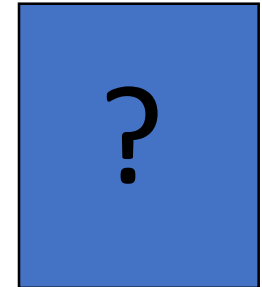
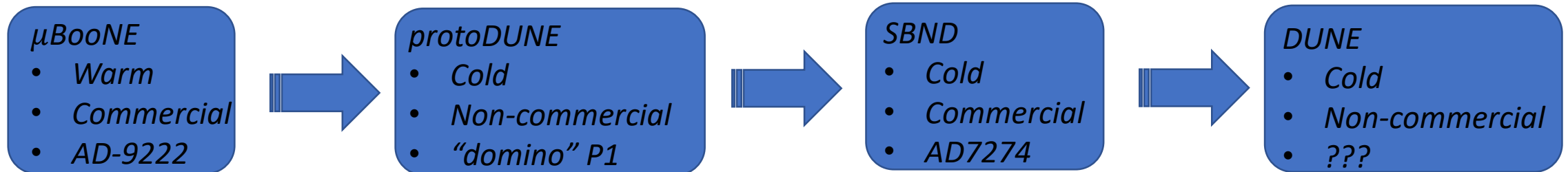


ADC Calibration in LArTPC

Wenqiang Gu, Brookhaven National Laboratory
for the DUNE collaboration

*Workshop on Calibration and Reconstruction for LArTPCs
December 10th 2018*

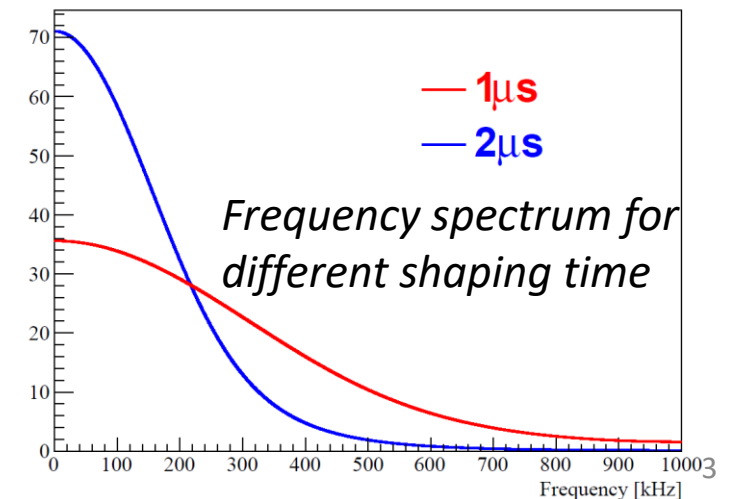
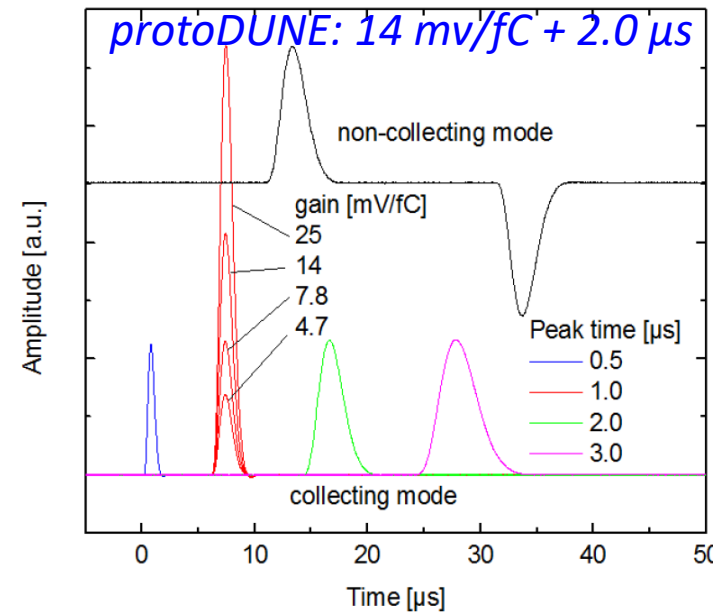
Evolution of ADC in LArTPC



- Cold ADC is essential for DUNE

How to choose an ADC?

- Dynamic range
 - 300 fC maximum ionized charge, 0.1 fC noise
→ 12-bit ADC
 - Preamp gain 4.7 mV/fC → ADC 0 ~ 1.6V
- Digitization frequency
 - Nyquist theorem: 1 MHz for 2 μ s shaping
 - 2 MHz oversampling is helpful for improving resolution, sticky code mitigation, etc.

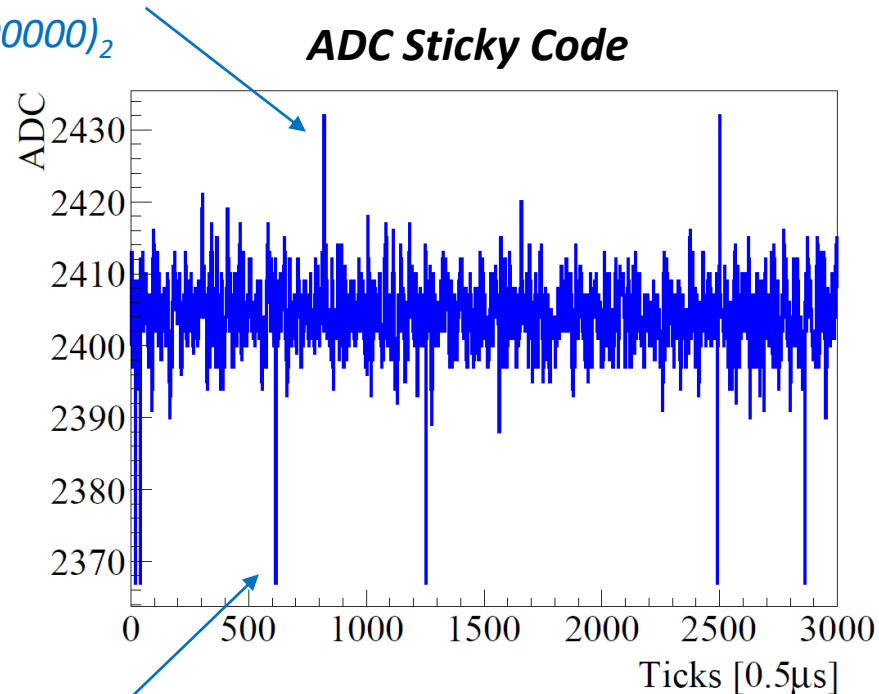


Precise ADC determination in protoDUNE

- However, given the cold environment in LAr, two problems occur for the precise determination of ADC

