

# Data acquisition and processing

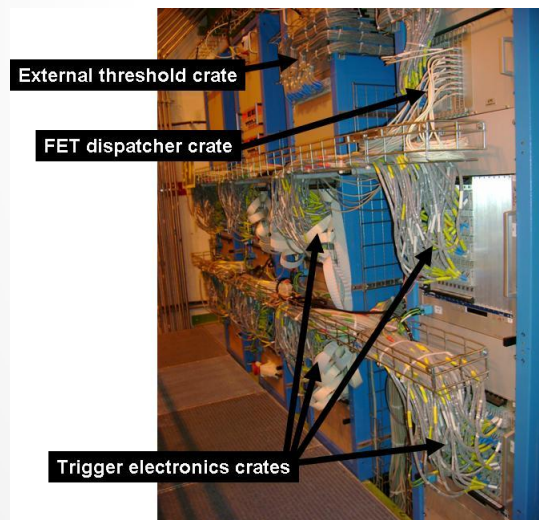
**Erik Schooneveld**

# Content

- Introduction
- Analogue approach
- Digital approach
- Digital signal processing

# Introduction

- ✓ Discuss approach for processing signals from detector for detector R&D
- ✗ Analysis of “sample related properties”, like d-spacing or  $S(Q, \omega)$  type data analysis



- ✗ “Back-end” electronics / computer interface
- ✗ Varies too much from institute to institute

# Introduction

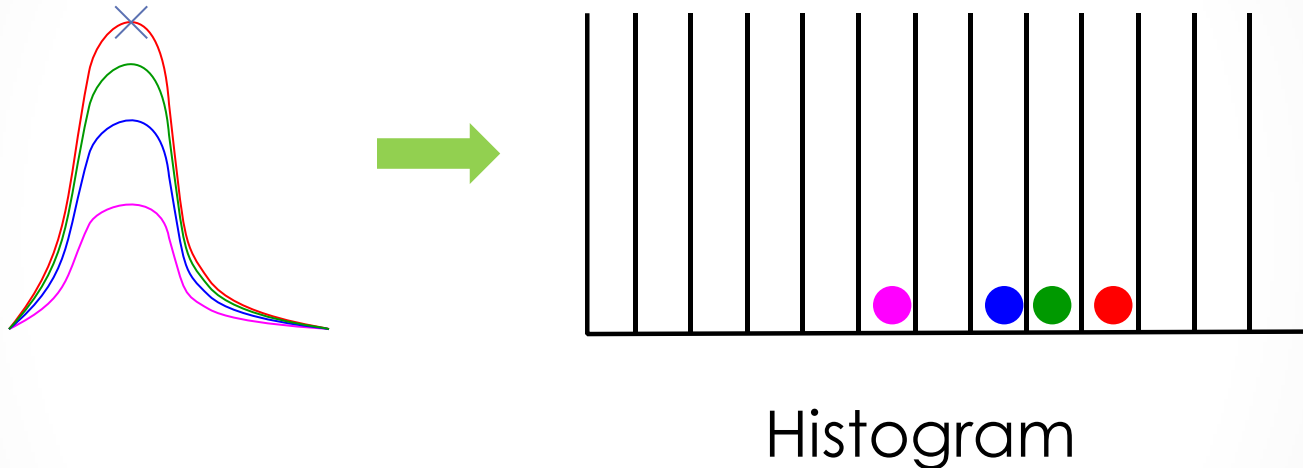
- Which properties of the detector do we want to measure during detector R&D?
  - 1) Neutron detection efficiency
  - 2) Gamma sensitivity
  - 3) Position resolution
  - 4) Count-rate capability (dead time)
  - 5) Stability
  - 6) Timing resolution?

# Introduction

- How do we measure these properties?
  - 1) Pulse height spectrum
  - 2) Counter/scaler
  - 3) ?Time to Digital Converter (TDC)
  - 4) Digitiser/Oscilloscope

# Introduction

- What is a pulse height spectrum?

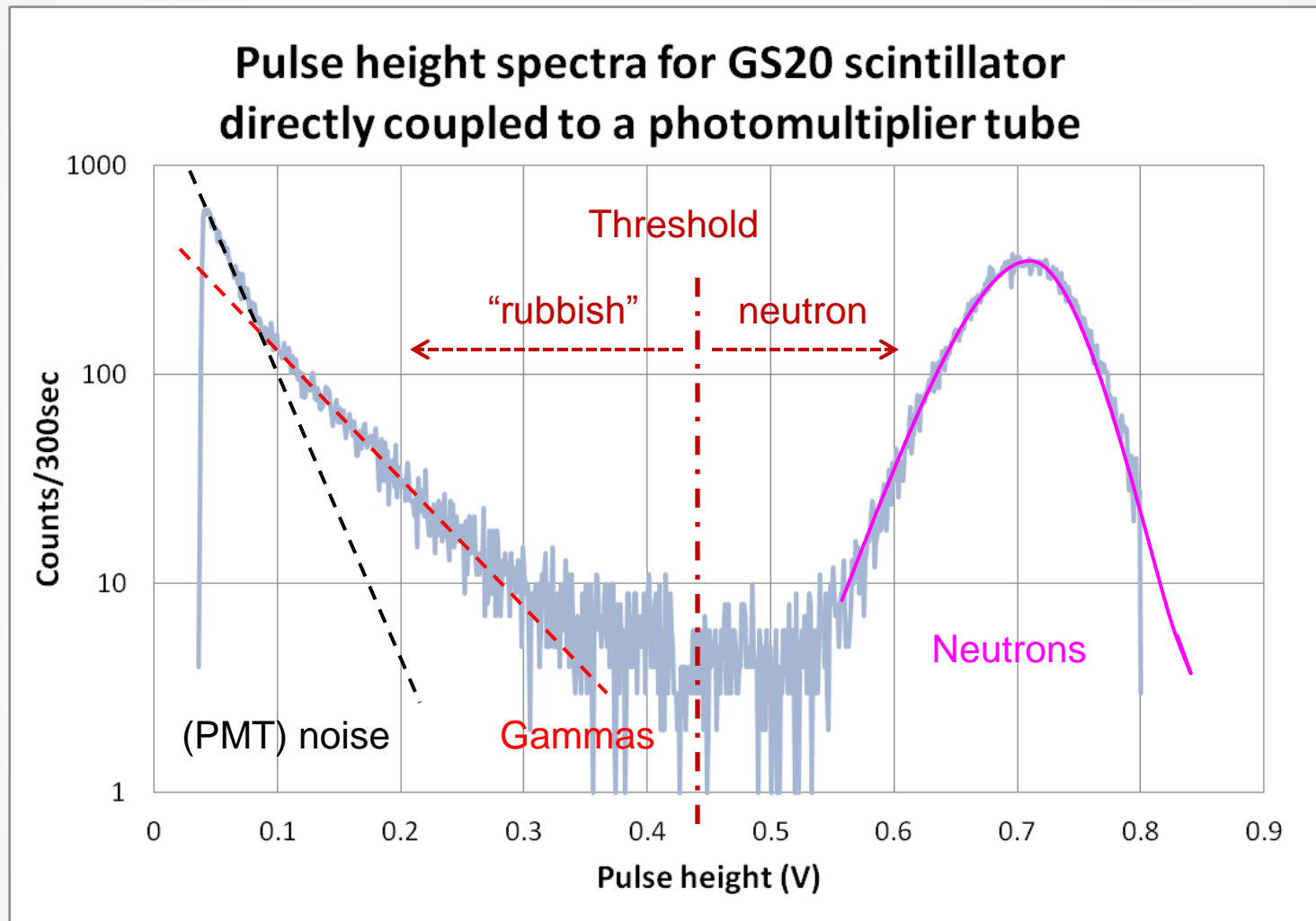


- Histogram of the distribution of signal amplitudes

# Introduction

- Neutrons usually produce a peak in a pulse height spectrum
- Noise, gammas and cosmic radiation usually produce an exponentially decaying tail.
- Pulse height spectrum tells how easy it is to discriminate neutron from unwanted background ⇒ pulse height analysis one of most important tools.

