

#### Signals and Systems (Lab)

#### Lab 4: The Continuous-Time Fourier Transform

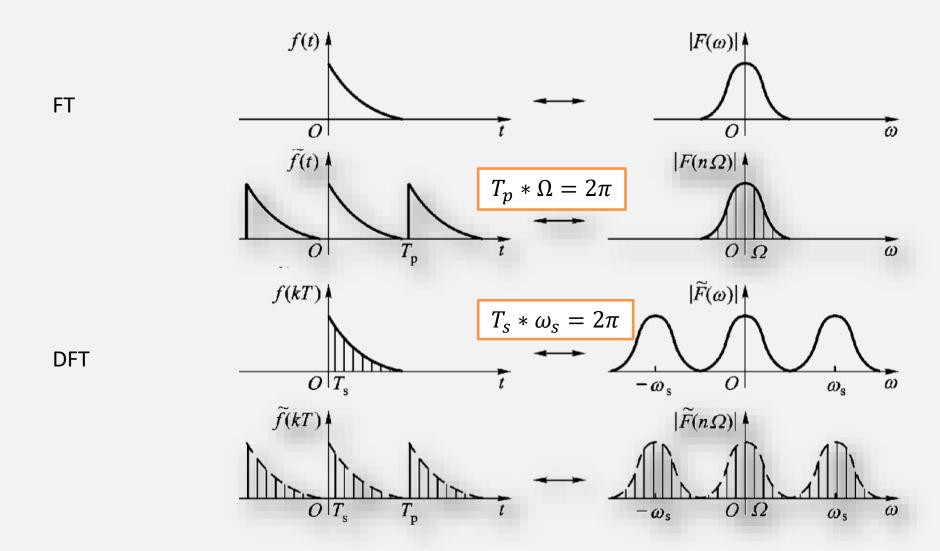
Dr. Wu Guang

wug@sustech.edu.cn

**Electrical & Electronic Engineering Southern University of Science and Technology** 

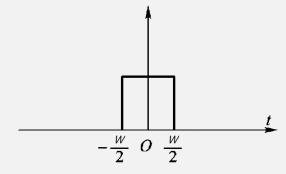
#### Review

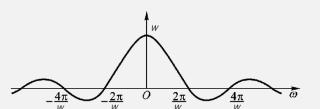
- ✓ 1. How to calculate the output of DT LTI system in frequency domain
- ✓ 2. How to calculate the output of CT LTI system
- √ 3. How to calculate the *DTFS* of signal via Matlab



#### Overview

- > CT Fourier Transform (CTFT):
  - > How to calculate via Matlab?
- > Frequency Response:
  - > How to calculate via Matlab?
  - > How to convert to impulse response
- > Application of CTFT:
  - ➤ Analysis of amplitude modulation (AM)



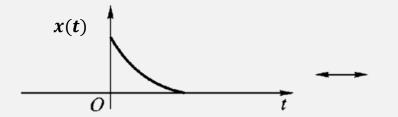


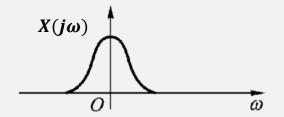
#### CT Fourier Transform

Definition of CT Fourier transform:

$$\mathbf{X}(\boldsymbol{j}\boldsymbol{\omega}) = \int_{-\infty}^{\infty} x(t)e^{-\boldsymbol{j}\boldsymbol{\omega}t}dt$$

- Function of continuous frequency  $\omega$
- Represent the "spectrum" of a signal





#### Use fft() to calculate the CTFT

CT Fourier Transform:  $X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t}dt$ 



DT Fourier Series:  $a_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-jk(\frac{2\pi}{N})n}$ 

Let 
$$t = n\tau$$

**CT Fourier Transform:** 

$$X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t}dt$$
$$= \lim_{\tau \to 0} \tau \sum_{n=-\infty}^{\infty} x(n\tau)e^{-j\omega\tau n}$$



**DT Fourier Series:** 

$$a_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-jk\left(\frac{2\pi}{N}\right)n}$$

