

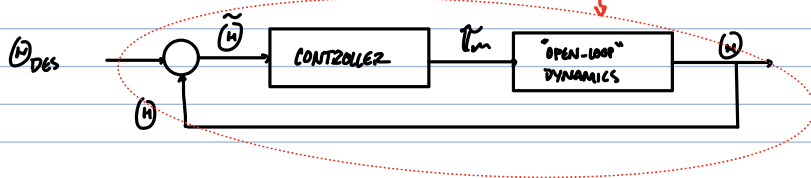
06 CLASSICAL CONTROL REVIEW

FEEDBACK CONTROL

WE WANT OUR DYNAMICS IN THESE TERMS

$$\tau = H(\theta)\ddot{\theta} + V(\theta, \dot{\theta}) + G(\theta) + F(\dot{\theta})$$

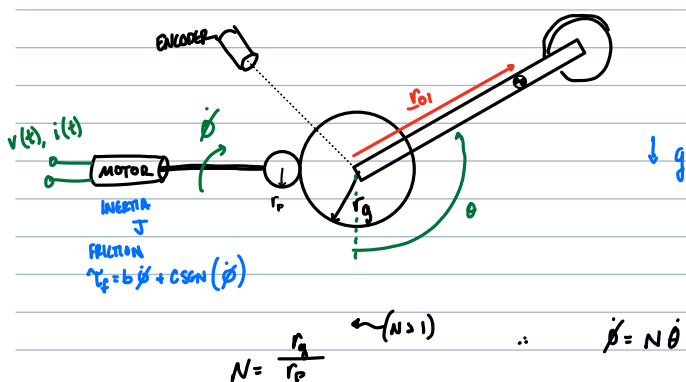
CONTROL = MODIFYING DYNAMICS IN A DESIRED FASHION



EQUAL TO



1 DOF EXAMPLE (SEVOZ MODULE)



DYNAMICS FOR 1-DOF

$$\tau_{\text{ARM}} = I_{\text{LINK}} \ddot{\theta} + \underline{r}_{01} \times m_{\text{LINK}} g$$

$$\tau_m = J \ddot{\theta} + b \dot{\theta} + c \text{sgn}(\dot{\theta}) + \frac{1}{N} \tau_{\text{ARM}}$$

OR IN TERMS OF θ

$$N \tau_m = N J (N \ddot{\theta}) + N b N \dot{\theta} + N c \text{sgn}(N \dot{\theta}) + I \ddot{\theta} + m g r_{01} \sin \theta$$

$$N \tau_m = \underbrace{(I + N^2 J)}_{\text{INERTIA}} \ddot{\theta} + \underbrace{m g r_{01} \sin \theta}_{\text{GRAVITY}} + \underbrace{b N^2 \dot{\theta} + c N \text{sgn}(\dot{\theta})}_{\text{FRICTION}}$$

IF NO CONTROL $i(t)$, THEN $\tau_m = K_T i$

$$N K_t \ddot{\theta} = \underbrace{(I + N^2 J)}_{\substack{I_e = \text{EFFECTIVE} \\ \text{INERTIA}}} \ddot{\theta} + \underbrace{mg r_{o1}}_{\substack{b_e = \text{EFFECTIVE} \\ \text{DAMPING}}} S\theta + \underbrace{b N^2 \dot{\theta} + c N \text{sgn}(\dot{\theta})}_{\substack{b_e = \text{EFFECTIVE} \\ \text{DAMPING}}}$$

$A_m = \text{EFFECTIVE GAIN}$

OPEN LOOP DYNAMICS:

$$A_m \ddot{\theta} = I_e \ddot{\theta} + b_e \dot{\theta} + mg r_{o1} S\theta + c N \text{sgn}(\dot{\theta})$$

TO USE CLASSICAL CONTROL, WE NEED TO LINEARIZE DYNAMICS ABOUT SOME OPERATING POINT

$$\theta = 0 \quad \dot{\theta} = 0$$

TAKE A TAYLOR SERIES:

$$f(x) = f(x_0) + \left. \frac{\partial f}{\partial x} \right|_{x=x_0} (x-x_0) + \left. \frac{\partial^2 f}{\partial x^2} \right|_{x=x_0} \frac{(x-x_0)^2}{2!} + \dots$$

IF WE ASSUME DEVIATION FROM O.P.

