

Bounded Input/Bounded Output:

LTV System: $\dot{x} = A(t)x + B(t)u$

$$y = C(t)x + D(t)u$$

When $x(0) = 0$ — Soln:

$$y_f(t) = \int_0^t C(t) \Phi(t, \tau) B(\tau) u(\tau) d\tau + D(t)u(t)$$

Def: The system is BIBO stable if there exists $0 < g < \infty$, s.t. for every input $u(t)$, the forced response satisfies:

$$\sup_{t \in [0, \infty)} \|y_f(t)\| \leq g \sup_{t \in [0, \infty)} \|u(t)\|$$

Supremum = least upper bound.
upper value for those included
in the set.

* Max is largest # w/ in the set.
Sup bounds the set. (may not or
may be part of the set)

Note: In the definition the I.C. are assumed to be zero.
If they are not zero, then still only use forced response.

Conditions for BIBO stability: (equivalent statements) (9.2)

1. The LTV system is uniformly BIBO stable
2. Every entry of $D(t)$ is uniformly bounded and

$$\sup_{t \geq 0} \int_0^t |g_{ij}(t, \tau)| d\tau < \infty$$

for every entry $g_{ij}(t, \tau)$ of $C(t) \Phi(t, \tau) B(\tau)$

Impulse Response
of the system

Proof:

$$(2) \Rightarrow (1)$$

$$\|y_f(t)\| \leq \int_0^t \|C(t)\Phi(t,\tau)B(\tau)\| \|u(\tau)\| d\tau + \|D(t)\| \|u(t)\|$$

$$\text{let } \mu := \sup_{t \in [0, \infty)} \|u(t)\| \quad \delta := \sup_{t \in [0, \infty)} \|D(t)\|$$

$$\|y_f(t)\| \leq \underbrace{\left(\int_0^t \|C(t)\Phi(t,\tau)B(\tau)\| d\tau + \sup_{t \geq 0} \|D(t)\| \right)}_g \sup_{t \geq 0} \|u(t)\|$$

For g to be finite we need:

1) $\sup_{t \geq 0} \|D(t)\| < \infty$, i.e. every $D(t)$ is uniformly bounded.

$$2) \int_0^t \|C(t)\Phi(t,\tau)B(\tau)\| d\tau < \infty$$

But $\|P\| \leq \sum_{i,j} |p_{ij}|$, why?

$$\|A+B\| \leq \|A\| + \|B\|$$

$$\begin{aligned} \|P\| &= \left\| \begin{pmatrix} p_{11} & p_{12} & \dots & p_{1n} \\ \vdots & \vdots & \ddots & \vdots \\ p_{ni} & \dots & \dots & p_{nn} \end{pmatrix} \right\| = \left\| \begin{pmatrix} p_{11} & 0 & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & \dots & \dots & 0 \end{pmatrix} + \begin{pmatrix} 0 & p_{12} & 0 & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & \dots & \dots & \dots & 0 \end{pmatrix} \right. \\ &\quad \left. + \dots + \begin{pmatrix} 0 & \dots & 0 & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & \dots & p_{nn} & \dots & 0 \end{pmatrix} \right\| \leq \sum \left\| \begin{pmatrix} 0 & \dots & 0 & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & \dots & p_{ij} & \dots & 0 \end{pmatrix} \right\| \\ &= \sum |p_{ij}| \end{aligned}$$

$$\therefore g = \sup_{t \geq 0} \int_0^t \|C(t)\Phi(t,\tau)B(\tau)\| d\tau + \sup_{t \geq 0} \|D(t)\|$$

$$\leq \sup_{t \geq 0} \sum_{i,j} \int_0^t |g_{ij}(t,\tau)| d\tau + \sup_{t \geq 0} \|D(t)\| < \infty$$

(1) \Rightarrow (2) (Prove by implication, i.e. $\neg(2) \Rightarrow \neg(1)$,
2nd statement false \Rightarrow 1st statement is false)

* Suppose at least one element of $D(t)$ is unbounded.

Let $d_{ij}(t)$ = the unbounded element

- Pick an arbitrary time t and consider the step input

$$u_T = \begin{cases} 0 & 0 \leq \tau < T \\ e_j & \tau \geq T \end{cases} \quad \forall \tau \geq 0$$

\uparrow
 $e_j \in \mathbb{R}^k$ is the j th vector in the
canonical basis of \mathbb{R}^k

Note that $\sup_{t \geq 0} \|u_T(t)\| = 1$

And @ time T we have —

$$y_f(T) = \int_0^T C(\tau) \Phi(T, \tau) B(\tau) u_T(\tau) d\tau + D(T) u_T(T) = D(T) u_T(T)$$

$$\Rightarrow \sup_{t \in [0, \infty)} \|y_f(t)\| \geq \|y_f(T)\| = \|D(T) u_T(T)\|$$

$$= \|D(T) e_j\| = \sqrt{\sum_{i=1}^m |d_{ij}(T)|^2} \geq |d_{ij}(T)|$$

\Rightarrow the norm must be larger
or equal to a single entry of
the j th col of D .