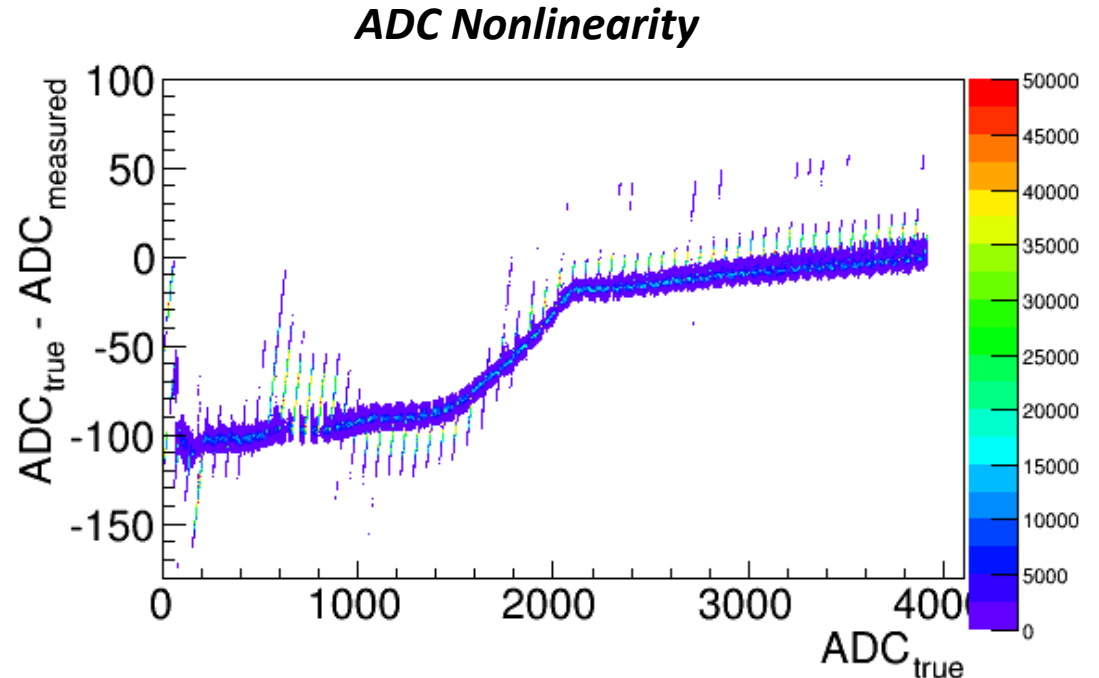
2432

$= (100110\ 000000)_2$



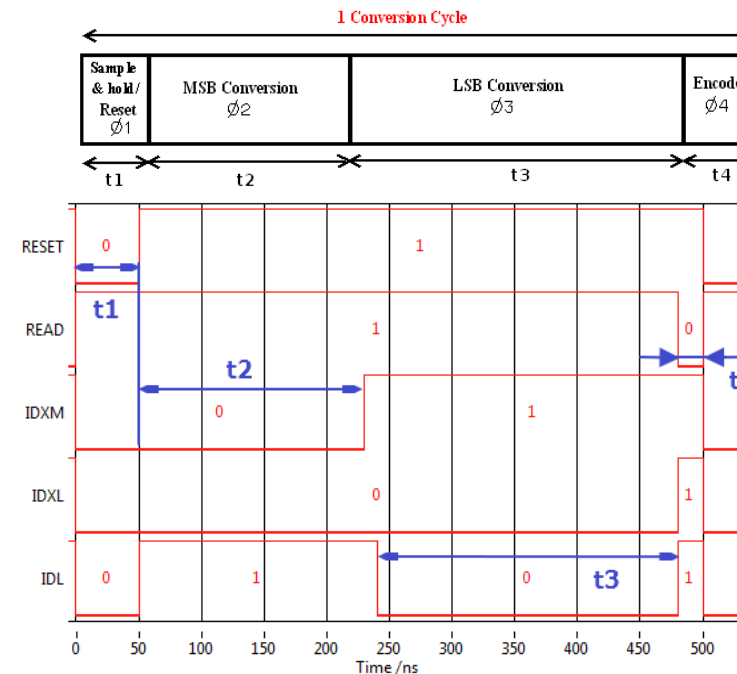
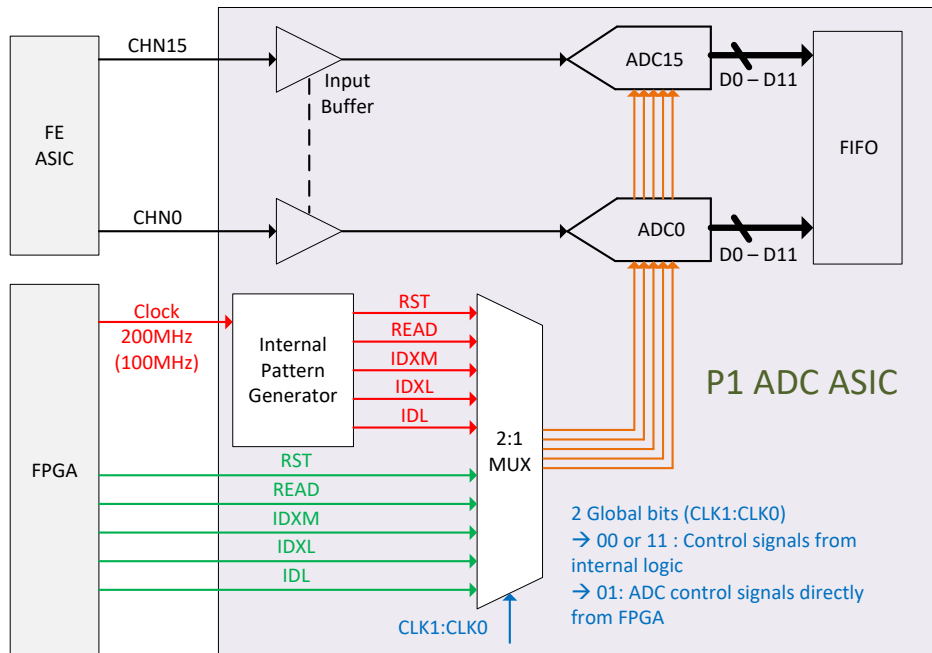
2367