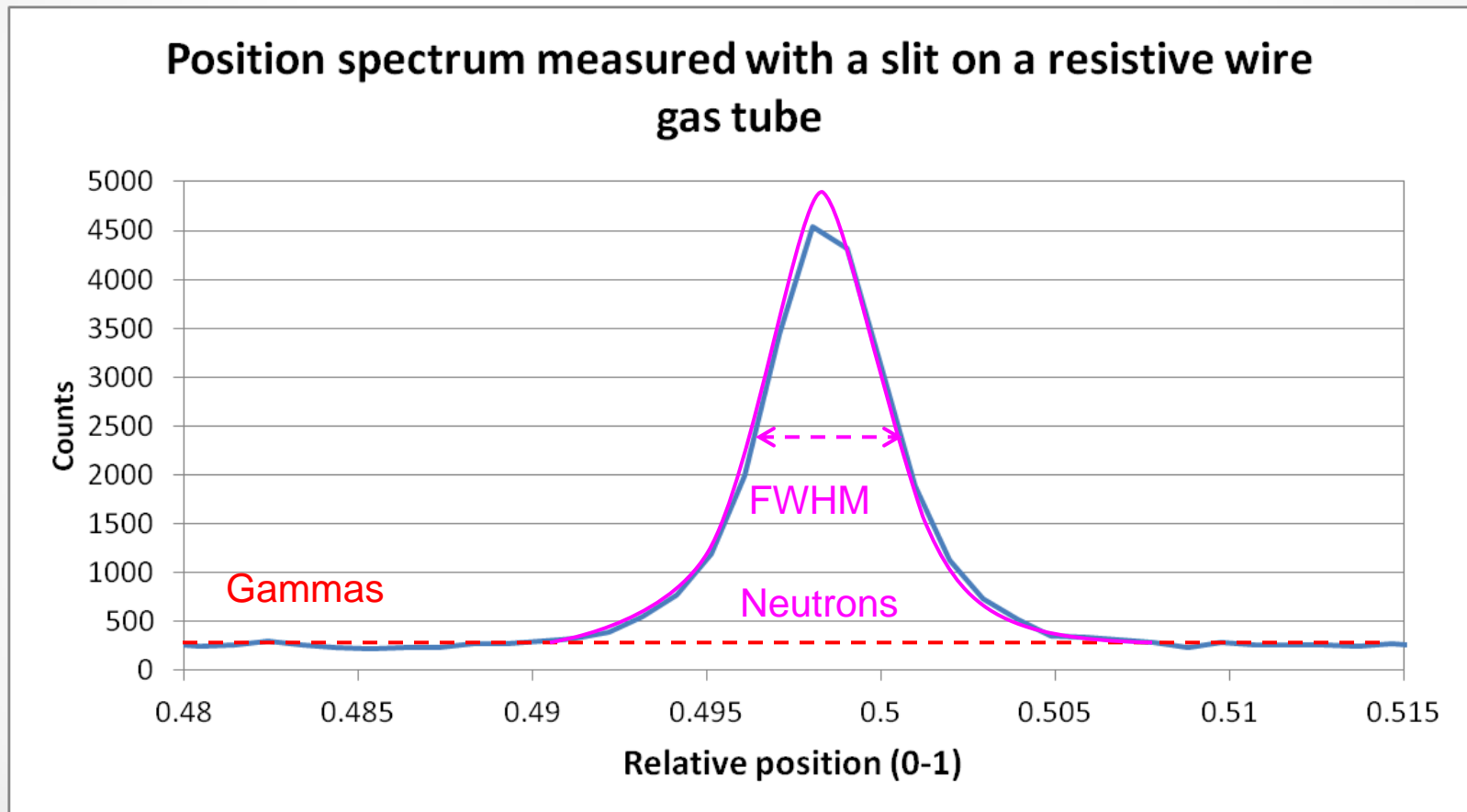
# Introduction





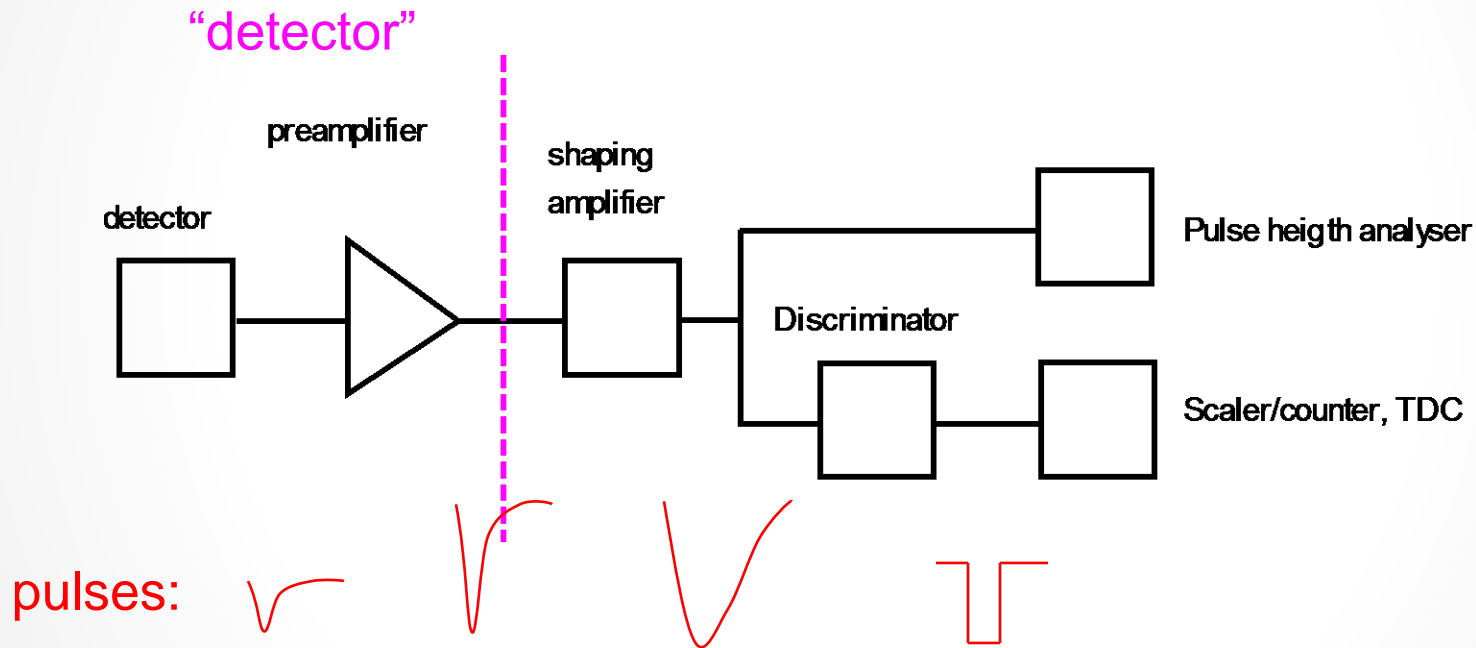
# Introduction

- Position spectrum: tells where the neutrons were detected



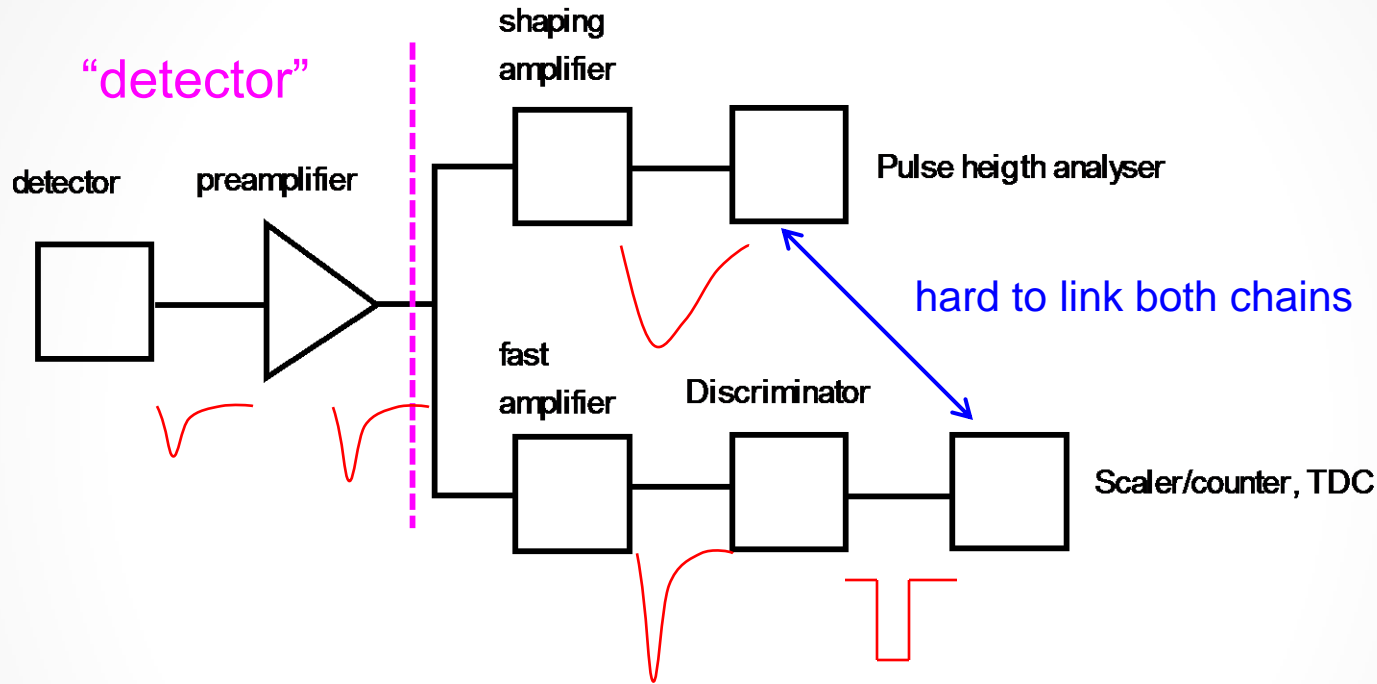
# Analogue approach

- Typical analogue data acquisition chain (text book):



# Analogue approach

- For fast timing some people use this circuit:

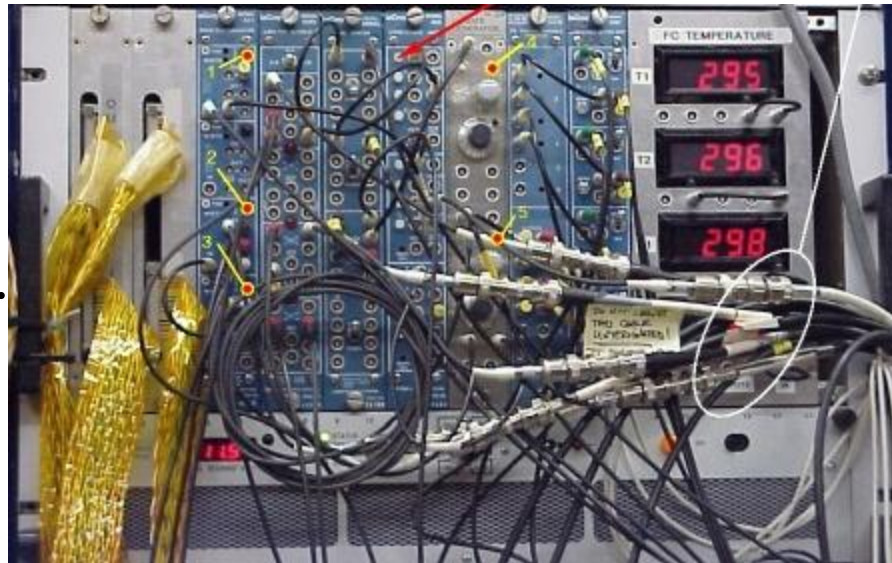


- (Personally) I don't like it because the pulse height spectrum doesn't tell you anything about how to set the threshold for the discriminator