IS SMALL, ERROR GET SMALLER + SMALLER

$$\text{IF } \|x - x_0\| \ll 1$$

$$f(x) \approx f(x_0) + \left. \frac{\partial f}{\partial x} \right|_{x=x_0} (x-x_0)$$

$$S\theta \approx \left. \sin(\theta) \right|_{\theta=0} + \left. \cos(\theta) \right|_{\theta=0} (\theta - \theta_0)$$

$$= \theta - \theta_0$$

↑ ASSUMED TO BE SMALL

$$S\theta \approx \theta$$

$$\therefore A_m \ddot{\theta} = I_e \ddot{\theta} + b_e \dot{\theta} + \underbrace{mg r_{o1}}_{\substack{\text{NOW} \\ \text{LINEAR} \ddot{\theta}}} \theta + \underbrace{c N \text{sgn}(\dot{\theta})}_{\substack{\text{CAN WE} \\ \text{LINEARIZE THIS?}}}$$

THE DERIVATIVE MUST EXIST AT OUR OPERATING POINT IF WE WANT TO USE A T.S.

• DISCONTINUITY AT $\dot{\theta} = 0$

$\frac{\partial f}{\partial x}$ DOESN'T EXIST \Rightarrow NO LINEARIZING ABOUT OUR OPERATING POINT

\therefore LET'S ASSUME WE CAN NEGLECT COULOMB FRICTION

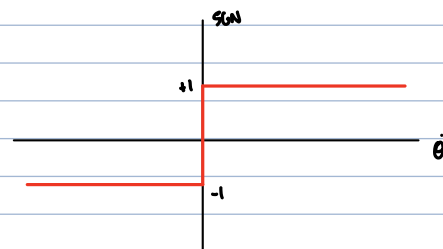
$$A_m \ddot{\theta} \approx I_e \ddot{\theta} + b_e \dot{\theta} + \underbrace{mg r_{o1}}_{\substack{K_e = \text{EFFECTIVE STIFFNESS}}}$$

OPEN-LOOP DYNAMICS FOR CONTROL

$$\Rightarrow \boxed{A_m \ddot{\theta} = I_e \ddot{\theta} + b_e \dot{\theta} + K_e \theta}$$

↑ LOOKS LIKE A LINEAR MASS-SPRING DAMPER SYSTEM

SGN FUNCTION



ANALYSIS

$$\mathcal{L}\{A_m i\} = \mathcal{L}\{I_e \ddot{\theta} + b_e \dot{\theta} + k_e \theta\}$$

$$A_m i(s) = I_e s^2 \theta(s) + b_e s \theta(s) + k_e \theta(s)$$

$$A_m i(s) = [I_e s^2 + b_e s + k_e] \theta(s)$$

SOLVE FOR $\theta(s)$

$$\frac{\theta(s)}{i(s)} = \frac{I_e s^2 + b_e s + k_e}{A_m} = \frac{A_m/I_e}{s^2 + \left(\frac{b_e}{I_e}\right)s + \left(\frac{k_e}{I_e}\right)}$$

OPEN-LOOP TRANSFER FUNCTION

$$= \frac{A_m/I_e}{s^2 + 2\zeta\omega_n s + \omega_n^2} = \frac{A_m/I_e}{(s+p_1)(s+p_2)}$$

ζ = DAMPING RATIO

ω_n = NATURAL FREQUENCY

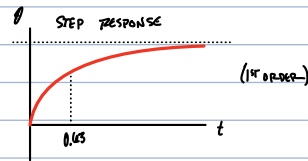
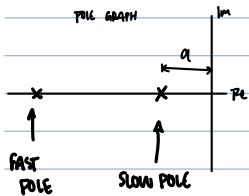
p_1, p_2 = POLES/ROOTS OF OPEN LOOP SYSTEM

