

# Lecture 9: LQG/Loop Transfer Recovery (LTR)

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Big picture  
Loop transfer recovery  
Target feedback loop  
Fictitious KF

## Big picture

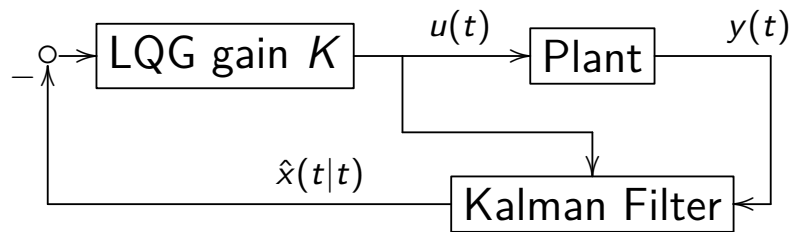
Where are we now?

- ▶ LQ: optimal control, guaranteed robust stability under basic assumptions in stationary case
- ▶ KF: optimal state estimation, good properties from the duality between LQ and KF
- ▶ LQG: LQ+KF with separation theorem
- ▶ frequency-domain feedback design principles and implementations

Stability robustness of LQG was discussed in one of the homework problems: the nice robust stability in LQ (good gain and phase margins) is lost in LQG.

LQG/LTR is one combined scheme that uses many of the concepts learned so far.

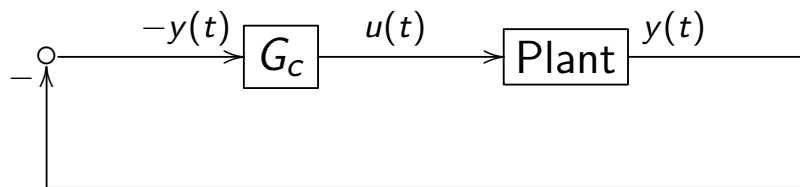
# Continuous-time stationary LQG solution



$$u(t) = -K\hat{x}(t|t)$$

$$\begin{aligned} \frac{d\hat{x}(t|t)}{dt} &= A\hat{x}(t|t) + Bu(t) + F(y(t) - C\hat{x}(t|t)) \\ &= (A - BK - FC)\hat{x}(t|t) + Fy(t) \end{aligned}$$

$\Leftrightarrow$

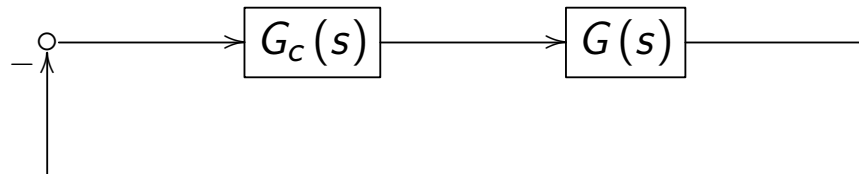


$$G_c(s) = K(sI - A + BK + FC)^{-1}F \quad (1)$$

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## Loop transfer recovery (LTR)



### Theorem (Loop Transfer Recovery (LTR))

If a  $m \times m$  dimensional  $G(s)$  has only minimum phase transmission zeros, then the open-loop transfer function

$$G(s)G_c(s) = \left[ C(sI - A)^{-1}B \right] \left[ K(sI - A + BK + FC)^{-1}F \right] \xrightarrow{\rho \rightarrow 0} C(sI - A)^{-1}F \quad (2)$$

$K$  and  $\rho$  are from the LQ  $[(A, B)$  controllable,  $(A, C)$  observable]

