

# Network-Based Line-of-Sight Path Tracking of Underactuated Unmanned Surface Vehicles With Experiment Results

Wentao Wu<sup>✉</sup>, Zhouhua Peng<sup>✉</sup>, *Senior Member, IEEE*, Dan Wang<sup>✉</sup>, *Senior Member, IEEE*,  
Lu Liu<sup>✉</sup>, *Member, IEEE*, and Qing-Long Han<sup>✉</sup>, *Fellow, IEEE*

**Abstract**—This article deals with the problem of network-based path-tracking control of an underactuated unmanned surface vehicle subject to model uncertainties and unknown disturbances over a wireless network. A two-level network-based control architecture is proposed, including a local inner loop and a remote outer loop. In the remote outer loop, an event-triggered line-of-sight guidance law is designed to achieve path tracking while reducing the network burden for the remote control at the kinematic level. In the local inner loop, an extended state observer is employed to estimate the unknown disturbances due to the model uncertainties and environmental disturbances. Based on the estimated information from the extended state observer, an event-triggered anti-disturbance control law is developed to reduce the execution rate of actuators at the kinetic level. The stability of the closed-loop path-tracking system is proved based on the input-to-state stability and cascade stability theory. The effectiveness of the proposed network-based method for path tracking of the USV is verified via experiments.

**Index Terms**—Event-trigger, extended state observer (ESO), line of sight (LOS), path tracking, unmanned surface vehicle (USV).

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Wentao Wu is with the School of Marine Electrical Engineering, Dalian Maritime University, Dalian 116026, China, and also with the Science and Technology on Underwater Vehicle Technology, Harbin Engineering University, Harbin 150001, China (e-mail: wuwentao@dlmu@gmail.com).

Zhouhua Peng, Dan Wang, and Lu Liu are with the School of Marine Electrical Engineering, Dalian Maritime University, Dalian 116026, China (e-mail: zhpeng@dlmu.edu.cn; dwang@dlmu.edu.cn; wendaerji@163.com).

Qing-Long Han is with the School of Software and Electrical Engineering, Swinburne University of Technology, Melbourne, VIC 3122, Australia (e-mail: qhan@swin.edu.au).

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## I. INTRODUCTION

IN RECENT years, unmanned surface vehicles (USVs) have received considerable attention due to their ubiquitous advantages in several military and scientific applications [1], [2]. A variety of motion control scenarios is considered, such as trajectory tracking [3], [4]; target tracking [5]–[7]; path following [8]–[12]; target enclosing [13], [14]; and path tracking [15]–[17]. In order to achieve these motion scenarios, kinematic guidance law and kinetic control law are two critical components of any motion controllers. In the guidance loop, typical guidance methods includes line of sight (LOS) [18]–[20] and constant bearing guidance [21]. In the kinetic control loop, adaptive control [22], [23]; neural-network control [24], [25]; fuzzy control [21], [26]; sliding-mode control [27]; model predictive control [28], [29]; and extended-state-observer-based control [30]–[32] methods are widely used to account for model uncertainties and ocean disturbances. However, there are a few results which consider motion control of marine surface vehicles from a network standpoint.

With the fast development of unmanned and networked technologies, network-based control [33], [34] is the main trend for motion control of unmanned systems, including USVs. Some attempts have been made on motion control of USVs in network environments [26], [35]–[38]. In [26], a network-based fuzzy model is proposed for dynamic positioning of a USV subject to network-induced delays and packet dropouts. In [35], a network-based model is designed for a USV subject to actuator faults and external disturbances where network-induced delays and packet dropouts occurring in sampler-to-control station or control station-to-actuator channel are overcome. In [36], a network-based model is employed to stabilize the heading angle and reduce the rudder oscillations. In [37], a sampled-data design is presented in the synchronized path following where the path variables are updated in a discrete-time manner. In [38], an event-triggered modular-ISS neural-network control method is developed to reduce the communication burden during distributed formation control of multiple USVs. However, few works are available for network-based path tracking of USVs.

Motivated by the observations above, this article aims to address the network-based path-tracking control of a USV in

network environments. The USV suffers from model uncertainties and external ocean disturbances. A network-based control architecture is proposed for path tracking of the USV, which involves a local inner loop design and a remote outer loop design. Specifically, in the remote outer loop, an event-triggered guidance law based on the LOS principle is designed to achieve network-based path tracking without periodic communication. The network burden during the remote-control process can be decreased and the desired performance of path tracking still can be maintained under the constrained network. In the local inner loop, an extended state observer (ESO) is designed to estimate an unknown total disturbance composed of model uncertainties and environmental ocean disturbances. Based on the estimated disturbances from the ESO, an event-triggered anti-disturbance control law is developed to reduce the actuation burden at the kinetic level. Through input-to-state stability and cascade stability analysis, the stability of the closed-loop path-tracking system is proved, and all signals are uniformly ultimately bounded. Experiments are carried out to show the effectiveness of the proposed network-based method for path tracking of the USV.

The salient features of the proposed design method are as follows. First, compared with the results in [3]–[12], [17]–[20], and [16], [39], [40] without involving networks between the guidance loop and control loop, a network-based two-level control architecture is proposed for a USV which enables the separation of the guidance loop and the control loop in space. The advantage of the architecture is that the kinematic motion scenario can be achieved from the remote command center. Next, in contrast to the existing LOS guidance laws in [8]–[10], [19], [20], [30], and [41]–[43] based on the periodical sampling or communication, an event-triggered LOS guidance law is designed with aperiodic sampling or communication. Finally, in contrast to some kinetic control laws in [30], [31], and [45], based on the time-triggered ESOs, an event-triggered kinetic control law is designed herein based on the event-triggered ESOs, which can reduce the actuation and communication burden in the control loop. Besides, the proposed control law is much simpler than the event-triggered neural-network control law in [44], which facilitates practical implementations.

The organization of this article is arranged as follows. Section II presents the problem formulation. Section III gives the network-based control law design. Section IV presents the stability of the closed-loop path-tracking system. Section V provides experiment results for illustrations. Section VI concludes this article.

**Notation:** The following notations are used in this article.  $\mathbb{N}^+$ ,  $\mathbb{R}$ ,  $\mathbb{R}^+$ , and  $\mathbb{R}^{n \times m}$  present a positive integer set, a real set, a positive real set, and an  $n \times m$ -dimensional matrix set, respectively.  $\text{diag}\{\dots\}$  and  $\text{col}(\dots)$ , respectively, denote a block-diagonal matrix and a column vector.  $A^T$  describes the transpose of a matrix  $A$ .  $|a|$  and  $\|b\|$  present the absolute value of a real number  $a$  and the Euclidean norm of a vector  $b$ , respectively.  $\lambda_{\max}(B)$  and  $\lambda_{\min}(B)$  are the maximum eigenvalue and the minimum eigenvalue of a symmetric matrix  $B$ , respectively;  $I_n$  represents the identity matrix with  $n$  dimensions.

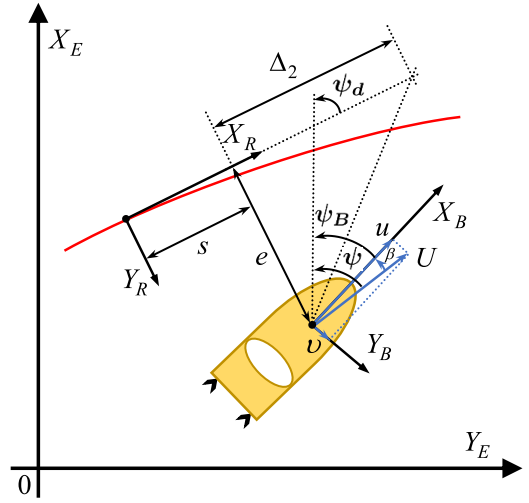


Fig. 1. LOS guidance.

## II. PROBLEM FORMULATION

As shown in Fig. 1, the mathematical model of a USV can be expressed in the north-east-down reference frame  $X_E - Y_E$  and the body-fixed reference frame  $X_B - Y_B$ . The kinematic equation of the USV is expressed as [46]

$$\begin{cases} \dot{x} = U \cos \psi \\ \dot{y} = U \sin \psi \\ \dot{\psi} = r + \beta_d \end{cases} \quad (1)$$

where  $x \in \mathbb{R}$  and  $y \in \mathbb{R}$  are the position of the USV in  $X_E - Y_E$ ;  $\psi$  presents its course angle in  $X_E - Y_E$ ;  $U = \sqrt{u^2 + v^2}$  is the total speed, where  $u$  and  $v$  denote the surge and sway speed expressed in  $X_B - Y_B$ , respectively;  $r = \dot{\psi}_B$  denotes the angular velocity with  $\psi_B$  denoting the yaw angle;  $\beta = \text{atan2}(v, u) \in (-\pi, \pi]$  is the sideslip angle in the reference frame  $X_B - Y_B$ , where  $\text{atan2}(v, u)$  is the four-quadrant version of  $\arctan(v/u) \in (-\pi/2, \pi/2)$ ; and  $\beta_d = \dot{\beta}$  presents the time derivative of the sideslip angle.

