# Time Domain

#### Time Domain

# 1 Definition

 $f_s$  = sampling rate

 $\lambda$  = wavelength

T = period

 $\omega$  = angular frequency

## 2 Wavelength $\lambda$

$$\lambda = \frac{c}{f}$$

### 3 Period T

$$T = 1 ms = 1000 Hz$$

### 4 Angular Frequency $\omega$

$$\omega = 2\pi f$$
$$= \frac{2\pi}{T}$$

$$\omega_0 = 2\pi T$$
$$= \pi \frac{f_0}{f s}$$

### 5 Unit Pulse & Unit Step

unit pulse:

$$\delta(n) = \begin{cases} 1, & \text{if } n = 0 \\ 0 & \text{otherwise} \end{cases}$$

unit step:

$$u(n) = \begin{cases} 1, & \text{if } n \ge 0 \\ 0 & \text{otherwise} \end{cases}$$

### 6 Harmonic Signals

$$n = [1:t \cdot f_s]$$

$$x[n] = sin(\omega_0 n + \varphi)$$

$$= sin(2\pi f_0 T + \varphi)$$

$$= sin(2\pi \frac{f_0}{fs} n + \varphi)$$

$$x[n] = e^{-i\omega_0 n}$$
  
=  $c o s(\omega_0 n) + i s i n(\omega_0 n)$ 

### Frequency Domain

#### 1 Definition

 $f_s = \text{sampling rate}$ 

 $\lambda$  = wavelength

T = period

 $\omega$  = angular frequency

### LTI Systems

$$y[n] = T\{x[n]\}$$

Linear time-invariant systems (LTI systems) are a class of systems used in signals and systems that are both linear and time-invariant. Linear systems are systems whose outputs for a linear combination of inputs are the same as a linear combination of individual responses to those inputs. Time-invariant systems are systems where the output does not depend on when an input was applied. These properties make LTI systems easier to represent and understand.

- 1. memory freedom
- 2. causality
- 3. stability (BIBO)

### Filter Design

### 1 In General

The lowpass general, standard form is given by

$$|H(j\omega)| = \left| \frac{A_0}{1 + \epsilon^2 A_N^2(\frac{\omega}{\omega_g})} \right|$$

with the design criteria defined by

#### 2 Butterworth

The Butterworth filter is a type of signal processing filter designed to have a frequency response as flat as possible in the passband. It is also referred to as a maximally flat magnitude filter, given by

$$A_N(\frac{\omega}{\omega_g}) = (\frac{\omega}{\omega_g})^N$$

so that

$$|H(j\omega)| = \left| \frac{A_0}{1 + \epsilon^2 (\frac{\omega}{\omega_g})^{2N}} \right|$$

The order N is therefore given by

$$\begin{split} \epsilon & (\frac{\omega}{\omega_g})^N = \lambda \\ \Rightarrow & (\frac{\omega}{\omega_g})^N = \frac{\lambda}{\epsilon} \\ \Rightarrow & N \cdot log_{10}(\frac{\omega_s}{\omega_g}) = log_{10}(\frac{\lambda}{\epsilon}) \\ \Rightarrow & N \geq \frac{log_{10}(\eta_s)}{log_{10}(\frac{\omega_s}{\omega_g})} \end{split}$$

for  $\eta_s = \frac{\omega_s}{\omega_g}$  and  $\eta = \frac{\omega}{\omega_g}$ . To find the poles we use

$$s_{\infty n} = \omega_g e^{-\frac{1}{N}} e^{j\frac{\pi}{2}(\frac{(2n-1)}{N}+1)}$$

and find 2N equidistant poles in a circle with radius  $r = \omega_{\sigma} e^{\frac{1}{N}}$ .

#### 3 Tchebychev I & II

The tchebychev filter is given by

$$\left|H(j\omega)\right| = \left|\frac{A_0}{1 + \epsilon^2 T_N^2(\frac{\omega}{\omega_g})}\right|$$

and uses the tchebychev polynomial  $T_N$  defined by

$$T_N(\eta) = \left\{ \begin{array}{ll} cos[N \cdot arccos(\eta)], & \text{for } |\eta| \le 1 \\ cosh[N \cdot arccosh(\eta)], & \text{for } |\eta| > 1 \end{array} \right\}$$