

# ECE 5210 Midterm 2

*Week of: March 28, 2024*

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You have 2 hours for 5 problems.

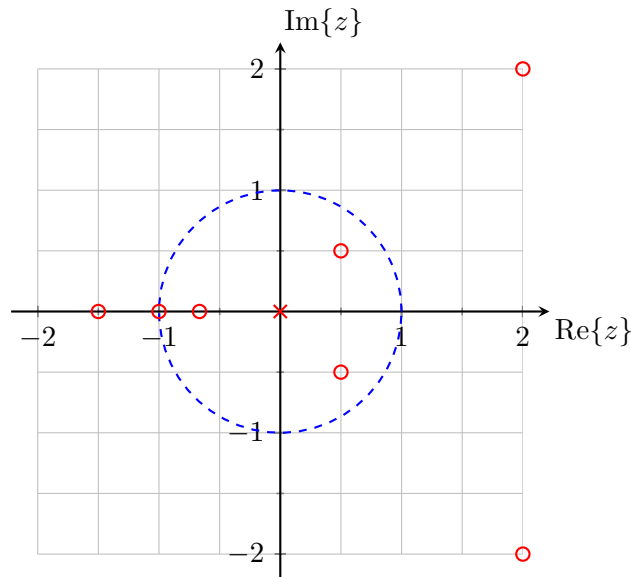
- Show enough (neat) work in the clear spaces on this exam to convince us that you derived, not guessed, your answers.
- Put your final answers in the boxes at the bottom of the page.

You are allowed TWO pages of notes (front and back) for this exam. You may use a graphing calculator of your choice. Consulting with any third party is considered cheating.

Problem	Score	Possible Points
1		20
2		20
3		20
4		20
5		20
<b>Total score</b>		100

# 1 Short answer

- (a) Consider the pole-zero plot for some system  $H(z)$ . Note that the pole at the origin is repeated seven times.



- (i) Is the system stable?

yes	no
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- (ii) Does the system have minimum phase?

yes	no
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- (iii) Does the system have linear phase? If so, what type of linear phase filter is this?

Type I	Type II	Type III	Type IV	Not linear phase
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- (iv) Is the system causal?

yes	no
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(b) The signal

$$x_c(t) = \cos(2\pi(600)t)$$

was sampled with a sampling period  $T = 1/300$  s to obtain a discrete-time signal  $x[n]$ . What is the resulting sequence  $x[n]$ ?

$$x[n] =$$

(c) Suppose we had a continuous signal  $x_c(t)$  and wanted to filter it with some impulse response

$$h(t) = 2e^{-t/3} \cos(12\pi t)u(t).$$

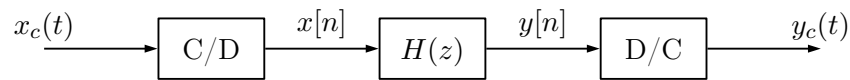
(i) If you implemented this on a discrete-time system using a sampling rate  $T = 0.1$ , what would the discrete-time impulse response  $h[n]$  be?

$$h[n] =$$

(i) What is the issue with implementing this system using this sampling rate?

## 2 Sampling and reconstruction

Consider the signal processing system below, where the C/D and D/C converters are ideal. Let the input  $x_c(t) = 2\cos(30\pi t)$  and the sampling rate be  $f_s = 40$  Hz.