The kinetic equation of the USV is described as [46]

$$\begin{cases} m_{11}\dot{u} = f_u(u, v, r) + \tau_u(t) + \tau_{uw}(t) \\ m_{22}\dot{v} = f_v(u, v, r) + \tau_{vw}(t) \\ m_{33}\dot{r} = f_r(u, v, r) + \tau_r(t) + \tau_{rw}(t) \end{cases} \quad (2)$$

where  $m_{11}$ ,  $m_{22}$ , and  $m_{33}$  are the inertia of the USV;  $f_u(u, v, r)$ ,  $f_v(u, v, r)$ , and  $f_r(u, v, r)$  are the unknown functions, including Coriolis terms, damping terms, and unmodeled dynamics;  $\tau_{uw}$ ,  $\tau_{vw}$ , and  $\tau_{rw}$  are the unknown sea loads induced by wind, waves, and currents; and  $\tau_u$  and  $\tau_r$  are the control input torque and moment, respectively.

Note that  $u = U \cos \beta$  and  $v = U \sin \beta$ . It follows from (2) that:

$$\begin{cases} m_{11}\dot{U} = \cos \beta (f_u(u, v, r) + \tau_{uw}) \\ \quad + \sin \beta (\tau_{vw} + f_v(u, v, r)) \frac{m_{11}}{m_{22}} - 2\tau_u \sin^2\left(\frac{\beta}{2}\right) \\ \quad - m_{11}\beta_d (v \cos \beta + u \sin \beta) + \tau_u \\ m_{33}\dot{r} = f_r(u, v, r) + \tau_{rw} + \tau_r. \end{cases} \quad (3)$$

Consider a path expressed as  $p_d = \text{col}(x_d(\theta), y_d(\theta)) \in \mathbb{R}^2$ , where  $\theta \in \mathbb{R}$  is a path variable. Define two kinematic tracking errors as

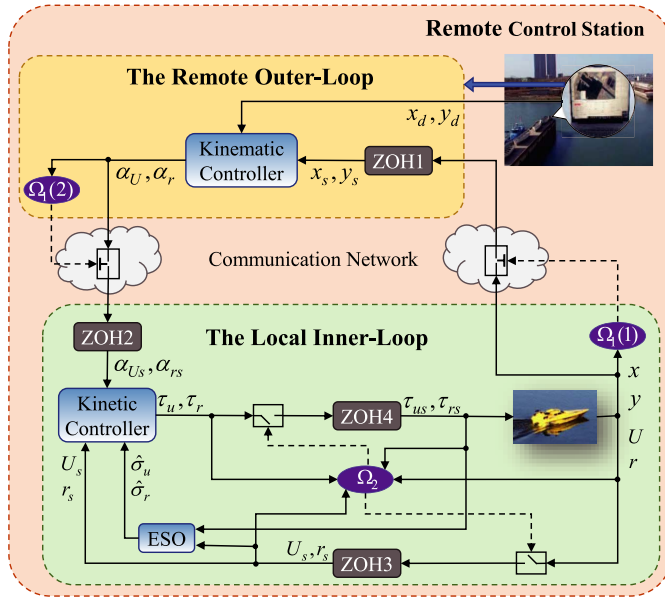


Fig. 2. Two-level network-based guidance and control architecture.

$$\begin{bmatrix} s \\ e \end{bmatrix} = \begin{bmatrix} \cos(\psi_d) & -\sin(\psi_d) \\ \sin(\psi_d) & \cos(\psi_d) \end{bmatrix}^T \begin{bmatrix} x - x_d(\theta) \\ y - y_d(\theta) \end{bmatrix} \quad (4)$$

where  $s$  and  $e$  are the along-track error and cross-track error;  $\psi_d = \text{atan2}(\dot{y}_d(\theta), \dot{x}_d(\theta))$  denotes the path-tangential angle with  $\dot{y}_d(\theta) = \partial y_d(\theta)/\partial \theta$  and  $\dot{x}_d(\theta) = \partial x_d(\theta)/\partial \theta$ . The following assumption is needed in this article.

*Assumption 1:* The signals of the parameterized path  $p_d(\theta)$   $\partial p_d(\theta)/\partial \theta$  and  $\partial^2 p_d(\theta)/\partial^2 \theta$  are bounded.

The control objective of this article is to develop a network-based path-tracking control law for the USV described by (1) and (2) such that

$$\lim_{t \rightarrow \infty} |s| \leq l_s^*, \lim_{t \rightarrow \infty} |e| \leq l_e^* \quad (5)$$

where  $l_s^*$  and  $l_e^*$  are positive constants.

### III. DESIGN AND ANALYSIS

In this section, an event-triggered two-level network-based control architecture is presented. This architecture includes a local inner loop design and a remote outer loop design shown in Fig. 2.

#### A. Remote Outer Loop Guidance Law

In this section, a remote outer loop guidance law is first proposed based on an LOS guidance scheme and an event-triggered scheme. Then, some triggering conditions are presented for scheduling the data transmissions. Finally, the input-to-state stability of the guidance loop is analyzed.

1) *Guidance Law Design:* Taking the time derivative of (4) and using (1), it follows that:

$$\begin{cases} \dot{s} = U \cos(\psi - \psi_d) + er_d - u_0 U_d \\ \dot{e} = U \sin(\psi - \psi_d) - sr_d \\ \dot{\psi} = r + \beta_d \end{cases} \quad (6)$$

where  $U_d = \sqrt{\dot{x}_d^2(\theta) + \dot{y}_d^2(\theta)}$ ,  $r_d = \dot{\psi}_d$ , and  $\dot{\theta} = u_0$  with ( $u_0 \in \mathbb{R}^+$ ) being a given update velocity of the path variable.

Define  $\epsilon_U = \alpha_{U_s} - \alpha_U$ ,  $q_U = U - \alpha_{U_s}$ ,  $\epsilon_r = \alpha_{r_s} - \alpha_r$ ,  $q_r = r - \alpha_{r_s}$ , and  $\psi_e = \psi - \alpha_\psi$ , where  $\alpha_U$ ,  $\alpha_r$ , and  $\alpha_\psi$  are the desired guidance signals; and  $\alpha_{U_s}$  and  $\alpha_{r_s}$  denote the sampling signals stored in the zero order holder (ZOH). The signal is maintained as a constant in the ZOH for the whole sampling period. The sampling signal stored in the ZOH will be updated until the corresponding event-triggered condition is satisfied. For a sampling period  $t \in [t_i^\eta, t_{i+1}^\eta)$  from the  $i$ th to the  $(i+1)$ th releasing state of the USV over the wireless network, the kinematic tracking error dynamics in (6) can be written as

$$\begin{cases} \dot{s} = \alpha_U + \epsilon_U + q_U + er_d - u_0 U_d \\ \quad - 2U \sin^2\left(\frac{\psi - \psi_d}{2}\right) \\ \dot{e} = U \sin(\alpha_\psi - \psi_d) + \epsilon - sr_d \\ \dot{\psi}_e = \alpha_r + \epsilon_r + q_r + \beta_d - \dot{\alpha}_\psi \end{cases} \quad (7)$$

where  $\epsilon = 2U \sin(\psi_e/2) \cos((\psi + \alpha_\psi - 2\psi_d)/2)$ . Based on (4), the sampling path tracking errors  $s_s$  and  $e_s$  are defined as

$$\begin{bmatrix} s_s \\ e_s \end{bmatrix} = \begin{bmatrix} \cos(\psi_d) & -\sin(\psi_d) \\ \sin(\psi_d) & \cos(\psi_d) \end{bmatrix}^T \begin{bmatrix} x_s - x_d(\theta) \\ y_s - y_d(\theta) \end{bmatrix}. \quad (8)$$

To stabilize (7), an event-triggered LOS guidance law is proposed as

$$\begin{cases} \alpha_U = -\frac{k_1 s_s}{\sqrt{s_s^2 + \Delta_1^2}} + u_0 U_d \\ \alpha_\psi = -\arctan\left(\frac{e_s}{\Delta_2}\right) + \psi_d \\ \alpha_r = -\frac{k_2 \psi_e}{\sqrt{\psi_e^2 + \Delta_3^2}} + r_d \end{cases} \quad (9)$$

where  $k_1, k_2 \in \mathbb{R}^+$  are control gains;  $\Delta_1, \Delta_3 \in \mathbb{R}^+$  are constants;  $\Delta_2 \in \mathbb{R}^+$  is a look-ahead distance.