# Analogue approach

- ✗ Different type of measurements require different types of electronics  $\Rightarrow$  have to buy a lot of equipment

✗ and .....

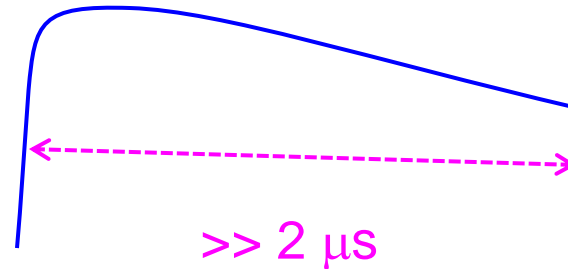


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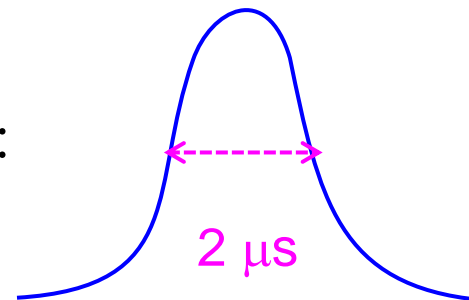
# Analogue approach

- Pulse height analysers, a.k.a. Multi-Channel Analysers (MCAs)
- Realize that pulse height analysers are mainly used for spectroscopy applications  $\Rightarrow$  dealing with slow signals:

pre-amplifier signal:

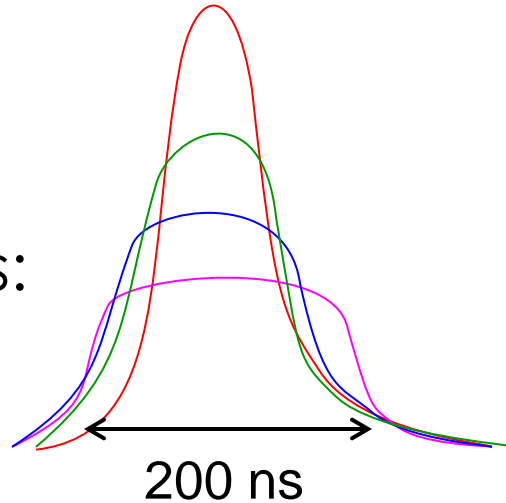


shaping amplifier signal:



# Analogue approach

- Can they handle this:

