



Xilinx ISE 10 Tutorial

A Tutorial on Using the Xilinx ISE Software to Create
FPGA Designs for the XESS XSA Boards

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What This Is and Is Not

There are numerous requests on newgroups that go something like this:

"I am new to using programmable logic like FPGAs and CPLDs. How do I start? Is there a tutorial and some free tools I can use to learn more?"

XILINX has released a free version of their ISE software on the web (they call it *WebPACK*) so that anyone can download a set of tools for CPLD and FPGA-based logic designs. And XESS Corp. has written this tutorial that attempts to give you a gentle introduction to using the ISE tools.

This tutorial shows the use of the ISE tools on three simple design examples: 1) an LED decoder, 2) a counter which displays its current value on a seven-segment LED and 3) a reprogrammable combination lock. Along the way, you will see:

- How to start an FPGA project.
- How to target a design to a particular type of FPGA.
- How to describe a logic circuit using VHDL and/or schematics.
- How to detect and fix VHDL syntactical errors.
- How to synthesize a netlist from a circuit description.
- How to fit the netlist into an FPGA.
- How to check device utilization and timing for an FPGA.
- How to generate a bitstream for an FPGA.
- How to download a bitstream into an FPGA.
- How to test the programmed FPGA.

That said, it is important to say what this tutorial will *not* teach you:

- It will not teach you how to design logic with VHDL.
- It will not teach you how to choose the best type of FPGA or CPLD for your design.
- It will not teach you how to arrange your logic for the most efficient use of the resources in an FPGA.
- It will not teach you what to do if your design doesn't fit in a particular FPGA.

- It will not show you every feature of the ISE software and discuss how to set every option and property.
- It will not show you how to use the variety of peripheral devices available on the XSA Boards.

In short, this is just a tutorial to get you started using the XILINX ISE FPGA tools. After you go through this tutorial you should be able to move on to more advanced topics.

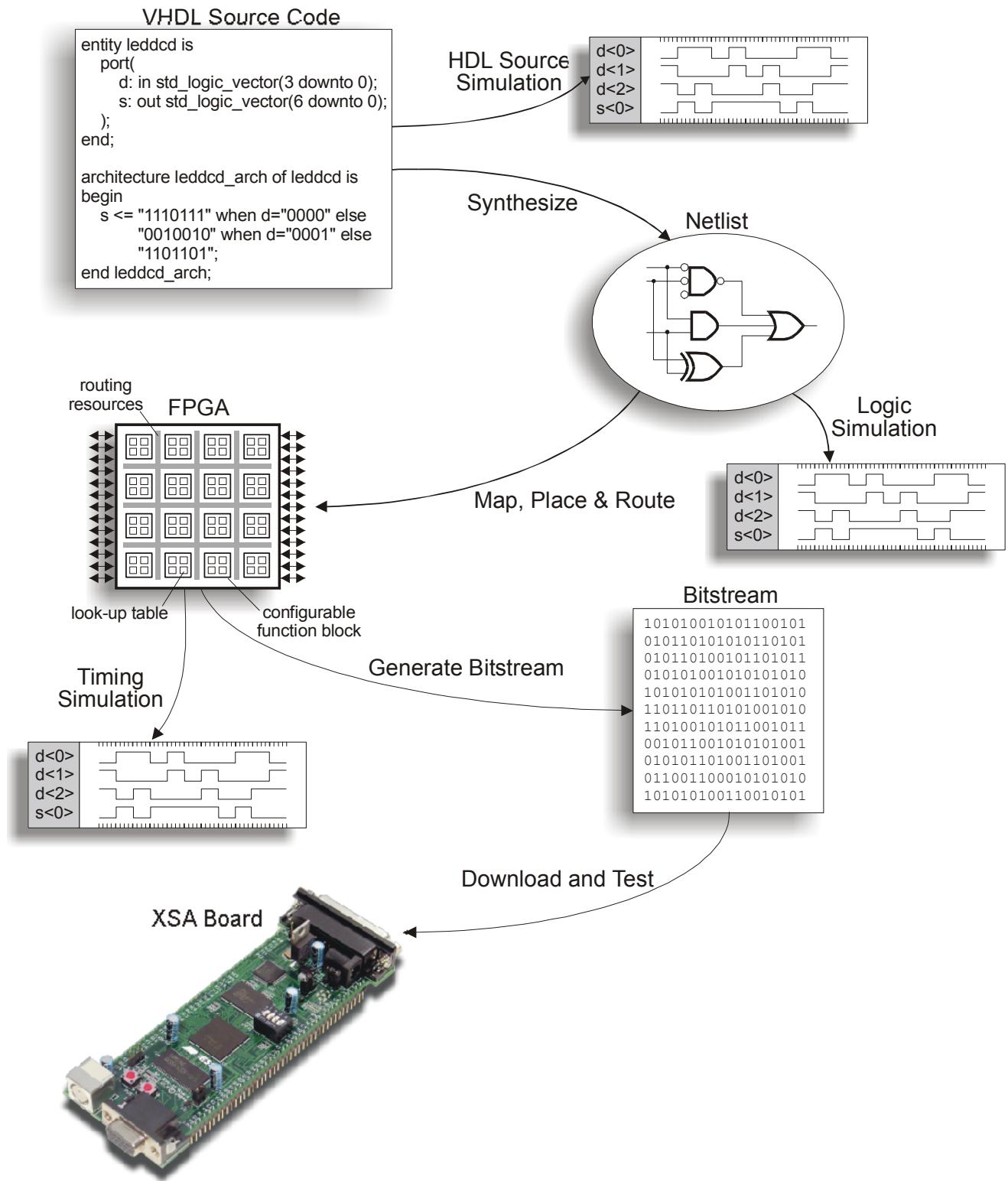
1

FPGA Programming

Implementing a logic design with an FPGA usually consists of the following steps (depicted in the figure which follows):

1. You enter a description of your logic circuit using a *hardware description language* (HDL) such as VHDL or Verilog. You can also draw your design using a schematic editor.
2. You use a *logic synthesizer* program to transform the HDL or schematic into a *netlist*. The netlist is just a description of the various logic gates in your design and how they are interconnected.
3. You use the *implementation tools* to map the logic gates and interconnections into the FPGA. The FPGA consists of many *configurable logic blocks*, which can be further decomposed into *look-up tables* that perform logic operations. The CLBs and LUTs are interwoven with various *routing resources*. The mapping tool collects your netlist gates into groups that fit into the LUTs and then the place & route tool assigns the groups to specific CLBs while opening or closing the switches in the routing matrices to connect them together.
4. Once the implementation phase is complete, a program extracts the state of the switches in the routing matrices and generates a *bitstream* where the ones and zeroes correspond to open or closed switches. (This is a bit of a simplification, but it will serve for the purposes of this tutorial.)
5. The bitstream is *downloaded* into a physical FPGA chip (usually embedded in some larger system). The electronic switches in the FPGA open or close in response to the binary bits in the bitstream. Upon completion of the downloading, the FPGA will perform the operations specified by your HDL code or schematic.

That's really all there is to it. XILINX ISE provides the HDL and schematic editors, logic synthesizer, fitter, and bitstream generator software. The XSTOOLS from XESS provide utilities for downloading the bitstream into the FPGA on the XSA Board.



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Installing ISE

Getting ISE

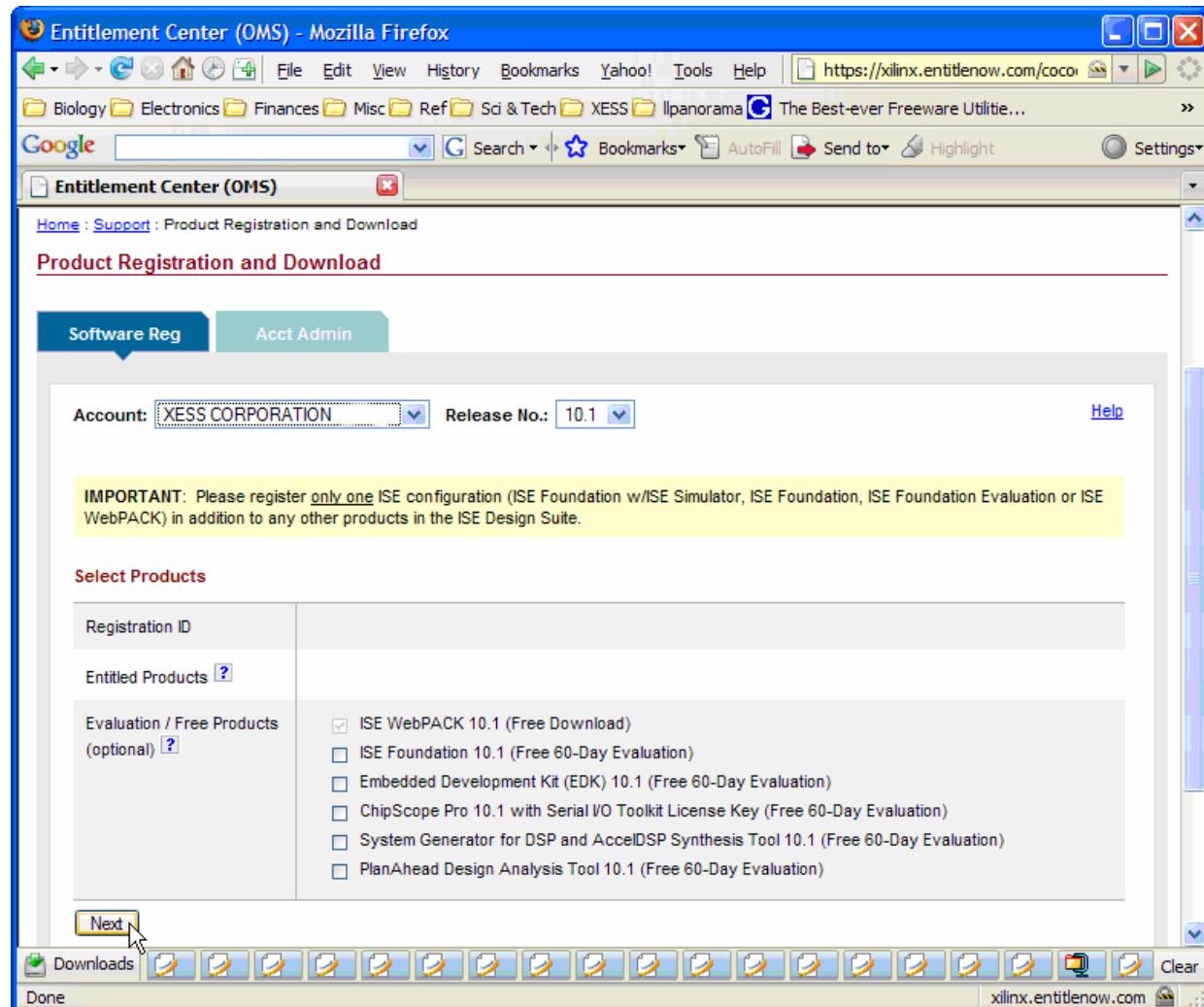
You can download the free ISE WebPACK software from this location:

http://www.xilinx.com/ise/logic_design_prod/webpack.htm. Click on the link to download the software as shown below.

The screenshot shows a Mozilla Firefox browser window with the following details:

- Title Bar:** Xilinx : ISE Foundation - Mozilla Firefox
- Address Bar:** http://www.xilinx.com/ise/logic_design_prod/webpack.htm
- Toolbar:** Back, Forward, Stop, Home, Refresh, Stop, Bookmarks, Help, Settings.
- Search Bar:** Google, Search, Bookmarks, AutoFill, Send to, Highlight.
- Page Content:**
 - Xilinx Logo:** XILINX®
 - Header Navigation:** Technology Solutions, Product & Services (highlighted), Market Solutions, Support, Online Store, About Xilinx.
 - Sub-navigation:** Silicon Devices | Design Tools | Intellectual Property | Boards & Kits | Training | Services | Third Party Alliances |
 - Current Path:** Home : Products & Services : Design Tools : Logic Design : ISE WebPACK Software
 - ISE WebPACK Software Section:**
 - Image:** ISE WebPACK logo.
 - Description:** ISE WebPACK software is the industry's only FREE, fully featured front-to-back FPGA design solution for Linux, Windows XP, and Windows Vista.
 - Text:** ISE™ WebPACK™ software is the ideal downloadable solution for FPGA and CPLD design offering HDL synthesis and simulation, implementation, device fitting, and JTAG programming.
 - Text:** ISE WebPACK provides the tools and features along with the same easy-to-use design environment as our award winning ISE™ Foundation™ design tools providing instant access to the ISE features and functionality at no cost. Xilinx has created a solution that allows convenient productivity by providing a design solution that is always up to date with error-free downloading and single file installation.
 - Call-to-Action:** Download ISE WebPACK Now!
 - [Download ISE WebPACK software for Windows and Linux](#) (with a cursor icon pointing to it).
 - Device Family Support:**
 - > Virtex™-5 LX, LXT, SXT, FXT
 - > Virtex-4 FX, LX, SX
 - > Spartan™-3A DSP
 - > Spartan-3A, 3AN
 - Right Sidebar:**
 - Buy ISE Design Suite
 - Request WebPACK DVD
 - Subscribe to Free Xilinx Newsletter
 - Documentation
 - ISE Product Brief
 - Tcl Reference Card
 - Software Manuals
 - Training
 - Learn ISE
 - Related Information
 - ISE Operating System Support
 - Bottom:** Downloads toolbar, URL https://secure.xilinx.com/webreg/register.do?group=esd_oms

You will have to create an account and choose a user ID and password before you are allowed to enter the download section of the site. Once you do that, you will get to a page where you can select the software that you want to download. In this case, check the ISE WebPACK 10.1 box and click the Next button.



Now you are finally at the page where you can download the ISE WebPACK software. Expand the Download Files Individually option, and then click on the Download link. This will download the entire install file for ISE WebPACK. With a size of 2.25 GB, it will take at least an hour to download, even with a 5 Mb/s link.

The screenshot shows a Mozilla Firefox browser window with the following details:

- Title Bar:** Entitlement Center (OMS) - Mozilla Firefox
- Address Bar:** https://xilinx.entitlenow.com/coco...
- Toolbar:** Back, Forward, Stop, Home, Refresh, etc.
- Menu Bar:** File, Edit, View, History, Bookmarks, Yahoo!, Tools, Help
- Google Search Bar:** Google Search, Bookmarks, AutoFill, Send to, Highlight, Settings
- Content Area:**
 - Section:** Product Registration and Download
 - Software Reg** (selected tab)
 - Acct Admin**
 - XESS CORPORATION - Software Release 10.1**
 - Download Products**
 - Download** section:
 - Registration ID:** Enter this Registration ID into the ISE Design Suite Installation Program.
 - DOWNLOAD** button (WebInstall Client 47.6 MB)
 - Download Alternatives:**
 - [Download Files Individually](#) (Each product must be downloaded and installed separately.)
 - [Request a DVD](#)
 - Included Products** section:
 - Evaluation Products**
 - ISE WebPACK 10.1 (Free Download)
 - Documentation** section: The following documents include information for all Entitled Products.
- Bottom Bar:** Downloads, Clear, javascript:download('15690799','15690801','15690800','333932','ISE WebPACK 10.1 (Free Download)')
- Page URL:** xilinx.entitlenow.com

Installing ISE

After the ISE WebPACK software download is complete, unpack the WebPACK_SFD.tar file and double-click the setup.exe file to start the installer. During the installation, you will have to enter the registration ID that you will receive in an email from Xilinx. Throughout the rest of the installation, accept the default settings for everything and you shouldn't have any problems. The total installation will take at least an hour.

Getting XSTOOLS

If you are going to download your FPGA bitstreams into an XESS FPGA Board, then you will need to get the XSTOOLS software from <http://www.xess.com/h007000.html>. Just download the latest XSTOOLS setup file.

Installing XSTOOLS

Double-click the XSTOOLS setup file. The installation script will run and install the software. Accept the default settings for everything.

Getting the Design Examples

You can download the project files for the design examples shown in this tutorial from <http://www.xess.com/projects/ise-10.zip>.

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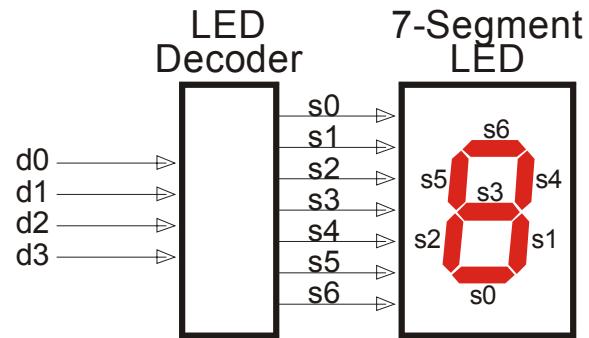
Our First Design

An LED Decoder

The first FPGA design you will try is an LED decoder. An LED decoder takes a four-bit input and outputs seven signals, which drive the segments of an LED digit. The LED segments will be driven to display the digit corresponding to the hexadecimal value of the four input bits as follows:

Four-bit Input	Hex Digit	LED Display
0000	0	0
0001	1	1
0010	2	2
0011	3	3
0100	4	4
0101	5	5
0110	6	6
0111	7	7
1000	8	8
1001	9	9
1010	A	A
1011	B	b
1100	C	c
1101	D	d
1110	E	E
1111	F	F

A high-level diagram of the LED decoder looks like this:

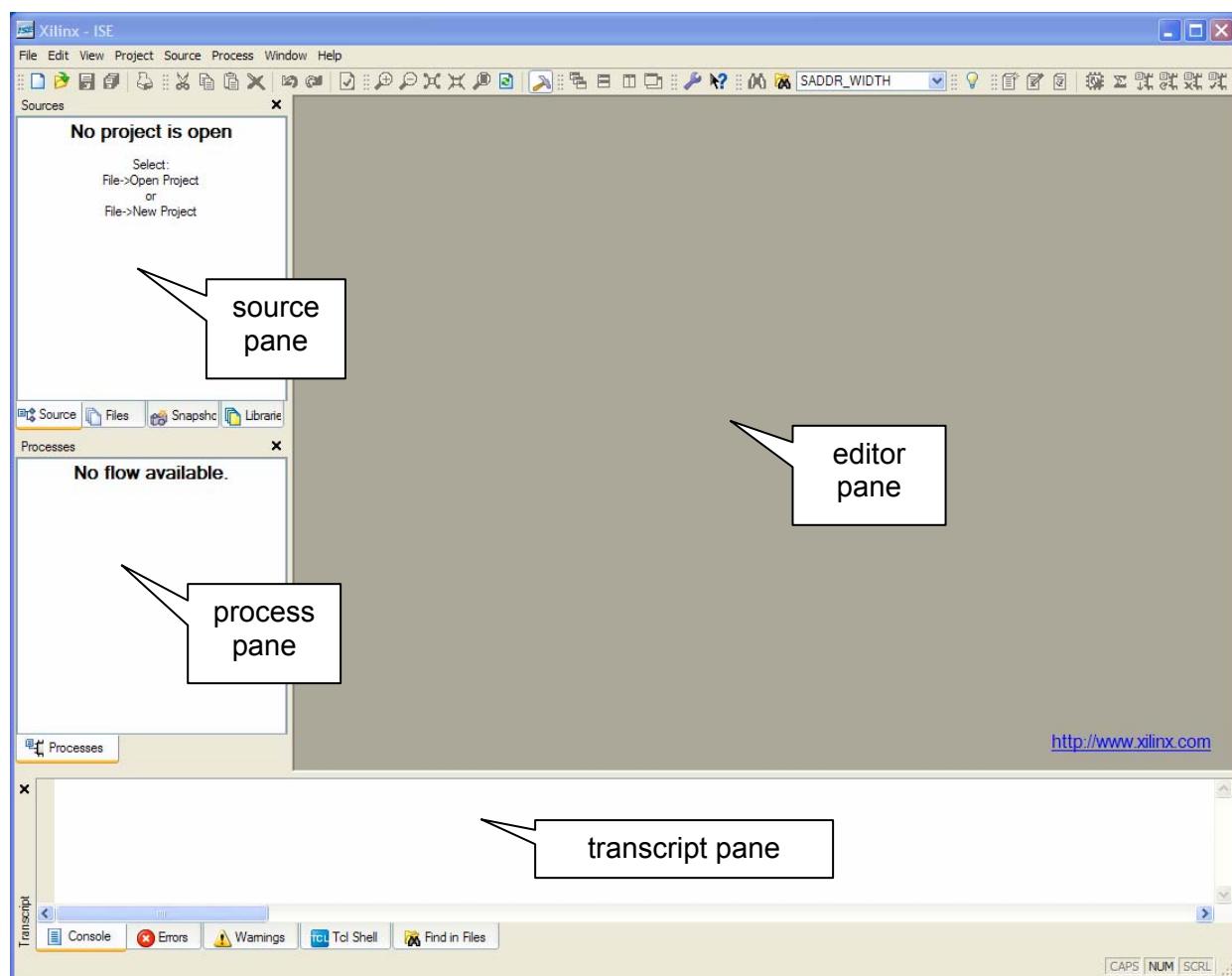


Starting ISE Project Navigator



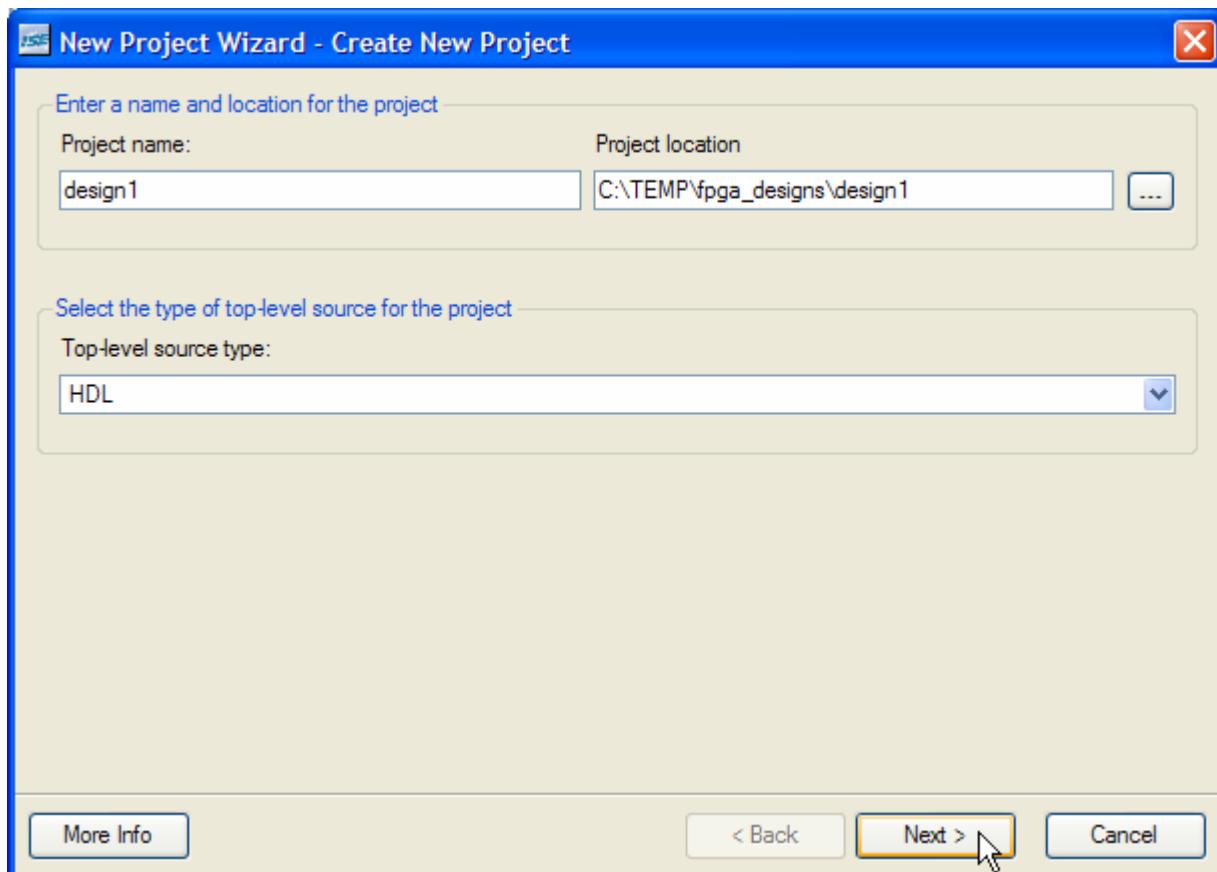
You start ISE by double-clicking the **Xilinx ISE 10.1** icon on the desktop. This will bring up an empty project window as shown below. The window has four panes:

1. A **source pane** that shows the organization of the source files that make up your design. There are four tabs so you can view the functional modules, source files, different snapshots (or versions) of your project, or the HDL libraries for your project.
2. A **process pane** that lists the various operations you can perform on a given object in the source pane.
3. A **transcript pane** that displays the various messages from the currently running process.
4. An **editor pane** where you can enter HDL code, schematics, state diagrams, etc.



To start your design, create a new project by selecting the **File→New Project** item from the menu bar. This brings up the **New Project Wizard** window where you can enter the name of your project, the location of your project files, and the style in which you will describe your design at the top level. I have given my project the descriptive name of **design1** and will place the files

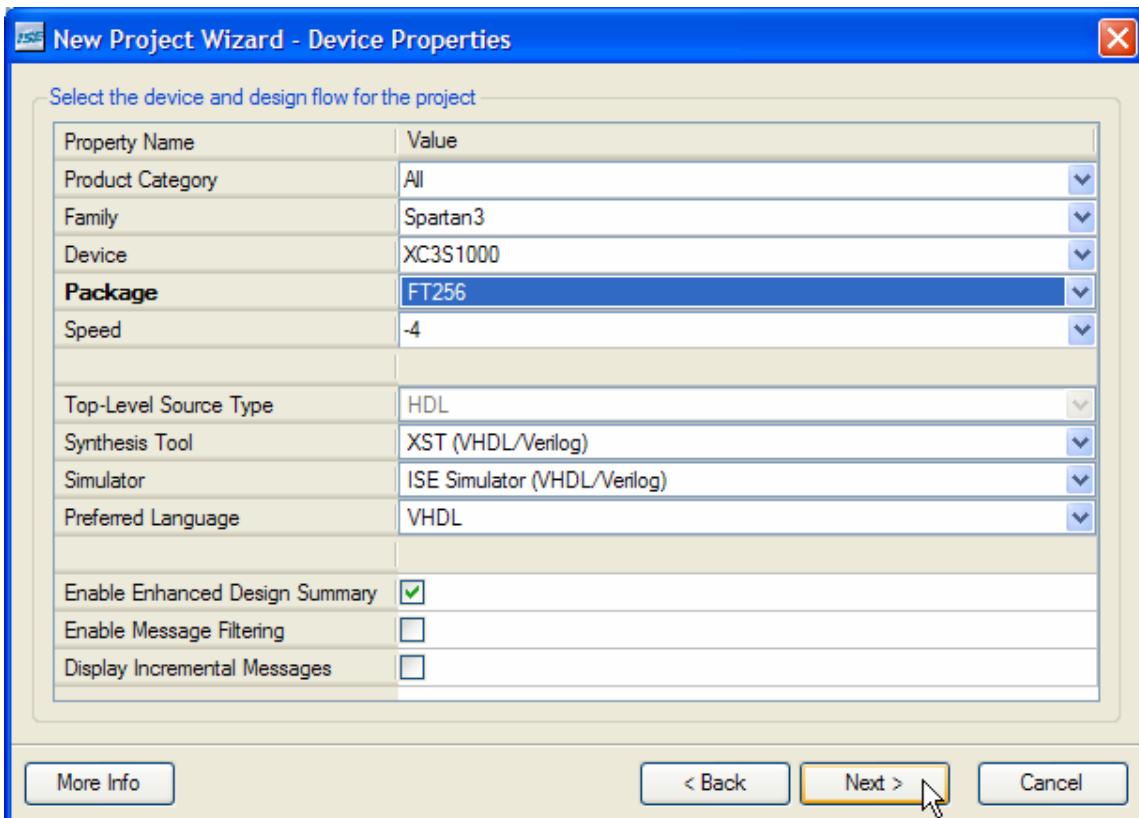
in the C:\TEMP\fpga_designs\design1 folder. (You may choose differently.) I am going to describe the LED decoder using VHDL, so I have set the top-level type (there will only be one level) to HDL (which would also be used if the design was done with Verilog). Click Next to continue creating your project.



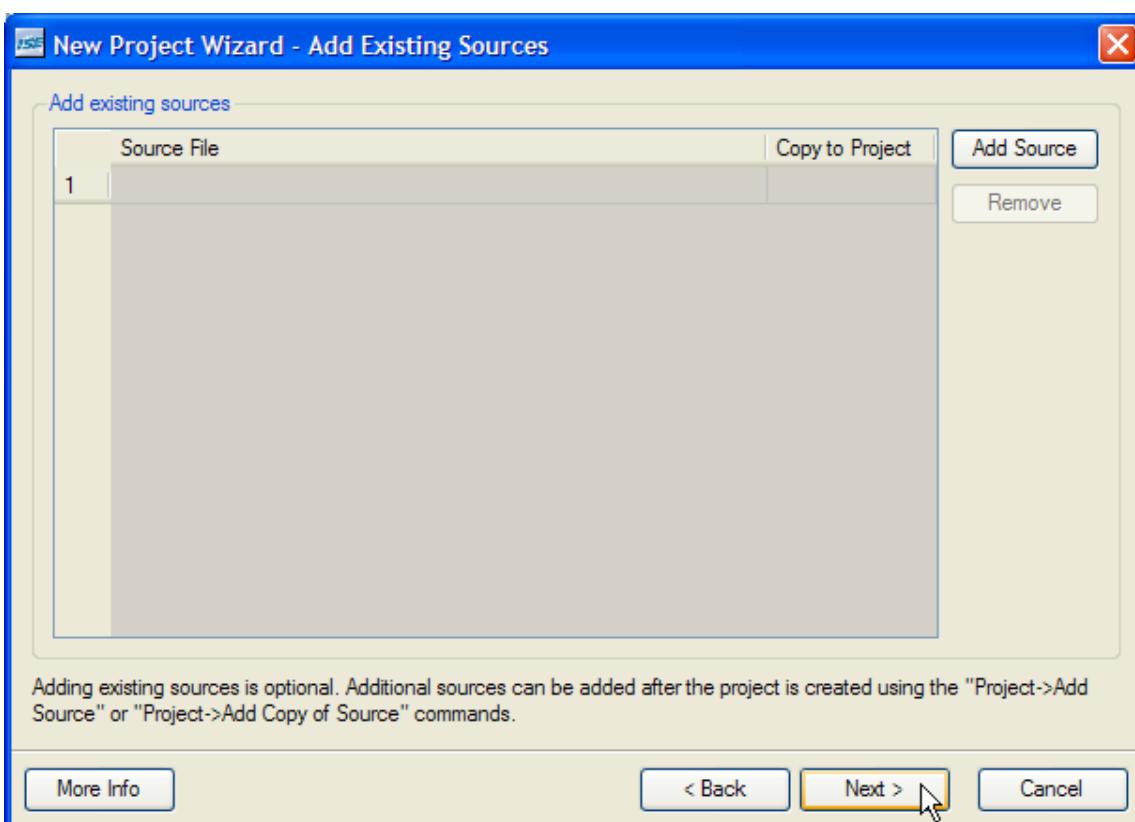
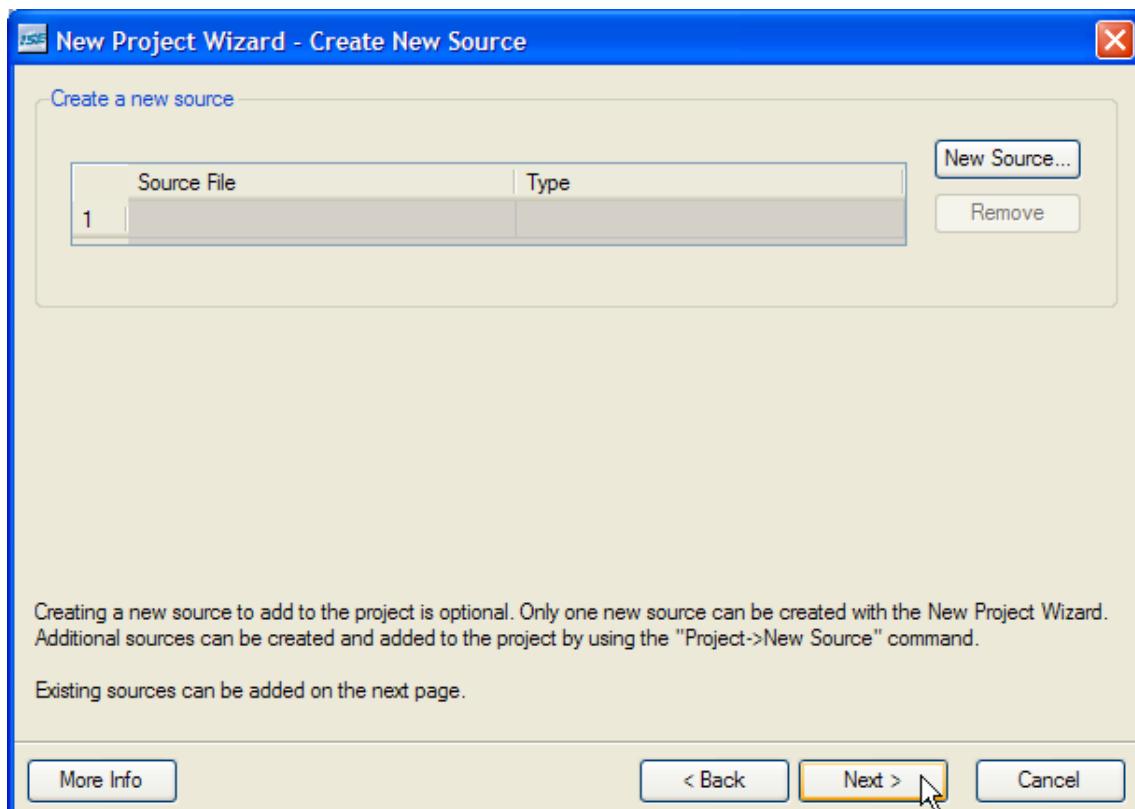
Now you need to tell ISE what FPGA you are going to use for your design. The device family, family member, package and speed grade for the FPGA on each model of XSA Board are shown below.

XSA Board	Device Family	Device	Package	Speed Grade
XSA-50	Spartan2	XC2S50	TQ144	-5
XSA-100	Spartan2	XC2S100	TQ144	-5
XSA-200	Spartan2	XC2S200	FG256	-5
XSA-3S1000	Spartan3	XC3S1000	FT256	-4

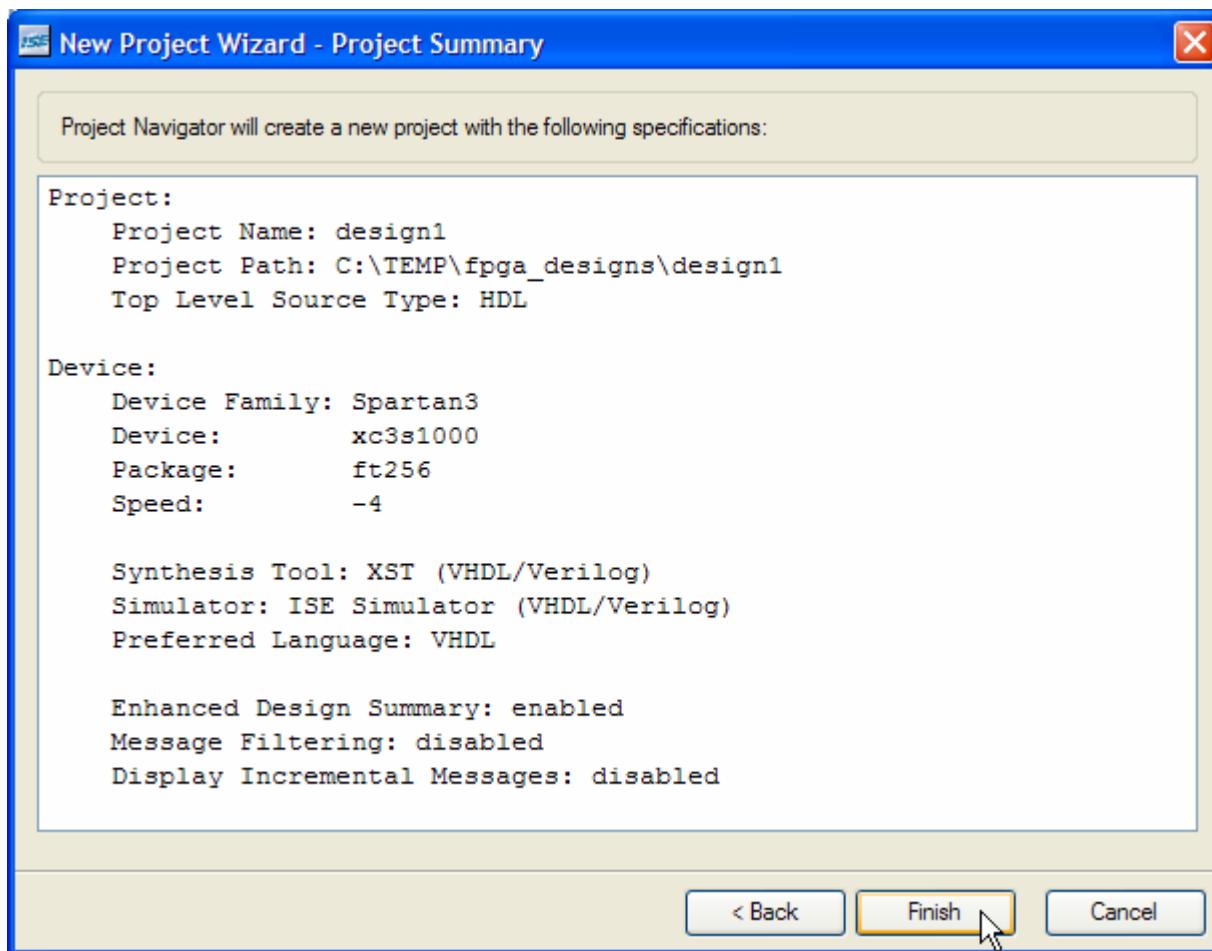
For this tutorial, I will target my design to the XSA-3S1000 Board so I have set the Value field of the Family, Device, Package and Speed properties as shown below. (Set these fields to whatever values are appropriate for your particular board using the table shown above.) The other fields can be left at their default values, so you can just click on the Next button to continue creating the project.



Click the Next button in the following two windows for creating or adding source files. (You will create the VHDL source code for the LED decoder at a later step.)

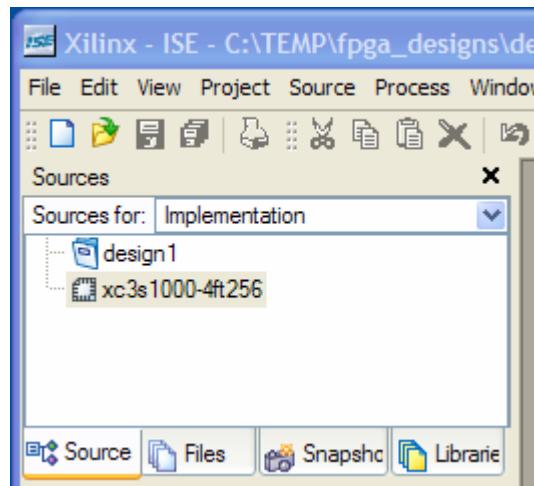


The final screen shows the pertinent information for the new project. Click on the Finish button to complete the creation of the project.



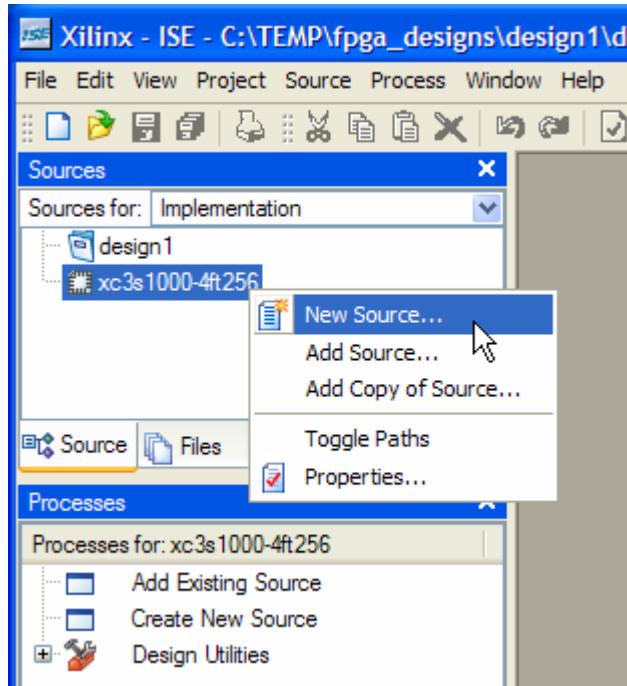
Now the Sources pane contains two items:

1. A project object called design1.
2. A chip object called xc3s1000-4ft256.

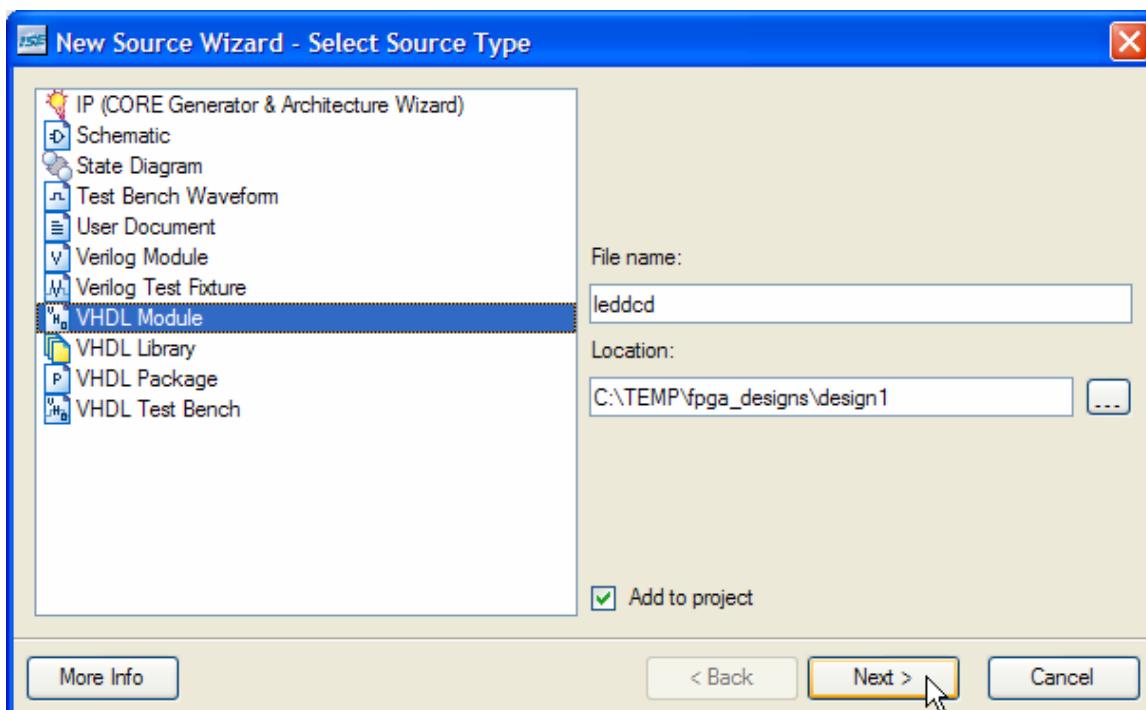


Describing Your Design With VHDL

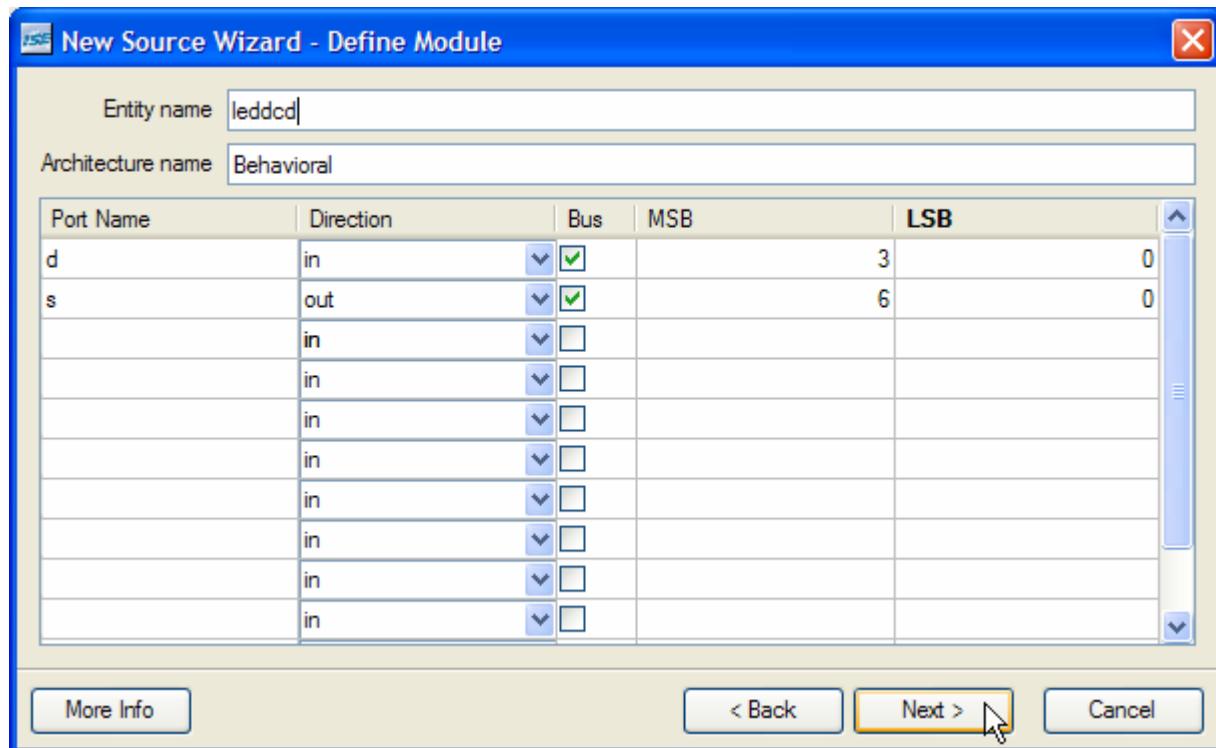
Once all the project set-up is complete, you can begin to actually design your LED decoder circuit. Start by adding a VHDL file to the **design1** project. Right-click on the xc3s1000-4ft256 object in the Sources pane and select New Source ... from the pop-up menu as shown below.



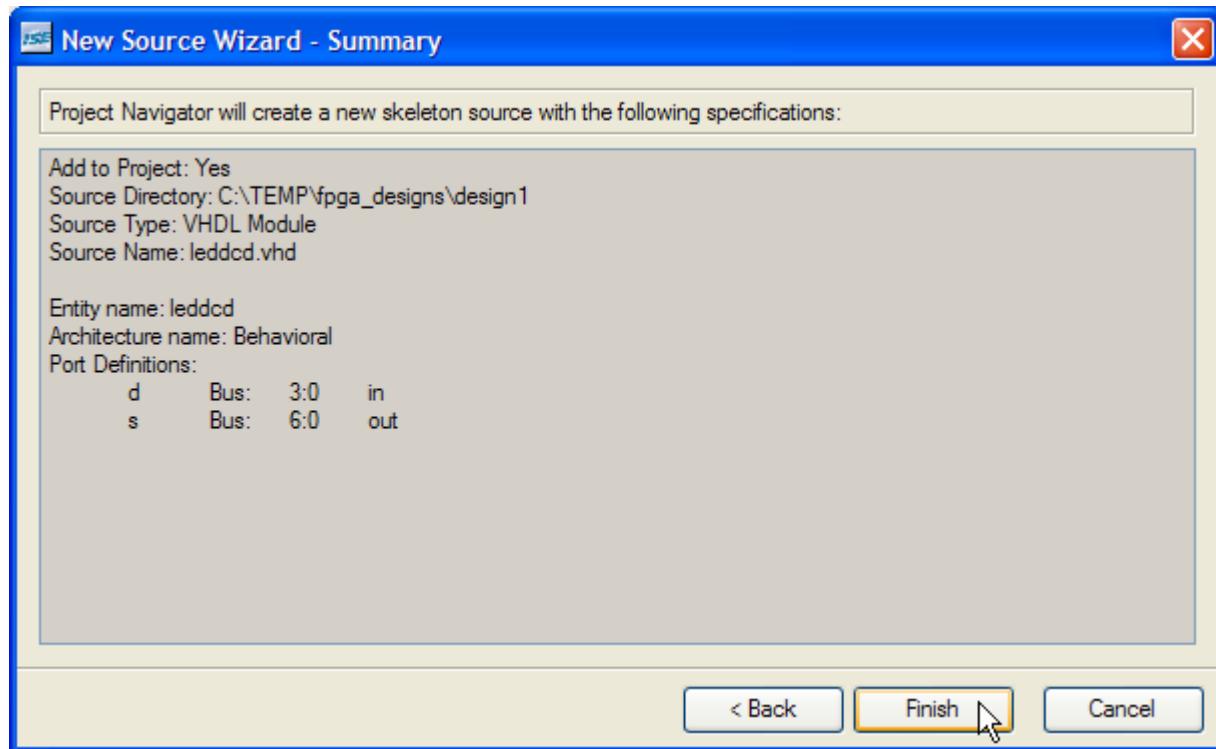
A window appears where you must select the type of source file you want to add. Since you are describing the LED decoder with VHDL, highlight the VHDL Module item. Then type the name of the module, **leddcd**, into the File name field and click on Next.



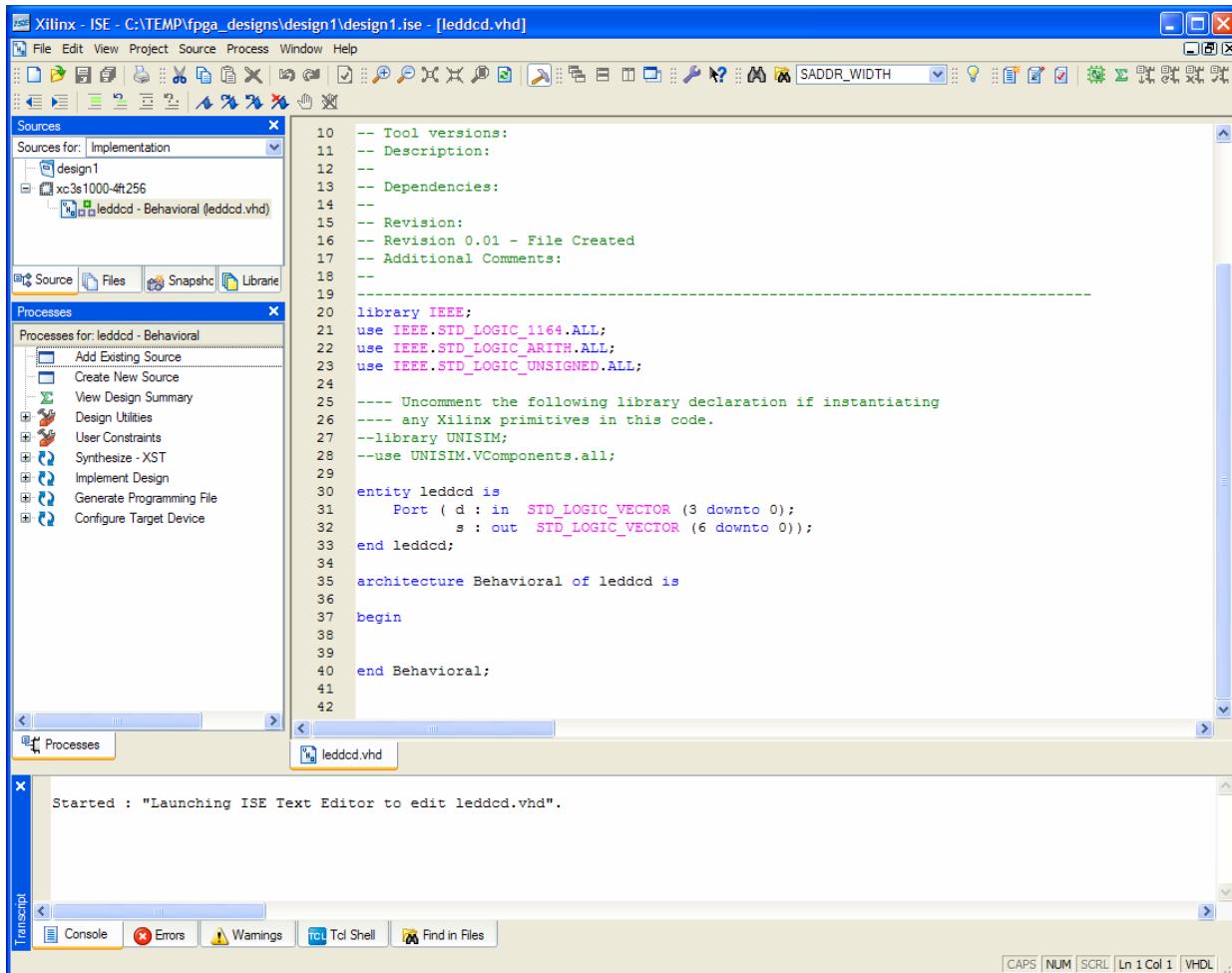
The **Define Module** window now appears where you can declare the inputs and outputs to the LED decoder circuit. In the first row, click in the Port Name field and type in **d** (the name of the inputs to the LED decoder). The **d** input bus has a width of four, so check the Bus box and type 3 in the MSB field while leaving 0 in the LSB field. Perform the same operations to create the seven-bit wide **s** output bus that drives the LEDs.