Substituting (9) into (7), the dynamics of the kinematic tracking errors are written as

$$\begin{cases} \dot{s} = -k_1(s + \epsilon_s) + \epsilon_U + q_U + er_d \\ \quad - 2U \sin^2\left(\frac{\psi - \psi_d}{2}\right) \\ \dot{e} = -k_2(e + \epsilon_e) + \epsilon - sr_d \\ \dot{\psi}_e = -k_3 \psi_e + \epsilon_r + q_r + \beta_d \end{cases} \quad (10)$$

where  $k_1 = k_1/\sqrt{s_s^2 + \Delta_1^2}$ ,  $k_2 = U/\sqrt{e_s^2 + \Delta_2^2}$ , and  $k_3 = k_2/\sqrt{\psi_e^2 + \Delta_3^2}$  which can be considered as time-varying gains;  $\epsilon_s = s_s - s$  and  $\epsilon_e = e_s - e$  are errors between the triggered path tracking errors and practical path tracking errors.

2) *Kinematic Event-Triggered Data Transmission*: As shown in Fig. 2, there are two network channels between the remote outer loop and the local inner loop. To reduce the communication burden, the triggering conditions  $\Omega_1$  are described as

$$\begin{cases} \Omega_1(1) : \|\eta_s(t) - \eta(t)\| \leq \epsilon_\eta^* \\ \Omega_1(2) : \|\alpha_s(t) - \alpha(t)\| \leq \epsilon_\alpha^* \end{cases} \quad (11)$$

where  $\eta(t) = \text{col}(x(t), y(t), \psi(t))$  and  $\alpha(t) = \text{col}(\alpha_U(t), \alpha_r(t))$ ;  $\eta_s(t)$  and  $\alpha_s(t)$  present the sampling signals stored in ZOH1.  $\epsilon_\eta^* \in \mathbb{R}^+$  and  $\epsilon_\alpha^* \in \mathbb{R}^+$  are the predefined triggering thresholds.

To be more specific, the variable  $t_i^\eta \in \mathbb{R}^+$  with  $i \in \mathbb{N}^+$  denotes the  $i$ th triggering time with the initial time  $t_0^\eta$ . The event detector in the local loop continuously compares the tracking errors by the  $\Omega_1(1)$ . If  $\Omega_1(1)$  is false at  $t_{i+1}^\eta$ , the state information at  $t_{i+1}^\eta$  is transmitted to store the ZOH1 over the wireless network. During  $[t_i^\eta, t_{i+1}^\eta)$ , the wireless network associated with the ZOH2 can be triggered to transmit the desired commands for the USV only if  $\Omega_1(2)$  is not satisfied. Similarly, let  $t_j^\alpha \in \mathbb{R}^+$  with  $j \in \mathbb{N}^+$  present the releasing moment of the desired signals.

3) *Stability Analysis*: We first analyze the stability of the event-triggered kinematic subsystem (10).

*Lemma 1*: Under Assumption 1, the event-triggered kinematic subsystem (10):  $[z_e, \alpha_e, q, U, \beta_d] \mapsto z_1$  is input-to-state stable (ISS).

*Proof*: Consider a Lyapunov function as

$$V_1 = \frac{1}{2}(s^2 + e^2 + \psi_e^2). \quad (12)$$

Taking the time derivative of  $V_1$  for  $t \in [t_i^\eta, t_{i+1}^\eta)$  along (10) yields

$$\begin{aligned} \dot{V}_1 = & -k_1 s^2 - k_2 e^2 - k_3 \psi_e^2 - k_1 s \epsilon_s - k_2 e \epsilon_e \\ & + s \epsilon_U + \psi_e \epsilon_r + s q_U + \psi_e q_r + e \epsilon + \psi_e \beta_d \\ & - 2sU \sin^2\left(\frac{\psi - \psi_d}{2}\right). \end{aligned} \quad (13)$$

Let  $z_1 = \text{col}(s, e, \psi_e)$ . Rewrite (13) as

$$\begin{aligned} \dot{V}_1 \leq & -c_1 \|z_1\|^2 + \|z_1\|(\lambda_{\max}(K)\|z_e\| + \|\alpha_e\| \\ & + \|q\| + |\beta_d| + 2|U|) \end{aligned} \quad (14)$$

where  $K = \text{diag}\{k_1, k_2, k_3\}$ ,  $c_1 = \lambda_{\min}(K)$ ,  $z_e = \text{col}(\epsilon_s, \epsilon_e)$ ,  $\alpha_e = \text{col}(\epsilon_U, \epsilon_r)$ , and  $q = \text{col}(q_U, q_r)$ .

As

$$\|z_1\| \geq \frac{\lambda_{\max}(K)\|z_e\| + \|\alpha_e\| + \|q\| + |\beta_d| + 2|U|}{c_1 \kappa_1} \quad (15)$$

it renders that

$$\dot{V}_1 \leq -c_1(1 - \kappa_1)\|z_1\|^2 \quad (16)$$

where  $0 < \kappa_1 < 1$ .

Thus, it is concluded that the even-triggered kinematic subsystem (10) is ISS [47], and the ultimate bound is given by

$$\|z_1\| \leq \max\left\{\|z_1(t_0)\|e^{-c_1(1-\kappa_1)(t-t_0)}, \frac{\|\alpha_e\|}{c_1 \kappa_1} + \frac{\lambda_{\max}(K)\|z_e\| + \|q\| + |\beta_d| + 2|U|}{c_1 \kappa_1}\right\}. \quad (17)$$

The proof is completed.  $\blacksquare$

### B. Local Inner Loop Control Law

In this section, an event-triggered kinetic control law is first developed by using the estimated information from ESOs. Then, the event-triggering conditions are presented for the inner loop such that the actuation times can be reduced. Finally, the stability of the inner loop is analyzed.

1) *Event-Triggered Control Law Design*: To facilitate the ESO design, rewrite (3) as

$$\begin{cases} \dot{U} = \sigma_u(u, v, r, t) + \frac{\tau_u}{m_{11}} \\ \dot{r} = \sigma_r(u, v, r, t) + \frac{\tau_r}{m_{33}} \end{cases} \quad (18)$$

where  $\sigma_u(u, v, r, t) = \cos\beta(f_u(u, v, r) + \tau_{uw})/m_{11} + \sin\beta(f_v(u, v, r) + \tau_{vw})/m_{22} - \beta_d(v\cos\beta + u\sin\beta) - 2\tau_u\sin^2(\beta/2)/m_{11}$  and  $\sigma_r(u, v, r, t) = (f_r(u, v, r) + \tau_{rw})/m_{33}$ .

Let  $\hat{U}$ ,  $\hat{\sigma}_u$ ,  $\hat{r}$ , and  $\hat{\sigma}_r$  be the estimates of  $U$ ,  $\sigma_u$ ,  $r$ , and  $\sigma_r$ , respectively. Two ESOs based on the triggered signals  $U_s$  and  $r_s$  are used to estimate  $\sigma_u$  and  $\sigma_r$  as

$$\begin{cases} \dot{\hat{U}} = -k_1^u(\hat{U} - U_s) + \hat{\sigma}_u + \frac{\tau_{us}}{m_{11}} \\ \dot{\hat{\sigma}}_u = -k_2^u(\hat{U} - U_s) \\ \dot{\hat{r}} = -k_1^r(\hat{r} - r_s) + \hat{\sigma}_r + \frac{\tau_{rs}}{m_{33}} \\ \dot{\hat{\sigma}}_r = -k_2^r(\hat{r} - r_s) \end{cases} \quad (19)$$

where  $k_1^u, k_2^u, k_1^r$ , and  $k_2^r \in \mathbb{R}^+$  are observer parameters. Define the estimation errors  $\vartheta_U = \hat{U} - \alpha_{U_s}$  and  $\vartheta_r = \hat{r} - \alpha_{r_s}$ . Taking the time derivative of  $\vartheta_U$  and  $\vartheta_r$  over a control interval  $t \in [t_m^\tau, t_{m+1}^\tau)$ , respectively, we have

$$\begin{cases} \dot{\vartheta}_U = -k_1^u(\tilde{U} - U_e) + \hat{\sigma}_u + \frac{\tau_{us}}{m_{11}} \\ \dot{\vartheta}_r = -k_1^r(\tilde{r} - r_e) + \hat{\sigma}_r + \frac{\tau_{rs}}{m_{33}} \end{cases} \quad (20)$$

where  $\tilde{U} = \hat{U} - U$  and  $\tilde{r} = \hat{r} - r$ .