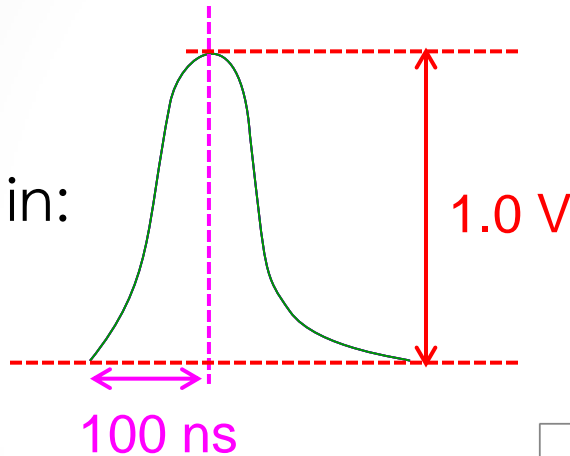


- Answer: not very well

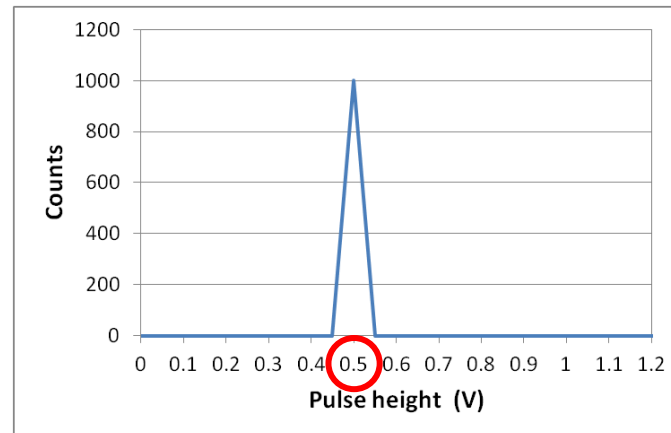
# Analogue approach

- Peaking time: nearly all MCAs require peaking time  $\geq 250\text{ns}$

- MCA in:



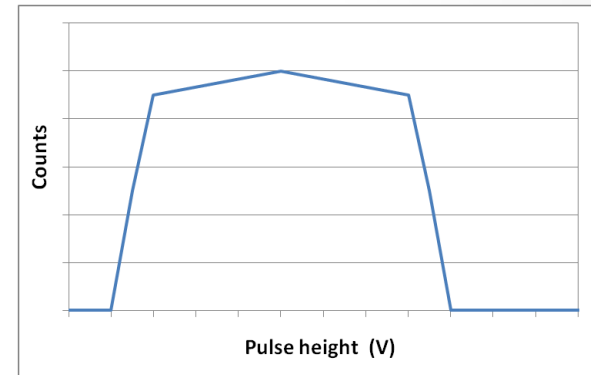
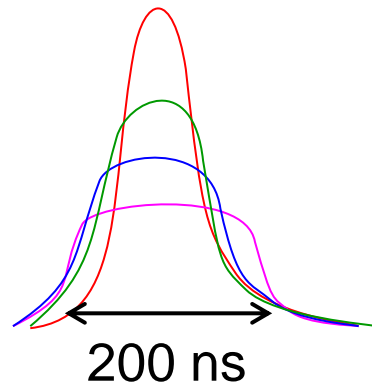
- What will happen?
- Pulse height spectrum:



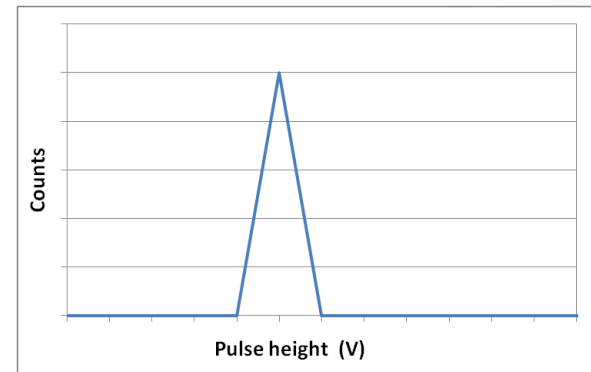
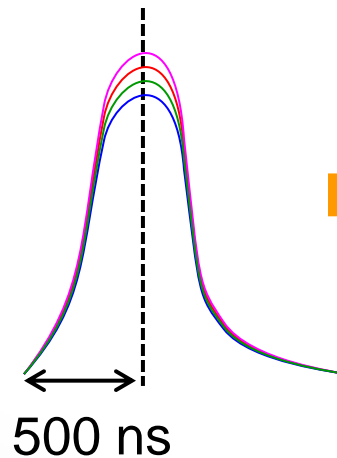
# Analogue approach

- Let's slow down signal so that peaking time  $> 250\text{ns}$

- Shaping amplifier in:

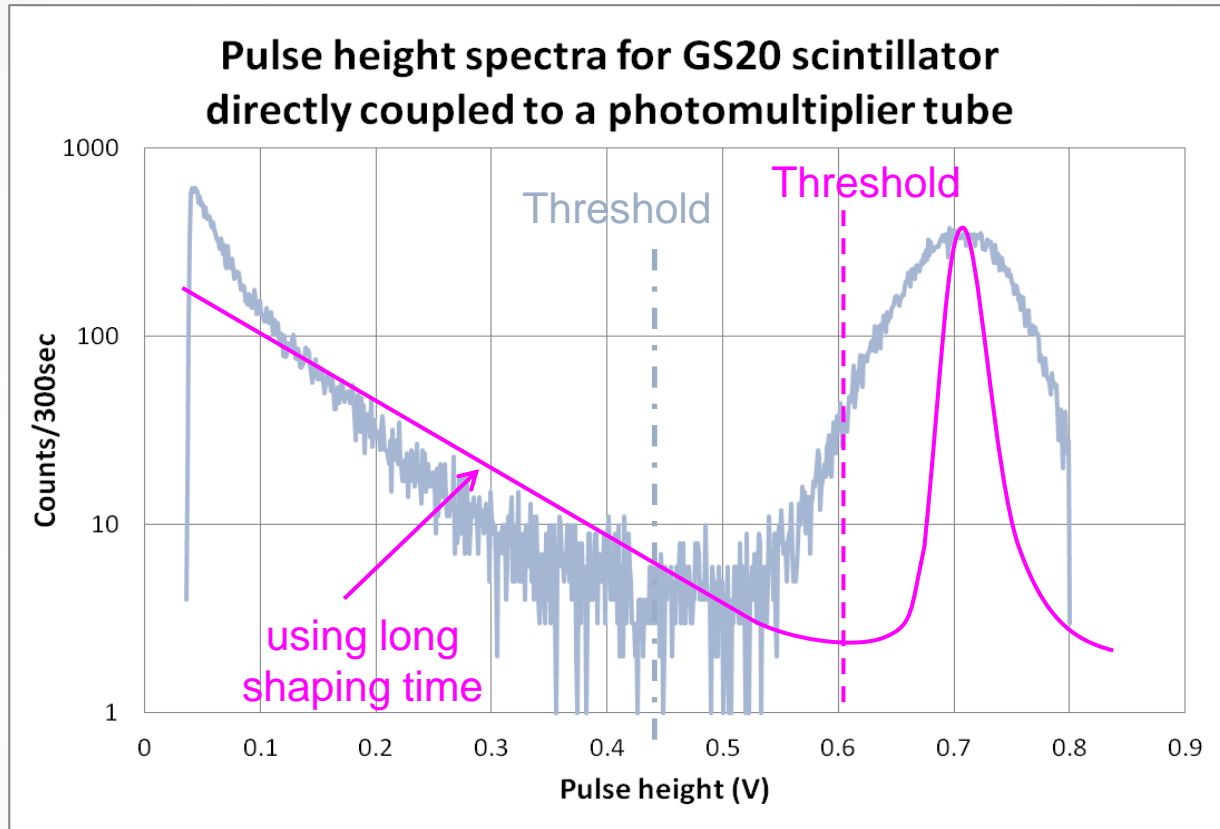


- Shaping amplifier out:





# Introduction

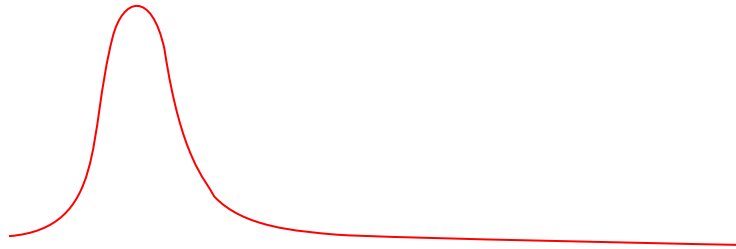



- Much less stable with incorrect threshold

# Analogue approach

- Have you ever wanted to veto the MCA processing an analogue signal like this?

