



Application Note: SY8113B

High Efficiency, 500kHz, 3A, 18V Input Synchronous Step Down Regulator

General Description

The SY8113B is a high efficiency 500 kHz synchronous step-down DC-DC converter capable of delivering 3A current. The SY8113B operates over a wide input voltage range from 4.5V to 18V and integrates main switch and synchronous switch with very low $R_{DS(ON)}$ to minimize the conduction loss.

Low output voltage ripple and small external inductor and capacitor sizes are achieved with 500 kHz switching frequency. It adopts the instant PWM architecture to achieve fast transient responses for high step down applications

Ordering Information

SY8113

Temperature Code
Package Code
Optional Spec Code

Ordering Number	Package type	Note
SY8113BADC	TSOT23-6	--

Features

- low $R_{DS(ON)}$ for internal switches (top/bottom): 80mΩ/40mΩ
- 4.5-18V input voltage range
- 3A output current capability
- 500 kHz switching frequency
- Instant PWM architecture to achieve fast transient responses.
- Cycle-by-cycle peak current limitation
- Internal softstart limits the inrush current
- Hic-cup mode output short circuit protection
- $\pm 1.5\%$ 0.6V reference
- TSOT23-6 package

Applications

- Set Top Box
- Portable TV
- Access Point Router
- DSL Modem
- LCD TV

Typical Applications

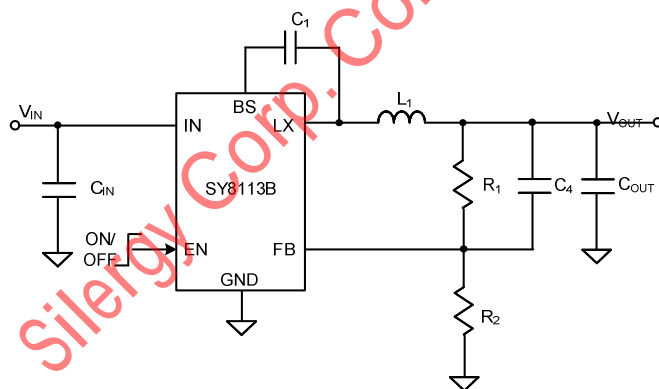


Figure 1. Schematic Diagram

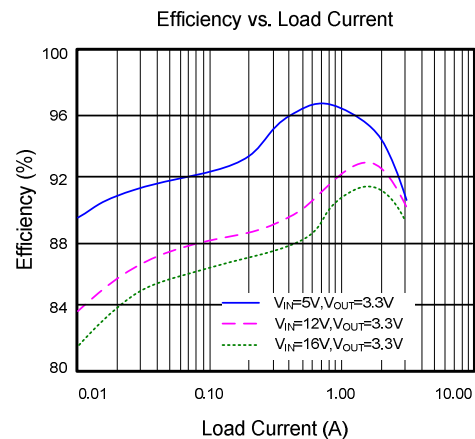
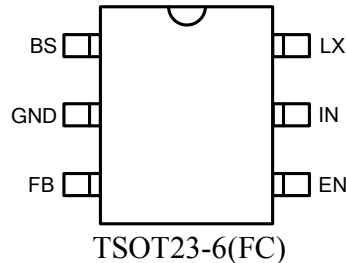


Figure 2. Efficiency Figure

Pinout (top view)



Top Mark: WCxyz, (Device code: WC, *x=year code*, *y=week code*, *z=lot number code*)

Pin Name	Pin Number	Pin Description
BS	1	Boot-Strap Pin. Supply high side gate driver. Decouple this pin to LX pin with 0.1uF ceramic cap.
GND	2	Ground pin
FB	3	Output Feedback Pin. Connect this pin to the center point of the output resistor divider (as shown in Figure 1) to program the output voltage: $V_{out}=0.6*(1+R1/R2)$
EN	4	Enable control. Pull high to turn on. Do not float.
IN	5	Input pin. Decouple this pin to GND pin with at least 1uF ceramic cap
LX	6	Inductor pin. Connect this pin to the switching node of inductor

