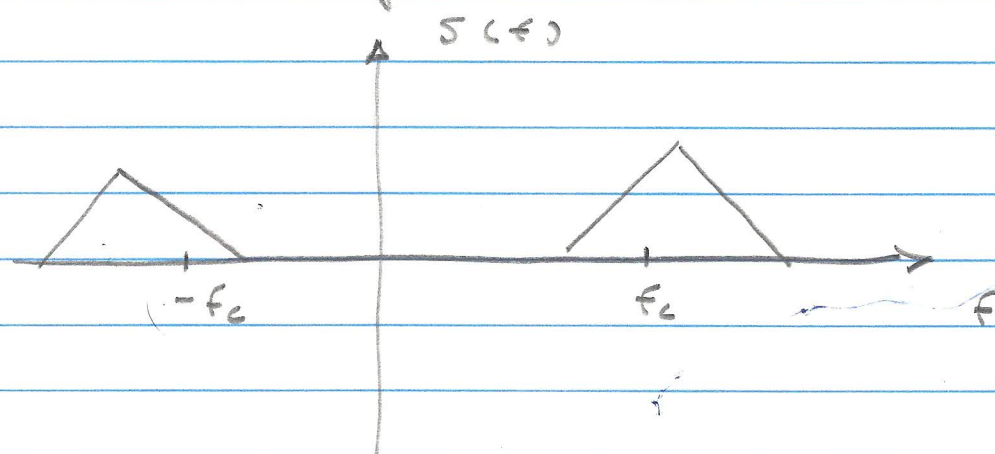


5

Bandpass signals.

- There are 3 ways to characterize BP signals
- Most digitally modulated signals are BP
- Let $u(\cdot)$ be the step function
- Look at a general situation:



f_c does not need to be in the middle!

Pre-envelope.

$$S_+(f) \triangleq 2 \cdot u(f) \cdot S(f), \text{ then}$$

$$s_+(t) = s(t) + j \cdot \hat{s}(t) \quad (\text{why?})$$

$$\hat{s}(t) = \frac{1}{\pi t} * s(t) \quad (\text{Hilbert transform})$$

$$\hat{S}(f) = -j \operatorname{sgn}(f) \cdot S(f)$$

Then, $s(t) = \text{Re} \{ s_+(t) \}$,

however

$$s_+(t) = s_e(t) \cdot \exp(+j2\pi f_c t)$$

Thus,

$$s(t) = \text{Re} \{ s_e(t) \exp(+j2\pi f_c t) \}$$

Also, $s_e(t) = s_I(t) + j s_Q(t)$, thus

$$s(t) = s_I(t) \cdot \cos(2\pi f_c t) - s_Q(t) \cdot \sin(2\pi f_c t),$$

and

$$s(t) = r(t) \cos(2\pi f_c t + \Theta(t)),$$

with

$$r(t) = \{ s_I^2(t) + s_Q^2(t) \}^{1/2}$$

and

$$\Theta(t) = \tan^{-1} \left(\frac{s_Q(t)}{s_I(t)} \right)$$

Note :

$$\begin{aligned}
 S(f) &= \mathcal{F}\{s(t)\} = \\
 &= \int_{-\infty}^{+\infty} s(t) \cdot \exp(-j2\pi ft) dt = \\
 &= \int_{-\infty}^{+\infty} \operatorname{Re}\{s_e(t) \exp(+j2\pi f_c t)\} \exp(-j2\pi ft) dt \\
 &= \frac{1}{2} \int_{-\infty}^{+\infty} s_e(t) \exp(-j2\pi (f-f_c)t) dt \\
 &\quad + \frac{1}{2} \int_{-\infty}^{+\infty} s_e^*(t) \exp(-j2\pi (f+f_c)t) dt = \\
 &= \frac{1}{2} [S_e(f-f_c) + S_e^*(-f-f_c)]
 \end{aligned}$$

Signal energy:

$$\begin{aligned}
 \mathcal{E}_S &= \int_{-\infty}^{+\infty} |s(t)|^2 dt = \int_{-\infty}^{+\infty} |S(f)|^2 df = \\
 &= \frac{1}{4} \cdot \int_{-\infty}^{+\infty} |S_e(f-f_c)|^2 df + \frac{1}{4} \int_{-\infty}^{+\infty} |S_e(-f-f_c)|^2 df \\
 &= \frac{1}{2} \int_{-\infty}^{+\infty} |S_e(f)|^2 df = \frac{1}{2} \mathcal{E}_{S_e}
 \end{aligned}$$

LTI BP Systems

$$h(t) = 2 \cdot \text{Re} \{ h_e(t) \exp(+j2\pi f_c t) \}$$

↗

$$x(t) = \text{Re} \{ x_e(t) \exp(+j2\pi f_c t) \}$$

$$y(t) = \int_{-\infty}^{+\infty} h(u) x(t-u) du =$$

$$= \int_{-\infty}^{+\infty} 2 \cdot \frac{1}{2} \{ h_e(u) \exp(+j2\pi f_c u) + h_e^*(u) \exp(-j2\pi f_c u) \} \cdot$$

$$\frac{1}{2} \{ x_e(t-u) \exp(+j2\pi f_c (t-u)) + x_e^*(t-u) \exp(-j2\pi f_c (t-u)) \} du =$$

$$= \frac{1}{2} \cdot \left(\int_{-\infty}^{+\infty} h_e(u) \cdot x_e(t-u) du \right) \cdot \exp(+j2\pi f_c t)$$

$$+ \frac{1}{2} \cdot \left(\int_{-\infty}^{+\infty} h_e^*(u) \cdot x_e^*(t-u) du \right) \cdot \exp(-j2\pi f_c t) =$$