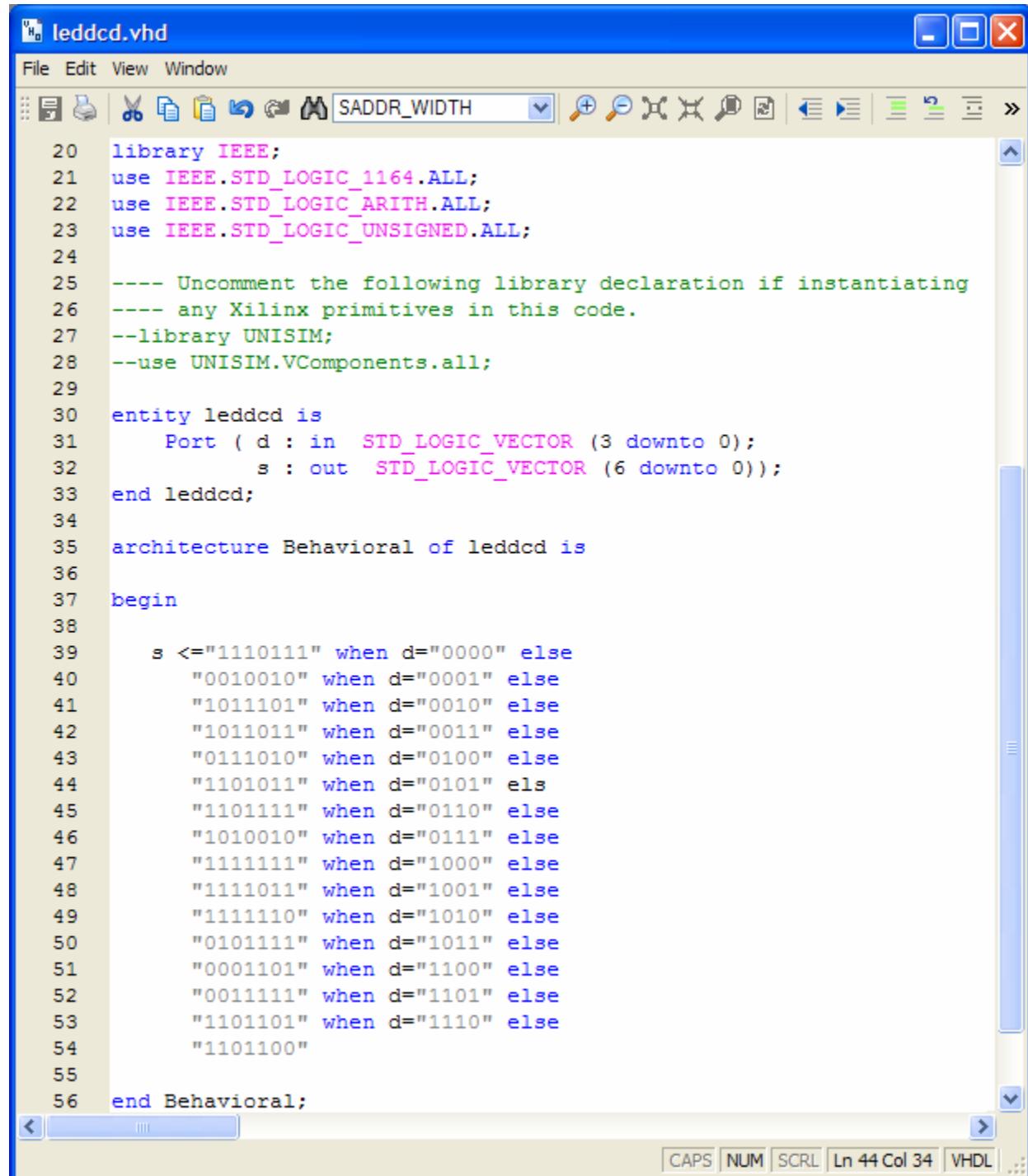
Click on Next in the **Define Module** window to get a summary of the information you just typed in:



After clicking on Finish, the editor pane displays a design summary and a VHDL skeleton for the LED decoder. (You can also see the ledgcd.vhd file has been added to the Sources pane.) Click on the ledgcd tab at the bottom of the editor pane and then scroll to the bottom of the VHDL skeleton. Lines 20-23 create links to the IEEE library and packages that contain various useful definitions for describing a design. The LED decoder inputs and outputs are declared in the VHDL entity on lines 30-33. You will describe the logic operations of the decoder in the architecture section between lines 37 and 40.



The completed VHDL file for the LED decoder is shown below. The architecture section contains a single statement which assigns a particular seven-bit pattern to the **s** output bus for any given four-bit input on the **d** bus (lines 39-54).



```

library IEEE;
use IEEE.STD_LOGIC_1164.ALL;
use IEEE.STD_LOGIC_UNSIGNED.ALL;
use IEEE.STD.TEXT.all;

entity ledcd is
    Port ( d : in STD_LOGIC_VECTOR (3 downto 0);
           s : out STD_LOGIC_VECTOR (6 downto 0));
end ledcd;

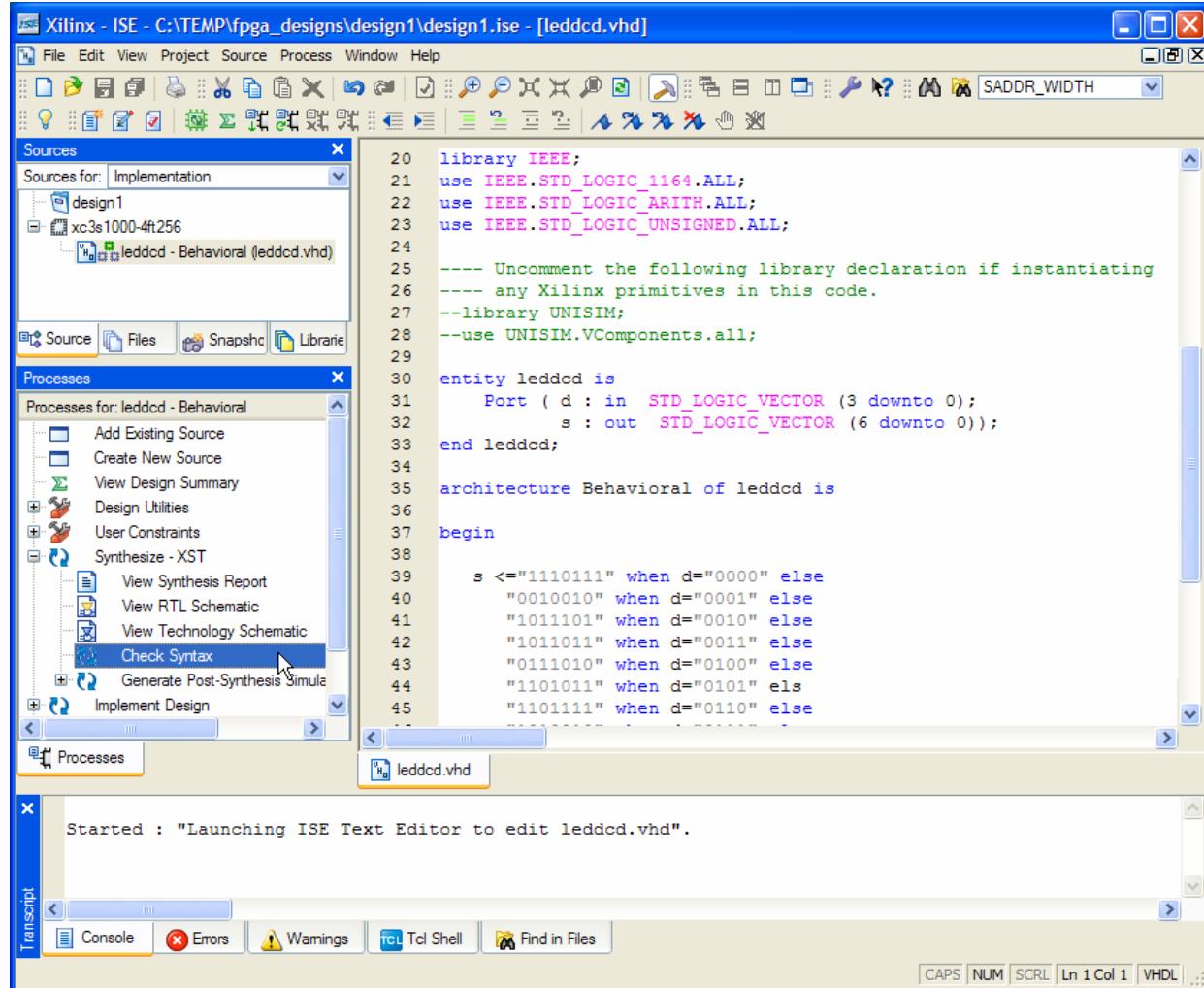
architecture Behavioral of ledcd is
begin
    s <="1110111" when d="0000" else
        "0010010" when d="0001" else
        "1011101" when d="0010" else
        "1011011" when d="0011" else
        "0111010" when d="0100" else
        "1101011" when d="0101" else
        "1101111" when d="0110" else
        "1010010" when d="0111" else
        "1111111" when d="1000" else
        "1111011" when d="1001" else
        "1111110" when d="1010" else
        "0101111" when d="1011" else
        "0001101" when d="1100" else
        "0011111" when d="1101" else
        "1101101" when d="1110" else
        "1101100";
end Behavioral;

```

Once the VHDL source is entered, click on the button to save it in the ledcd.vhd file.

Checking the VHDL Syntax

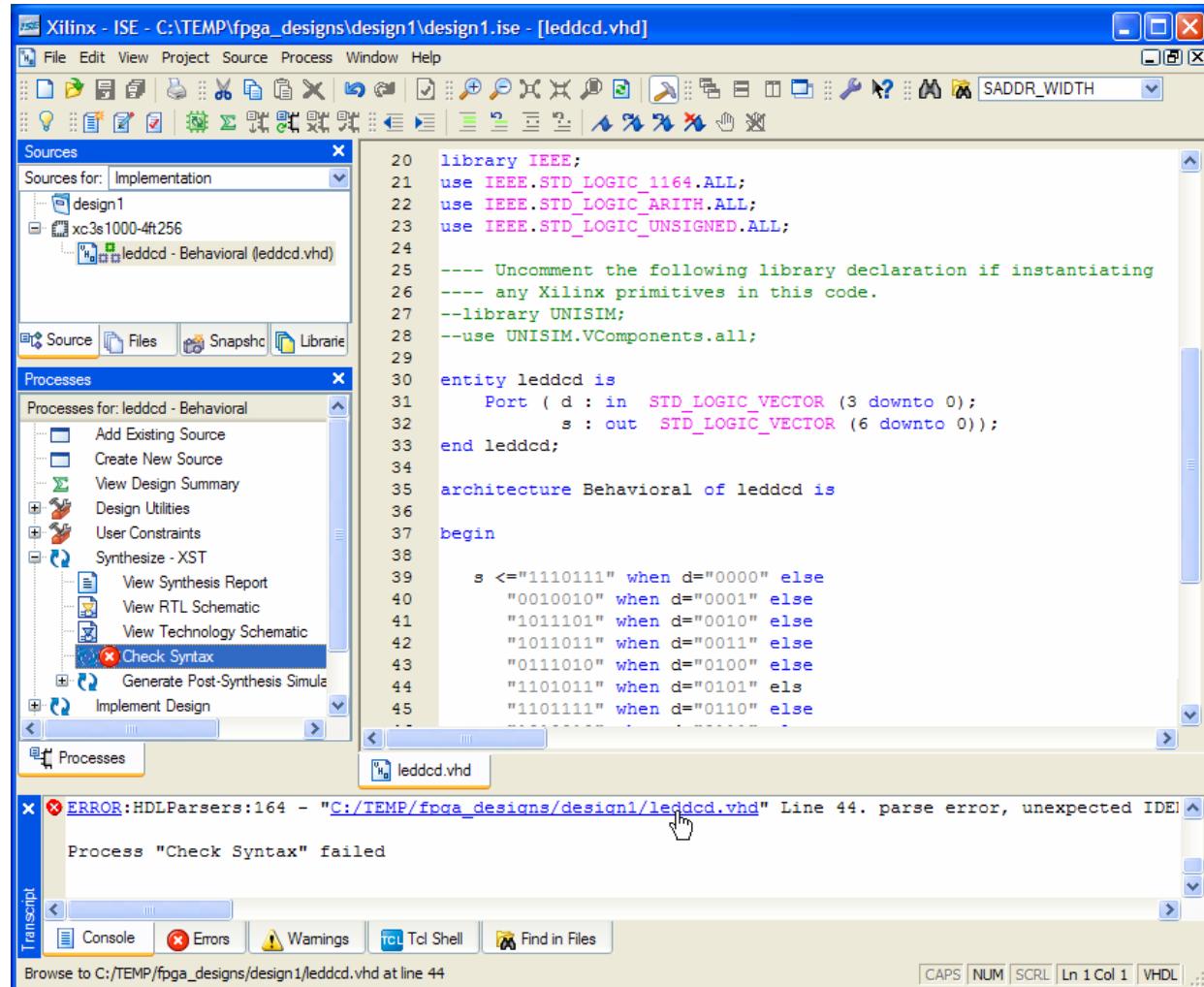
You can check for errors in our VHDL by highlighting the ledgcd object in the Sources pane and then double-clicking on Check Syntax in the Process pane as shown below.



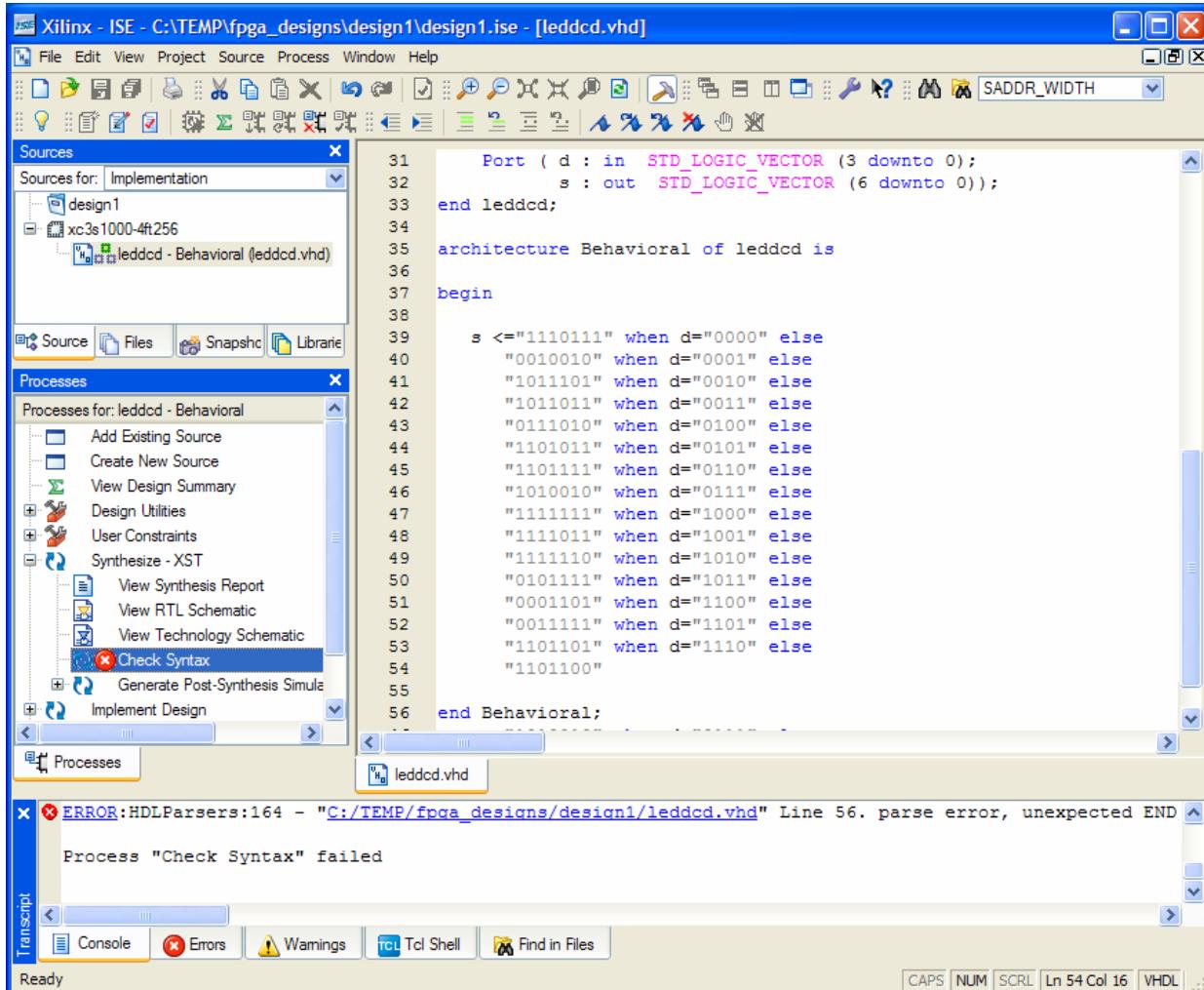
The syntax checking tool grinds away and then displays the result in the process pane. In this case, an error was found as indicated by the next to the Check Syntax process. But what is the error and where is it?

Fixing VHDL Errors

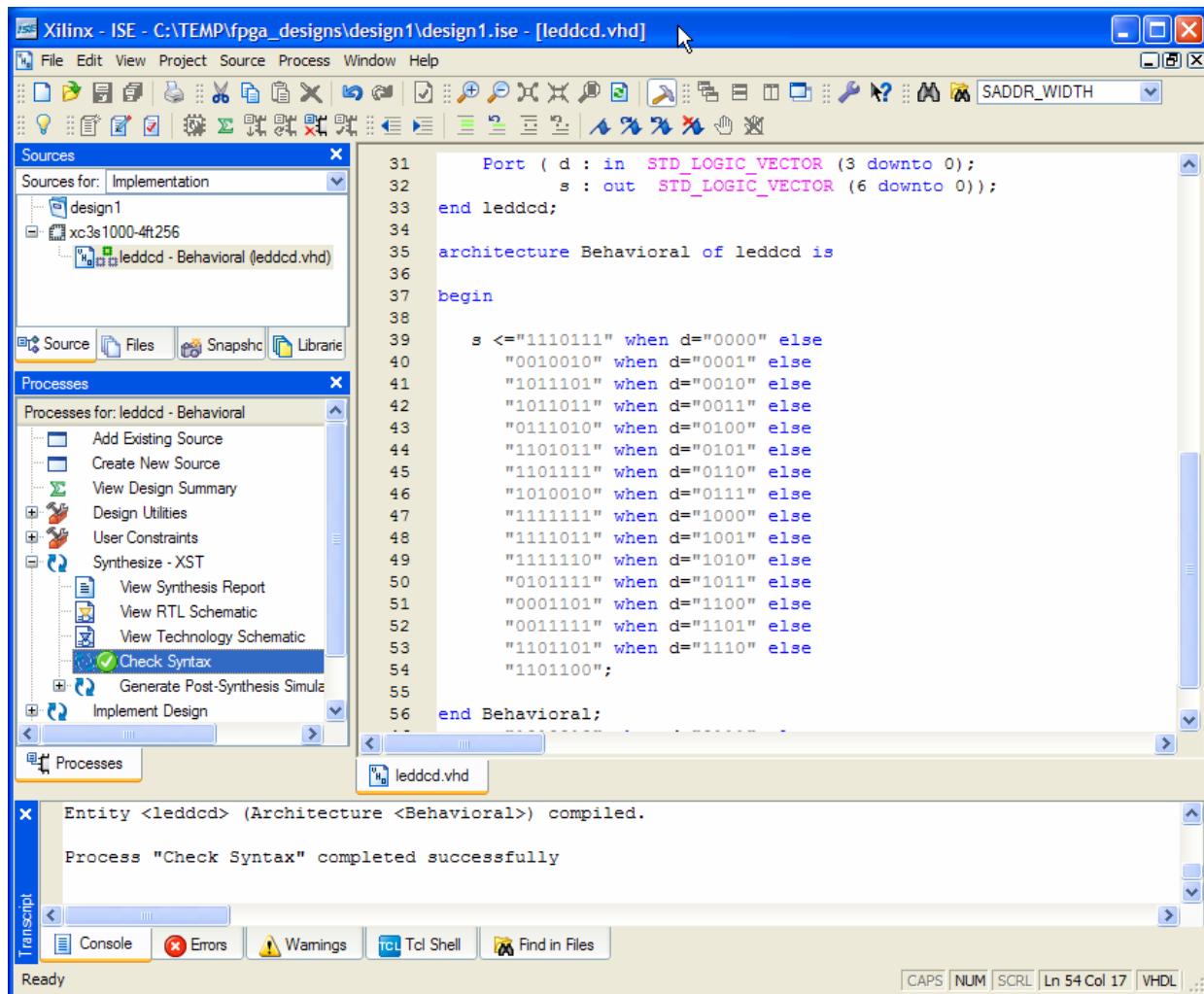
You can find the location of the error by clicking on the Errors tab at the bottom of the transcript pane. In this case, the error is located on line 44 and you can manually scroll there. You can also click on the error message in the log pane to go directly to the erroneous source. (This is most useful in more complicated projects consisting of multiple source files.) You can also get a more detailed explanation of the error by clicking on the [ERROR](#) hyperlink at the beginning of the message.



On line 44, you can see that the 'e' was left off the end of the `else` keyword. After correcting this error and saving the file, double-click the on Check Syntax in the Process pane to re-check the VHDL code. The syntax checker now finds another error on line 56 of the VHDL code.

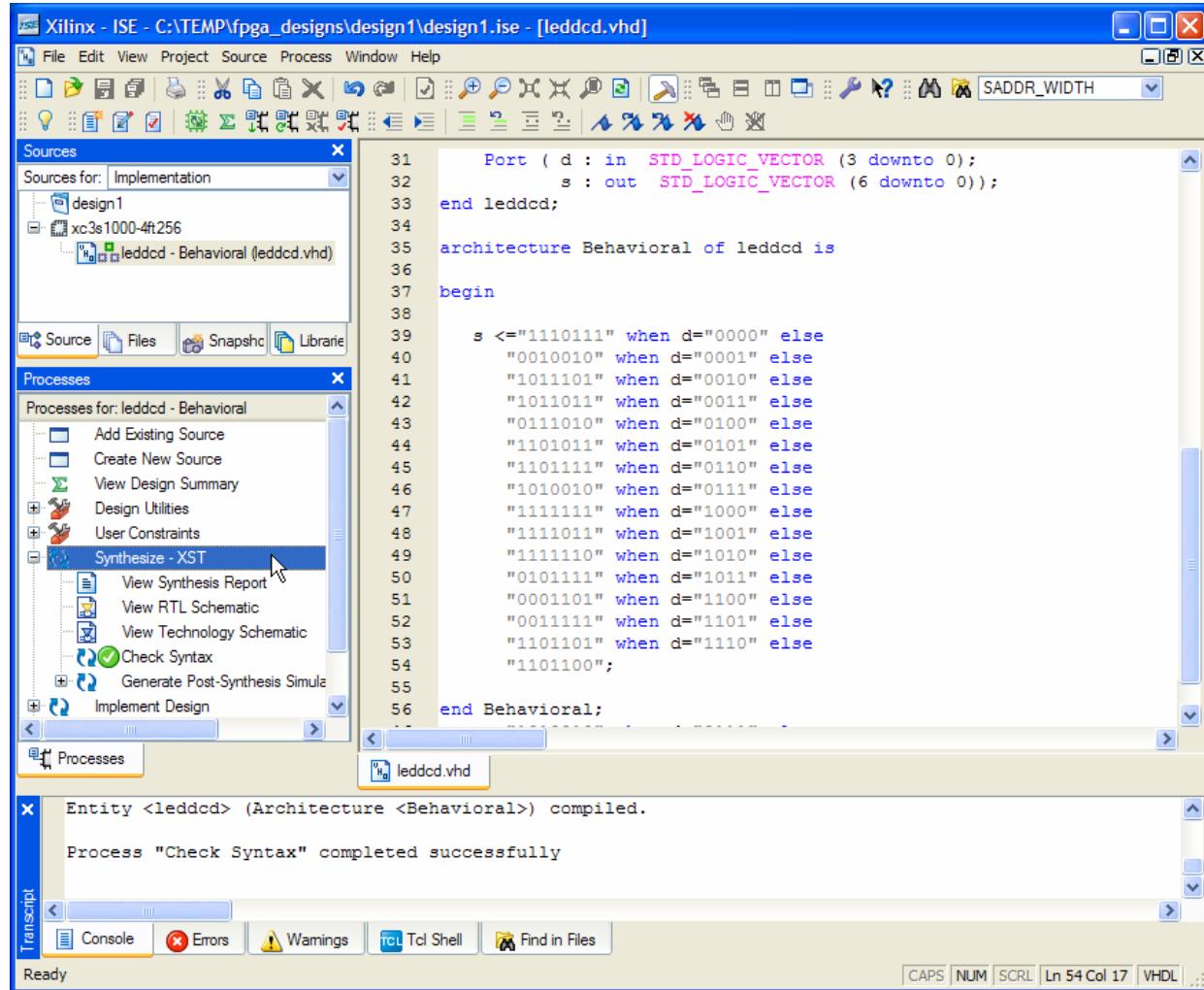


Examining line 56, you can see it is just the end statement for the architecture section. The actual error occurred on line 54. The VHDL syntax checker was expecting to find a ';' but it is missing. After adding the semicolon and saving the file, the Check Syntax process runs without errors and displays a  .



Synthesizing the Logic circuitry for Your Design

Now that you have valid VHDL for your design, you need to convert it into a logic circuit. This is done by highlighting the ledcd object in the Sources pane and then double-clicking on the Synthesize-XST process as shown below.

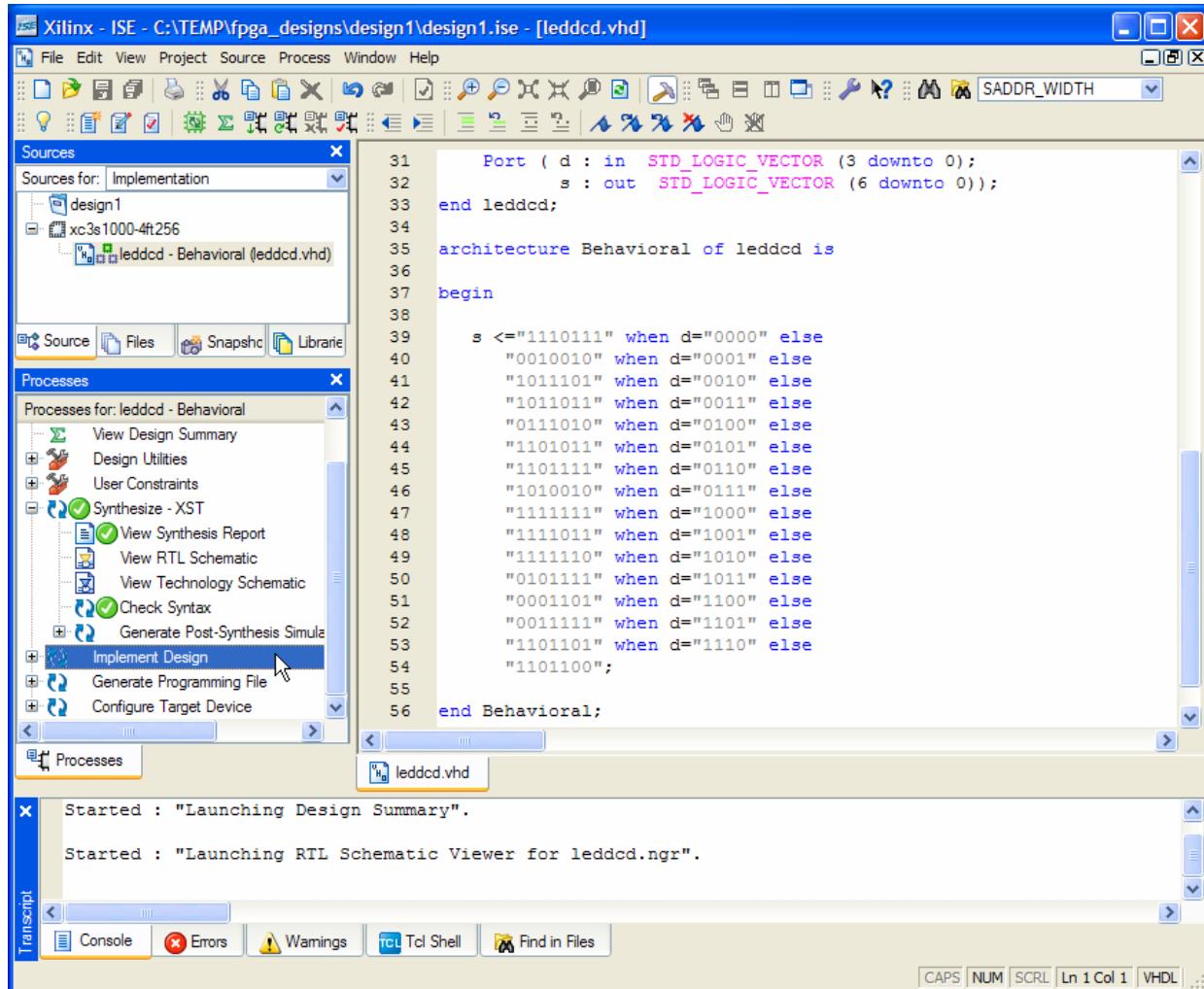


The synthesizer will read the VHDL code and transform it into a netlist of gates. This will take less than a minute. If no problems are detected, a will appear next to the Synthesize process. You can double-click on the View Synthesis Report to see the various synthesizer options that were enabled and some device utilization and timing statistics for the synthesized design. (The device utilization is also viewable on the Design Summary tab in the editor pane that appears when you click on the Project→View Design Summary menu item.) You can also double-click on View RTL Schematic to see the schematic that was derived from the VHDL source code, but it's not very interesting in this case.

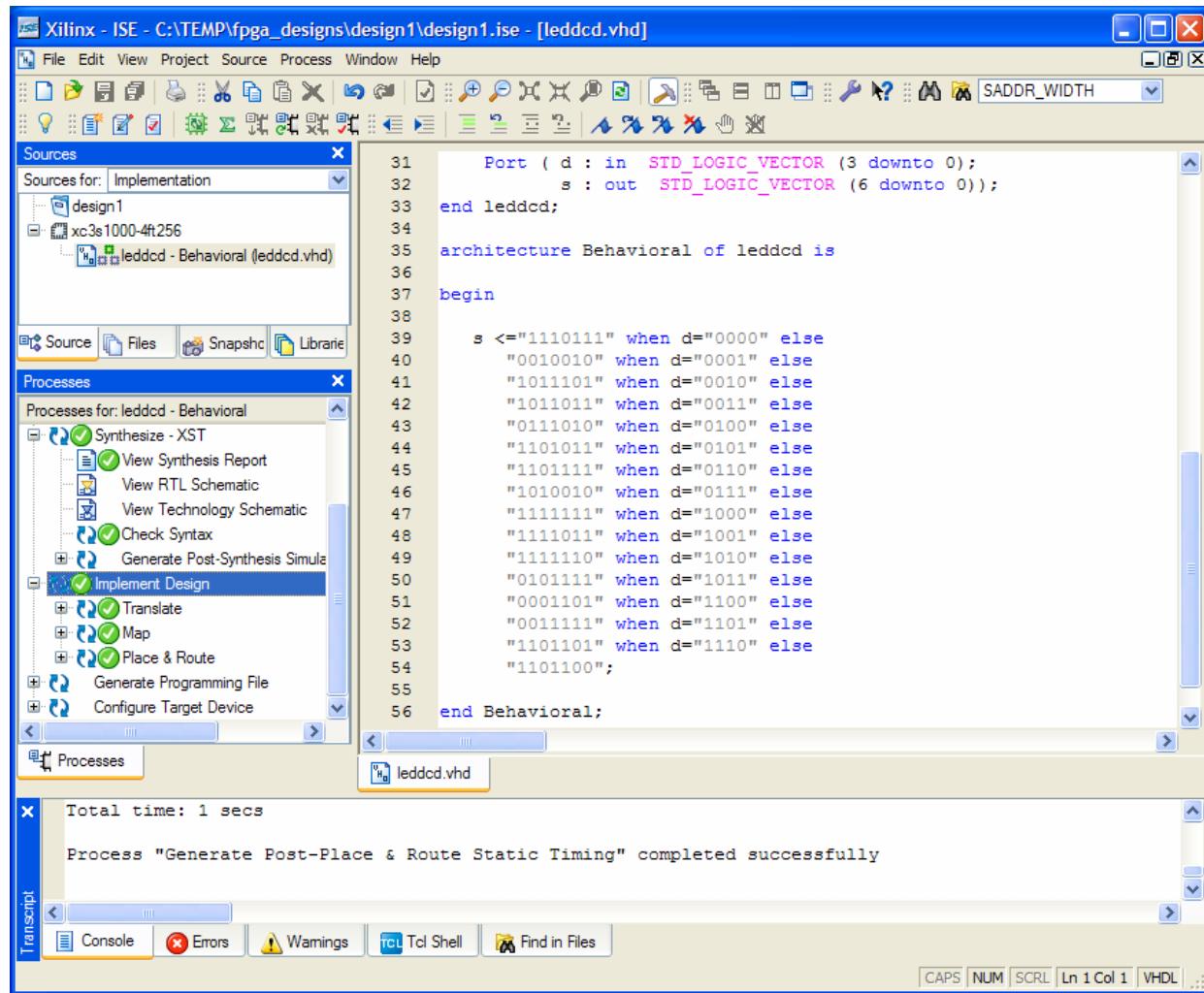
Implementing the Logic Circuitry in the FPGA

You now have a synthesized logic circuit for the LED decoder, but you need to translate, map and place & route it into the logic resources of the FPGA in order to actually use it. Start this

process by highlighting the ledcdc object in the Sources pane and then double-click on the Implement Design process.

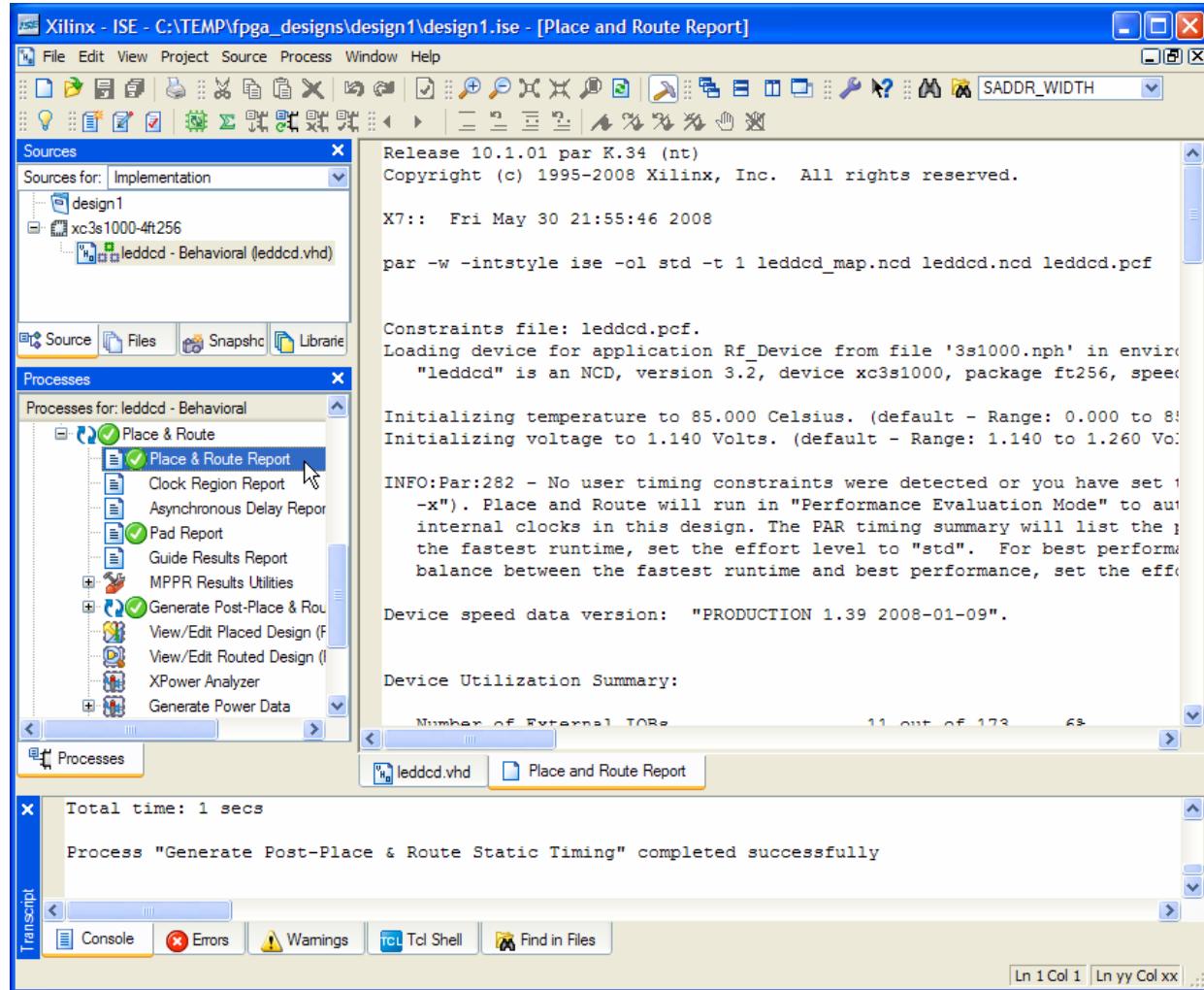


You can watch the progress of the implementation process in the console tab of the transcript pane. For a simple design this, the implementation is completed in less than 30 seconds (on a 2.0 GHz Athlon 64 X2 PC with 3 Gbytes of RAM). A successful implementation is indicated by the next to the Implement Design process. You can expand the Implement Design process to see the subprocesses within it. The Translate process converts the netlist output by the synthesizer into a Xilinx-specific format and annotates it with any design constraints you may specify (more on that later). The Map process decomposes the netlist and rearranges it so it fits nicely into the circuitry elements contained in the specified FPGA device. Then the Place & Route process assigns the mapped elements to specific locations in the FPGA and sets the switches to route the logic signals between them. If the Implement Design process had failed, a would appear next to the subprocess where the error occurred. You may also see a that indicates a successful completion but some warnings were issued or not all the subprocesses were enabled.



Checking the Implementation

You have your design fitted into the FPGA, but how much of the chip does it use? Which pins are the inputs and outputs assigned to? You can find answers to these questions by double-clicking on the Place & Route Report and the Pad Report in the Process pane.



The device utilization of the LED decoder circuit can be found near the top of the place & route report (or in the Design Summary tab). The circuit only uses four of the 7680 available slices in the XC3S1000 FPGA. Each slice contains two CLBs and each CLB can compute the logic function for one LED segment output.

Device utilization summary:

Number of External IOBs	11 out of 173	6%
Number of LOCed External IOBs	0 out of 11	0%
Number of Slices	4 out of 7680	1%
Number of SLICEMs	0 out of 3840	0%

The pad report shows what pins the LED decoder inputs and outputs use to enter and exit the FPGA. (The pad report was edited to remove unused pins and fields so it would fit into this document.)

Pin Number	Signal Name	Pin Usage	Direction	IO Standard
A7	s<4>	IOB	OUTPUT	LVCMOS25
A8	s<0>	IOB	OUTPUT	LVCMOS25
B6	s<6>	IOB	OUTPUT	LVCMOS25
B7	d<0>	IOB	INPUT	LVCMOS25
B8	s<1>	IOB	OUTPUT	LVCMOS25
C6	d<3>	IOB	INPUT	LVCMOS25
C7	d<2>	IOB	INPUT	LVCMOS25
C8	s<2>	IOB	OUTPUT	LVCMOS25
D7	s<5>	IOB	OUTPUT	LVCMOS25
D8	d<1>	IOB	INPUT	LVCMOS25
E7	s<3>	IOB	OUTPUT	LVCMOS25

Assigning Pins with Constraints

The problem now is that the inputs and outputs for the LED decoder were assigned to pins picked by the implementation process, but these are not the pins you actually want to use on the FPGA. You want the inputs assigned to pins on the FPGA that you can force high and low so as to test the LED decoder operation for each possible input pattern. In addition, the outputs should be attached to a seven-segment LED to make it easy to verify the correct operation of the design.

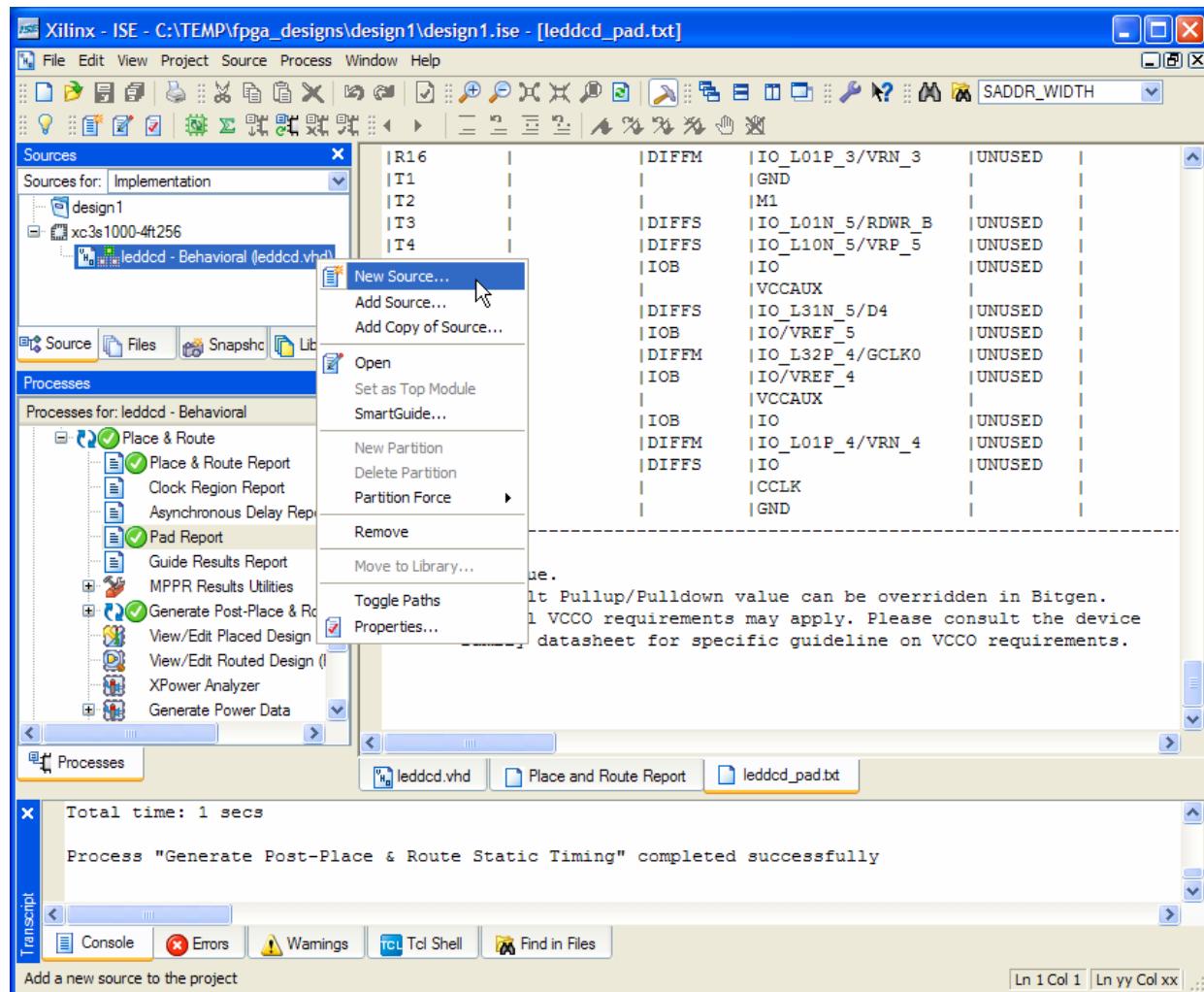
The GXSPORT utility lets you set the levels on the eight data outputs of the PC parallel port. The parallel port data pins attach to a group of eight specific pins on the FPGA of each model of XSA Board. You should assign the LED decoder inputs to four of these so that you can control the inputs using GXSPORT. The four pins I selected from the group of eight on each XSA Board are shown below:

LED Decoder Input	XSA-50	XSA-100	XSA-200	XSA-3S1000
d0	P50	P50	E13	N14
d1	P48	P48	C16	P15
d2	P42	P42	E14	R16
d3	P47	P47	D16	P14

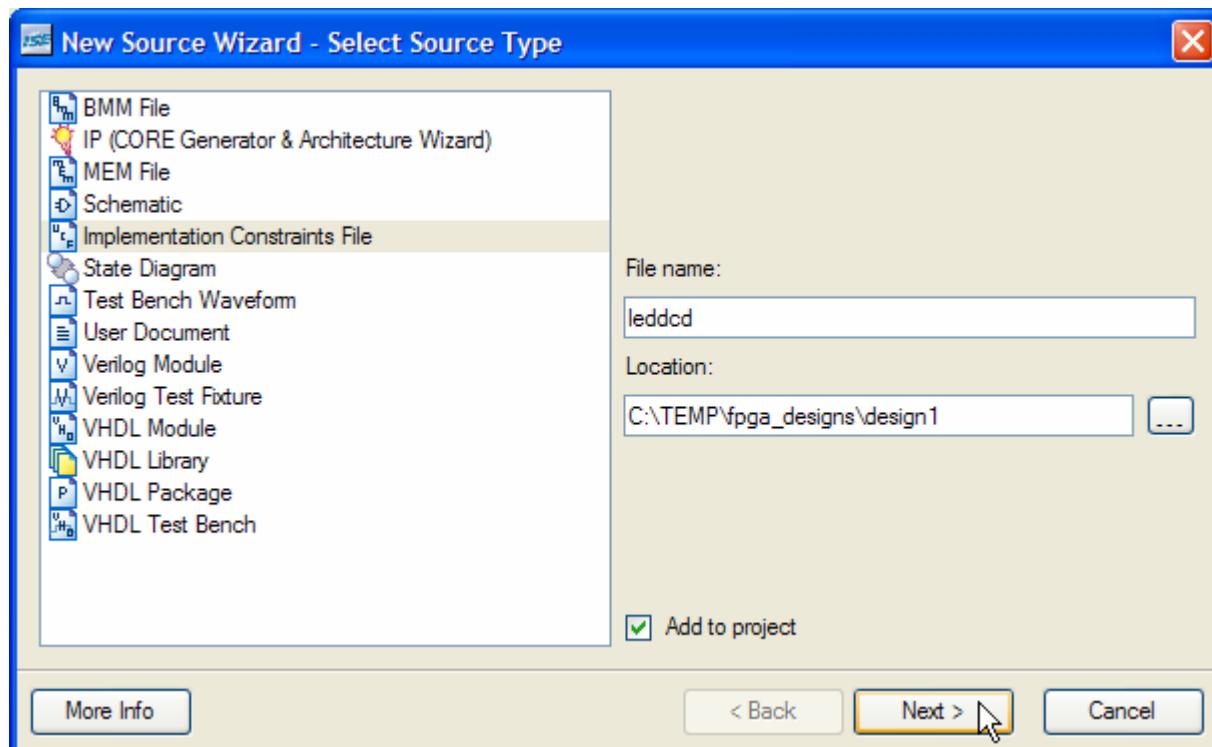
Likewise, each XSA Board has a seven-segment LED attached to the following pins of the FPGA:

LED Decoder Output	XSA-50	XSA-100	XSA-200	XSA-3S1000
s0	P67	P67	N14	M6
s1	P39	P39	D14	M11
s2	P62	P62	N16	N6
s3	P60	P60	M16	R7
s4	P46	P46	F15	P10
s5	P57	P57	J16	T7
s6	P49	P49	G16	R10

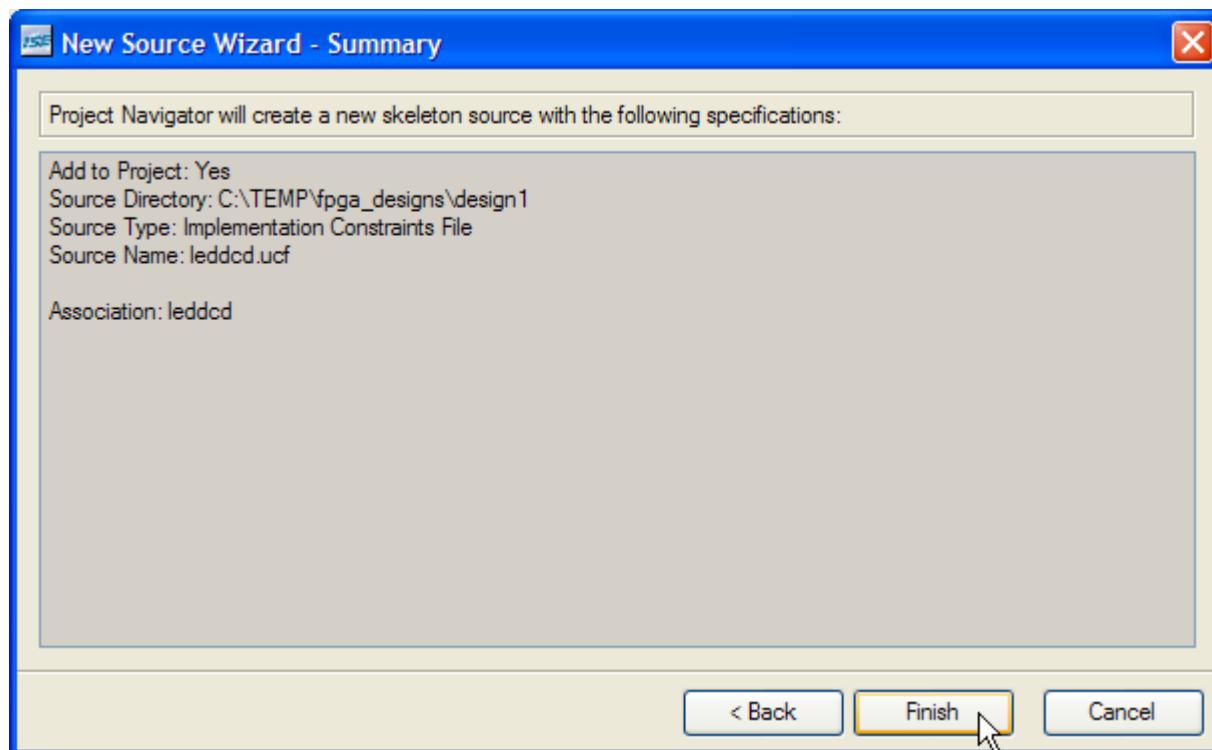
How do you direct the implementation process so it assigns the inputs and outputs to the pins you want to use? This is done by using *constraints*. In this case, you are constraining the implementation process so it assigns the inputs and outputs only to the pins shown in the previous tables. Start creating these constraints by right-clicking the ledcd object in the Sources pane and selecting New Source... from the pop-up menu.



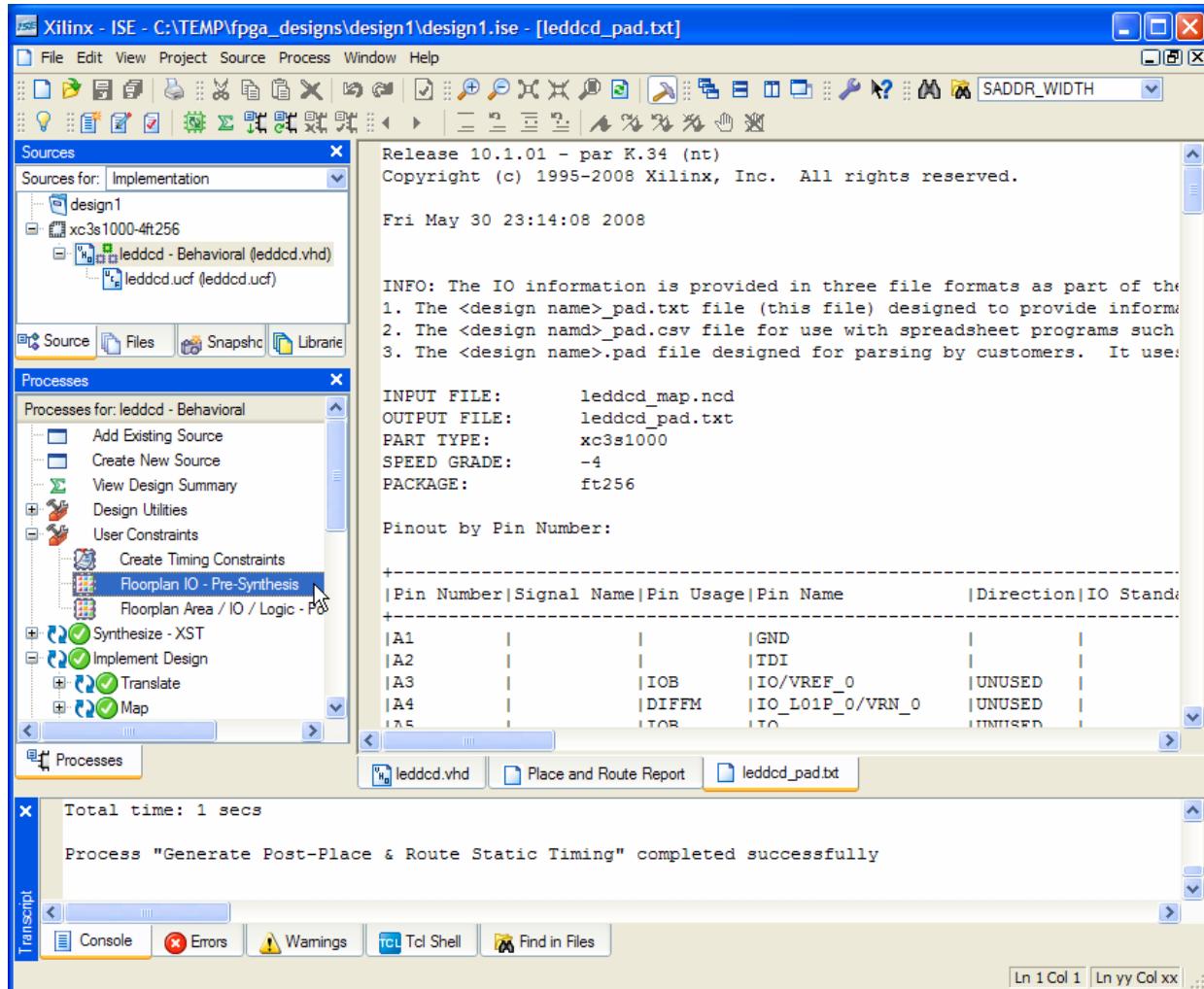
Select Implementation Constraints File as the type of source file you want to add and type `leddcd` in the File Name field. Then click on the Next button.



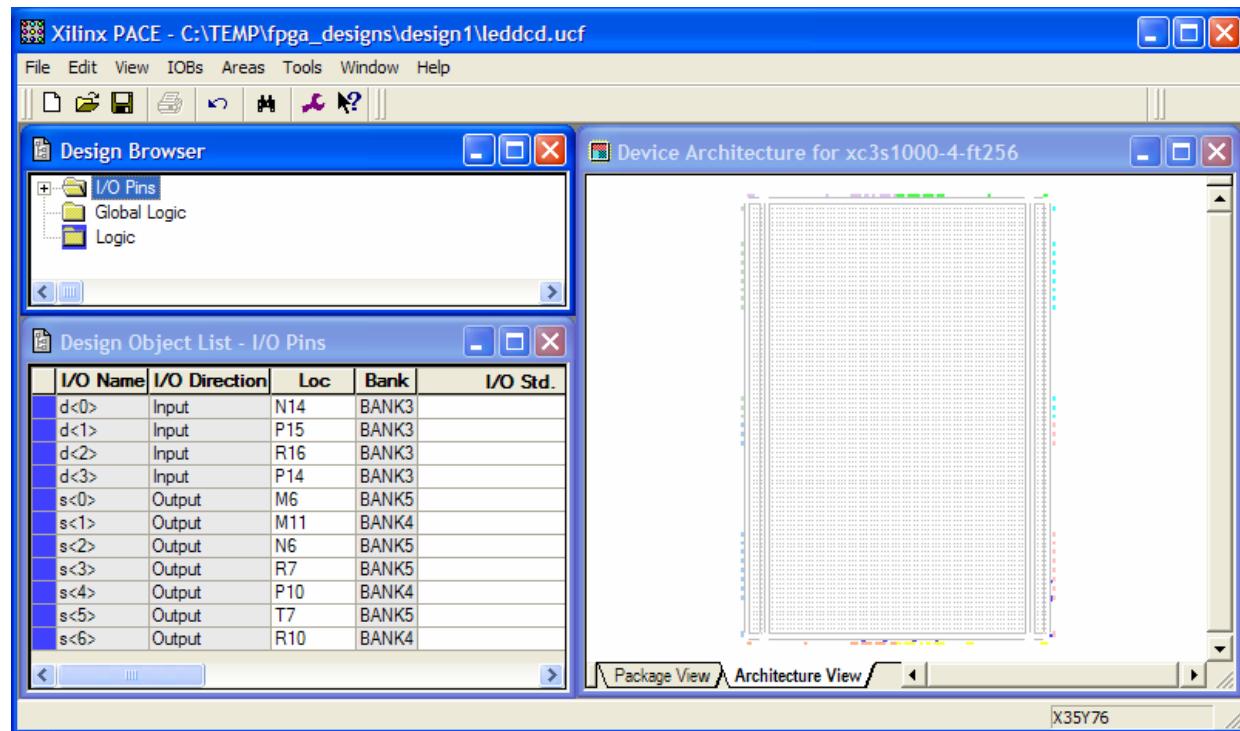
You will receive a feedback window that shows the name and type of the file you created and the file to which the constraints apply (`leddcd.vhd`). Click on the Finish button to complete the addition of the `leddcd.ucf` constraint file to this project.