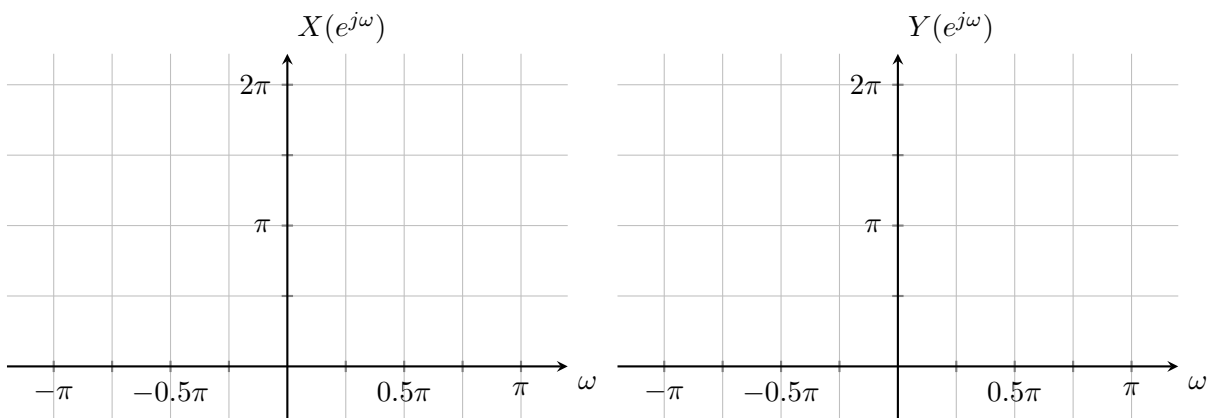


The system  $H(z)$  is described by the input/output relationship  $y[n] = x^2[n]$ .

- (a) Is the subsystem (and only the subsystem)  $H(z)$  linear and/or time-invariant?

Circle one:	linear	non-linear
Circle one:	time-invariant	time-variant

- (b) Sketch  $X(e^{j\omega})$  and  $Y(e^{j\omega})$  from  $\omega = -\pi$  to  $\pi$ .



(c) What is the output  $y_c(t)$ ?

$y_c(t) =$

(d) For which frequencies of  $\Omega$  of  $x_c(t) = 2 \cos(\Omega t)$  will  $y_c(t) = x_c^2(t)$ ?

$\Omega :$

### 3 Minimum phase and all-pass systems

Consider a FIR system with one zero at  $z = 0.5$ , one zero at  $z = 0.8e^{j\frac{\pi}{5}}$ , and all associated zeros needed for a Type I linear phase filter.

- (a) What is the transfer function  $H(z)$  of the overall system? You do not need to simplify the transfer function beyond second-order stages.

$H(z) =$

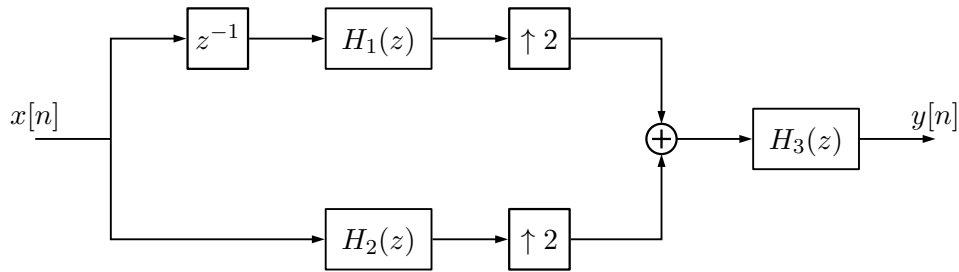
- (b) Decompose this system to a minimum phase system  $H_{\min}(z)$  and an all-pass system  $H_{\text{ap}}(z)$ .

$$H_{\min}(z) =$$

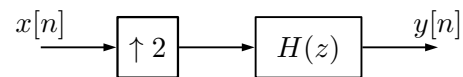
$$H_{\text{ap}}(z) =$$

## 4 Interpolation

Consider the system below.



This system can also be implemented as a cascade as an expander followed by a simple FIR filter  $H(z)$ . This simplified system is shown below.



The filter  $H(z)$  is given as

$$H(z) = 1 + z^{-1} + z^{-2} + z^{-3} + z^{-4} + z^{-5}.$$

- (a) Please find  $H(z)$  in terms of  $H_1$ ,  $H_2$ ,  $H_3$ .

$H(z) =$



(b) The filters  $H_1(z)$ ,  $H_2(z)$ , and  $H_3(z)$  each have two taps. Thus, we can represent

$$H_1(z) = h_1[0] + h_1[1]z^{-1}$$

$$H_2(z) = h_2[0] + h_2[1]z^{-1}$$

$$H_3(z) = h_3[0] + h_3[1]z^{-1}.$$

Find the values  $h_1[0]$ ,  $h_1[1]$ ,  $h_2[0]$ ,  $h_2[1]$ ,  $h_3[0]$ , and  $h_3[1]$ . There are multiple correct solutions. I will check your answer in a Python script to verify it works.

