real Number!



Imaginary Numbers when **squared** give a **negative** result.

Normally this doesn't happen, because:

- when we square a positive number we get a positive result, and
- when we square a negative number we also get a positive result (because a negative times a negative gives a positive), for example  $-2 \times -2 = +4$

# Complex Number

## Imaginary number

The "unit" imaginary number (like 1 for Real Numbers) is  $i$ , which is the square root of  $-1$

$$i = \sqrt{-1}$$

Because when we square  $i$  we get  $-1$

$$i^2 = -1$$

Examples of Imaginary Numbers:

$$3i$$

$$1.04i$$

$$-2.8i$$

$$3i/4$$

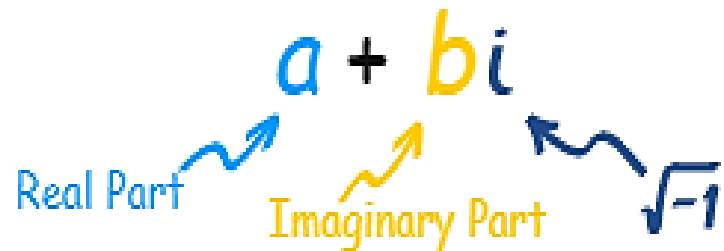
$$(\sqrt{2})i$$

$$1998i$$

And we keep that little "i" there to remind us we need to multiply by  $\sqrt{-1}$

# Complex Number

When we combine a Real Number and an Imaginary Number we get a **Complex Number**:



The diagram shows the general form of a complex number,  $a + bi$ . The letter  $a$  is blue and labeled "Real Part" with a blue wavy arrow pointing to it. The letter  $b$  is yellow and labeled "Imaginary Part" with a yellow wavy arrow pointing to it. The letter  $i$  is blue and labeled with  $\sqrt{-1}$  with a blue wavy arrow pointing to it.

Examples:

$$1 + i$$

$$39 + 3i$$

$$0.8 - 2.2i$$

$$-2 + \pi i$$

$$\sqrt{2} + i/2$$

# Complex Number

## Either Part Can Be Zero

So, a Complex Number has a real part and an imaginary part.

But either part can be **0**, so all Real Numbers and Imaginary Numbers are also Complex Numbers.

Complex Number	Real Part	Imaginary Part	
$3 + 2i$	3	2	
5	5	0	Purely Real
$-6i$	0	-6	Purely Imaginary

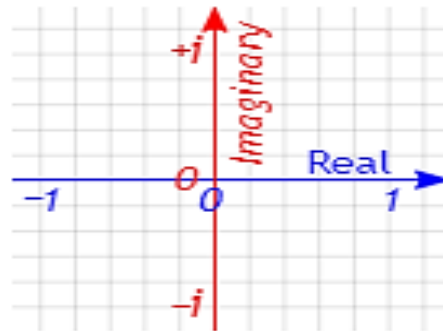


# Complex Number

## A Visual Explanation

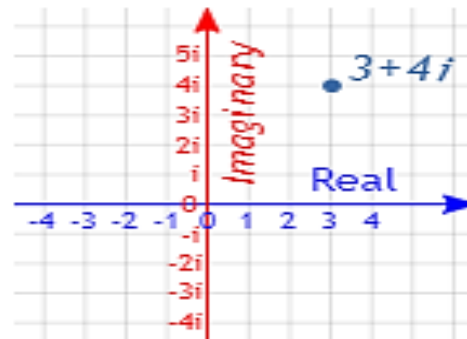
You know how the number line goes **left-right**?

Well let's have the imaginary numbers go **up-down**:



And we get the Complex Plane

A complex number can now be shown as a point:



The complex number  $3 + 4i$

# Complex Number

## Adding

To add two complex numbers we add each part separately:

$$(a+bi) + (c+di) = (a+c) + (b+d)i$$

Example: add the complex numbers  $3 + 2i$  and  $1 + 7i$

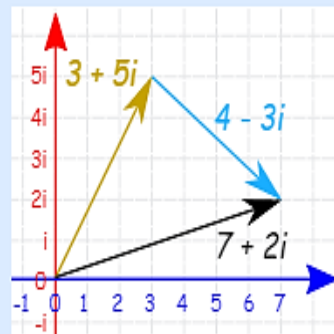
- add the real numbers, and
- add the imaginary numbers:

$$\begin{aligned}(3 + 2i) + (1 + 7i) \\&= 3 + 1 + (2 + 7)i \\&= 4 + 9i\end{aligned}$$

Example: add the complex numbers  $3 + 5i$  and  $4 - 3i$

$$\begin{aligned}(3 + 5i) + (4 - 3i) \\&= 3 + 4 + (5 - 3)i \\&= 7 + 2i\end{aligned}$$

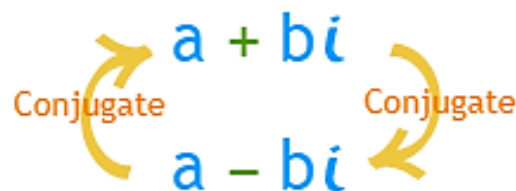
On the complex plane it is:



# Complex Number

## Conjugates

A conjugate is where we **change the sign in the middle** like this:



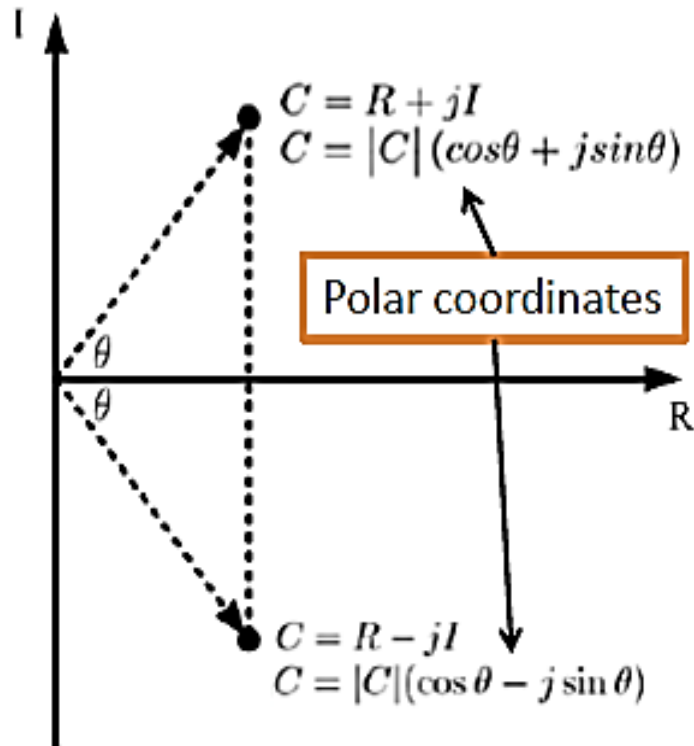
A conjugate is often written with a bar over it:

Example:

$$\overline{5 - 3i} = 5 + 3i$$

# Complex Number Representations (important)

In technical disciplines, we represent a complex number as  $\mathbf{j}$  instead of  $\mathbf{i}$  to avoid the confusion for the electric current representation( $\mathbf{i}$ ).



$$\begin{aligned}\text{Cartesian Form: } C &= R + jI \\ \text{Polar Form: } C &= |C|(\cos\theta + j\sin\theta) = |C|e^{j\theta} \\ \text{Euler's Relation: } e^{j\theta} &= \cos\theta + j\sin\theta\end{aligned}$$

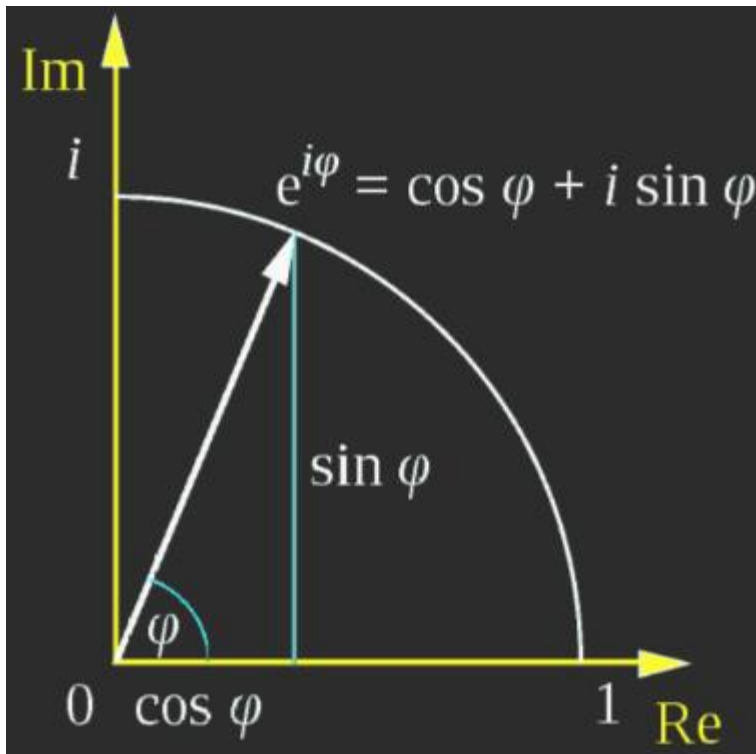
$$|C| = \sqrt{R^2 + I^2} \quad \text{Magnitude}$$

$$\theta = \arctan\left(\frac{I}{R}\right) \approx \tan^{-1}\left(\frac{I}{R}\right)$$

Complex Plane

# Euler's Formula

- One of the applications of Euler's formula is for **complex analysis**.
- **Euler's formula allows us to express complex numbers as exponentials.**



$\varphi$  = any real number

$i$  = imaginary unit ( i.e.,  $\sqrt{-1}$  )