Absolute Maximum Ratings (Note 1)

Supply Input Voltage	19V
Enable Voltage	$V_{IN} + 0.3V$
FB Voltage	4V
Power Dissipation, PD @ TA = 25°C TSOT23-6 (FC)	1.5W
Package Thermal Resistance (Note 2)	
θ_{JA}	66°C /W
θ_{JC}	15°C /W
Junction Temperature Range	150°C
Lead Temperature (Soldering, 10 sec.)	260°C
Storage Temperature Range	-65°C to 150°C

Recommended Operating Conditions (Note 3)

Supply Input Voltage	4.5V to 18V
Junction Temperature Range	-40°C to 125°C
Ambient Temperature Range	-40°C to 85°C



Electrical Characteristics

($V_{IN} = 12V$, $V_{OUT} = 1.2V$, $L = 2.2\mu H$, $C_{OUT} = 47\mu F$, $T_A = 25^\circ C$, $I_{OUT} = 1A$ unless otherwise specified)

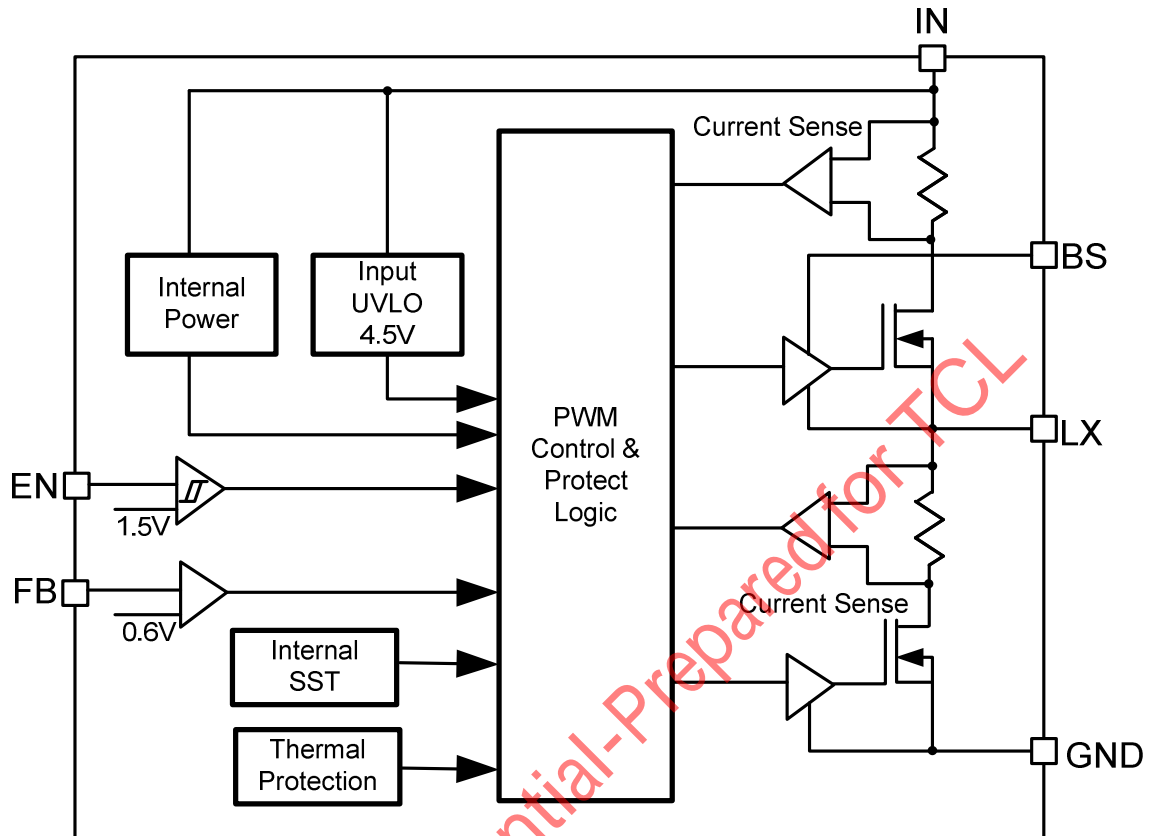
Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Input Voltage Range	V_{IN}		4.5		18	V
Quiescent Current	I_Q	$I_{OUT}=0$, $V_{FB}=V_{REF}*105\%$		100		μA
Shutdown Current	I_{SHDN}	$EN=0$		5	10	μA
Feedback Reference Voltage	V_{REF}		0.591	0.6	0.609	V
FB Input Current	I_{FB}	$V_{FB}=3.3V$	-50		50	nA
Top FET RON	$R_{DS(ON)1}$			80	90	m Ω
Bottom FET RON	$R_{DS(ON)2}$			40	50	m Ω
Top FET Peak Current Limit	$I_{LIM, TOP}$		4.8	5.4	6	A
Bottom FET Valley Current Limit	$I_{LIM, BOTTOM}$		3	3.6	4.25	A
EN Rising Threshold	V_{ENH}		1.5			V
EN Falling Threshold	V_{ENL}				0.4	V
Input UVLO Threshold	V_{UVLO}		3.8	4.2	4.5	V
UVLO Hysteresis	V_{HYS}			0.3		V
Min ON Time			50	80	120	ns
Min OFF Time			140	170	220	ns
Switching Frequency				500		kHz
Soft-start Time	t_{SS}			800		μS
Thermal Shutdown Temperature	T_{SD}			150		$^\circ C$
Thermal Shutdown Hysteresis	T_{HYS}			15		$^\circ C$