\rightarrow Since at least one element of d_{ij} is unbounded, we
can make $\sup_{t \in [0, \infty)} \|y_f(t)\|$ arbitrarily large using a
bounded input. \Rightarrow Not BIBO stable.

* Now suppose (2) is false because $\sup_{t \geq 0} \int_0^t |g_{ij}(\tau)| d\tau$ is
unbounded. ($D(t)$ is bounded)

For some i, j
let $u_T(t) = \begin{cases} +e_j \\ -e_j \end{cases}$

— Again pick an arbitrary T and consider the switching function.

$$u_T(\tau) = \begin{cases} e_j & g_{ij}(t, \tau) \geq 0 \\ -e_j & g_{ij}(t, \tau) < 0 \end{cases} \quad \tau \geq 0$$

$$y_f(t) = \begin{pmatrix} \int_0^t |g_{1j}(t, \tau)| d\tau \\ \vdots \\ \int_0^t |g_{nj}(t, \tau)| d\tau \end{pmatrix} + \begin{pmatrix} d_{1j}(t) \\ \vdots \\ d_{nj}(t) \end{pmatrix}$$

Note that the output.

$$\sup_{t \in [0, \infty)} \|y_f(t)\| \geq \|y_f(T)\| \geq \underbrace{\int_0^T |g_{ij}(T, \tau)| d\tau}_{\text{— choose this to be arbitrarily large (unbounded)}} + d_{ij}(T)$$

$\Rightarrow \sup_{t \in [0, \infty)} \|y_f(t)\|$ is not bounded even though the input is bounded.

\therefore (9.2) must hold for a system to be BIBO stable.

$$(1) \Rightarrow (2)$$

For LTI Systems : $\dot{x} = Ax + Bu$
 $y = Cx + Du$

$$(9.2) \quad C\Phi(t, \tau)B \Rightarrow C e^{A(t-\tau)} B$$

$$\Rightarrow \sup_{t \geq 0} \int_0^t \underbrace{|\bar{g}_{ij}(t-\tau)|}_{\text{LTI can look @ time difference of systems.}} d\tau < \infty$$

- LTI can look @ time difference of systems.

$$p = t - \tau$$

$$\Leftrightarrow \int_0^\infty |\bar{g}_{ij}(p)| dp < \infty$$

Freq. Domain Characteristics :

(Theorem 9.3) For LTI systems $\dot{x} = Ax + Bu$
 $y = Cx + Du$

where TF $\hat{G}(s) = \mathcal{L}\{C(sI - A)^{-1}B\}$ (leaving off D matrix, but D is constant & won't change the poles)
 The following statements are equivalent — \rightarrow TF of impulse response.

- 1) The system is uniformly BIBO stable
- 2) Every pole of every entry in $\hat{G}(s)$ has a strictly negative real part.

Proof — We need to show that :

$$\int_0^\infty |\bar{g}_{ij}(p)| dp < \infty \iff \text{the poles of } \hat{g}_{ij}(s) \text{ are in the left hand plane.}$$

\rightarrow We know that all the elements of $\hat{G}(s)$ are strictly proper (will be true even if we included the D matrix because D will be constant & won't change the poles).

Strictly ^{rational} proper = fewer zeros than poles.

$$\Rightarrow \hat{g}_{ij}(s) = \frac{a_0 s^q + a_1 s^{q-1} + \dots + a_{q-1} s + a_q}{(s-\lambda_1)^{m_1} (s-\lambda_2)^{m_2} \dots (s-\lambda_k)^{m_k}} \quad \sum m_k > q$$

$$= \frac{a_{11}}{s-\lambda_1} + \frac{a_{12}}{(s-\lambda_1)^2} + \dots + \frac{a_{1m_1}}{(s-\lambda_1)^{m_1}} + \dots$$

$$+ \frac{a_{k1}}{(s-\lambda_k)} + \frac{a_{k2}}{(s-\lambda_k)^2} + \dots + \frac{a_{km_k}}{(s-\lambda_k)^{m_k}}$$

The inverse Laplace transform is

$$g_{ij}(t) = a_{11} e^{\lambda_1 t} + a_{12} t e^{\lambda_1 t} + \dots + a_{1m_1} t^{m_1-1} e^{\lambda_1 t} + \dots$$

$$+ a_{k1} e^{\lambda_k t} + \dots + a_{km_k} t^{m_k-1} e^{\lambda_k t}$$

for each term we have

$$\int_0^\infty a_{ij} t^{j-1} e^{\lambda_i t} dt < \infty \text{ iff } \operatorname{Re}\{\lambda_i\} < 0$$

\Rightarrow If one of these has a positive or even a zero pole then the terms don't converge to zero and $|g_{ij}(t)|$ will not converge to zero because the integral will keep growing. \rightarrow system not BIBO stable.

