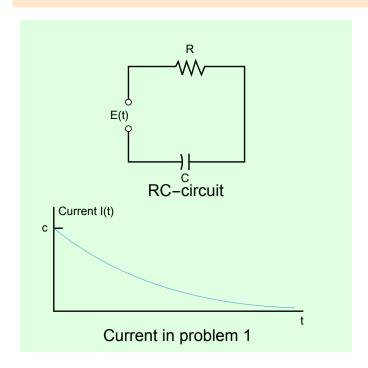
## 1 - 6 RLC-Circuits: special cases

1. RC-Circuit. Model the RC-Circuit in the figure below. Find the current due to a constant E.



## ClearAll["Global`\*"]

The problem is asking for a look at RC circuit, not RLC.

The site https://www.intmath.com/differential - equations/6 - rc - circuits.php assumes a constant voltage source, just what the problem specifies. Below: There is no inductance here, only R and C.

eqnw = rR (D[eye[t], t]) + eye[t] / cC == 0  

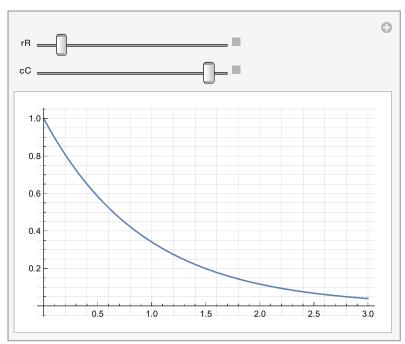
$$\frac{\text{eye[t]}}{\text{cC}} + \text{rR eye'[t]} == 0$$

Within a certain range of capacitance and resistance, the plot resembles the one in the problem description, and can be manipulated to imitate changing parameters, with the voltage remaining constant.

$$\left\{\left\{\text{eye} \rightarrow \text{Function}\left[\left\{t\right\}, \ e^{-\frac{t}{cC\,rR}}\,C[1]\right]\right\}\right\}$$

It looks like the current is normalized to 1 at t=0, and the fraction of its max value at a given time needs to be estimated from the underlying grid.

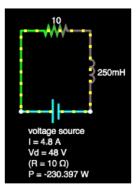
Manipulate 
$$\left[ \text{Plot} \left[ e^{-\frac{t}{\text{CC} \, rR}}, \left\{ t, \, 0, \, 3 \right\}, \, \text{PlotRange} \rightarrow \text{All}, \, \text{GridLines} \rightarrow \text{All} \right], \, \left\{ rR, \, 0.2, \, 10 \right\}, \, \left\{ cC, \, 0.01, \, 1 \right\} \right]$$



A random scrap from a different perspective, kept as interesing junk.

$$\{ind, cap, res\} = \{li'[t] == v_1[t], v_c'[t] == 1/ci[t], ri[t] == v_r[t]\}; \\ kirchhoff = v_1[t] + v_c[t] + v_r[t] == v_s[t];$$

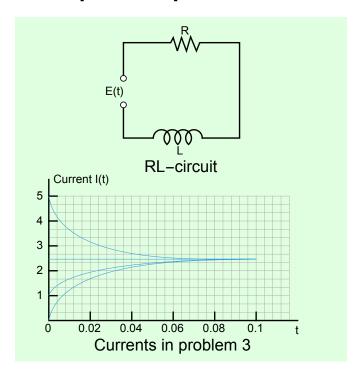
3. RL-Circuit. Model the RL-circuit in the figure below. Find a general solution when R, L, E are any constants. Graph or sketch solutions when L = 0.25 H,  $R = 10 \Omega$ , and E = 48 V.



The above screenshot came from the online app at https://falstad.com/circuit/. The current it shows agrees with the old formula for current, I=E/R, and was captured after the resistance had plenty of time to decay. And that's all it is, except that there is a time constant to apply. The time constant becomes ever smaller as the operation time increases. Since the

problem description talks in terms of a constant state, it seems the time constant would become vanishingly small, leaving merely I=E/R=4.8 amps.