POSSIBLE BEHAVIORS

OVERDAMPED

$$\zeta > 1$$

REAL POLES



$$\begin{aligned} \tau_c &= \text{TIME CONSTANT} \\ &= \text{TIME TO 63\% OF SS VALUE} \\ &= \frac{1}{a} \end{aligned}$$

$$\frac{1}{a} = \text{TIME CONSTANT}$$

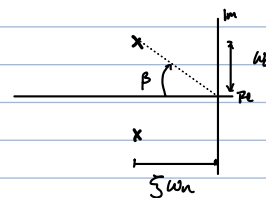
WE NEGLECT FAST POLE IF

5 X LARGER THAN SLOW POLE

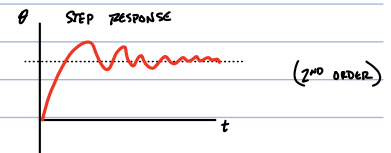
UNDERDAMPED

$$\zeta < 1$$

COMPLEX CONJUGATE POLES



$$\begin{aligned} \cos \beta &= \zeta \\ \therefore 0.5 &= \zeta \end{aligned}$$

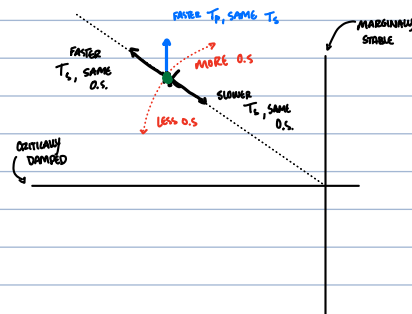


$$\begin{aligned} T_p &= \text{TIME TO PEAK} \\ &= \frac{\pi}{\omega_d} \\ T_s &= \text{TIME TO SETTLE (98\% SS)} \\ &= \frac{4}{\zeta\omega_n} \end{aligned}$$

$$\zeta\omega_n = \text{TIME TO SS.}$$

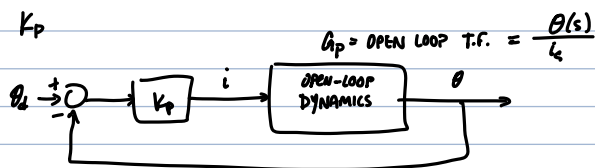
ω_d = DAMPED NATURAL FREQUENCY

MODIFYING POLES



CONTROL TECHNIQUES

PROPORTIONAL CONTROL



TO GET CLOSED LOOP TRANSFER FUNCTION... $\frac{\theta(s)}{\theta_d(s)}$

$$\theta = G_p i$$

$$i = K_p (\theta_d - \theta)$$

$$\therefore \theta = G_p K_p (\theta_d - \theta)$$

$$\theta (1 + K_p G_p) = K_p G_p \theta_d$$

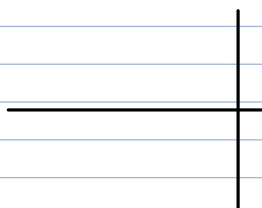
$$\Rightarrow \boxed{\frac{\theta}{\theta_d} = \frac{K_p G_p}{1 + K_p G_p}} \quad \text{FEEDBACK FORMULA}$$

$$\frac{\theta(s)}{\theta_d(s)} = K_p \frac{G_p}{1 + K_p G_p} = \frac{K_p A_m}{I_0 s^2 + b_0 s + (k_0 + K_p A_m)}$$

STIFFNESS WE
CAN MODIFY
(CHANGING NATURAL FREQUENCY)

ROOT LOCUS

GRAPH OF ALL CLOSED LOOP POLE LOCATIONS



AS WE INCREASE K_p , $\omega_d \uparrow$, T_c STAYS THE SAME

FOR A STEP INPUT IN θ_d

$$\theta \Big|_{t \rightarrow \infty} = \lim_{s \rightarrow 0} s \theta(s) = \theta_d(s) \frac{\theta(s)}{\theta_d(s)} = \lim_{s \rightarrow 0} \frac{1}{s} \cdot \overset{\text{STEP}}{\frac{\theta(s)}{\theta_d(s)}}$$

$$= \lim_{s \rightarrow 0} \frac{\theta(s)}{\theta_d(s)} = \frac{K_p A_m}{k_0 + K_p A_m}$$

$$\therefore e = \frac{\theta_d - \theta}{\theta_d} =$$

WE WANT TO INCREASE S.S. ERROR

IF YOU HAVE ZERO ERROR, CURRENT WILL

NO LONGER FLOW \Rightarrow NEVER CORRECT