BIBO vs. Lyapunov Stability: (CLTI systems)

$$\left[\begin{array}{c} \text{exp stability} \\ \text{in the sense of} \\ \text{Lyapunov} \end{array} \right] \Rightarrow \left[\text{BIBO stability} \right]$$

Proof —

Exp stability \Rightarrow All e-values of A are in ~~RHP~~ LHP

\Rightarrow all poles of $C(sI-A)^{-1}B$ are in ~~RHP~~ LHP

\Rightarrow BIBO stability

$$\left[\text{BIBO stability} \right] \not\Rightarrow \left[\begin{array}{c} \text{exp stability} \\ \text{in the sense of} \\ \text{Lyapunov} \end{array} \right]$$

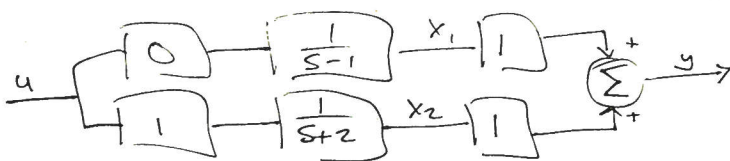
— this can happen when Ce^{At} or $e^{At}B$ cancel terms in e^{At} that are not converging exp. fast.

$$\dot{x} = \begin{bmatrix} 1 & 0 \\ 0 & -2 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u \quad y = \begin{bmatrix} 1 & 1 \end{bmatrix} x$$

$$e^{At} = \begin{bmatrix} e^t & 0 \\ 0 & e^{-2t} \end{bmatrix} \rightarrow \text{not exp. stable because of } e^t$$

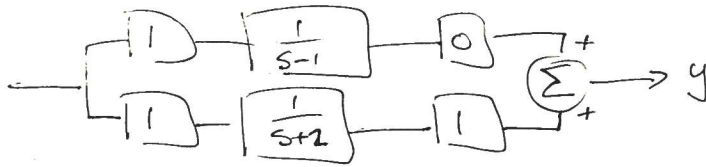
$$\text{But } Ce^{At}B = \begin{bmatrix} 1 & 1 \end{bmatrix} \begin{bmatrix} e^t & 0 \\ 0 & e^{-2t} \end{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix} = e^{-2t}$$

2-situations



$$G(s) = \frac{1 \cdot 1}{s+2} + \frac{1 \cdot 0}{s-1} = \frac{1}{s+2} = \frac{Y(s)}{U(s)}$$

\hookrightarrow This is unstable, but mode is not effected by the input.



$$\dot{x}_1 = x_1 + u$$

$$\dot{x}_2 = -2x_2 + u$$

$$y = x_2$$

$$G(s) = \frac{1}{s+2}$$

→ Same TF, but the output is not "seeing" the "run-away" mode. The output sensor does not measure this.

Two problems:

- ① Input does not effect all the modes \Rightarrow this system lacks "controllability"
- ② Sensors do not see all of the modes \Rightarrow this system lacks "observability"

→ System must be completely controllable + observable in order for BIBO stability to imply internal stability.

Discrete Time:

$$x(t+1) = A(t)x(t) + B(t)u(t)$$

$$y(t) = C(t)x(t) + D(t)u(t)$$

Forced response:

$$y_f(t) = \sum_{\tau=0}^{t-1} C(t) \underbrace{\Phi(t, \tau+1)}_{\text{discrete state transition matrix}} B(\tau) u(\tau) d\tau \quad \forall t \geq 0$$

Definition of BIBO stability is almost identical to cont. case

$$\sup_{t \in \mathbb{N}} \|y_f(t)\| \leq g \sup_{t \in \mathbb{N}} \|u(t)\| \quad w/ \quad g > 0$$

Thm 9.5 the following 2 statements are equivalent

1. the DLTV system is BIBO stable
2. Every entry of $D(t)$ is uniformly bounded and

$$\sup_{t \geq 0} \sum_{\tau=0}^{t-1} |g_{ij}(t, \tau)| < \infty$$

$$\text{Where } G(t, \tau) = C \Phi(t, \tau) B(\tau)$$

Thm 9.6 The following three statements are equivalent

- 1) The time-invariant system

$$x^+ = Ax + Bu, \quad y = Cx + Du$$

is BIBO stable

- 2) Every entry of $G(p) = CA^p B$ satisfies

$$\sum_{p=1}^{\infty} |g_{ij}(p)| < \infty$$

- 3) Every rule of $\hat{G}(z) = Z\{CA^p B\}$ has magnitude in the open unit circle.