## In[8]:= ClearAll["Global`\*"]



When there are a lot of variables to watch, the Manipulate command is the only way I know to get an overview. The box below is based on the material at https://www.electronics-tutorials.ws/inductor/lr-circuits.html and may not agree with the text in detail.

$$ln[9] = eye[vee\_, are\_, ell\_, tee\_] = \frac{vee}{are} \left(1 - e^{-\frac{are tee}{ell}}\right)$$

$$Out[9] = \frac{\left(1 - e^{-\frac{are tee}{ell}}\right) vee}{are}$$

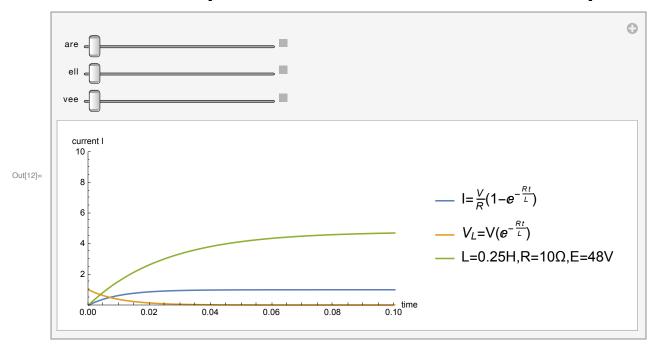
It takes some time for the current to reach its max value. From t=0.4 on in the green grid below, the circuit current is nominal.

 $log_{10} = Grid[Table[{tee, eye[48, 10, 0.25, tee]}, {tee, 0, 0.6, 0.1}], Frame <math>\rightarrow All]$ 

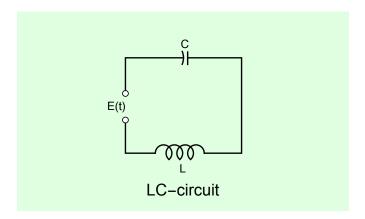
	0.	0.	
	0.1	4.71208	
	0.2	4.79839	
0]=	0.3	4.79997	
	0.4	4.8	
	0.5	4.8	
	0.6	4.8	

Out[1

```
\label{eq:local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_local_
```



5. LC-Circuit. This is an RLC-circuit with negligibly small R (analog of an undamped mass-spring system). Find the current when L=0.5~H,~C=0.005~F,~and~E=Sin[t]~V,~assuming~zero~initial~current~and~charge.



I ran across a couple of snippets, including one from the *Mathematica* documentation, suggesting that state space modeling would be a good way to look at circuits in Mathematica. I use it here.

Here I put in the given parameters, taking the opportunity to equate the resistance with zero.

ms = m1 /. {cC 
$$\rightarrow$$
 0.005, eL  $\rightarrow$  0.5, aR  $\rightarrow$  0}
$$\begin{pmatrix}
0 & 1 & 0 \\
-400. & 0. & 2. \\
\hline
0 & 1 & 0
\end{pmatrix}$$

The way to get output from a state space model is to use the command **OutputResponse**. Since the voltage depends on a periodic function, I drop the V for the input field, the voltage, because it is just a label.

```
outz = OutputResponse[{ms}, Sin[t], t]
\{(1.46082 \times 10^{-17} + 0.0526316 i)
   (0. + 0.0952381 i) \cos[20.t] - (0. + 1.i) \cos[19.t] \cos[20.t] +
      (0. + 0.904762 i) \cos[20.t] \cos[21.t] -
      (1.66533 \times 10^{-16} - 7.21645 \times 10^{-17} \text{ i}) \cos[20.t] \sin[19.t] -
       (2.24688 \times 10^{-17} + 6.60847 \times 10^{-19} i) \sin[20.t] +
       (2.35922 \times 10^{-16} + 6.93889 \times 10^{-18} i) \cos[19.t] \sin[20.t] -
       (2.13454 \times 10^{-16} + 6.27805 \times 10^{-18} i) \cos[21.t] \sin[20.t] +
       (5.96745 \times 10^{-17} - 1. i) Sin[19.t] Sin[20.t] +
       (1.50673 \times 10^{-16} - 6.52917 \times 10^{-17} \text{ i}) \cos[20. \text{ t}] \sin[21. \text{ t}] -
      (5.39912 \times 10^{-17} - 0.904762 i) Sin[20.t] Sin[21.t])
```