$e$  = base of the natural logarithm

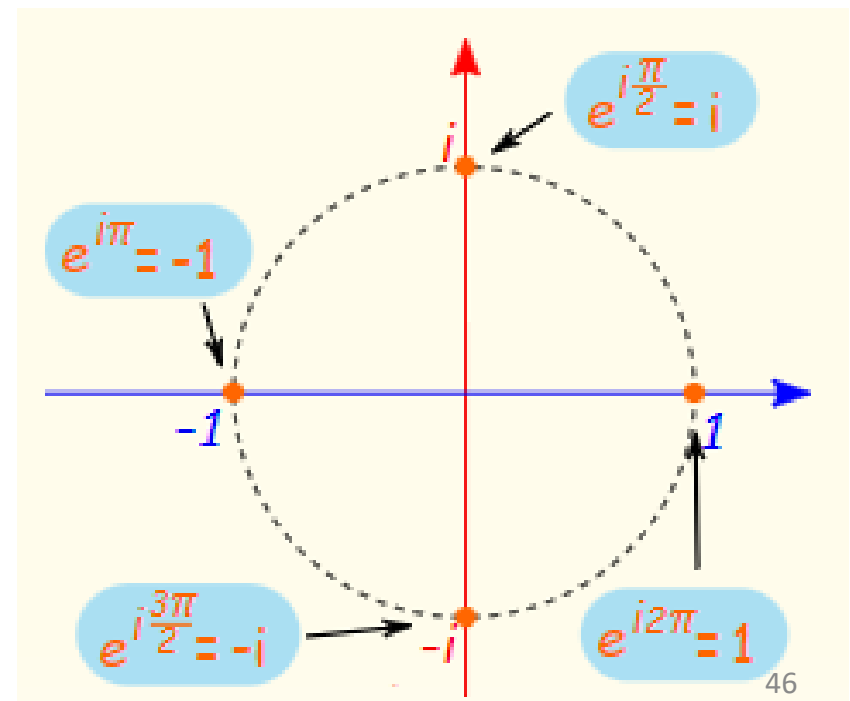
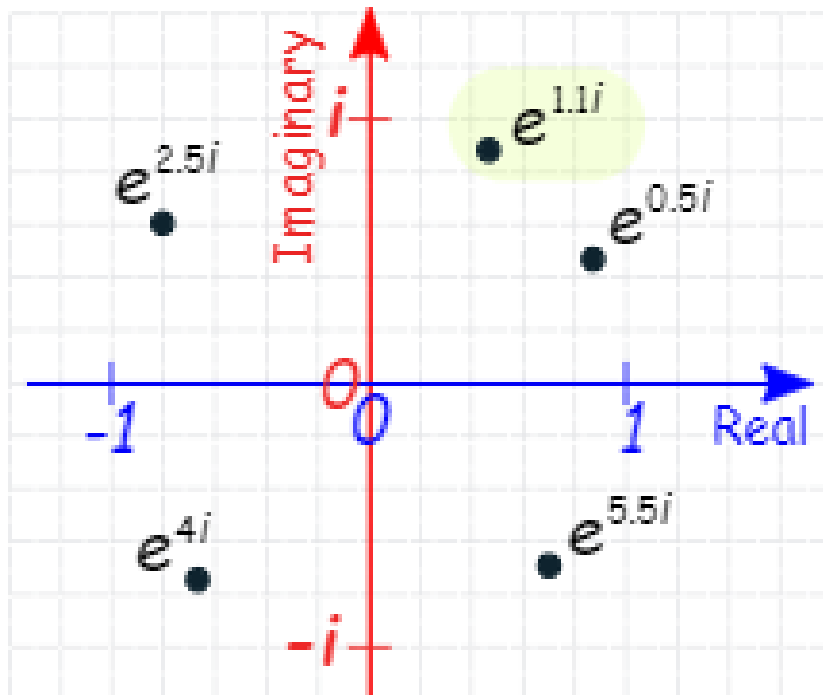
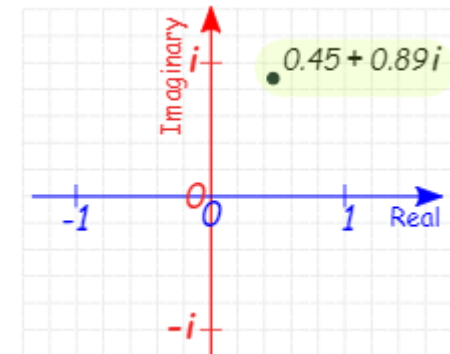
# Euler's Formula

Example: when  $x = 1.1$

→  $e^{ix} = \cos x + i \sin x$

→  $e^{1.1i} = \cos 1.1 + i \sin 1.1$

→  $e^{1.1i} = 0.45 + 0.89i$  (to 2 decimals)



# Euler's Formula - Cartesian to Polar conversion

Example: the number  $3 + 4i$

To turn  $3 + 4i$  into  $re^{j\theta}$  form we do a [Cartesian to Polar conversion](#) :

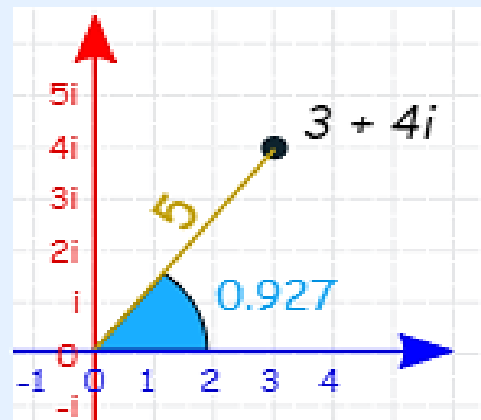
- $r = \sqrt{3^2 + 4^2} = \sqrt{9+16} = \sqrt{25} = 5$
- $\theta = \tan^{-1} ( 4 / 3 ) = 0.927$  (to 3 decimals)

$$|C|e^{j\theta}$$

$$|C| = \sqrt{R^2 + I^2} \quad \text{Magnitude}$$

$$\theta = \arctan\left(\frac{I}{R}\right) \approx \tan^{-1}\left(\frac{I}{R}\right)$$

So  $3 + 4i$  can also be  $5e^{0.927 i}$



# Fourier Series

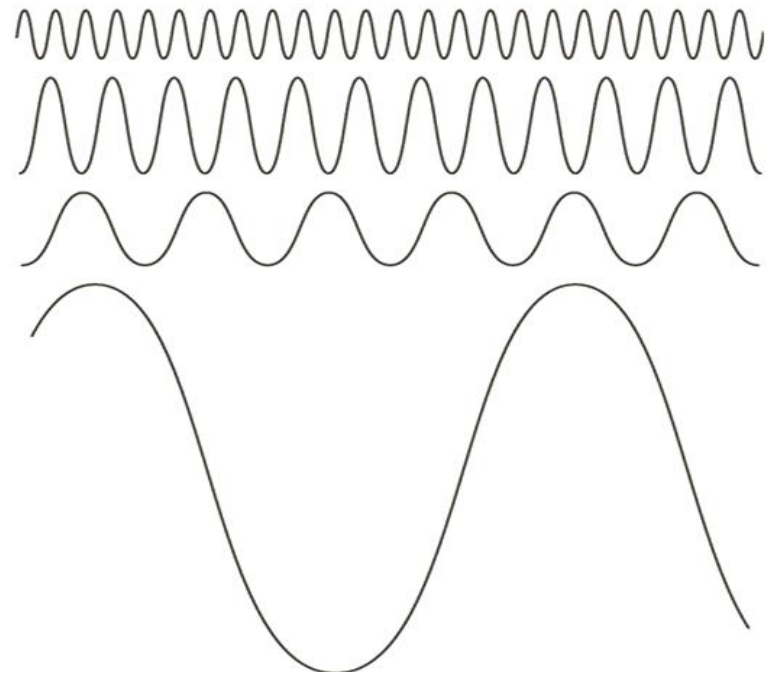


# Background

- **Jean Baptiste Joseph Fourier** (in 1807) stated that: **any periodic function** can be decomposed in to a sum of **sines and/or cosines** of different frequencies, each multiplied by a different coefficient (**Fourier Series**).