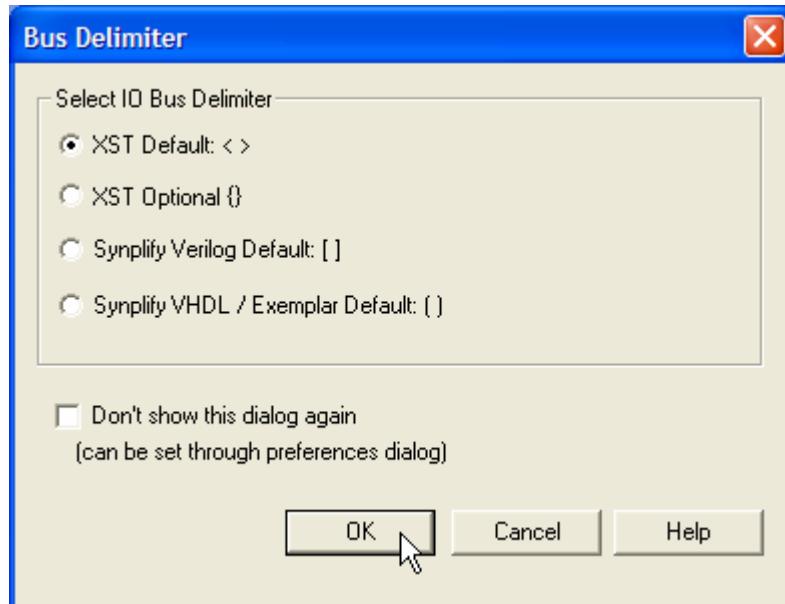
Now highlight the ledgcd object in the Sources pane and double-click the Floorplan IO – Pre-Synthesis item in the Process pane to begin adding pin assignment constraints to the design.



The **Xilinx PACE** window now appears. Click on the I/O Pins item in the Design Browser pane. A list of the current inputs and outputs for the LED decoder will appear in the Design Object List – I/O Pins pane. You can change your pin assignments here. You start by clicking in the Loc field for the **d<0>** input. Then just type in the pin assignment for this input: N14. Do this for all of the inputs and outputs using the pin assignments from the previous table.



After the pin assignments are entered, save the file. A dialog window will appear requesting that you select the delimiter for the I/O buses. Select <> since the XST synthesizer is being used for this project, and then click on OK. Then close the **Xilinx PACE** window.



Now you can re-implement your design by highlighting the ledcc object in the Sources pane and double-clicking on the Implement Design process. After the implementation process completes,

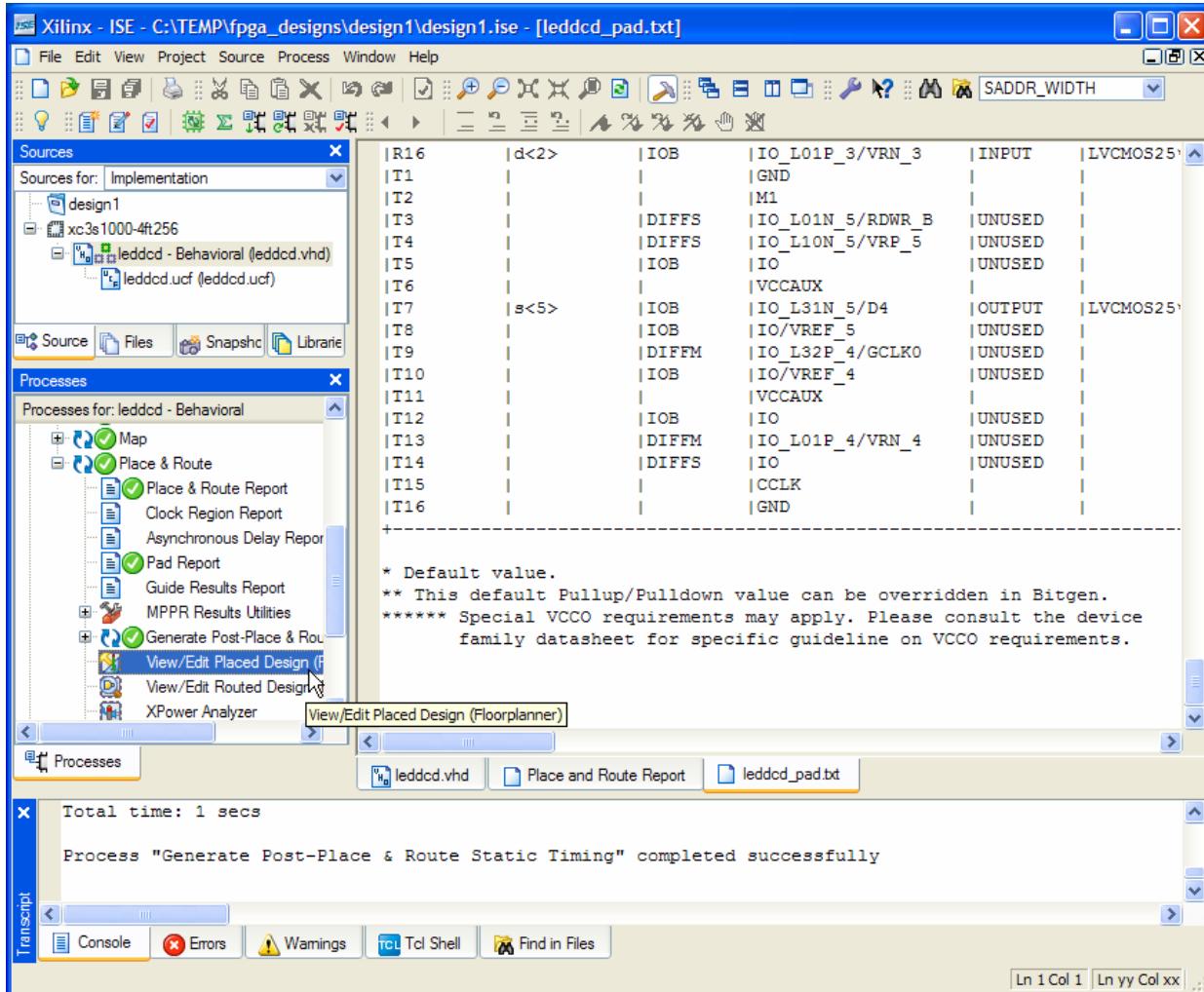
double-click on Pad Report to view the pin assignments. Now the pad report shows the following pin assignments:

Pin Number	Signal Name	Pin Usage	Direction	IO Standard
M6	s<0>	IOB	OUTPUT	LVCMOS25
M11	s<1>	IOB	OUTPUT	LVCMOS25
N6	s<2>	IOB	OUTPUT	LVCMOS25
N14	d<0>	IOB	INPUT	LVCMOS25
P10	s<4>	IOB	OUTPUT	LVCMOS25
P14	d<3>	IOB	INPUT	LVCMOS25
P15	d<1>	IOB	INPUT	LVCMOS25
R7	s<3>	IOB	OUTPUT	LVCMOS25
R10	s<6>	IOB	OUTPUT	LVCMOS25
R16	d<2>	IOB	INPUT	LVCMOS25
T7	s<5>	IOB	OUTPUT	LVCMOS25

The reported pin assignments match the assignments made in the **Xilinx Pace** window, so it appears the I/O have been constrained to the appropriate pins.

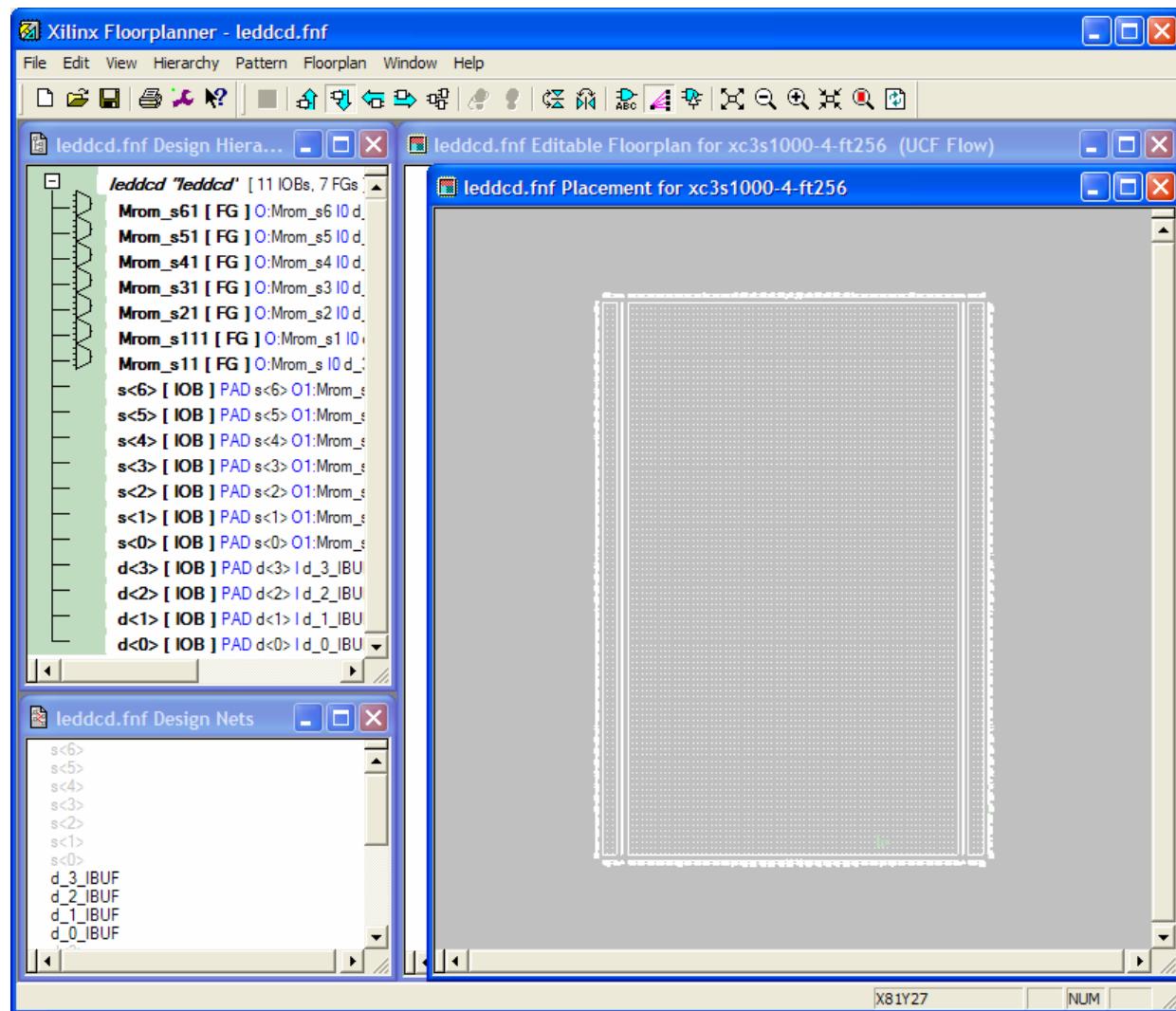
Viewing the Chip

After the implementation process completes, you can get a graphical depiction of how the logic circuitry and I/O are assigned to the FPGA CLBs and pins. Just highlight the ledcdc object in the Sources pane and then double-click the View/Edit Placed Design (FloorPlanner) process.

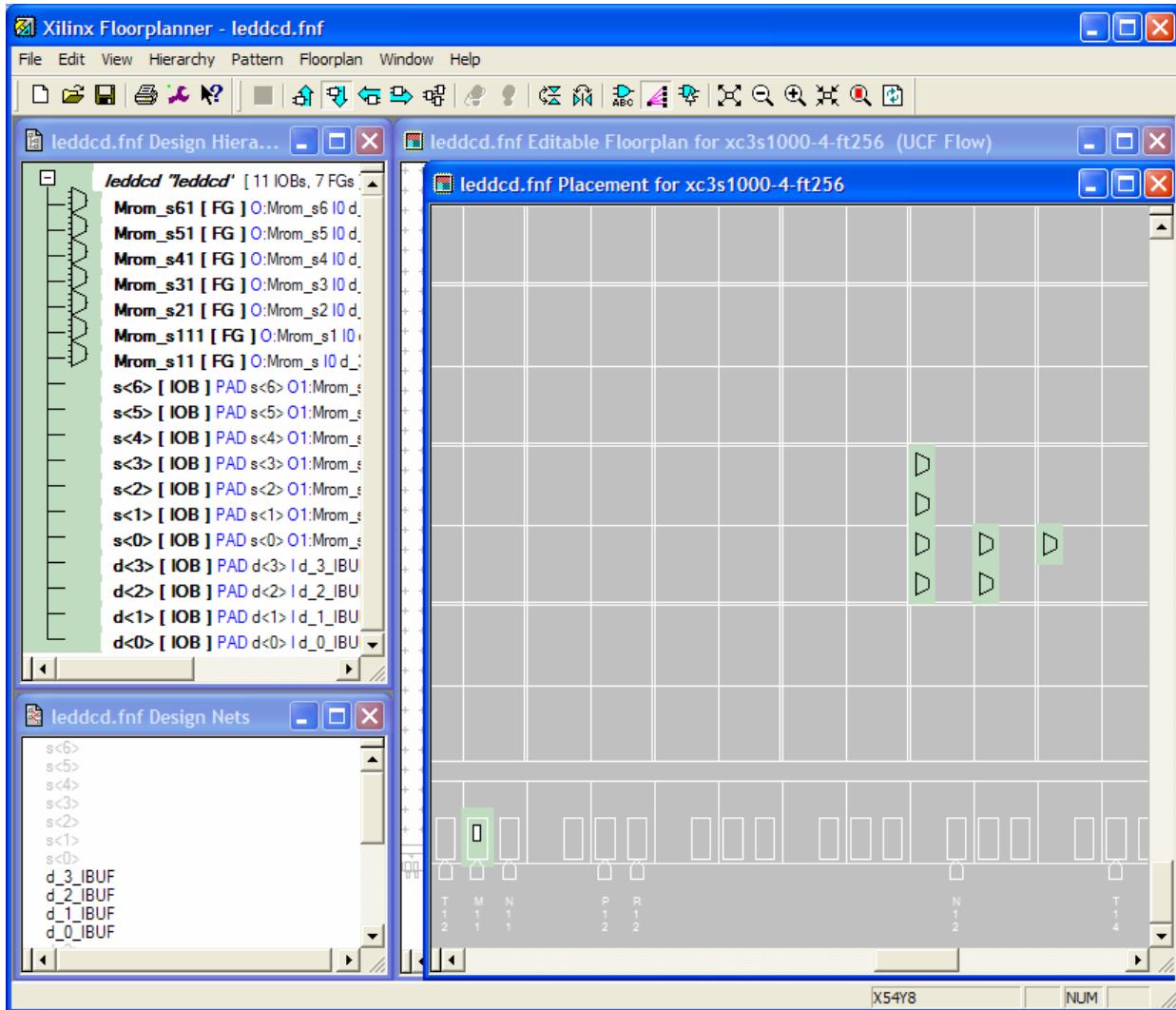


The **FloorPlanner** window will appear containing four panes:

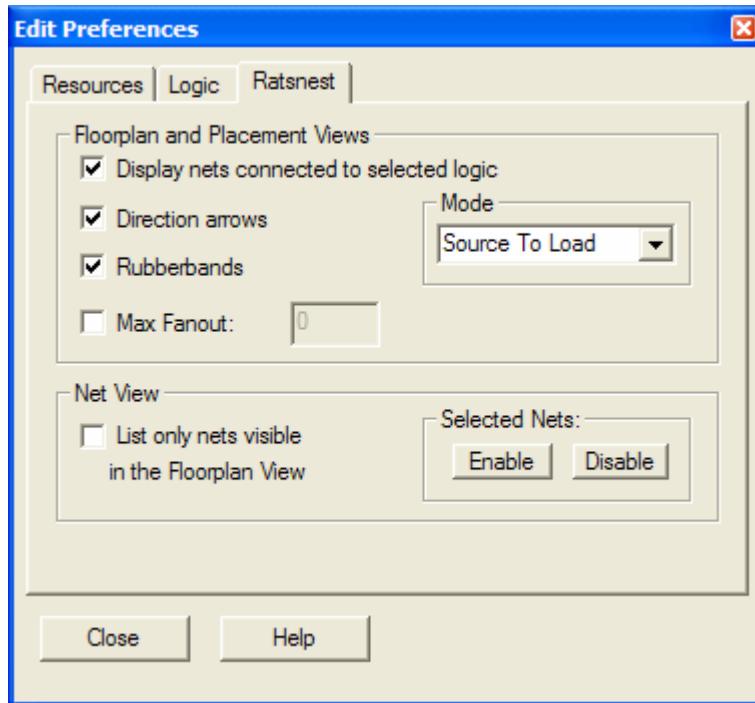
1. The Design Hierarchy pane lists the LED decoder inputs, outputs and LUTs assigned to the various CLBs in the FPGA.
2. The Design Nets pane lists the various signal nets in the LED decoder.
3. The Placement pane shows the array of CLBs in the FPGA. The I/O pins are also shown around the periphery. (The pins used for Vcc, GND, and programming are not shown.)
4. The Editable Floorplan pane allows you to change the pin assignments of the I/O signals by dragging them to different pin locations.



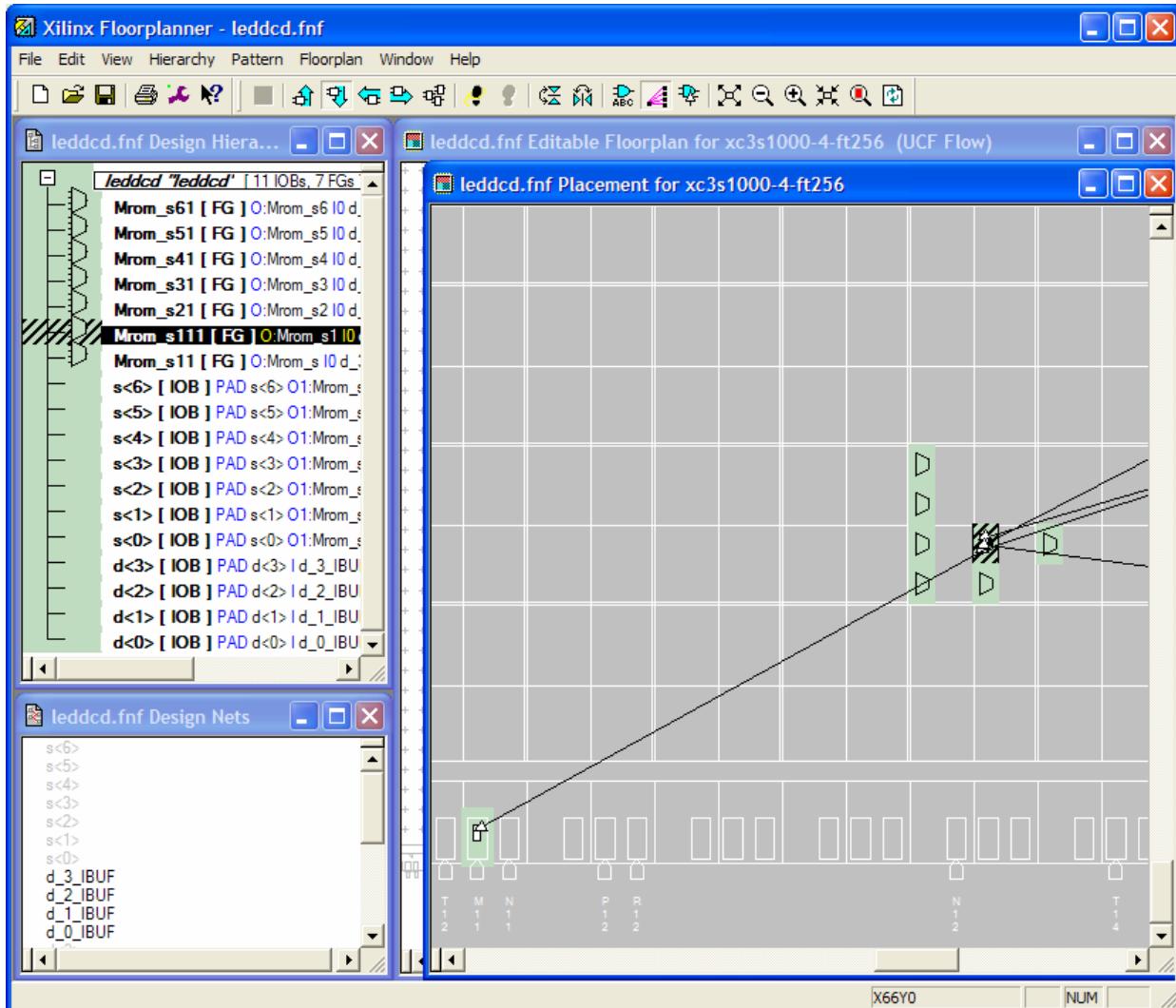
The CLBs used by the LED decoder circuit are highlighted in light-green and are clustered near the lower right-hand edge of the CLB array. To enlarge this region of the array, click on the button and then draw a rectangle around the highlighted CLBs in the Placement pane. The enlarged view of the CLBs used by the LED decoder appear as shown below.



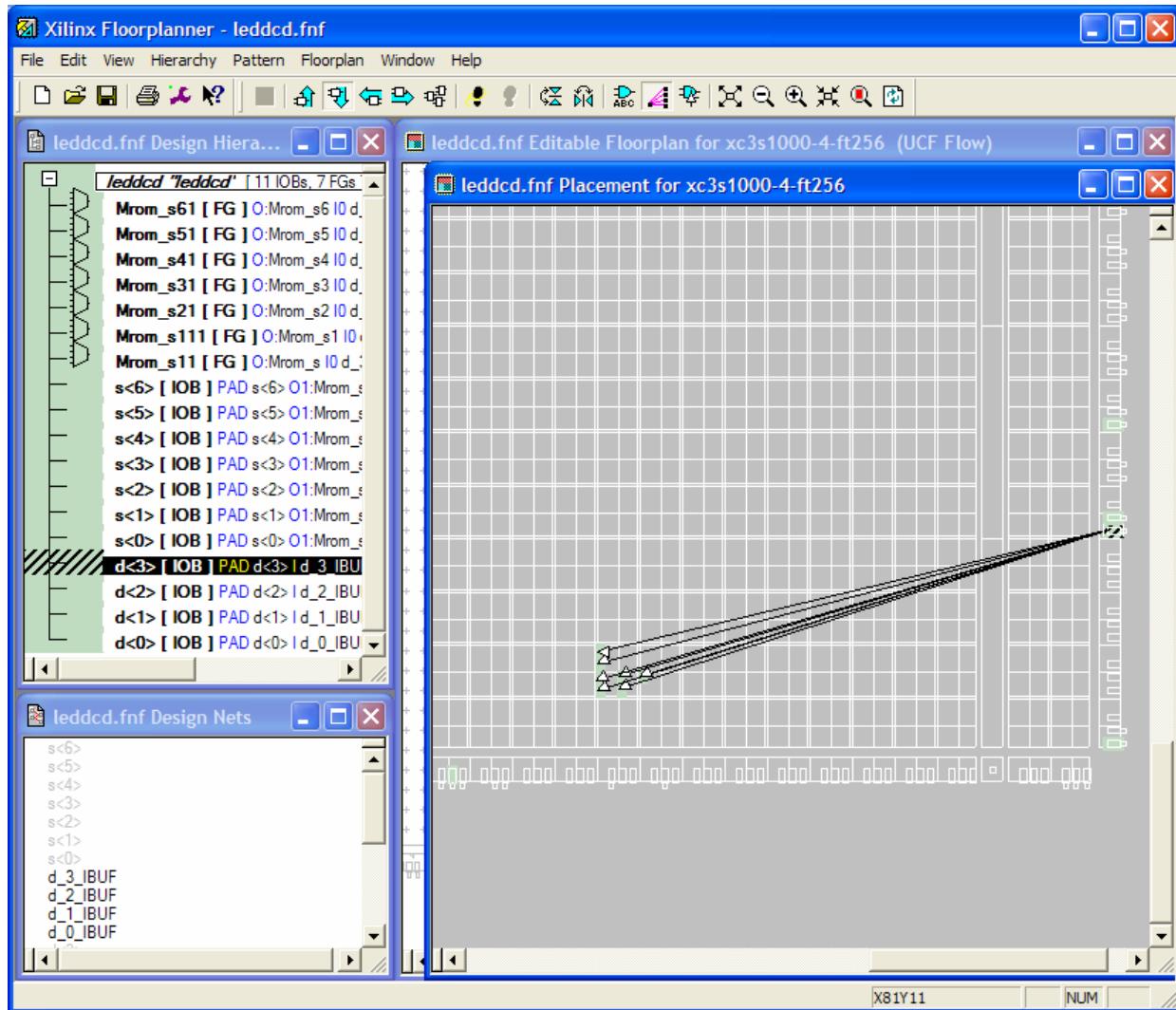
You can enable the display of the connections between I/O pins and CLBs by selecting the **Edit**→**Preferences** menu item and then checking the boxes in the Ratsnest tab of the **Edit Preferences** window as shown below.



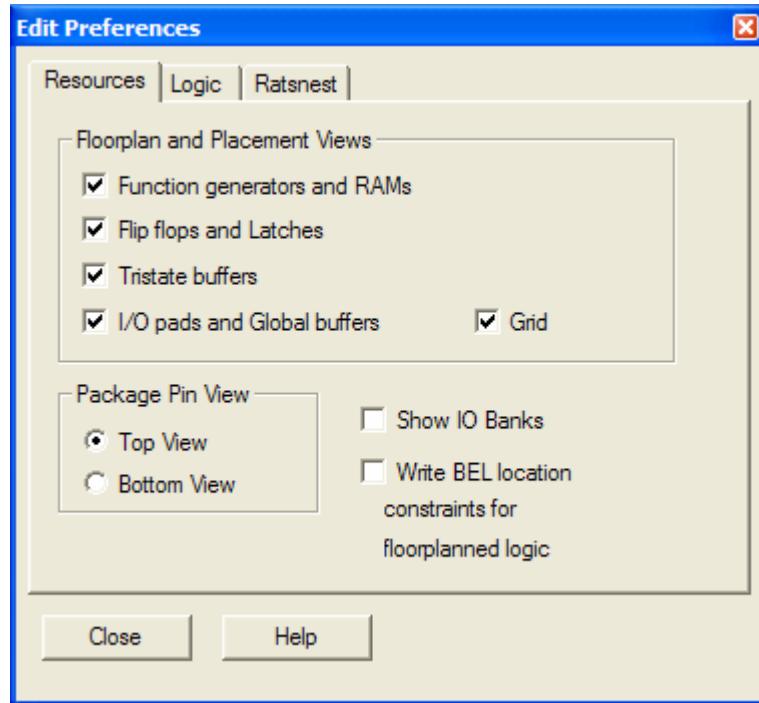
Now clicking on a CLB will highlight the nets connecting the inputs and output to the CLB.



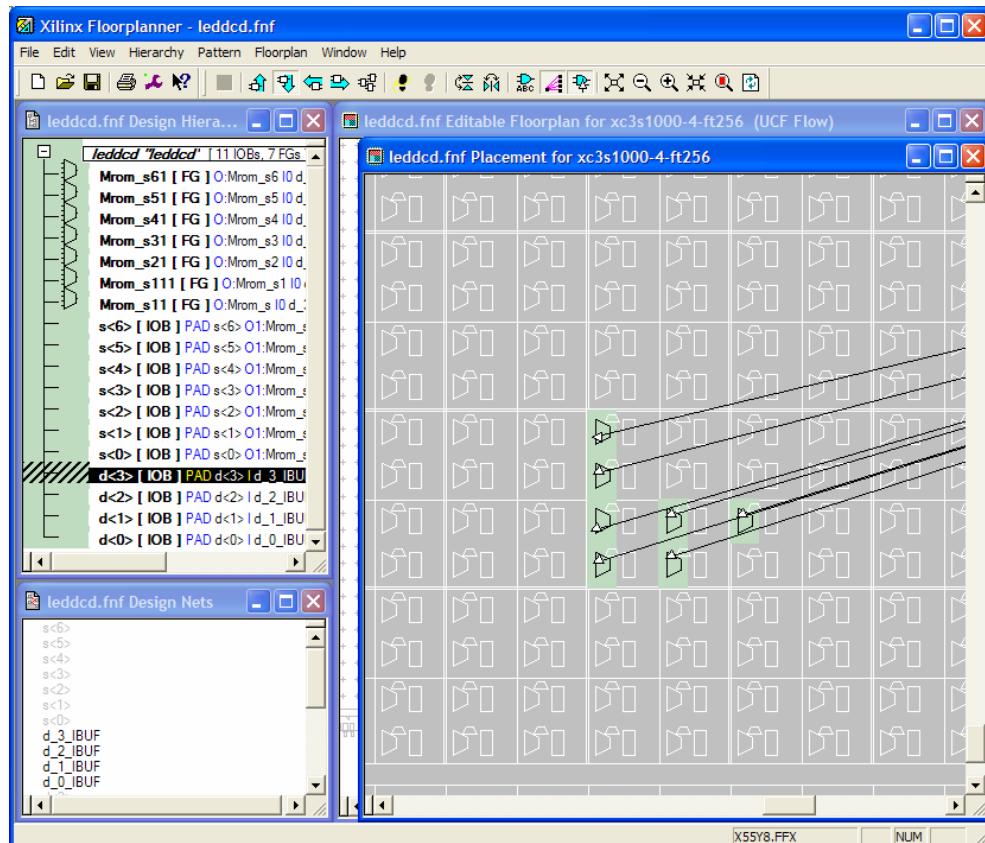
In an analogous manner, you can click on an input pin to highlight which CLBs are dependent on that input. (For this design, each input affects every CLB.)



To view all the available FPGA resources (not just those that are used by the design), select Edit→Preferences and check all the boxes in the Resources tab as shown below.



Now the Placement pane shows all the LUTs, RAMs, buffers, etc. that are available in the FPGA.

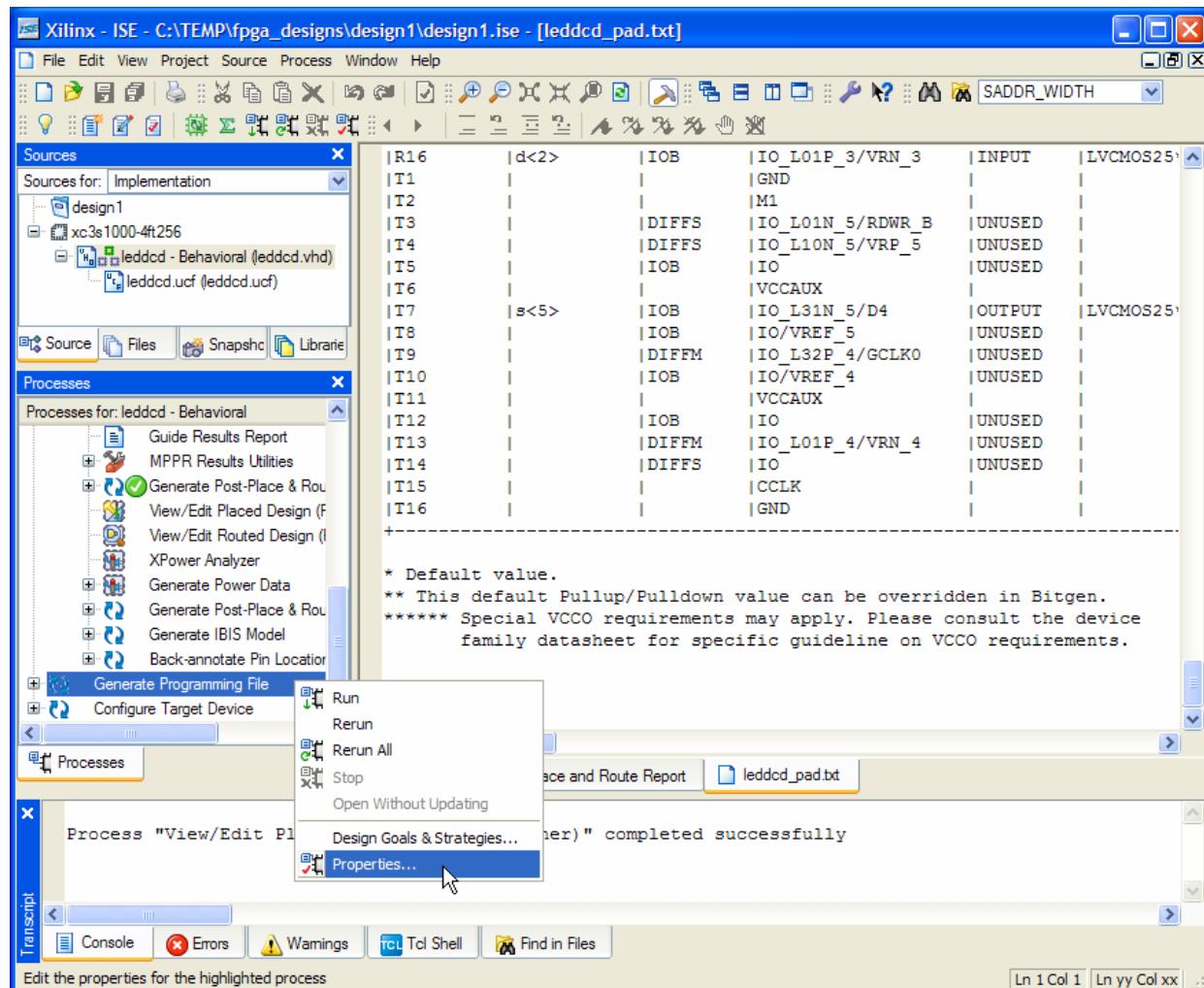


Viewing the placement of circuit elements after the place & route process can give you insights into the resource usage of certain VHDL language constructs. In addition to viewing the placement of the design, the Floorplanner can be used to re-arrange and optimize the placement. This is akin to the software technique of hand-optimizing assembly code output by a compiler. You won't do this here, but it is an option for designs which push at the limits of the capabilities of FPGAs.

Generating the Bitstream

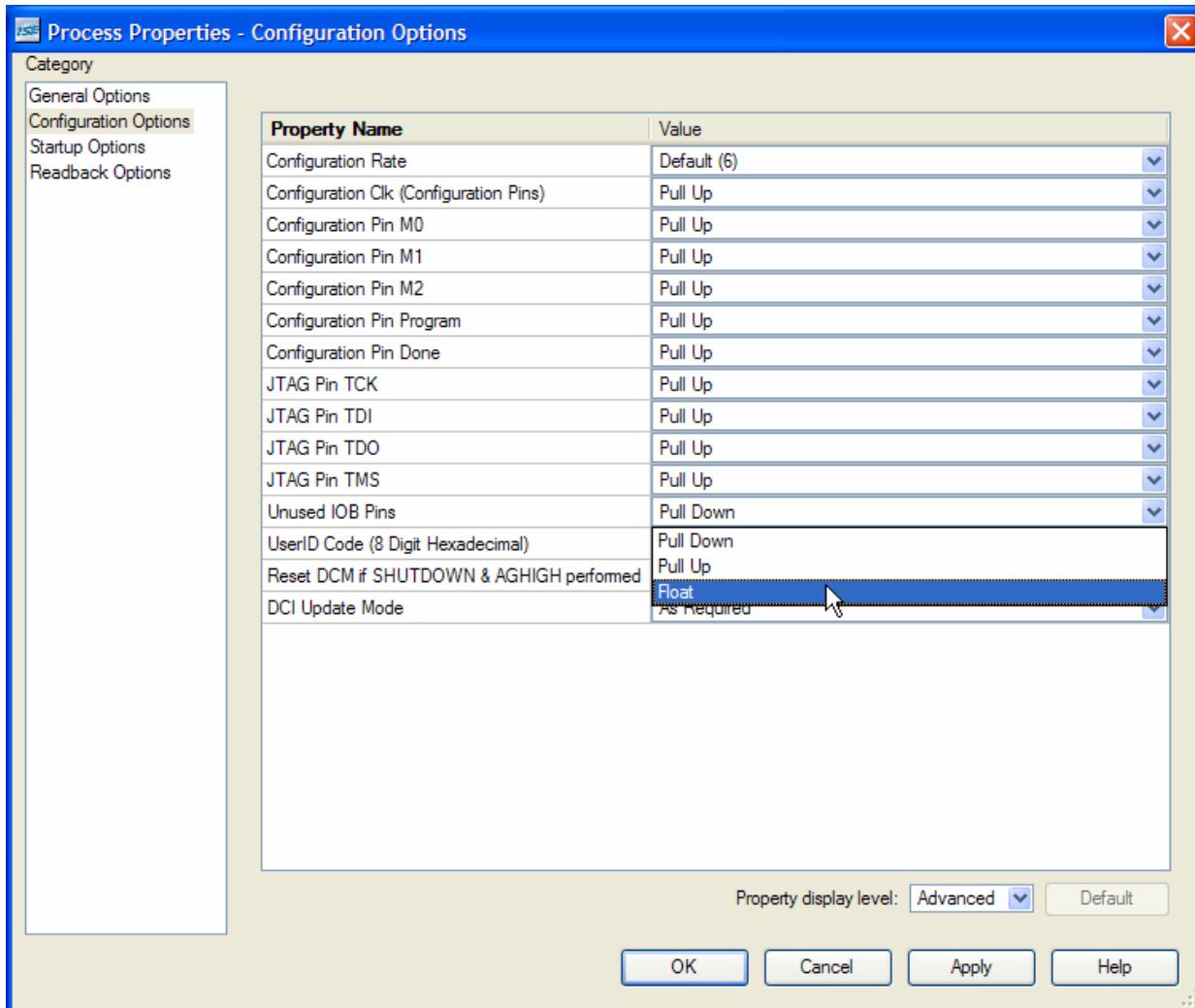
Now that you have synthesized our design and mapped it to the FPGA with the correct pin assignments, you are ready to generate the bitstream that is used to program the actual chip.

When using the XSA-3S1000 Board, you need to set some options before generating the bitstream. (This is not needed if you are using the XSA-50, XSA-100 or XSA-200 Board, but it wouldn't hurt, either.) Right-click on the Generate Programming File process and select Properties... from the pop-up menu.

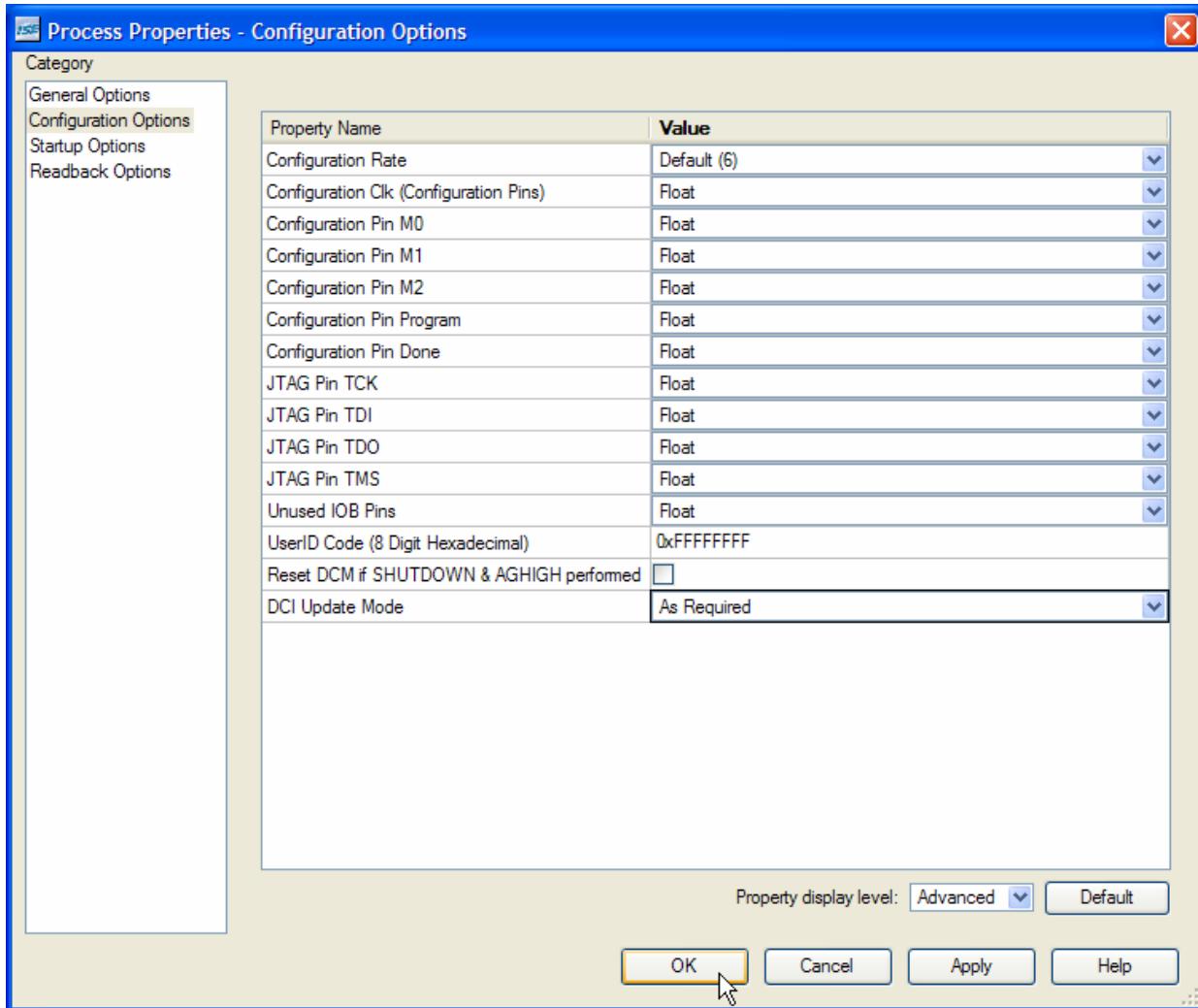


Now select the Configuration Options tab in the **Process Properties** window and set all the Pull Up and Pull Down values to Float to disable the internal resistors of the FPGA. (At the minimum, the Unused IOB Pins value must be set to Float, but it is best to set them all to Float.) If they are not

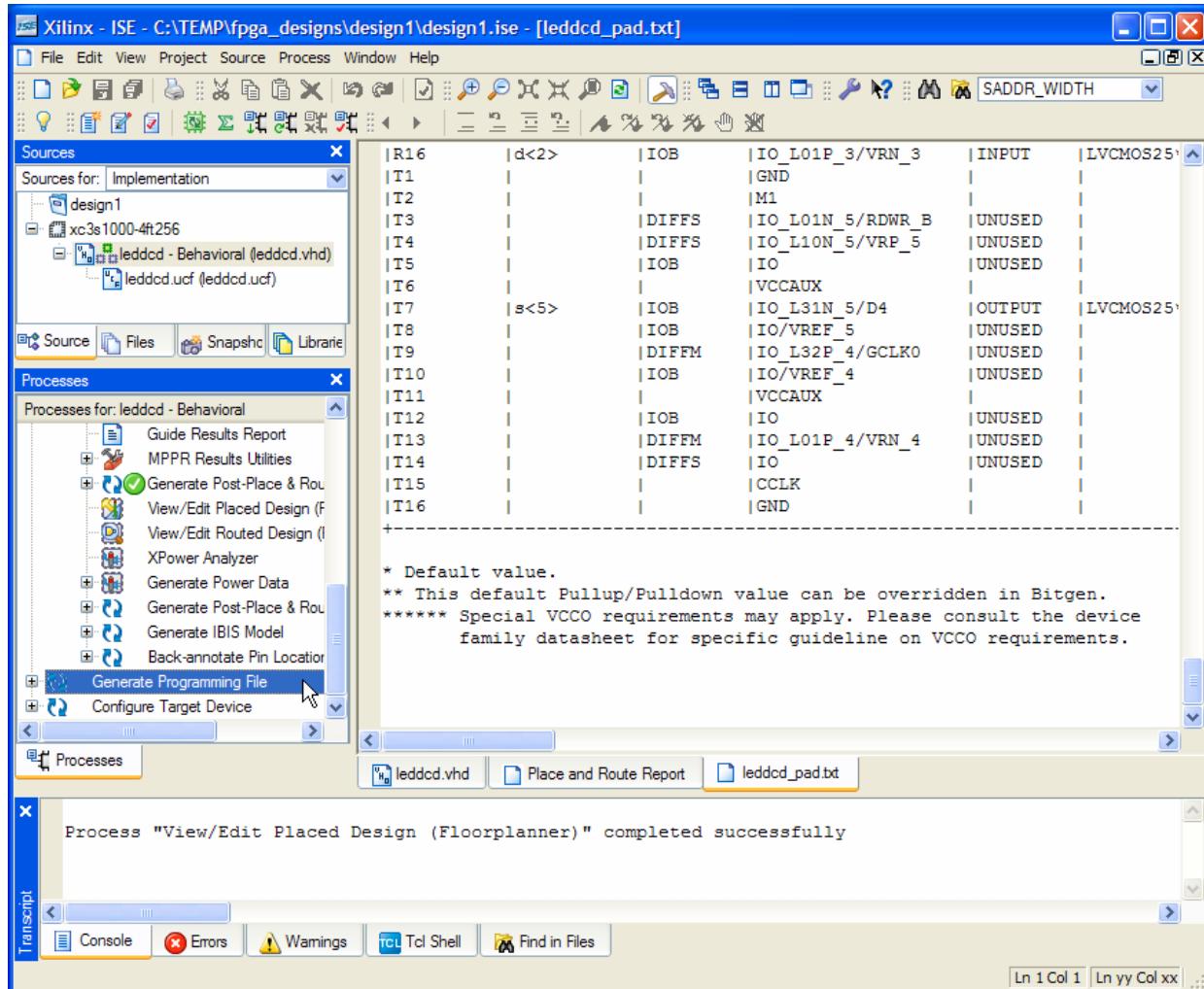
disabled, the strong internal pull-up and pull-down resistors in the Spartan3 FPGA will overpower the external resistors on the XSA Board.



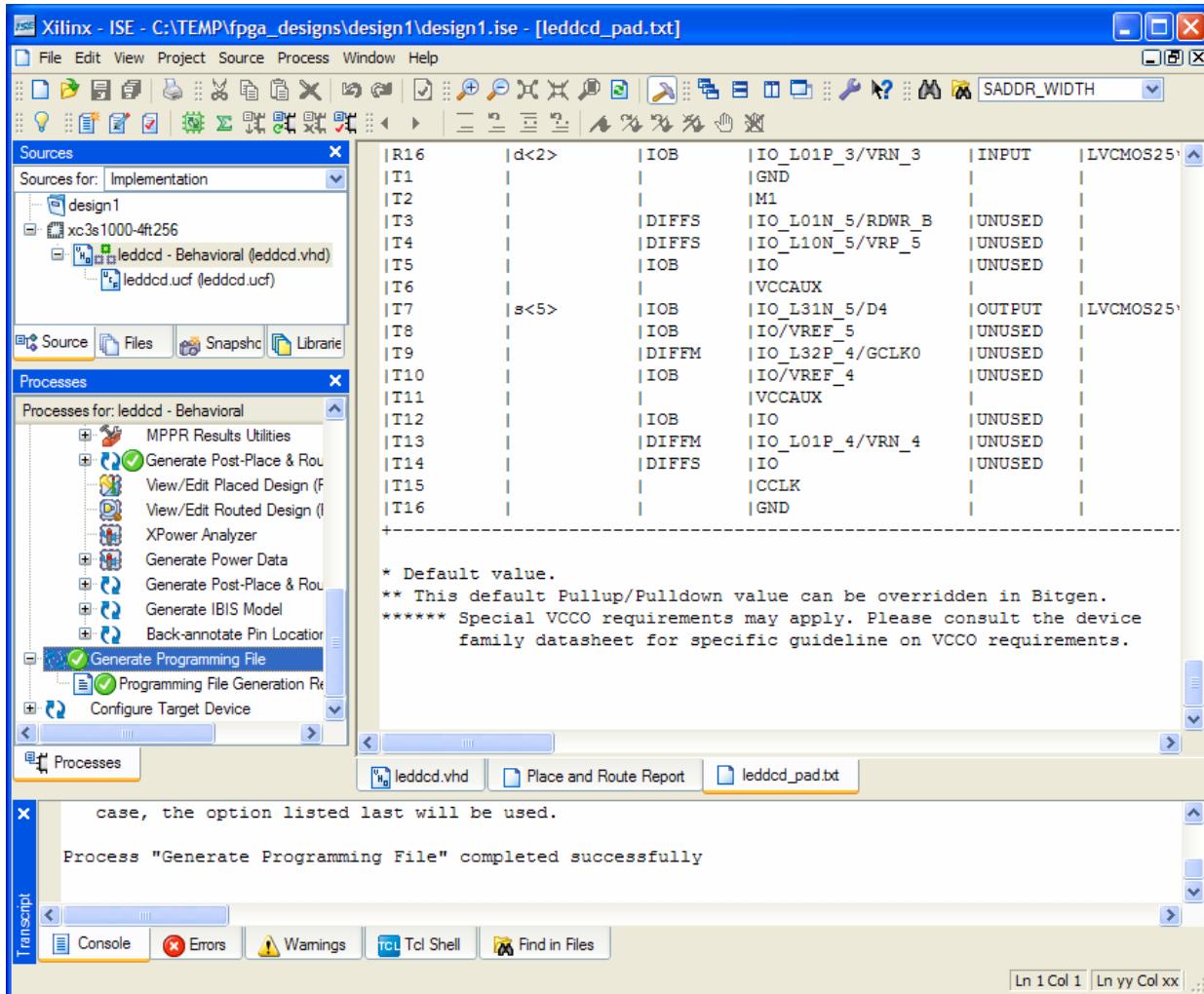
After setting all the values as shown below, click on OK.



Once you have set the bitstream generation options, highlight the ledgcd object in the Sources pane and double-click on the Generate Programming File process.

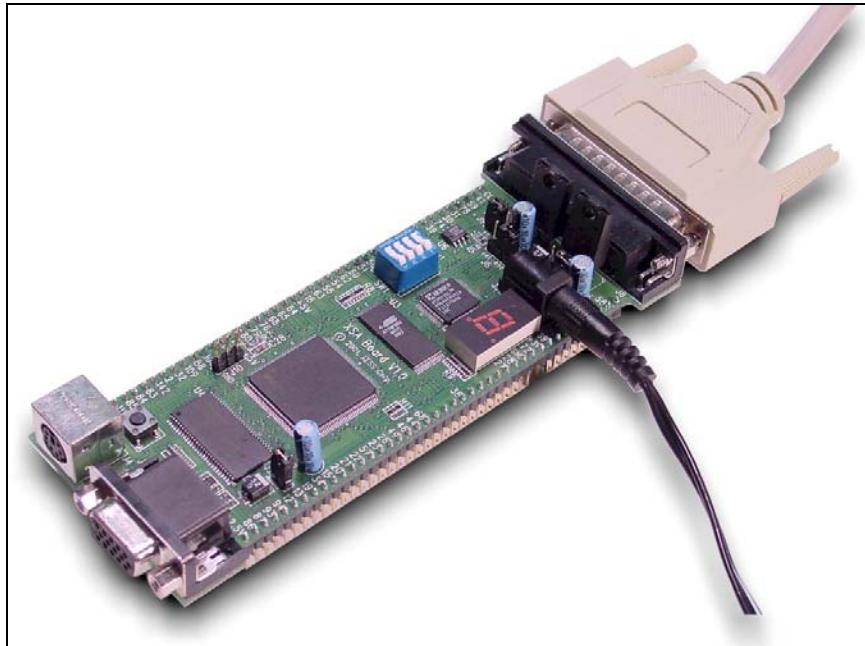


Within a few seconds, a will appear next to the Generate Programming File process and a file detailing the bitstream generation process will be created. A bitstream file named ledcdc.bit can now be found in the design1 folder.

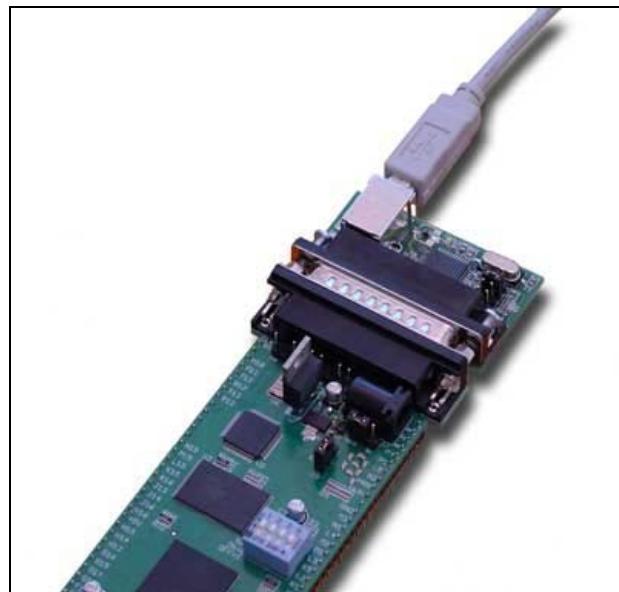


Downloading the Bitstream

Now you have to download the bitstream file into the FPGA on the XSA Board. The XSA Board is powered with a DC power supply and is attached to the PC parallel port with a standard 25-wire cable as shown below.

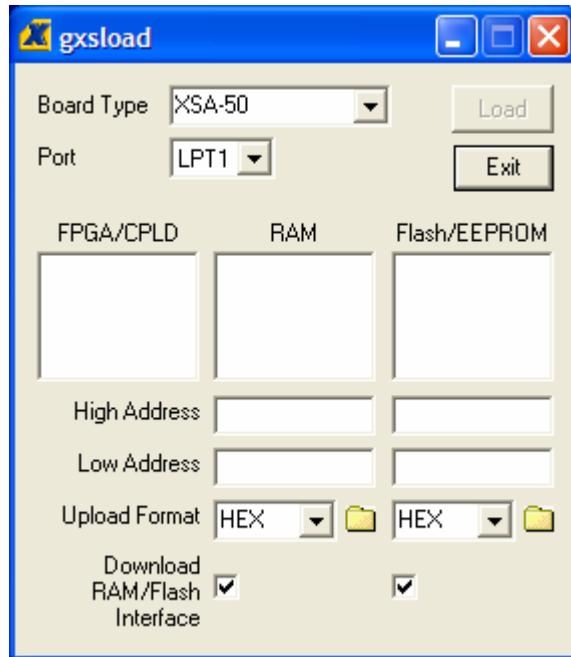


(You can also communicate with the XSA Board through a USB port by attaching an XSUSB interface as shown below. Downloading the bitstream is done in the same way regardless of the choice of interface.)