> Suppose the dominant region of x(t) is in [0,T], then we can use the following approximation

> 
$$X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t}dt \approx \int_{0}^{T} x(t)e^{-j\omega t}dt \approx \sum_{n=0}^{N-1} x(n\tau)e^{-j\omega n\tau}\tau$$

 $\triangleright$  It could be observed that  $\stackrel{\triangleright}{\sim}$  Slice into N intervals, each with length  $\tau$ =T/N

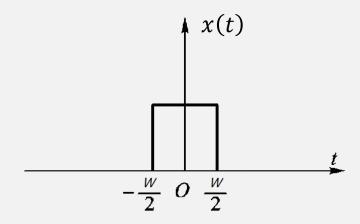
$$\succ X(j\frac{2\pi k}{N\tau}) \approx \tau \sum_{n=0}^{N-1} x(n\tau)e^{-j\frac{2\pi kn}{N}}$$

$$k = \frac{1-N}{2}, ..., -1, 0, 1, ..., \frac{N-1}{2}$$

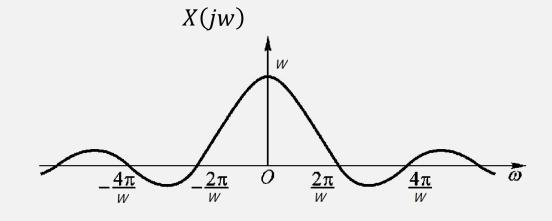
 $\rightarrow$  fft() of [x(0),  $x(\tau)$ ,  $x(2\tau)$ , ...,  $x((N-1)\tau)$ ]

Conclusion: we can use fft() to calculate CTFT approximately, which could reduce the computation complexity

## Example-sinc function



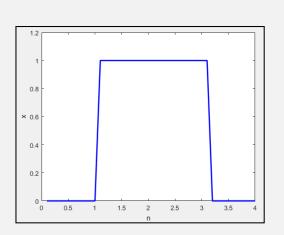
$$x(t) = \begin{cases} 1, t < |\frac{W}{2}| \\ 0, t > |\frac{W}{2}| \end{cases}$$

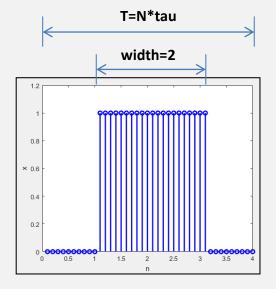


$$X(jw) = \left[\frac{\sin(\frac{\omega W}{2})}{\frac{\omega W}{2}}\right] = WSinc(\frac{\omega W}{2})$$

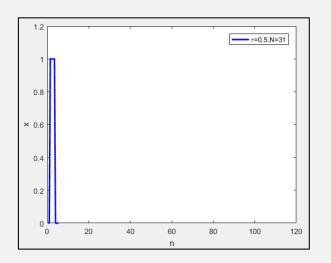
```
tau=0.5;
N=201; % N should be odd number,
N*tau=T>3
x=[zeros(1,1/tau),ones(1,2/tau+1),zeros(
1,N-3/tau-1)];
figure(1)
plot([1:N]*tau,x,'r-','LineWidth',2)
xlabel('n');ylabel('x');
axis([0 4 0 1.2])
legend('\tau=0.5,N=201')
y=fftshift(fft(x));
X=tau*y;
lb = (1-N)*pi/N/tau;
ub = (N-1)*pi/N/tau;
step = 2*pi/N/tau;
figure(2)
plot(lb:step:ub, abs(X),'r-','LineWidth',2);
xlabel('\omega');ylabel('abs(X)');
legend('\tau=0.5,N=201')
```

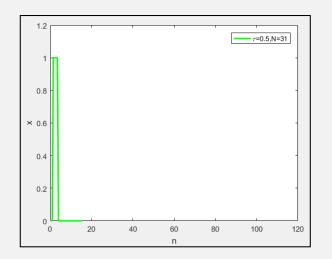
- **➤** Single rectangular wave with width=2
- > T should be larger than 3

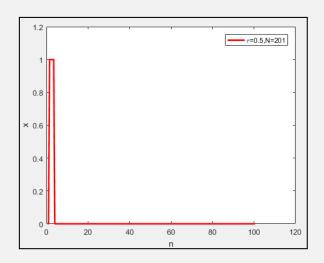




What happen if we change tau and N? Why?