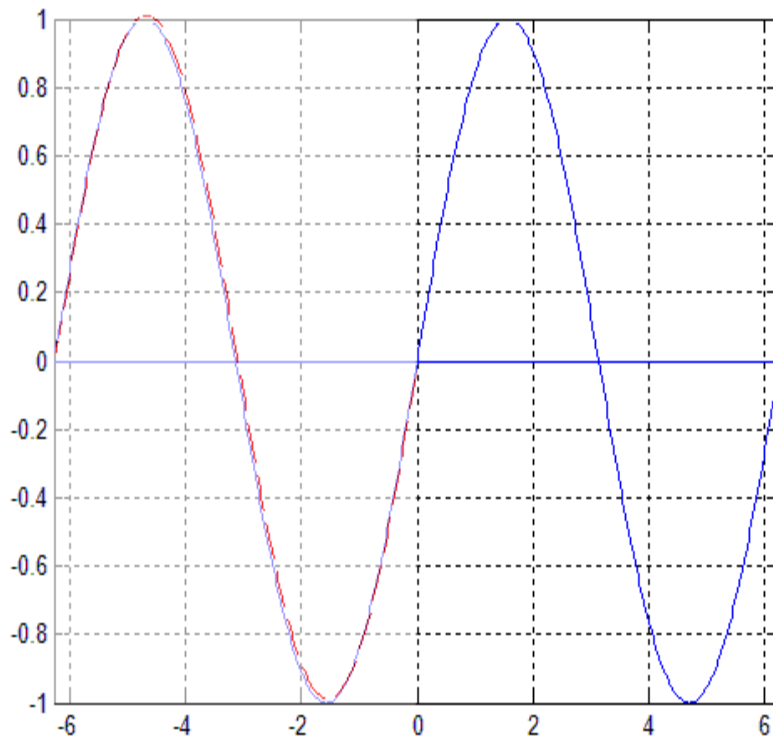


Periodic function



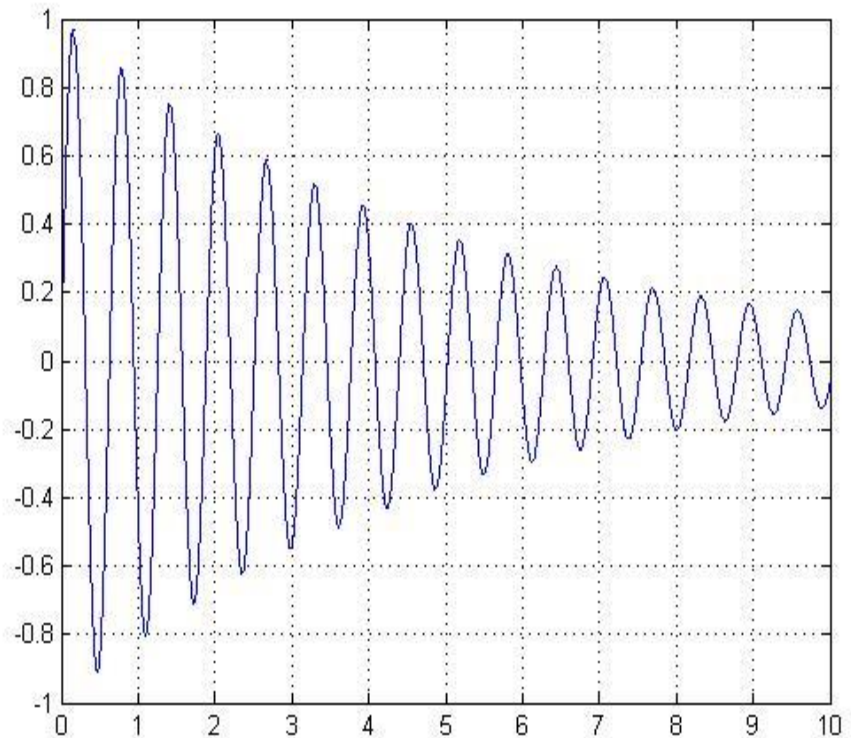
Sines and Cosines

# Background



Periodic function

A periodic function remain self-similar for all integer multiples of its period.



Non-periodic function

A non-periodic function does not remain self-similar for all integer multiples of its period.

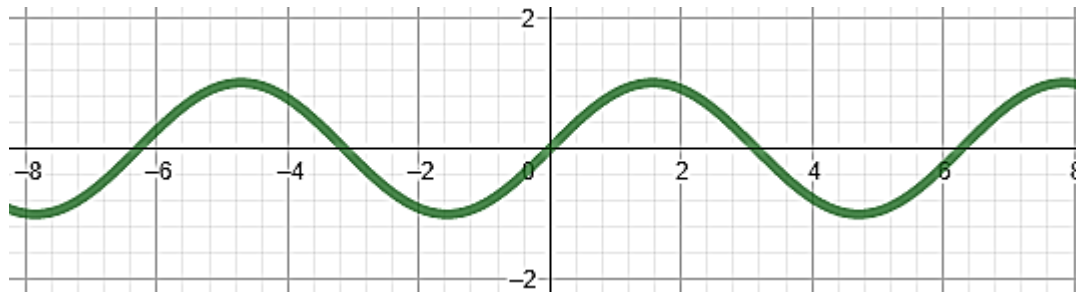
# Background

- Fourier Series
  - Any **periodic functions**
  - Expressed as **sum of sine and/or cosine functions** of different frequencies, each multiplied by a different coefficient
- Fourier Transform
  - Any **Non-periodic functions**, with area under the curve being **finite** (bounded)
  - Expressed as **integral of sine and/or cosine functions**, multiplied by a weighing function

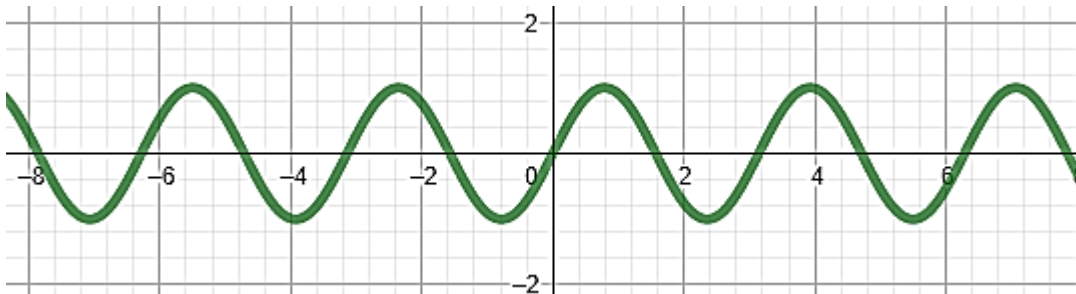
# Adding periodic functions - example

- sine and cosine waves can make other periodic functions.

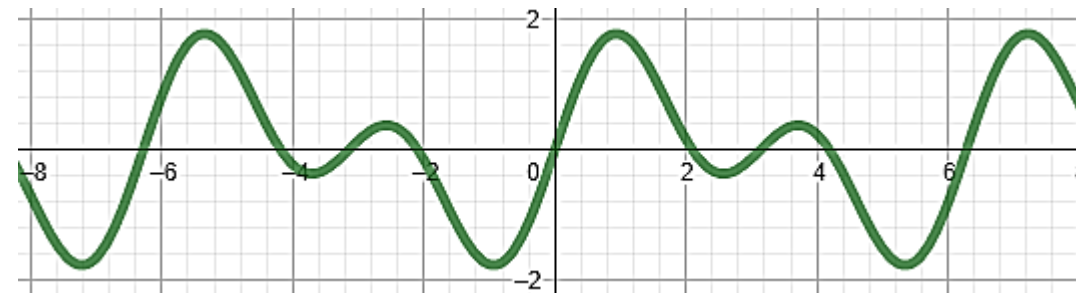
$\sin(x)$



$\sin(2x)$



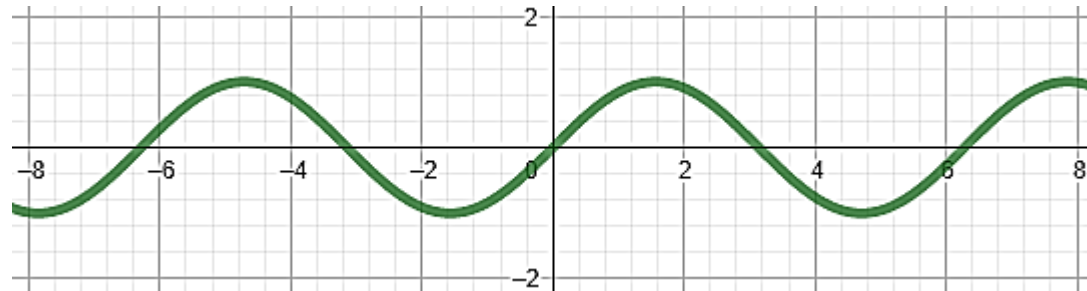
$\sin(x) + \sin(2x)$



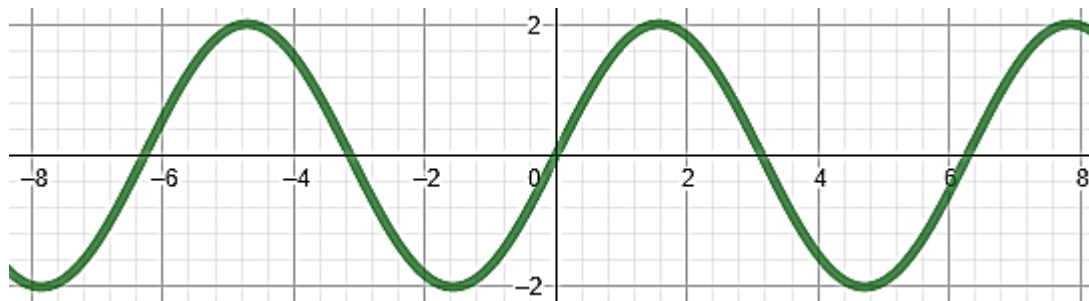
# Adding periodic functions - example

- sine and cosine waves can make other periodic functions.