$= (100100\ 111111)_2$

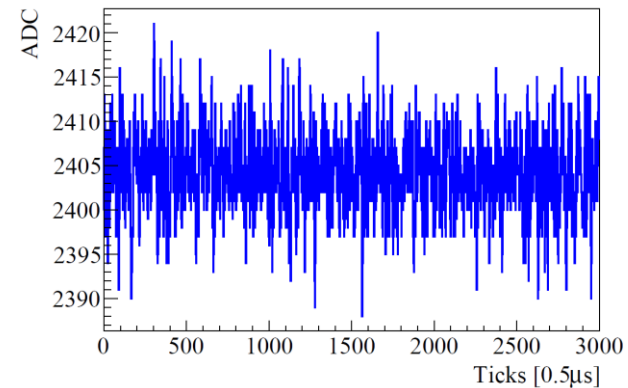
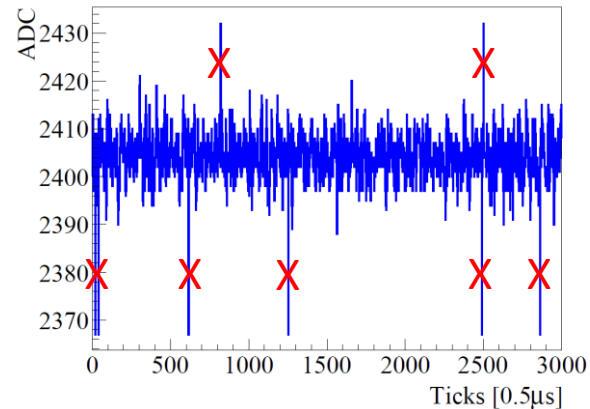
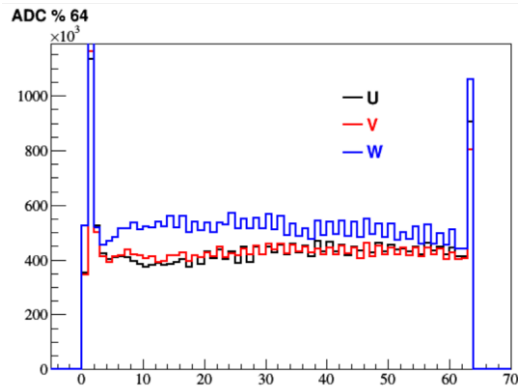


ADC readout scheme in protoDUNE

- Read/write logic synchronized through five control signals
- In cold environment, an instability in bit conversion results in sticky code and nonlinearity

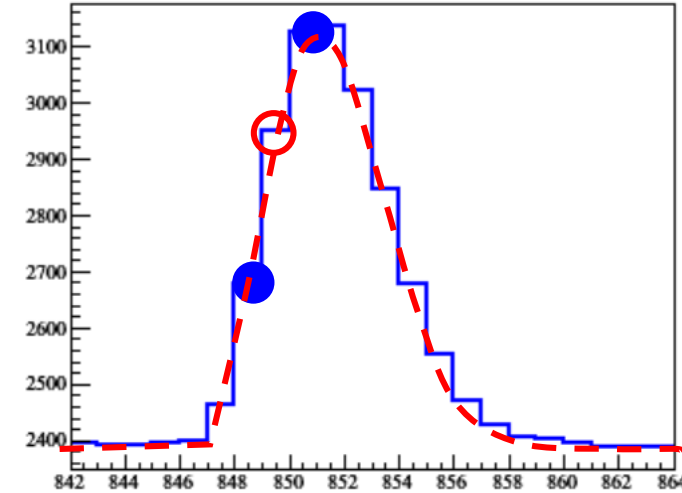
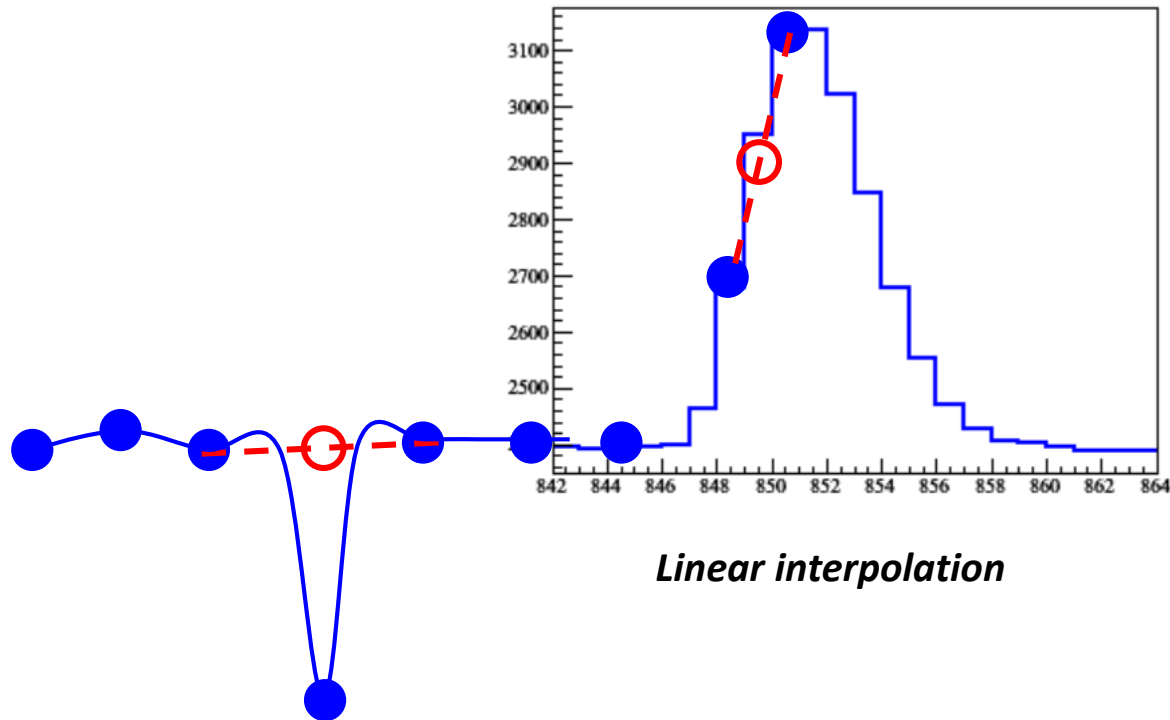