$$h_1[0] =$$

$$h_1[1] =$$

$$h_2[0] =$$

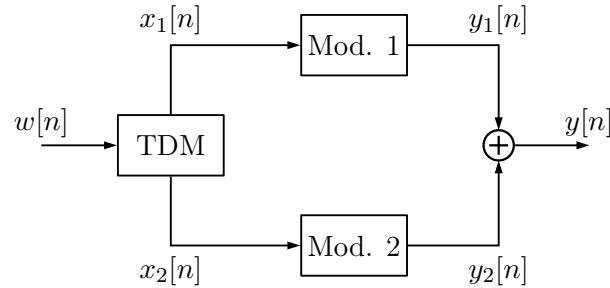
$$h_2[1] =$$

$$h_3[0] =$$

$$h_3[1] =$$

## 5 Time-division multiplexing

Communication systems often require conversion from time-division multiplexing (TDM) to frequency-division multiplexing (FDM). In this problem, we examine a simple example of such a system. The full block diagram of the system is given below.



- (a) The TDM input,  $w[n]$ , is the sequence of interleaved samples of two signals,  $x_1[n]$  and  $x_2[n]$ , such that

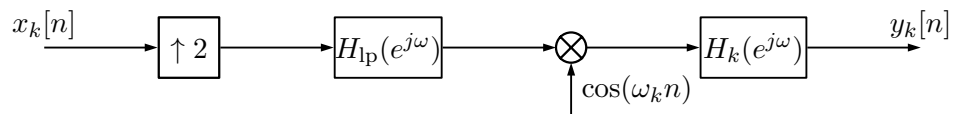
$$w[n] = \begin{cases} x_1[n/2] & \text{if } n \text{ is even} \\ x_2[(n-1)/2] & \text{if } n \text{ is odd.} \end{cases}$$

Design and draw a block diagram of a TDM demultiplexer that separates  $w[n]$  into  $x_1[n]$  and  $x_2[n]$ . You may use only the following blocks:

- delay
- decimator by a factor of  $M$  ( $\downarrow M$ , if used specify  $M$ )
- expander by a factor of  $L$  ( $\uparrow L$ , if used specify  $L$ )
- ideal low-pass filter with cutoff frequency  $\omega_c$ , if used specify  $\omega_c$

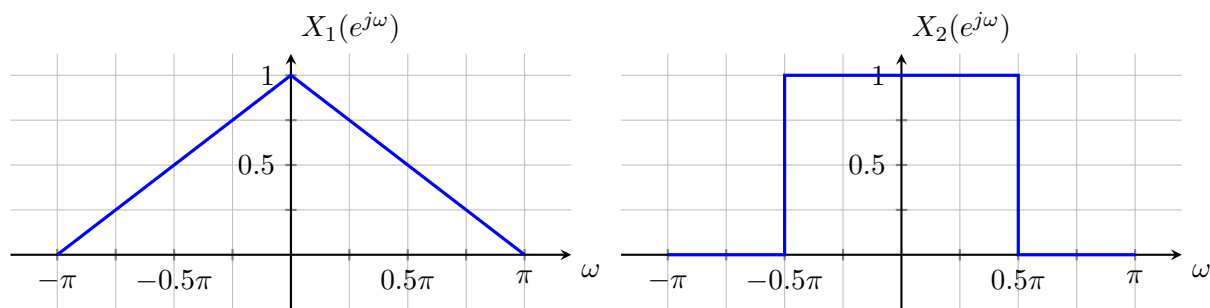
Your work continued...

(b) Each modulator system ( $k = 1$  or  $2$ ) is given by the block diagram below.



All filters are ideal. The low-pass filter  $H_{\text{lpf}}(e^{j\omega})$  is the same for both channels and has a gain of 2 and a cutoff frequency  $\omega_c = \pi/2$ . The high pass filters  $H_k(e^{j\omega})$  have unity gain and cutoff frequencies  $\omega_1 = \pi/4$  and  $\omega_2 = 3\pi/4$  for  $k = 1$  and  $k = 2$ , respectively.

The DTFT of  $x_1[n]$  and  $x_2[n]$  are plotted from  $\omega \in [-\pi, \pi]$ . Sketch the DTFT of the full system  $y[n]$ . Sketch the DTFT at each point in the modulator system for partial credit.



Your work continued...