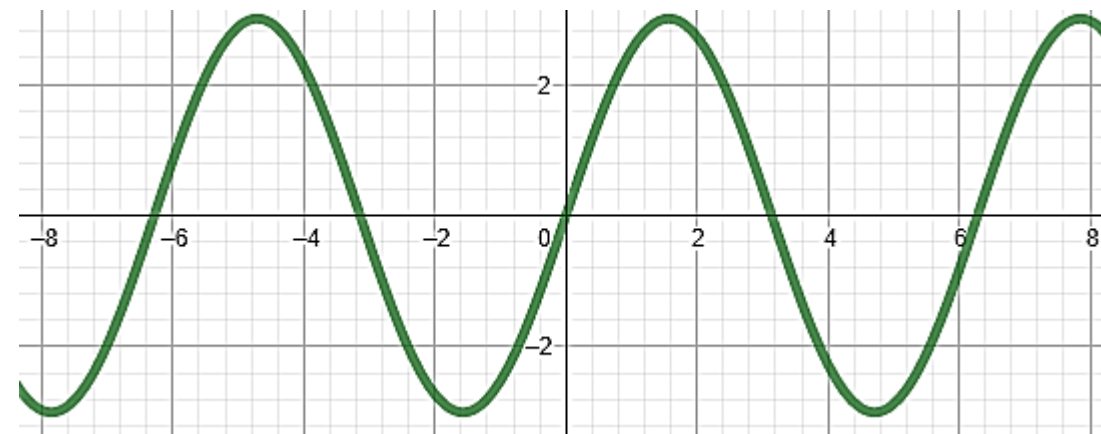
**$\sin(x)$**



**$2 \sin(x)$**



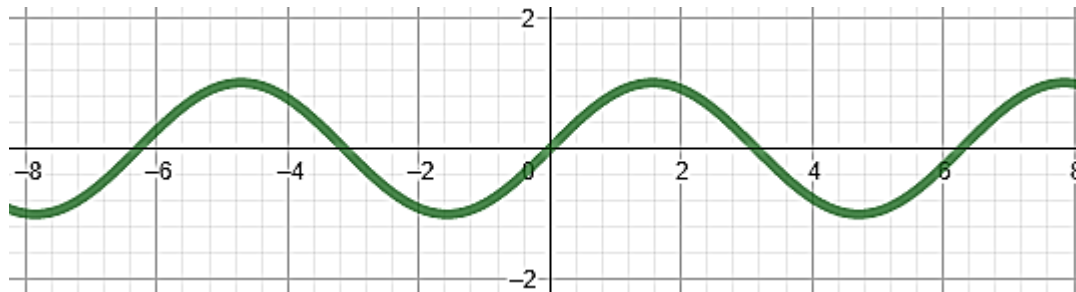
**$\sin(x) + 2 \sin(x)$**



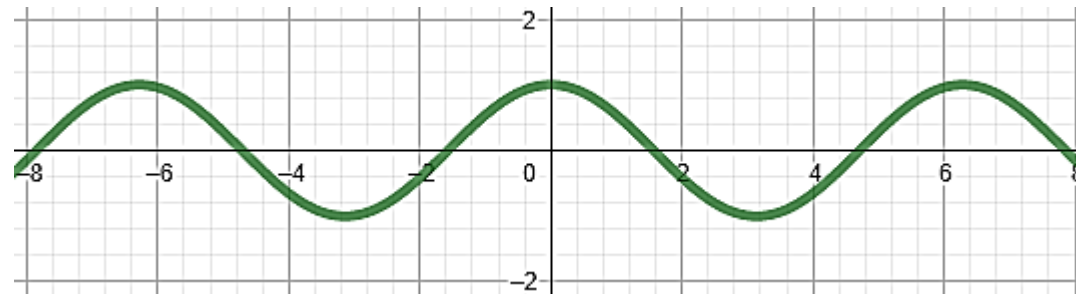
# Adding periodic functions - example

- sine and cosine waves can make other periodic functions.

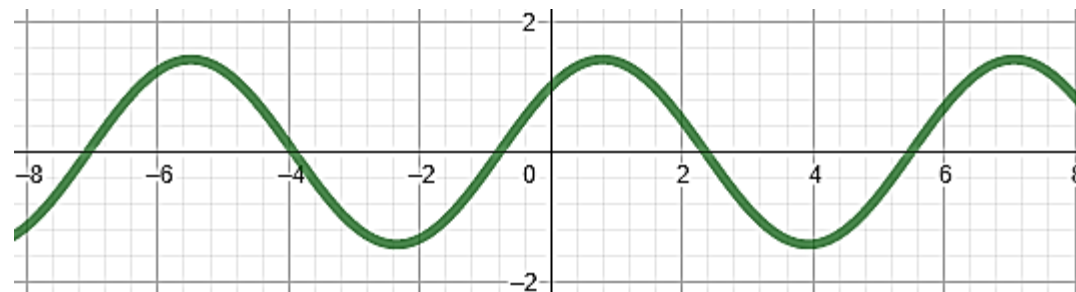
$\sin(x)$



$\cos(x)$



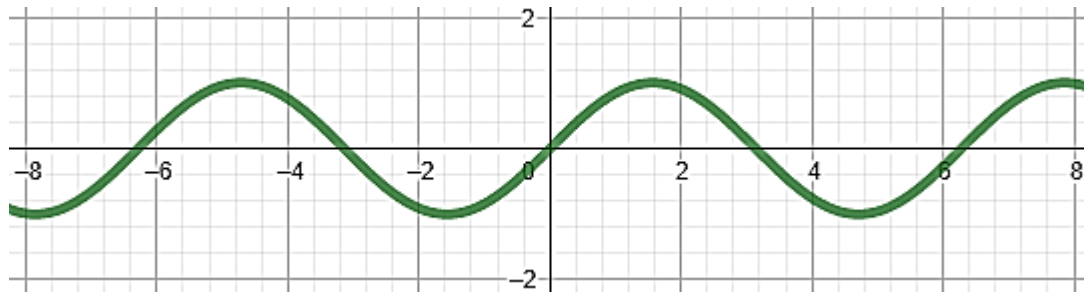
$\sin(x) + \cos(x)$



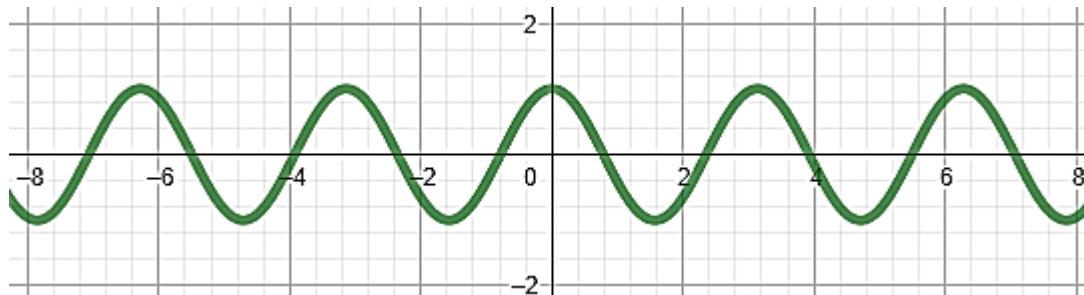
# Adding periodic functions - example

- sine and cosine waves can make other periodic functions.

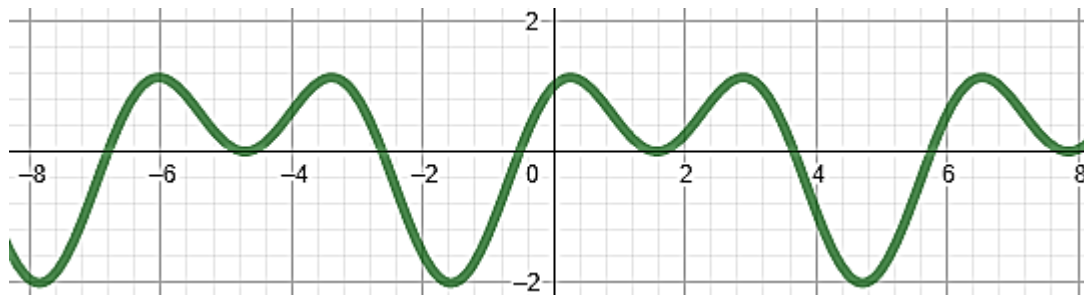
$\sin(x)$



$\cos(2x)$



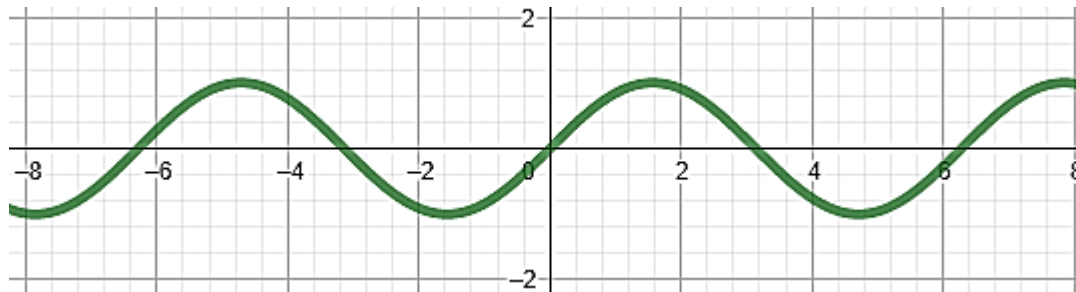
$\sin(x) + \cos(2x)$



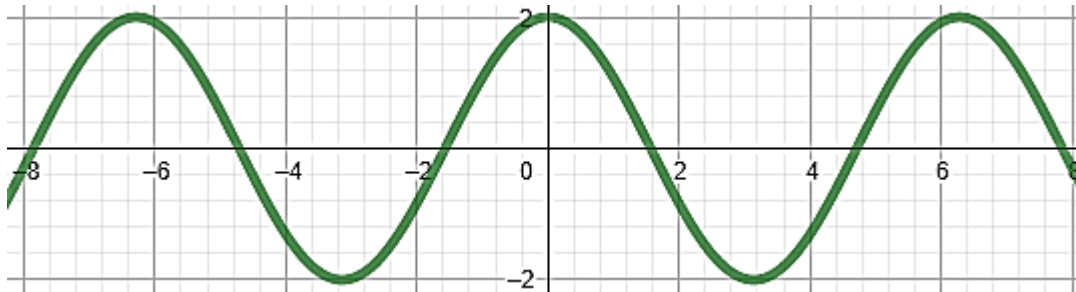
# Adding periodic functions - example

- sine and cosine waves can make other periodic functions.

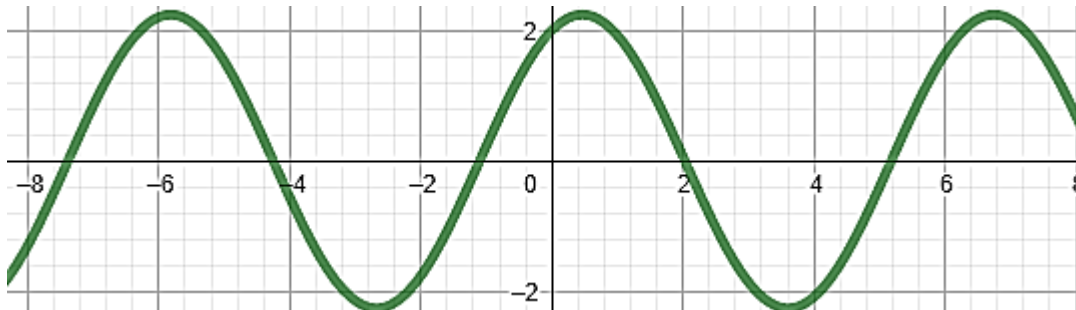
$\sin(x)$



$2 \cos(x)$



$\sin(x) + 2 \cos(x)$





# Fourier Series

- A function  $f(t)$  of a continuous variable  $t$ , that is periodic with a period  $T$ , can be expressed as the **sum** of *sines* and *cosines* multiplied by appropriate coefficients.