Note 1: Stresses beyond the “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Note 2: θ_{JA} is measured in the natural convection at $T_A = 25^\circ C$ on a two-layer Silergy Evaluation Board.

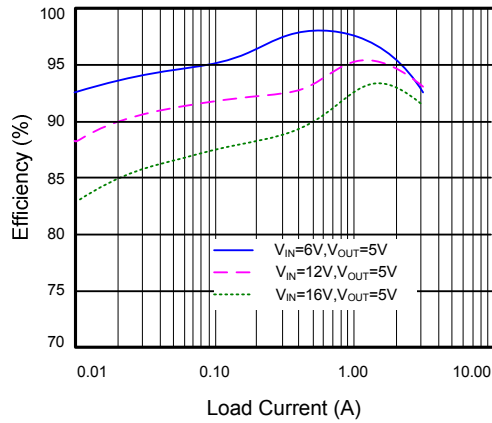
Note 3: The device is not guaranteed to function outside its operating conditions

Block Diagram

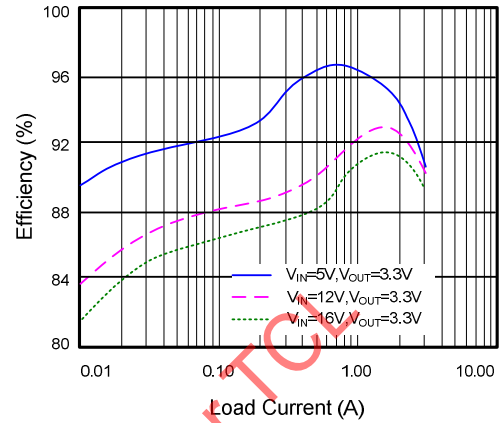


Typical Performance Characteristics

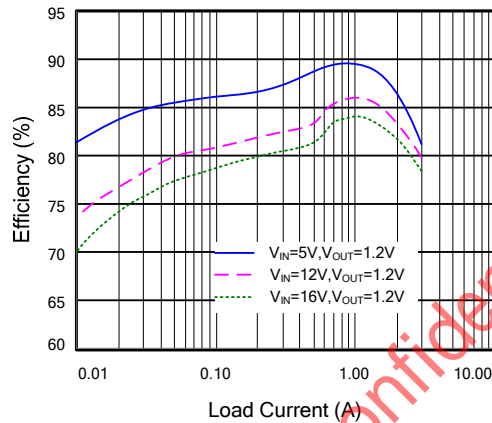
Efficiency vs. Load Current



Efficiency vs. Load Current