An anti-disturbance kinetic control law is designed to stabilize the dynamics of  $\vartheta_U$  and  $\vartheta_r$  as

$$\begin{cases} \tau_u = m_{11}(-k_u(U_s - \alpha_{U_s}) - \hat{\sigma}_u) \\ \tau_r = m_{33}(-k_r(r_s - \alpha_{r_s}) - \hat{\sigma}_r) \end{cases} \quad (21)$$

where  $k_u \in \mathbb{R}^+$  and  $k_r \in \mathbb{R}^+$  are control gains.

Let  $\tau_{ue} = \tau_{us} - \tau_u$  and  $\tau_{re} = \tau_{rs} - \tau_r$ . Substituting (21) into (20), we have the time derivative of  $\vartheta_U$  and  $\vartheta_r$  as

$$\begin{cases} \dot{\vartheta}_U = -k_u \vartheta_U - k_e^u (\tilde{U} - U_e) + \frac{\tau_{ue}}{m_{11}} \\ \dot{\vartheta}_r = -k_r \vartheta_r - k_e^r (\tilde{r} - r_e) + \frac{\tau_{re}}{m_{33}} \end{cases} \quad (22)$$

where  $k_e^u = k_1^u - k_u$  and  $k_e^r = k_1^r - k_r$ .

Define  $\tilde{\sigma}_u = \hat{\sigma}_u - \sigma_u$  and  $\tilde{\sigma}_r = \hat{\sigma}_r - \sigma_r$ . To make it easier to analyze the stability of the kinetic subsystem, the error dynamics of the (19) can be expressed as

$$\begin{cases} \dot{\tilde{U}} = -k_1^u \tilde{U} + \tilde{\sigma}_u + k_1^u U_e + \frac{\tau_{ue}}{m_{11}} \\ \dot{\tilde{\sigma}}_u = -k_2^u \tilde{U} + k_2^u U_e - \dot{\sigma}_u \\ \dot{\tilde{r}} = -k_1^r \tilde{r} + \tilde{\sigma}_r + k_1^r r_e + \frac{\tau_{re}}{m_{33}} \\ \dot{\tilde{\sigma}}_r = -k_2^r \tilde{r} + k_2^r r_e - \dot{\sigma}_r. \end{cases} \quad (23)$$

Let  $z_2 = \text{col}(\tilde{U}, \tilde{\sigma}_u, \tilde{r}, \tilde{\sigma}_r)$ . It follows from (23) that:

$$\dot{z}_2 = A_1 z_2 + B_1 v_e + B_2 \tau_e - B_3 \dot{\sigma} \quad (24)$$

where  $v_e = v_s(t) - v(t) = \text{col}(U_e, r_e)$ ,  $\tau_e = \tau_s(t) - \tau(t) = \text{col}(\tau_{ue}, \tau_{re})$ ,  $\dot{\sigma} = \text{col}(\dot{\sigma}_u, \dot{\sigma}_r)$ , and

$$\begin{cases} A_1 = \begin{bmatrix} -k_1^u & 1 & 0 & 0 \\ -k_2^u & 0 & 0 & 0 \\ 0 & 0 & -k_1^r & 1 \\ 0 & 0 & -k_2^r & 0 \end{bmatrix} \\ B_1 = \begin{bmatrix} k_1^u & k_2^u & 0 & 0 \\ 0 & 0 & k_1^r & k_2^r \end{bmatrix}^T \\ B_2 = \begin{bmatrix} \frac{1}{m_{11}} & 0 & 0 & 0 \\ 0 & 0 & \frac{1}{m_{33}} & 0 \end{bmatrix}^T \\ B_3 = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}^T. \end{cases} \quad (25)$$

Note that the matrix  $A_1$  is a Hurwitz matrix. There exists a positive-definite matrix  $P_1$  satisfying the following inequality:

$$A_1^T P_1 + P_1 A_1 \leq -c_2 I_1 \quad (26)$$

where  $c_2 \in \mathbb{R}^+$ .

2) *Kinetic Event-Triggered Data Transmission*: As shown in Fig. 2, an event-triggered condition  $\Omega_2$  is presented to determine the event-triggered schedule to transmit the data. Two triggering thresholds are given as

$$\begin{cases} \Omega_2(1) : \|v_s(t) - v(t)\| \leq \epsilon_v^* \\ \Omega_2(2) : \|\tau_s(t) - \tau(t)\| \leq \epsilon_\tau^* \\ \Omega_2(3) : v_s \text{ is transmitted into the controller (21)} \end{cases} \quad (27)$$

where  $v(t) = \text{col}(U(t), r(t))$ , and  $\tau(t) = \text{col}(\tau_u(t), \tau_r(t))$ ,  $v_s(t)$  and  $\tau_s(t)$  denote the sampling signal,  $\epsilon_v^* \in \mathbb{R}^+$  and  $\epsilon_\tau^* \in \mathbb{R}^+$  are the triggering thresholds.

In the local loop, define the data transmission moment  $t_m^v \in \mathbb{R}^+$  with  $m \in \mathbb{N}^+$  from the actuator to controller. Under  $\Omega_2(1)$  false condition, the previous data  $v(t_m^v)$  stored in the ZOH3 is updated at  $t_{m+1}^v$ . The estimator (19) and controller (21) will recalculate the estimated disturbances and

control inputs of the USV using data updated in ZOH3. If the triggering condition  $\Omega_2(2)$  is not satisfied, the control inputs in the ZOH4 will be updated over  $[t_m^v, t_{m+1}^v)$ . Let  $t_n^r \in \mathbb{R}^+$  with  $n \in \mathbb{N}^+$  represent the triggering time if  $\Omega_2(2)$  is false.

3) *Stability Analysis*: We now analyze the stability of the kinetic subsystem (22) and the ESO subsystem (24). The following assumption is needed.

*Assumption 2*: The time derivatives of  $\sigma_u(\cdot)$  and  $\sigma_r(\cdot)$  are bounded and satisfy  $|\dot{\sigma}_u| \leq \dot{\sigma}_u^*$  with  $\dot{\sigma}_u^* \in \mathbb{R}^+$  and  $|\dot{\sigma}_r| \leq \dot{\sigma}_r^*$  with  $\dot{\sigma}_r^* \in \mathbb{R}^+$ .

*Lemma 2*: Under Assumption 2, the ESO subsystem (24):  $[v_e, \tau_e, \dot{\sigma}] \mapsto [z_2]$  is ISS.

*Proof*: Construct a Lyapunov function as

$$V_2 = \frac{1}{2} z_2^T P_1 z_2. \quad (28)$$

Taking the time derivative of  $V_2$  along (24) over  $[t_n^r, t_{n+1}^r)$ , it follows that:

$$\dot{V}_2 = z_2^T P_1 (A_1 z_2 + B_1 v_e + B_2 \tau_e - B_3 \dot{\sigma}). \quad (29)$$

Using (26), it follows that:

$$\dot{V}_2 \leq -\frac{c_2 \|z_2\|^2}{2} + \|z_2\| (\|P_1 B_1\| \|v_e\| + \|P_1 B_2\| \|\tau_e\| + \|P_1 B_3\| \|\dot{\sigma}\|) \quad (30)$$

where  $0 < \kappa_2 < 1$ .

As

$$\|z_2\| \geq \frac{2(\|P_1 B_1\| \|v_e\| + \|P_1 B_2\| \|\tau_e\|)}{c_2 \kappa_2} + \frac{2\|P_1 B_3\| \|\dot{\sigma}\|}{c_2 \kappa_2} \quad (31)$$

under (30), it follows that:

$$\dot{V}_2 \leq -\frac{c_2}{2} (1 - \kappa_2) \|z_2\|^2. \quad (32)$$

Therefore, the ESO subsystem (24) is ISS and the ultimate bound is given by

$$\|z_2\| \leq \max \left\{ \|z_2(t_0)\| e^{-\frac{c_2(1-\kappa_2)(t-t_0)}{2}}, 2\sqrt{\lambda_{\max}(P_1)} \times \frac{\|P_1 B_1\| \|v_e\| + \|P_1 B_2\| \|\tau_e\| + \|P_1 B_3\| \|\dot{\sigma}\|}{c_2 \kappa_2 \sqrt{\lambda_{\min}(P_1)}} \right\}. \quad (33)$$

The proof is completed. ■

*Lemma 3*: The kinetic subsystem (22):  $[\tilde{U}, \tilde{r}, v_e, \tau_e] \mapsto [z_3]$  is ISS.