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{j\frac{2\pi n}{T}t}$$

Where,

$$c_n = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-j\frac{2\pi n}{T}t} dt \quad \text{for } n = 0, \pm 1, \pm 2, \dots$$

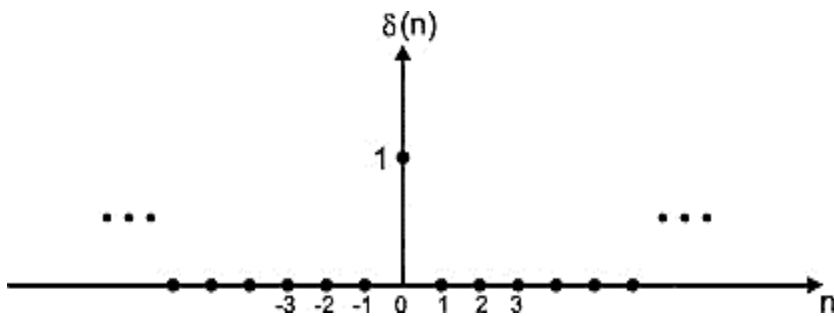
are the coefficients

**Small  $n \rightarrow$  Longer period and vice versa**

# Recall

- Discrete unit impulse signal  $\delta(n)$ : signal contains a single **1** with the rest being **0**s.

unit impulse

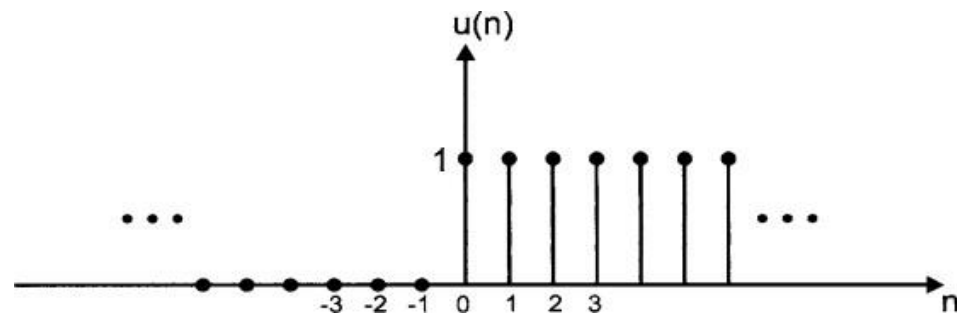


$$\delta(n) = \begin{cases} 1, & n = 0 \\ 0, & n \neq 0 \end{cases}$$

or

$$\delta(n) = \{\dots, 0, 0, 1, 0, 0, \dots\}$$

unit step



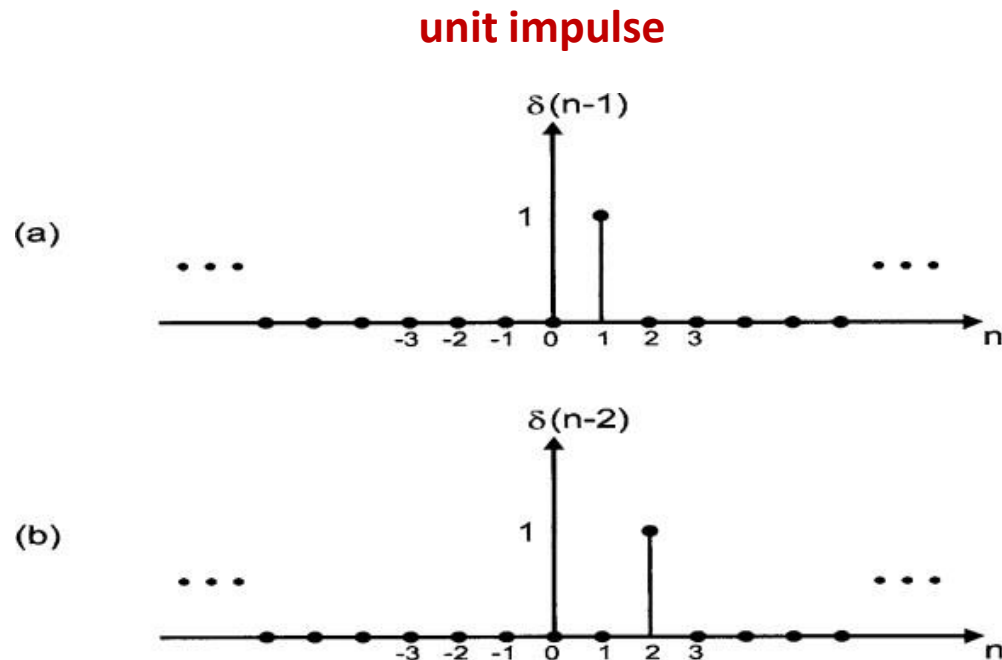
$$u(n) = \begin{cases} 0, & n < 0 \\ 1, & n \geq 0 \end{cases}$$

or

$$u(n) = \{\dots, 0, 0, 1, 1, 1, 1, \dots\}$$

# Recall

- Discrete unit impulse signal  $\delta(n)$ : signal contains a single **1** with the rest being **0**s.



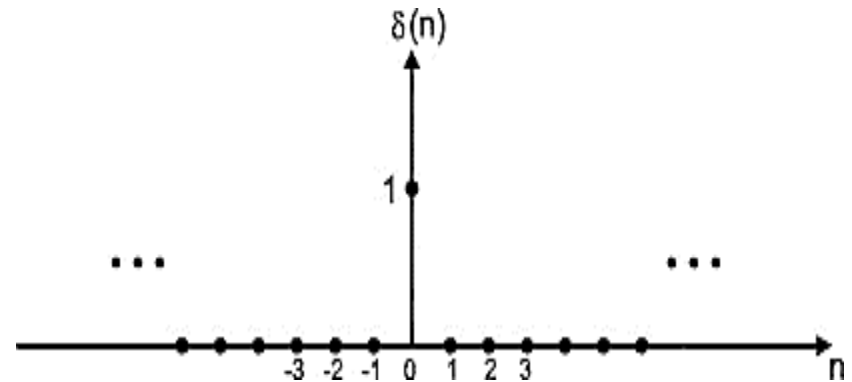
Delayed/shifted versions of the unit impulse sequence

# Recall

## Difference between an **unit impulse signal** in **discrete time domain** and **continuous time domain**

- Discrete time Unit Impulse signal

$$\delta(n) = 1 \text{ for } n = 0$$
$$0 \text{ otherwise}$$

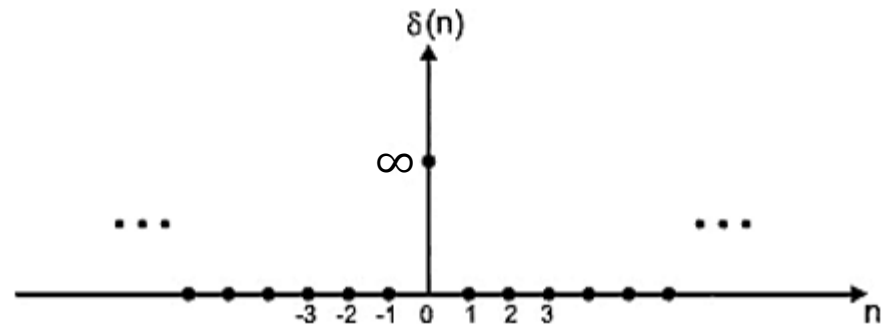


- Continuous time Unit Impulse signal

$$\delta(n) = \infty \text{ for } n = 0$$
$$0 \text{ otherwise}$$

**and**

$$\int_{-\infty}^{\infty} \delta(n) dn = 1 \text{ \{ its area is unity \}}$$



# Recall

- Correlation of a signal with a discrete unit impulse
  - Outputs a **rotated version of the signal**, centered at the impulse location.
- Convolution of a signal with a discrete unit impulse
  - Outputs a **copy of the same signal**, centered at the impulse location.

# Sifting Property of Impulses (important)

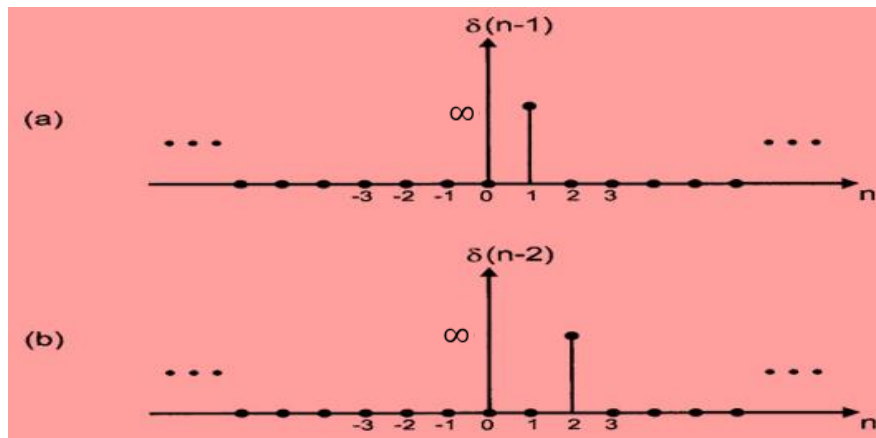
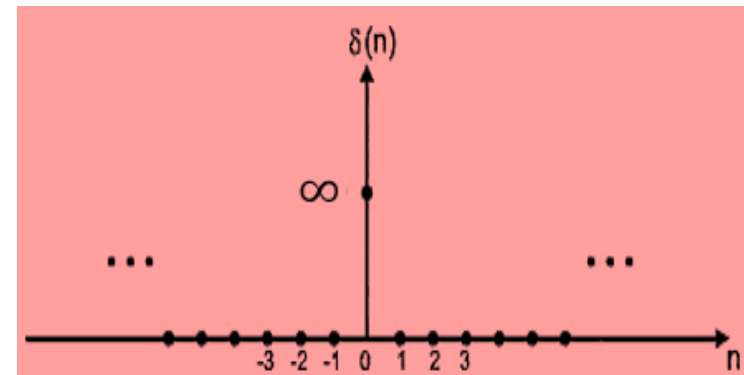
$$\int_{-\infty}^{\infty} f(t)\delta(t)dt = f(0)$$

provided  $f(t)$  is continuous at  $t = 0$



Generalize

$$\int_{-\infty}^{\infty} f(t)\delta(t - t_0)dt = f(t_0)$$



# Sifting Property of Impulses (**important**)

$$\int_{-\infty}^{\infty} f(t)\delta(t - t_0)dt = f(t_0)$$

Sifting yields the *value of the function* at the *location of the impulse*.