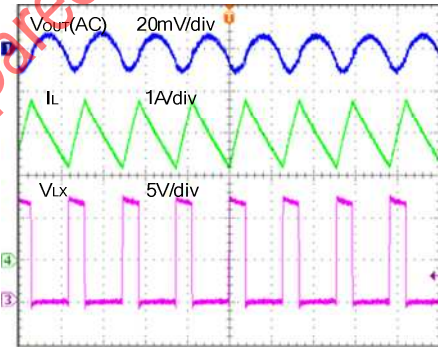


Efficiency vs. Load Current



Output Ripple

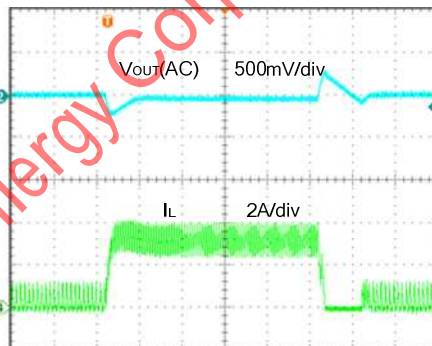
($V_{IN}=12V, V_{OUT}=3.3V, I_{LOAD}=3A$)



Time (2μs/div)

Load Transient

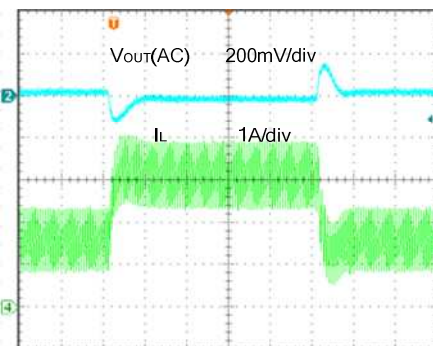
($V_{IN}=12V, V_{OUT}=3.3V, I_{LOAD}=0.3-3A$)



Time (40μs/div)

Load Transient

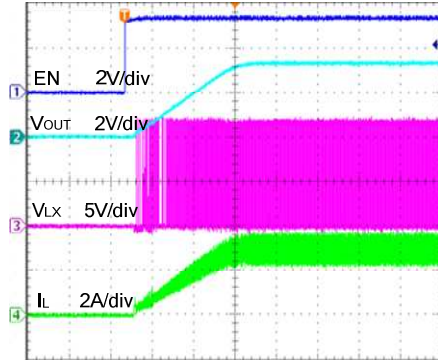
($V_{IN}=12V, V_{OUT}=3.3V, I_{LOAD}=1.5-3A$)



Time (40μs/div)

Startup

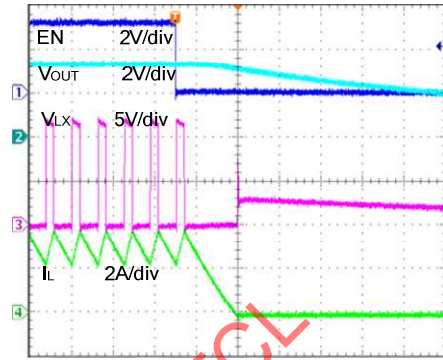
($V_{IN}=12V$, $V_{OUT}=3.3V$, $I_{LOAD}=3A$)



Time (400μs/div)