It is necessary to clean up the result with a small **Chop**.

```
outt = Chop[ComplexExpand[Re[outz]], 10<sup>-16</sup>] // FullSimplify
\{0.00501253 \cos[1.t] - 0.00501253 \cos[20.t] + 3.46945 \times 10^{-18} \cos[39.t]\}
```

Recognizing the periodic value of cosine, I can get the expression ready for a second chop by doing

```
outtf = outt /. \cos[39.t] \rightarrow 1
{3.46945 \times 10^{-18} + 0.00501253 \cos[1.t] - 0.00501253 \cos[20.t]}
```

And then the **Chop**.

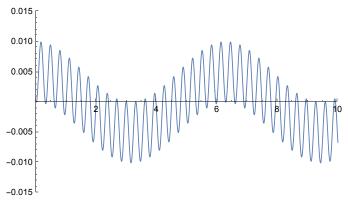
```
outtff = Chop \left[\%, 10^{-17}\right]
 {0.00501253 Cos[1.t] - 0.00501253 Cos[20.t]}
```

Testing the identity of those coefficients

```
1/0.005012531328320802
199.5
```

I find that the answer matches the text answer, justifying the green coloration above. The plot is interesting.

Plot[outtff,  $\{t, 0, 10\}$ , ImageSize  $\rightarrow$  350, AspectRatio  $\rightarrow$  0.6, PlotRange  $\rightarrow \{\{-0.01, 10\}, \{-0.015, 0.015\}\}, PlotStyle \rightarrow Thickness[0.003]]$ 



## 7 - 18 General RLC-circuits

7. Tuning. In tuning a sterio system to a radio station, we adjust the tuning control (turn a knob) that changes C (or perhaps L) in an RLC-circuit so that the amplitude of the steady-state current, numbered line (5), p. 95 becomes maximum. For what C will this happen?

It is where the particular solution of the homogeneous equation is maximized. Numbered line (5) looks like

$$I_p(t) = I_0 \sin[\omega t - \theta]$$

The quantity  $\theta$  is known as the phase lag, and, I suppose, the signal is best,  $I_p$  maximized, when  $\theta$  equals zero.

8 - 14 Find the steady-state current in the RLC-circuit in the figure below for the given data.

9. 
$$R = 4 \Omega$$
,  $L = 0.1 H$ ,  $C = 0.05 F$ ,  $E = 110 V$ 

$$LD[q[t], \{t, 2\}] + RD[q[t], t] - \frac{1}{c}q[t] = v[t]$$

eqn = 
$$0.1 q''[t] + 4 q'[t] - \frac{1}{0.05}q[t] == 110$$

$$-20. q[t] + 4 q'[t] + 0.1 q''[t] == 110$$

sol = DSolve[eqn, q, t]

$$\{ \{q \rightarrow Function[\{t\}, -5.5 + e^{-44.4949 t} C[1] + e^{4.4949 t} C[2]] \} \}$$

If C[1]=C[2]=0, then the green cell above matches the text answer.