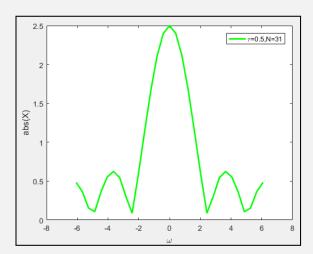




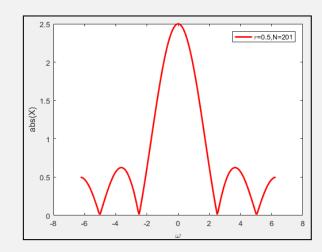
0.5

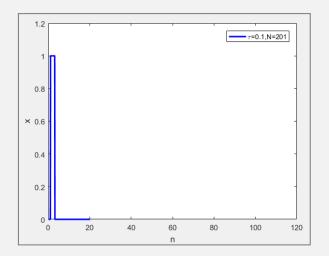


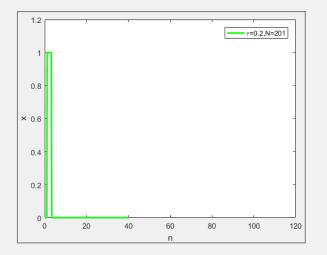
N=31

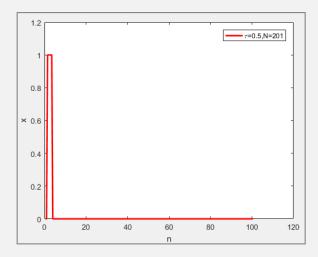


N=201

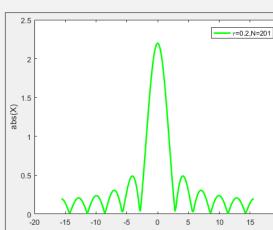




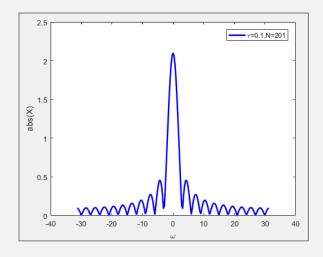


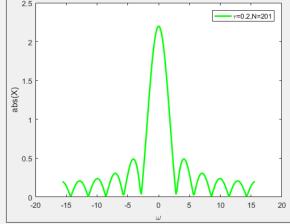


$$\tau = 0.1$$

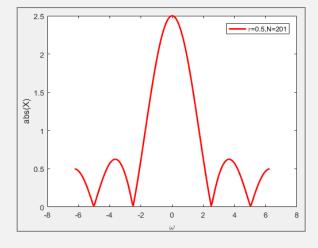


$$\tau = 0.5$$





 $\tau = 0.2$ 



## What's Frequency Response?

#### Frequency response

- A function of frequency
- System gain in frequency domain

$$e^{j\omega_0 t}$$
  $\Longrightarrow$  CT LTI System: H(jw)  $\Longrightarrow$   $H(j\omega_0)e^{j\omega_0 t}$ 

#### Math definition

Fourier transform of impulse response

$$H(j\omega) = \int_{-\infty}^{\infty} h(t)e^{-j\omega t}dt = Y(j\omega)/X(j\omega)$$

## Differential Equation and Frequency Response

$$\sum_{k=0}^{N} a_k \frac{d^k y(t)}{dt^k} = \sum_{k=0}^{M} b_k \frac{d^k x(t)}{dt^k}$$

$$H(j\omega) = \frac{b_M (j\omega)^M + b_{M-1} (j\omega)^{M-1} + \dots + b_1 (j\omega) + b_0}{a_N (j\omega)^N + a_{N-1} (j\omega)^{N-1} + \dots + a_1 (j\omega) + a_0}$$