For example, if  $f(t) = \cos(t)$ , using the impulse  $\delta(t - \pi)$  yields the result:  
 $f(\pi) = \cos(\pi) = -1$ .

# Sifting Property of Impulses (**important**)

A **unit impulse** of a discrete variable  $x$ , located at  $x = 0$ , denoted  $\delta(x)$  is defined as:

$$\delta(x) = \begin{cases} 1 & \text{if } x = 0 \\ 0 & \text{if } x \neq 0 \end{cases} \text{ and subject to: } \sum_{x=-\infty}^{\infty} \delta(x) = 1$$

Sifting Property in discrete variable  $\int_{-\infty}^{\infty} \delta(n)dn = 1$

$$\sum_{x=-\infty}^{\infty} f(x)\delta(x) = f(0)$$



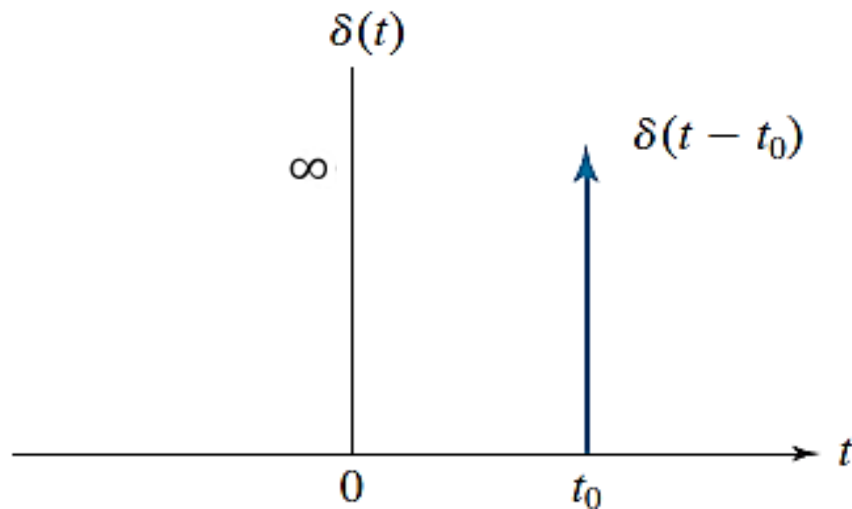
Generalize

$$\sum_{x=-\infty}^{\infty} f(x)\delta(x - x_0) = f(x_0)$$

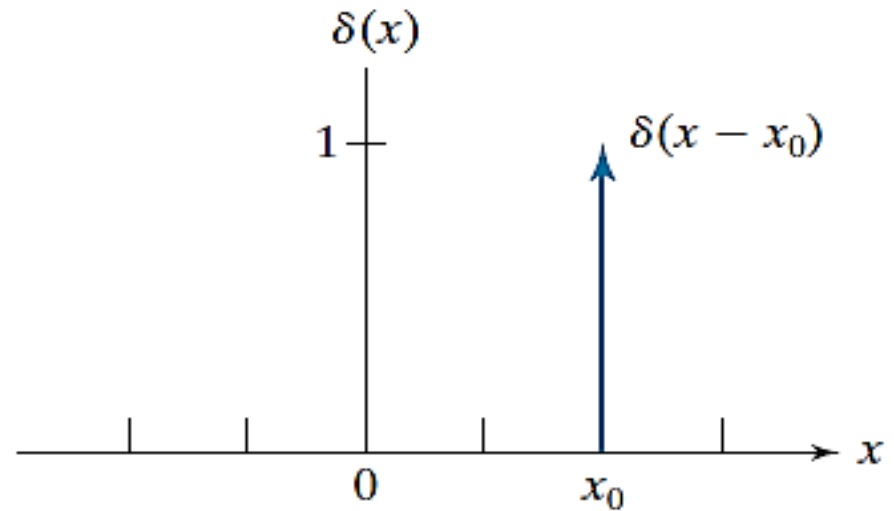
Sifting yields the *value of the function* at the *location of the impulse*.



# Sifting Property of Impulses (**important**)



Continuous impulse located  
at  $t = t_0$

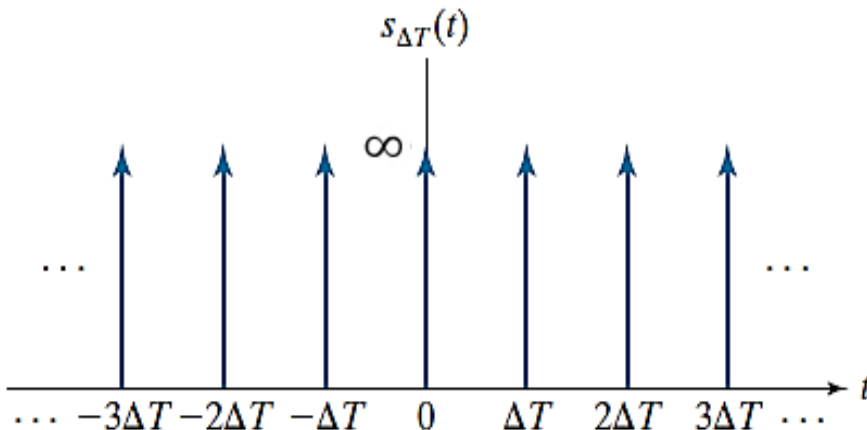


Discrete impulse located at  
 $x = x_0$

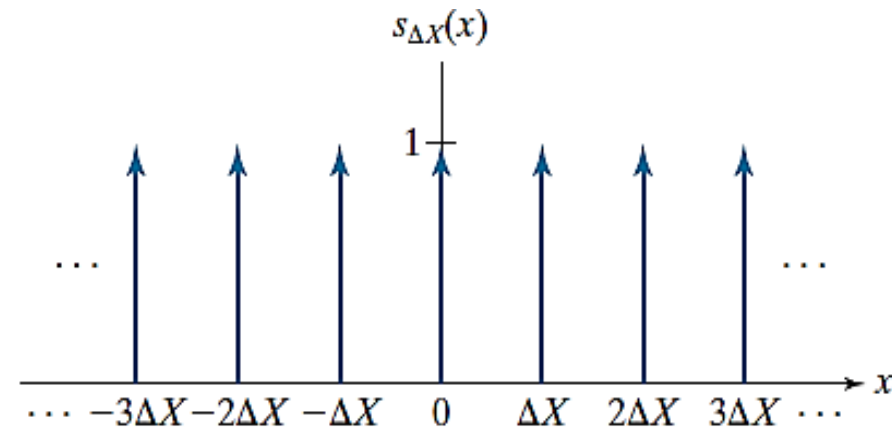
# Impulse Train

The **impulse train**  $s_{\Delta T}(t)$  is defined as the sum of infinitely many impulses which are  **$T$  units apart**:

$$s_{\Delta T}(t) = \sum_{k=-\infty}^{\infty} \delta(t - k\Delta T)$$



**Continuous impulse train**



**Discrete impulse train**

# 1D Continuous Fourier Transform

- The **Fourier transform** of a **continuous function**  $f(t)$  of a **continuous variable**  $t$ , denoted  $\mathfrak{F}\{f(t)\}$  is defined by the equation:

$$\mathfrak{F}\{f(t)\} = \int_{-\infty}^{\infty} f(t) e^{-j2\pi\mu t} dt$$

where  $\mu$  is a continuous variable (defines frequency) also.

- Because  $t$  is integrated out,  $\mathfrak{F}\{f(t)\}$  is a function only of  $\mu$ . Therefore,

$$\mathfrak{F}\{f(t)\} = F(\mu) \quad \text{so,}$$

$$F(\mu) = \int_{-\infty}^{\infty} f(t) e^{-j2\pi\mu t} dt$$

# 1D Continuous Fourier Transform

$$\left. \begin{aligned} F(\mu) &= \int_{-\infty}^{\infty} f(t) e^{-j2\pi\mu t} dt \\ f(t) &= \int_{-\infty}^{\infty} F(\mu) e^{j2\pi\mu t} d\mu \end{aligned} \right\} \text{Fourier Transform Pair}$$
$$f(t) \Leftrightarrow F(\mu)$$

$$F(\mu) = \int_{-\infty}^{\infty} f(t) e^{-j2\pi\mu t} dt = \int_{-\infty}^{\infty} f(t) [\cos(2\pi\mu t) - j \sin(2\pi\mu t)] dt$$

- Here,  $f(t)$  is **real** and  $F(\mu)$  in general have **complex terms**.
- **Fourier transform  $F(\mu)$  is an expansion of  $f(t)$  multiplied by sinusoidal terms whose frequencies are determined by the values of  $\mu$ .**
- As  $\mu$  (*frequency*) is the only variable left after integration, the domain of  $F(\mu)$  is called the ***frequency domain***.