11. 
$$R = 12 \Omega$$
,  $L = 0.4 H$ ,  $C = \frac{1}{80} F$ ,  $E = 220 Sin[10 t] V$ 

The state space method has been working where former methods I tried did not, so it makes sense to stick with it.

eqns = 
$$\{eLq''[t] + aRq'[t] + \frac{1}{cC}q[t] = Vee[t]\};$$

m1 = StateSpaceModel[eqns,

$$\{\{q[t], 0\}, \{q'[t], 0\}\}, \{\{Vee[t], 0\}\}, \{q'[t]\}, t]$$

$$\left(\begin{array}{c|cccc}
0 & 1 & 0 \\
-\frac{1}{cC \ eL} & -\frac{aR}{eL} & \frac{1}{eL} \\
\hline
0 & 1 & 0
\end{array}\right) S$$

Here I put in the given parameters.

ms = m1 /. 
$$\{cC \rightarrow \frac{1}{80}, eL \rightarrow 0.4, aR \rightarrow 12\}$$

$$\left(\begin{array}{c|cccc} 0 & 1 & 0 \\ -200. & -30. & 2.5 \\ \hline 0 & 1 & 0 \end{array}\right)^{S}$$

The way to get output from a state space model is to use the command **OutputResponse**.

outz = OutputResponse[{ms}, 220 Sin[10 t], t] 
$$\left\{ 0. + e^{-30.t} \left( 22. e^{10.t} - 27.5 e^{20.t} - 7.10543 \times 10^{-15} e^{20.t} Cos[10.t] + 5.5 e^{30.t} Cos[10.t] + 7.10543 \times 10^{-15} e^{20.t} Sin[10.t] + 16.5 e^{30.t} Sin[10.t] + 7.10543 \times 10^{-15} e^{40.t} Sin[10.t] \right) \right\}$$

It is necessary to clean up the result with a **Chop**.

outt = Chop[outz, 
$$10^{-14}$$
] // FullSimplify 
$$\{22. e^{-20.t} - 27.5 e^{-10.t} + 5.5 \cos[10.t] + 16.5 \sin[10.t] \}$$

I guess the e factors can be dropped if they are small enough, say, at 3 seconds.

$$N[-27.50000000000007^e^{-10.t}]/.t \rightarrow 3$$
  
 $-2.57335 \times 10^{-12}$ 

Evidently the text considers that size to be negligible, leaving

$$5.5 \cos[10.t] + 16.5 \sin[10.t]$$

as the answer. The plot looks routine.

13. R = 12, L = 1.2 H, C = 
$$\frac{20}{3}$$
\*10<sup>-3</sup> F, E = 12,000 Sin[25 t] V

$$C = \frac{20}{3} * \frac{1}{1000} = \frac{20}{3000} = \frac{2}{300}$$

ClearAll["Global`\*"]

eqns = 
$$\{eLq''[t] + aRq'[t] + \frac{1}{cC}q[t] = Vee[t]\};$$

$$\left(\begin{array}{ccc|c}
0 & 1 & 0 \\
-\frac{1}{cC eL} & -\frac{aR}{eL} & \frac{1}{eL} \\
\hline
0 & 1 & 0
\end{array}\right) S$$

Here I put in the given parameters.

ms = m1 /. 
$$\left\{ cC \rightarrow \frac{20}{3} * 10^{-3}, eL \rightarrow 1.2, aR \rightarrow 12 \right\}$$

$$\left( \begin{array}{c|cccc}
0 & 1 & 0 \\
-125. & -10. & 0.833333 \\
\hline
0 & 1 & 0
\end{array} \right)$$