- MCA in:

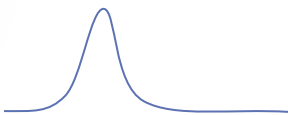


- Veto signal: 

- Veto signal is too late.
- Very hard to delay fast analogue signals for microseconds  $\Rightarrow$  cannot veto MCA.

# Analogue approach

- Other things to keep in mind when using MCAs:
- Most MCAs can handle only positive signals

OK: 

not: 

- MCAs usually have  $\sim 1\text{ k}\Omega$  input impedance  $\Rightarrow$  need  $50\ \Omega$  termination resistor at input.

# Analogue approach

- My opinion:

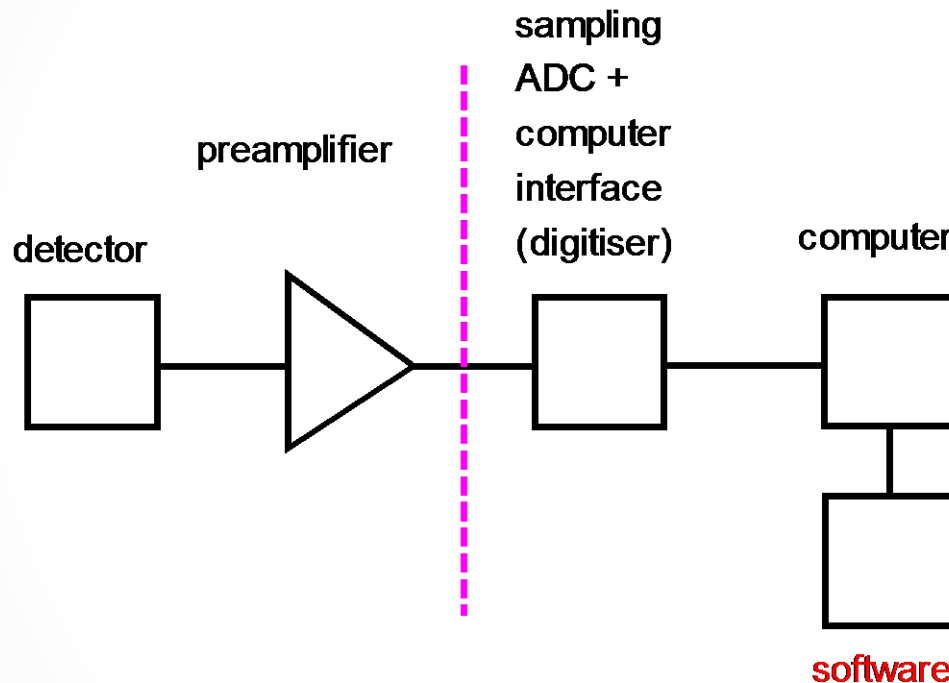
- Generation:



(for neutron detector development)

# Digital approach

- Typical digital data acquisition chain:

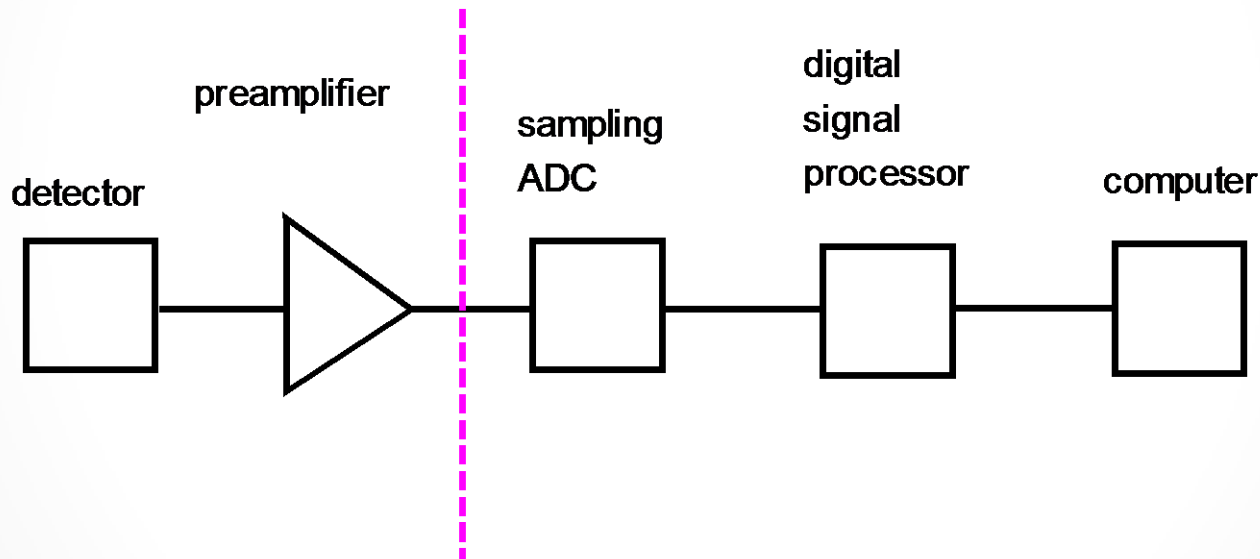


# Digital approach

- That is all the equipment you need for whatever type of measurement you want to do
- All analysis is done in software!
- Extremely flexible and versatile. Compact.
- Hardware is cheaper. But, to make use of full potential, software development needs resources. **Do not underestimate this.**
- Digitisers with dedicated signal processing hardware are commercially available

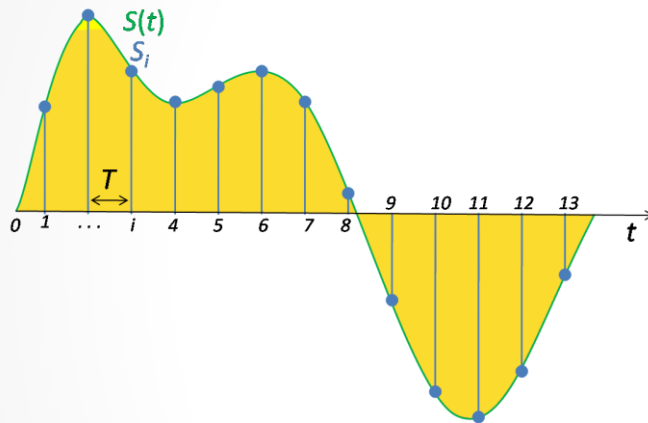
# Digital approach

- Could do analysis in dedicated hardware for equipment “permanently” installed on beam line.



# Digital approach

- Sampling Analogue to Digital Converter (ADC).