Proof: 
$$\mathcal{F}\left[\sum_{k=0}^{N}a_{k}\frac{d^{k}y(t)}{dt^{k}}\right] = \sum_{k=0}^{N}\mathcal{F}\left[a_{k}\frac{d^{k}y(t)}{dt^{k}}\right] = \sum_{k=0}^{N}a_{k}(j\omega)^{k}Y(j\omega)$$

$$\mathcal{F}\left[\sum_{k=0}^{M}b_{k}\frac{d^{k}x(t)}{dt^{k}}\right] = \sum_{k=0}^{M}\mathcal{F}\left[b_{k}\frac{d^{k}x(t)}{dt^{k}}\right] = \sum_{k=0}^{M}b_{k}(j\omega)^{k}X(j\omega)$$

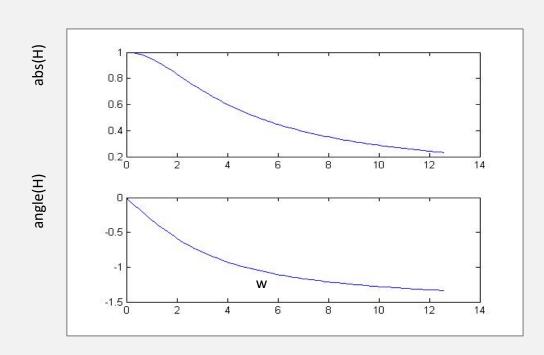
## Matlab Function: freqs()

- > **Description:** generate the frequency response of LTI system
- > Syntax: freqs(b,a,w)
- > Example:

$$\frac{dy(t)}{dt} + 3y(t) = 3x(t)$$

```
a=[1 3];
b=3;
w=linspace(0,4*pi);
H=freqs(b,a,w);
subplot(2,1,1), plot(w,abs(H));
subplot(2,1,2),
plot(w,angle(H));
```

**High-pass or Low-pass?** 



## Frequency Response and Impulse Response





> Transform pair:

$$\geq e^{-at}u(t) \quad Re\{a\} > 0 \quad \Leftrightarrow \quad \frac{1}{i\omega + a}$$

> Example:

$$>\frac{3}{j\omega+3} \iff 3e^{-3t}u(t)$$

How about general fraction?

$$H(j\omega) = \frac{b_M(j\omega)^M + b_{M-1}(j\omega)^{M-1} + \dots + b_1(j\omega) + b_0}{a_N(j\omega)^N + a_{N-1}(j\omega)^{N-1} + \dots + a_1(j\omega) + a_0}$$

#### Partial Fraction Expansion

> Partial Fraction Expansion (No identical poles):

$$H(j\omega) = \frac{b_{M}(j\omega + z_{1})(j\omega + z_{2}) \dots (j\omega + z_{M})}{a_{N}(j\omega + p_{1})(j\omega + p_{2}) \dots (j\omega + p_{N})}$$

$$= \frac{A_{1}}{j\omega + p_{1}} + \frac{A_{2}}{j\omega + p_{2}} + \dots + \frac{A_{N}}{j\omega + p_{N}}$$

> Partial Fraction Expansion (with identical poles):

$$H(j\omega) = \frac{b_{M}(j\omega + z_{1})(j\omega + z_{2}) \dots (j\omega + z_{M})}{a_{N}(j\omega + p_{1})^{k_{1}}(j\omega + p_{2})^{k_{2}} \dots (j\omega + p_{n})^{k_{n}}} =$$

$$= \frac{A_{1,1}}{(j\omega + p_{1})^{k_{1}}} + \frac{A_{1,2}}{(j\omega + p_{1})^{k_{1}-1}} + \dots \frac{A_{1,k_{1}}}{(j\omega + p_{1})}$$

$$+ \dots +$$

$$\frac{A_{n,1}}{(j\omega + p_{n})^{k_{n}}} + \frac{A_{n,2}}{(j\omega + p_{n})^{k_{n}-1}} + \dots \frac{A_{n,k_{n}}}{(j\omega + p_{n})}$$