*Proof*: Define a Lyapunov function as

$$V_3 = \frac{1}{2} (\vartheta_U^2 + \vartheta_r^2). \quad (34)$$

Taking the time derivative of  $V_3$  and using (22) over  $[t_m^v, t_{m+1}^v)$ , it renders

$$\begin{aligned} \dot{V}_3 = & -k_u \vartheta_U^2 - k_r \vartheta_r^2 - k_e^u \tilde{U} \vartheta_U - k_e^r \tilde{r} \vartheta_r + k_e^u U_e \vartheta_U \\ & + k_e^r r_e \vartheta_r + \frac{\tau_{ue}}{m_{11}} \vartheta_U + \frac{\tau_{re}}{m_{33}} \vartheta_r. \end{aligned} \quad (35)$$

Noting that  $z_3 = \text{col}(\vartheta_U, \vartheta_r)$  and  $c_3 = \lambda_{\min}\{k_u, k_r\}$ , one has

$$\begin{aligned} \dot{V}_3 \leq & -c_3 \|z_3\|^2 + (k_e^u + k_e^r) \|z_2\| \|z_3\| \\ & + (\lambda_{\max}(K_e^v) \|v_e\| + \lambda_{\max}(K_m) \|\tau_e\|) \|z_3\| \end{aligned} \quad (36)$$

where  $K_e^v = \text{diag}\{k_e^u, k_e^r\}$  and  $K_m = \text{diag}\{1/m_{11}, 1/m_{33}\}$ .  
Since

$$\|z_3\| \geq \frac{(k_e^u + k_e^r)\|z_2\| + \lambda_{\max}(K_e^v)\|v_e\|}{\kappa_3 c_3} + \frac{\lambda_{\max}(K_m)\|\tau_e\|}{\kappa_3 c_3} \quad (37)$$

it follows that:

$$\dot{z}_3 \leq -c_3(1 - \kappa_3)\|z_3\|^2 \quad (38)$$

where  $0 < \kappa_3 < 1$ .

Therefore, one can conclude that the kinetic subsystem (22) is ISS by [47], and the ultimate bound is given by

$$\|z_3\| \leq \max \left\{ \|z_3(t_0)\| e^{-\frac{c_3(1-\kappa_3)(t-t_0)}{2}}, \frac{k_e^u + k_e^r}{\kappa_3 c_3} \|z_2\| + \frac{\lambda_{\max}(K_e^v)\|v_e\| + \lambda_{\max}(K_m)\|\tau_e\|}{\kappa_3 c_3} \right\}. \quad (39)$$

The proof is completed.  $\blacksquare$

#### IV. STABILITY ANALYSIS

In the previous section, the stability of the subsystems (10), (22), and (24) are analyzed. In this section, the stability of the cascade system consisted of kinematic guidance subsystem (10) and kinetic control subsystem (22) is stated in the following theorem.

*Theorem 1:* Consider the underactuated USV described by (1) and (2), the event-triggered network-based path-tracking guidance law in (9), the ESOs in (19), the kinetic control law in (22), and the event-triggered conditions in (11) and (27). Under Assumptions 1 and 2, the closed-loop system is ISS and all error signals in the closed-loop network system are bounded.

*Proof:* Note that

$$\begin{cases} q_U = \vartheta_U - \tilde{U} \leq |\vartheta_U| + |\tilde{U}| \\ q_r = \vartheta_r - \tilde{r} \leq |\vartheta_r| + |\tilde{r}|. \end{cases} \quad (40)$$

We have

$$q = z_3 - \tilde{v} \leq \|z_3\| + \|\tilde{v}\| \leq \|z_2\| + \|z_3\|. \quad (41)$$

Note also that

$$\begin{aligned} U &= q_U + \alpha_U + \epsilon_U \\ &\leq |\vartheta_U| + |\tilde{U}| + |\alpha_U| + |\epsilon_U| \\ &\leq \|z_2\| + \|z_3\| + |\alpha_U| + |\epsilon_U|. \end{aligned} \quad (42)$$

We have

$$\|z_1\| \leq \max \left\{ \|z_1(t_0)\| e^{-c_1(1-\kappa_1)(t-t_0)}, \frac{\|\alpha_e\| + |\beta_d|}{c_1 \kappa_1} + \frac{\lambda_{\max}(K)\|z_e\| + |\alpha_U| + |\epsilon_U| + 3(\|z_2\| + \|z_3\|)}{c_1 \kappa_1} \right\}. \quad (43)$$

According to [47] and Lemmas 1–3, the closed-loop system is ISS. As  $t \rightarrow \infty$ , it follows that:

$$\begin{cases} \|z_1\| \leq 6(\|P_1 B_1\| \epsilon_v^* + \|P_1 B_2\| \epsilon_\tau^* + \|P_1 B_3\| \dot{\sigma}^*) \\ \quad \times \frac{(k_e^u + k_e^r + c_3 \kappa_3) \sqrt{\lambda_{\max}(P_1)}}{c_1 c_2 c_3 \kappa_1 \kappa_2 \kappa_3 \sqrt{\lambda_{\min}(P_1)}} \\ \quad + \frac{3(\lambda_{\max}(K_e^v) \epsilon_v^* + \lambda_{\max}(K_m) \epsilon_\tau^*)}{c_1 c_3 \kappa_1 \kappa_3} \\ \quad + \frac{\lambda_{\max}(K) \epsilon_z^* + 2\epsilon_\alpha^* + k_1 + u_0 U_d^* + \beta_d^*}{c_1 \kappa_1} \\ \|z_2\| \leq \frac{2(\|P_1 B_1\| \epsilon_v^* + \|P_1 B_2\| \epsilon_\tau^*) \sqrt{\lambda_{\max}(P_1)}}{c_2 \kappa_2 \sqrt{\lambda_{\min}(P_1)}} \\ \quad + \frac{2\|P_1 B_3\| \sigma^* \sqrt{\lambda_{\max}(P_1)}}{c_2 \kappa_2 \sqrt{\lambda_{\min}(P_1)}} \\ \|z_3\| \leq 2(\|P_1 B_1\| \epsilon_v^* + \|P_1 B_2\| \epsilon_\tau^* + \|P_1 B_3\| \dot{\sigma}^*) \\ \quad \times \frac{(k_e^u + k_e^r) \sqrt{\lambda_{\max}(P_1)}}{c_2 c_3 \kappa_2 \kappa_3 \sqrt{\lambda_{\min}(P_1)}} + \frac{\lambda_{\max}(K_e^v)}{c_3 \kappa_3} \epsilon_v^* \\ \quad + \frac{\lambda_{\max}(K_m)}{c_3 \kappa_3} \epsilon_\tau^*. \end{cases} \quad (44)$$

As a result, all error signals in the cascade system are uniformly ultimately bounded.  $\blacksquare$

*Remark 1:* It is worth noting that the sampling virtual guidance law  $\dot{\alpha}_s = 0$  over the period  $[t_j^\alpha, t_{j+1}^\alpha)$  such that the repeated differentiation of the virtual control law is not required. ZOHs are used to hold the sampling data and thus there is a jump on the sampling moment  $t_s$ . The difference of the sampled data on  $t_s^-$  and  $t_s^+$  is a bounded measurement such that the jump bounce exists. As a result, the jump produced by the ZOHs can be considered as a bounded perturbation.

*Remark 2:* By Theorem 1, the tracking performance of the closed-loop system relates to the event-triggered thresholds. Upper bounds of these errors are determined by the corresponding triggered thresholds. Hence, there is a trade-off between the desired performance and the triggering thresholds.

*Theorem 2:* Under the event-triggered mechanism (11) and (27), there exist positive constants  $\varrho_\eta$ ,  $\varrho_\alpha$ ,  $\varrho_v$ , and  $\varrho_\tau$  such that the communication intervals  $t_{i+1}^\eta - t_i^\eta \geq \varrho_\eta$ ,  $t_{j+1}^\alpha - t_j^\alpha \geq \varrho_\alpha$ ,  $t_{m+1}^v - t_m^v \geq \varrho_v$ , and  $t_{n+1}^\tau - t_n^\tau \geq \varrho_\tau$ , and Zeno behaviors are excluded.