- ADC measures the amplitude of the signal at fixed intervals ( $T$  sampling period)
- Normally specified as sampling rate ( $1/T$ )
- ADC gives amplitude in discrete values
- Number of values =  $2^{\text{number of bits of ADC}}$



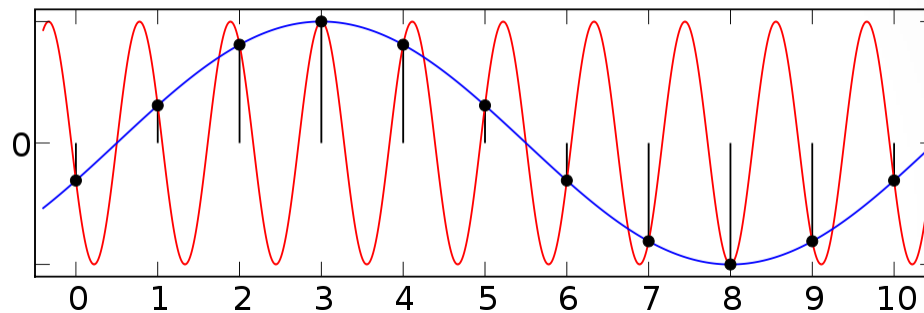
# Digital approach

- Two “digitisation” errors:
  - 1) Time
  - 2) Amplitude
- Offset errors
- Time digitisation/sampling
- Nyquist–Shannon sampling theorem:  
**Perfect reconstruction of a signal is possible when the sampling frequency is greater than twice the maximum frequency of the signal being sampled (Nyquist rate)**

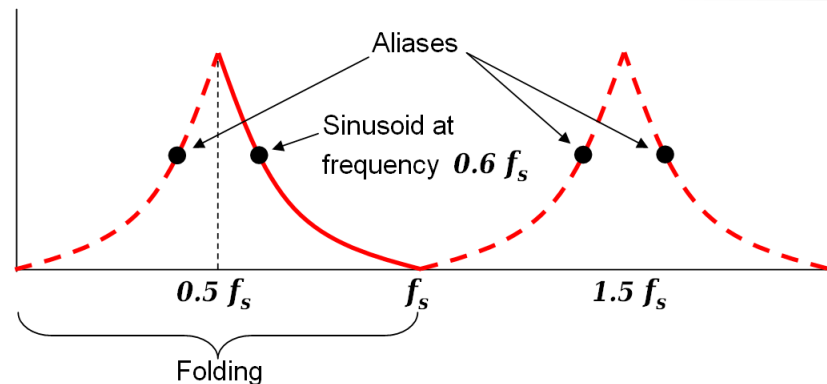
# Digital approach

- If the sampling rate is less than the Nyquist rate, aliasing will occur.

- Time domain:



- Frequency domain:  
(sampling frequency  $f_s$ )



# Digital approach



- Equivalent for spatially undersampling image: Moiré pattern

# Digital approach

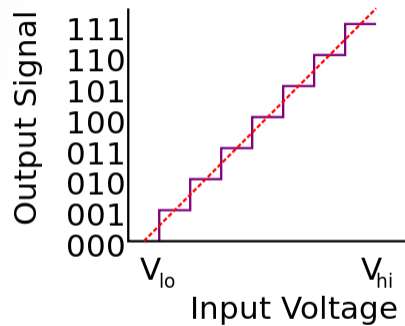
- Aliases hard/impossible to remove with digital signal shaping filters (they are really present in signal).
  - Detector signals are intrinsically bandwidth limited  $\Rightarrow$  do not really need anti-aliasing filter.
  - Usually better to chose sampling rate  $\sim 2$  times higher than Nyquist rate because of digital signal filtering.
  - To give you an idea: for resistive wire gas tubes we use a 33MHz sample rate for the ADCs
- Time digitization errors are easy to deal with in practice

# Digital approach

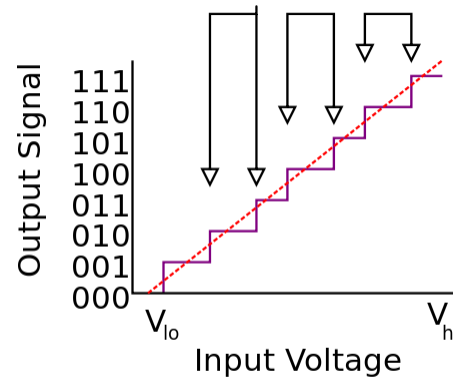
- Amplitude digitisation
- Needs some attention
- Number of bits.
- 10 **effective** bits (not actual number of bits) is good enough for most applications.
- Effective number of bits is less than actual number of bits because of noise and other error sources.
- Most errors specified in Least Significant Bit (LSB)  $\Rightarrow$  helps to have ADC with higher number bits.

# Digital approach

- Differential Non-Linearity (DNL):



A. Linear

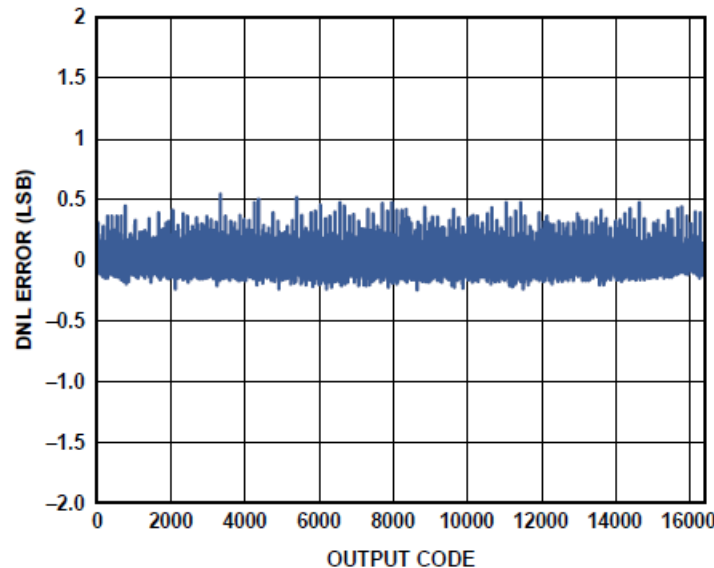


B. Non-Linear

- “Short range” non-linearity (how much it is varies from neighbour to neighbour)

# Digital approach

- AD9648: Sampling ADC, 14-bit, 125 MHz maximum sampling rate

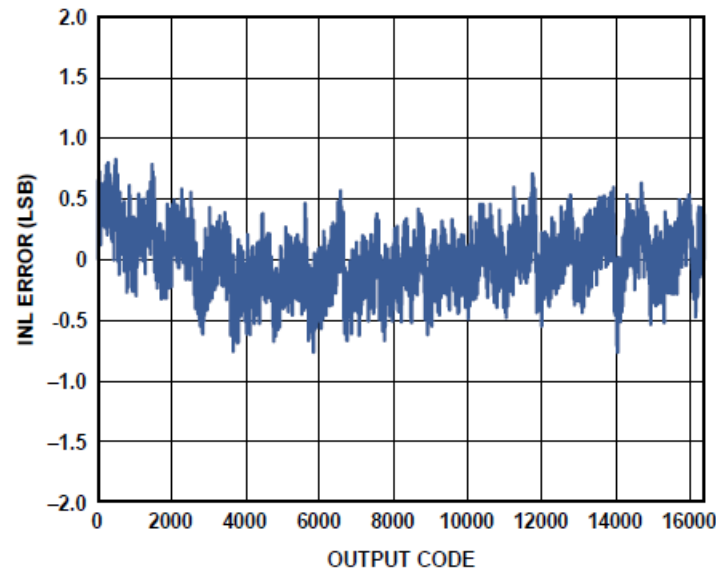


- Usually specified for slow signals. Non-linearity noticeably higher for fast signals

# Digital approach

- Integral Non-Linearity (INL):
- Deviation between actual output value and perfect/theoretical output value.