Shutdown

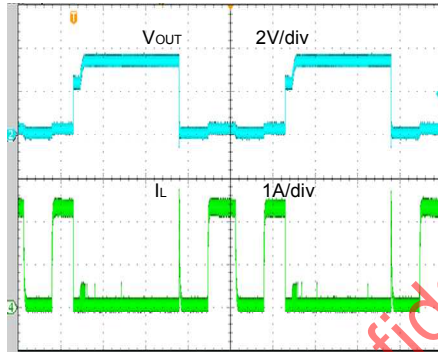
($V_{IN}=12V$, $V_{OUT}=3.3V$, $I_{LOAD}=3A$)



Time (4μs/div)

Short Circuit Protection

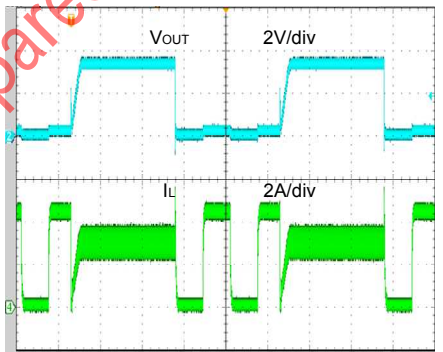
($V_{IN}=12V$, $V_{OUT}=3.3V$, Open to Short)



Time (4ms/div)

Short Circuit Protection

($V_{IN}=12V$, $V_{OUT}=3.3V$, 3A to Short)



Time (4ms/div)

Operation

SY8113B is a synchronous buck regulator IC that integrates the PWM control, top and bottom switches on the same die to minimize the switching transition loss and conduction loss. With ultra low $R_{DS(ON)}$ power switches and proprietary PWM control, this regulator IC can achieve the highest efficiency and the highest switch frequency simultaneously to minimize the external inductor and capacitor size, and thus achieving the minimum solution footprint.

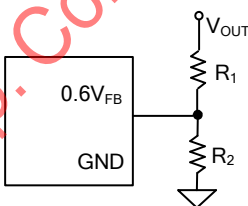
SY8113B provides protection functions such as cycle by cycle current limiting and thermal shutdown protection. SY8113B will sense the output voltage conditions for the fault protection.

Applications Information

Because of the high integration in the SY8113B IC, the application circuit based on this regulator IC is rather simple. Only input capacitor C_{IN} , output capacitor C_{OUT} , output inductor L and feedback resistors (R_1 and R_2) need to be selected for the targeted applications specifications.

Feedback resistor dividers R_1 and R_2 :

Choose R_1 and R_2 to program the proper output voltage. To minimize the power consumption under light loads, it is desirable to choose large resistance values for both R_1 and R_2 . A value of between $10k\Omega$ and $1M\Omega$ is highly recommended for both resistors. If V_{OUT} is 3.3V, $R_1=100k$ is chosen, then using following equation, R_2 can be calculated to be 22.1k:

$$R_2 = \frac{0.6V}{V_{OUT} - 0.6V} R_1$$


Input capacitor C_{IN} :

The ripple current through input capacitor is calculated as:

$$I_{CIN_RMS} = I_{OUT} \cdot \sqrt{D(1-D)}$$

To minimize the potential noise problem, place a typical X5R or better grade ceramic capacitor really close to the IN and GND pins. Care should be taken to minimize the loop area formed by C_{IN} , and IN/GND

pins. In this case, a 10uF low ESR ceramic capacitor is recommended.

Output capacitor C_{OUT} :

The output capacitor is selected to handle the output ripple noise requirements. Both steady state ripple and transient requirements must be taken into consideration when selecting this capacitor. For the best performance, it is recommended to use X5R or better grade ceramic capacitor greater than 22uF capacitance.

Output inductor L :

There are several considerations in choosing this inductor.

- 1) Choose the inductance to provide the desired ripple current. It is suggested to choose the ripple current to be about 40% of the maximum output current. The inductance is calculated as:

$$L = \frac{V_{OUT}(1 - V_{OUT}/V_{IN,MAX})}{F_{SW} \times I_{OUT,MAX} \times 40\%}$$

where F_{sw} is the switching frequency and $I_{OUT,MAX}$ is the maximum load current.