*Proof:* The time derivation of the triggering condition  $\Omega_1(1)$  with  $t \in [t_i^\eta, t_{i+1}^\eta)$ ,  $\Omega_1(2)$  with  $t \in [t_j^\alpha, t_{j+1}^\alpha)$ ,  $\Omega_2(1)$  with  $t \in$



$[t_m^\nu, t_{m+1}^\nu)$  and  $\Omega_2(2)$  with  $t \in [t_n^\tau, t_{n+1}^\tau)$  are as follows:

$$\begin{cases} \|\dot{\eta}_e\| = \|\dot{\eta}_s(t) - \dot{\eta}(t)\| \\ \leq \|\text{col}(U\cos\psi, U\sin\psi, r + \beta_d)\| \\ \|\dot{\alpha}_e\| = \|\dot{\alpha}_s(t) - \dot{\alpha}(t)\| \\ \leq \|\text{col}(u_0\dot{U}_d, |k_3\dot{\psi}_e|(\Delta_3^2 - \psi_e^2) + \dot{r}_d)\| \\ \|\dot{v}_e\| = \|\dot{v}_s(t) - \dot{v}(t)\| \\ \leq \|\sigma\| + \lambda_{\max}(K_m)\|\tau\| \\ \|\dot{\tau}_e\| = \|\dot{\tau}_s(t) - \dot{\tau}(t)\| \\ \leq \lambda_{\max}(K_m)\|\dot{\sigma}\|. \end{cases} \quad (45)$$

Under Assumptions 1, 2, and by Theorem 1,  $U$ ,  $\dot{U}_d$ ,  $r$ ,  $\dot{r}_d$ ,  $\beta_d$ ,  $\psi_e$ ,  $\sigma$ ,  $\tau$ , and  $\dot{\sigma}$  are bounded. Therefore,  $\dot{\eta}_e$ ,  $\dot{\alpha}_e$ ,  $\dot{v}_e$ , and  $\dot{\tau}_e$  all have upper bounds denoting  $\bar{h}_\eta$ ,  $\bar{h}_\alpha$ ,  $\bar{h}_v$ , and  $\bar{h}_\tau$ , respectively. The initial conditions are satisfied as follows:

$$\begin{cases} \lim_{t \rightarrow t_i^{\eta+}} \|\eta_s(t) - \eta(t)\| = 0 \\ \lim_{t \rightarrow t_j^{\alpha+}} \|\alpha_s(t) - \alpha(t)\| = 0 \\ \lim_{t \rightarrow t_m^{\nu+}} \|v_s(t) - v(t)\| = 0 \\ \lim_{t \rightarrow t_n^{\tau+}} \|\tau_s(t) - \tau(t)\| = 0. \end{cases} \quad (46)$$

The errors can be rendered as

$$\begin{cases} \|\eta_s(t) - \eta(t)\| \leq \bar{h}_\eta(t - t_i^\eta), & t \in [t_i^\eta, t_{i+1}^\eta) \\ \|\alpha_s(t) - \alpha(t)\| \leq \bar{h}_\alpha(t - t_j^\alpha), & t \in [t_j^\alpha, t_{j+1}^\alpha) \\ \|v_s(t) - v(t)\| \leq \bar{h}_v(t - t_m^\nu), & t \in [t_m^\nu, t_{m+1}^\nu) \\ \|\tau_s(t) - \tau(t)\| \leq \bar{h}_\tau(t - t_n^\tau), & t \in [t_n^\tau, t_{n+1}^\tau). \end{cases} \quad (47)$$

When the triggering conditions  $\Omega_1(1)$ ,  $\Omega_1(2)$ ,  $\Omega_2(1)$ , and  $\Omega_2(2)$  are not satisfied, we have

$$\begin{cases} \lim_{t \rightarrow t_{i+1}^{\eta-}} \|\eta_s(t) - \eta(t)\| = \epsilon_\eta^* \\ \lim_{t \rightarrow t_{j+1}^{\alpha-}} \|\alpha_s(t) - \alpha(t)\| = \epsilon_\alpha^* \\ \lim_{t \rightarrow t_{m+1}^{\nu-}} \|v_s(t) - v(t)\| = \epsilon_v^* \\ \lim_{t \rightarrow t_{n+1}^{\tau-}} \|\tau_s(t) - \tau(t)\| = \epsilon_\tau^*. \end{cases} \quad (48)$$

As a result, it follows that:

$$\begin{cases} t_{i+1}^\eta - t_i^\eta \geq \varrho_\eta \\ t_{j+1}^\alpha - t_j^\alpha \geq \varrho_\alpha \\ t_{m+1}^\nu - t_m^\nu \geq \varrho_\nu \\ t_{n+1}^\tau - t_n^\tau \geq \varrho_\tau \end{cases} \quad (49)$$

where  $\varrho_\eta = \epsilon_\eta^*/\bar{h}_\eta$ ,  $\varrho_\alpha = \epsilon_\alpha^*/\bar{h}_\alpha$ ,  $\varrho_\nu = \epsilon_v^*/\bar{h}_v$ , and  $\varrho_\tau = \epsilon_\tau^*/\bar{h}_\tau$ . This means that there always exists a positive low upper bound of minimum interevent interval time during the period of communication and actuation. As a result, the Zeno phenomenon will not occur. ■

## V. EXPERIMENTAL RESULTS

In this section, experimental results are provided to illustrate the efficacy of the proposed event-triggered network-based path-tracking method for a USV. A small type of USV called CSICET-DH01 is designed with a motion control system, including micro control unit, a motion coprocessor, a global navigation satellite system receiver, an attitude sensor, and a ZigBee communication network. The entire experiment platform can be seen in Fig. 3. Experiments are conducted at the harbor basin of Dalian Maritime University at the Lingshui

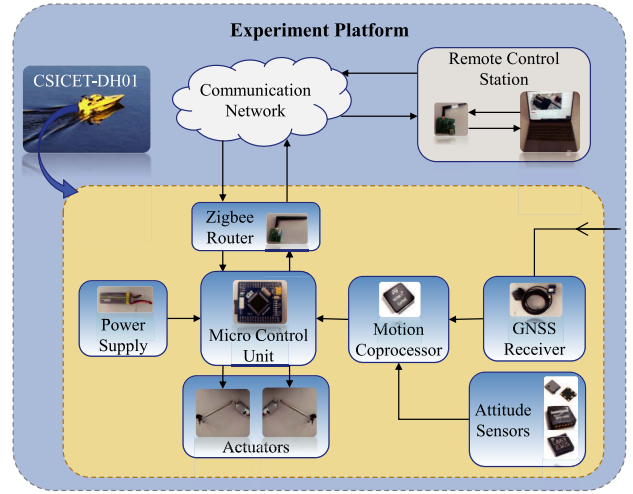


Fig. 3. Experiment platform of the CSICET-DH01 USV.

Port. In experiments, the USV is required to track a series of straight-line paths. The parametrized paths are set to

$$\begin{cases} p_{d1} = \text{col}(21.6 - 0.4856\theta, 1302.5 + 0.8742\theta) \\ p_{d2} = \text{col}(-22.1 - 0.8746(\theta - 90), 1381.2 \\ \quad - 0.4848(\theta - 90)) \\ p_{d3} = \text{col}(-100.8 + 0.4856(\theta - 180), 1337.63 \\ \quad - 0.8742(\theta - 180)) \\ p_{d4} = \text{col}(-61.7 + 0.8746(\theta - 260), 1267.7 \\ \quad + 0.4848(\theta - 580)) \\ p_{d5} = \text{col}(8.3 - 0.4856(\theta - 340), 1306.5 \\ \quad + 0.8742(\theta - 340)) \\ p_{d6} = \text{col}(-25.7 - 0.8746(\theta - 410), 1367.7 \\ \quad - 0.4848(\theta - 410)) \\ p_{d7} = \text{col}(-86.9 + 0.4856(\theta - 480), 1333.8 \\ \quad - 0.8742(\theta - 480)) \\ p_{d8} = \text{col}(-57.8 + 0.8746(\theta - 540), 1281.3 \\ \quad + 0.4848(\theta - 540)) \\ p_{d9} = \text{col}(-5.3 - 0.4856(\theta - 600), 1310.4 \\ \quad + 0.8742(\theta - 600)) \\ p_{d10} = \text{col}(-29.6 - 0.8746(\theta - 650), 1354.1 \\ \quad - 0.4848(\theta - 650)) \\ p_{d11} = \text{col}(-73 + 0.4856(\theta - 700), 1329.9 \\ \quad - 0.8742(\theta - 700)) \\ p_{d12} = \text{col}(-53.6 + 0.8746(\theta - 740), 1295 \\ \quad + 0.4848(\theta - 740)) \\ p_{d13} = \text{col}(-18.6 - 0.4856(\theta - 780), 1314.4 \\ \quad + 0.8742(\theta - 780)) \\ p_{d14} = \text{col}(33.2 - 0.8742(\theta - 800), 1340.5 \\ \quad - 0.4848(\theta - 800)) \\ p_{d15} = \text{col}(-59.4 + 0.4856(\theta - 830), 1325.8 \\ \quad - 0.8742(\theta - 830)). \end{cases} \quad (50)$$