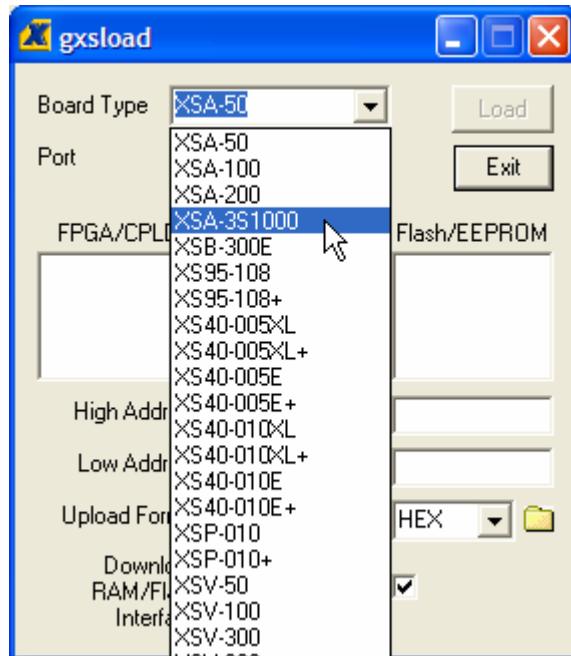




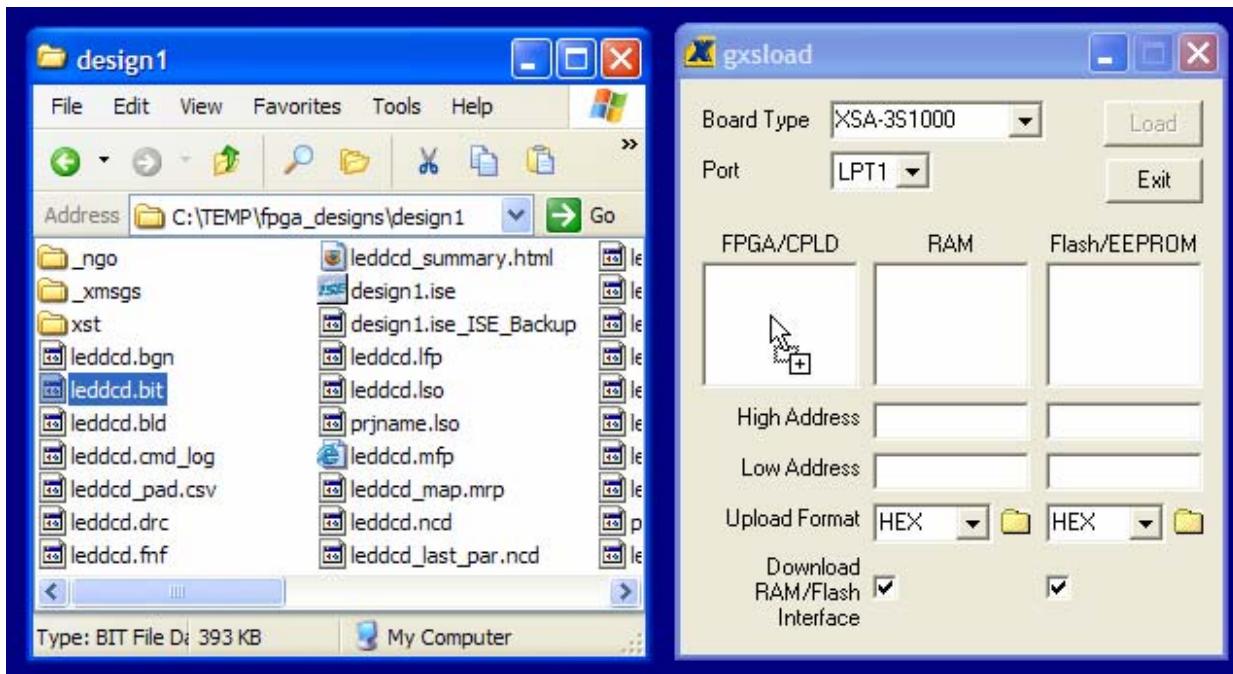
The XSA Boards are programmed using the gxsload utility. Double click the **GXSLOAD** icon to bring up the **gxsload** window:



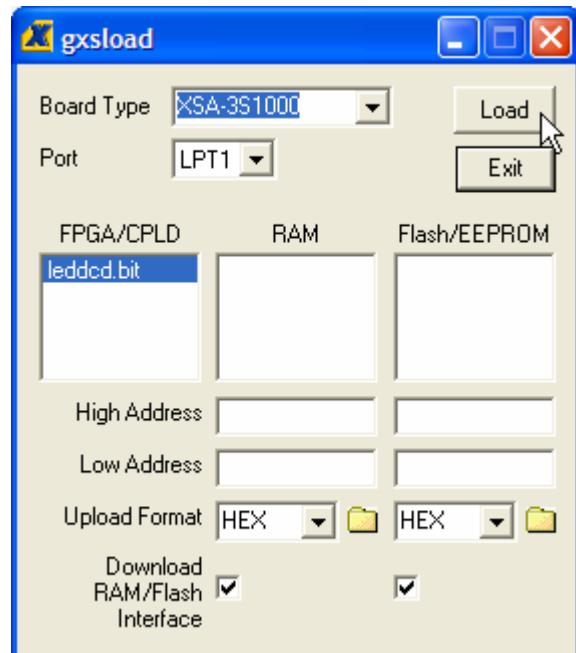
Then click in the Board Type field and select XSA-3S1000 from the drop-down menu since this is the board you are going to load with the bitstream.



Then open a window that shows the contents of the folder where you stored the LED decoder design (C:\TEMP\fpga_designs\design1 in my case). You just drag-and-drop the leddcd.bit file from the **design1** window into the FPGA/CPLD pane of the **gxload** window.



Then you click on the Load button to initiate the programming of the FPGA. Downloading the leddcd.bit file to the XSA Board takes only a few seconds.

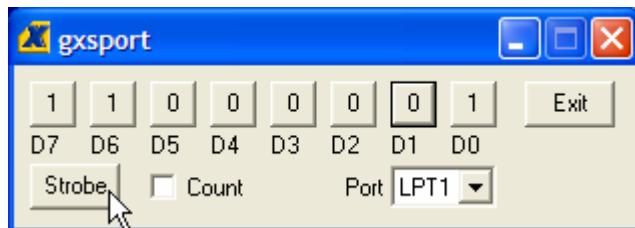


Testing the Circuit

Once the FPGA on the XSA Board is programmed, you can begin testing the LED decoder. The eight data pins of the PC parallel port connect to the FPGA through the downloading cable. You have assigned the inputs of the LED decoder to pins, which are connected to the parallel port data pins. The gxsport utility lets you control the logic values on these pins. By placing different bit patterns on the pins, you can observe the outputs of the LED decoder through the seven-segment LED on the XSA Board.



Double-clicking the **gxsport** icon initiates the gxsport utility. The **d0**, **d1**, **d2**, and **d3** inputs of the LED decoder are assigned to the pins controlled by the D0, D1, D2, and D3 buttons of the **gxsport** window. To apply a given input bit pattern to the LED decoder, click on the D buttons to toggle their values. Then click on the Strobe button to send the new bit pattern to the pins of the parallel port and on to the FPGA. For example, setting (D3,D2,D1,D0) = (1,1,1,0) will cause E to appear on the seven-segment LED of the XSA Board.



If you check the Count box in the **gxsport** window, then each click on the Strobe button increments the eight-bit value represented by D7-D0. This makes it easy to check all sixteen input combinations.

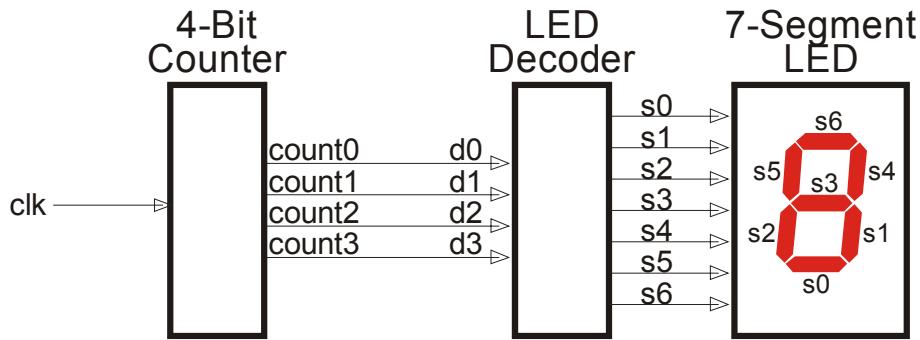
NOTE: Bit D7 of the parallel port controls the /PROGRAM pin of the FPGA. Do not set D7 to 0 or you will erase the configuration of the FPGA. Then you will have to download the bitstream again to continue testing your design.

4

Hierarchical Design

A Displayable Counter

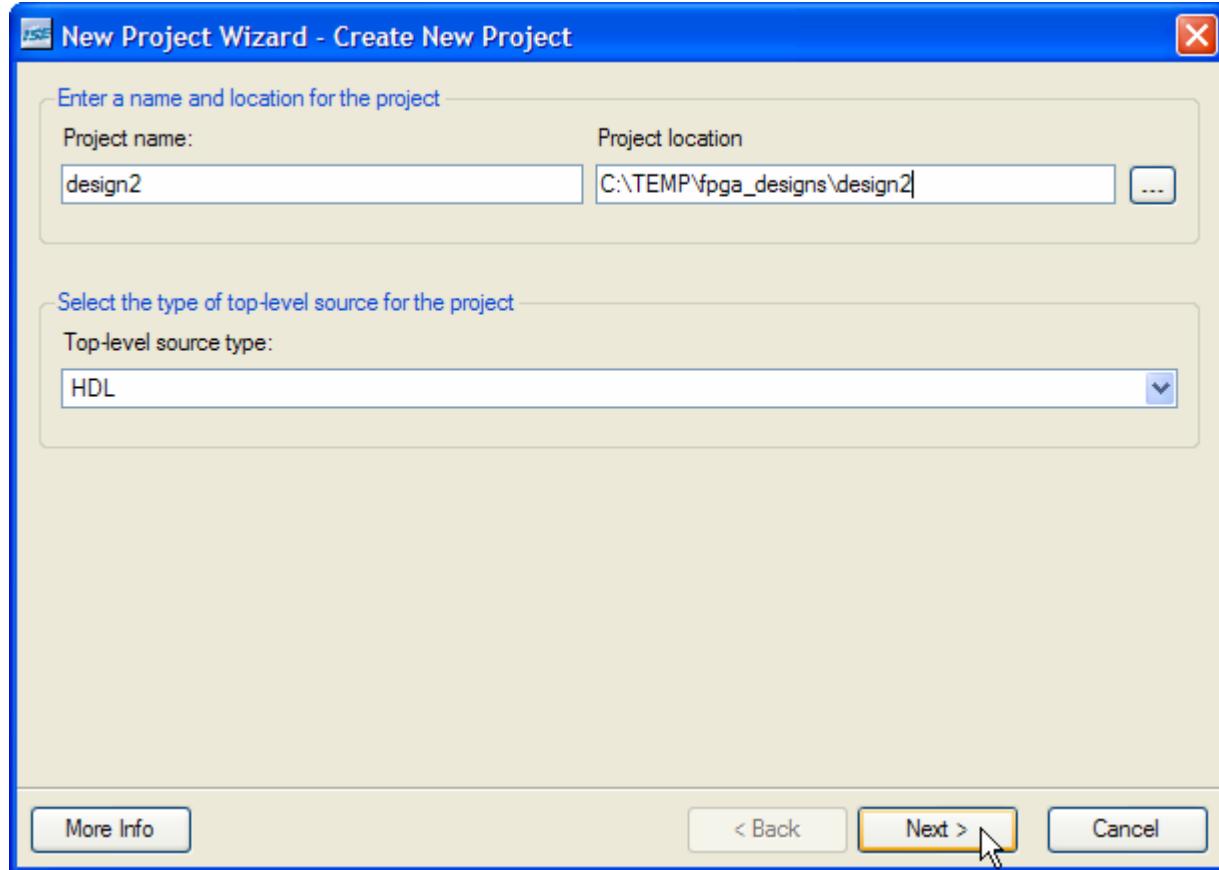
You went through a lot of work for your first FPGA design, so you will reuse it in this design: a four-bit counter whose value is displayed on a seven-segment display. The counter will increment on the falling edge of the clock. The four-bit output from the counter enters the LED decoder whereupon the counter value is displayed on the seven-segment LED. A high-level diagram of the displayable counter looks like this:



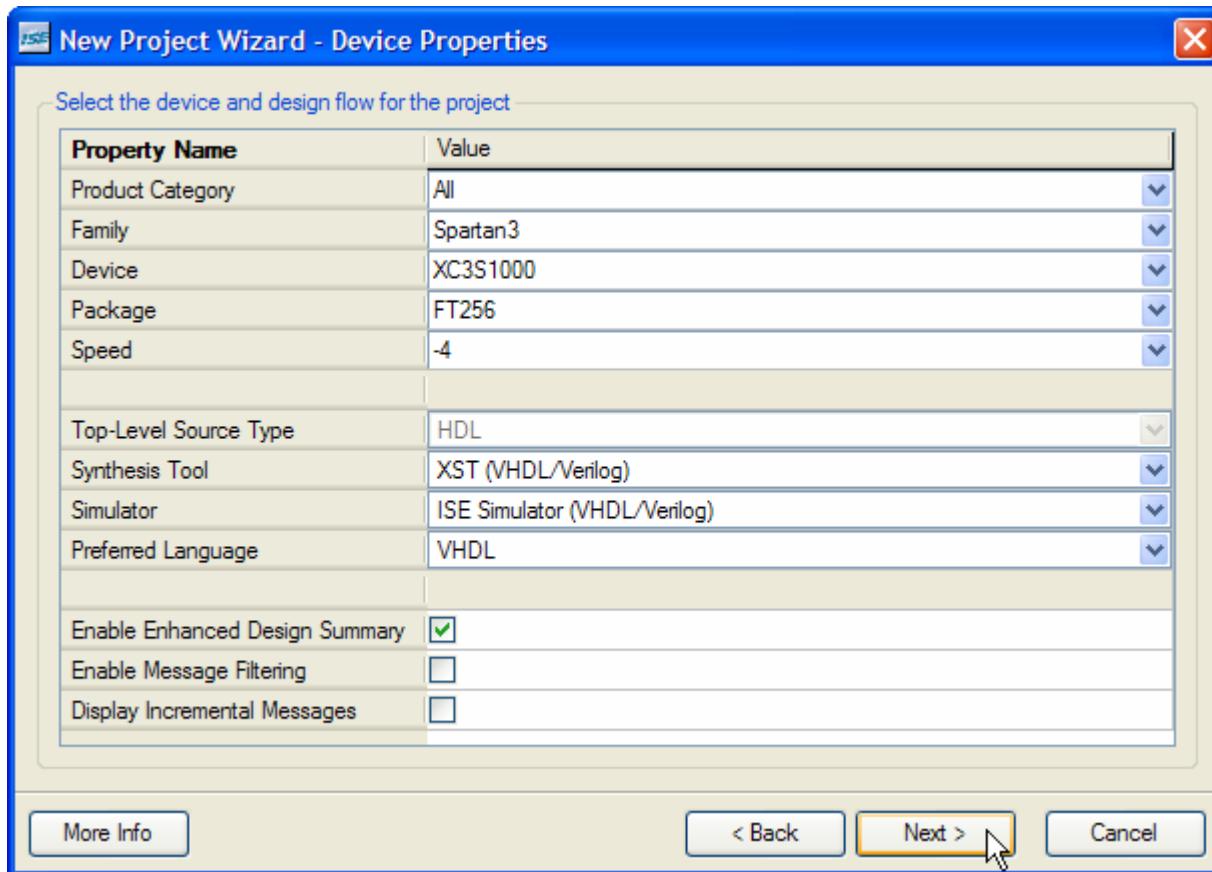
This design is hierarchical in nature. The LED decoder and counter are modules, which are interconnected within a top-level module.

Starting a New Design

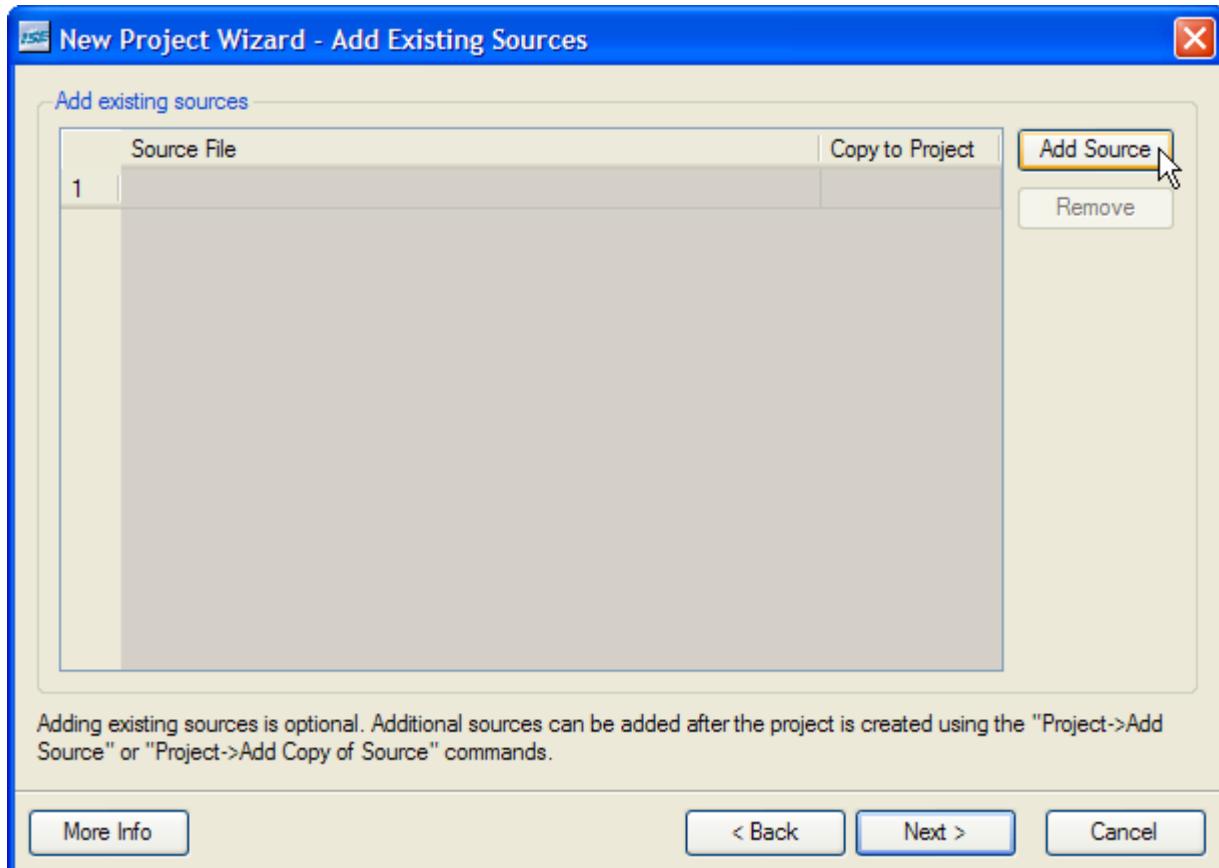
You can start a new project using the File→New Project... menu item. Name the project **design2** and store it in the same folder as the previous design: C:\TEMP\fpga_designs. Then click on the Next button.



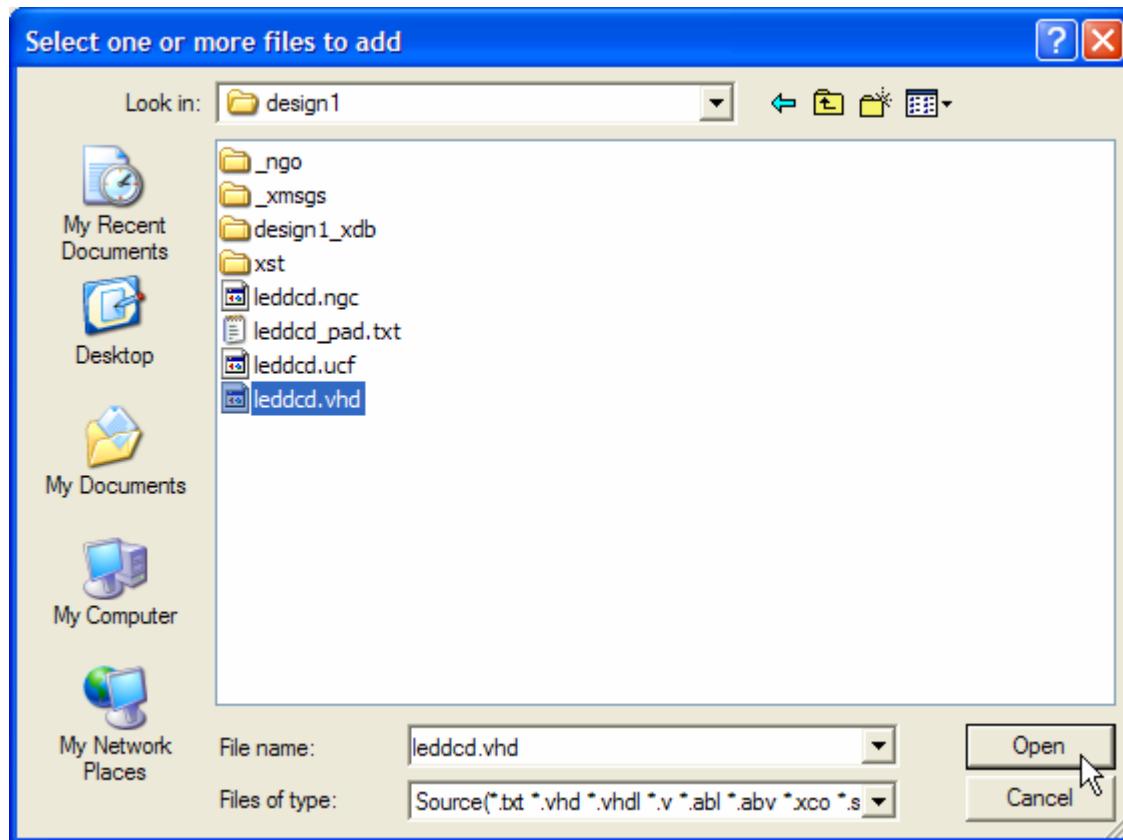
You will target the same FPGA on the XSA Board, so the other properties in the **New Project** window retain the same values you set in the previous project.



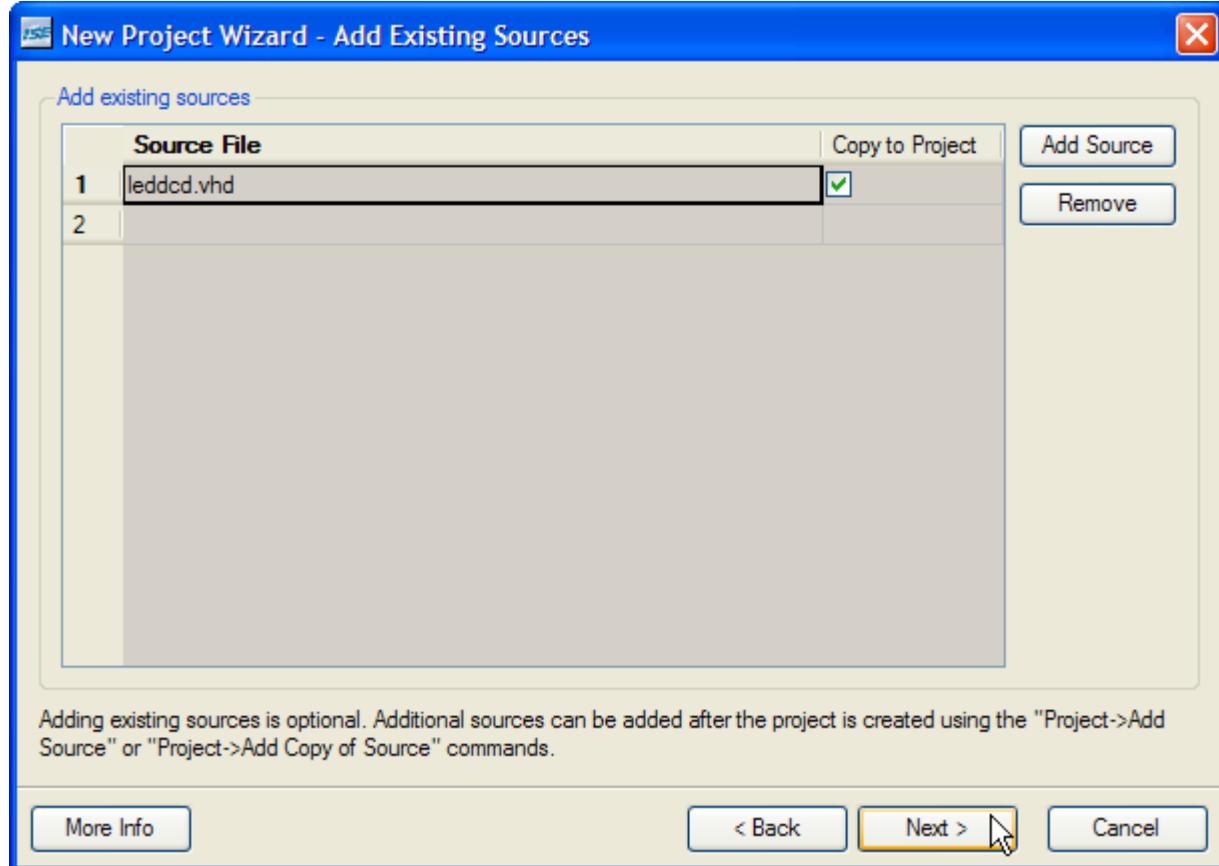
You won't add any new source files at the moment, so just click on the Next button on the **Create New Source** window. This brings up the **Add Source** window shown below that lets you add existing source files to your new project. It will save time if you re-use the LED decoder from the previous design, so click on the Add Source... button.



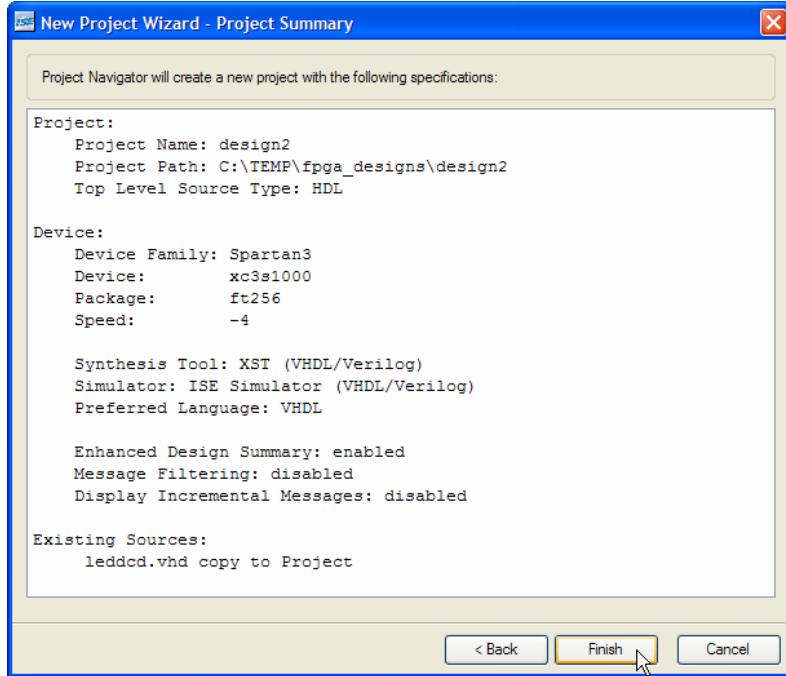
The **Add Existing Sources** window appears. Move to the C:\TEMP\fpga_designs\design1 folder and highlight the ledgcd.vhd file that contains the VHDL source code for the LED decoder.



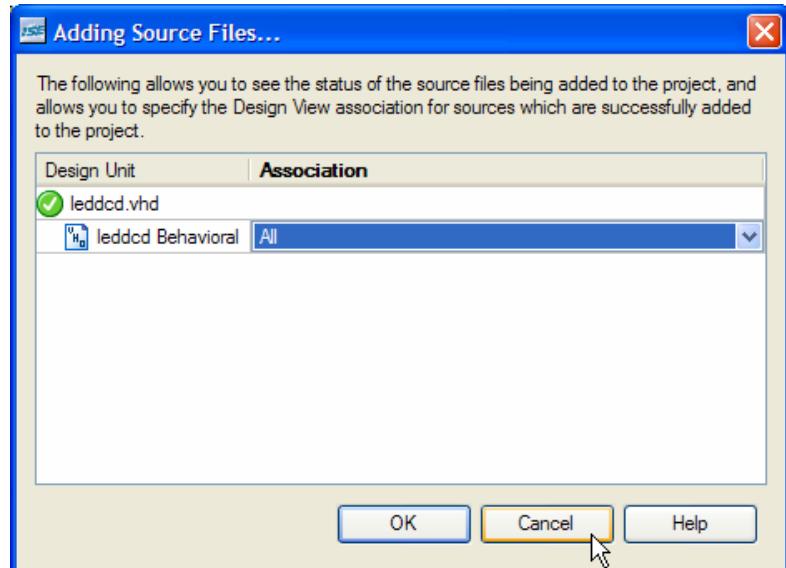
The **New Project** window now shows that a copy of the leddcd.vhd file will be added to the project. The Copy to Project box is checked, so the leddcd.vhd file will be copied from the design1 folder to the design2 folder. Unchecking this box will cause the **design2** project to link directly to the leddcd.vhd file in the design1 folder, so any change in this file will affect both projects. For this example, I have chosen to make a separate copy so the box remains checked. This is the only existing file you need to add, so click on the Next button to move on to the next window.



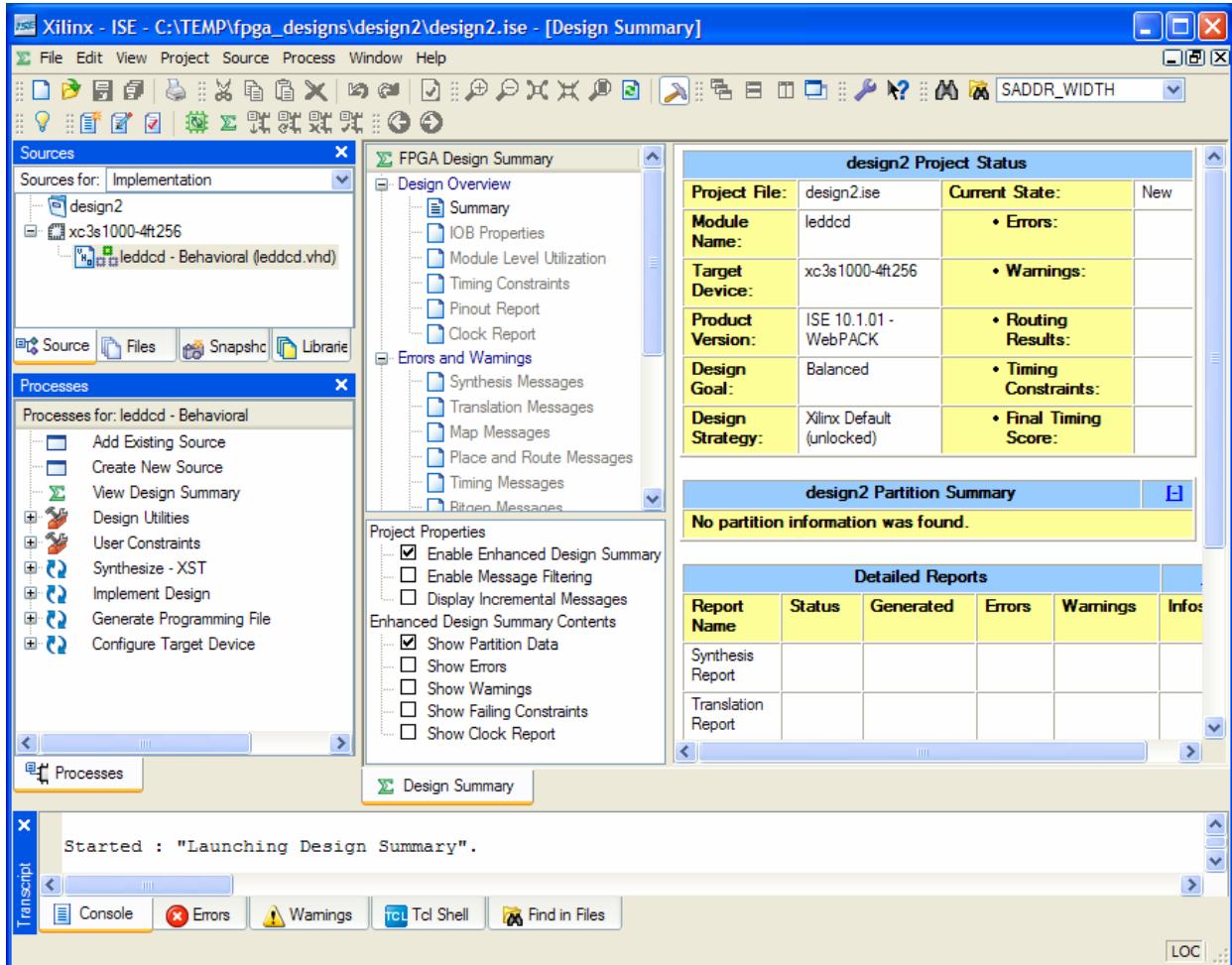
The final screen shows the pertinent information for the new project. Click on the Finish button to complete the creation of the project.



An **Adding Source Files...** window will appear that lets you specify the intended use for the files you are adding. Files can be used during synthesis and implementation of an FPGA bitstream as we did in the previous design. They can also be used only during simulation, such as testbench files that specify stimulus patterns that are driven onto the inputs of a design. Or they can be for both synthesis/implementation and simulation (as most are) because they describe the function and/or structure of the design in both these domains. There are also files, such as text files containing design documentation that are added to a project but are not used in either domain. I selected All for the leddcd.vhd file even though it will only be used for synthesis and implementation in this example.

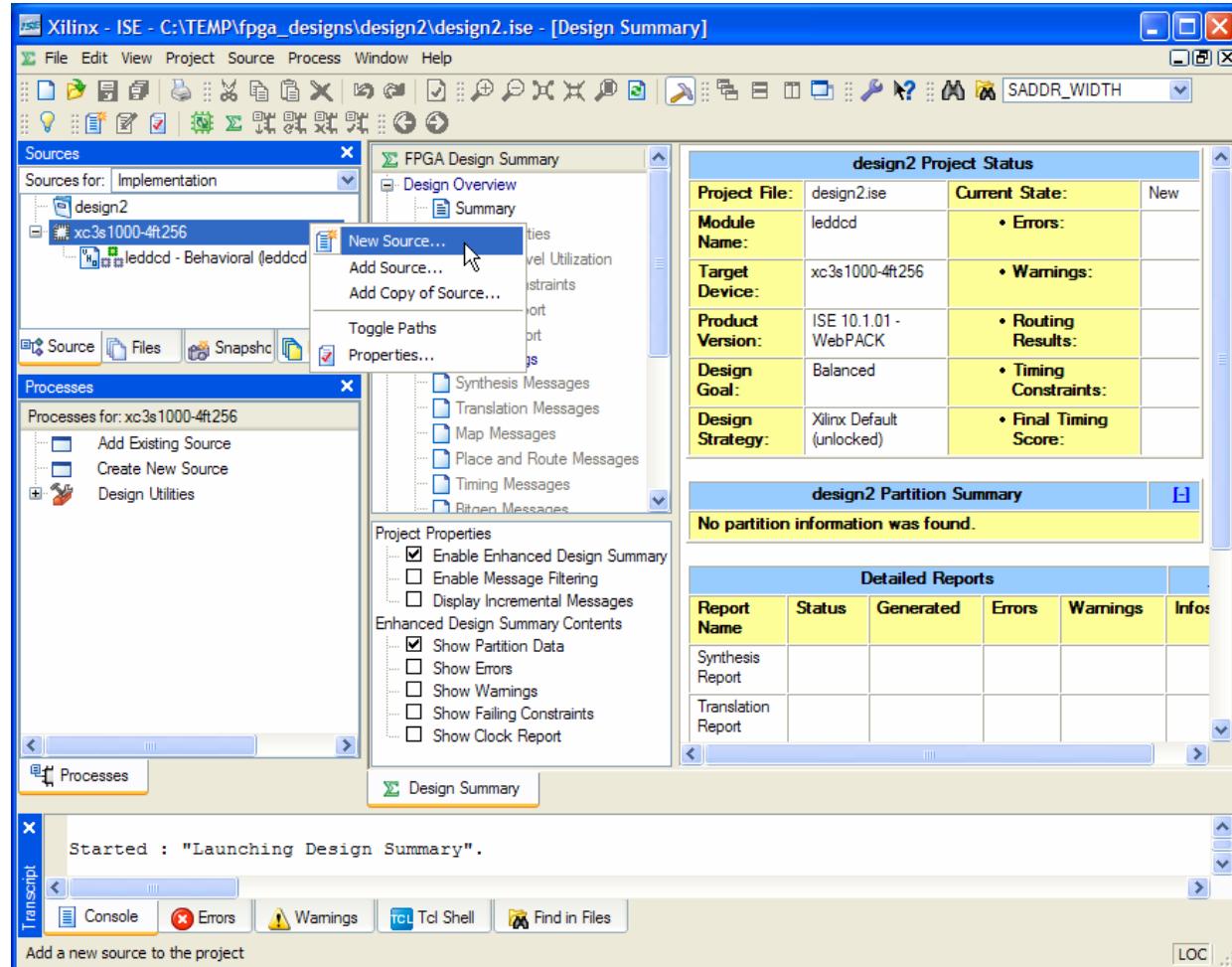


After you click on Finish in the **Adding Source Files...** window, the Sources pane contains only the single ledcdc.vhd source file.

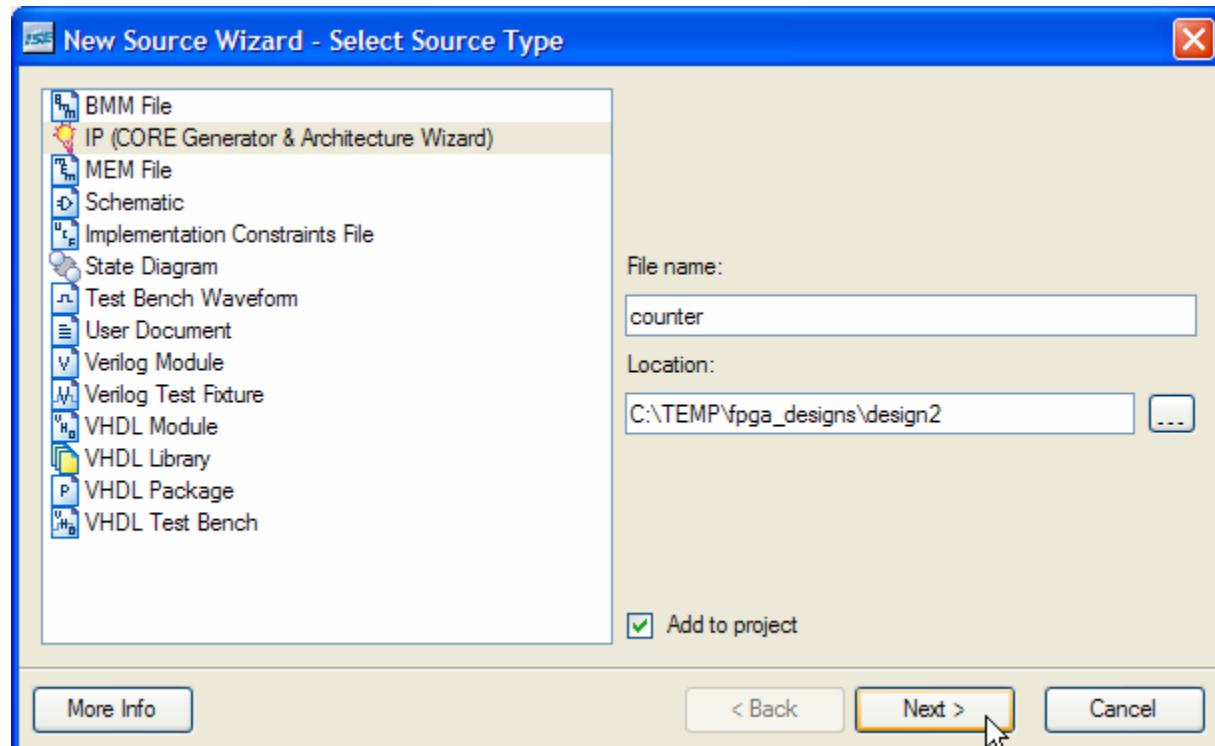


Adding a Counter

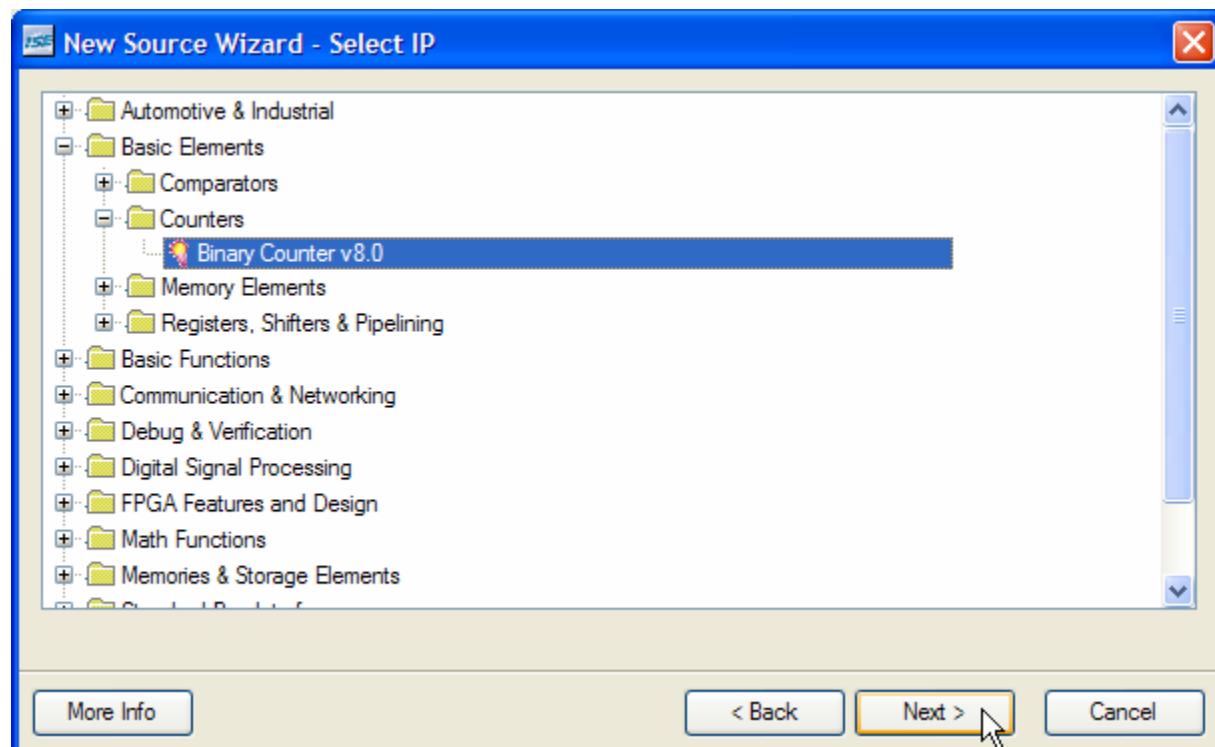
Now you have to add the counter to the design. There is no counter module yet, so you have to build one. Right-click on the xc3s1000-4ft256 object and select New Source... from the pop-up menu.



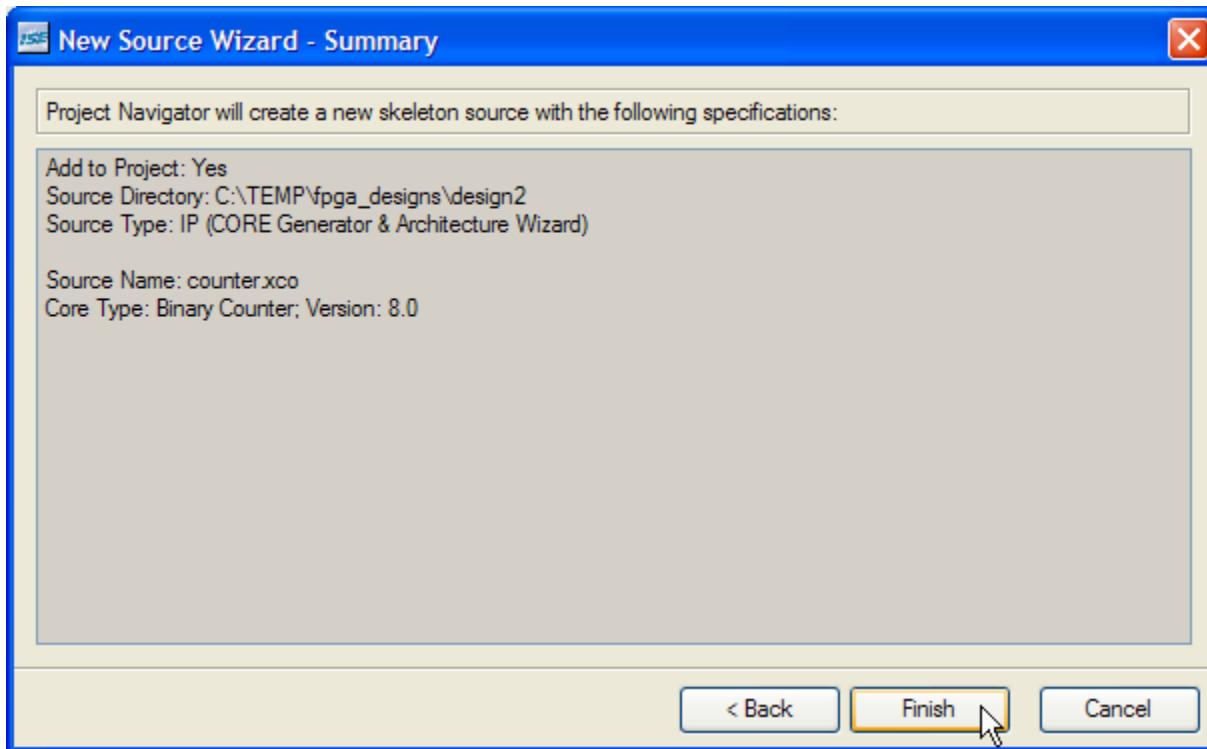
As in the previous example, you are prompted for the type of file to add to the project. This time, choose the IP (Coregen & Architecture Wizard) menu item. This will allow you to select a counter from a library of pre-built, configurable components. Then type counter into the File Name field and click on the Next button.



In the next window, expand the library tree until you find the binary counter component. Then click on Next.



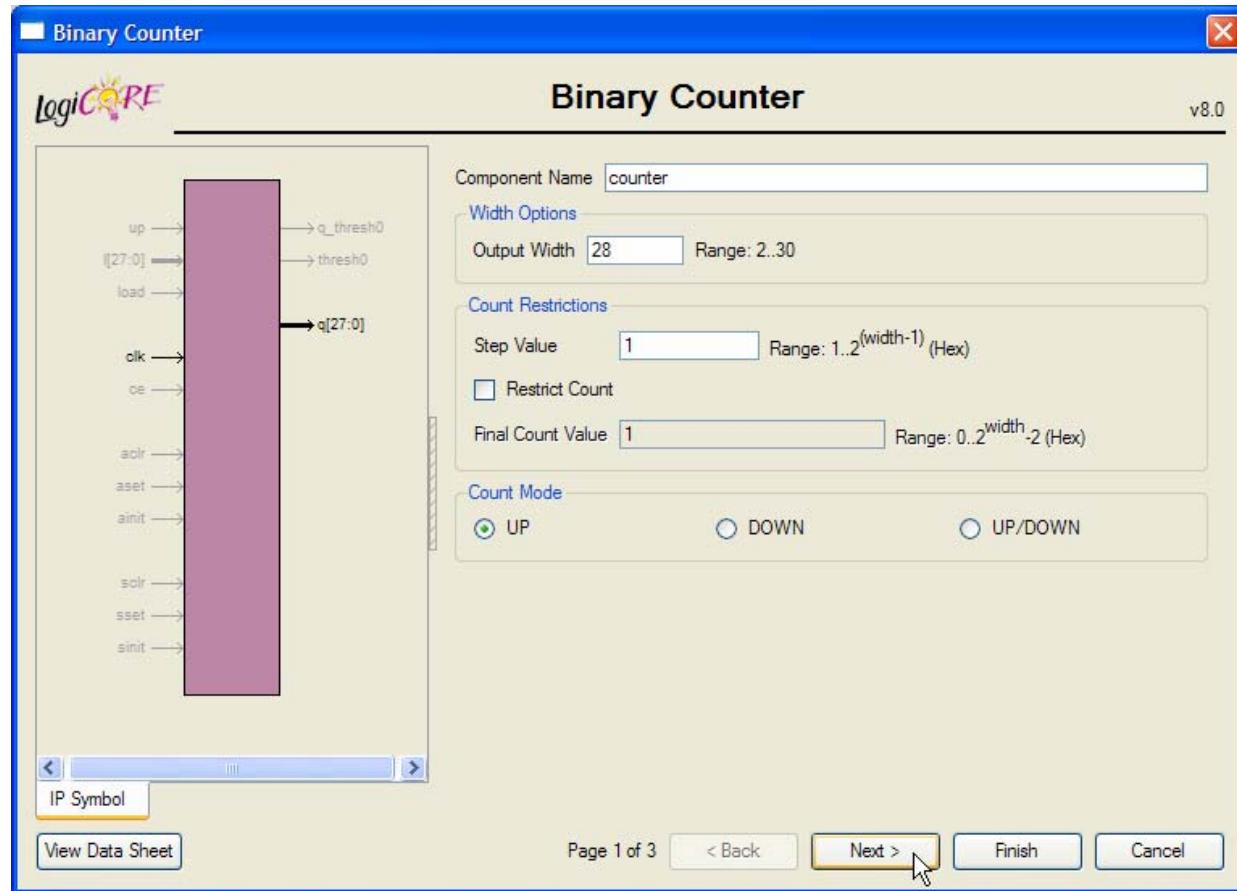
A very uninteresting summary of your choice appears. Click Finish to move on to the screens that will let you configure the counter component for this design.



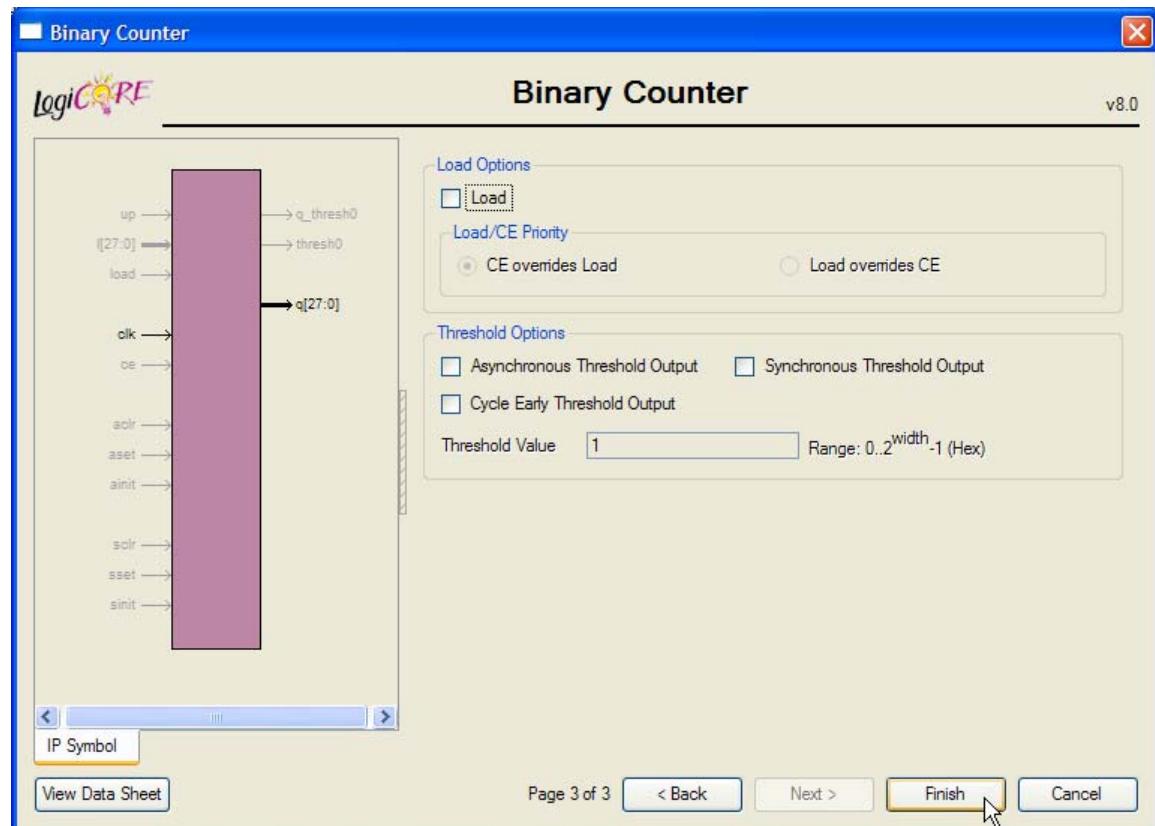
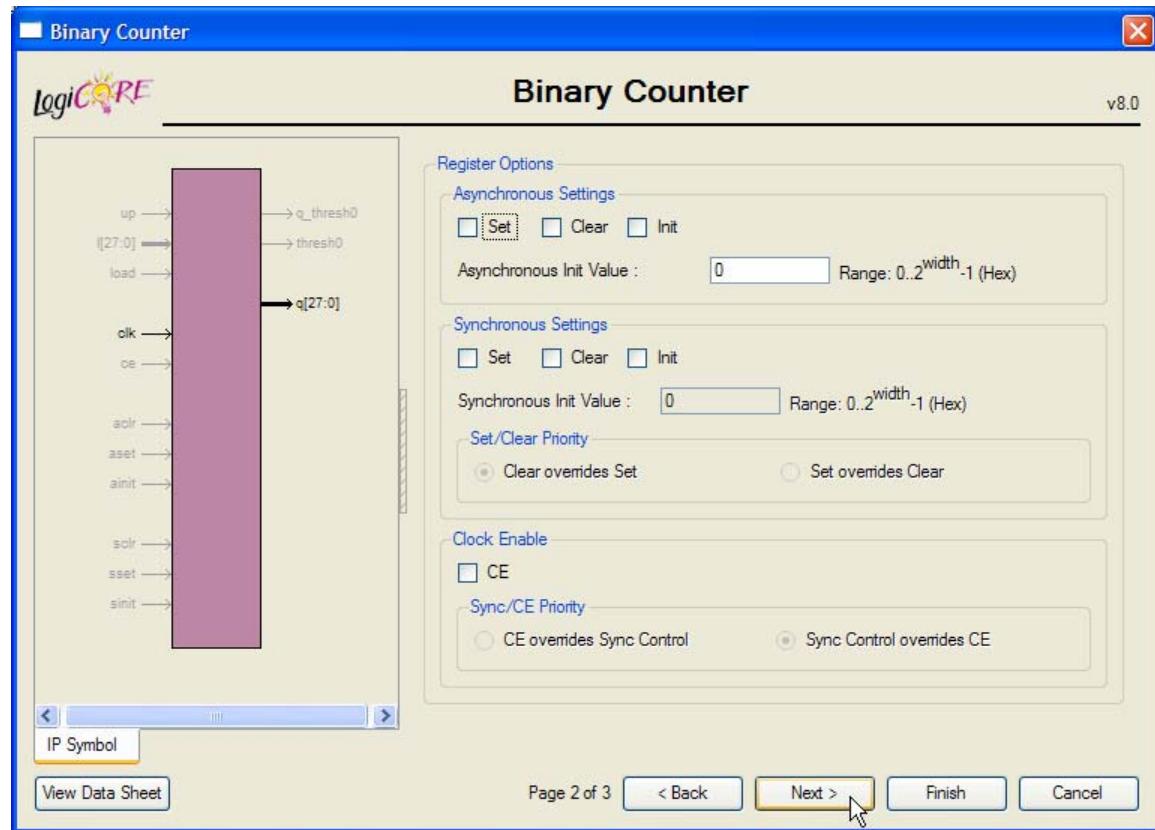
The initial configuration window specifies a sixteen-bit counter that increments its value upward by one on each clock cycle until it reaches 65,535 after which it rolls-over back to 0. You can change many of the features of this counter on the following screens. (Click on the View Data Sheet button to see all the details on how this counter can be configured.)