Procedure of sticky code mitigation



Sticky code interpolation

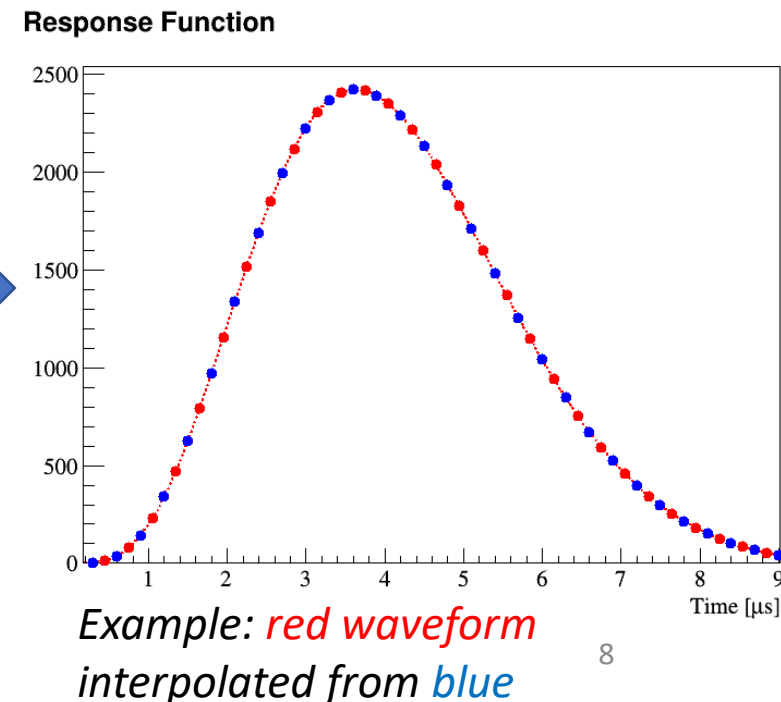
- Linear interpolation between “un-sticky” codes is a good first step
 - *2 MHz oversampling is helpful for interpolation*
- However, linear interpolation may not be sufficient for signal region
- A correction w.r.t. the electronics response function would be better



FFT interpolation w.r.t electronics response

- However, some facts makes it difficult for using the response function
 - A few percent channel-to-channel variation in response function
 - Coupled with ADC nonlinearity (discussed shortly)
- Instead, a **FFT interpolation** is proposed by
 - i) Linear interpolation as a base correction
 - ii) Once a “sticky” code found in an **even-binned** tick, apply phase shift to **odd-binned ticks** to cover **even-binned ticks**, and *vice versa*

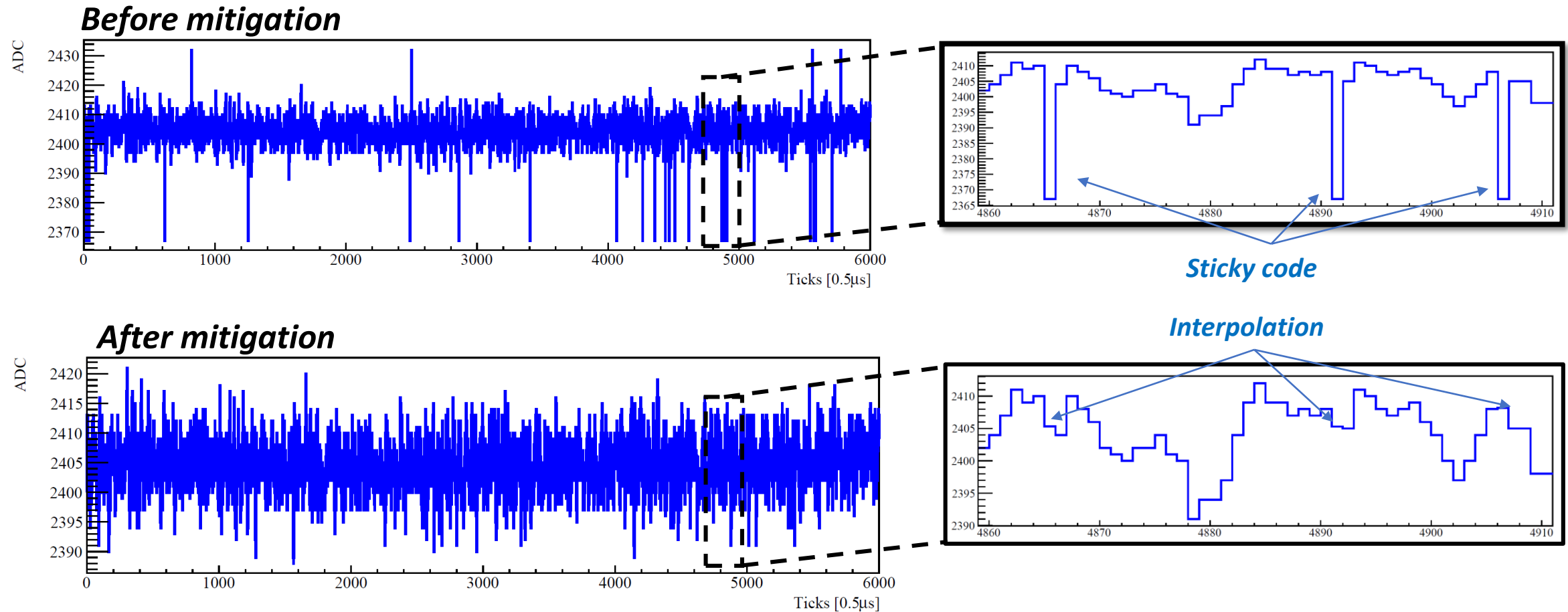
FT property	Time domain	Frequency domain
	$f(t)$	$F(\omega)$
Phase shift	$f(t - t_0)$	$F(\omega)e^{-j\omega t_0}$



Advantages of such FT interpolation

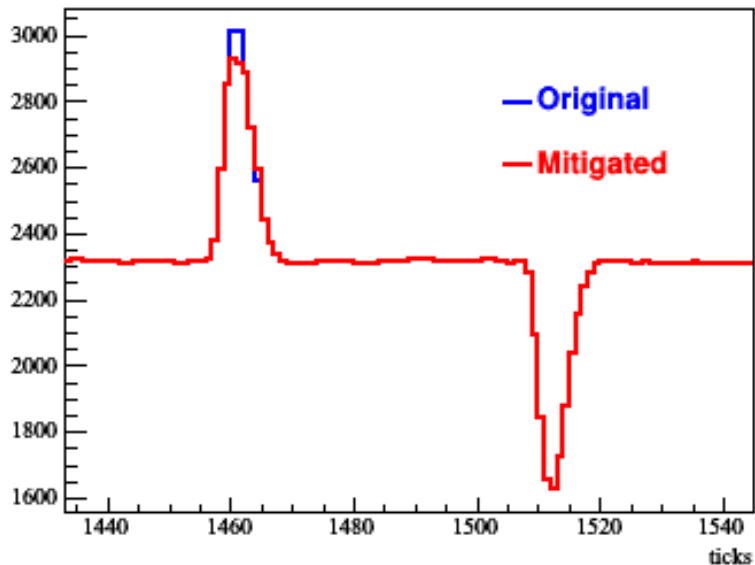
- Only the phase changed, while no changes of the magnitude in the frequency domain
 - Still respect the shape of the electronics response function
- Sometimes, good code tagged as “sticky”, the FT interpolation presumably minimize the biases
 - Balance of efficiency and accuracy for sticky code tagging