$$= \text{Re} \left\{ \underbrace{\int_{-\infty}^{+\infty} h_e(u) x_e(t-u) du}_{\text{}} \exp(+j2\pi f_c t) \right\}$$

$$\text{Thus, } y_e(t) = \int_{-\infty}^{+\infty} h_e(u) \cdot x_e(t-u) du$$

↗ Note the way: $\mathcal{L} y_e = 2 \mathcal{L} y$ ↖

Also,

$$\begin{aligned} Y(f) &= \frac{1}{2} \left\{ Y_L(f-f_0) + Y_L^*(-f-f_0) \right\} = \\ &= \frac{1}{2} \left\{ H_L(f-f_0) \cdot X_L(f-f_0) + \right. \\ &\quad \left. + H_L^*(-f-f_0) \cdot X_L^*(-f-f_0) \right\}. \end{aligned}$$

Band-pass stationary stochastic process:

$$n(t) = \operatorname{Re} \{ n_e(t) \cdot \exp(+j 2\pi f_c t) \}$$

⇒ For $n(t)$ to be WSS, we need

$$n_e(t) = n_I(t) + j n_Q(t), \quad \text{to have:}$$

a) $n_I(\cdot)$, $n_Q(\cdot)$ are WSS, jointly

$$b) R_{n_I}(\tau) = R_{n_Q}(\tau)$$

$$c) R_{n_I n_Q}(\tau) = -R_{n_Q n_I}(\tau)$$

Then,

$$R_n(\tau) = R_{n_I}(\tau) \cos(2\pi f_c \tau) - R_{n_Q n_I}(\tau) \sin(\cdot)$$

Thus,

$$R_n(\tau) = \text{Re} \{ R_{ne}(\tau) \cdot \exp(+j2\pi f_c \tau) \}$$

with $R_{ne}(\tau) = R_{nI}(\tau) + j R_{nQI}(\tau)$

Now, consider:

$$z(t) = n_I(t) + j n_Q(t)$$

$$R_z(\tau) = 2 \cdot (R_{nI}(\tau) + j R_{nQI}(\tau))$$

More properties of BPS R.P.R.

$$\begin{aligned} 1. \quad S_n(f) &= \mathcal{F} \{ R_z(\tau) \exp(+j2\pi f_c \tau) \} = \\ &= \frac{1}{2} \{ S_z(f - f_c) + S_z^*(-f - f_c) \} \end{aligned}$$

$$\begin{aligned} 2. \quad R_z(\tau) &= \mathbb{E} \{ z(t+\tau) \cdot z^*(t) \} = \\ &= R_z^*(-\tau) \Rightarrow R_z^*(\tau) = \underline{R_z(-\tau)} \end{aligned}$$

$$\Rightarrow S_z^*(-f) = S_z(-f) \Leftrightarrow S_z \text{ is } \underline{\text{real}}.$$

Thus, $S_n(f) = \frac{1}{2} \{ S_z(f - f_c) + S_z(-f - f_c) \}$

$$3. \quad R_{nQI}(\tau) = -R_{nIQ}(\tau) \Rightarrow$$

$\Rightarrow n_I(t), n_Q(t)$ are uncorrelated

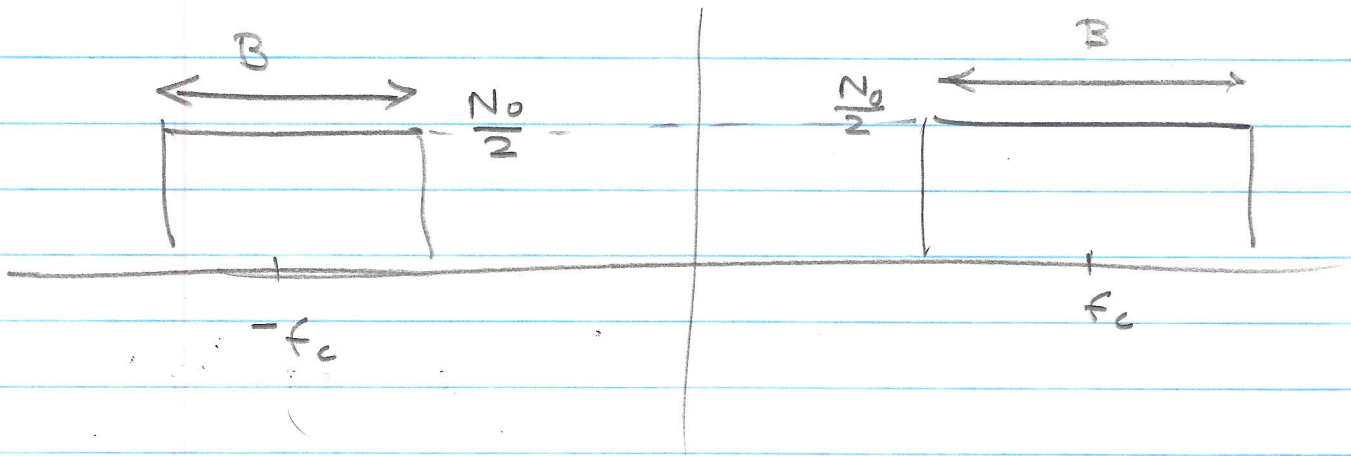
• White noise

Idealized concept

$$S_N(f) = \frac{N_0}{2} \quad \text{W. Hz}^{-1}$$

for all f .

BP white noise



$$R_z(\tau) = \begin{cases} 2N_0 & |f| < \frac{B}{2} \\ 0 & \text{else} \end{cases}$$

(no normalization by $1/2$)

Thus, $R_{n_{aI}}(\tau) = 0$ for all τ .