The kinematic guidance law (9) is implemented in the remote control center and the kinetic control law (21) is implemented in the micro control unit in the USV. The control parameters are selected as:  $k_1 = 0.2$ ,  $k_2 = 1$ ,  $\Delta_1 = \Delta_3 = 1$ ,  $\Delta_2 = 2$ ,  $u_0 = 0.5$  m/s,  $k_1^u = 1$ ,  $k_2^u = 0.25$ ,  $k_1^r = 0.1$ ,  $k_2^r = 0.3$ ,  $k_u = 2$ ,  $k_r = -0.5$ ,  $m_{11} = m_{33} = 1$ ,  $\epsilon_\alpha^* = 0.4372$ ,  $\epsilon_\tau^* = 0.1414$ ,  $\epsilon_v^* = 0.03333$ .



Fig. 4. Tracking performance of the CSICET-DH01 USV in the experiment.

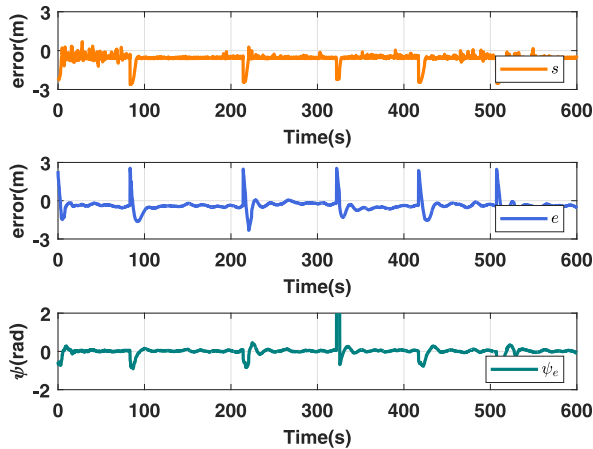


Fig. 5. Kinematic tracking errors in the experiment.

The experiments results are shown in Figs. 4–11. Fig. 4 depicts the trajectories of the USV and it can be seen that the USV is able to follow the path by using the proposed event-triggered network-based guidance law and control law with aperiodic communications and actuation. Fig. 5 plots the along-track, cross-track and yaw errors and it shows that they converge to a neighborhood of the origin. According to (4), the tracking errors suddenly increase when switching paths. As the positioning accuracy of the adopted GPS device is about 1.5 m, these tracking errors are acceptable in the sea environment. Fig. 6 shows the tracking performance of speed and yaw rate by the proposed event-triggered anti-disturbance control law. It can be observed that they are able to follow the commanded signals from the remote control law over wireless network. Fig. 7 plots the surge force  $\tau_u$  and control moment  $\tau_r$ , and they all bounded. The estimation performance of the ESOs are shown in Fig. 8. During transitions between different paths, the USV has to speed up and the total disturbance will increase. Note that the estimated disturbance of ESOs increase accordingly. This implies that the ESOs play their roles. Fig. 9 presents the event-triggered state of commanded speed and yaw rate from the remote control center. The event-triggered states of control inputs, actual total speed and yaw rate can be observed from Figs. 10 and 11. The triggering numbers in the steady phase are less than those in the transient phase. In Table I, the triggering times of  $\alpha_U$ ,  $\alpha_r$ ,  $\tau_u$ ,  $\tau_r$ ,

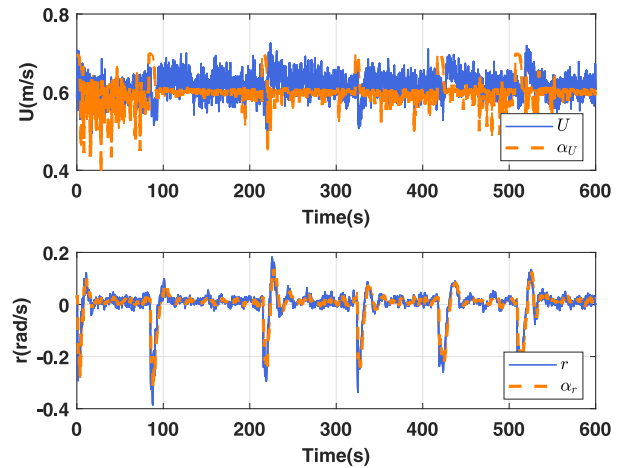


Fig. 6. Tracking performance of total speed and angular speed in the experiment.

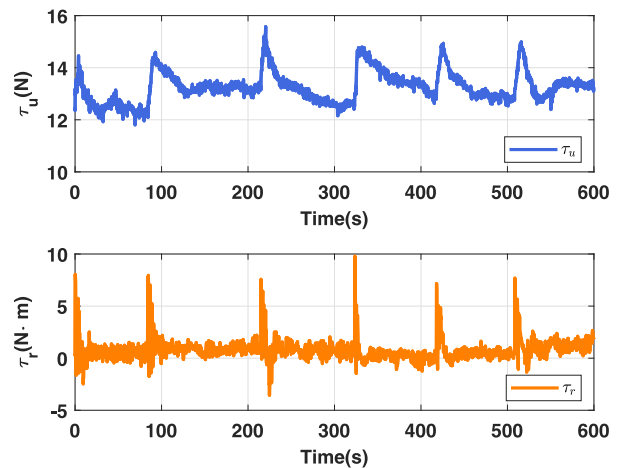


Fig. 7. Control inputs of the USV in the experiment.

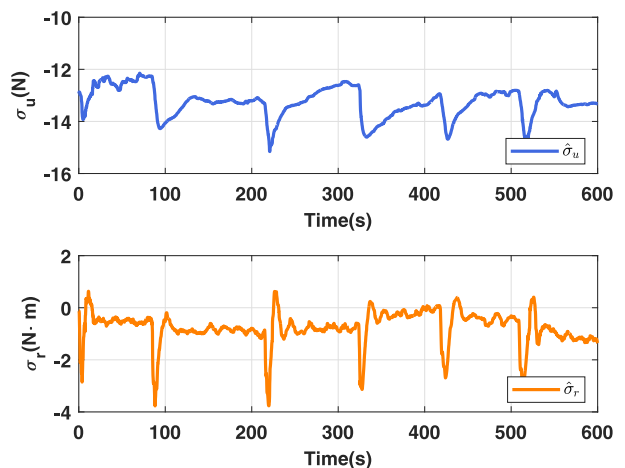


Fig. 8. ESO estimation of uncertainties in the experiment.

$U$ , and  $r$  and their percentages with respect to periodic triggering times are listed. It is shown that the highest ratio is 21.7%, which implies that the proposed method can reduce the communication and actuation burden.



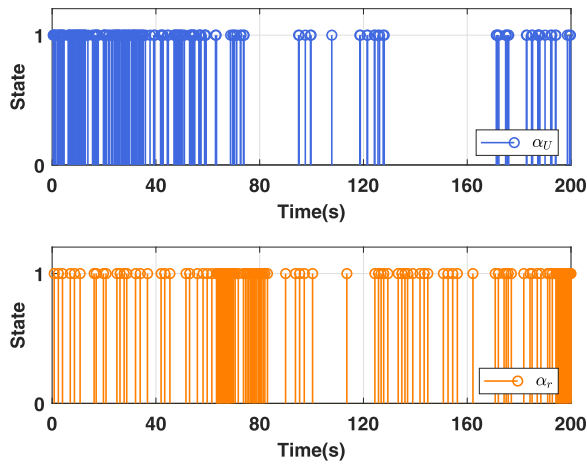


Fig. 9. Event-triggered state of desired commands in the experiment.

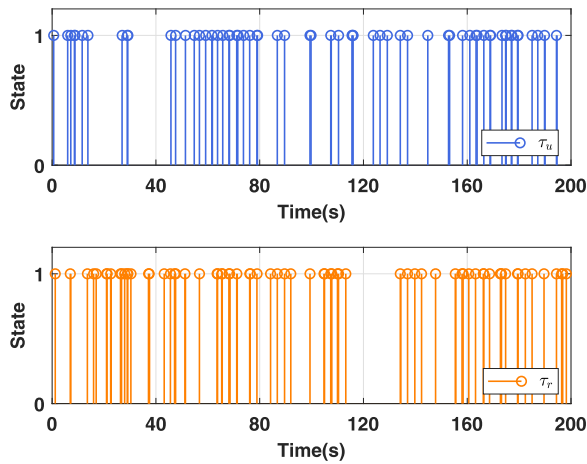


Fig. 10. Event-triggered state of control inputs in the experiment.