The only change that needs to be made for this design is to increase the counter width to 28 bits. Why build a 28-bit counter when only the upper four bits are used? The counter will be driven by a clock signal on the XSA Board that has a frequency of 50 MHz. The LED display would be changing much too quickly to see at this frequency. By connecting the LED decoder to the upper four bits of the 28-bit counter, the display will only change once in every 2^{24} clock cycles. So the LED display will change every $2^{24} / (50 \times 10^6) = 0.336$ seconds which is slow enough to read.

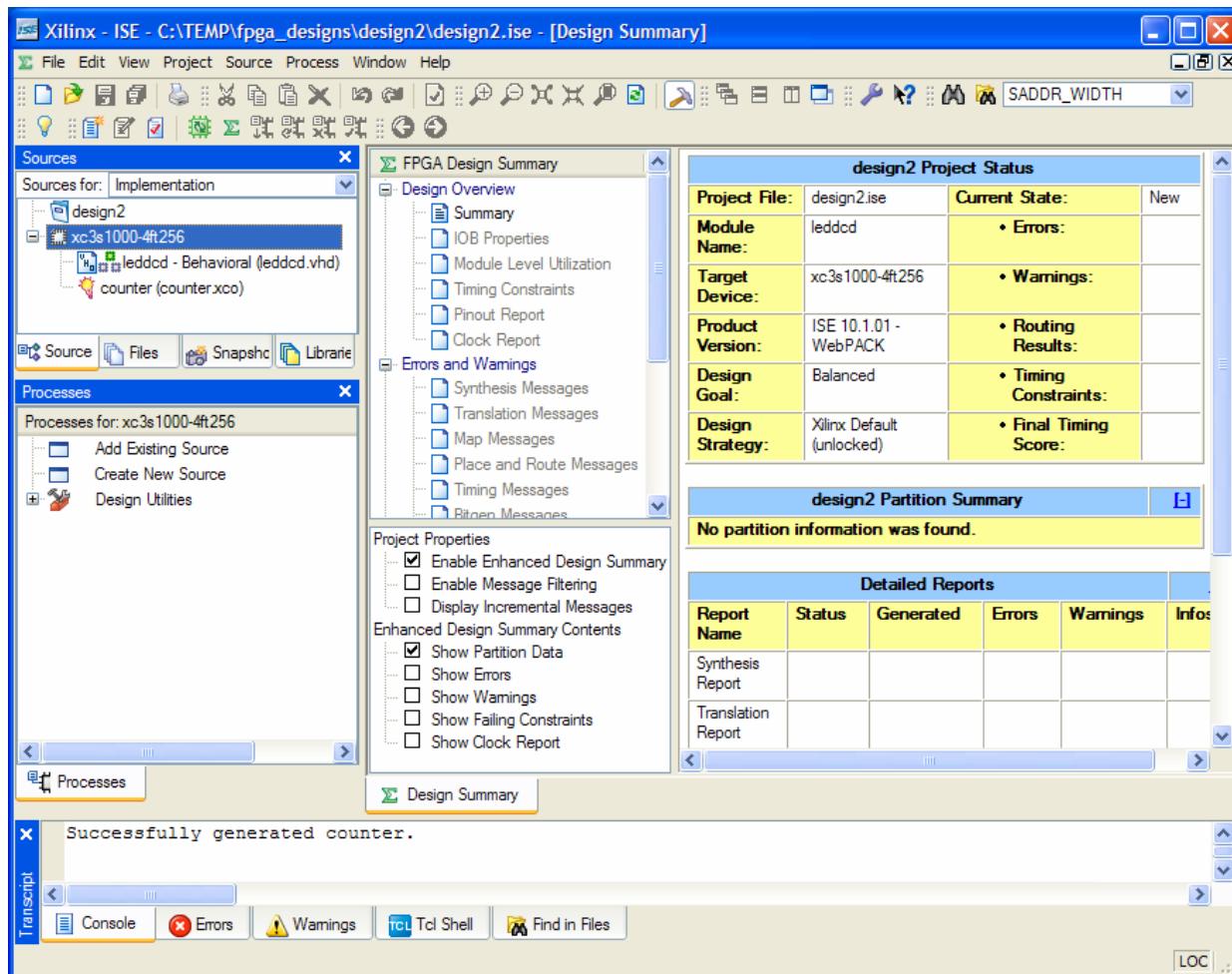
After changing the counter width, click on Next to continue configuring the counter.



None of the advanced configuration options for the counter are needed in this design, so click Next and Finish in the next two windows that appear.

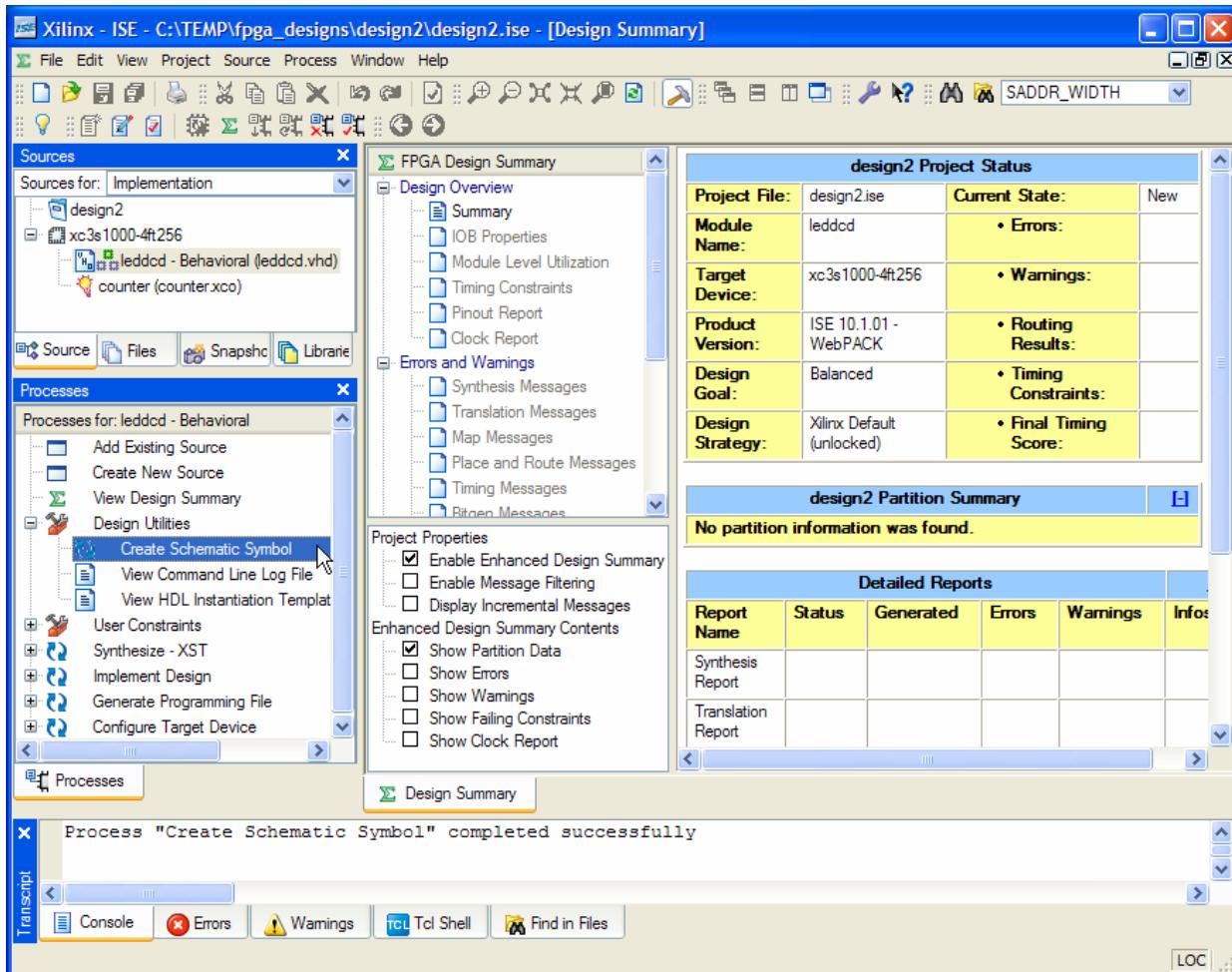


After clicking Finish in the final configuration window, the **counter** module appears in the Sources pane. The counter is stored as a Xilinx Coregen object (.xco file suffix). If you need to change the operation of the counter, just double-click the counter module and make your changes in the Coregen wizard windows that appear.

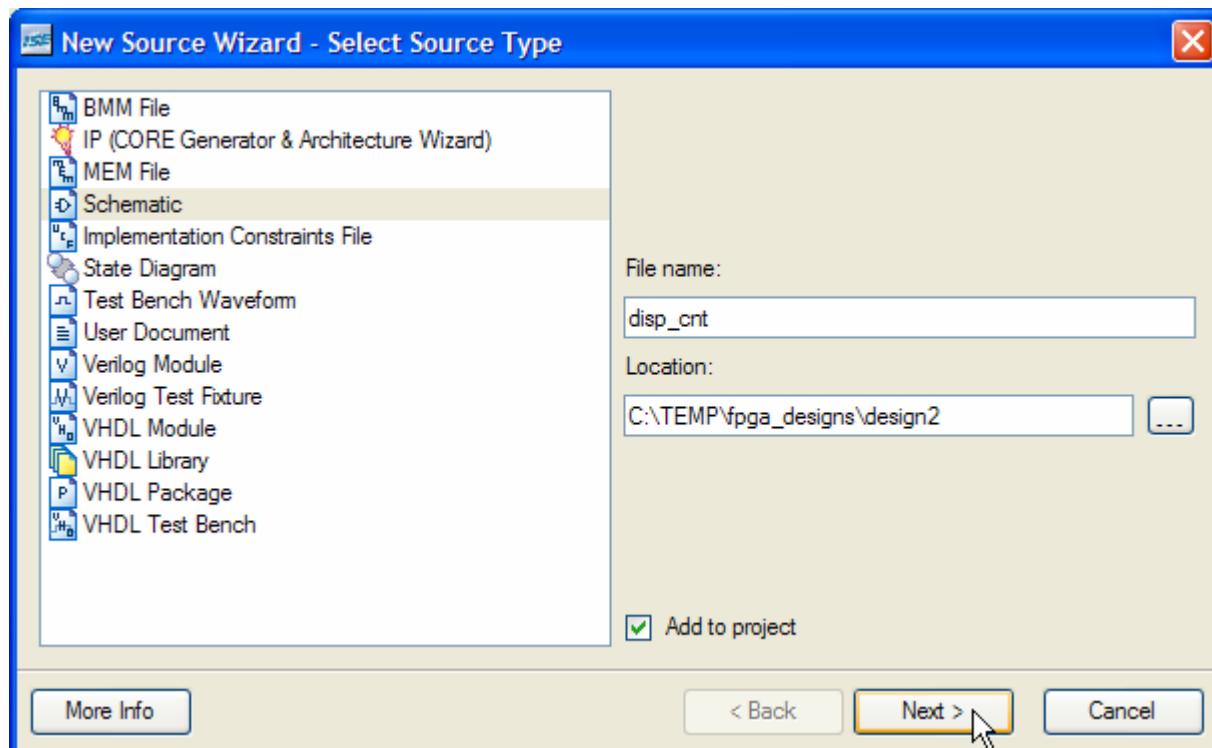


Tying Them Together

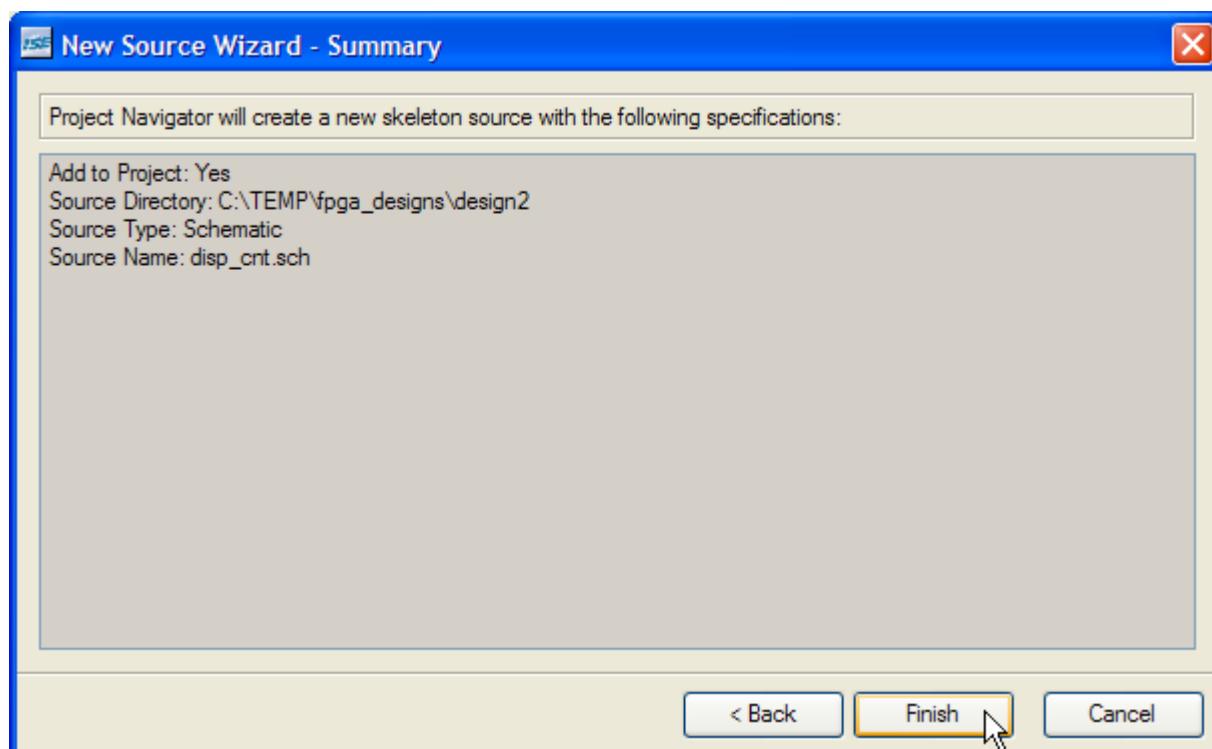
You have the LED decoder and the counter, but now you need to tie them together to build the displayable counter. You will do this by connecting the counter to the LED decoder in a top-level schematic. Before you can do this, you have to create a schematic symbol for the LED decoder module from its VHDL source code. (The counter module already has a schematic symbol since it was created using Coregen.) To create the LED decoder schematic symbol, highlight the ledgcd object in the Sources pane and then double-click the Create Schematic Symbol process. A message indicating the schematic symbol has been created will appear in the Console tab of the Transcript pane.



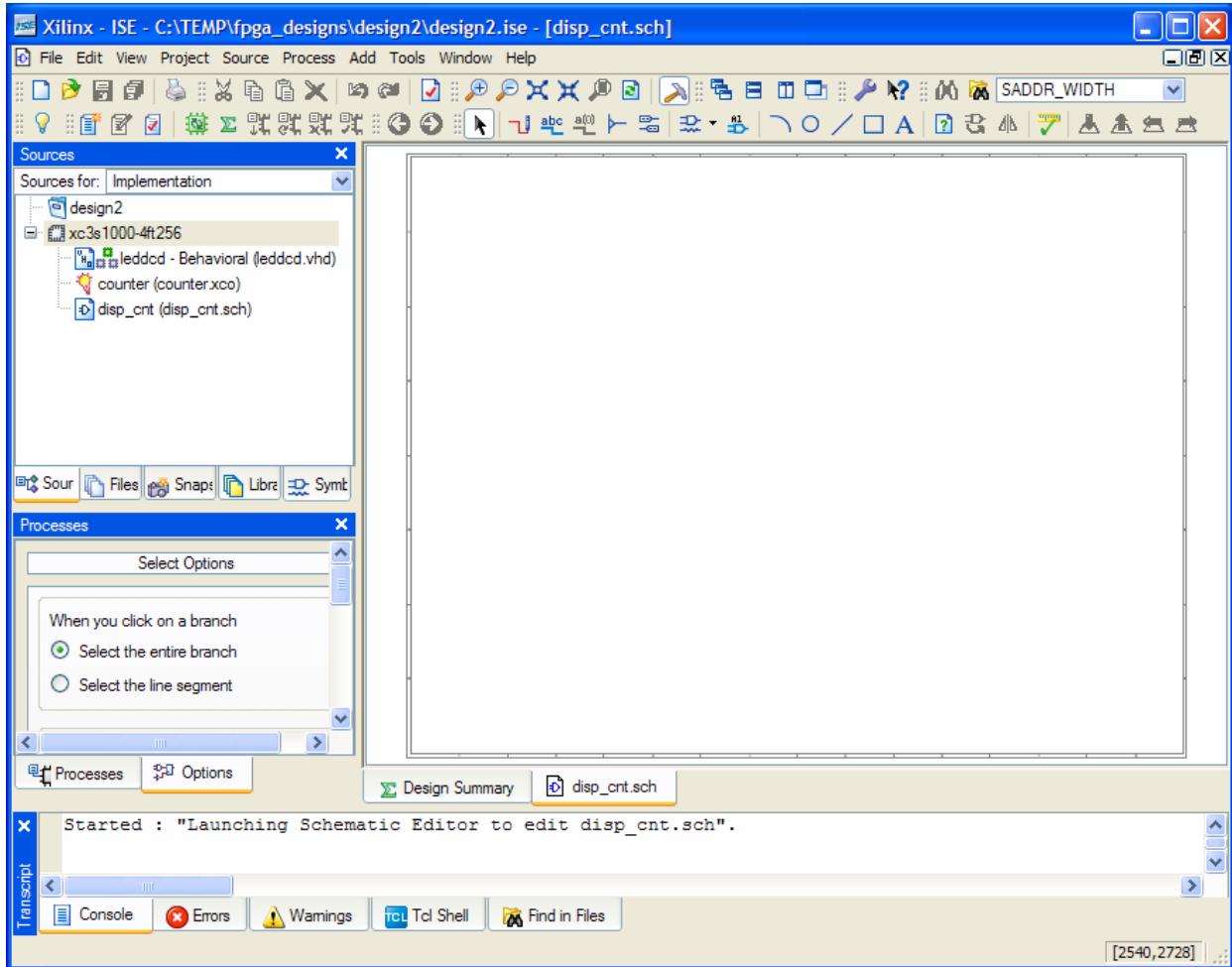
Now it is time to create the top-level schematic that will hold the counter and LED decoder symbols. Right-click on the xc3s1000-4ft256 object and select New Source... from the pop-up menu. Then highlight the Schematic entry in the **New Source** window and name the schematic **disp_cnt**. Then click on Next.



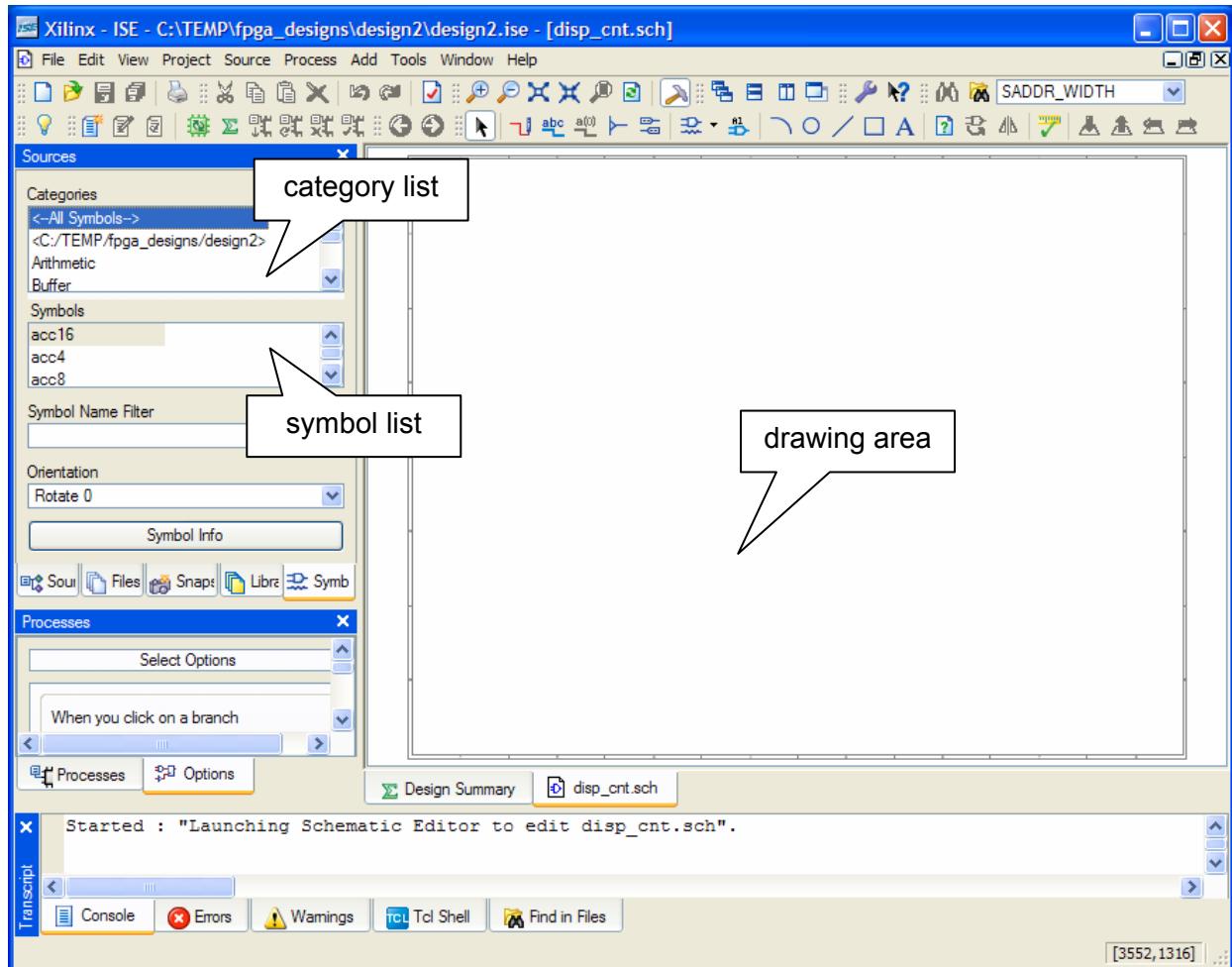
There is very little to do when initializing a schematic, so just click on the Finish button in the **Summary** window that appears.



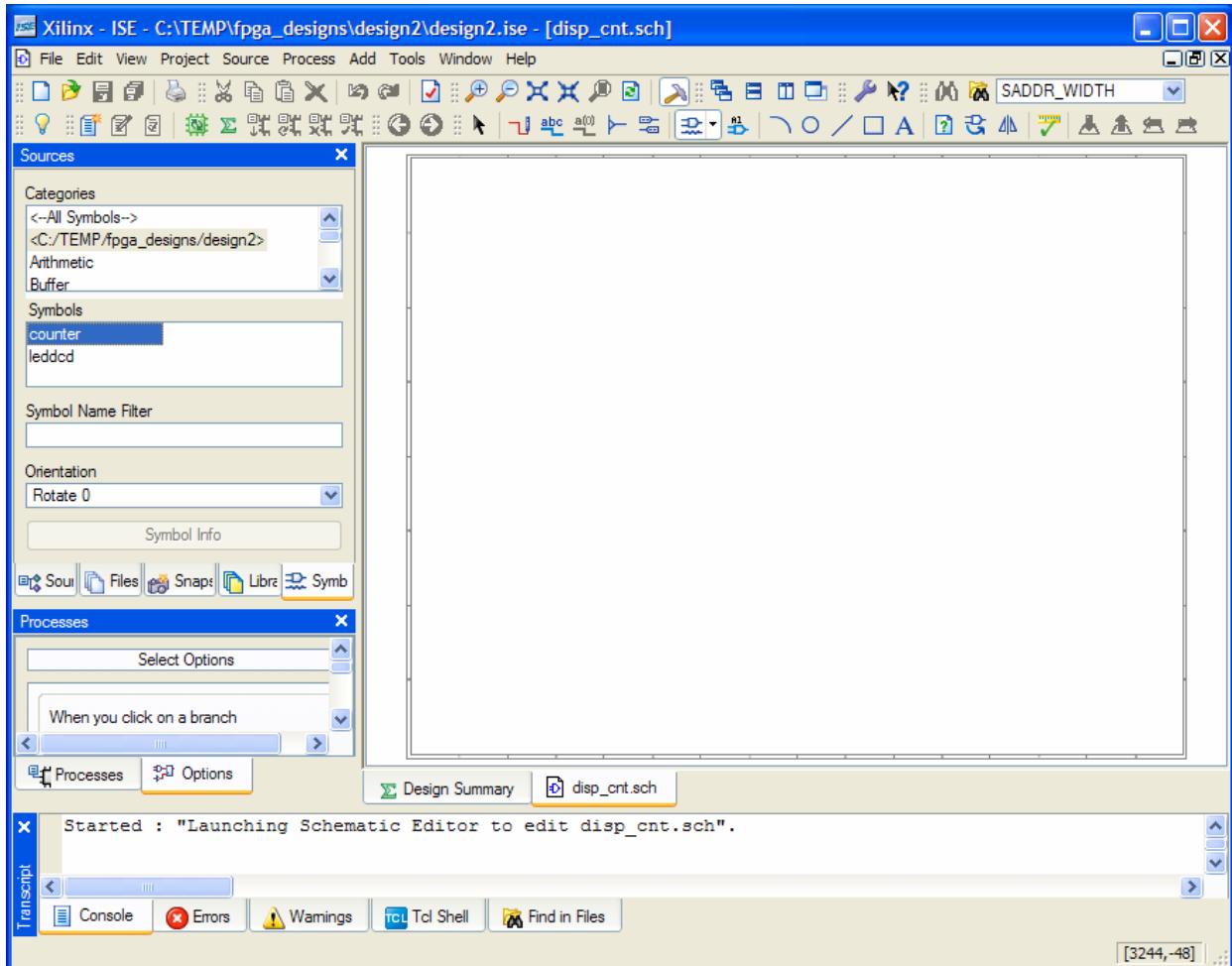
The disp_cnt schematic object has now been added to the Sources pane. You can double-click it to begin creating the schematic, but a schematic editor window should open automatically once the file is created.



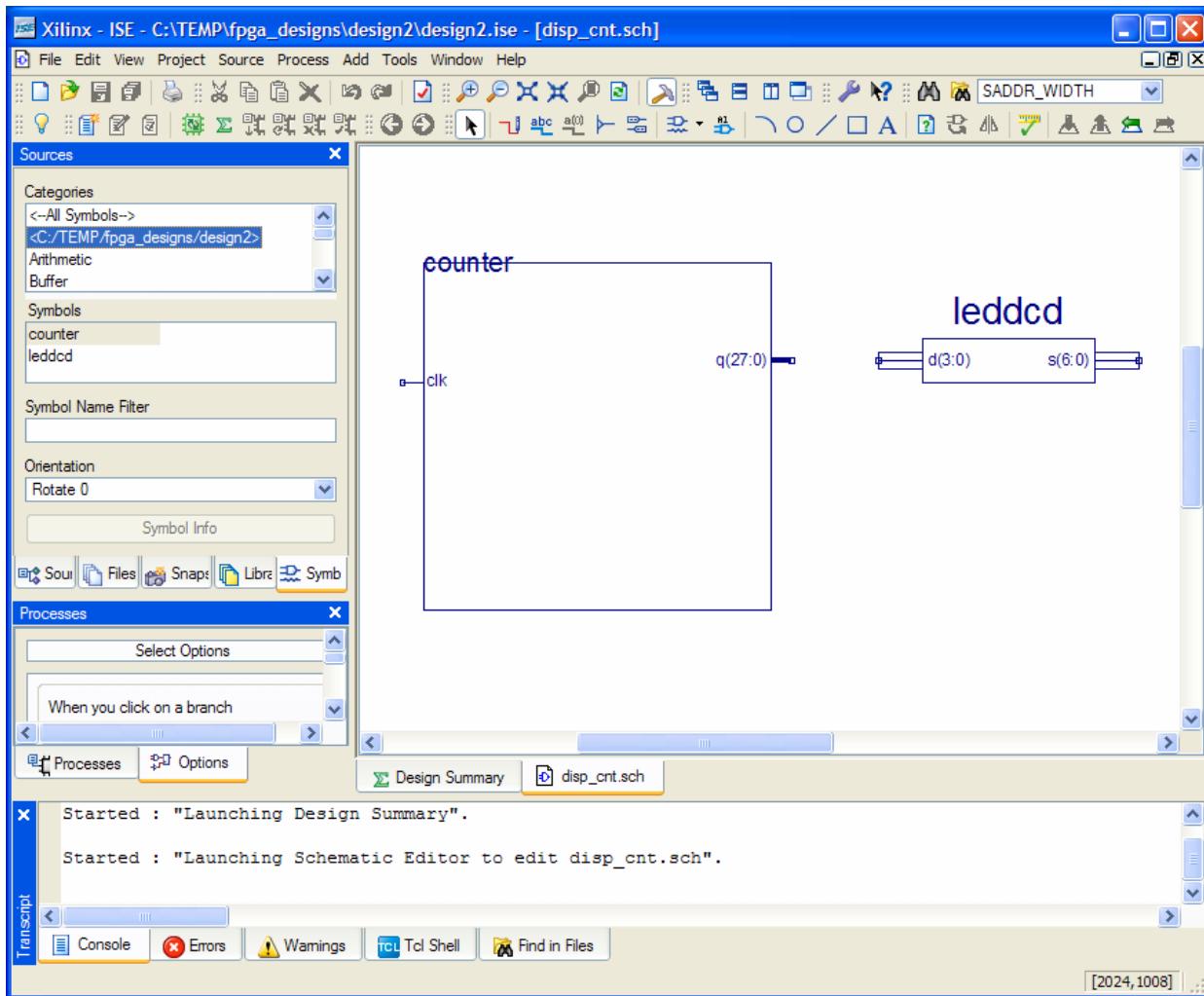
Click on the Symbols tab at the bottom of the Sources pane. The Symbols tab contains a list of categories for various logic circuit elements that can be used in a schematic. Below that is the list of circuit element symbols in the highlighted category. A symbol can be selected from this list and dropped into the drawing area to the right.



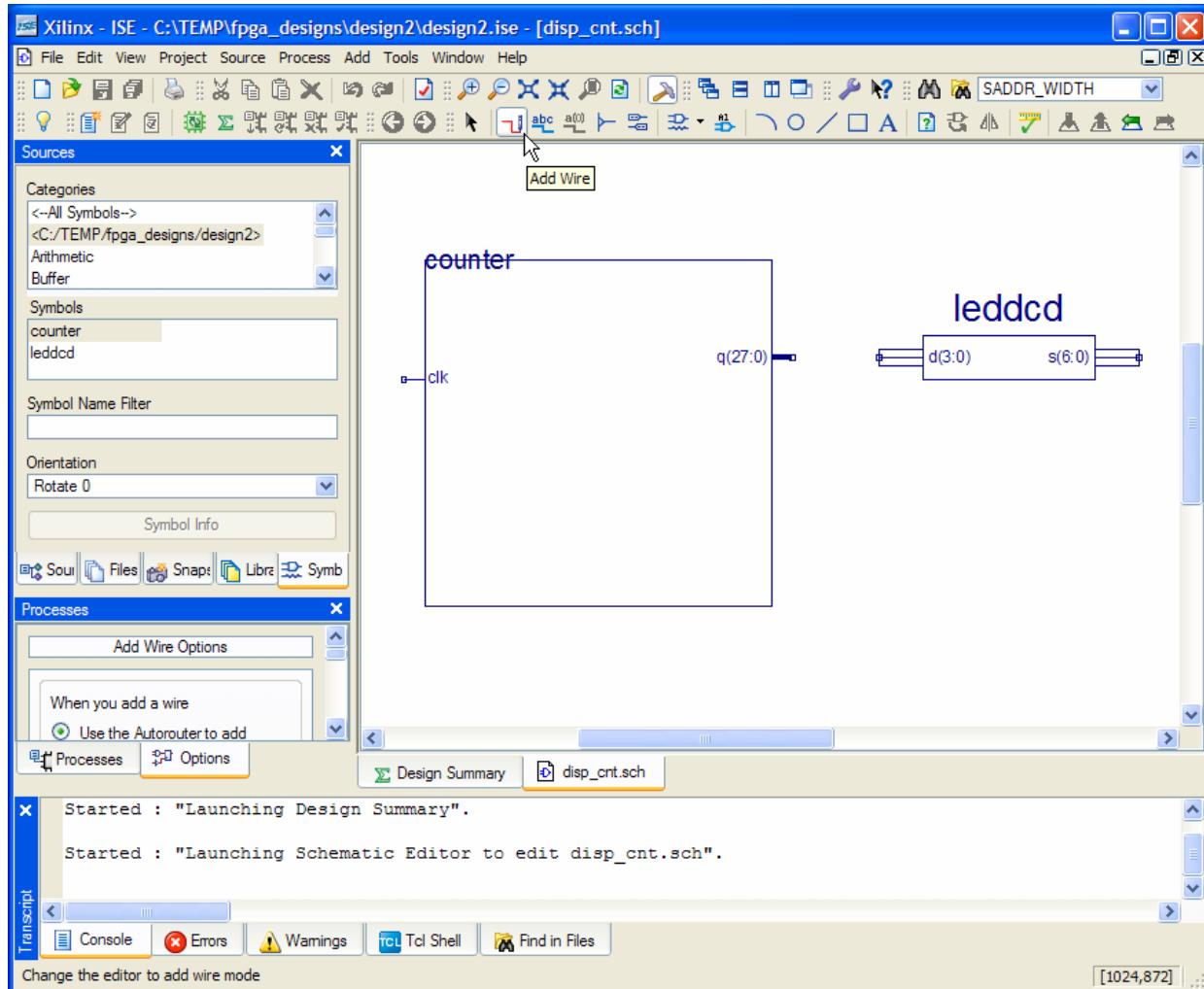
To start creating the top-level schematic, highlight the second entry in the category list. The c:/TEMP/fpga_designs/design2 category contains the schematic symbols for the **design2** project's counter and LED decoder modules. You can see the names of these modules in the symbol list.



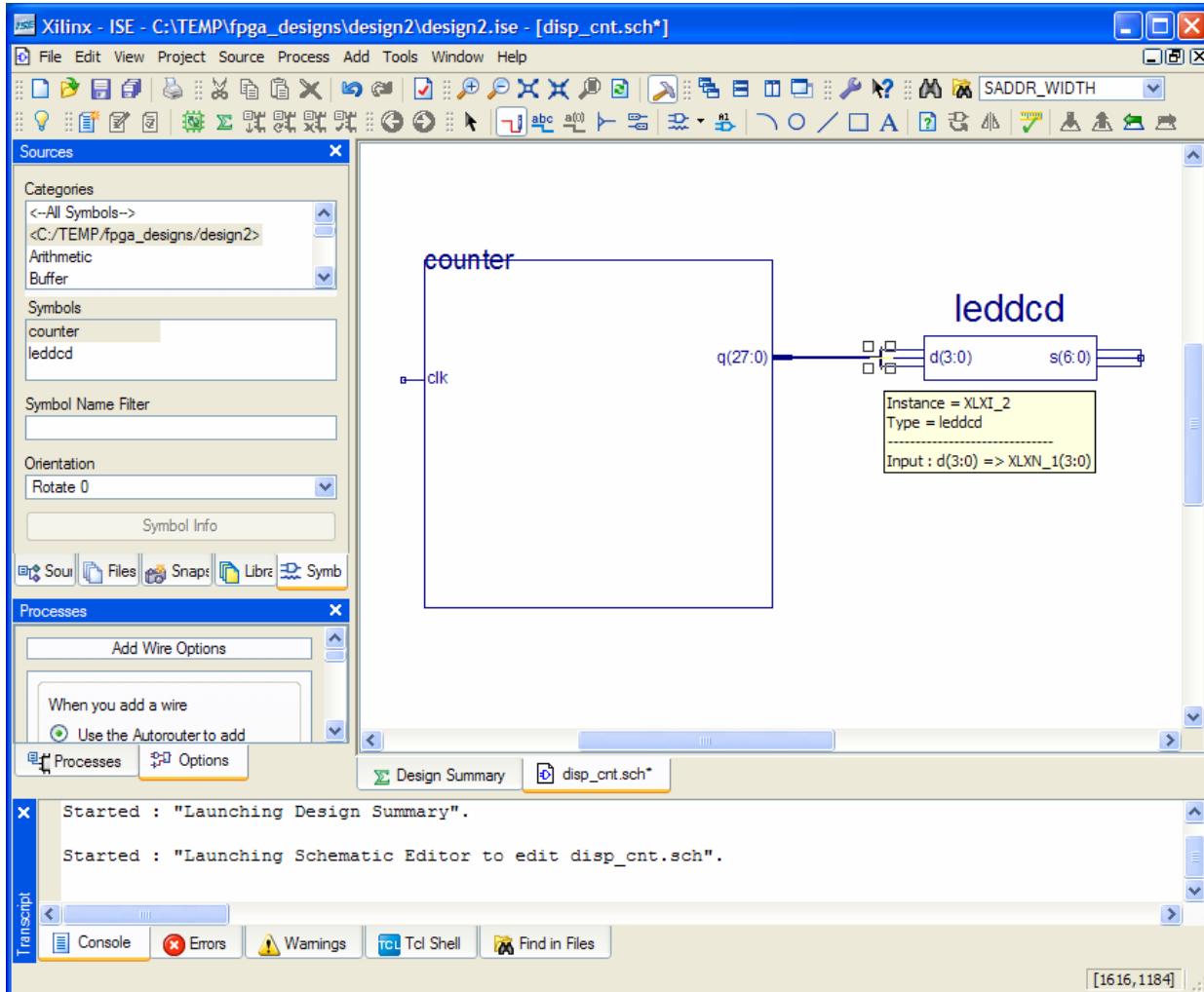
Click on the counter entry in the Symbols list. Then move the mouse cursor into the drawing area and left-click to place an instance of the **counter** module into the schematic. Repeat this process with the **leddcd** module to arrive at the arrangement shown below. (For clarity, I used the Edit→Preferences command to turn off the display of the grid in the schematic drawing area.)



Next, click on the  button to begin adding wires to the schematic.



Left-click the mouse on the **q(27:0)** bus on the right-hand edge of the **counter** module. Then left-click on the **d(3:0)** bus on the left-hand edge of the **leddcd** module. This creates a four-bit bus between the output of the counter and the input of the LED decoder.



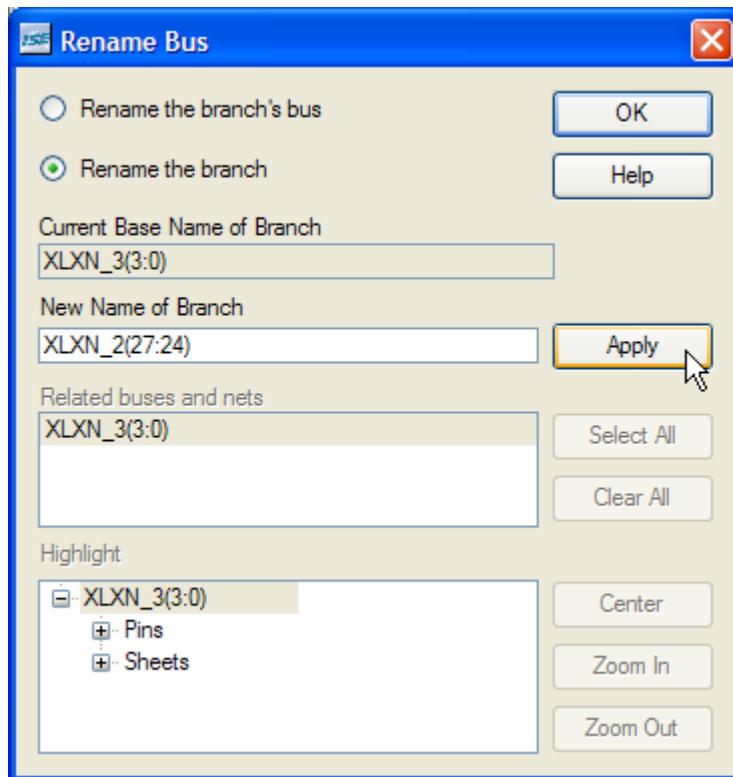
However, the four-bit bus causes a problem. Click on the button to run a design-rule check. An error is reported in the Console tab:

```
Error:DesignEntry:20 - disp_cnt.sch: Pin 'q(27:0)' is connected to a
bus of a different width.
```

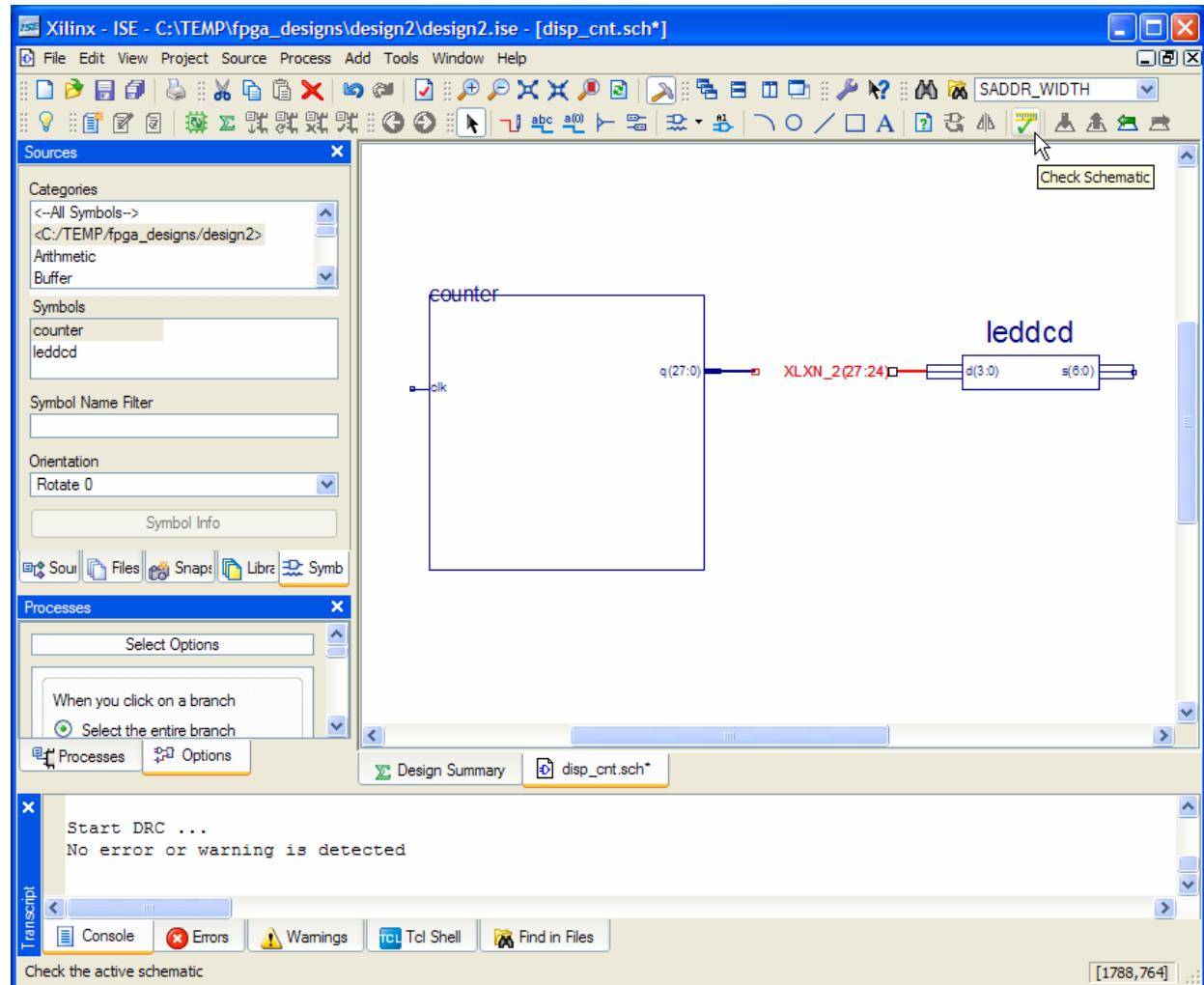
You can highlight the place in the schematic where the error occurs by clicking on the `disp_cnt.sch` text in the error message. The error is caused by the mismatch between the 28-bit counter output and the four-bit bus. This problem is fixed as follows:

1. Click on the selection tool button: .
2. Right-click on the existing bus and select Delete from the pop-up menu to remove this bus.

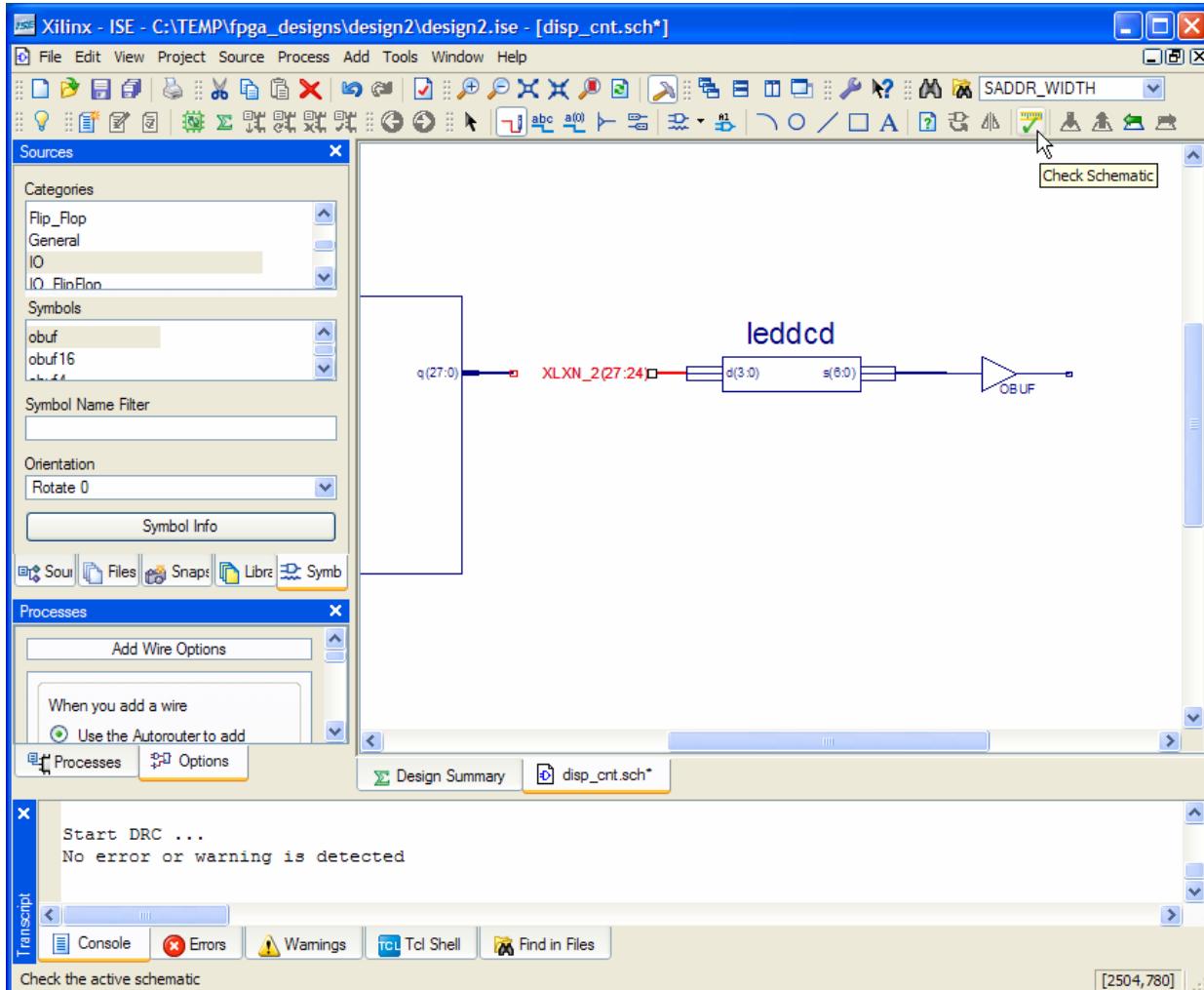
3. Click on the wiring tool, , click on the counter output, and draw a small bus stub outward. Terminate the bus with a double-click. This will create a 28-bit bus connected to the counter outputs.
4. Repeat the previous step to attach a four-bit bus to the LED decoder input.
5. Click on the selection tool, , and then hover the mouse pointer over the bus connected to the counter. The name of the bus will be shown. In my project, the bus is named **XLXN_2(27:0)**.
6. Right-click on the bus connected to the LED decoder and select Rename Selected Bus... in the pop-up menu.
7. Rename the LED decoder input bus from **XLXN_3(3:0)** to **XLXN_2(27:24)** as shown below and click on the Apply button followed by the OK button. This will connect the LED decoder inputs to the four most-significant bits of the counter output bus.



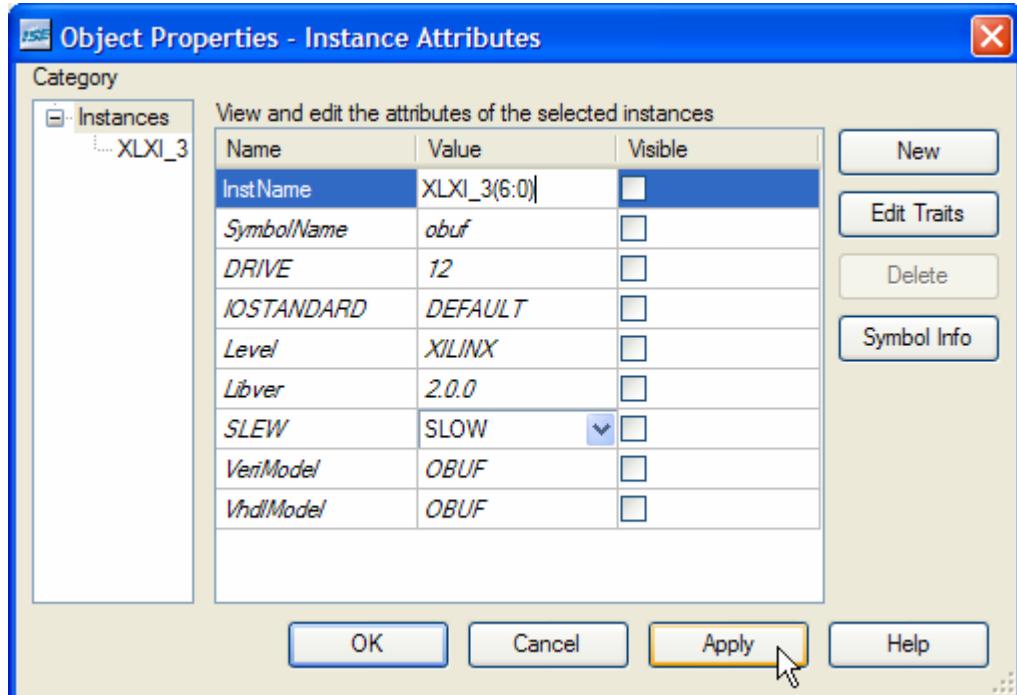
At this point, the schematic should appear as shown below. The design-rule checker should no longer detect any problems.



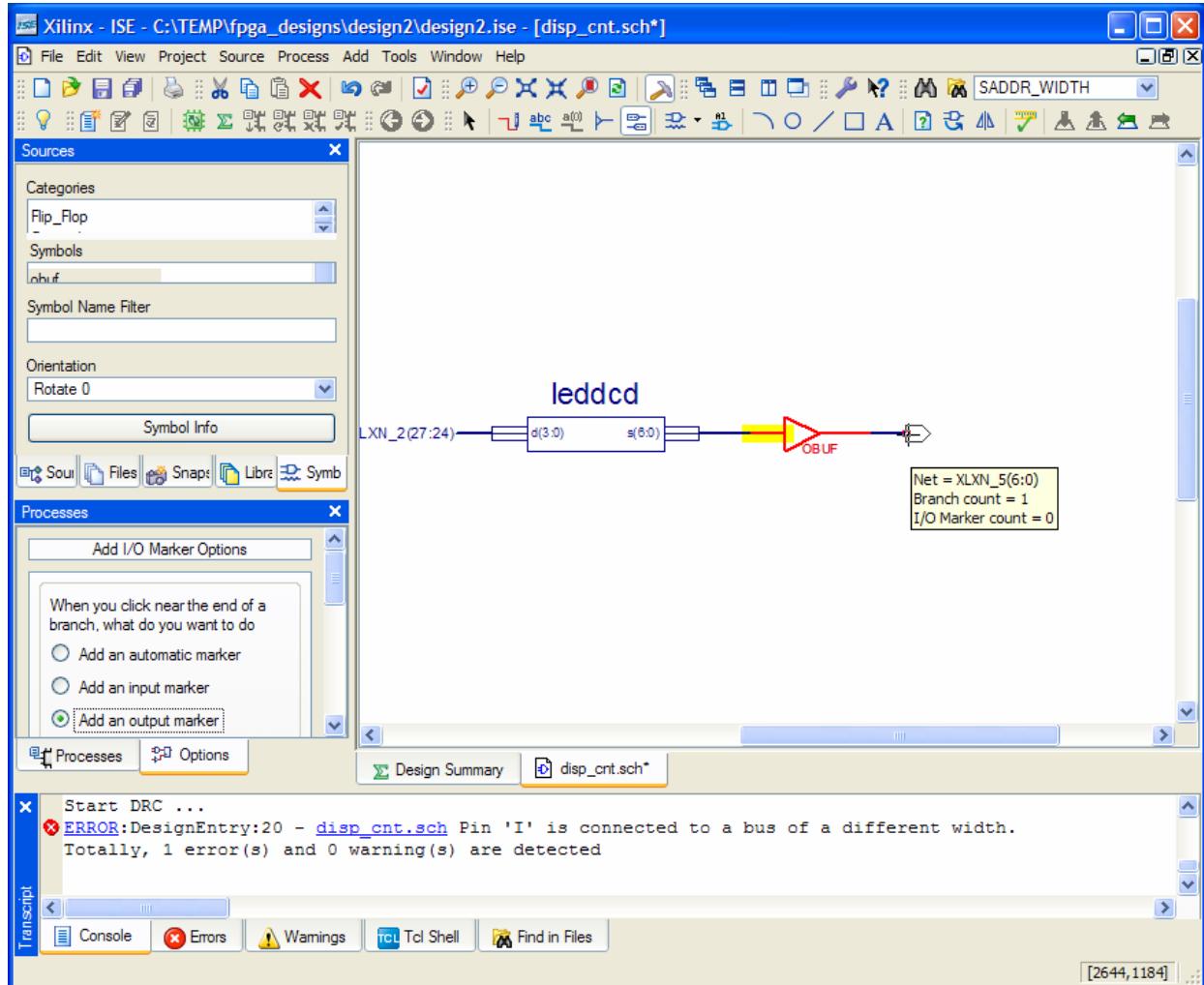
Now highlight the IO category and select a single-bit output buffer (obuf) from the list of symbols. Attach the output buffer to the output of the LED decoder as shown below. Then run the design-rule checker.