- AD9648:

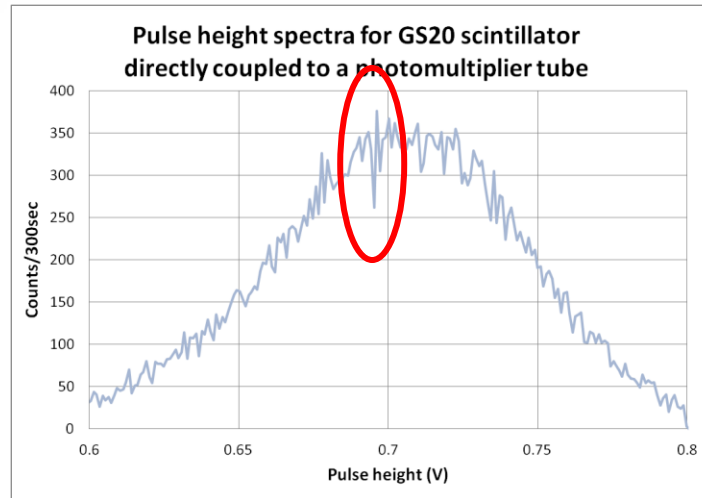


- See as medium to long range effects



# Digital approach

- Effect of non-linearity on pulse height spectrum:



- Dips usually appear at predictable bins (128, 256, 512)
- Dips/peaks will be greatly reduced by using low pass filter with (very) short RC time

# Digital approach

- Offset voltages
- ADC will give non-zero reading even with 0V at input.
- Intrinsic in ADC  $\Rightarrow$  not good enough to remove DC component from analogue signal before ADC.
- Originate from analogue parts of ADC  $\Rightarrow$  don't get better (in absolute value) for ADCs with more bits.
- ✗ Offsets devastating for resistive wire gas tube electronics.

# Digital approach

- Offsets vary with temperature  $\Rightarrow$  one off calibration is not good enough.
- ✓ Offsets are easily removed with digital signal processing.

# Digital signal processing

- I will discuss only digital signal filtering.
- Two types of digital (signal) filters:
  - 1) Finite Impulse Response (FIR)
  - 2) Infinite Impulse Response (IIR)

# Digital signal processing

- Infinite Impulse Response (IIR)
- General formulae (time domain):

$$y[n] = \sum_{i=0}^P b_i x[n-i] + \sum_{i=1}^Q a_i y[n-i]$$

- Current value of output of filter  $y[n]$  depends on:
  - 1) the previous input values of the filter  $x[n-i]$
  - 2) the previous output values of the filter  $y[n-i]$   
(feedback)  $\Rightarrow$  recursive
- P and Q are the order of the 2 parts

# Digital signal processing

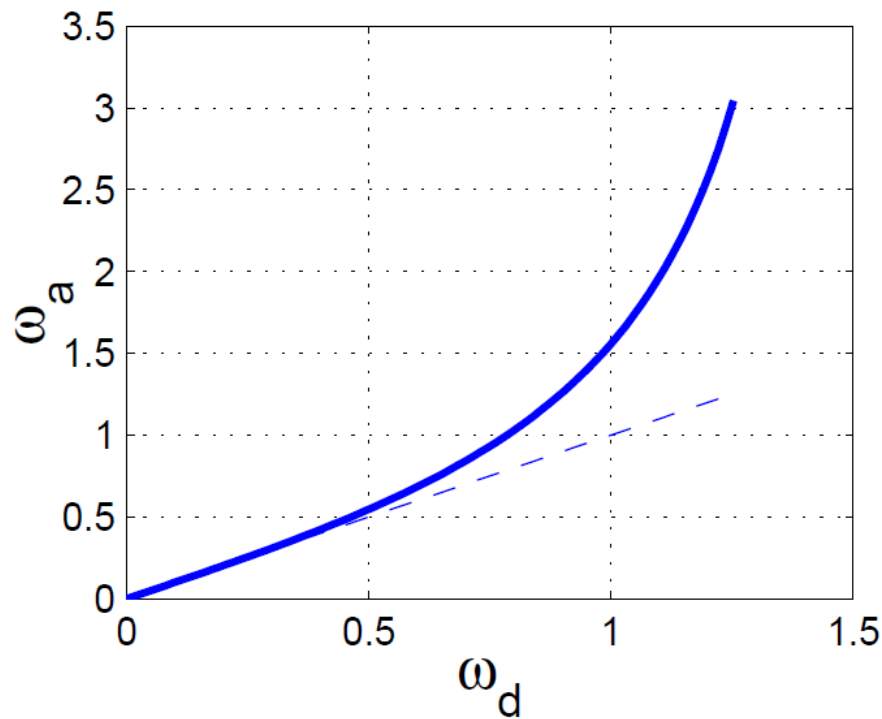
- IIR filters are equivalent of analogue filters (RC, CR, Butterworth and so on) and are usually designed starting from their analogue counterpart.
- They are easy to implement in software and hardware.
- Because of feedback, signals are “infinitely” long, like the capacitor in an RC filter will never get completely discharged.
- Can be unstable, but 1<sup>st</sup> order RC filter is unconditionally stable, provided that  $0 < a_1 < 1$

# Digital signal processing

- One (theoretical) complication:
- Maximum frequency in analogue systems is infinite.
- Maximum frequency in digital systems is the Nyquist frequency (finite).

# Digital signal processing

- Bilinear transform is simplest way to map (infinite) analogue frequency range onto (finite) digital frequency range (frequency warping).





# Digital signal processing

- I prefer to make my digital filters by cascading 1<sup>st</sup> order RC and CR filters.

- 1<sup>st</sup> order filter:  $P=1$  and  $Q=1$

$$y[n] = \sum_{i=1}^Q a_i y[n-i] + \sum_{i=0}^P b_i x[n-i] \quad \Rightarrow$$

$$y[n] = a_1 y[n-1] + b_1 x[n-1] + b_0 x[n]$$

- This web-site can calculate coefficients for most common filters (warped and unwarped):  
<http://www.cs.york.ac.uk/~fisher/mkfilter>

# Digital signal processing

- Problem for first sample of trace ( $n=0$ ) in:

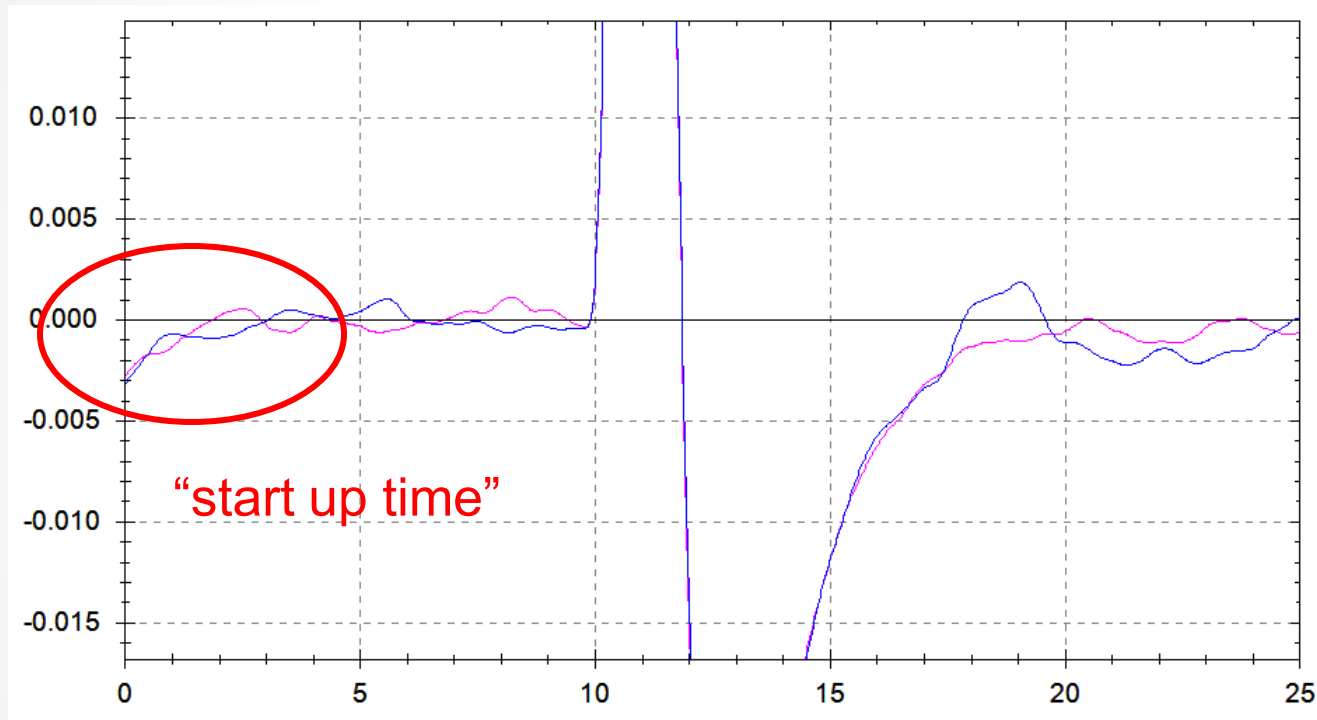
$$y[n] = a_1 y[n-1] + b_1 x[n-1] + b_0 x[n]$$

