

TRIPLE OPERATIONAL AMPLIFIER

The TCA220 is a monolithic integrated circuit, consisting of three identical high-gain amplifiers.

The amplifiers have a differential input stage and an emitter-follower output stage, which can supply a current up to 100 mA.

The unity-gain frequency with 6 dB/octave compensation is 5 MHz minimum. No latch-up occurs if the input voltage range is exceeded.

QUICK REFERENCE DATA

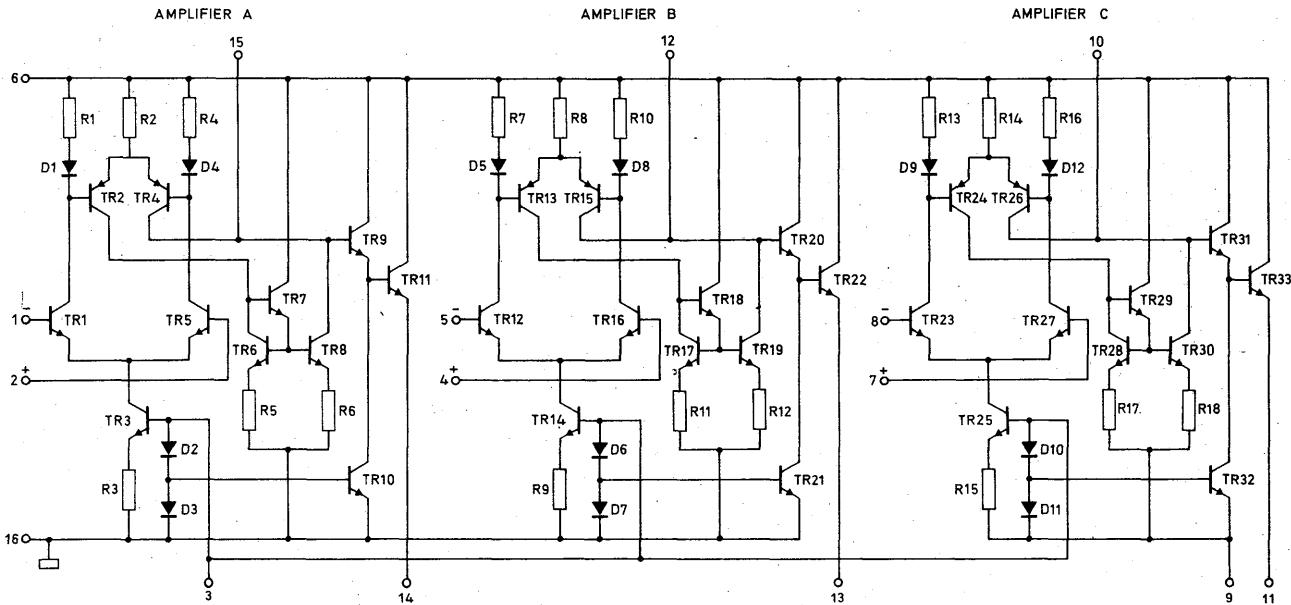
Positive supply voltage	V_P	nom.	6	V
Negative supply voltage	V_N	nom.	6	V

Voltage gain	G_v	typ.	4000	
Common mode rejection ratio	CMRR	typ.	90	dB
Supply voltage rejection ratio	SVRR	typ.	200	$\mu V/V$
Input offset voltage	V_{io}	typ.	2	mV
Input offset current	I_{io}	typ.	0,2	μA

PACKAGE OUTLINE plastic 16-lead dual in-line (see general section).