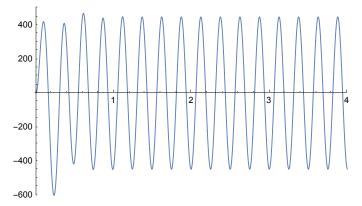
The way to get output from a state space model is to use the command **OutputResponse**.

```
outz = OutputResponse[{ms}, 12000 Sin[25t], t]
\{(0. + 0. i) - (400. + 1.56319 \times 10^{-14} i) e^{-5.t}\}
    (-1.+0.i) Cos[10.t] + (1.+0.i) e<sup>5.t</sup> Cos[10.t]<sup>2</sup> Cos[25.t] +
       (0.75 - 4.80505 \times 10^{-16} i) Sin[10.t] -
       (3.19744 \times 10^{-16} - 3.21521 \times 10^{-16} i) e^{5.t} Cos[10.t] Cos[25.t]
        Sin[10.t] + (1. + 2.45581 \times 10^{-16} i) e^{5.t} Cos[25.t] Sin[10.t]^{2} -
       (0.5 - 7.49623 \times 10^{-17} i) e^{5.t} Cos[10.t]^2 Sin[25.t] +
       (1.42109 \times 10^{-16} - 9.97247 \times 10^{-17} i) e<sup>5.t</sup> Cos[10.t] Sin[10.t]
        Sin[25.t] - (0.5 - 1.39035 \times 10^{-16} i) e^{5.t} Sin[10.t]^{2} Sin[25.t])
It is necessary to clean up the result with a Chop.
outt = Chop[ComplexExpand[Re[outz]], 10<sup>-15</sup>] // Simplify
-300. e^{-5.t} Sin[10.t] +
  Cos[10.t] (400. e^{-5.t} + 1.27898 \times 10^{-13} Cos[25.t] Sin[10.t]) -
   2.84217 \times 10^{-14} \sin[20.t] \sin[25.t] +
   Cos[10.t]^2 (-400. Cos[25.t] + 200. Sin[25.t]) +
  Sin[10.t]^2 (-400.Cos[25.t] + 200.Sin[25.t])
There is a \sin^2 + \cos^2 trig identity in the above, but I'm going to have to pull it out by hand.
outhnd = -300.e^{-5.t} Sin[10.t] +
  Cos[10.t] (400.e^{-5.t}) + (-400.Cos[25.t] + 200.Sin[25.t])
400. e^{-5.t} Cos[10.t] - 400. Cos[25.t] - 300. e^{-5.t} Sin[10.t] + 200. Sin[25.t]
outhnd2 = Collect[outhnd, e<sup>-5.t</sup>]
Clear["Global`*"]
```

While I was pulling things out by hand, I pulled out a choppable term. The text constant B is equal to -300. The text constant A is equal to 1 in one position and 400 in another position. That makes my answer wrong, technically. I guess I should make it yellow, though I don't feel it is a just action to do so. I feel like it is correct.

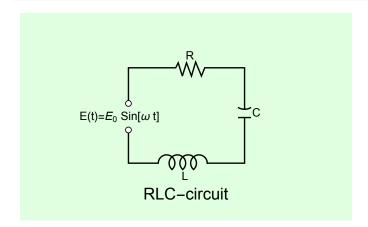
 $-400. \cos[25.t] + e^{-5.t} (400. \cos[10.t] - 300. \sin[10.t]) + 200. \sin[25.t]$ 

$$\begin{split} & \text{Plot} \Big[ -400. \text{`Cos}[25. \text{`t}] + \text{e}^{-5. \text{`t}} \text{ (400. \text{`Cos}[10. \text{`t}] - 300. \text{`Sin}[10. \text{`t}]) + } \\ & 200. \text{`Sin}[25. \text{`t}], \text{ {t, 0, 4}, ImageSize} \rightarrow 350, AspectRatio} \rightarrow 0.6, \\ & \text{PlotRange} \rightarrow \text{ { -0.01, 4}, {-600, 500}}, \text{PlotStyle} \rightarrow \text{Thickness}[0.003] \Big] \\ \end{aligned}$$



15. Cases of damping. What are the conditions for an RLC-circuit to be (I) overdamped, (II) critically damped, (III) underdamped? What is the critical resistance  $R_{\text{crit}}$  (the analog of the critical damping constant 2  $\sqrt{m k}$ ?

16 - 18 Solve the initial value problem for the RLC-circuit shown below, with the given data, assuming zero initial current and charge. Graph or sketch the solution.