Performance of the ADC mitigation

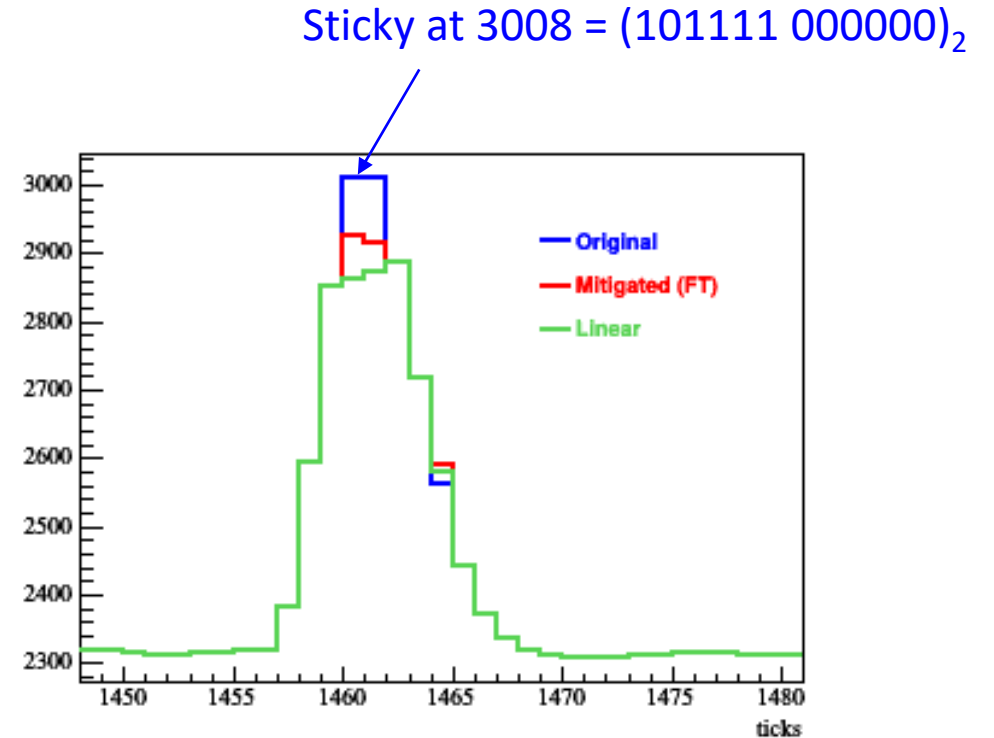
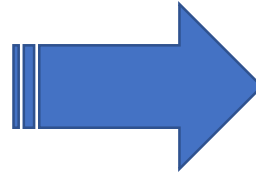


A special case

Preamplifier waveform



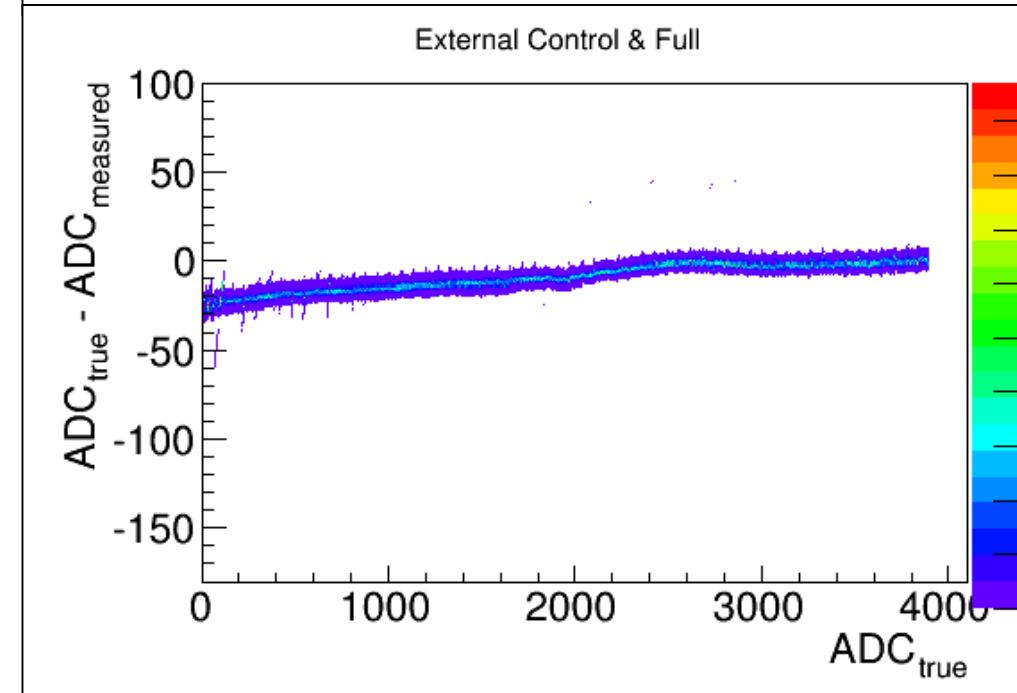
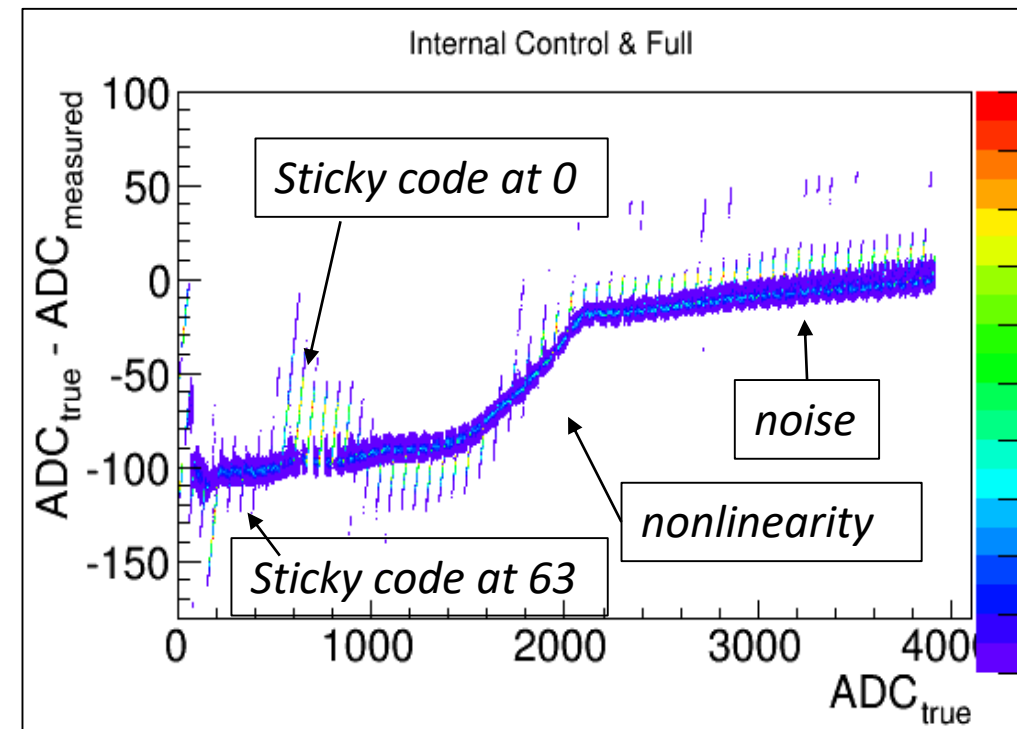
Zoom-in