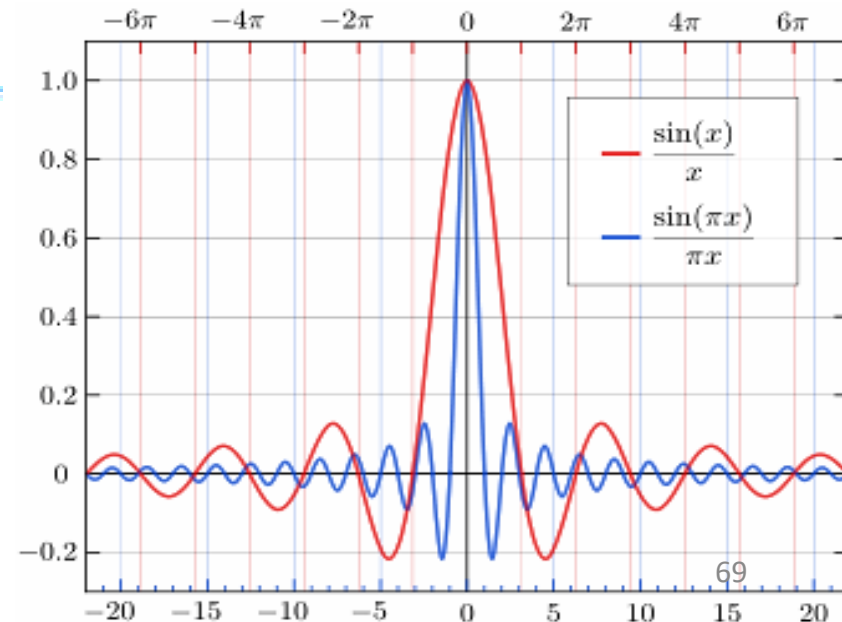
# Cardinal Sine Function – **sinc( )**

- The **unnormalized** *sinc* function is given as:

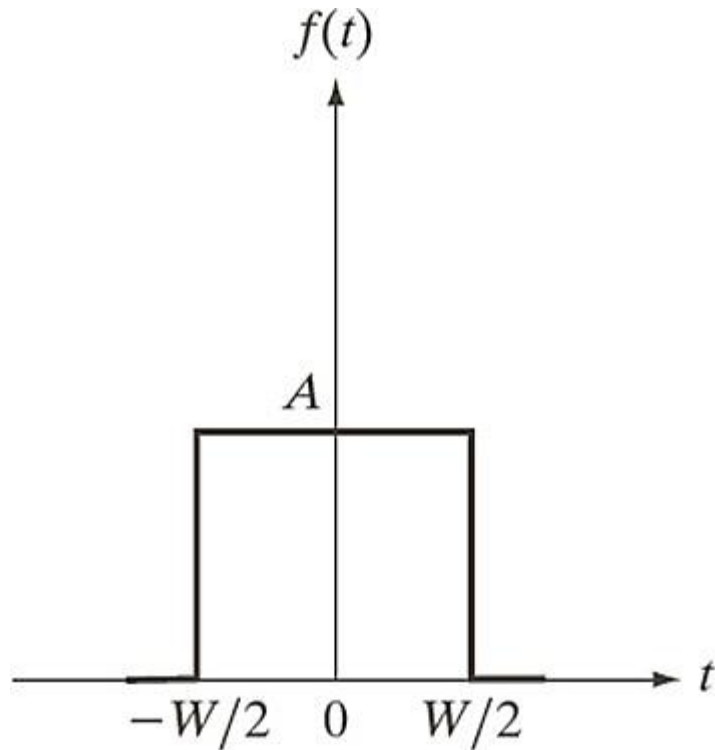
$$\text{sinc}(x) = \begin{cases} 1 & \text{for } x = 0 \\ \frac{\sin(x)}{x} & \text{otherwise} \end{cases}$$

- The **normalized** *sinc* function used in the context of *digital signal processing* is given as:

$$\text{sinc}(\pi x) = \begin{cases} 1 & \text{for } x = 0 \\ \frac{\sin(\pi x)}{\pi x} & \text{otherwise} \end{cases}$$



# Computing Fourier Transform – of a **box function**



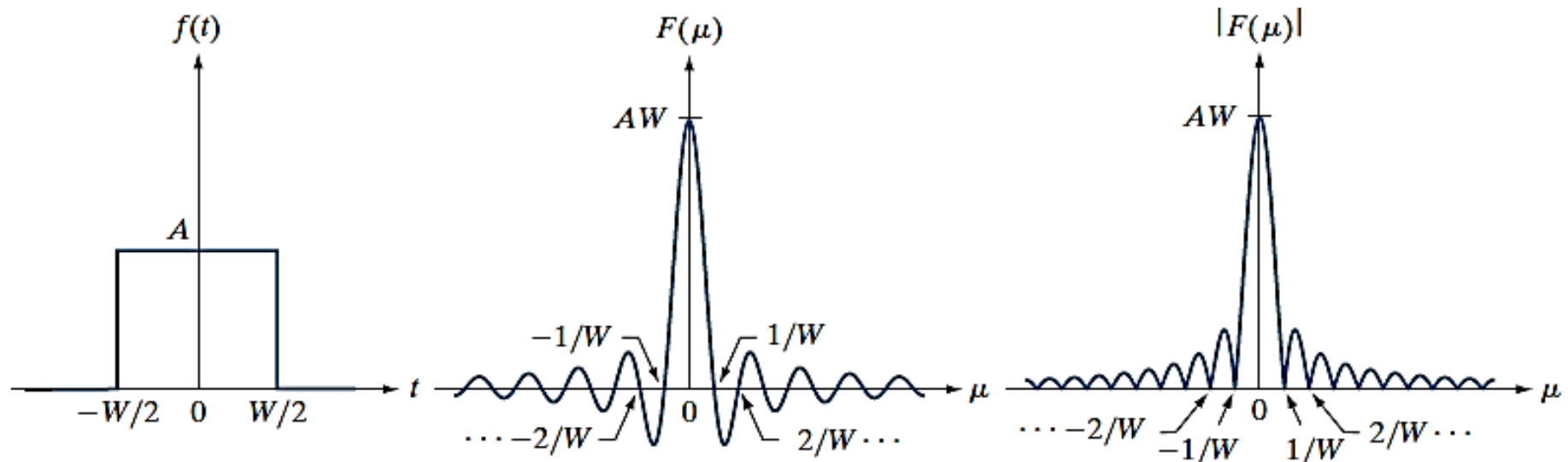
**A box function**

$$f(t) = \begin{cases} A & \text{if } -\frac{W}{2} \leq t \leq \frac{W}{2} \\ 0 & \text{otherwise} \end{cases}$$

- Compute Fourier Transform for the box function

# Compute Fourier Transform – of a **box function**

$$\begin{aligned}
 F(\mu) &= \int_{-\infty}^{\infty} f(t) e^{-j2\pi\mu t} dt = \int_{-W/2}^{W/2} A e^{-j2\pi\mu t} dt \quad , \text{ since } \int_a^b e^{kt} dt = \frac{e^{kt}}{k} \Big|_a^b \\
 &= \frac{-A}{j2\pi\mu} \left[ e^{-j2\pi\mu t} \right]_{-W/2}^{W/2} = \frac{-A}{j2\pi\mu} \left[ e^{-j\pi\mu W} - e^{j\pi\mu W} \right] \\
 &= \frac{A}{j2\pi\mu} \left[ e^{j\pi\mu W} - e^{-j\pi\mu W} \right] \\
 &= AW \frac{\sin(\pi\mu W)}{(\pi\mu W)} \quad \text{where the trigonometric identity } \sin \theta = (e^{j\theta} - e^{-j\theta}) / 2j
 \end{aligned}$$



**A box function**

**its Fourier transform**

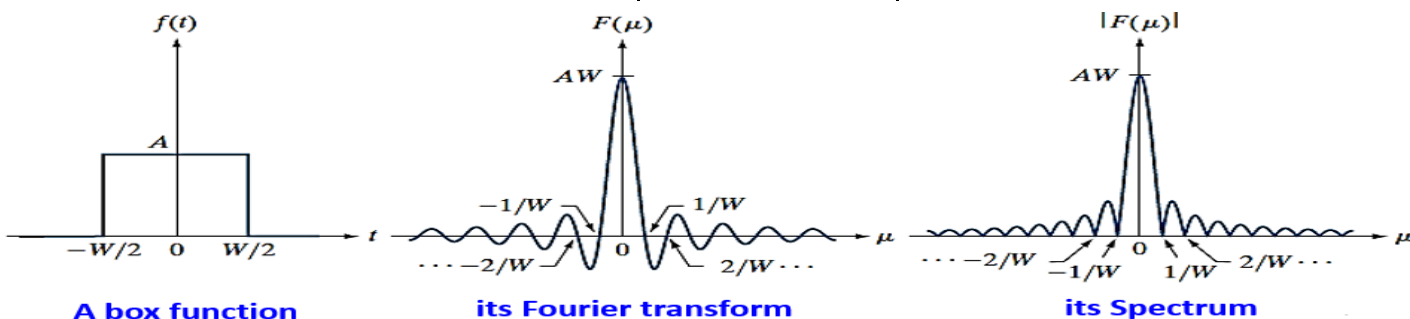
**its Spectrum**

# Compute Fourier Transform – of a **box function**

## Some Observations:

- Locations of zeroes in  $F(\mu)$  and  $|F(\mu)|$  are **inversely proportional** to  $W$ , the width of the “box” function.
- The height of the function **decreases** away from the origin.
- The function extends to **infinity** in both directions for variable  $\mu$ .
- The Fourier transform contains complex terms, and for display purposes, we work with the **magnitude of the transform** (a real quantity), which is called the *Fourier spectrum* or the *frequency spectrum*:

$$|F(\mu)| = AW \left| \frac{\sin(\pi\mu W)}{(\pi\mu W)} \right|$$





# Compute Fourier Transform – of an **impulse**

- The Fourier transform of a **unit impulse** located at the origin is:

$$\mathfrak{F}\{\delta(t)\} = F(\mu) = \int_{-\infty}^{\infty} \delta(t) e^{-j2\pi\mu t} dt = \int_{-\infty}^{\infty} e^{-j2\pi\mu t} \delta(t) dt = e^{-j2\pi\mu}$$

where we used the sifting property  $\int_{-\infty}^{\infty} f(t) \delta(t) dt = f(0)$