$$y[0] = a_1 y[-1] + b_0 x[0] \quad \text{where } y[-1] \text{ is undefined}$$

- Simple solution: start with second sample ( $n=1$ ) and assign  $y[0] = y[1]$  after calculating  $y[1]$
- Issue for calculating  $y[1]$ : have to use  $y[0]$  which is not calculated previously (filtered)  $\Rightarrow$  filter needs some samples to start up.

# Digital signal processing

- Effect:



- Leave enough samples in trace before start of signal!!

# Digital signal processing

- How to implement a simple (unwarped) 1<sup>st</sup> order low-pass RC filter (integrator)

$$a_1 = e^{-T/\tau} \quad b_0 = 1 - e^{-T/\tau} \quad b_1 = 0$$

- where  $T$  is sampling period and  $\tau$  the RC time of the analogue RC filter.
- Remember that  $\tau = \frac{1}{2\pi f_c}$  with  $f_c$  cutoff frequency  
(mkfilter web-site wants  $f_c$  and not  $2\pi f_c$ )

# Digital signal processing

- How to implement a simple (unwarped) 1<sup>st</sup> order high-pass RC filter (differentiator)

$$a_1 = e^{-T/\tau} \quad b_0 = 0.5 (1 + e^{-T/\tau}) \quad b_1 = -0.5 (1 + e^{-T/\tau})$$

- where T is sampling period and  $\tau$  the RC time of the analogue RC filter.

# Digital signal processing

- Finite Impulse Response (FIR)

- General formulae (time domain):

$$y[n] = \sum_{i=0}^P b_i x[n-i]$$

- P is the order of the filter
- Current value of output of filter  $y[n]$  only depends on the input values of the filter  $x[n-i]$ .

# Digital signal processing

- FIR filters are always stable (no feedback).
- FIR filters tend to be of high order ( $\sim 20$ )  $\Rightarrow$  computational intensive.
- They don't have direct analogue equivalents.
- More complicated to calculate coefficients.
- FIR filtered signals return to baseline much faster because of finite length (number of terms).

# Digital signal processing

- Most interesting FIR filters:
  - 1) Moving average filter
  - 2) Trapezoidal filter
- Moving average filter:  $y[n] = \frac{1}{P+1} \sum_{i=0}^P x[n-i]$
- Low pass filter (integrator type).
- Order of filter determines rise/fall time of signal.
- Very efficient implementation using recursion.



# Digital signal processing

- Example 5<sup>th</sup> order filter

$$y[n] = \frac{1}{6} (x[n] + x[n-1] + x[n-2] + x[n-3] + x[n-4] + x[n-5])$$

$$y[n+1] = \frac{1}{6} (x[n+1] + x[n] + x[n-1] + x[n-2] + x[n-3] + x[n-4])$$

- after rewrite:

$$y[n+1] = y[n] + \frac{1}{6} (x[n+1] - x[n-5])$$

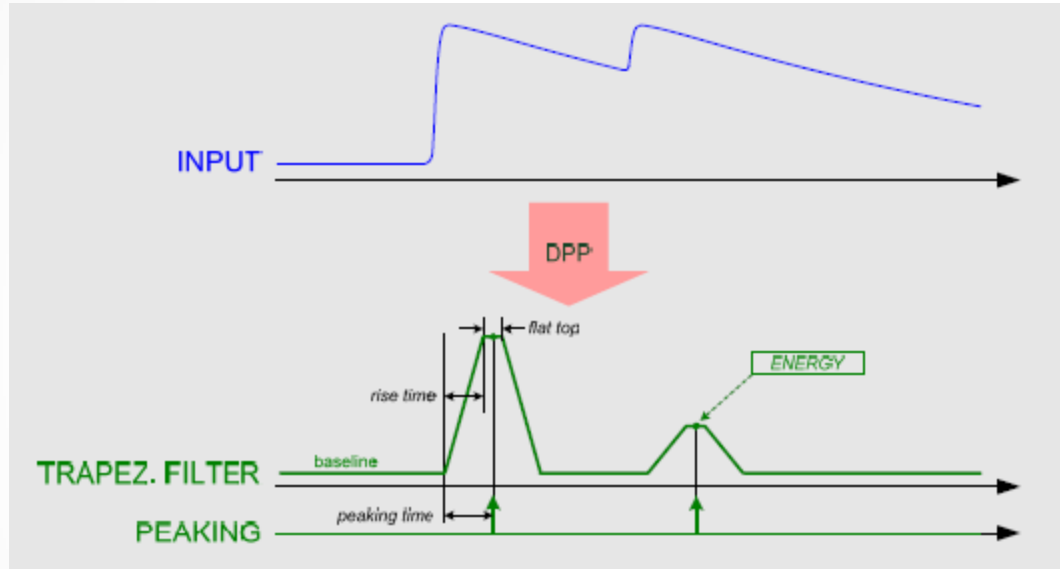
- ✓ Need only 1 addition and 1 subtraction, independent of order of filter.

# Digital signal processing

- ✗ Once  $y[n]$  gets corrupted output of filter will be corrupted for ever  $\Rightarrow$  best to add redundancy.
- Trapezoidal filter
- Band pass filter (combined low pass and high pass).
- ✓ Good timing resolution and energy resolution (spectroscopy).

# Digital signal processing

- Works best for signals with long fall time.



- ✗ Very hard to find good filter coefficients for fast signals.

# Digital signal processing

- Implementation: Basically 2 time-shifted moving average filters.
- Assume moving average filter:  $MA[n] = \frac{1}{P+1} \sum_{i=0}^P x[n-i]$
- Trapezoidal output is given by:
$$y[n] = MA[n] - MA[n - P - G]$$
- where  $G$  is the width of the flat part (in samples).
- Find it hard/impossible to suggest values for  $P$  and  $G$ .

# Digital signal processing

- Pros and cons of IIR compared to FIR.
- ✓ IIR does require much fewer terms (additions and multiplication) than most FIR.
- ✓ More intuitive/easier to design since there are direct equivalents of analogue filters.
- ✗ Require a long time to get back to baseline.
- I normally use IIR filters for my “electronics”.

# Summary

- Current generation of fast sampling ADC is good enough to be able to use digital approach.
- Very flexible, if you have the data analysis software.
- Signal filtering in digital electronics has many advantages.
- Modern Field Programmable Arrays (FPGAs) are capable of implementing signal processing algorithm that was optimised in software.



# The End