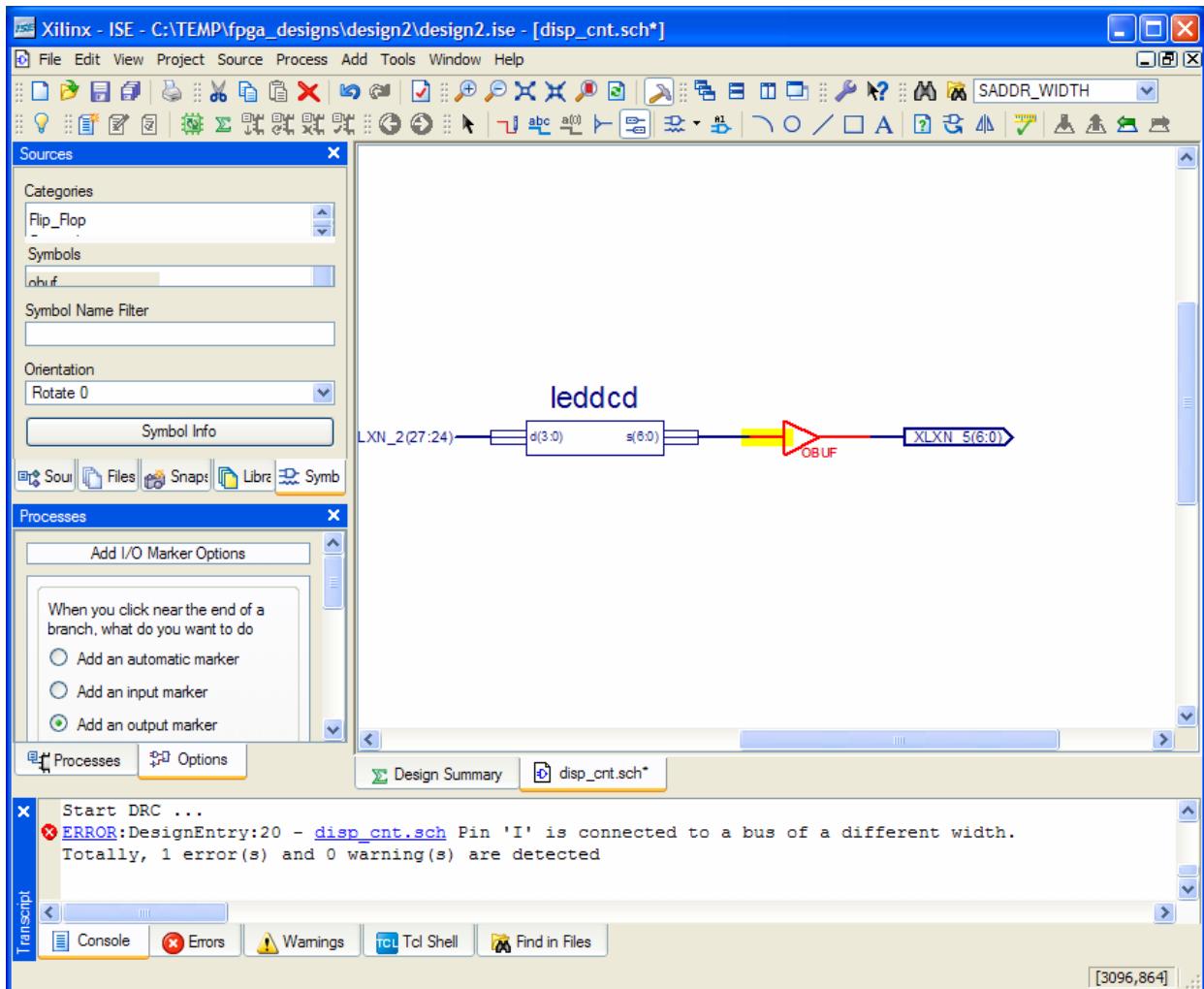
The design-rule checker will detect a width-mismatch error in the above schematic since a single output buffer has been attached to the seven-bit LED decoder output bus. This error can be fixed by right-clicking on the OBUF symbol and selecting Object Properties from the pop-up menu. Then append (6:0) to the instance name of the output buffer as shown below and click on Apply followed by OK. This will expand the single buffer into an array of seven buffers with each one connected to a bit of the LED decoder output bus.



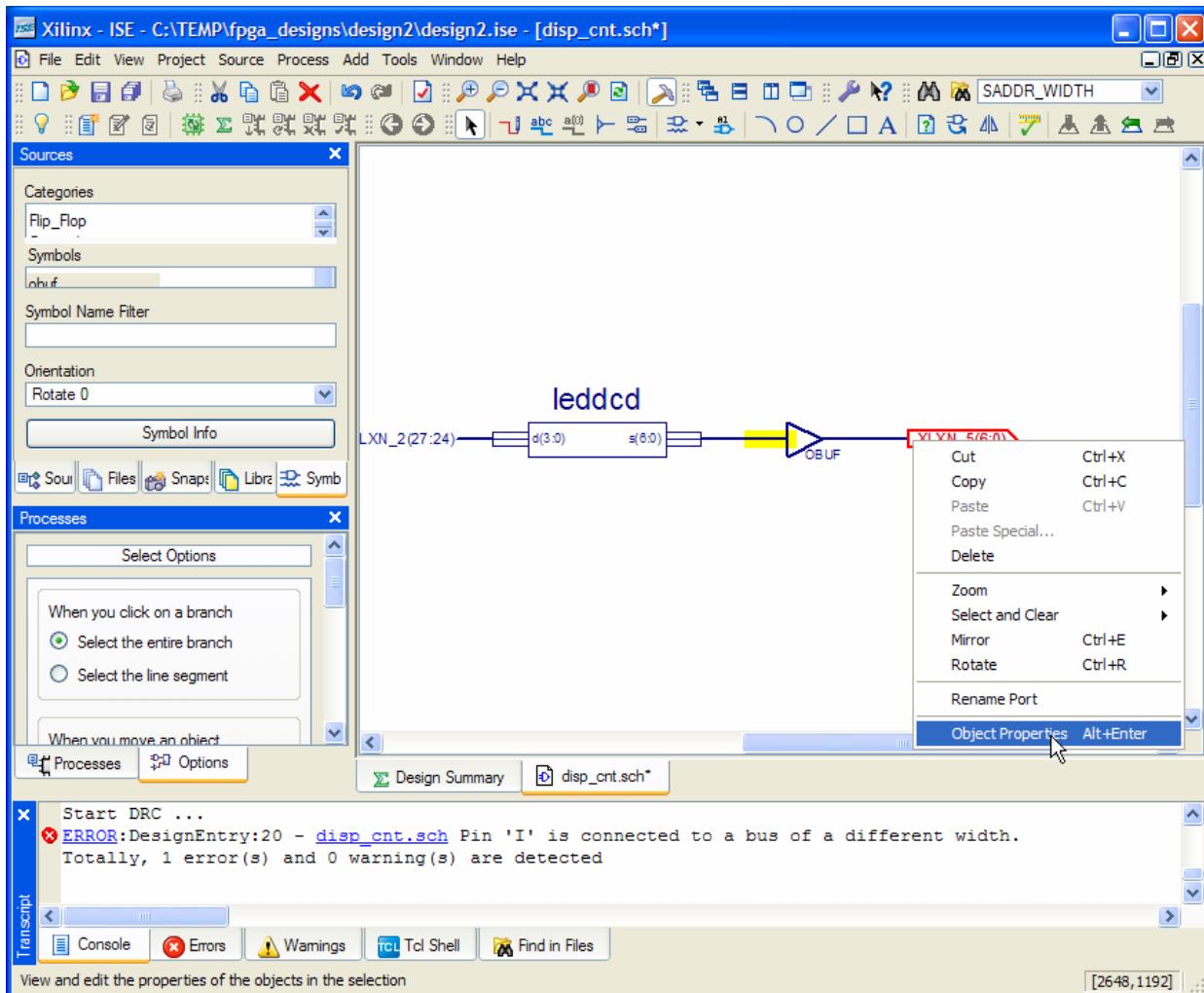
Next, attach a short bus segment to the output of the OBUF. Then click on the  button for adding I/O markers. Click on the Add an output marker button in the Options tab of the **Processes** pane and then click on the free end of the wire segment that you just added.



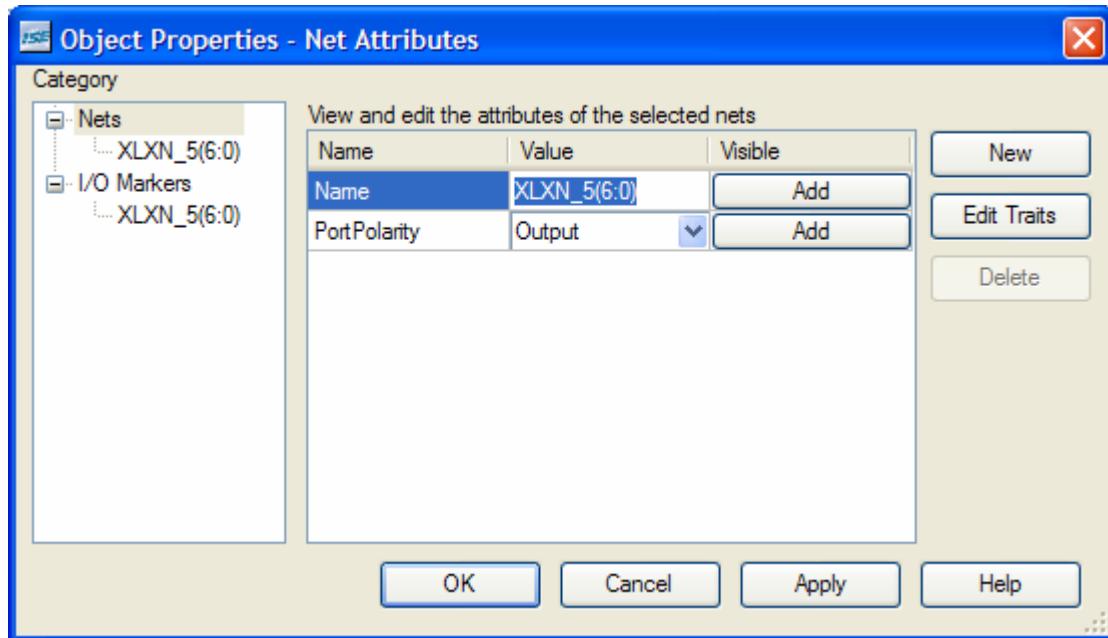
Clicking on the end of the wire creates a seven bit-wide set of output pins.



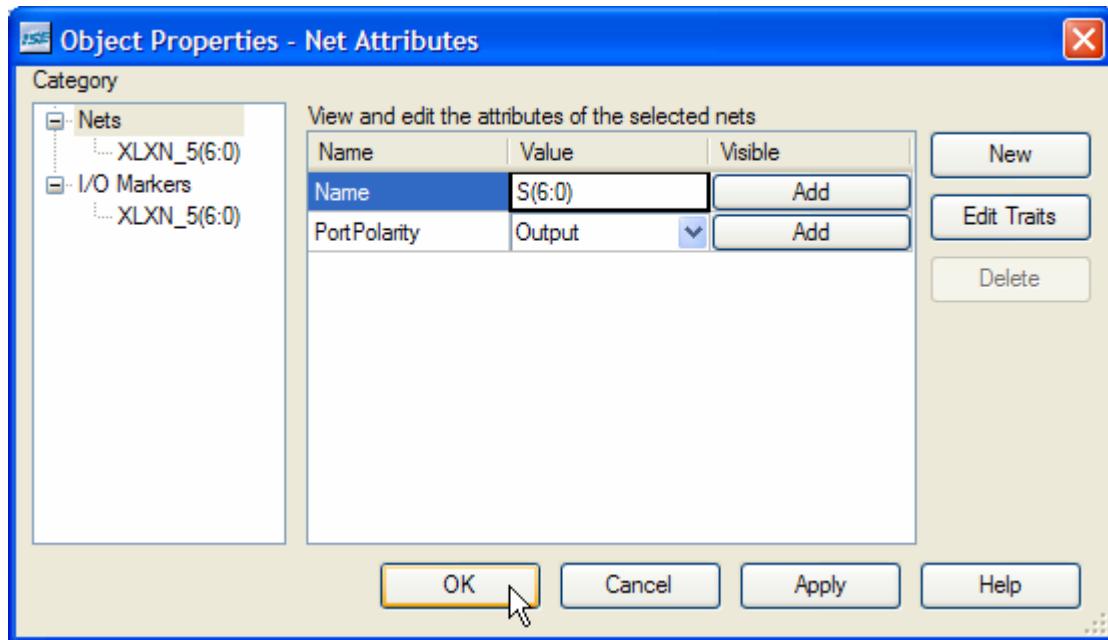
The output pins automatically assume the same name as the bus to which they are attached, but this name was automatically generated and doesn't carry a lot of meaning. To change the name of the outputs (and the associated bus), right-click on the I/O marker and select Object Properties... from the pop-up menu.



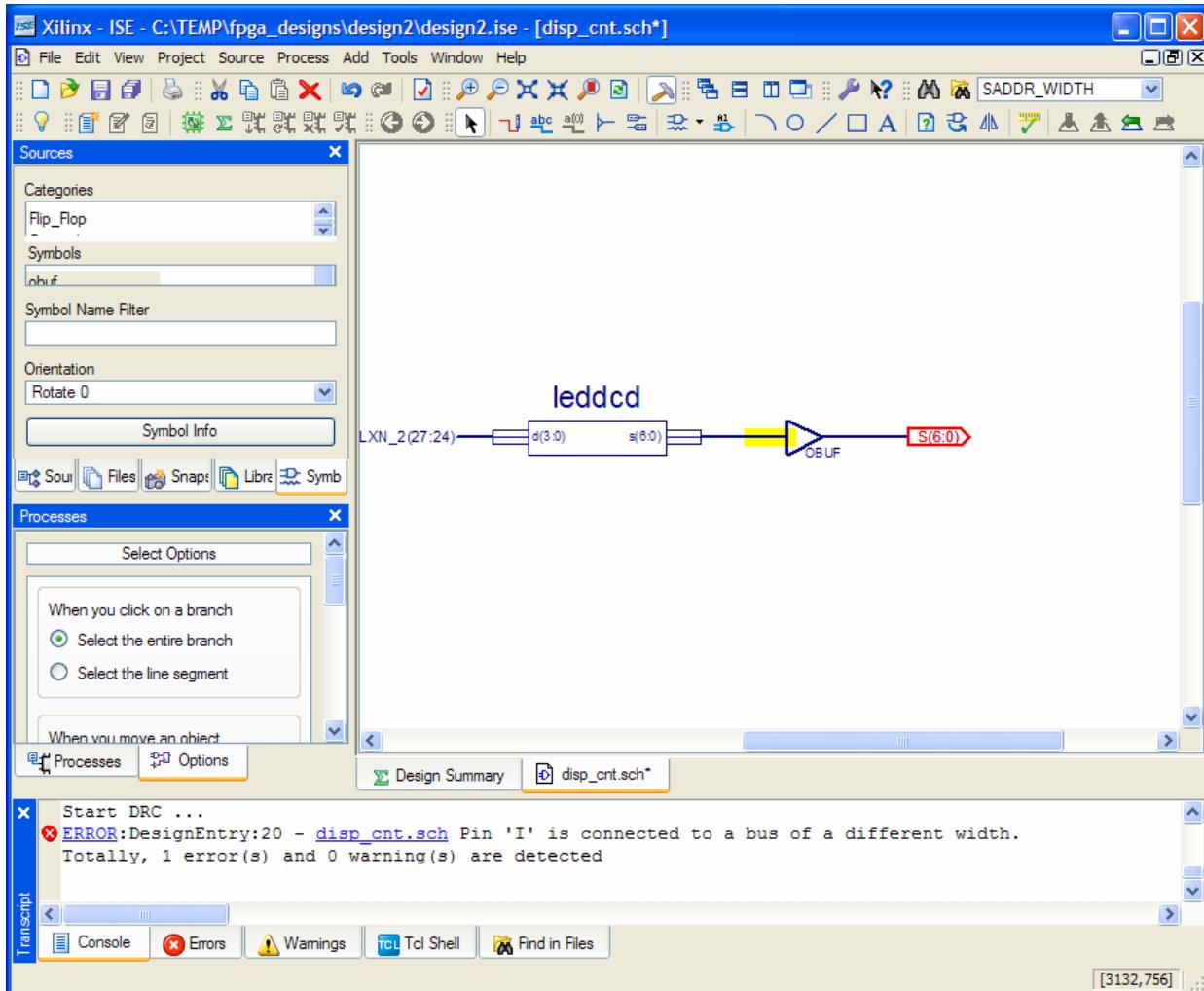
The **Object Properties** window allows you to set the name and direction of the pins.



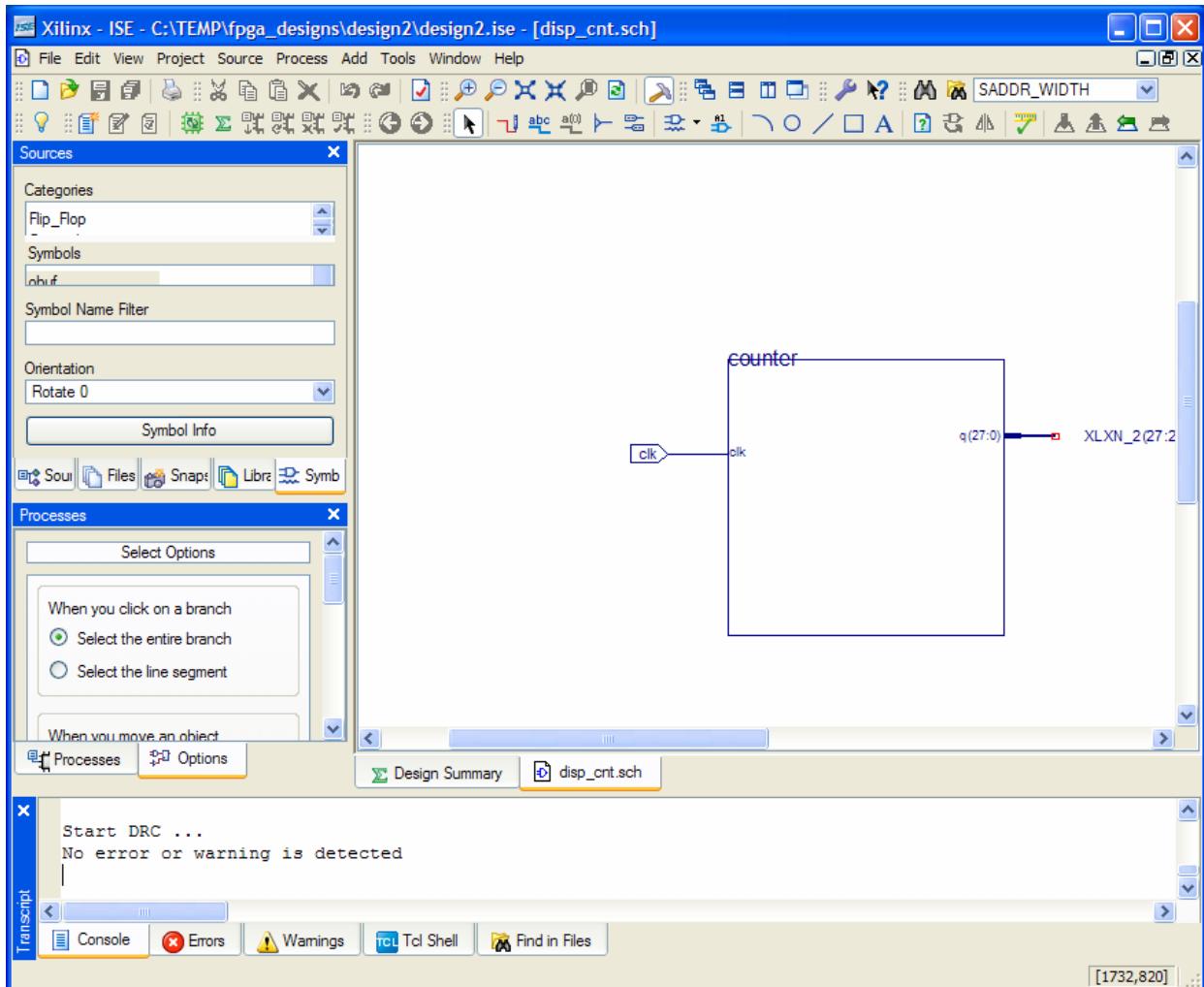
Double-click the existing bus name and replace it with **S(6:0)**. The direction of the bus pins is already set to Output so you can finish by clicking on the OK button.



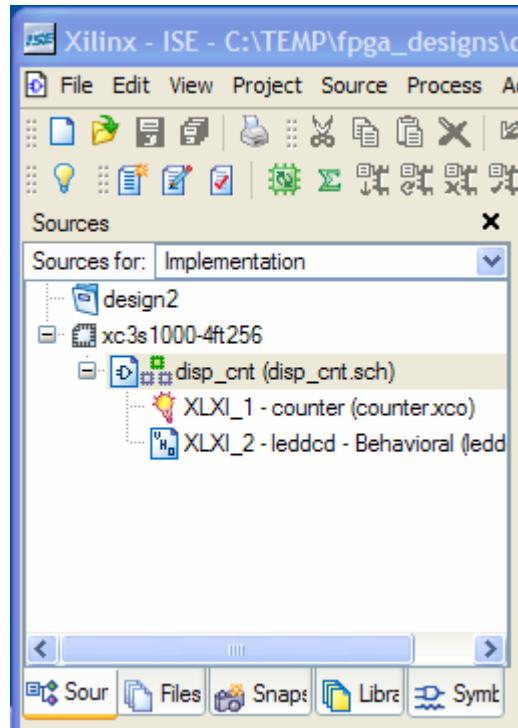
The output pins now appear with their new name.



Once the outputs from the circuit are in place, you can connect a single input I/O marker to the clock input of the counter. (No input buffer is needed because the clock signal will enter the FPGA through a dedicated global clock input.) Right-click (or double-click) on the I/O marker and rename it to `clk`. After this, perform another schematic check to detect any errors and save the schematic using the File→Save command.

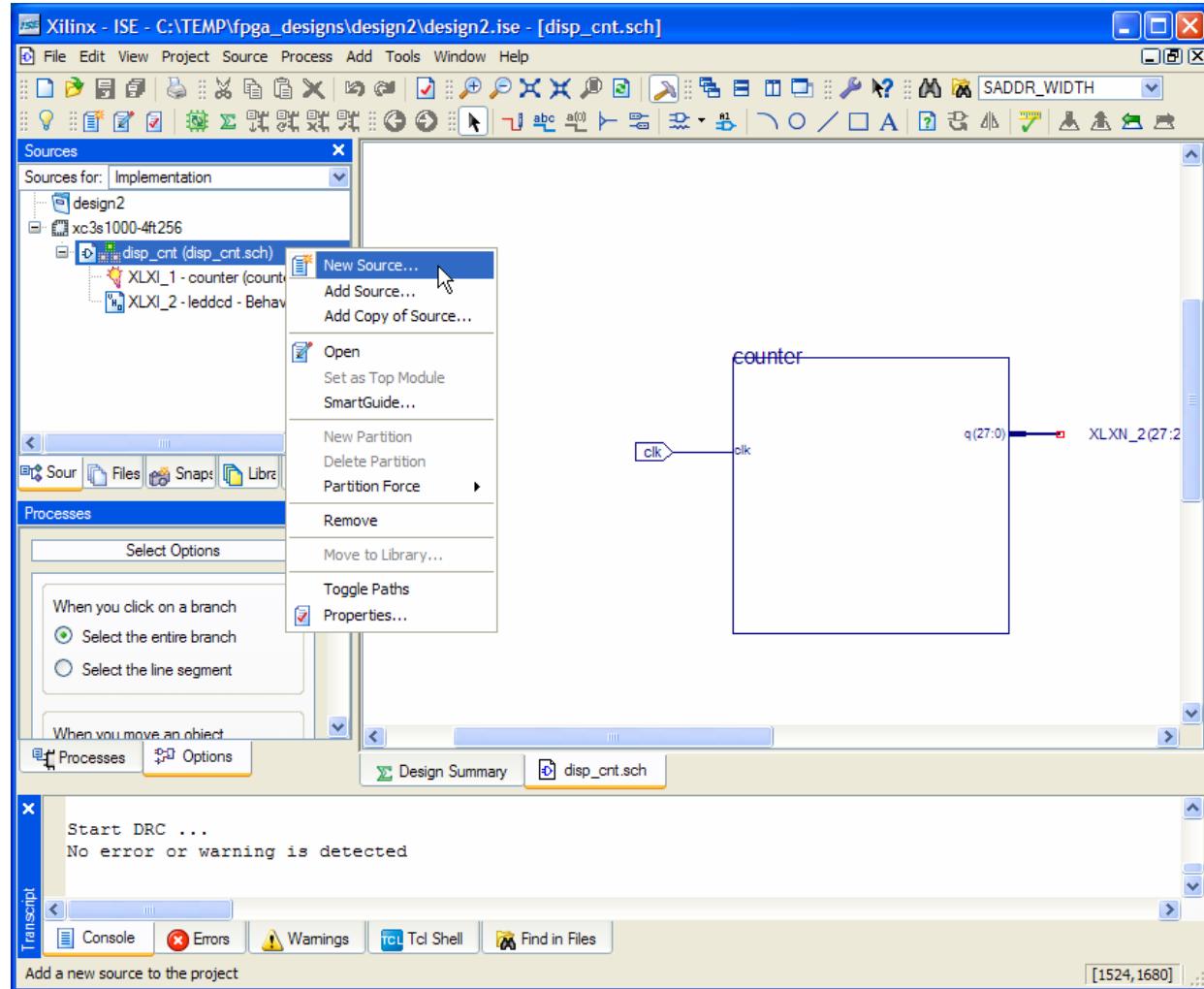


Once you save the schematic for the top-level module, the hierarchy in the Sources pane gets updated. Now the **counter** and **leddcd** modules are shown as lower-level modules that are included within the top-level **disp_cnt** module.

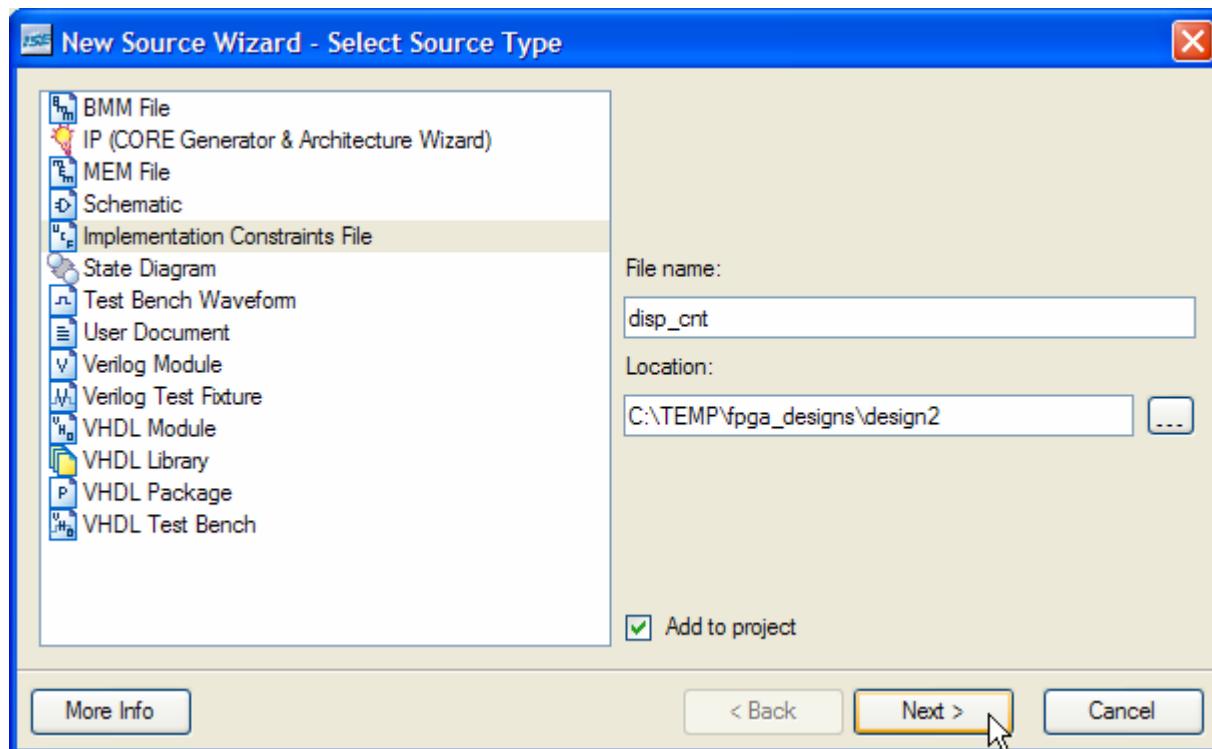


Constraining the Design

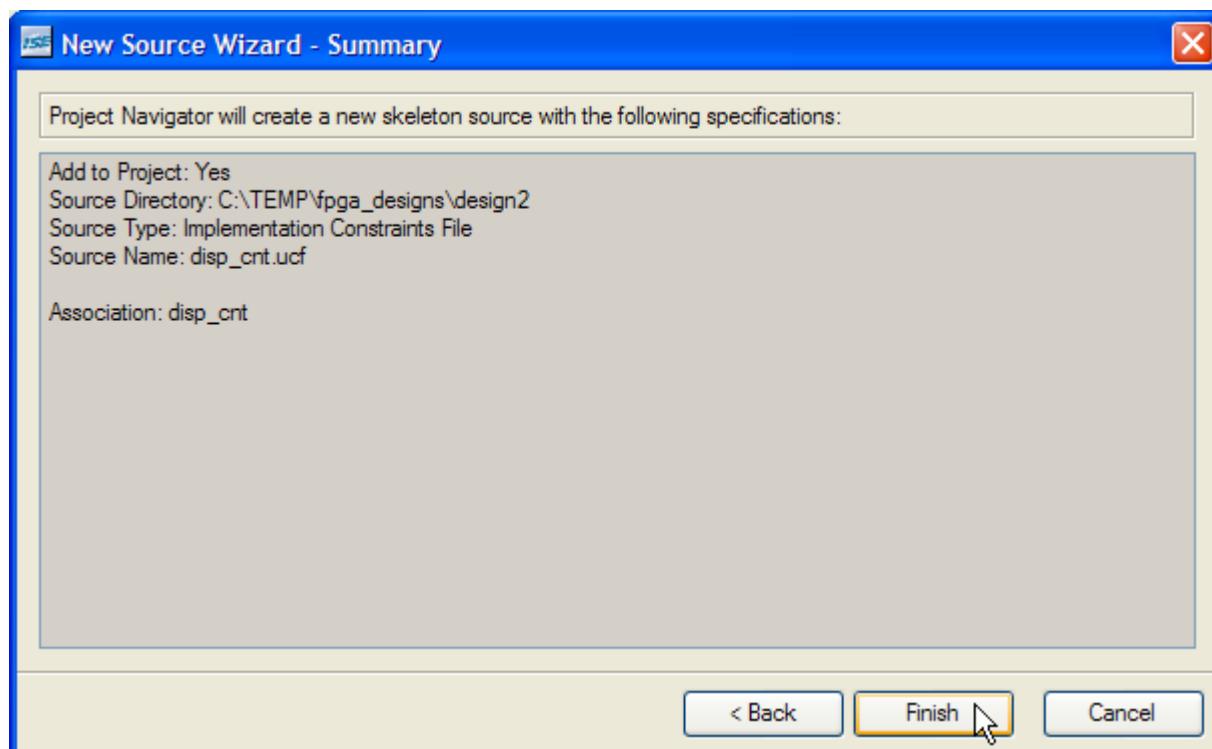
Before synthesizing the displayable counter, you need to assign pins to the inputs and outputs. Start by right-clicking the disp_cnt object in the Sources pane and selecting New Source... from the pop-up menu.



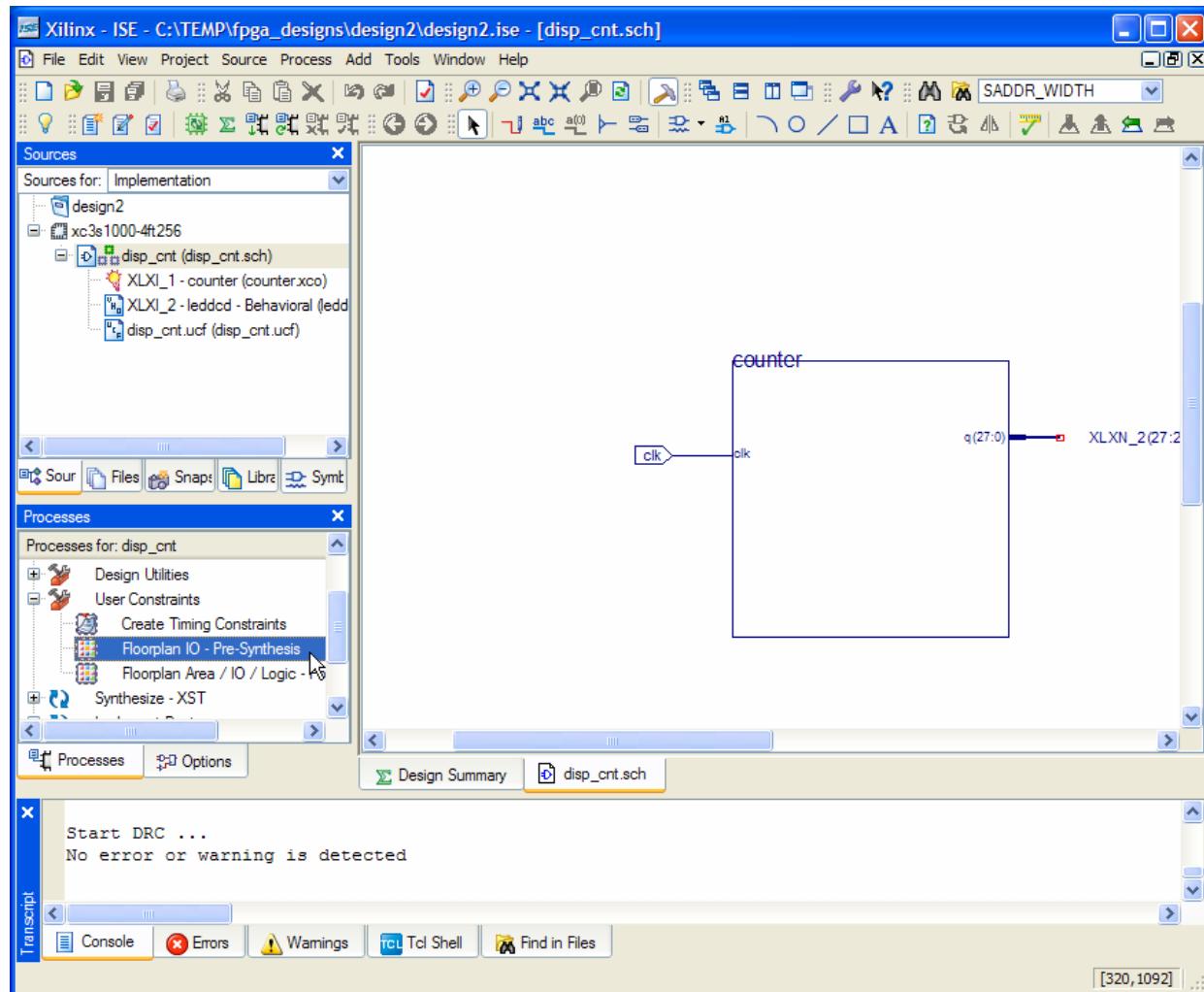
Select Implementation Constraints File as the type of source file to add and type `disp_cnt` in the File Name field. Then click on the Next button.



You will receive a feedback window that shows the name and type of the file you created and the file to which it is associated. The pin assignments will be made for the top-level module in the design hierarchy, so the constraints file is associated with the `disp_cnt` module. Click on the Finish button to complete the addition of the `disp_cnt.ucf` file to this project.



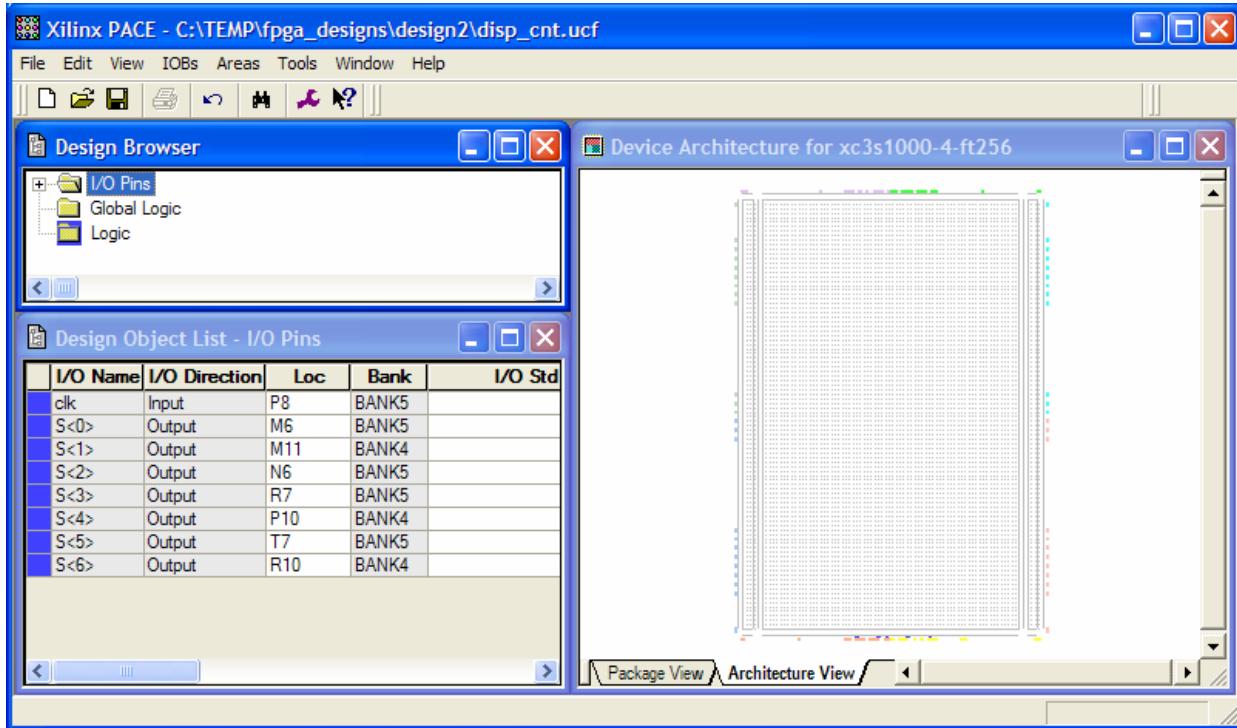
Now highlight the disp_cnt.sch object in the Sources pane and double-click the Floorplan IO – Pre-Synthesis process to begin adding pin assignment constraints to the design.



The appropriate pin assignments for each model of XSA Board are shown below. The **clk** input is assigned to a dedicated clock input on each FPGA to which a 50 MHz clock signal is applied. The seven-segment LED pin assignments are the same as in the previous design.

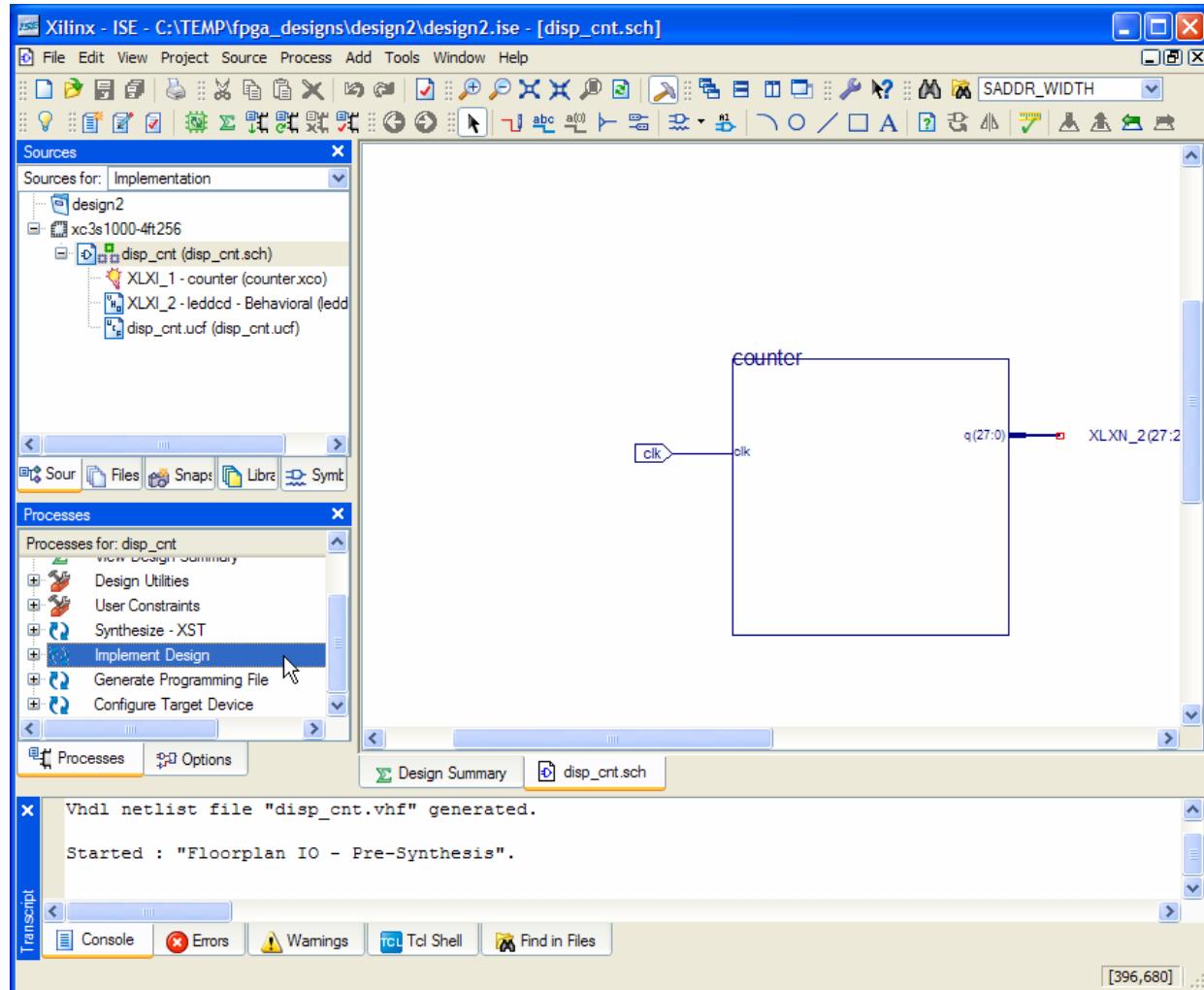
I/O Signal	XSA-50	XSA-100	XSA-200	XSA-3S1000
clk	P88	P88	B8	P8
s0	P67	P67	N14	M6
s1	P39	P39	D14	M11
s2	P62	P62	N16	N6
s3	P60	P60	M16	R7
s4	P46	P46	F15	P10
s5	P57	P57	J16	T7
s6	P49	P49	G16	R10

In the **Design Object List – I/O Pins** pane of the **Xilinx PACE** window that appears, set the pin assignments for the clock input and LED segment drivers as shown below. Then save the pin assignments and close the **PACE** window.



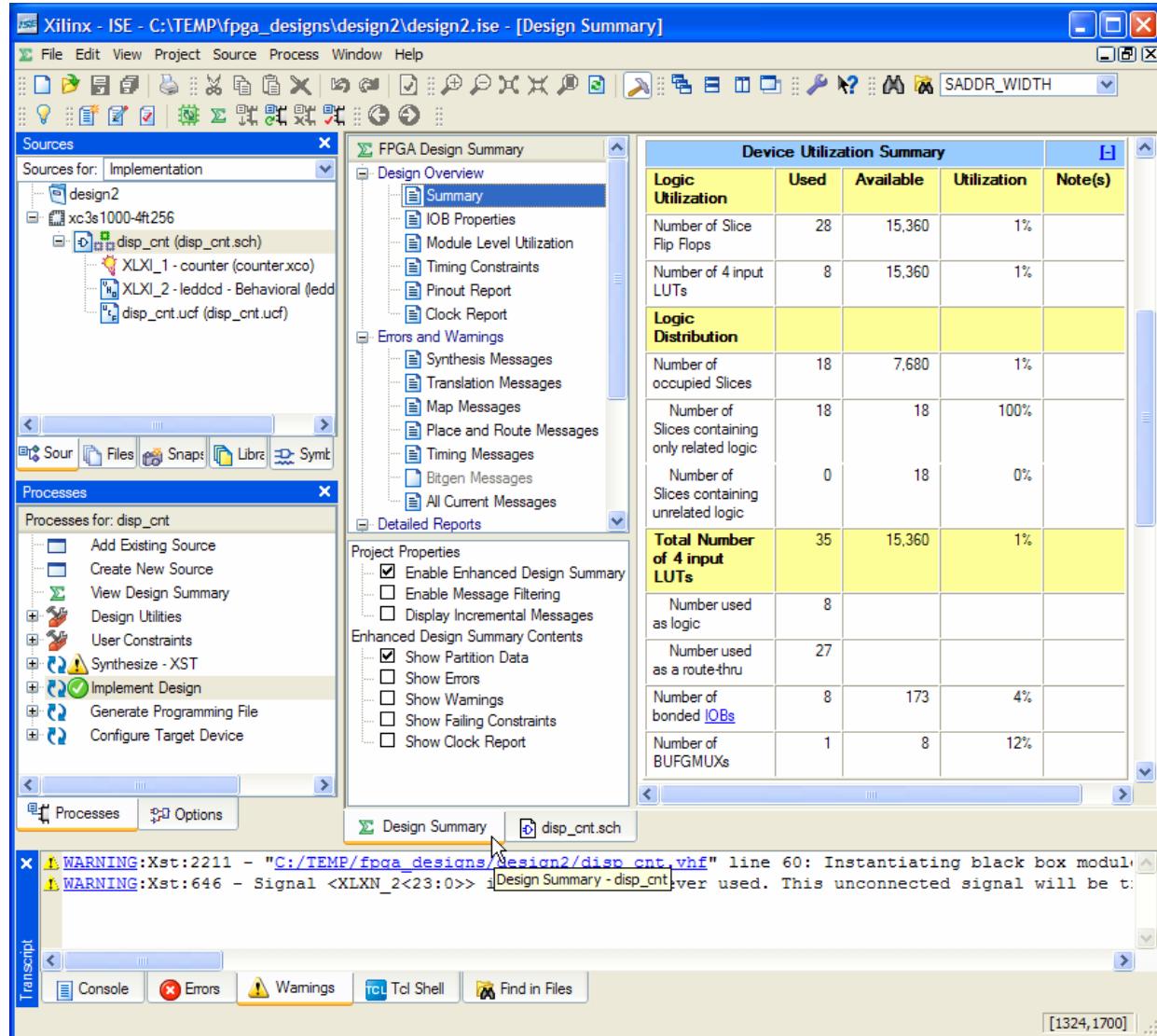
Synthesizing and Implementing the Design

Now you can synthesize the logic circuit netlist and translate, map and place & route it into the FPGA by highlighting the top-level disp_cnt module in the Sources pane and double-clicking the Implement Design process. The software will automatically invoke the synthesizer and then pass the synthesized netlist to the implementation tools. There should be no problems synthesizing and implementing the VHDL, Coregen and schematic files in the design. (There will be a few warnings that you can view in the Warnings tab of the Transcript pane. These warnings are mostly due to the unconnected wires in the 28-bit bus, but this is not an error.)



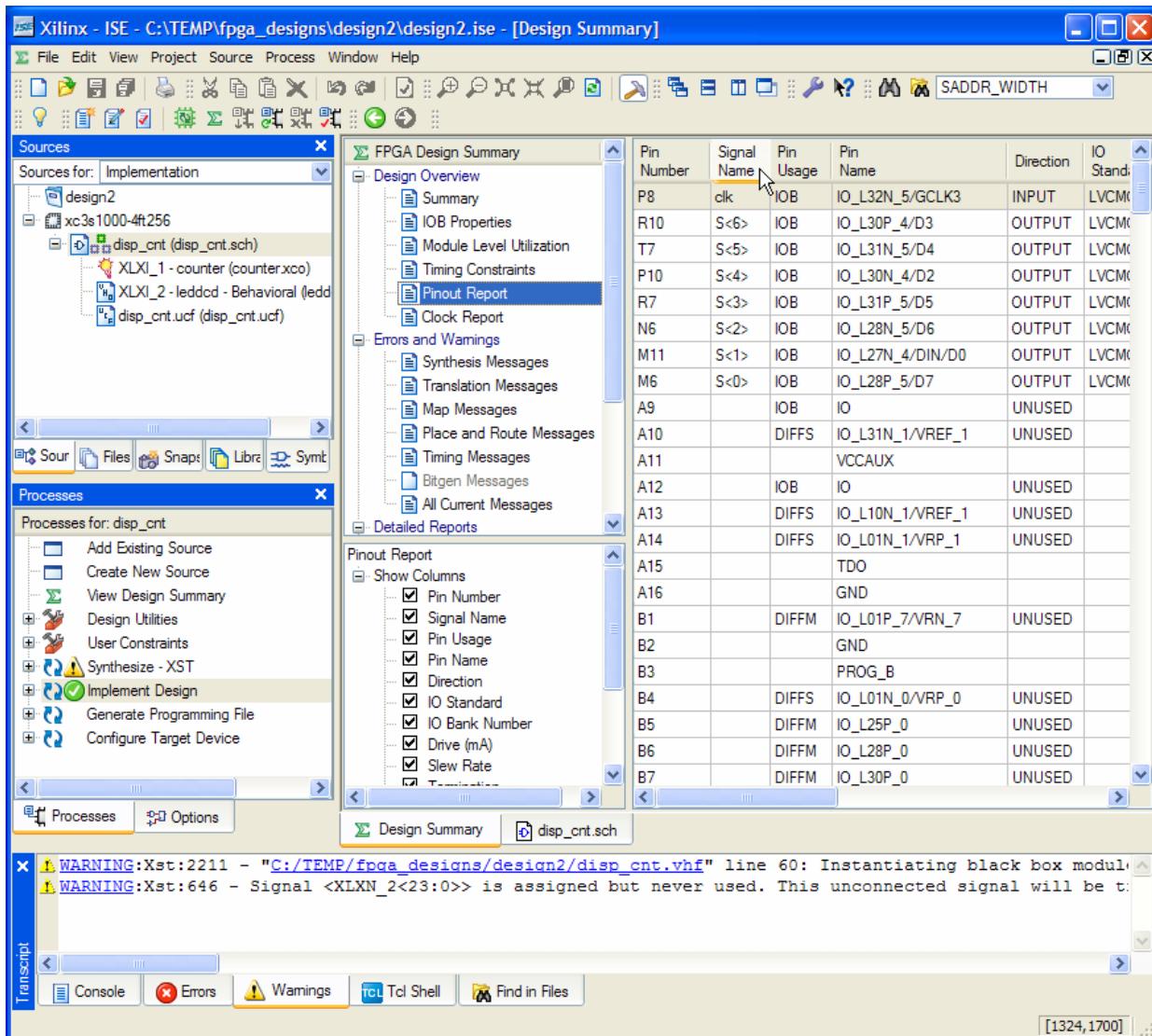
Checking the Implementation

After the implementation process is done, you can check the logic utilization by clicking on the Design Summary tab (or by double-clicking on the Place & Route Report).