CIRCUIT DIAGRAM

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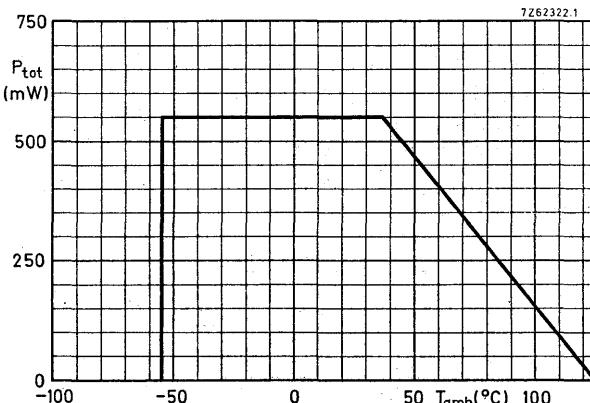
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RATINGS Limiting values in accordance with the Absolute Maximum System (IEC134)Voltages

Supply voltage	V_{6-16}	max.	18	V
Common mode input voltage	V_i	max.	18	$V^1)$
Differential input voltages	$\pm V_{1-2}$ $\pm V_{5-4}$ $\pm V_{8-7}$	max.	5,0	V
Pin No. 9 voltage	V_{9-16}	max.	0	$V^2)$

Currents

Input currents (pins, 1, 2, 4, 5, 7, 8)	$I_1; I_2$ $I_4; I_5$ $I_7; I_8$	max.	0,5	mA
Output currents (pins 14, 13, 11)	$-I_{14}; -I_{13}; -I_{11}$	max.	100	mA
Bias current (pin 3)	I_3	max.	5,0	mA

Total power dissipationTemperatures

Storage temperature	T_{stg}	-55 to +125	°C
Junction temperature	T_j	max.	125 °C

1) For a total supply voltage less than 18 V, the absolute maximum input voltage is equal to the supply voltage.

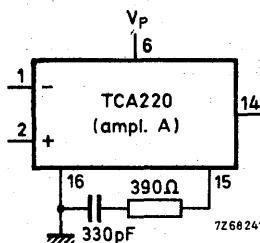
2) If amplifier C is used, pin 9 must be connected to pin 16.

THERMAL RESISTANCE

From junction to ambient	R_{th}	max.	160	$^{\circ}\text{C}/\text{W}$
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CHARACTERISTICS (each amplifier) at $V_p = 6 \text{ V}$; $-V_N = 6 \text{ V}$; $T_{\text{amb}} = 25 \text{ }^{\circ}\text{C}$
 $R_L = 10 \text{ k}\Omega$ (unless otherwise specified)

<u>Voltage gain</u> at $\pm V_{\text{OM}} = 3,5 \text{ V}$	G_v	typ.	4000	
<u>Input offset voltage</u> at $R_S \leq 200 \Omega$	V_{io}	{ typ. <	2 10	mV
<u>Input bias current</u>	I_i	{ typ. <	1,0 2,0	μA
<u>Input offset current</u>	I_{io}	typ.	0,2	μA
<u>Common mode rejection ratio</u> at $R_S = 2 \text{ k}\Omega$	CMRR	typ.	90	dB
<u>Input voltage range</u>	V_i	-4,3 to +5,6		V
<u>Differential input resistance</u>	R_i	>	25	$\text{k}\Omega$
<u>Supply voltage rejection ratio</u> at $R_S = 2 \text{ k}\Omega$	SVRR	typ.	200	$\mu\text{V/V}$
<u>Peak output voltage swing</u>	V_{OM}	-6 to +3,5		V
Total current at $V_o = 0$; $R_L = 10 \text{ k}\Omega$ at $V_o = 0$; $R_L = \infty$	I_{tot}	typ.	1,0	mA
<u>Slew rate</u> (unity-gain)		typ.	0,4	$\text{V}/\mu\text{s}$
<u>Bias current</u> (all three amplifiers together)	I_3	>	200	$\mu\text{A}^1)$
<u>Channel separation</u> between amplifiers A and B between amplifiers A and C between amplifiers B and C		typ. typ. typ.	94 130 110	$\text{dB}^{2)}$

Frequency compensation circuit

- 1) The voltage at pin 3 is always 2 diode voltages (approx. 1,5 V) above the negative supply voltage; if the bias current is obtained from the positive supply voltage a dropping resistor $R_p \leq \frac{V_p - V_N - 1,5}{200 \cdot 10^{-6}}$

gives minimum power consumption.

- 2) Channel separation defined as $20 \log \frac{V_{oA}}{V_{oB}} \times G_B$, if G_B is the closed loop gain of amplifier B.