- However, when two adjacent sticky codes happens on the peak region, the mitigation does not work well
- Need to improve this special case
 - Maybe ignore the base correction from linear interpolation

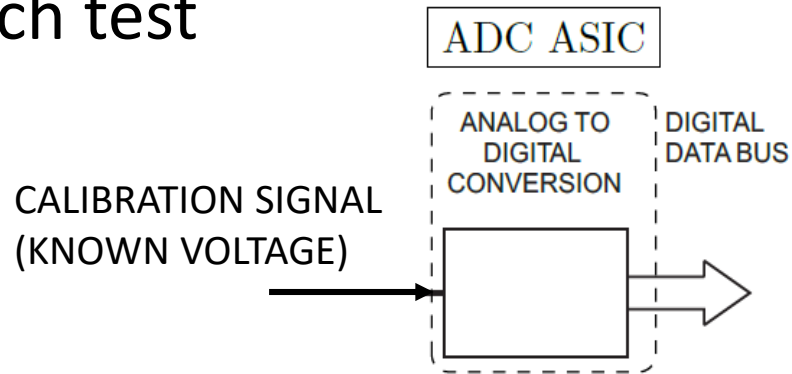
ADC nonlinearity (NL)

- Low temperature degrades the electronics and read/write logics
- External clock eases NL as well as sticky code
- NL is sensitive to clock settings
- A precise determination of ADC NL is very important for the extraction of ionized electrons



Difficulties from a bench test to protoDUNE

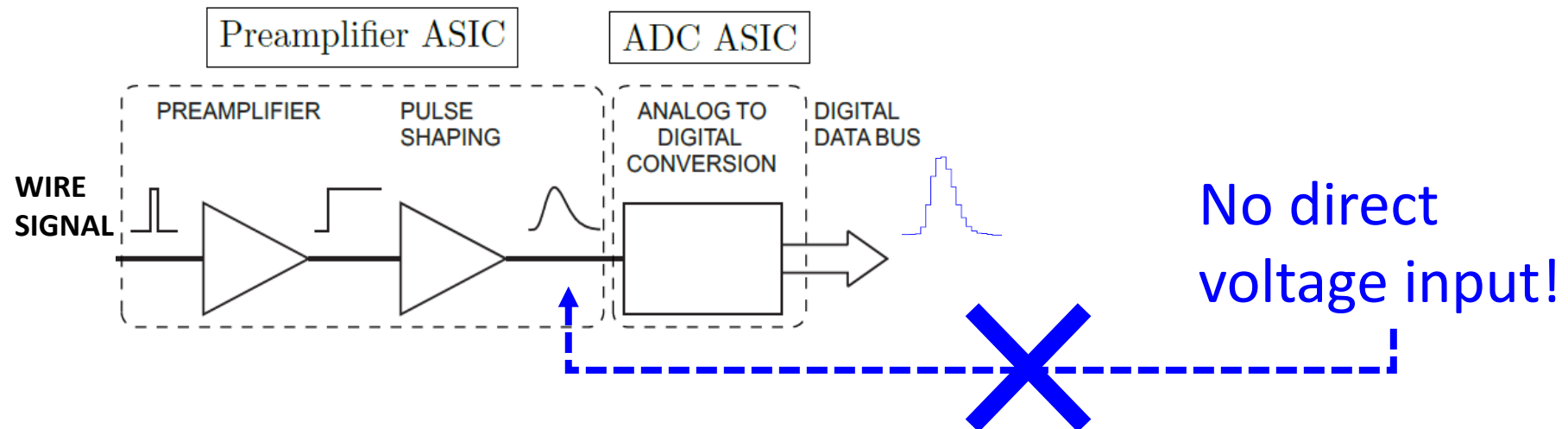
- Bench test



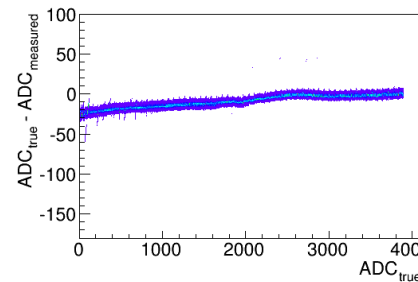
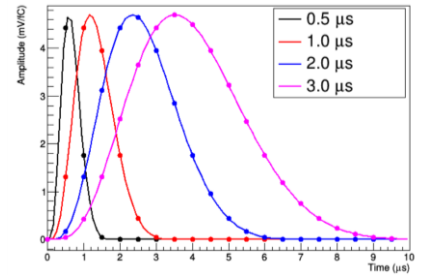
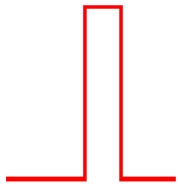
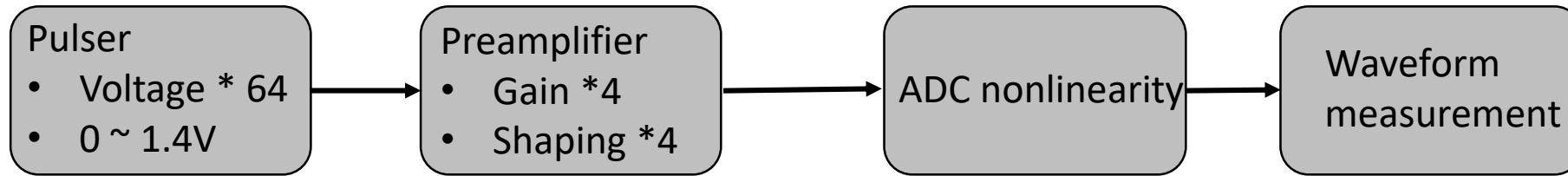
NL is sensitive to clock settings