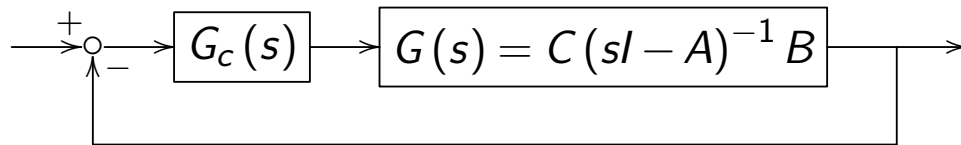
$$J = \int_0^\infty \left( x^T(t)C^T Cx(t) + \rho u^T(t)Nu(t) \right) dt \quad (3)$$

$$\dot{x}(t) = Ax(t) + Bu(t) \quad (4)$$

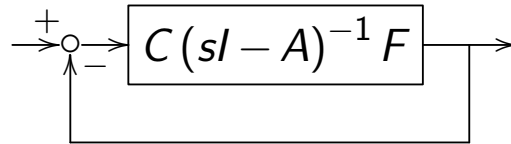
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# Loop transfer recovery (LTR)



converges, as  $\rho \rightarrow 0$ , to the *target feedback loop*



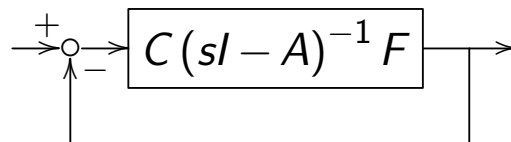
key concepts:

- ▶ regard LQG as an output feedback controller
- ▶ will design  $F$  such that  $C(sI - A)^{-1}F$  has a good loop shape
- ▶ not a conventional optimal control problem
- ▶ not even a stochastic control design method

## Selection of $F$ for the target feedback loop

standard KF procedure: given noise properties ( $W$ ,  $V$ , etc), KF gain  $F$  comes from RE

*fictitious* KF for target feedback loop design: want to have good behavior in



select  $W$  and  $V$  to get a desired  $F$  (hence a *fictitious* KF problem):

$$\begin{aligned} \dot{x}(t) &= Ax(t) + Lw(t), & E[w(t)w^T(t+\tau)] &= I\delta(\tau) \\ y(t) &= Cx(t) + v(t), & E[v(t)v^T(t+\tau)] &= \mu I\delta(\tau) \end{aligned}$$

which gives

$$F = \frac{1}{\mu} MC^T, \quad AM + M^T A + LL^T - \frac{1}{\mu} MC^T CM = 0, \quad M \succ 0 \quad (5)$$

# The target feedback loop from fictitious KF

$$\begin{aligned}\dot{x}(t) &= Ax(t) + Lw(t), & E[w(t)w^T(t+\tau)] &= I\delta(\tau) \\ y(t) &= Cx(t) + v(t), & E[v(t)v^T(t+\tau)] &= \mu I\delta(\tau)\end{aligned}$$

Return difference equation for the fictitious KF is

$$[I_m + G_F(s)][I_m + G_F(-s)]^T = I_m + \frac{1}{\mu} [C\Phi(s)L][C\Phi(-s)L]^T$$

where  $G_F(s) = C(sI - A)^{-1}F$  and  $\Phi(s) = (sI - A)^{-1}$ . Then

$$\begin{aligned}\sigma[I_m + G_F(j\omega)] &= \sqrt{\lambda \left\{ [I_m + G_F(j\omega)][I_m + G_F(-j\omega)]^T \right\}} \\ &= \sqrt{1 + \frac{1}{\mu} \{ \sigma[C\Phi(j\omega)L] \}^2} \geq 1\end{aligned}$$

# The (nice) target feedback loop from fictitious KF

$$\begin{aligned}\sigma[I_m + G_F(j\omega)] &= \sqrt{\lambda \left\{ [I_m + G_F(j\omega)][I_m + G_F(-j\omega)]^T \right\}} \\ &= \sqrt{1 + \frac{1}{\mu} \{ \sigma[C\Phi(j\omega)L] \}^2} \geq 1\end{aligned}$$

gives:

- ▶  $\sigma_{\max} S(j\omega) = \sigma_{\max} [I + G_F(j\omega)]^{-1} \leq 1$ , namely

no disturbance amplification at any frequency

- ▶  $\sigma_{\max} T(j\omega) = \sigma_{\max} [I - S(j\omega)] \leq 2$ , hence,

guaranteed closed loop stable if  $\sigma_{\max} \Delta(j\omega) < 1/2$