17. 
$$R = 6 \Omega$$
,  $L = 1 H$ ,  $C = 0.04 F$ ,  $E = 600(Cos[t] + 4 Sin[t])V$ 

ClearAll["Global`\*"]

eqns = 
$$\left\{eL\ q''[t] + aR\ q'[t] + \frac{1}{cC}\ q[t] = Vee[t]\right\};$$

Here I put in the given parameters.

ms = m1 /. {cC 
$$\rightarrow$$
 0.04, eL  $\rightarrow$  1, aR  $\rightarrow$  6}  

$$\begin{pmatrix}
0 & 1 & | & 0 \\
-25. & -6 & | & 1 \\
\hline
0 & 1 & | & 0
\end{pmatrix}$$

The way to get output from a state space model is to use the command **OutputResponse**.

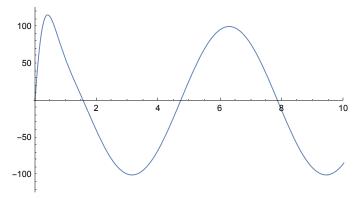
```
outz = OutputResponse[{ms}, 600 (Cos[t] + 4 Sin[t]), t]
\{(0. + 0. i) +
  e^{-3.t} ((-100. -1.11022×10<sup>-14</sup> i) Cos[4.t] + (100. +1.11022×10<sup>-14</sup> i)
        e^{3.t} Cos[t] Cos[4.t]<sup>2</sup> - (1.87214×10<sup>-14</sup> - 1.65445×10<sup>-14</sup> i)
        e^{3.t} \cos[4.t]^2 \sin[t] + (75. + 1.52656 \times 10^{-14} i) \sin[4.t] -
        (8.65974 \times 10^{-15} + 1.80411 \times 10^{-14} i) e^{3.t} Cos[t] Cos[4.t] Sin[4.t] +
        (2.27374 \times 10^{-13} + 2.91161 \times 10^{-14} i) e^{3.t} Cos[4.t] Sin[t] Sin[4.t] +
        (100. - 1.14492 \times 10^{-14} i) e^{3.t} Cos[t] Sin[4.t]^2 -
        (0. + 7.91555 \times 10^{-14} i) e^{3.t} Sin[t] Sin[4.t]^{2})
outt = Chop[ComplexExpand[Re[outz]]] // Simplify
\{-100. e^{-3.t} \cos[4.t] + 100. \cos[t] \cos[4.t]^2 +
   Sin[4.t] (75. e^{-3.t} + 100. Cos[t] Sin[4.t])
outtf = Collect[outt, e<sup>-3.t</sup>]
\{100. \cos[t] \cos[4.t]^2 + 100. \cos[t] \sin[4.t]^2 +
   e^{-3.t} (-100. Cos[4.t] + 75. Sin[4.t])
```

I can see the  $\sin^2 + \cos^2$  identity in the above, but will have to take it out by hand.

```
100. \cos[t] + e^{-3.t} (-100. \cos[4.t] + 75. \sin[4.t])
```

And with that, the above cell matches the text answer.

Plot[100. Cos[t] + 
$$e^{-3.t}$$
 (-100. Cos[4.t] + 75. Sin[4.t]),  
{t, 0, 10}, ImageSize  $\rightarrow$  350, AspectRatio  $\rightarrow$  0.6,  
PlotRange  $\rightarrow$  {{-0.01, 10}, {-125, 125}}, PlotStyle  $\rightarrow$  Thickness[0.003]



19. Writing report. Mechanical-electrical analogy. Explain table 2.2 (reproduced below) in a 1 - 2 page report with examples, e.g. the analog (with L = 1 H) of a mass-spring system of mass 5 kg, damping constant 10 kg/sec, spring constant 60 kg/sec<sup>2</sup>, and driving force 220 cos 10t kg/sec.

Electrical System	Mechanical System
Inductance L	Mass m
Reciprocal $\frac{1}{c}$ of capacitance	Spring modulus k
Derivative $\mathbf{E}_0\omega$ $Cos[\omega]$	Driving force $F_0Cos[\omega t]$
t] of electromotive force	
Current I(t)	Displacement y(t)

Out[30]=