- (bench test) clock is tuned for each ADC
- (protoDUNE) one clock shared by four ADC circuits

- ProtoDUNE



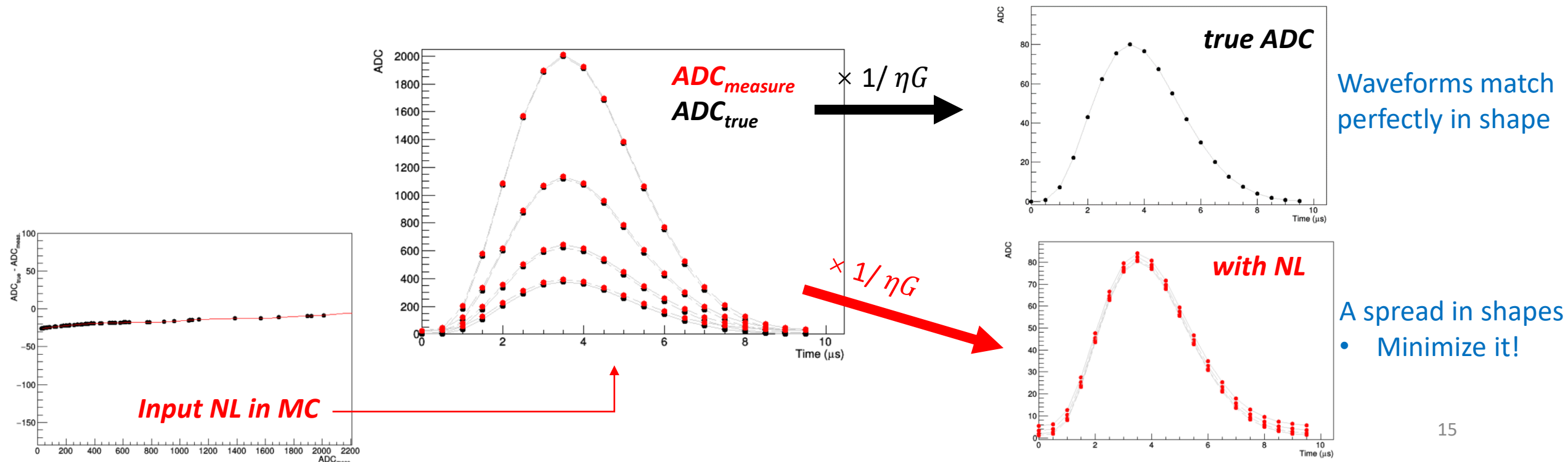
ADC calibration setup



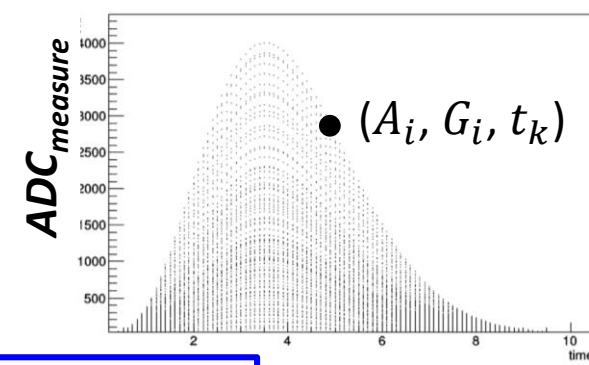
- Similar setup as in MicroBooNE electronics calibration

Calibration scheme independent of the shape of response function

- Assume pulse voltage (η) and preamp gain (G) **do not change the shape of electronics response**
- NL distorts the shape differently for high ADC and low ADC



χ^2 minimization



index of the α -th iteration

G_i = pulse voltage \times preamp gain
(A normalization of charge input)

$$\chi^{2(\alpha)} = \sum_i \left(A_i + f^{(\alpha)}(A_i) - R^{(\alpha-1)}(t_k) G_i \right)^2$$

Time tick for A_i

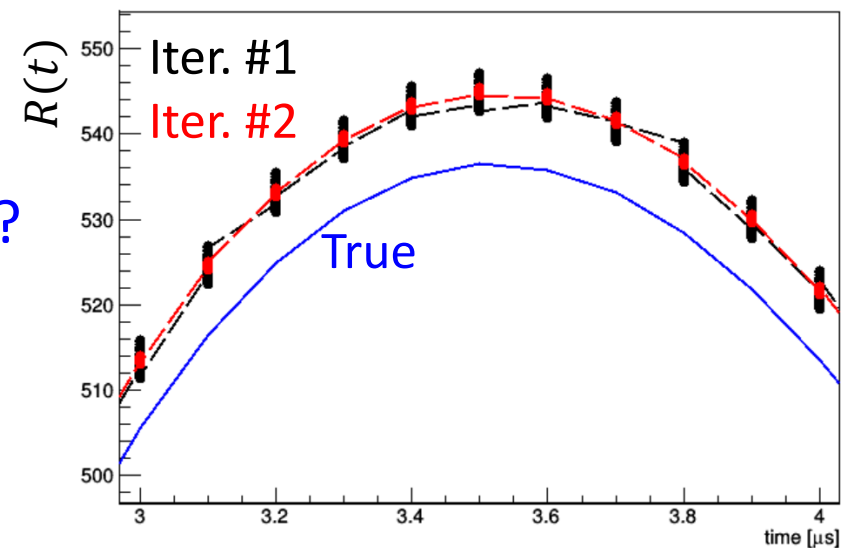
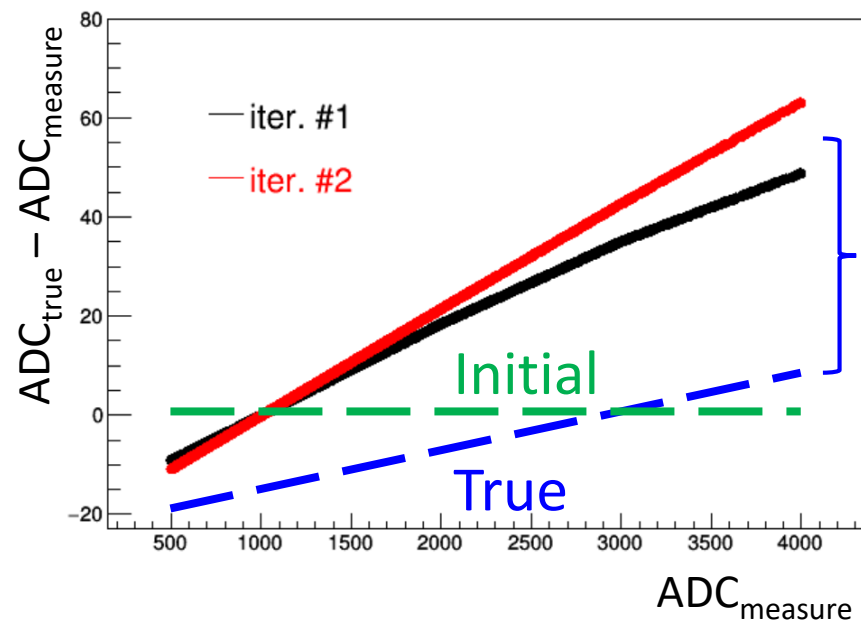
i -th data point