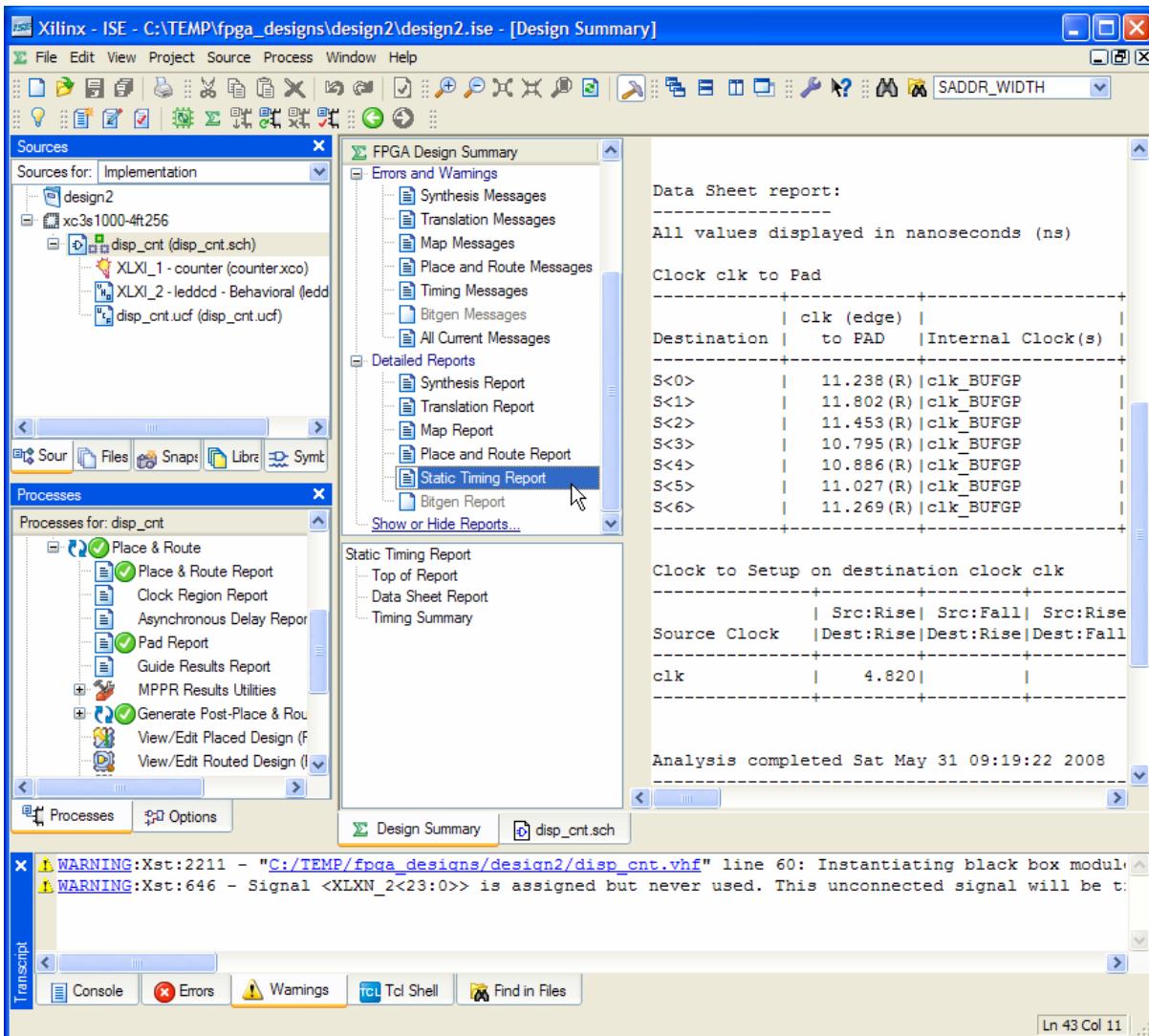
The displayable counter consumes 18 of the 7680 slices in the FPGA. Each slice contains two CLBs, so the displayable counter uses a maximum of 36 CLBs. The 28-bit counter requires at least 28 CLBs and the LED decoder requires 7 CLBs so this totals to 35 CLBs. So the device utilization statistics make sense.

As a precaution, you should also click on the Pinout Report in the Design Summary pane and check that the pin assignments for the clock input and LED decoder outputs match the assignments made with PACE. (You can make the comparison easier by clicking on the Signal Name column header to bring all the signal-pin assignments to the top.)



Checking the Timing

You have the displayable counter synthesized and implemented in the FPGA with the correct pin assignments. But how fast can the counter run? To find out, double-click on the Generate Post-Place & Route Static Timing process. This will determine the maximum delays between logic elements in the design taking into account logic and wiring delays for the routed circuit. (If a is already visible, then the static timing analysis has already been done.)

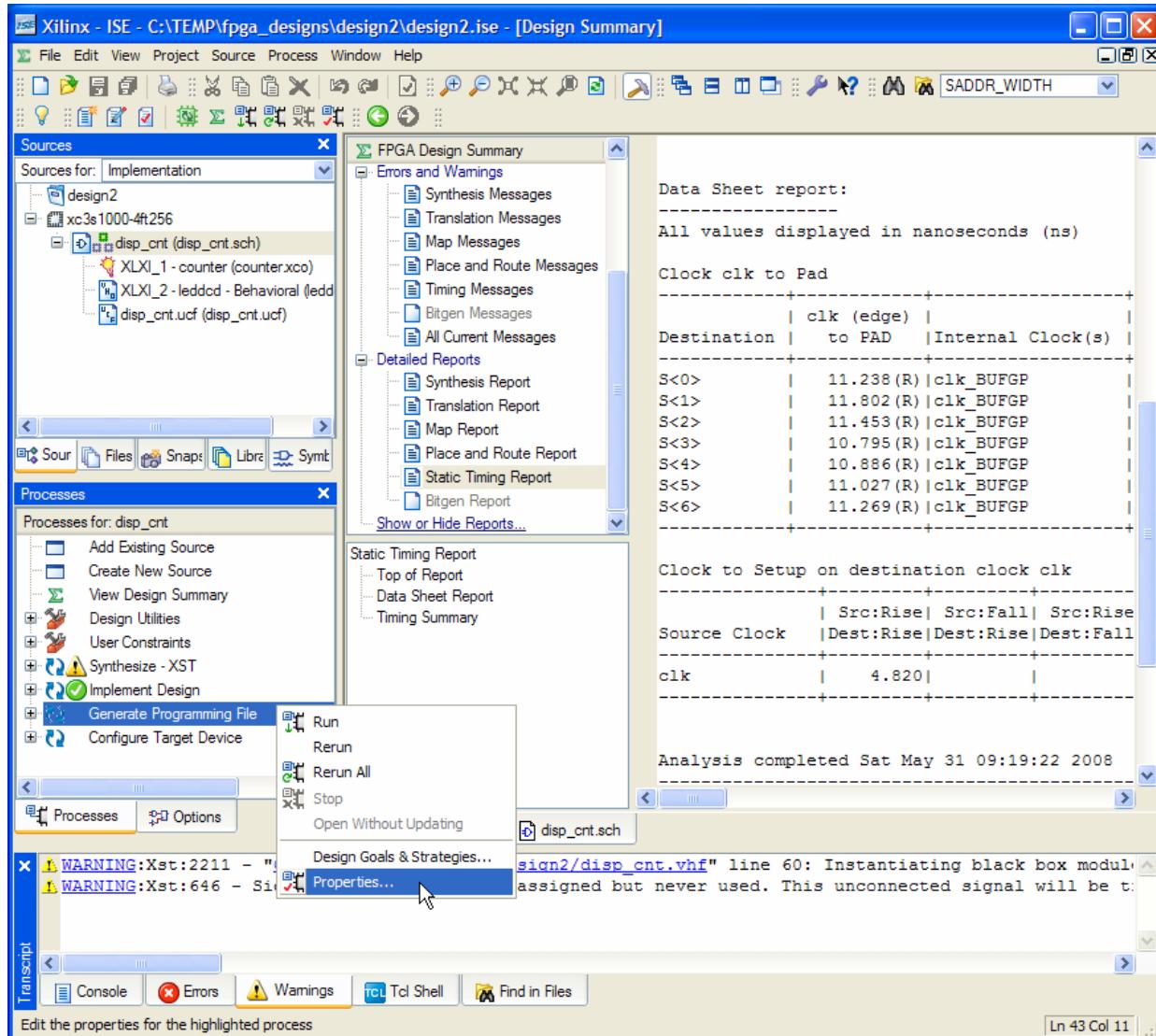


After the static timing delays are calculated, click on the Static Timing Report item in the Design Summary pane to view the results of the analysis. From the information shown at the bottom of the timing report, the minimum clock period for this design is 4.820 ns which means the maximum clock frequency is 207.5 MHz. The clock frequency on the XSA Board is 50 MHz which is well below the maximum allowable frequency for this design.

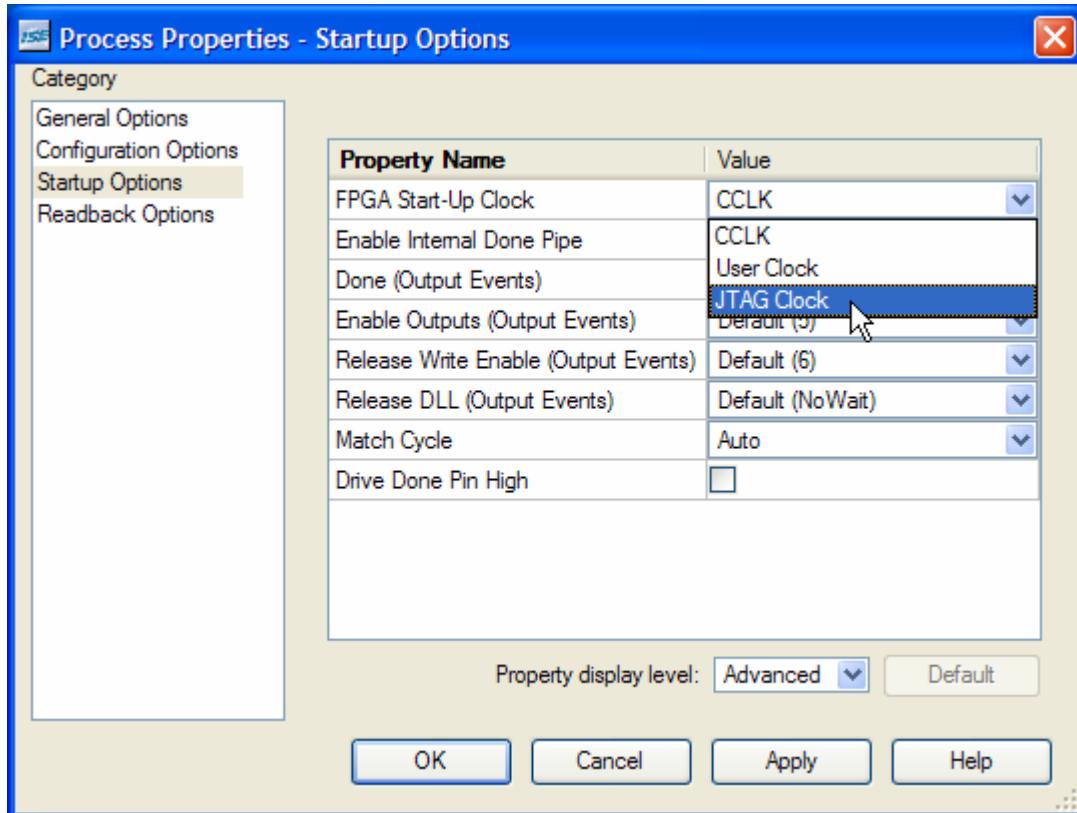
Generating the Bitstream

Now that you have synthesized our design and mapped it to the FPGA with the correct pin assignments, you are ready to generate the bitstream that is used to program the actual chip.

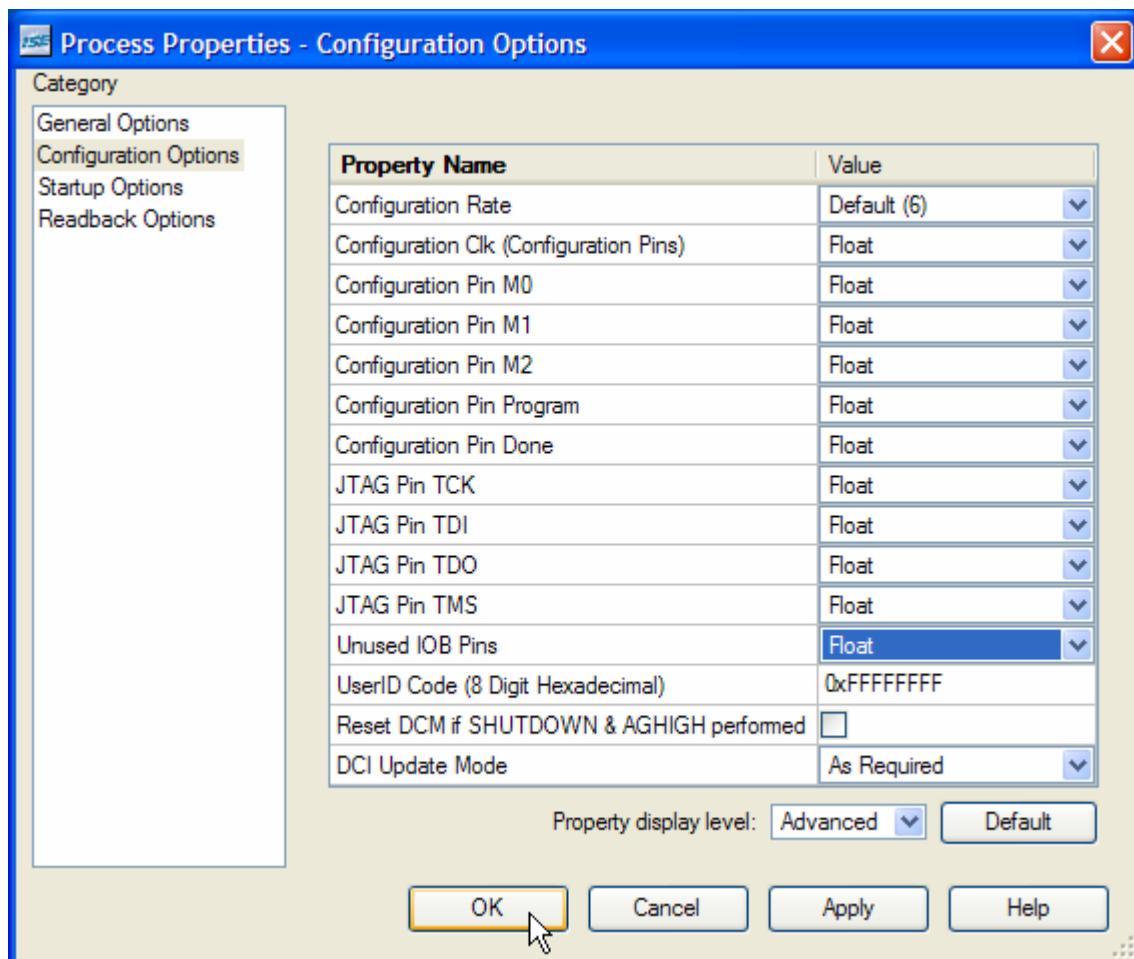
In this example, rather than use the gxlload utility you will employ the downloading utilities built into ISE. The iMPACT programming tool downloads the bitstream through the JTAG interface of the FPGA, so you need to adjust the way the bitstream is generated to account for this. Right click on the Generate Programming File process and select the Properties... entry from the pop-up menu.



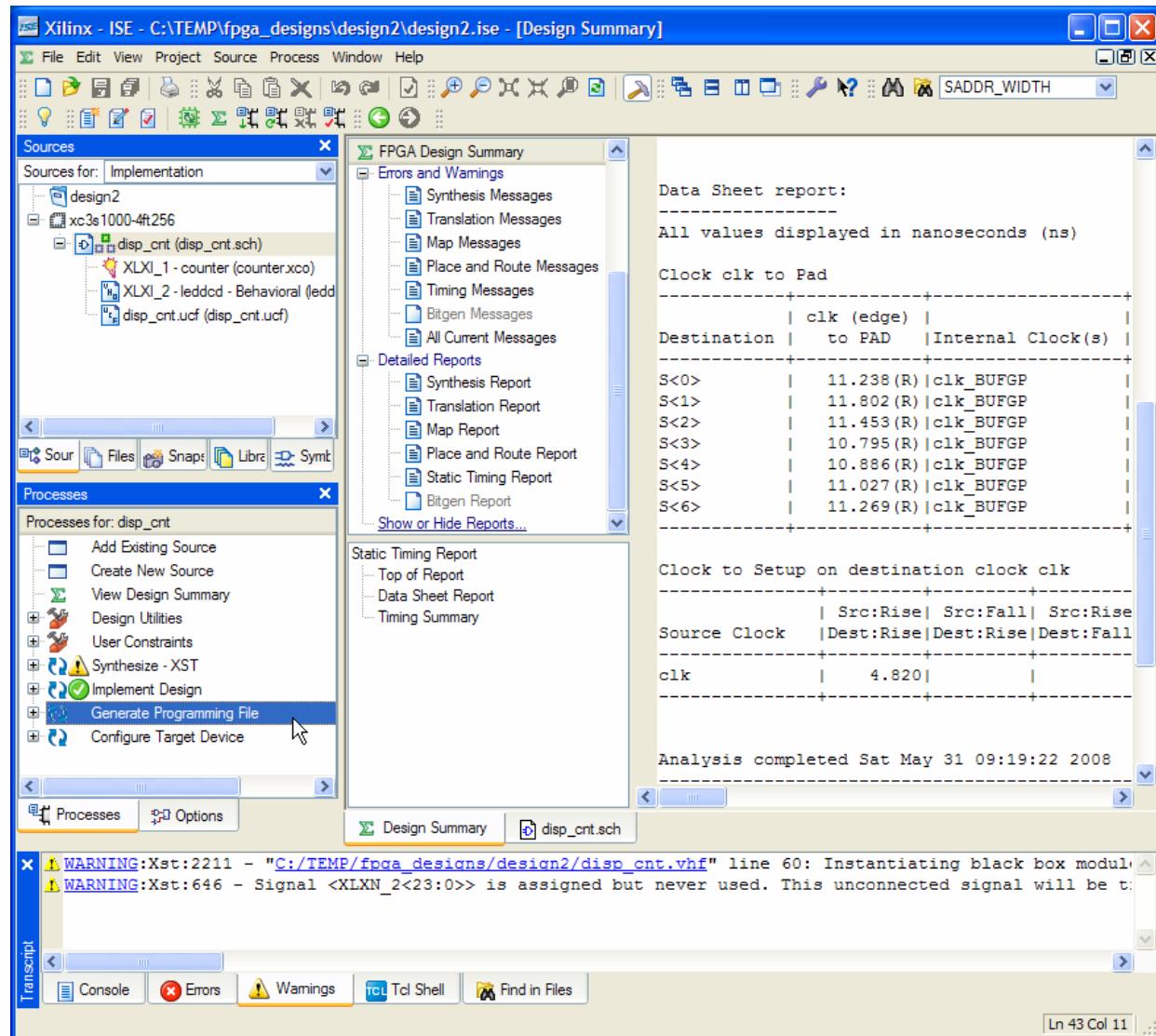
Select the Startup Options tab of the **Process Properties** window. Change the FPGA Start-Up Clock property to JTAG Clock so the FPGA will react to the clock pulses put out by the iMPACT tool during the final phase of the downloading process. If this option is not selected, the FPGA will not finish its configuration process and it will fail to operate after the downloading completes. Note that the startup clock is only used to complete the configuration process; it has no affect on the clock that is used to drive the actual circuit after the FPGA is configured.



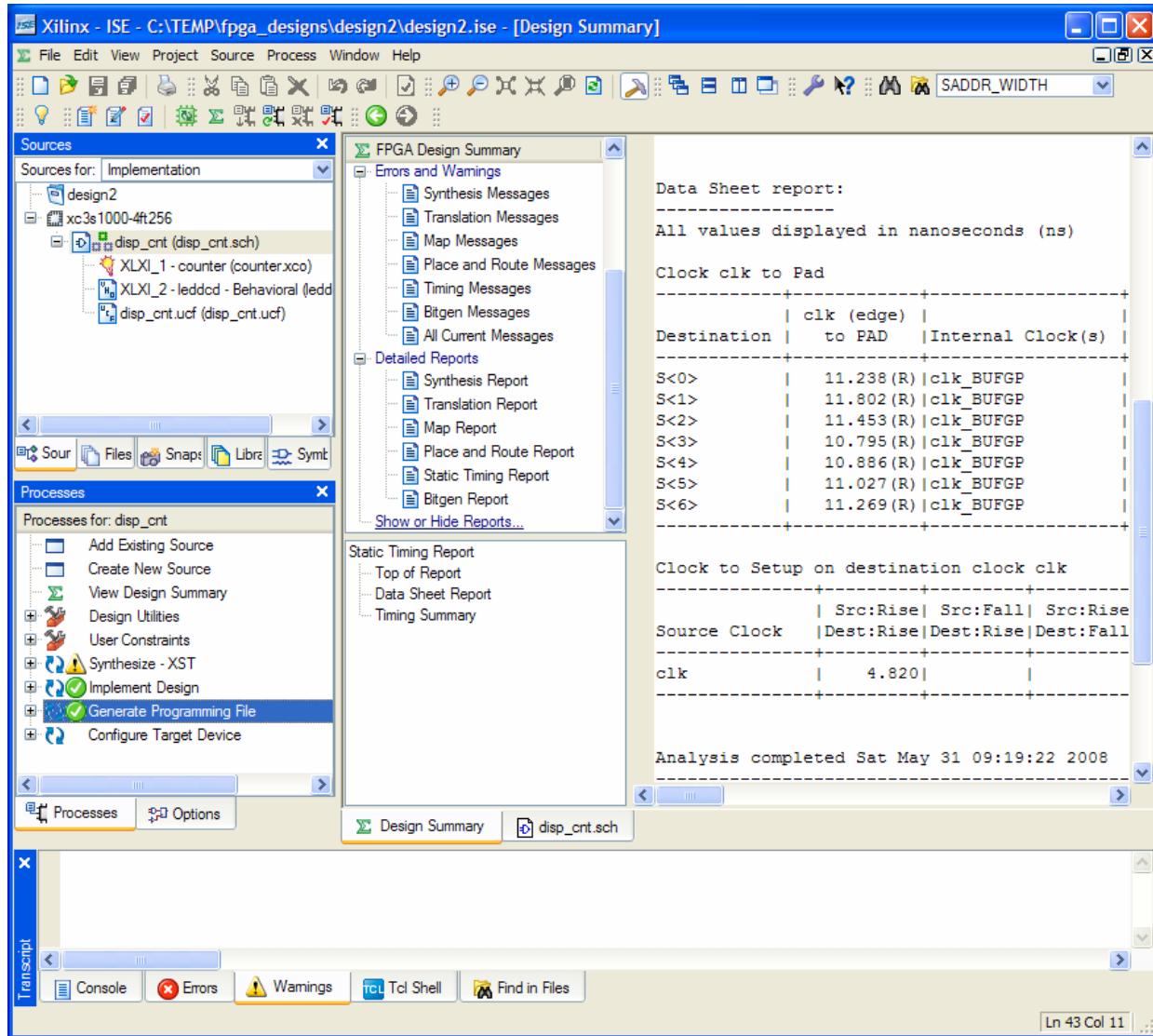
Next, click on the Configuration Options tab and disable all the internal pull-up and pull-down resistors in the FPGA as we did in the previous design. Then click OK.



Now that the bitstream generation options are set, highlight the disp_cnt object in the Sources pane and double-click on the Generate Programming File process to create the bitstream file.



Within a few seconds, a will appear next to the Generate Programming File process and a file detailing the bitstream generation process will be created. A bitstream file named disp_cnt.bit is placed in the design2 folder.

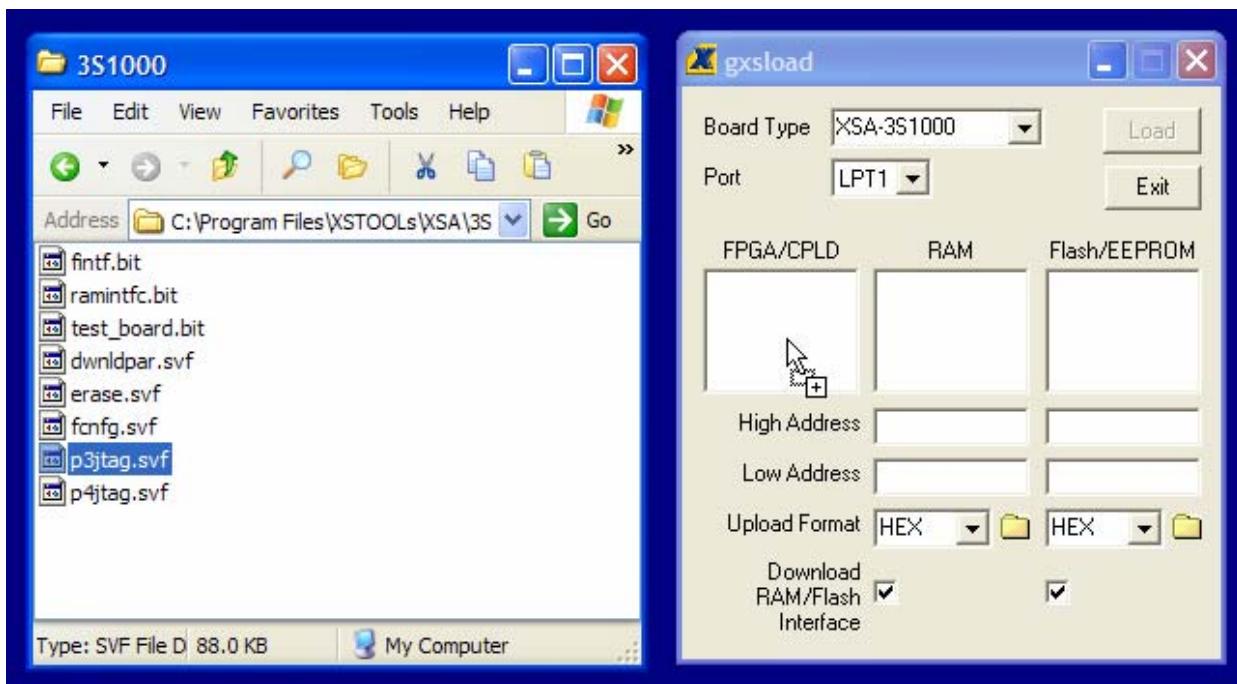


Downloading the Bitstream

Before downloading the disp_cnt.bit file, you must configure the interface CPLD on the

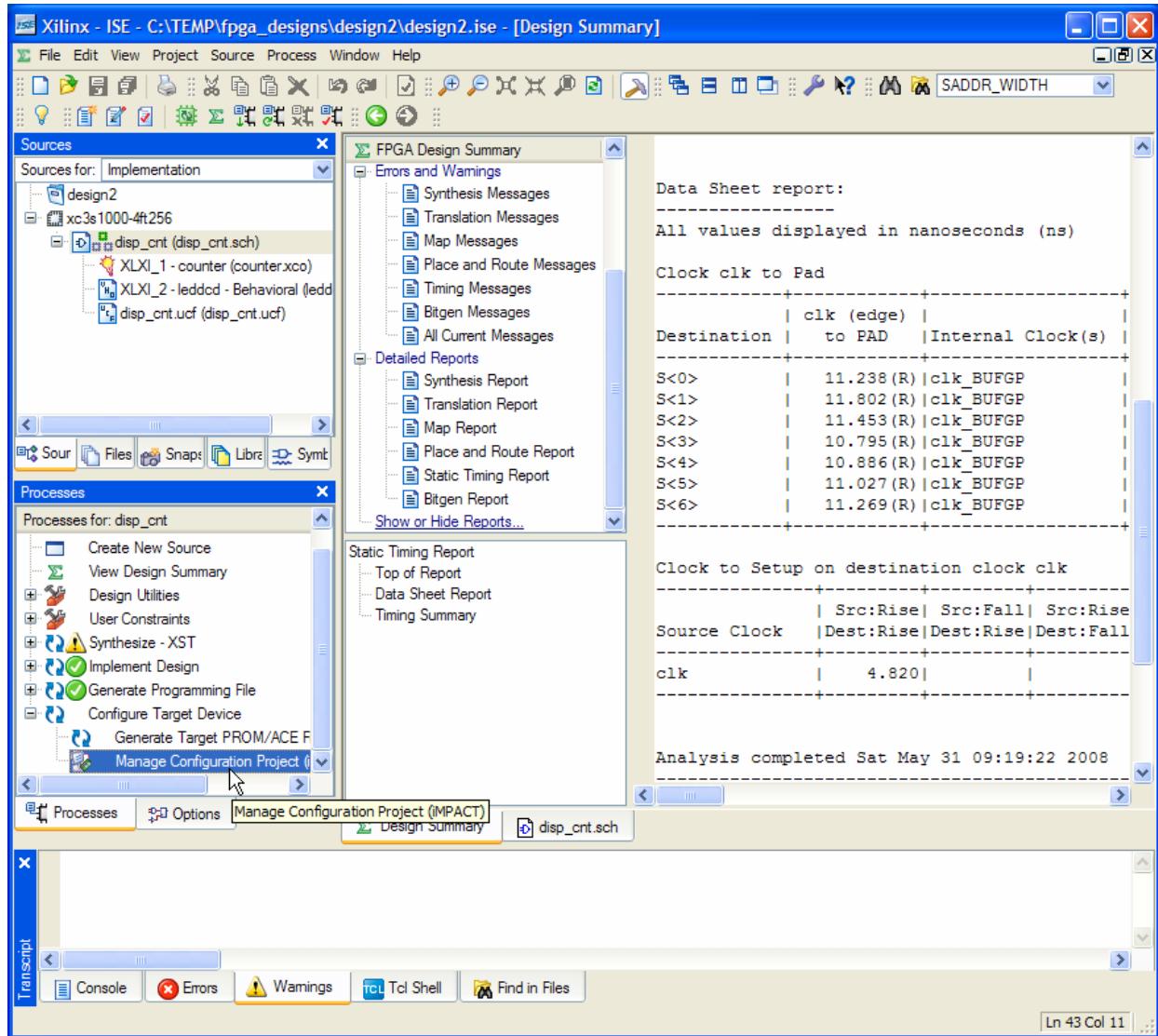


XSA-3S1000 board so it will work with the iMPACT programming tool. Double click the [GXSLOAD](#) icon and then drag & drop the p3jtag.svf file from the C:\Program Files\XSTOOLS\XSA\3S1000 folder into the FPGA/CPLD pane of the **gxsload** window. Then click on the Load button and the CPLD will be reprogrammed in less than a minute.

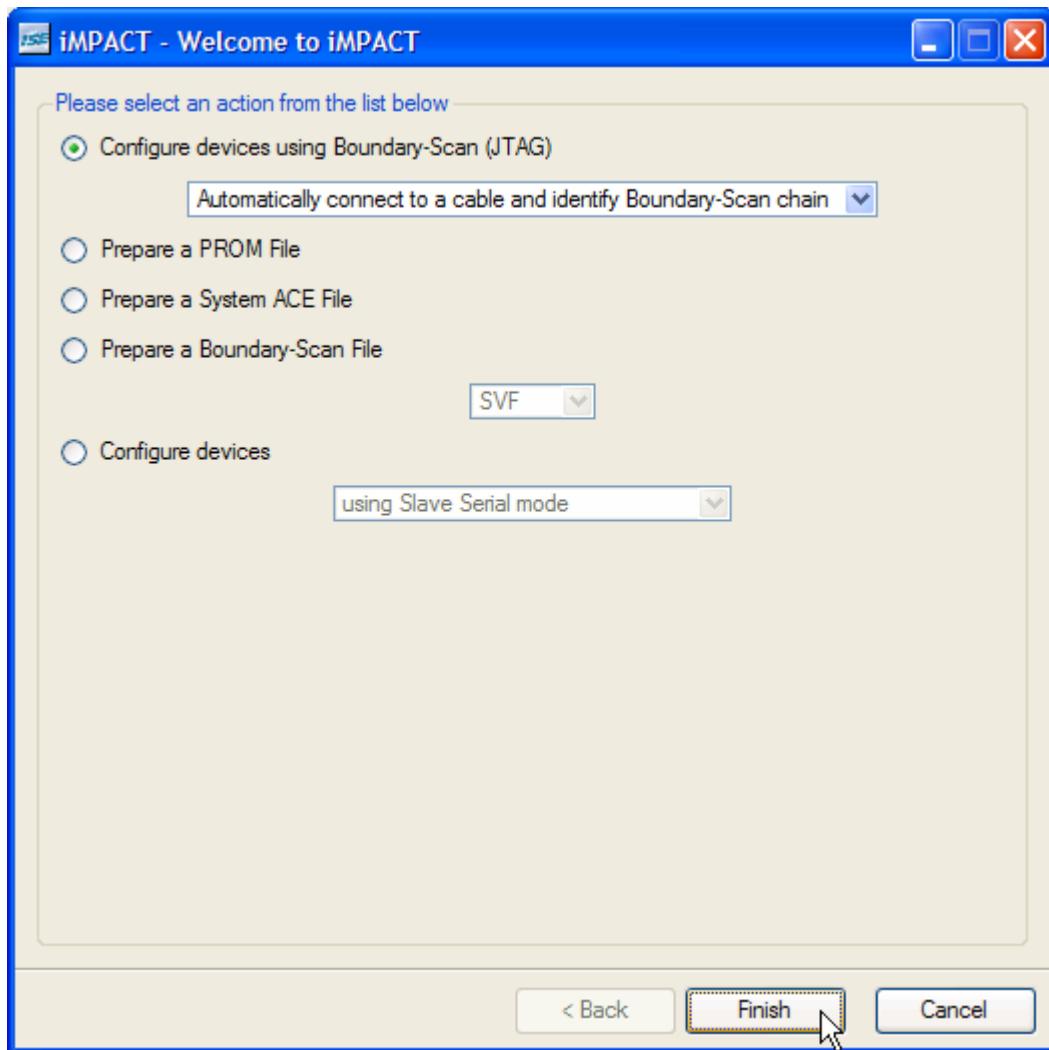


After the p3jtag.svf file is loaded into the XSA Board, move the shunt on jumper J9 from the **xs** to the **xi** position. The XSA Board is now setup so the FPGA can be configured through its JTAG boundary-scan pins with the iMPACT programming tool. Note that this process only needs to be done once because the CPLD on the XSA Board will retain its configuration even when power is removed from the board. (If you want to go back to using the gxsload programming utility, you must move the shunt on J9 back to the **xs** position and download the dwnldpar.svf file into the CPLD.)

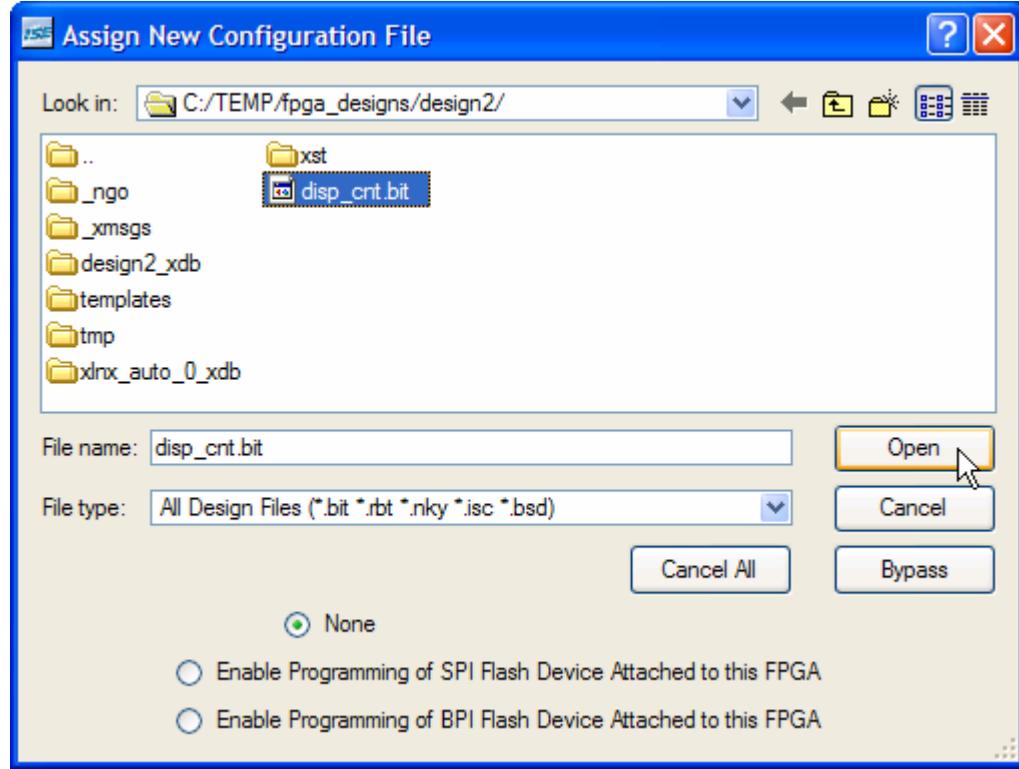
Now double-click on the Manage Configuration Project (iMPACT) process.



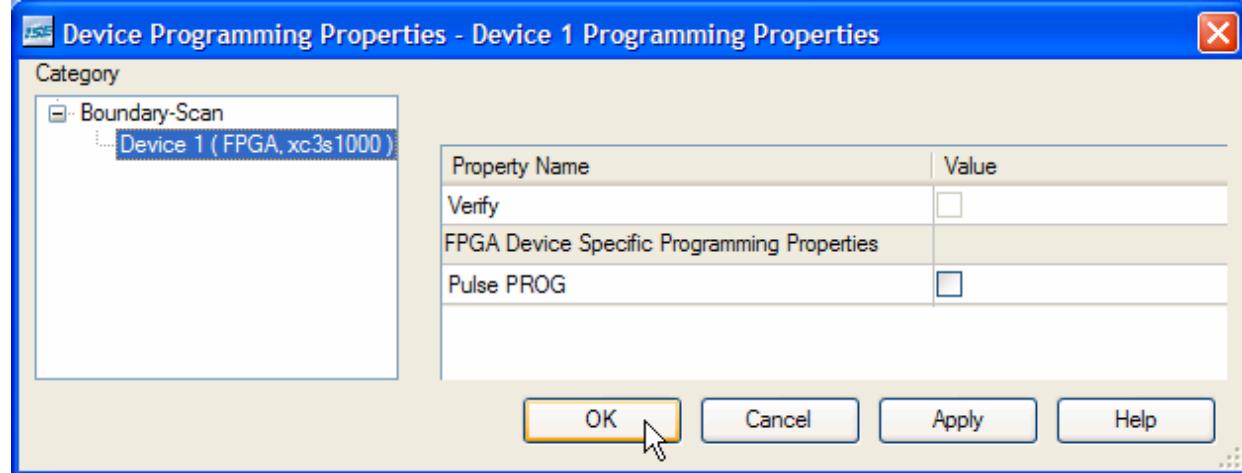
The **iMPACT – Welcome to iMPACT** window now appears. Make sure the Configure Devices using Boundary-Scan (JTAG) radio button is selected. Boundary-scan mode allows the configuration of multiple FPGAs connected together in a chain. To accomplish this, the iMPACT software needs to know the types of the FPGAs in the chain. There is just a single FPGA on the XSA Board and you could easily describe this to iMPACT. But iMPACT can also probe the boundary-scan chain and automatically identify the types of the FPGAs. This is even easier, so select the Automatically connect to a cable and identify Boundary-Scan chain option and click on the Next button.



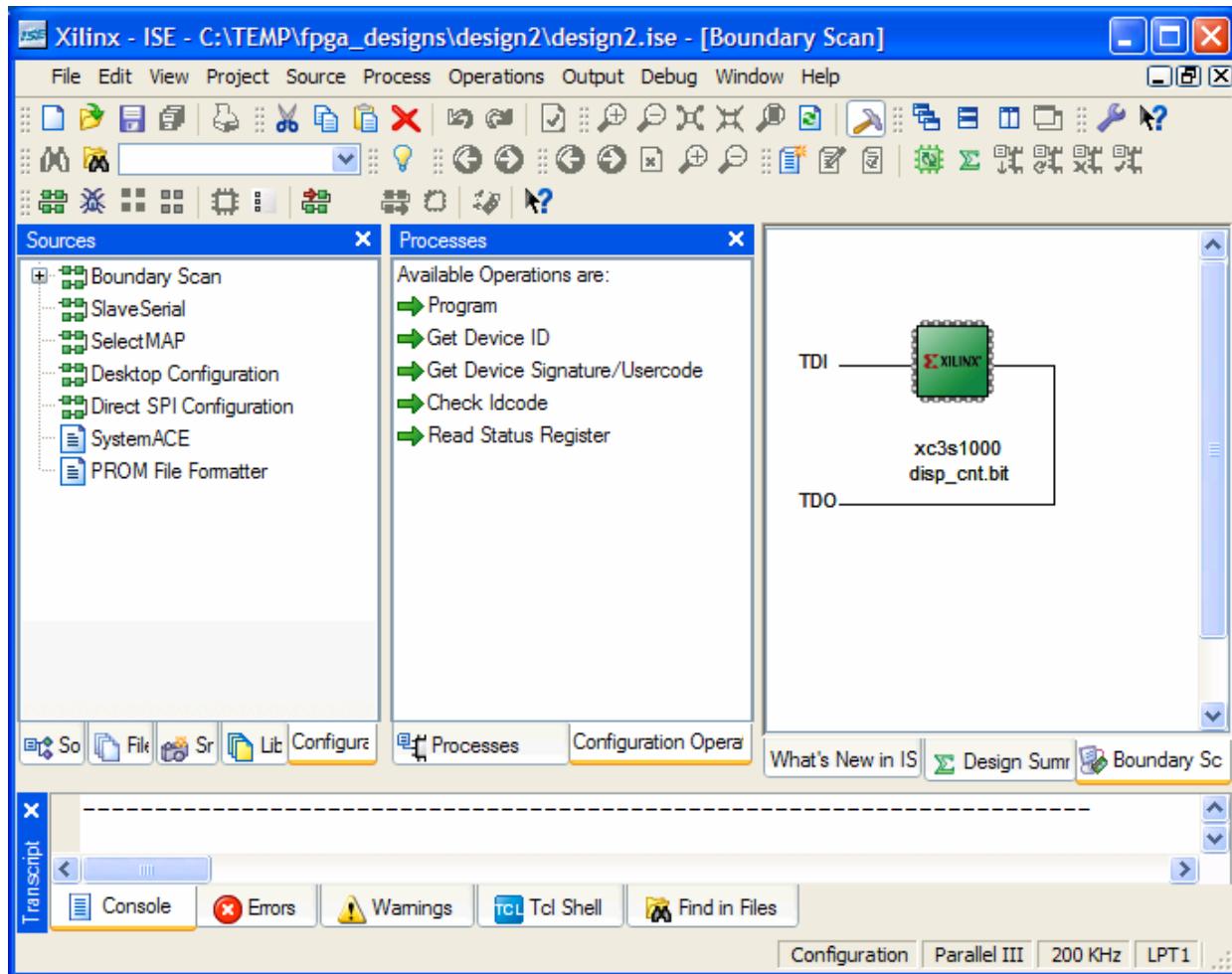
The **Assign New Configuration File** window now appears. You need to tell iMPACT what bitstream file to download into the FPGA. Go to the folder where your **design2** project is stored and highlight the disp_cnt.bit file. Then click on the Open button.



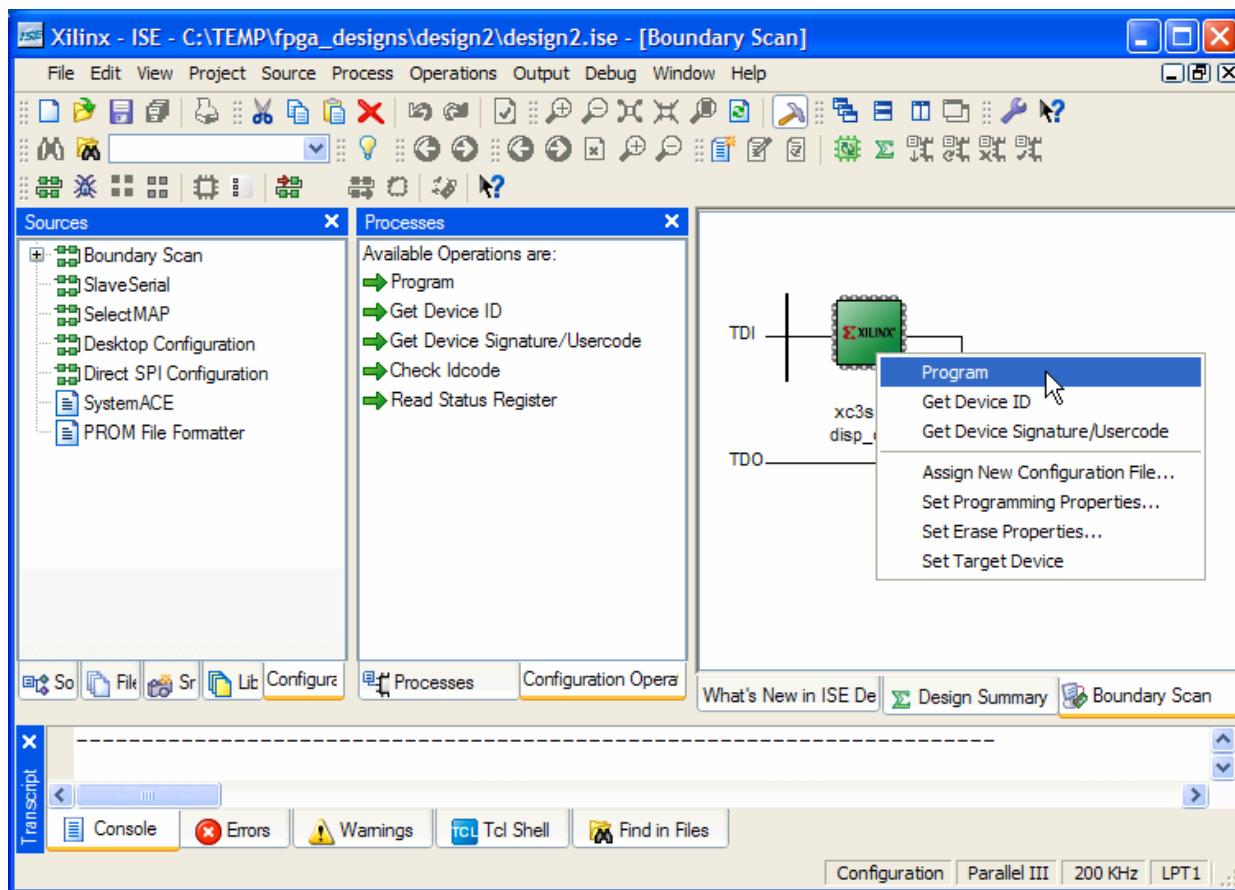
The **Device Programming Properties** appears next. You can check the Verify box if you want to make iMPACT readback the bitstream from the FPGA after the download to make sure it was sent correctly. Then click on the OK button.



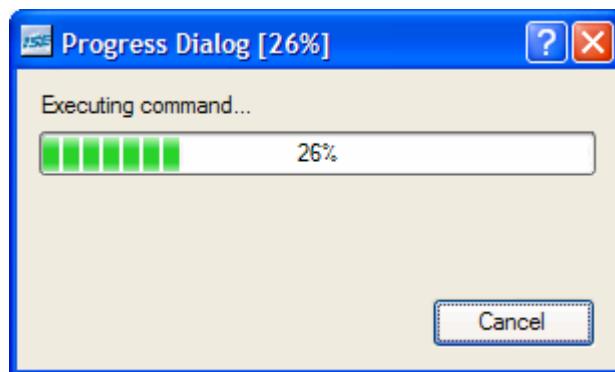
The iMPACT software will probe the boundary-scan chain and the main iMPACT window will appear showing a boundary-scan chain consisting of a single XC3S1000 FPGA.



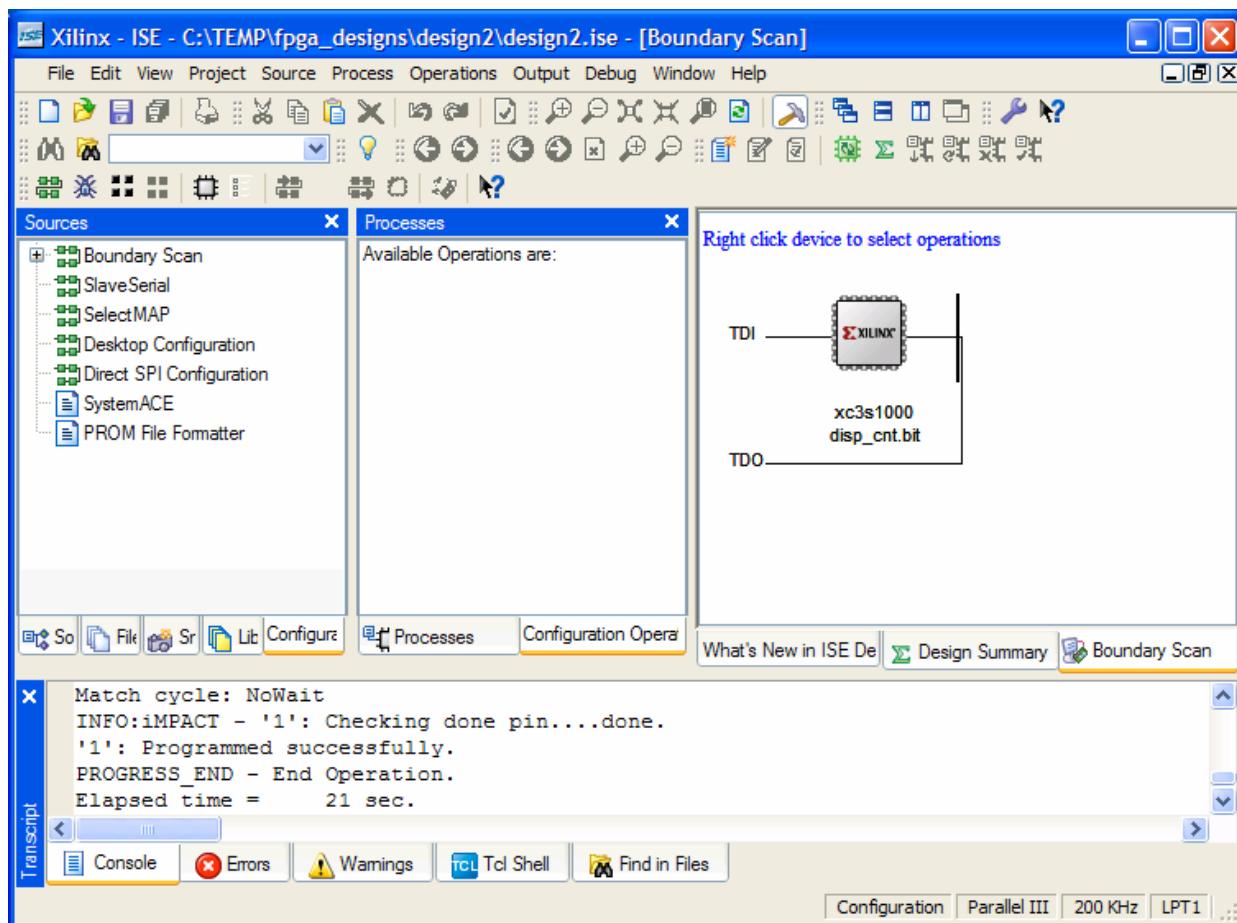
Now right-click on the xc3s1000 icon and select the Program... item on the pop-up menu. This will initiate the download of the bitstream to the FPGA.



The progress of the bitstream download will be displayed. The download operation should take less than a minute.



After the download operation completes, you can check the status messages in the bottom pane of the iMPACT window to see if the FPGA was configured successfully.



Testing the Circuit

Once the FPGA on the XSA Board is programmed, the circuit will begin operating without any further action from you. The LED display should repeatedly count through the sequence

0, 1, 2, 3, 4, 5, 6, 7, 8, 9, A, b, c, d, E, F with a complete cycle taking 5.4 seconds.

5

State Machine Design

Finite State Machines

A simple finite state machine (FSM) uses one or more flip-flops to store its internal state. The pattern of ones and zeroes on the flip-flop outputs are the current state. In a synchronous FSM, the current state is replaced with the next state on the rising edge of a clock signal. The next state is computed by a combinational logic circuit that accepts the current state and possibly some external signals as inputs. So a synchronous FSM is composed basically of a set of flip-flops fed by a combinatorial circuit that accepts feedback from the flip-flops on every clock cycle.

In this chapter we will build an FSM that acts like a combination lock. The requirements for this digital combination lock are:

1. The user enters a combination as a sequence of n key presses on a keyboard.
2. The combination lock stores a particular combination as a sequence of n key presses.
3. The combination lock will open if the user enters an n -key sequence that matches the combination. Otherwise, the lock stays locked.
4. The user must enter an entire n -key sequence before the lock either accepts or rejects the sequence.
5. Once the combination lock is unlocked, the user can relock it or enter a new combination as a sequence of key presses.
6. The lock will require the user to verify any new combination that is entered before it replaces the previous combination.

A hierarchical view of the combination lock and its lower-level modules is shown below. The combination lock consists of:

Keyboard interface: This module accepts a serial data stream and clock signal from a standard PS/2 PC keyboard and converts it into a parallel scancode with an associated ready signal that indicates the presence of the scancode.

Lock&key mechanism: This module accepts scancodes from the keyboard interface and determines whether or not the correct combination has been entered and manages the entry of new combinations.

The combination lock accepts the keyboard serial data and clock as inputs along with a main clock that synchronizes the operations of both modules. There is also a reset input to initialize

the entire FSM upon startup. The combination lock visually indicates its current status on a seven-segment LED.

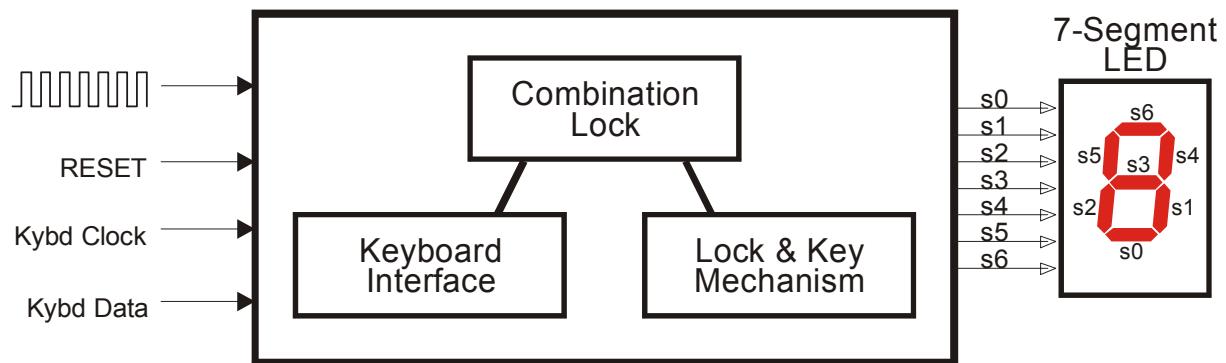
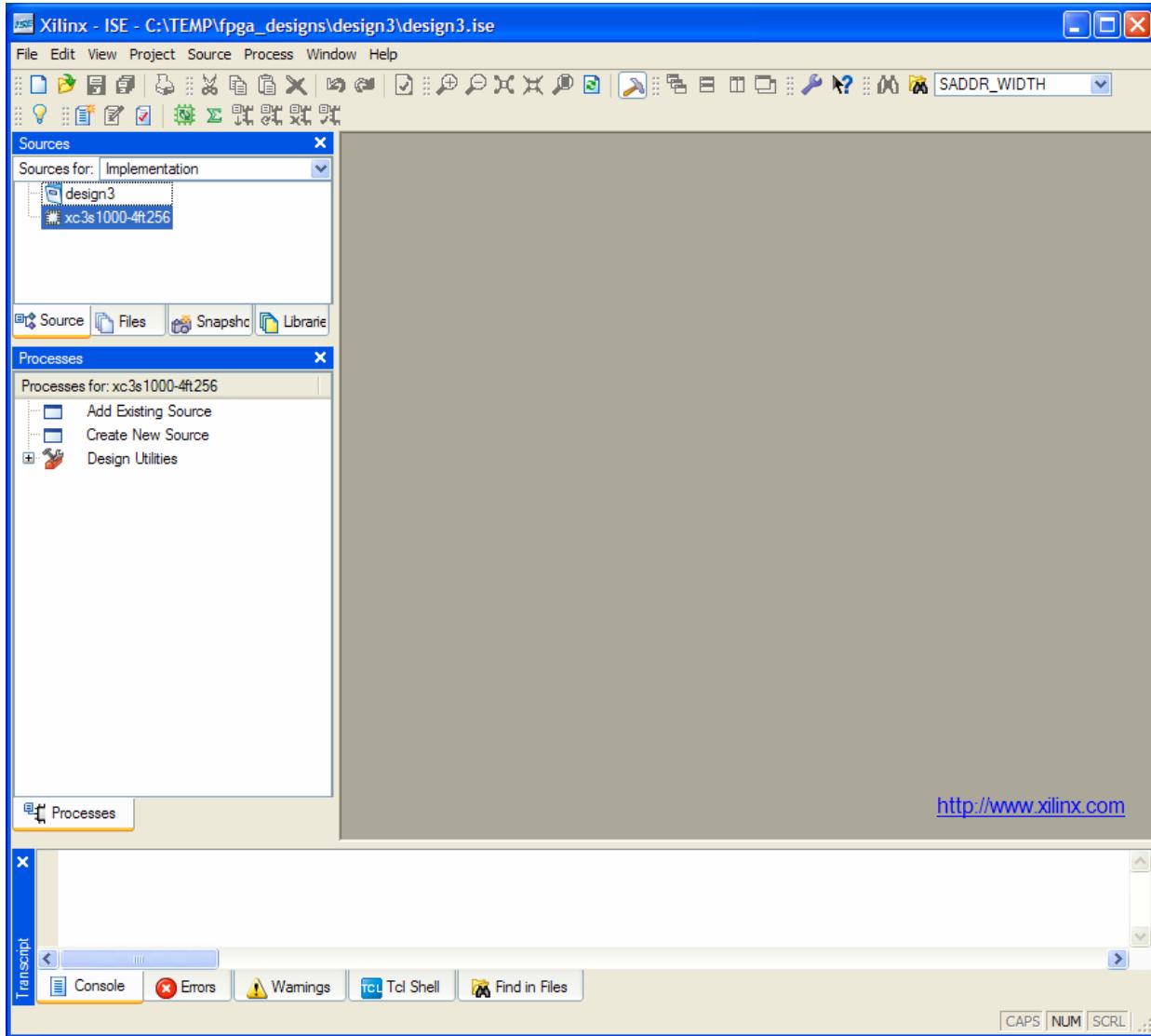


Figure 1: Design hierarchy for a combination lock.