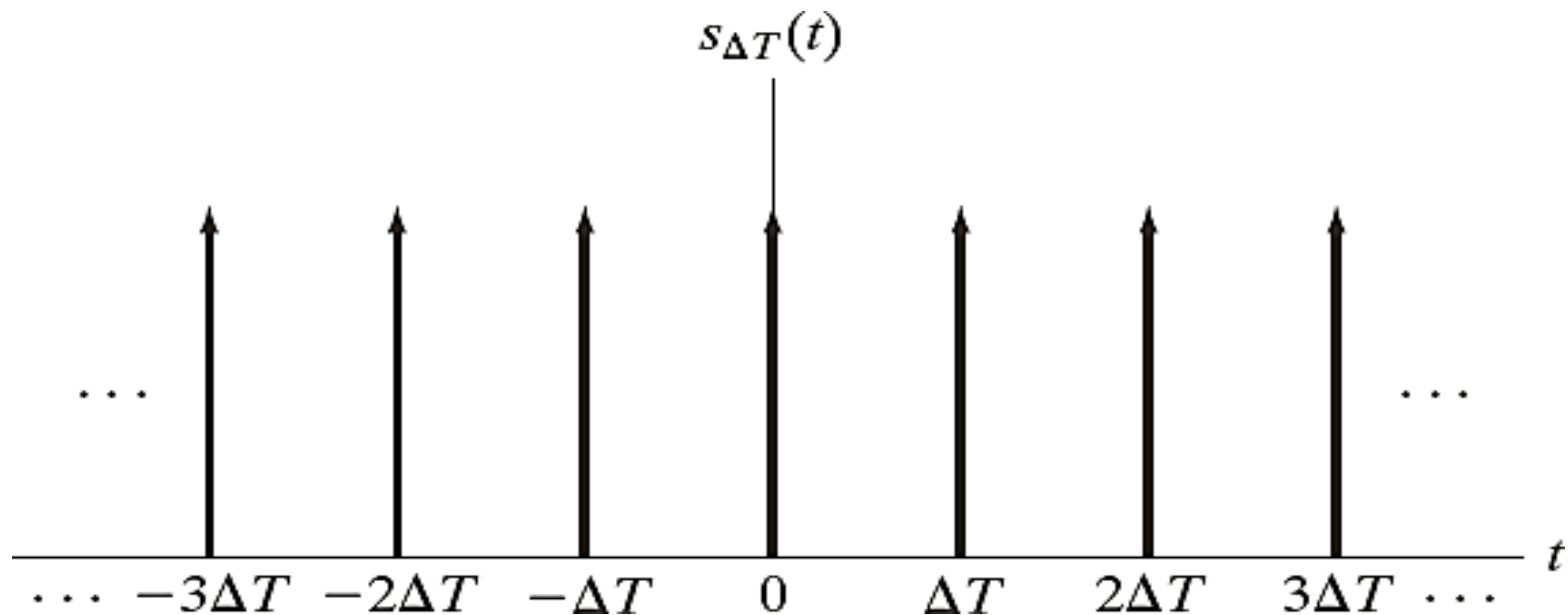
- Similarly, the Fourier transform of an impulse located at  $t = t_0$  is:

$$\mathfrak{F}\{\delta(t - t_0)\} = F(\mu) = \int_{-\infty}^{\infty} \delta(t - t_0) e^{-j2\pi\mu t} dt = \int_{-\infty}^{\infty} e^{-j2\pi\mu t} \delta(t - t_0) dt = e^{-j2\pi\mu t_0}$$
$$\int_{-\infty}^{\infty} f(t) \delta(t - t_0) dt = f(t_0)$$

The Fourier transform of an **impulse** located at the **origin** of the *spatial domain* is a **constant** in the *frequency domain*.

# Compute Fourier Transform – of a **periodic impulse train**

**Recall:** Impulse Train



$$s_{\Delta T}(t) = \sum_{k=-\infty}^{\infty} \delta(t - k\Delta T)$$

# Compute Fourier Transform – of a **periodic impulse train**

Fourier Series is given by:

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{j\frac{2\pi n}{T}t}$$

Where,

$$c_n = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-j\frac{2\pi n}{T}t} dt \quad \text{for } n = 0, \pm 1, \pm 2, \dots$$

The impulse train  $s_{\Delta T}(t)$  can be expressed as a the following Fourier series:

$$s_{\Delta T}(t) = \sum_{n=-\infty}^{\infty} c_n e^{j\frac{2\pi n}{\Delta T}t}$$

Where,

$$c_n = \frac{1}{\Delta T} \int_{-\Delta T/2}^{\Delta T/2} s_{\Delta T}(t) e^{-j\frac{2\pi n}{\Delta T}t} dt$$

# Compute Fourier Transform – of a **periodic impulse train**

Simplify  $c_n$  and we have:  $c_n = \frac{1}{\Delta T}$

The [Fourier Series](#) then becomes:

$$s_{\Delta T}(t) = \sum_{n=-\infty}^{\infty} c_n e^{j\frac{2\pi n}{\Delta T}t} = \frac{1}{\Delta T} \sum_{n=-\infty}^{\infty} e^{j\frac{2\pi n}{\Delta T}t}$$

Finally, the [Fourier transform](#) of this Fourier series becomes:

$$S(\mu) = \mathfrak{F}\{s_{\Delta T}(t)\} = \mathfrak{F}\left\{\frac{1}{\Delta T} \sum_{n=-\infty}^{\infty} e^{j\frac{2\pi n}{\Delta T}t}\right\} = \frac{1}{\Delta T} \mathfrak{F}\left\{\sum_{n=-\infty}^{\infty} e^{j\frac{2\pi n}{\Delta T}t}\right\} = \frac{1}{\Delta T} \sum_{n=-\infty}^{\infty} \delta\left(\mu - \frac{n}{\Delta T}\right)$$

# Compute Fourier Transform – of a **periodic impulse train**

Compare Impulse train **with its** Fourier transform:

$$s_{\Delta T}(t) = \sum_{k=-\infty}^{\infty} \delta(t - k\Delta T) \quad \text{Vs.} \quad \mathfrak{F}\{s_{\Delta T}\} = \frac{1}{\Delta T} \sum_{n=-\infty}^{\infty} \delta\left(\mu - \frac{n}{\Delta T}\right)$$

- The Fourier Transform of an impulse train with period  **$\Delta T$**  is also an **impulse train**, whose period is  **$1/\Delta T$**
- This *inverse proportionality* between the periods of  $s_{\Delta T}(t)$  and  $s(\mu)$  is analogous to the “box” function transform where *Zero-crossings* were inversely proportional to  $W/2$ .

# Compute Fourier Transform – of a **periodic impulse train**

Compare Impulse train **with its** Fourier transform:

$$s_{\Delta T}(t) = \sum_{k=-\infty}^{\infty} \delta(t - k\Delta T) \quad \text{Vs.} \quad S(\mu) = \mathfrak{F}\{s_{\Delta T}\} = \frac{1}{\Delta T} \sum_{n=-\infty}^{\infty} \delta\left(\mu - \frac{n}{\Delta T}\right)$$

- The Fourier Transform of an impulse train with period  **$\Delta T$**  is also an **impulse train**, whose period is  **$1/\Delta T$**
- There is *inverse proportionality* between the periods of  $s_{\Delta T}(t)$  and  $s(\mu)$ .

# Convolution of Continuous Functions

# Convolution

- Convolution is a **sliding window** representation in **spatial domain**
  - Rotate by  $180^\circ$  (flip) one function (**kernel**) and **slide it over** the second function.
  - **At each displacement in the sliding process, compute the sum of products in the local neighborhood and replace the value at the origin of the kernel with this new value.**
- We are now interested in the **convolution** of **two spatial domain continuous functions of one continuous variable**, **t**, say  $f(t)$  and  $h(t)$  **and its equivalent in the frequency domain.**



# Convolution

- Let  $f(t)$  and  $h(t)$  be the *two spatial domain continuous functions of one continuous variable,  $t$* .
- The convolution of these two functions is defined as:

$$(f \star h)(t) = \int_{-\infty}^{\infty} f(\tau) h(t - \tau) d\tau$$

where the **minus sign** accounts for the *flipping* of kernel function,  $t$  is the *displacement* needed to slide one function past the other, and  $\tau$  is a *dummy variable* that is integrated out.

- What is the Fourier transform of  $(f \star h)(t)$ ?

# Convolution

$$\begin{aligned}\mathfrak{F}\{(f \star h)(t)\} &= \int_{-\infty}^{\infty} \left[ \int_{-\infty}^{\infty} f(\tau) h(t - \tau) d\tau \right] e^{-j2\pi\mu t} dt \\ &= \int_{-\infty}^{\infty} f(\tau) \left[ \int_{-\infty}^{\infty} h(t - \tau) e^{-j2\pi\mu t} dt \right] d\tau\end{aligned}$$

$$F(\mu) = \int_{-\infty}^{\infty} f(t) e^{-j2\pi\mu t} dt$$

**claim:**  $\mathfrak{F}\{h(t - \tau)\} = H(\mu) e^{-j2\pi\mu\tau}$  where  $H(\mu)$  is the Fourier transform of  $h(t)$ .

$$\begin{aligned}\mathfrak{F}\{(f \star h)(t)\} &= \int_{-\infty}^{\infty} f(\tau) [H(\mu) e^{-j2\pi\mu\tau}] d\tau \\ &= H(\mu) \int_{-\infty}^{\infty} f(\tau) e^{-j2\pi\mu\tau} d\tau \\ &= H(\mu) F(\mu) \\ &= (H \bullet F)(\mu)\end{aligned}$$

$$\mathfrak{F}\{(f \star h)(t)\} = (H \bullet F)(\mu)$$
 where **dot (.)** indicates **multiplication**

# Convolution

**Note:** The Fourier Transform of the **convolution of two functions in the spatial domain** is equal to the **product in the frequency domain** of the Fourier transforms of these two functions. **The converse statement is also true.**

**The Convolution Theorem:**

$$(f \star h)(t) \Leftrightarrow (H \bullet F)(\mu)$$

$$(f \bullet h)(t) \Leftrightarrow (H \star F)(\mu)$$

# Next Lecture

- 1-D Sampling
  - Sampling Revisited
  - Sampling Theorem
  - Signal Recovery
- 2-D Sampling
- Aliasing
- Aliasing in Images
  - How to reduce the effects of spatial aliasing?
  - Moiré Patterns
  - Halftoning