NL correction function
(To be fitted out)

$R^{(\alpha-1)}(t_k) = \frac{1}{N} \sum \frac{A_i + f^{(\alpha-1)}(A_i)}{G_i}$ is the
effective response function
(Calculated with the NL function from best fit of previous iteration)

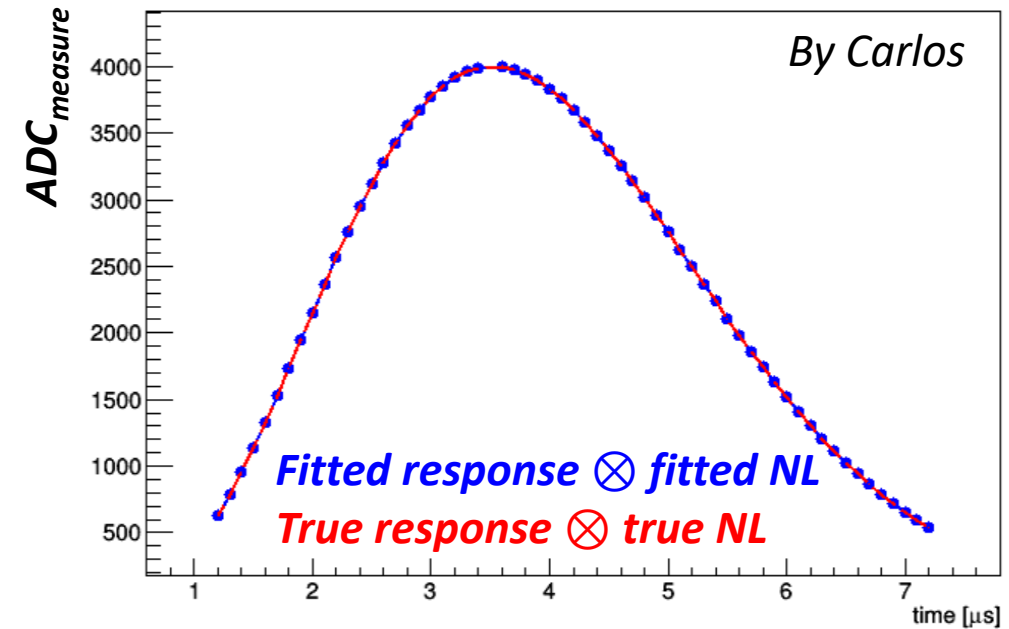
“Best-fit” $f(A_i)$ and $R(t)$ from simulation

- Given an initial value of NL correction function $f(A_i)$
- After a few iterations, “best-fit” NL $f(A_i)$ and effective response $R(t)$ tends to be stable
- The spread in $R(t)$ significantly shrinks after minimization



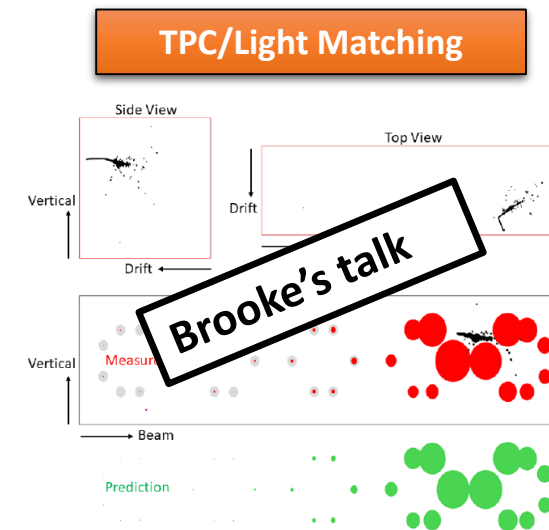
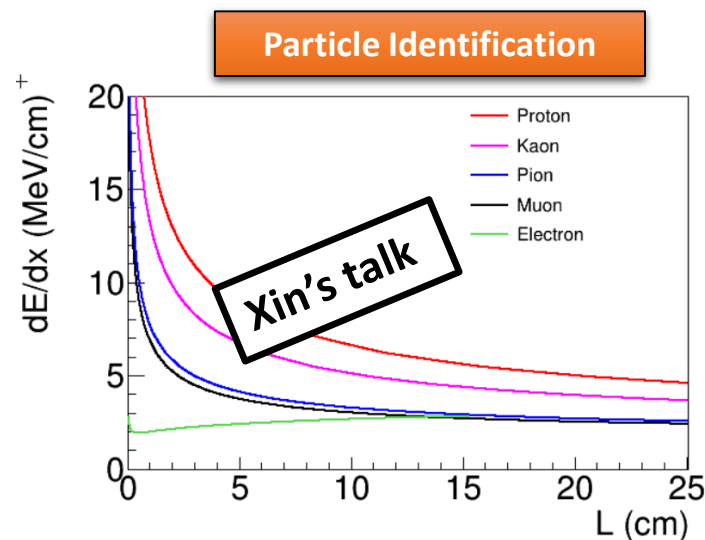
MC validation of the degeneracy

- Given a same charge input, the waveform predictions are close (<1 ADC) for
 - True response and NL
 - A “best-fit” effective response and NL
- NL bias in “best-fit” is not a problem!



Discussion: ADC impact on physics analysis

- ADC nonlinearity calibration is necessary for a precise extraction of ionized electrons
- Presumably, an important input for any analysis related with energy/charge
 - e.g., particle identification, TPC/light matching



Summary

- Cold electronics is essential for LArTPC experiments
- Sticky code mitigation
 - An interpolation approach through FT is proposed and studied in protoDUNE
- ADC nonlinearity
 - A calibration strategy was proposed and preliminarily studied in simulation