The SY8113B regulator IC is quite tolerant of different ripple current amplitude. Consequently, the final choice of inductance can be slightly off the calculation value without significantly impacting the performance.

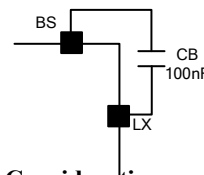
- 2) The saturation current rating of the inductor must be selected to be greater than the peak inductor current under full load conditions.

$$I_{SAT, MIN} > I_{OUT, MAX} + \frac{V_{OUT}(1 - V_{OUT}/V_{IN,MAX})}{2 \cdot F_{SW} \cdot L}$$

- 3) The DCR of the inductor and the core loss at the switching frequency must be low enough to achieve the desired efficiency requirement. It is desirable to choose an inductor with $DCR < 50m\Omega$ to achieve a good overall efficiency.

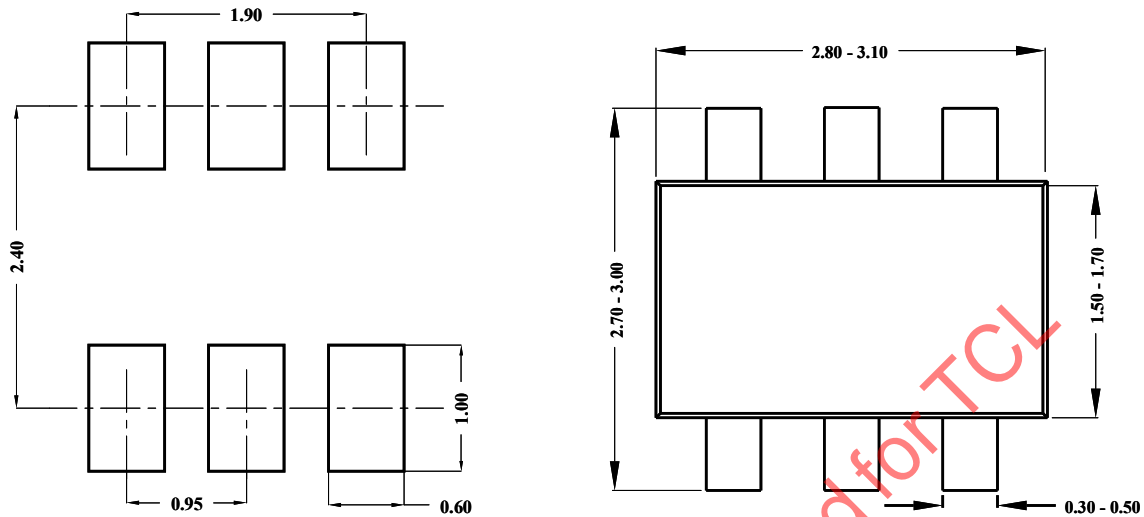
External Bootstrap Cap

This capacitor provides the gate driver voltage for internal high side MOSEFET. A 100nF low ESR ceramic capacitor connected between BS pin and LX pin is recommended.

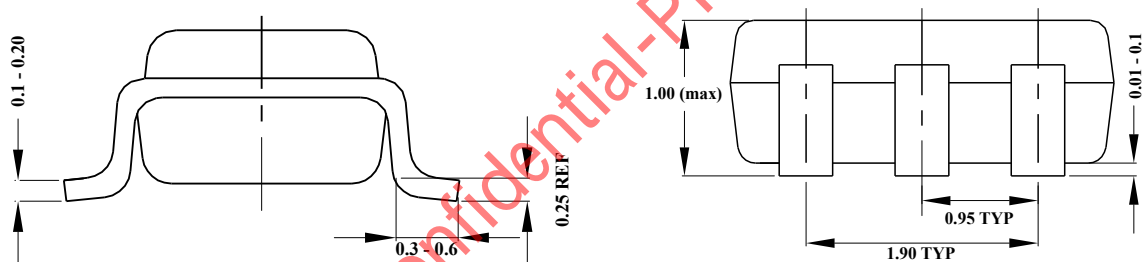


Load Transient Considerations:

TSOT23-6L (FC) Package outline & PCB layout



Recommended Pad Layout

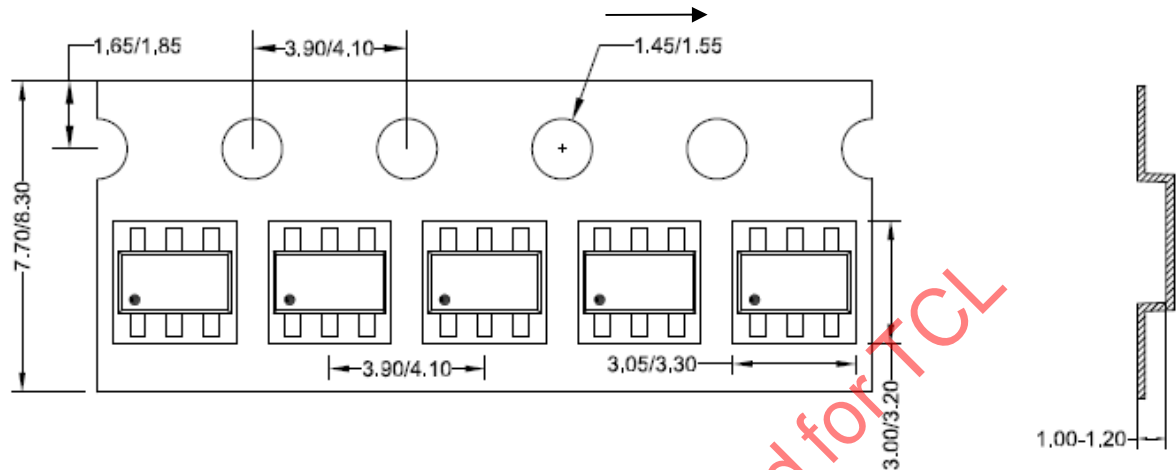


Notes: All dimension in MM
 All dimension don't not include mold flash & metal burr

Taping & Reel Specification

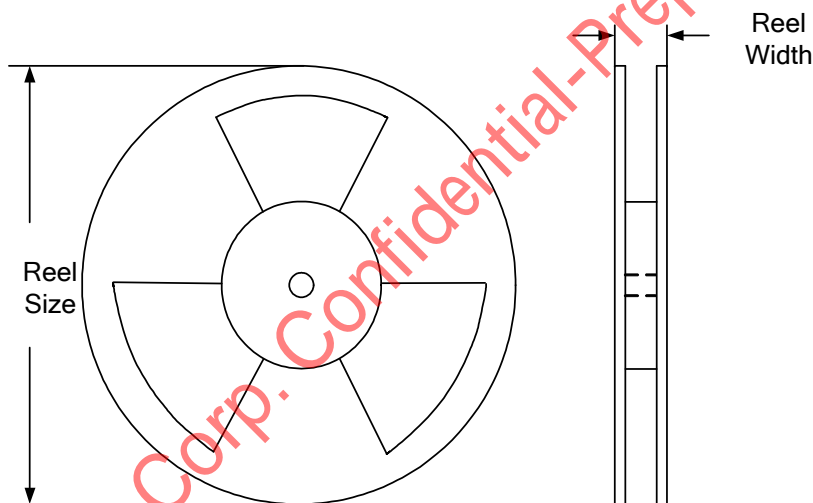
1. Taping orientation

TSOT23-6



Feeding direction

2. Carrier Tape & Reel specification for packages



Package types	Tape width (mm)	Pocket pitch(mm)	Reel size (Inch)	Reel width(mm)	Trailer length(mm)	Leader length (mm)	Qty per reel
TSOT23-6	8	4	7	8.4	400	160	3000

3. Others: NA