Starting the Combination Lock Project

We will begin the design of the combination lock by creating an HDL-based project for the Spartan3 FPGA as we did before. We will describe the lower-level keyboard interface and the lock&key modules using the HDL Editor, and then tie these modules together with a top-level schematic.



Creating the Keyboard Interface Module

A PS/2 keyboard connects to an XSA Board through two signals:

psData: This signal carries the serial data stream as each key is pressed and released. Each key is assigned an eight-bit scancode that is transmitted least-significant bit to most-significant bit with a preceding start bit and a terminating parity bit and stop bit.

psClk: The falling edge of this signal indicates when the psData signal is valid.

The keyboard interface will accept the serial data stream and will output the eight-bit scancode in parallel along with an **rdy** pulse that indicates a valid scancode is available. The **rdy** pulse will be generated when the **psClk** signal goes high and stays there. The timing of the **psData**, **psClk**, and **rdy** signals is shown in Figure 2.

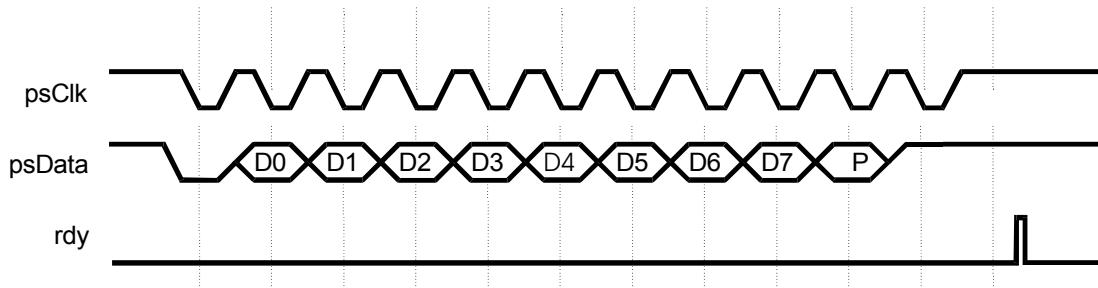
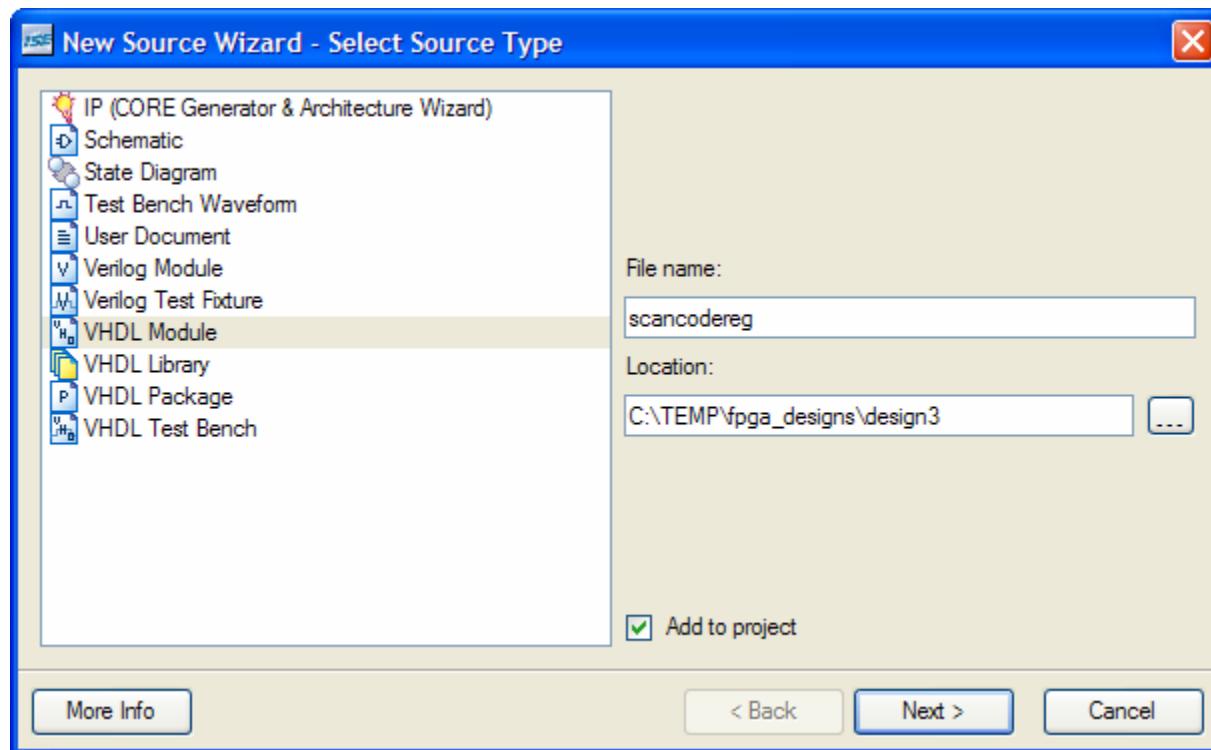


Figure 2: PS/2 keyboard waveforms.

A single scancode is transmitted when a key is pressed. But two scancodes are transmitted when the key is released: an initial scancode of 11110000 to indicate the key release, and then the scancode for the key is sent again. The keyboard interface will be designed such that the **rdy** signal pulses only after the key has been released.

To begin the keyboard interface, add a VHDL module as shown below.



The VHDL code for the keyboard interface is shown in Listing 1. The functions of the code for the **scancodeReg** module are as follows:

Lines 36–41: The module receives the **psData** and **psClk** inputs from the keyboard and outputs the eight-bit **scancode** and the **rdy** signals that were described above. A master clock

is also provided which synchronizes the operations of this module with the lock&key module. A reset signal initializes the module when it first powers up.

Line 46: This line declares a 10-bit shift-register that holds the scancode value as it arrives from the keyboard. The start bit, eight scancode bits, parity bit, and stop bit will enter the most-significant bit of the **sc_r** register and shift towards the least-significant bit. By the end of a scancode transmission the start bit will have shifted completely out of the register and be lost while the scancode will end up in the lower eight bits of **sc_r**. The stop and parity bits will be in the uppermost two bits.

Lines 47–48: These lines define a counter that is used to determine when the **psClk** signal is no longer pulsing. The timeout value (line 47) is determined by dividing the main clock frequency (defined in the GENERIC section on line 32) by the frequency of the **psClk** (line 33). If the main clock is 50 MHz and the keyboard clock is 10 KHz, then the timeout value is 5000 which means it will take 5000 pulses of the main clock to determine if the **psClk** signal is static. The timeout counter register is defined on line 48 as a natural that can hold a value as large as the timeout value. By defining the counter register in this way, we can change the frequency of the main clock or the keyboard clock and the timeout counter will be automatically resized by the synthesizer with exactly the number of bits needed to store the timeout value.

Lines 56–66: This process parallelizes the serial keyboard data. If the reset input is active, the scancode shift register is cleared to all zeroes. Otherwise, on falling edges of the keyboard clock the value on the keyboard data signal is placed into the most-significant bit of the shift register and the upper nine bits of the register are shifted one bit position downward. By the end of a scancode transmission, the start bit will have shifted completely out of the register and be lost while the scancode will end up in the lower eight bits of **sc_r**. The stop and parity bits will be in the uppermost two bits.

Line 69: The eight lower bits of the **sc_r** register are output as the scancode output of the module.

Lines 73–94: This process detects when the **psClk** signal has stopped pulsing and indicates that a scancode is available. The timeout counter and scancode ready flag are cleared when the module is reset. Then the counter is incremented as long as the **psClk** is at logic 1 and the counter has not reached its timeout value yet. The counter is reset to zero if **psClk** is ever low because that indicates the keyboard clock is still pulsing so the scancode cannot be complete. But if the counter ever reaches the value **TIMEOUT-1**, then the scancode ready flag is pulsed high for a single clock cycle.

Lines 98–114: This process checks the **scRdy_r** flag and looks for the scancode that matches the **KEYRELEASE** scancode defined on line 52. After seeing the key release scancode, this process looks for the next following scancode. Then it sets the flag that indicates the scancode received after the **KEYRELEASE** code is ready for output.

Line 116: The ready flag from the previous process is output from the module.

Listing 1: VHDL code for the keyboard interface.

```

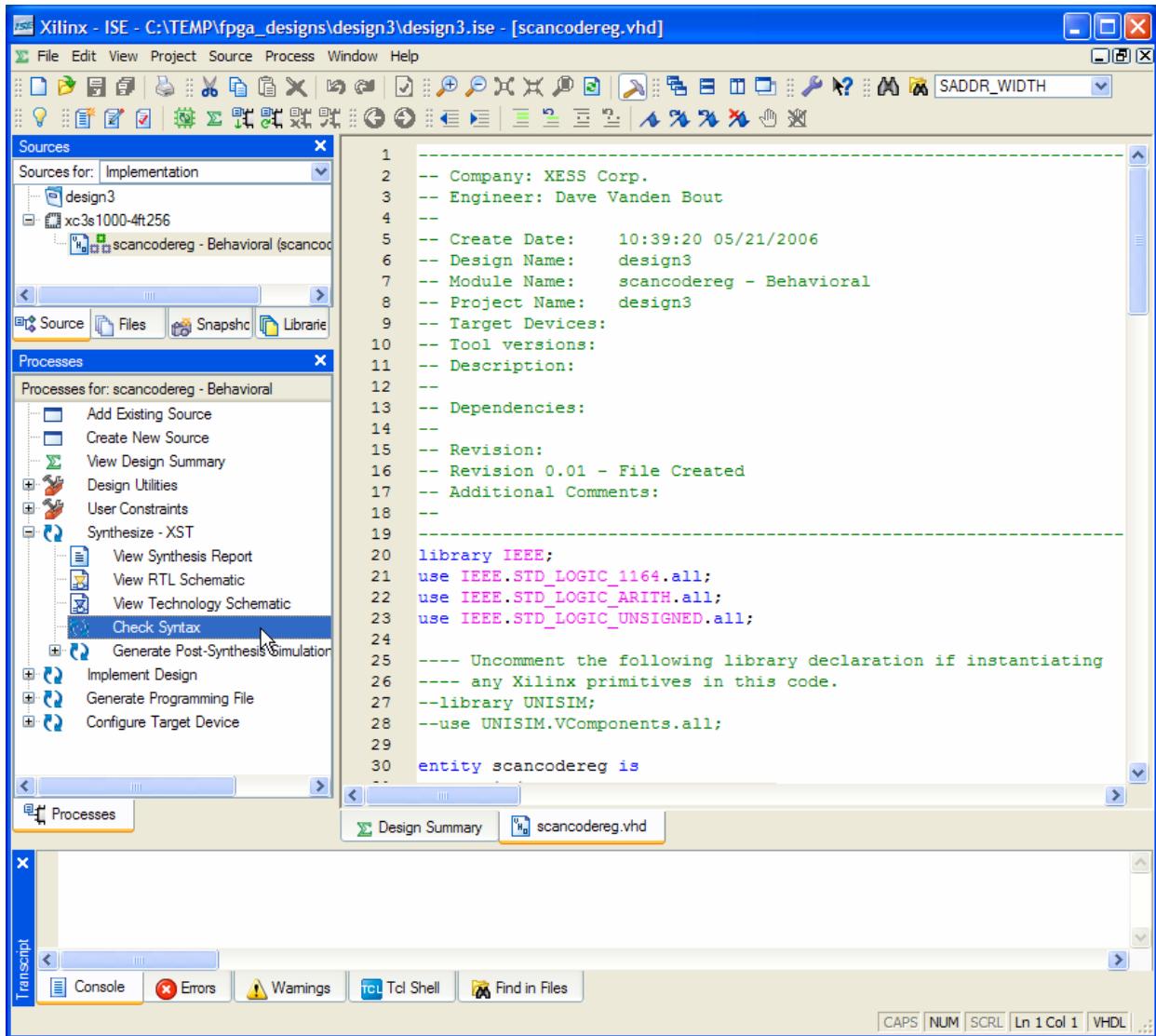
1 -----
2 -- Company: XESS Corp.
3 -- Engineer: Dave Vanden Bout
4 --
5 -- Create Date: 10:39:20 05/21/2006
6 -- Design Name: design3
7 -- Module Name: scancodereg - Behavioral
8 -- Project Name: design3
9 -- Target Devices:
10 -- Tool versions:
11 -- Description:
12 --
13 -- Dependencies:
14 --
15 -- Revision:
16 -- Revision 0.01 - File Created
17 -- Additional Comments:
18 --
19 -----
20 library IEEE;
21 use IEEE.STD_LOGIC_1164.all;
22 use IEEE.STD_LOGIC_ARITH.all;
23 use IEEE.STD_LOGIC_UNSIGNED.all;
24 --
25 ----- Uncomment the following library declaration if instantiating
26 ----- any Xilinx primitives in this code.
27 --library UNISIM;
28 --use UNISIM.VComponents.all;
29 --
30 entity scancodereg is
31     generic(
32         CLKFREQ : natural := 50_000; -- main clock freq (KHz)
33         PSCLKFREQ : natural := 10        -- keyboard clock freq (KHz)
34     );
35     port(
36         clk      : in std_logic;       -- main clock
37         rst      : in std_logic;       -- reset
38         psClk    : in std_logic;       -- keyboard clock
39         psData   : in std_logic;       -- keyboard data
40         scanCode : out std_logic_vector(7 downto 0); -- key scancode
41         rdy      : out std_logic;       -- true when scancode is ready
42     );
43 end scancodereg;
44 --
45 architecture Behavioral of scancodereg is
46     signal sc_r      : std_logic_vector(9 downto 0); -- scancode shift reg
47     constant TIMEOUT  : natural := CLKFREQ/PSCLKFREQ; -- psClk quiet timeout
48     signal cnt_r    : natural range 0 to TIMEOUT;      -- timeout counter
49     signal scRdy_r  : std_logic; -- scan code is ready flag
50     signal rdy_r    : std_logic; -- output scan code is ready flag
51     signal keyrel_r : std_logic; -- key has been released flag
52     constant KEYRELEASE : std_logic_vector(7 downto 0) := "11110000";
53 begin
54 
55     -- this process gathers the keybrd scancode into the shift register
56     process(psClk, rst)
57     begin
58         -- async. reset of scancode ready flag
59         if rst = '1' then
60             sc_r <= (others => '0');
61         -- accept keyboard data on falling edge of keyboard clock
62         elsif falling_edge(psClk) then
63             -- key data arrives LSB first so right-shift it into MSB of register

```

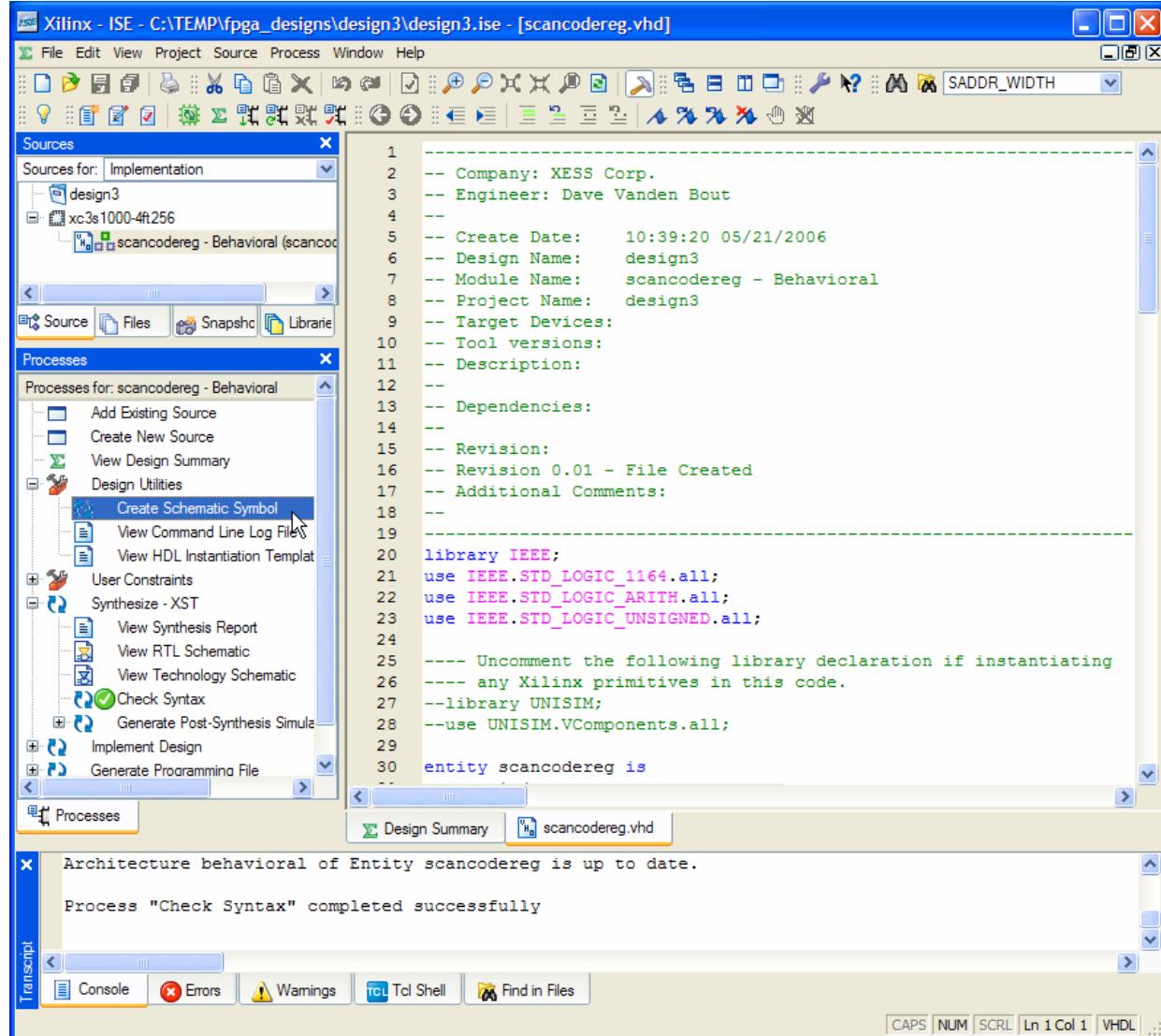
```

64      sc_r <= psData & sc_r(sc_r'high downto 1);
65    end if;
66  end process;
67
68  -- key scancode is in the lower 8-bits of the shift register
69  scancode <= sc_r(scancode'range);  -- output scancode
70
71  -- this process detects the end of the scancode by looking
72  -- for the absence of keyboard clock pulses
73  process(clk, rst)
74  begin
75    if rst = '1' then
76      cnt_r <= 0;  -- clear the timeout counter
77      scRdy_r <= '0';  -- clear the scancode ready flag
78    elsif rising_edge(clk) then
79      scRdy_r <= '0';  -- by default, no key scancode is ready for output
80      if psClk = '0' then
81        -- reset the timeout register whenever the keyboard clock pulses low
82        cnt_r <= 0;
83      elsif cnt_r /= TIMEOUT then
84        -- increment the timeout counter if the keyvoard clock is high
85        -- and the counter hasn't reached the timeout value yet
86        cnt_r <= cnt_r + 1;
87        if cnt_r = TIMEOUT-1 then
88          -- signal that a key scancode is ready when the counter is
89          -- equal to one less than the timeout value
90          scRdy_r <= '1';  -- rdy signal pulses for one clock period
91        end if;
92      end if;
93    end if;
94  end process;
95
96  -- this process detects when the keyboard key is released and
97  -- signals when the scancode for the released key is ready
98  process(clk)
99  begin
100    if rising_edge(clk) then
101      rdy_r <= '0';  -- by default, no key scancode is ready for output
102      if scRdy_r = '1' then
103        -- check the scancode register when a code is ready
104        if sc_r(7 downto 0) = KEYRELEASE then
105          -- set flag if the key release prefix is detected
106          keyrel_r <= '1';
107        elsif keyrel_r = '1' then
108          -- end up here on next scancode received after key release prefix
109          rdy_r <= '1';  -- released key scancode is in the scancode register
110          keyrel_r <= '0';  -- reset the key release flag
111        end if;
112      end if;
113    end if;
114  end process;
115
116  rdy <= rdy_r;  -- signal that a key scancode is ready
117
118 end Behavioral;
```

Once the VHDL code is entered and saved in the scancodereg.vhd file, you should check it for syntax errors.



Once any syntax errors are corrected, create a schematic symbol for the keyboard interface. You will use that later in the top-level schematic for this project.



Creating the Lock&Key Module

Now we can design the lock&key VHDL module. Add a new VHDL source module called **lock** and add the code shown in Listing 1.

The lock module follows the basic structure of almost all FSM's I create. There is a **combinatorial** process (lines 106-247) that computes the FSM's next state given the current state and inputs, and there is an **update** process (lines 251-277) that loads the computed values into the registers on the next clock edge.

I also follow a register naming convention that appends **_r** to a signal or variable name if it stores the current state, and **_x** is appended to the name if it stores the computed value that will be loaded into the register on the next clock edge. This naming convention also allows me to visually check my VHDL for errors because I know that only state registers end with **_x**

should be on the left-hand side of the assignment operator ($<=$) in the **combinatorial** process and registers with the **_r** register should only appear on the right-hand side because this process computes the next state values using the current state values as operands. The situation is reversed in the update process with the **_r** signals appearing on the left-hand side of the assignment and **_x** signals on the right.

The lock&key FSM has 5 states (declared on lines 47-51) with the operations in each state handled in a separate section of a `case` statement in the **combinatorial** process:

enterComb (lines 127-150): Upon entering this state, an *L* is displayed on the LED to indicate the lock is locked. The value on the LED is incremented as each keypress is received from the user (indicated by the **rdy** signal going high), and the scancode received from the **scancodereg** module is compared against the next scancode of the active lock combination stored in the **activeComb** array. If any key fails to match the corresponding entry in the combination, then the **match** flag is cleared. Once the last key in the combination is checked (as indicated by the **index** register that steps through the **activeComb** array), the FSM transitions to the **checkComb** state.

checkComb (lines 154-164): The **match** flag is checked in this state. If the flag is set, then the combination was entered correctly and the FSM moves to the **unlocked** state. Otherwise, it returns to the **enterComb** state and waits for the user to enter another key sequence.

unlocked (lines 167-180): The LED displays a *U* in this state to indicate the lock is unlocked. If the user presses a backspace key, the FSM transitions to the **enterNewComb** state so a new combination can be entered. (Thus, the user must know the current combination before he can enter a new combination.) Any other keypress relocks the lock and returns the FSM to the **enterComb** state.

enterNewComb (lines 185-204): The LED displays an *C* upon entering this state to indicate the combination is being replaced. Each keypress entered in this state is stored in the **newComb** array and the value on the LED is incremented. After all the keys for a complete combination are entered, the FSM goes to the **verifyNewComb** state.

verifyNewComb (lines 209-237): In this state, the user must re-type the new combination so it can be compared against the contents of the **newComb** array. If any mismatch occurs, the FSM immediately goes to the **enterComb** state and relocks the lock without changing the combination. If the new combination is repeated correctly, the contents of the **newComb** array are transferred to the **activeComb** array and the lock is relocked with the new combination.

The reset of the FSM and the actual state transitions is handled by the **update** process on each rising edge of the master clock:

Lines 255–265: The FSM is synchronously reset if the reset input (**rst**) is high or if the power-on reset flag (**rst_i**) is set. The power-on reset flag is initially set on line 67 so the FSM is guaranteed to be reset whenever power is first applied to the FPGA. The FSM always resets to the **enterComb** state and the lock is locked. If a power-on reset has occurred, then the active combination is not defined so it must be loaded with a default combination (defined as the key sequence “123456789” on lines 61-63). A reset caused by the reset input, however, will not reload the active combination so it will remain at whatever key sequence was set by the user.

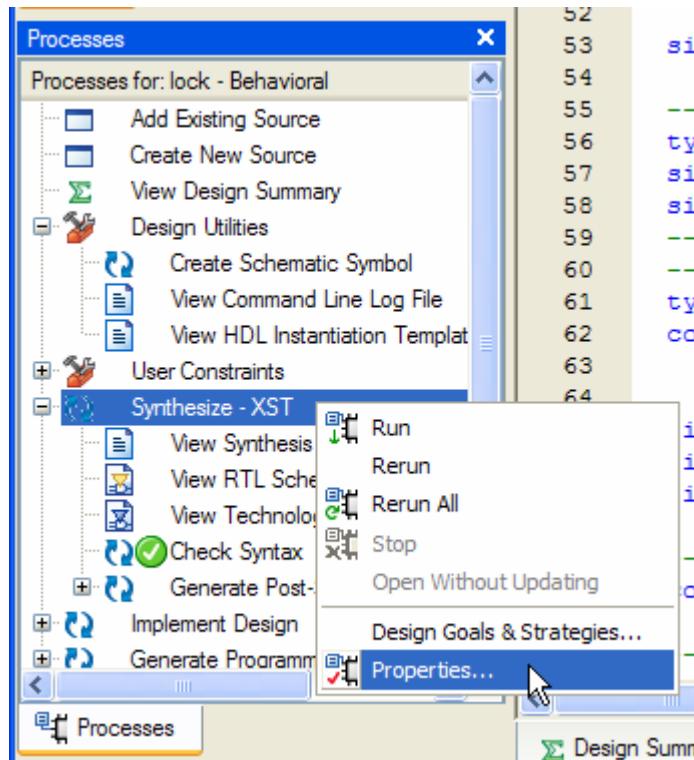
Lines 267-276: When the FSM is not being reset, then all the state registers are loaded with their next values here.

There are a few other points to note about the **lock** module:

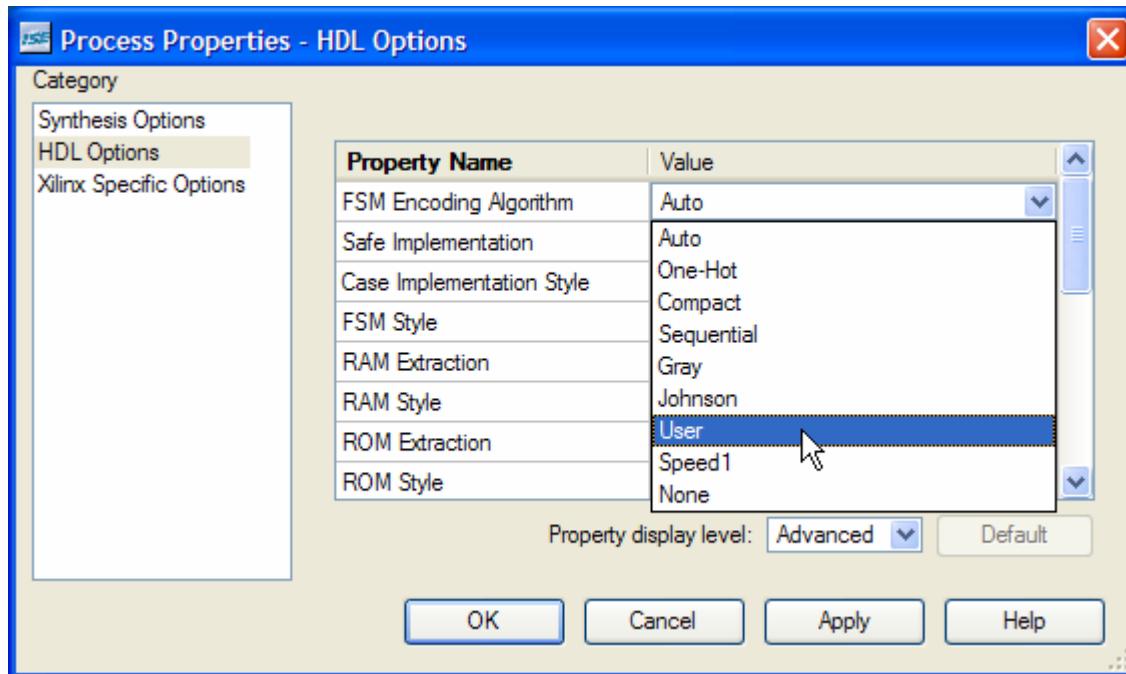
- The FSM requires a single clock cycle duration for the scancode ready signal. If the **rdy** output from the keyboard interface module stayed high for more than a single clock cycle on each key press, this would cause a transition between multiple states of the FSM.
- The **COMB_LENGTH** generic parameter on line 32 can be set to any value between 1 and 9 to change the number of keys in the combination. The use of an array to store the combination (lines 56-58) allows the module to resize itself automatically if you decide to change the length of the combination.
- The VHDL synthesizer automatically generates the bit encodings for the FSM states on lines 47-51. If you want to explicitly code the states, you can place lines such as the following below the state register declaration on line 53:

```
attribute ENUM_ENCODING: STRING;
attribute ENUM_ENCODING of lockStateType:type is
  "00001 00010 00100 01000 10000";
```

Next, highlight the **scancodereg** module in the Sources pane (since this was the first module added to the project, it is the top-level module by default), and then right-click on the Synthesize process and select Properties from the pop-up menu:



Then select User in the FSM Encoding Algorithm field:



Now your explicit state encodings will be used. You can check this by looking in the synthesis report to see:

```
=====
*           Advanced HDL Synthesis      *
=====

Optimizing FSM <XLXI_2/lockState_r> on signal <lockState_r[1:5]>
with one-hot encoding.
-----
 State          | Encoding
 -----
 entercomb     | 00001
 checkcomb    | 00010
 unlocked      | 00100
 enternewcomb  | 01000
 verifynewcomb | 10000
-----
```

Listing 2: VHDL code for the lock & key module.

```

1  --
2  -- Company:
3  -- Engineer:
4  --
5  -- Create Date: 09:36:57 06/03/2006
6  -- Design Name:
7  -- Module Name: lock - Behavioral
8  -- Project Name:
9  -- Target Devices:
10 -- Tool versions:
11 -- Description:
12 --
13 -- Dependencies:
14 --
15 -- Revision:
```

```

16 -- Revision 0.01 - File Created
17 -- Additional Comments:
18 --
19 -----
20 library IEEE;
21 use IEEE.STD_LOGIC_1164.ALL;
22 use IEEE.STD_LOGIC_ARITH.ALL;
23 use IEEE.STD_LOGIC_UNSIGNED.ALL;
24
25 ---- Uncomment the following library declaration if instantiating
26 ---- any Xilinx primitives in this code.
27 --library UNISIM;
28 --use UNISIM.VComponents.all;
29
30 entity lock is
31   Generic (
32     COMB_LENGTH : natural := 2  -- number of scancodes in a combination
33   );
34   Port (
35     rst : in STD_LOGIC;    -- reset
36     clk : in STD_LOGIC;    -- master clock
37     rdy : in STD_LOGIC;    -- true when a new keystroke is available
38     sc : in STD_LOGIC_VECTOR (7 downto 0); -- scancode for keystroke
39     led : out STD_LOGIC_VECTOR (6 downto 0) -- LED status indicators
40   );
41 end lock;
42
43 architecture Behavioral of lock is
44
45   -- State definitions for the FSM.
46   type lockStateType is (
47     enterComb,          -- user enters keystrokes which are checked against stored comb.
48     checkComb,          -- check the entered keystrokes to see if they match the comb.
49     unlocked,           -- user's keystrokes match combination, so unlock the lock
50     enterNewComb,       -- user enters a new combination
51     verifyNewComb      -- user re-types the new combination to verify it
52   );
53   signal lockState_r, lockState_x: lockStateType; -- current and next FSM state
54
55   -- The key scancodes composing a combination are stored in an array.
56   type combinationType is ARRAY (0 to COMB_LENGTH-1) of std_logic_vector(sc'range);
57   signal activeComb_r, activeComb_x : combinationType; -- active combination
58   signal newComb_r, newComb_x : combinationType; -- new combination
59   -- The default combination is an array the size of the longest allowable
60   -- combination filled with the scancodes for the keys "1","2",...,"9".
61   type defaultCombinationType is ARRAY (0 to 8) of std_logic_vector(sc'range);
62   constant defaultComb : defaultCombinationType :=
63     (x"16",x"1e",x"26",x"25",x"2e",x"36",x"3d",x"3e",x"46");
64
65   signal index_r, index_x : natural range COMB_LENGTH-1 downto 0; -- comb. array index
66   signal match_r, match_x : std_logic; -- true when user keystrokes = active comb.
67   signal rst_i : std_logic := '1'; -- power-on reset flag
68
69   -- Scancode for the key that initiates the entry of a new combination.
70   constant NEW_COMB_KEY: std_logic_vector(sc'range) := "01100110"; -- backspace
71
72   -- Function that displays a hex digit on the 7-segment LED.
73   subtype ledOutputType is std_logic_vector(6 downto 0);
74   function ledDecoder (digit: natural) return ledOutputType is
75   begin
76     case digit is
77       when 0 => return "1110111";    -- 0
78       when 1 => return "0010010";    -- 1
79       when 2 => return "1011101";    -- 2
80       when 3 => return "1011011";    -- 3
81       when 4 => return "0111010";    -- 4
82       when 5 => return "1101011";    -- 5
83       when 6 => return "1101111";    -- 6
84       when 7 => return "1010010";    -- 7
85       when 8 => return "1111111";    -- 8
86       when 9 => return "1111011";    -- 9

```

```

87      when 10 => return "1111110";      -- A
88      when 11 => return "0101111";      -- b
89      when 12 => return "0001101";      -- c
90      when 13 => return "0011111";      -- d
91      when 14 => return "1101101";      -- E
92      when 15 => return "1101100";      -- F
93      when others => return "1101101";
94  end case;
95 end ledDecoder;
96
97 -- 7-segment LED output for various conditions
98 constant LED_LOCKED      : ledOutputType := "0100101"; -- "L" - locked
99 constant LED_UNLOCKED     : ledOutputType := "0110111"; -- "U" - unlocked
100 constant LED_ENTERNEWCOMB : ledOutputType := "0001100"; -- "r" - replace comb
101 constant LED_VERIFYNEWCOMB : ledOutputType := "1100101"; -- "C" - check comb
102
103 begin
104
105 -- this process determines the next state given the current state and inputs
106 combinatorial: process(rdy,sc,lockState_r,index_r,match_r,activeComb_r,newComb_r)
107 begin
108
109 -- set default values for outputs and the next states of the registers
110 led        <= (others=>'0'); -- turn off all LEDs
111 match_x    <= match_r;       -- keep current value
112 index_x    <= index_r;       -- keep current value
113 lockState_x <= lockState_r; -- keep in current state
114 for i in 0 to COMB_LENGTH-1 loop
115   -- keep combinations unchanged
116   activeComb_x(i) <= activeComb_r(i);
117   newComb_x(i)     <= newComb_r(i);
118 end loop;
119
120 -- determine the next state and the outputs
121 case lockState_r is
122
123   -- FSM is in this state when the user is entering the combination
124   -- to unlock the lock. The index_r register points to the location
125   -- of the scancode in the combination that is being compared to the
126   -- current scancode entered by the user.
127 when enterComb =>
128   -- Show an "L" on the 7-segment LED before the user has pressed any keys.
129   -- After that, indicate the number of keypresses received from the user.
130   if index_r = 0 then
131     led <= LED_LOCKED; -- show "L" on LED to indicate lock is locked
132   else
133     led <= ledDecoder(index_r); -- index into comb. array = # of user keypresses
134   end if;
135   -- Wait until a new key scancode arrives.
136   if rdy = '1' then
137     -- Clear the match flag if the current scancode fails to match the
138     -- current scancode in the combination.
139     if sc /= activeComb_r(index_r) then
140       match_x <= '0';
141     end if;
142     -- Increment the index if the complete combination hasn't been entered yet.
143     if index_r /= COMB_LENGTH-1 then
144       index_x <= index_r + 1;
145     -- Otherwise, enough keys for a complete combination have been entered,
146     -- so see if it matches the active combination.
147     else
148       lockState_x <= checkComb;
149     end if;
150   end if;
151
152   -- FSM checks the match flag in this state to see if the user entered the
153   -- correct combination.
154 when checkComb =>
155   -- Reset the index and match flag.
156   index_x <= 0;
157   match_x <= '1';

```

```

158    -- If the match flag is still set, then all the user-entered scancodes
159    -- matched the active combination, so transition to the unlocked state.
160    if match_r = '1' then
161        lockState_x <= unlocked; -- keys match combination, so unlock the lock
162    else
163        lockState_x <= enterComb; -- otherwise, wait for another attempt
164    end if;
165
166    -- The lock is unlocked when the FSM is in this state.
167    when unlocked =>
168        led <= LED_UNLOCKED; -- show "U" on LED to indicate lock is unlocked
169        -- Wait until a new key scancode arrives.
170        if rdy = '1' then
171            -- After the active combination is entered, the user can hit the
172            -- backspace key to move to the state that allow entry of a new comb.
173            if sc = NEW_COMB_KEY then
174                lockState_x <= enterNewComb;
175                -- Otherwise, any other key will re-lock the lock and return to the state
176                -- that checks for the active combination to be entered.
177            else
178                lockState_x <= enterComb;
179            end if;
180        end if;
181
182    -- FSM is in this state when the user is entering a new combination
183    -- to replace the current active combination. The index_r register points to
184    -- the location where the next scancode will be stored in the new combination.
185    when enterNewComb =>
186        -- Show an "r" on the 7-segment LED before the user has pressed any keys.
187        -- After that, indicate the number of keypresses received from the user.
188        if index_r = 0 then
189            led <= LED_ENTERNEWCOMB; -- show "r" on LED to indicate comb. replacement
190        else
191            led <= ledDecoder(index_r); -- index into comb. array = # of user keypresses
192        end if;
193        -- Wait until a new key scancode arrives.
194        if rdy = '1' then
195            newComb_x(index_r) <= sc; -- store current scancode into new combination array
196            -- Increment the index if the new combination isn't complete yet.
197            if index_r /= COMB_LENGTH-1 then
198                index_x <= index_r + 1;
199                -- Otherwise, reset the index and get ready to verify the combination.
200            else
201                index_x <= 0;
202                lockState_x <= verifyNewComb;
203            end if;
204        end if;
205
206    -- FSM is in this state when the user is re-entering the new combination
207    -- to verify it against what was stored. The index_r register points to
208    -- the location of the next scancode that will be compared in the new comb.
209    when verifyNewComb =>
210        -- Show a "C" on the 7-segment LED before the user has pressed any keys.
211        -- After that, indicate the number of keypresses received from the user.
212        if index_r = 0 then
213            led <= LED_VERIFYNEWCOMB; -- show "C" on LED to indicate check of new comb.
214        else
215            led <= ledDecoder(index_r); -- index into comb. array = # of user keypresses
216        end if;
217        -- Wait until a new key scancode arrives.
218        if rdy = '1' then
219            -- Abort the operation as soon as a scancode does not match the new comb.
220            if sc /= newComb_r(index_r) then
221                index_x <= 0; -- reset the index
222                lockState_x <= enterComb; -- go wait for the user to enter the active comb.
223            else
224                -- Increment the index if the new comb. verification isn't complete yet.
225                if index_r /= COMB_LENGTH-1 then
226                    index_x <= index_r + 1;
227                    -- Otherwise, the user has correctly re-entered the new comb. so overwrite
228                    -- the active comb. with the new comb. and return to the locked state.

```

```

229      else
230          for i in 0 to COMB_LENGTH-1 loop
231              activeComb_x(i) <= newComb_r(i);
232          end loop;
233          index_x <= 0;
234          lockState_x <= enterComb;
235      end if;
236      end if;
237  end if;
238
239 -- Something wrong has happened if the FSM ever gets into this state, so
240 -- reset into the locked state.
241 when others =>
242     index_x <= 0;
243     match_x <= '1';
244     lockState_x <= enterComb;
245
246 end case;
247 end process;
248
249 -- This process just updates the various state registers with their next values
250 -- as computed by the previous process. It also handles the reset of the FSM.
251 update: process(rst,clk)
252 begin
253     if rising_edge(clk) then
254         -- Handle the reset input or the power-on reset.
255         if rst='1' or rst_i='1' then
256             rst_i <= '0'; -- clear power-on reset
257             index_r <= 0;
258             match_r <= '1';
259             lockState_r <= enterComb;
260             -- On power-on, initialize the active combination from the default value.
261             if rst_i = '1' then
262                 for i in 0 to COMB_LENGTH-1 loop
263                     activeComb_r(i) <= defaultComb(i);
264                 end loop;
265             end if;
266             -- Otherwise, update all registers with their next values.
267         else
268             index_r <= index_x;
269             match_r <= match_x;
270             lockState_r <= lockState_x;
271             for i in 0 to COMB_LENGTH-1 loop
272                 activeComb_r(i) <= activeComb_x(i);
273                 newComb_r(i) <= newComb_x(i);
274             end loop;
275         end if;
276     end if;
277 end process;
278
279 end Behavioral;
```

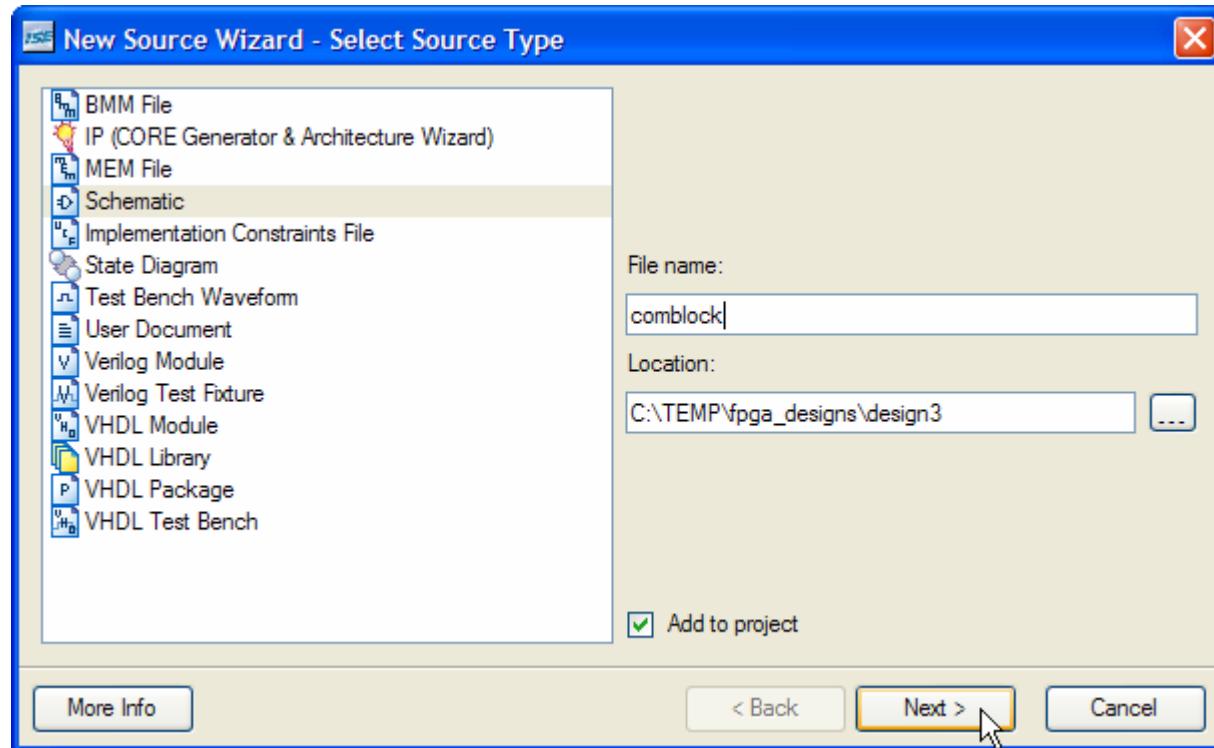
Once the VHDL for the **lock** module is entered, run a syntax check and generate a schematic symbol as you did for the **scancodereg** module.

The VHDL description of the FSM can be useful for two reasons:

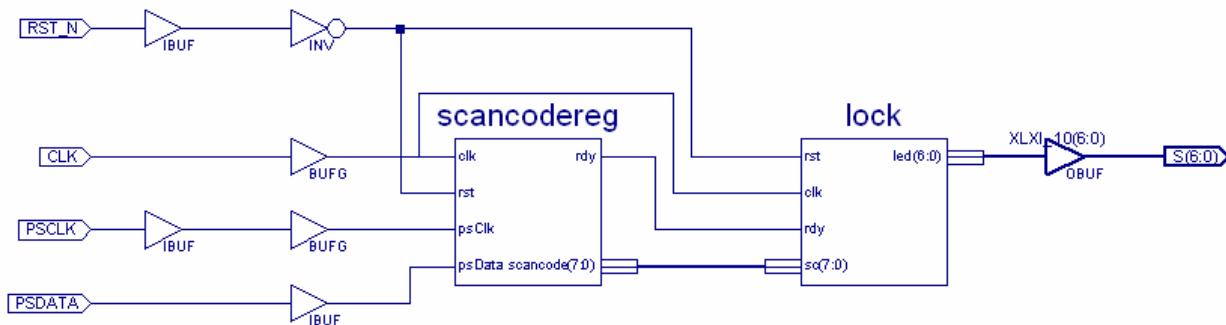
1. The editing area of the State Editor window gets very cluttered for complicated FSMs. You can use the State Editor to draw an initial, simplified version of your FSM and then add the rest of your description directly to the VHDL file. You cannot automatically back-annotate the additions to the VHDL file back into the State Editor, so the VHDL file must be used as the master design file for the FSM after you do this.
2. If you are unsure how to write FSM descriptions using VHDL, you can create simple FSMs in the State Editor and export them as VHDL to view the basic language constructs that are used.

Creating the Top-Level Module

The top-level module of the combination lock is built by connecting the keyboard interface and the lock&key modules together in a schematic. Right-click on the Sources pane and select New Source... from the pop-up menu. Then create a new schematic named **combLock** as shown below:



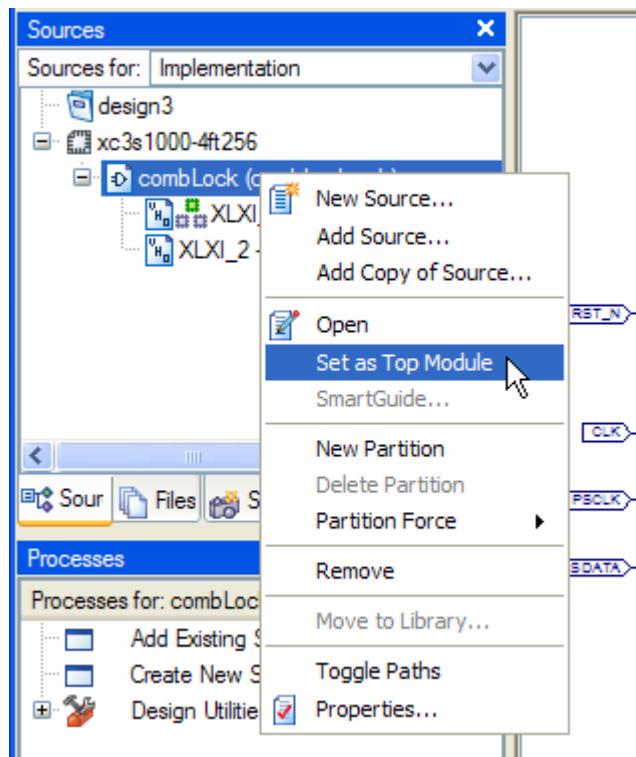
Enter the following components and connections into the schematic editor window:



Note the following:

1. The main clock (**CLK**) enters the FPGA on a dedicated clock pin (because that is the way it is connected on the XSA Board) so the input pad can connect directly to a general clock buffer (**BUFG**). Using the **BUFG** ensures that the clock signal reaches all the flip-flops in the design with minimal skew so they all change state at the same time.
2. The clock from the PS/2 keyboard (**PSCLK**) enters on a generic I/O pin so it must go through an input buffer (**IBUF**) before going through a **BUFG**.
3. The keyboard serial data signal (**PSDATA**) and the reset signal are standard, non-clock inputs so they just connect to **IBUFs**.
4. The reset signal (**RST_N**) is sourced by an active-low DIP-switch button on the XSA Board, so it passes through an inverter before going to the active-high reset inputs of the **scancodereg** and **lock** modules.
5. An array of seven output buffers (**OBUF**) as was created as in the design example of the previous chapter. The seven-bit led bus of the **lock** module connects to the inputs of the output buffers, and the buffer outputs connect to the output pads.

Once the components are connected to each other and the I/O, run a design-rule check to make sure there are no errors and then save the schematic. Then, make the schematic the top-level module of the design by right-clicking it in the Sources pane and selecting Set as Top Module from the pop-up menu.



Constraining the Design

Now you need to assign pins to the inputs and outputs, either by using PACE or by entering the pin assignments directly into the combLock.ucf constraints file. The appropriate pin assignments for each model of XSA Board are shown below. The **rst_n** input is driven by DIP-switch SW1-1 on the XSA Board. The **clk** input is assigned to a dedicated clock input on each FPGA to which a 50 MHz clock signal is applied. The **psclk** and **psdata** inputs are attached to the equivalent pins on the PS/2 port. The seven-segment LED pin assignments are the same as in the previous designs.

I/O Signal	XSA-50	XSA-100	XSA-200	XSA-3S1000
rst_n	P54	P54	P11	K4
clk	P88	P88	B8	P8
psclk	P94	P94	F4	B16
psdata	P93	P93	E1	E13
s0	P67	P67	N14	M6
s1	P39	P39	D14	M11
s2	P62	P62	N16	N6
s3	P60	P60	M16	R7
s4	P46	P46	F15	P10
s5	P57	P57	J16	T7
s6	P49	P49	G16	R10

Implementing the Design and Generating the Bitstream

Once you have specified the correct pin assignments, just double-click the Generate Programming File in the Process pane. Unfortunately, instead of a bitstream, you'll probably get an error like this:

```
ERROR:Place:1018 - A clock IOB/clock component pair have been found
that are not placed at an optimal clock IOB/clock site pair. The clock
component <XLXI_5> is placed at site <BUFGMUX5>. The IO component
<PSCLK> is placed at site <B16>. This will not allow the use of the
fast path between the IO and the Clock buffer. If this sub optimal
condition is acceptable for this design, you may use the
CLOCK_DEDICATED_ROUTE constraint in the .ucf file to demote this
message to a WARNING and allow your design to continue. However, the
use of this override is highly discouraged as it may lead to very poor
timing results. It is recommended that this error condition be
corrected in the design. A list of all the COMP.PINs used in this
clock placement rule is listed below. These examples can be used
directly in the .ucf file to override this clock rule.
```

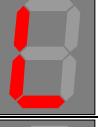
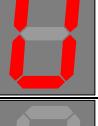
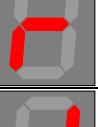
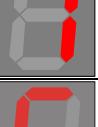
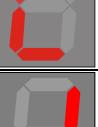
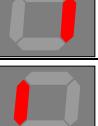
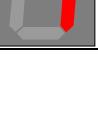
```
< NET "PSCLK" CLOCK_DEDICATED_ROUTE = FALSE; >
```

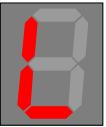
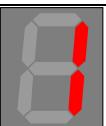
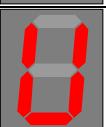
The Xilinx software tends to balk at using general-purpose inputs as clock sources for registers, and that's exactly what is being done with the clock from the PS/2 keyboard. The PS/2 clock has a very low frequency and it runs asynchronously with respect to the main logic clock, so it doesn't really matter that the PS/2 clock routing is sub-optimal. You can get around this error by placing the constraint listed above (strip off the '<' and '>' delimiters) into the comblock.ucf

file. Then you can re-run the synthesis and implementation processes to get an FPGA bitstream in the comblock.bit file.

Testing the Combination Lock

Attach a PS/2 keyboard to the six-pin mini-DIN socket at the bottom of the XSA Board. Then use GXSLOAD to download the bitstream to the XSA Board. Make sure DIP-switch SW1-1 is in the OFF position so the **rst_n** input is high. Now the combination lock should be ready to respond to key presses. A sequence of key presses and the results are shown below:

Press key...	LED displays...	New State...	This means...
None		enterComb	The combination lock is locked and is waiting for the default combination to be entered from the keyboard.
1		enterComb	The first key of the default combination was entered.
2		checkComb → unlocked	The second key of the default combination was entered. The lock is now unlocked
backspace		enterNewComb	The backspace initiates the replacement of the current combination with a new one.
q		enterNewComb	The first key of the new was entered.
w		verifyNewComb	The second key of the new combination was entered. Now the combination must be re-typed to check it.
q		verifyNewComb	The first key of the combination was re-typed correctly.
w		enterComb	The second key of the combination was entered correctly and the lock is now waiting for the new combination to be entered.
1		enterComb	The first key of the previous combination was entered.

2		enterComb	The second key of the previous combination was entered and the lock remains locked because this combination is no longer active.
q		enterComb	The first key of the new combination was entered.
w		checkComb → unlocked	The second key of the new combination was entered. The lock is now unlocked.

6

Going Further...

OK! You made it to the end! You have scratched the surface of programmable logic design, but how do you learn even more? Here are a few easy things to do:

- Select Help→Software Manuals. You will be presented with an Adobe Acrobat document that lists all the manuals for the ISE software. This includes a 300-page set of guidelines on synthesis and simulation techniques for FPGA designs.
- Select Help→Xilinx on the Web→Xilinx Application Notes. This will take you to a large set of interesting designs that have been done using Xilinx FPGAs.
- Get *Essential VHDL* (ISBN:0-9669590-0-0) or *The Designer's Guide to VHDL* (ISBN:1-55860-270-4) to learn more about VHDL for logic design.
- Read the *comp.arch.fpga* newsgroup for helpful questions and answers about programmable logic design.