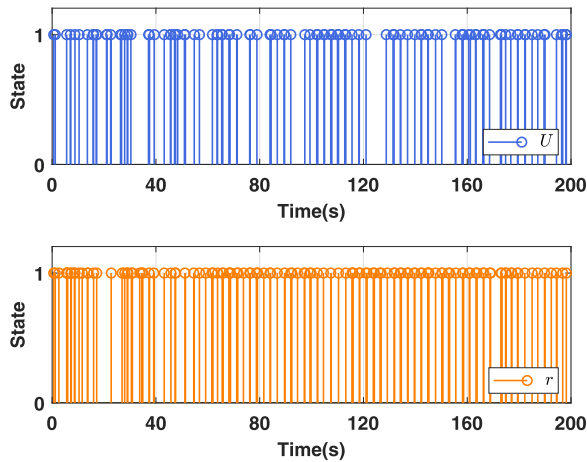


Fig. 11. Event-triggered state of actual total speed and angular speed in the experiment.

## VI. CONCLUSION

A network-based path-tracking control method is proposed for a USV over the wireless network. The USV is subject to model uncertainties and unknown environmental disturbances, including wind, wave, and current, and a two-level

TABLE I  
TRIGGERING TIMES AND THEIR PERCENTAGES WITH RESPECT TO THE PERIODIC COMMUNICATIONS OR ACTUATIONS

Variables	Triggering Times	Percentage
$\alpha_U$	217	21.7%
$\alpha_r$	142	14.2%
$\tau_u$	67	6.7%
$\tau_r$	80	8.0%
$U$	120	12.0%
$r$	143	14.3%

Times of periodic communications or actuation: 1000

network-based control architecture is presented, including a local inner loop and a remote outer loop. In the remote outer loop, an event-triggered guidance law based on the LOS principle is developed to track the predefined path and reduce the communication burden for the remote guidance side. The closed-loop guidance subsystem is proven to be ISS. In the local inner loop, two second-order ESOs are developed to estimate the uncertainties and unknown disturbances. An event-triggered control law based on the total estimated disturbances are presented for achieving the speed and angular speed tracking, which can also reduce actuation burden of the local inner loop. Then, the stability of the ESO subsystem and the kinetic subsystem are ISS by the input-to-state stability theory. The stability of the network-based closed-loop system is proved to be ISS by input-to-state and cascade stability theory. Finally, the proposed method is verified on an experiment platform of CSICET-DH01 USV. The experimental results demonstrated the effectiveness of the event-triggered network-based path-tracking control method.

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**Wentao Wu** received the B.E. degree in electrical engineering and automation from the Harbin University of Science and Technology, Harbin, China, in 2018. He is currently pursuing the M.E. degree in electrical engineering with Dalian Maritime University, Dalian, China.

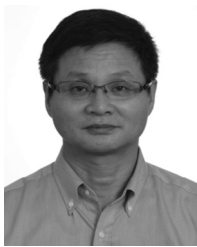
He conducted academic exchanges in the Science and Technology on Underwater Vehicle Technology, Harbin Engineering University, Harbin, in 2020. His current research interest is guidance and control of unmanned surface vehicles.



**Zhouhua Peng** (Senior Member, IEEE) received the B.E. degree in electrical engineering and automation, the M.E. degree in power electronics and power drives, and the Ph.D. degree in control theory and control engineering from Dalian Maritime University, Dalian, China, in 2005, 2008, and 2011, respectively.

In 2011, he joined the School of Marine Engineering, Dalian Maritime University, where he is currently a Professor with the School of Marine Electrical Engineering. From 2014 to 2018, he was a Postdoctoral Research Fellow with the School of Control Science and Engineering, Dalian University of Technology, Dalian. From 2016 to 2018, he was a Hong Kong Scholar with the Department of Computer Science, City University of Hong Kong, Hong Kong, where he was a Senior Research Fellow with the Department of Computer Science, in 2019. He has authored over 190 refereed publications. His research interest is cooperative control of multiple autonomous surface vehicles.

Prof. Peng was a recipient of the Science and Technology Award (First Class) from China Association of Oceanic Engineering in 2019, the Natural Science Awards (Second Class) from Liaoning Province in 2013 and 2017, the Hong Kong Scholar Award in 2016, and the Science and Technology Award for Youth from China Institute of Navigation in 2017. He won the Honor of the Distinguished Worker in the National Ministry of Transport of China in 2020, the Science and Innovation Leadership for Young and Middle-aged Scientists of the Ministry of Transport of the People's Republic of China in 2020, the "Bai-Qian-Wan" Talent (level Bai) from Liaoning Province in 2019, the Young Talent in Science and Technology from the Ministry of Transport of the People's Republic of China in 2017, and the Distinguished Young Talent in Science and Technology from Dalian in 2018.



**Dan Wang** (Senior Member, IEEE) received the B.E. degree in industrial automation engineering from the Dalian University of Technology, Dalian, China, in 1982, the M.E. degree in marine automation engineering from Dalian Maritime University, Dalian, in 1987, and the Ph.D. degree in mechanical and automation engineering from The Chinese University of Hong Kong, Hong Kong, in 2001.

He is a Professor with the School of Marine Electrical Engineering, Dalian Maritime University. From 2001 to 2005, he was a Research Scientist with Temasek Laboratories, National University of Singapore, Singapore. In 2012, he was a Visiting Professor with the Institute for Aerospace Studies, University of Toronto, Toronto, ON, Canada. He has authored over 220 refereed publications. His research interests include nonlinear control theory and its applications.



**Lu Liu** (Member, IEEE) received the B.E. degree in electrical engineering and automation and the Ph.D. degree in marine electrical engineering from Dalian Maritime University, Dalian, China, in 2012 and 2018, respectively.

In 2018, she joined the School of Marine Engineering, Dalian Maritime University, where she is currently an Associate Professor with the School of Marine Electrical Engineering and a Postdoctoral Research Fellow with the School of Electrical Information and Electric Engineering, Shanghai Jiaotong University, Shanghai, China. She has authored over 40 refereed publications. Her research interests include guidance and control of single/multiple marine surface vehicles.



**Qing-Long Han** (Fellow, IEEE) received the B.Sc. degree in mathematics from Shandong Normal University, Jinan, China, in 1983, and the M.Sc. and Ph.D. degrees in control engineering from the East China University of Science and Technology, Shanghai, China, in 1992 and 1997, respectively.

He is a Pro Vice-Chancellor (Research Quality) and a Distinguished Professor with Swinburne University of Technology, Melbourne, VIC, Australia. He held various academic and management positions with Griffith University, Gold Coast, QLD, Australia, and Central Queensland University, Rockhampton, QLD. His research interests include networked control systems, multi-agent systems, time-delay systems, smart grids, unmanned surface vehicles, and neural networks.

Dr. Han was the recipient of the 2021 M. A. Sargent Medal (the Highest Award of the Electrical College Board of Engineers Australia), the 2020 IEEE Systems, Man, and Cybernetics (SMC) Society Andrew P. Sage Best Transactions Paper Award, the 2020 IEEE Transactions on Industrial Informatics Outstanding Paper Award, and the 2019 IEEE SMC Society Andrew P. Sage Best Transactions Paper Award. He was a Highly Cited Researcher in Engineering and Computer Science (Clarivate Analytics, 2019–2020). He is a Co-Editor of *Australian Journal of Electrical and Electronic Engineering*, an Associate Editor for 12 international journals, including the IEEE TRANSACTIONS ON CYBERNETICS, the IEEE TRANSACTIONS ON INDUSTRIAL INFORMATICS, *IEEE Industrial Electronics Magazine*, the IEEE/CAA JOURNAL OF AUTOMATICA SINICA, *Control Engineering Practice*, and *Information Sciences*, and a guest editor for 13 special issues. He has served as an AdCom Member of IEEE Industrial Electronics Society (IES), a Member of IEEE IES Fellow Committee, and the Chair of IEEE IES Technical Committee on Networked Control Systems. He was one of Australia's Top Five Lifetime Achievers (Research Superstars) in Engineering and Computer Science (The Australian's 2020 Research Magazine). He is a Fellow of The Institution of Engineers Australia.