## Partial Fraction Expansion

Therefore, we should know the FT of

$$\frac{1}{(j\omega+a)^k}$$

We already know:

$$e^{-at}u(t)$$
  $Re\{a\} > 0 \iff \frac{1}{j\omega + a}$ 

**According to the FT property:** 

$$tx(t) \iff j\frac{d}{d\omega}X(j\omega)$$

We have:

$$te^{-at}u(t)$$
  $Re\{a\} > 0$   $\iff$   $\frac{1}{(j\omega + a)^2}$ 

•••

#### Matlab Function: residue()

- Convert between partial fraction expansion and polynomial coefficients
- [r,p,k] = residue(b,a); [r,p] = residue(b,a)

$$\frac{b(s)}{a(s)} = \frac{b_1 s^m + b_2 s^{m-1} + b_3 s^{m-2} + \dots + b_{m+1}}{a_1 s^n + a_2 s^{n-1} + a_3 s^{n-2} + \dots + a_{n+1}}$$

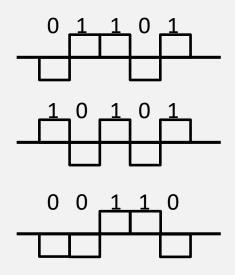
$$\frac{b(s)}{a(s)} = \frac{r_1}{s - p_1} + \frac{r_2}{s - p_2} + \dots + \frac{r_n}{s - p_n} + k(s)$$

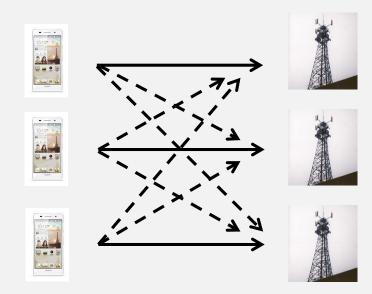
• Example:

$$\frac{b(s)}{a(s)} = \frac{5s^3 + 3s^2 - 2s + 7}{-4s^3 + 8s + 3}$$



#### What's Modulation





- > Each phone wants to deliver information to its base station
- > Therefore, there is cross-talk in the wireless channel
- > How can we solve this issue?

#### Amplitude Modulation

Signal: m(t)



Spectrum of m(t)

ightharpoonup Carrier:  $\cos(\omega_1 t) \Leftrightarrow \pi[\delta(\omega - \omega_1) + \delta(\omega + \omega_1)]$ 



**Spectrum of carrier** 

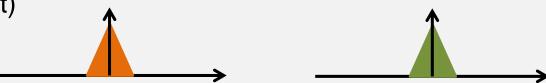
ightharpoonup Amplitude Modulation:  $y(t) = m(t) \times \cos(\omega_1 t)$ 



Shift by  $\omega_1$  and Scale by 1/2

**No information loss** 

 $\rightarrow$  Signals:  $m_1(t)$  and  $m_2(t)$ 



 $\triangleright$  Carriers:  $\cos(\omega_1 t)$  and  $\cos(\omega_2 t)$ 



ightharpoonup Amplitude Modulation:  $y(t) = m_1(t) \times \cos(\omega_1 t) + m_2(t) \times \cos(\omega_2 t)$ 



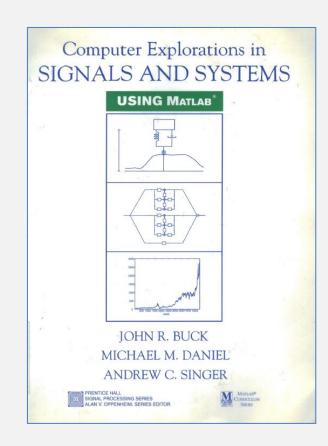
Signals are distinguished in frequency domain!

## Lab Assignments

- Read tutorial 4.1 by yourself
- 4.2 & 4.6

$$x(t) = m_1(t)\cos(2\pi f_1 t) + m_2(t)\sin(2\pi f_2 t) + m_3(t)\sin(2\pi f_1 t)$$

• Submit your report.



# Question ?

